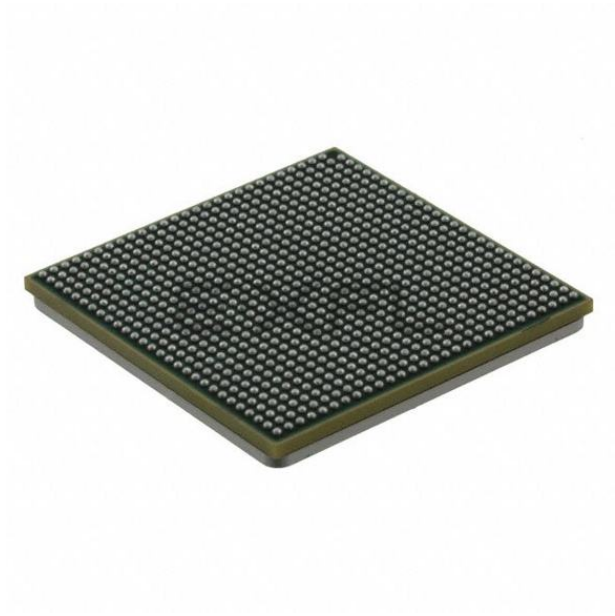

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Understanding Embedded - DSP (Digital Signal Processors)

Embedded - DSP (Digital Signal Processors) are specialized microprocessors designed to perform complex mathematical computations on digital signals in real-time. Unlike general-purpose processors, DSPs are optimized for high-speed numeric processing tasks, making them ideal for applications that require efficient and precise manipulation of digital data. These processors are fundamental in converting and processing signals in various forms, including audio, video, and communication signals, ensuring that data is accurately interpreted and utilized in embedded systems.

Applications of Embedded - DSP (Digital Signal Processors)

Details	
Product Status	Obsolete
Type	SC3850 Quad Core
Interface	Ethernet, I²C, PCI, RGMII, Serial RapidIO, SGMII, SPI, UART/USART
Clock Rate	1GHz
Non-Volatile Memory	ROM (96kB)
On-Chip RAM	576kB
Voltage - I/O	2.50V
Voltage - Core	1.00V
Operating Temperature	-40°C ~ 105°C (TJ)
Mounting Type	Surface Mount
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCPBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/msc8154tvt1000b

1 Pin Assignment

This section includes diagrams of the MSC8154 package ball grid array layouts and tables showing how the pinouts are allocated for the package.

1.1 FC-PBGA Ball Layout Diagram

The top view of the FC-PBGA package is shown in Figure 3 with the ball location index numbers.

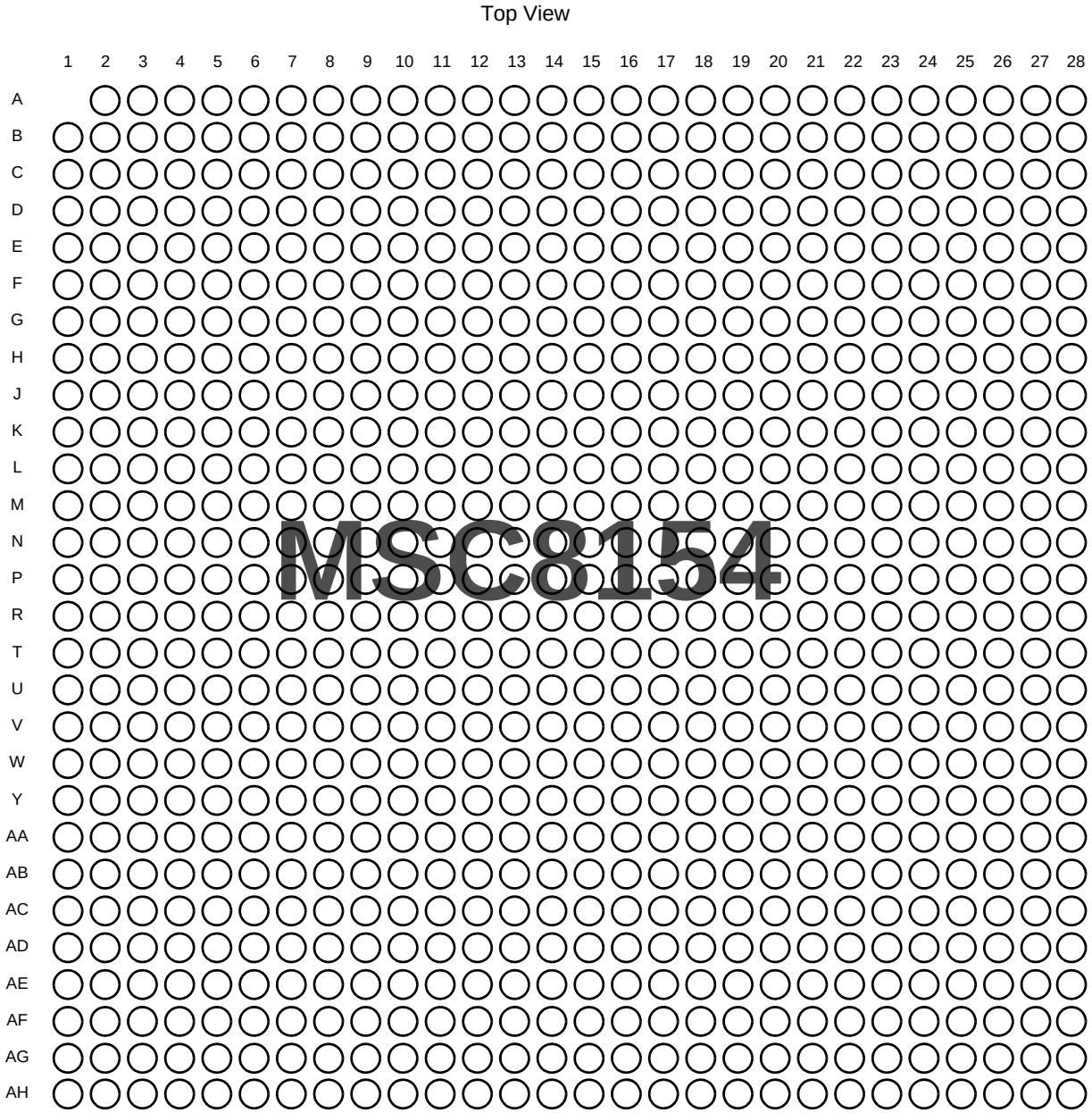


Figure 3. MSC8154 FC-PBGA Package, Top View

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
R13	VSS	Ground	N/A
R14	VDD	Power	N/A
R15	VSS	Ground	N/A
R16	MVDD	Power	N/A
R17	VSS	Ground	N/A
R18	VDD	Power	N/A
R19	VSS	Ground	N/A
R20	VSS	Non-user	N/A
R21	SXPVSS2	Ground	N/A
R22	SXPVDD2	Power	N/A
R23	SR2_TXD1/PE_TXD1[4]	O	SXPVDD2
R24	$\overline{\text{SR2\_TXD1}}/\overline{\text{PE\_TXD1}}$[4]	O	SXPVDD2
R25	SXCVSS2	Ground	N/A
R26	SXCVDD2	Power	N/A
R27	$\overline{\text{SR2\_RXD1}}/\overline{\text{PE\_RXD1}}$[4]	I	SXCVDD2
R28	SR2_RXD1/PE_RXD1[4]	I	SXCVDD2
T1	VSS	Ground	N/A
T2	TCK	I	QVDD
T3	$\overline{\text{SRESET}}$[6,7]	I/O	QVDD
T4	TDI	I	QVDD
T5	VSS	Ground	N/A
T6	TDO	O	QVDD
T7	VSS	Ground	N/A
T8	VSS	Ground	N/A
T9	QVDD	Power	N/A
T10	VSS	Ground	N/A
T11	VDD	Power	N/A
T12	VSS	Ground	N/A
T13	M3VDD	Power	N/A
T14	VSS	Ground	N/A
T15	VDD	Power	N/A
T16	VSS	Ground	N/A
T17	MVDD	Power	N/A
T18	VSS	Ground	N/A
T19	VDD	Power	N/A
T20	VSS	Ground	N/A
T21	VSS	Non-user	N/A
T22	SR2_IMP_CAL_RX	I	SXCVDD2
T23	SXPVSS2	Ground	N/A
T24	SXPVDD2	Power	N/A
T25	SR2_REF_CLK	I	SXCVDD2
T26	$\overline{\text{SR2\_REF\_CLK}}$	I	SXCVDD2
T27	Reserved	NC	—
T28	Reserved	NC	—
U1	M1DQ8	I/O	GVDD1
U2	VSS	Ground	N/A

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
Y11	GVDD1	Power	N/A
Y12	VSS	Ground	N/A
Y13	GVDD1	Power	N/A
Y14	VSS	Ground	N/A
Y15	GVDD1	Power	N/A
Y16	VSS	Ground	N/A
Y17	GVDD1	Power	N/A
Y18	VSS	Ground	N/A
Y19	GVDD1	Power	N/A
Y20	VSS	Ground	N/A
Y21	NVDD	Power	N/A
Y22	GPIO20/$\overline{\text{SPI\_SL}}$[5,8]	I/O	NVDD
Y23	GPIO17/SPI_SCK[5,8]	I/O	NVDD
Y24	GPIO14/DRQ0/$\overline{\text{IRQ14}}$/RC14[5,8]	I/O	NVDD
Y25	GPIO12/$\overline{\text{IRQ12}}$/RC12[5,8]	I/O	NVDD
Y26	GPIO8/$\overline{\text{IRQ8}}$/RC8[5,8]	I/O	NVDD
Y27	NVDD	Power	N/A
Y28	VSS	Ground	N/A
AA1	GVDD1	Power	N/A
AA2	VSS	Ground	N/A
AA3	M1DQ18	I/O	GVDD1
AA4	GVDD1	Power	N/A
AA5	VSS	Ground	N/A
AA6	M1DQ20	I/O	GVDD1
AA7	GVDD1	Power	N/A
AA8	VSS	Ground	N/A
AA9	M1A15	O	GVDD1
AA10	M1CK2	O	GVDD1
AA11	M1MDIC0	I/O	GVDD1
AA12	M1VREF	I	GVDD1
AA13	M1MDIC1	I/O	GVDD1
AA14	M1DQ46	I/O	GVDD1
AA15	M1DQ47	I/O	GVDD1
AA16	M1DQ45	I/O	GVDD1
AA17	M1DQ41	I/O	GVDD1
AA18	M1DQ62	I/O	GVDD1
AA19	M1DQ63	I/O	GVDD1
AA20	M1DQ61	I/O	GVDD1
AA21	VSS	Ground	N/A
AA22	GPIO21[5,8]	I/O	NVDD
AA23	GPIO18/SPI_MOSI[5,8]	I/O	NVDD
AA24	GPIO16/RC16[5,8]	I/O	NVDD
AA25	GPIO4/DDN1/$\overline{\text{IRQ4}}$/RC4[5,8]	I/O	NVDD
AA26	GPIO9/$\overline{\text{IRQ9}}$/RC9[5,8]	I/O	NVDD
AA27	GPIO6/$\overline{\text{IRQ6}}$/RC6[5,8]	I/O	NVDD
AA28	GPIO1/$\overline{\text{IRQ1}}$/RC1[5,8]	I/O	NVDD

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
AB1	$\overline{\text{M1DQS2}}$	I/O	GVDD1
AB2	M1DQS2	I/O	GVDD1
AB3	M1DQ19	I/O	GVDD1
AB4	M1DM2	O	GVDD1
AB5	M1DQ21	I/O	GVDD1
AB6	M1DQ22	I/O	GVDD1
AB7	M1CKE0	O	GVDD1
AB8	M1A11	O	GVDD1
AB9	M1A7	O	GVDD1
AB10	$\overline{\text{M1CK2}}$	O	GVDD1
AB11	M1APAR_OUT	O	GVDD1
AB12	M1ODT1	O	GVDD1
AB13	$\overline{\text{M1APAR\_IN}}$	I	GVDD1
AB14	M1DQ43	I/O	GVDD1
AB15	M1DM5	O	GVDD1
AB16	M1DQ44	I/O	GVDD1
AB17	M1DQ40	I/O	GVDD1
AB18	M1DQ59	I/O	GVDD1
AB19	M1DM7	O	GVDD1
AB20	M1DQ60	I/O	GVDD1
AB21	VSS	Ground	N/A
AB22	GPIO31/I2C_SDA[5,8]	I/O	NVDD
AB23	GPIO27/TMR4/RCW_SRC0[5,8]	I/O	NVDD
AB24	GPIO25/TMR2/RCW_SRC1[5,8]	I/O	NVDD
AB25	GPIO24/TMR1/RCW_SRC2[5,8]	I/O	NVDD
AB26	GPIO10/$\overline{\text{IRQ10}}$/RC10[5,8]	I/O	NVDD
AB27	GPIO5/$\overline{\text{IRQ5}}$/RC5[5,8]	I/O	NVDD
AB28	GPIO0/$\overline{\text{IRQ0}}$/RC0[5,8]	I/O	NVDD
AC1	VSS	Ground	N/A
AC2	GVDD1	Power	N/A
AC3	M1DQ16	I/O	GVDD1
AC4	VSS	Ground	N/A
AC5	GVDD1	Power	N/A
AC6	M1DQ17	I/O	GVDD1
AC7	VSS	Ground	N/A
AC8	GVDD1	Power	N/A
AC9	M1BA2	O	GVDD1
AC10	VSS	Ground	N/A
AC11	GVDD1	Power	N/A
AC12	M1A4	O	GVDD1
AC13	VSS	Ground	N/A
AC14	GVDD1	Power	N/A
AC15	M1DQ42	I/O	GVDD1
AC16	VSS	Ground	N/A
AC17	GVDD1	Power	N/A
AC18	M1DQ58	I/O	GVDD1

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
AE9	M1A8	O	GVDD1
AE10	GVDD1	Power	N/A
AE11	VSS	Ground	N/A
AE12	M1A0	O	GVDD1
AE13	GVDD1	Power	N/A
AE14	VSS	Ground	N/A
AE15	M1DQ39	I/O	GVDD1
AE16	GVDD1	Power	N/A
AE17	VSS	Ground	N/A
AE18	M1DQ54	I/O	GVDD1
AE19	GVDD1	Power	N/A
AE20	VSS	Ground	N/A
AE21	GPIO29/UART_TXD[5,8]	I/O	NVDD
AE22	TDM1TCK/GE2_RX_CLK[3]	I	NVDD
AE23	TDM1RSN/GE2_RX_CTL[3]	I/O	NVDD
AE24	VSS	Ground	N/A
AE25	TDM3RCK/GE1_GTX_CLK[3]	I/O	NVDD
AE26	TDM3TSN/GE1_RX_CLK[3]	I/O	NVDD
AE27	TDM2RSN/GE1_TD2[3]	I/O	NVDD
AE28	TDM2RDT/GE1_TD1[3]	I/O	NVDD
AF1	M1DQ28	I/O	GVDD1
AF2	M1DM3	O	GVDD1
AF3	M1DQ26	I/O	GVDD1
AF4	M1ECC4	I/O	GVDD1
AF5	M1DM8	O	GVDD1
AF6	M1ECC2	I/O	GVDD1
AF7	M1CKE1	O	GVDD1
AF8	M1CK0	O	GVDD1
AF9	$\overline{\text{M1CK0}}$	O	GVDD1
AF10	M1BA1	O	GVDD1
AF11	M1A1	O	GVDD1
AF12	$\overline{\text{M1WE}}$	O	GVDD1
AF13	M1DQ37	I/O	GVDD1
AF14	M1DM4	O	GVDD1
AF15	M1DQ36	I/O	GVDD1
AF16	M1DQ32	I/O	GVDD1
AF17	M1DQ55	I/O	GVDD1
AF18	M1DM6	O	GVDD1
AF19	M1DQ53	I/O	GVDD1
AF20	M1DQ52	I/O	GVDD1
AF21	GPIO28/UART_RXD[5,8]	I/O	NVDD
AF22	TDM0RSN/GE2_TD2[3]	I/O	NVDD
AF23	TDM0TDT/GE2_TD3[3]	I/O	NVDD
AF24	NVDD	Power	N/A
AF25	TDM2TSN/GE1_TX_CTL[3]	I/O	NVDD
AF26	GE1_RX_CTL	I	NVDD

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
AH17	$\overline{\text{M1DQS6}}$	I/O	GVDD1
AH18	M1DQS6	I/O	GVDD1
AH19	M1DQ48	I/O	GVDD1
AH20	M1DQ49	I/O	GVDD1
AH21	VSS	Ground	N/A
AH22	TDM0RCK/GE2_RD2[3]	I/O	NVDD
AH23	TDM0RDT/GE2_RD3[3]	I/O	NVDD
AH24	TDM0TSN/GE2_RD0[3]	I/O	NVDD
AH25	TDM1RCK/GE2_RD1[3]	I/O	NVDD
AH26	TDM3TDT/GE1_RD3[3]	I/O	NVDD
AH27	TDM3TCK/GE1_RD2[3]	I	NVDD
AH28	VSS	Ground	N/A

Notes:
1. Reserved signals should be disconnected for compatibility with future revisions of the device. Non-user signals are reserved for manufacturing and test purposes only. The assigned signal name is used to indicate whether the signal must be unconnected (Reserved), pulled down (VSS), or pulled up (VDD).
2. Signal function during power-on reset is determined by the RCW source type.
3. Selection of TDM versus RGMII functionality is determined by the RCW bit values.
4. Selection of RapidIO, SGMII, and PCI Express functionality is determined by the RCW bit values.
5. Selection of the GPIO function and other functions is done by GPIO register setup. For configuration details, see the *GPIO* chapter in the *MSC8154 Reference Manual*.
6. Open-drain signal.
7. Internal 20 KΩ pull-up resistor.
8. For signals with GPIO functionality, the open-drain and internal 20 KΩ pull-up resistor can be configured by GPIO register programming. See the *GPIO* chapter of the *MSC8154 Reference Manual* for configuration details.
9. Connect to power supply via external filter. See **Section 3.2**, *PLL Power Supply Design Considerations* for details.
10. Pin types are: Ground = all VSS connections; Power = all VDD connections; I = Input; O = Output; I/O = Input/Output; NC = not connected.

2.5 DC Electrical Characteristics

This section describes the DC electrical characteristics for the MSC8154.

2.5.1 DDR SDRAM DC Electrical Characteristics

This section describes the DC electrical specifications for the DDR SDRAM interface of the MSC8154.

Note: DDR2 SDRAM uses V_{DDDDR}(typ) = 1.8 V and DDR3 SDRAM uses V_{DDDDR}(typ) = 1.5 V.

2.5.1.1 DDR2 (1.8 V) SDRAM DC Electrical Characteristics

Table 6 provides the recommended operating conditions for the DDR SDRAM controller when interfacing to DDR2 SDRAM.

Note: At recommended operating conditions (see Table 3) with V_{DDDDR} = 1.8 V.

Table 6. DDR2 SDRAM Interface DC Electrical Characteristics

Parameter/Condition	Symbol	Min	Max	Unit	Notes
I/O reference voltage	MV_{REF}	0.49 × V_{DDDDR}	0.51 × V_{DDDDR}	V	2, 3, 4
Input high voltage	V_{IH}	MV_{REF} + 0.125	V_{DDDDR} + 0.3	V	5
Input low voltage	V_{IL}	–0.3	MV_{REF} – 0.125	V	5
I/O leakage current	I_{OZ}	–50	50	μA	6
Output high current (V_{OUT} (VOH) = 1.37 V)	I_{OH}	–13.4	—	mA	7
Output low current (V_{OUT} (VOL) = 0.33 V)	I_{OL}	13.4	—	mA	7

Notes:
1. V_{DDDDR} is expected to be within 50 mV of the DRAM V_{DD} supply voltage at all times. The DRAM and memory controller can use the same or different sources.
2. MV_{REF} is expected to be equal to 0.5 × V_{DDDDR} and to track V_{DDDDR} DC variations as measured at the receiver. Peak-to-peak noise on MV_{REF} may not exceed ±2% of the DC value.
3. V_{TT} is not applied directly to the device. It is the supply to which far end signal termination is made and is expected to be equal to MV_{REF} with a minimum value of MV_{REF} – 0.4 and a maximum value of MV_{REF} + 0.04 V. V_{TT} should track variations in the DC-level of MV_{REF}.
4. The voltage regulator for MV_{REF} must be able to supply up to 300 μA.
5. Input capacitance load for DQ, DQS, and DQS signals are available in the IBIS models.
6. Output leakage is measured with all outputs are disabled, 0 V ≤ V_{OUT} ≤ V_{DDDDR}.
7. Refer to the IBIS model for the complete output IV curve characteristics.

2.5.2.2 SerDes Reference Clock Receiver Characteristics

The SerDes reference clock inputs are applied to an internal PLL whose output creates the clock used by the corresponding SerDes lanes. The SerDes reference clock inputs are SR1_REF_CLK/$\overline{\text{SR1\_REF\_CLK}}$ or SR2_REF_CLK/$\overline{\text{SR2\_REF\_CLK}}$. Figure 5 shows a receiver reference diagram of the SerDes reference clocks.

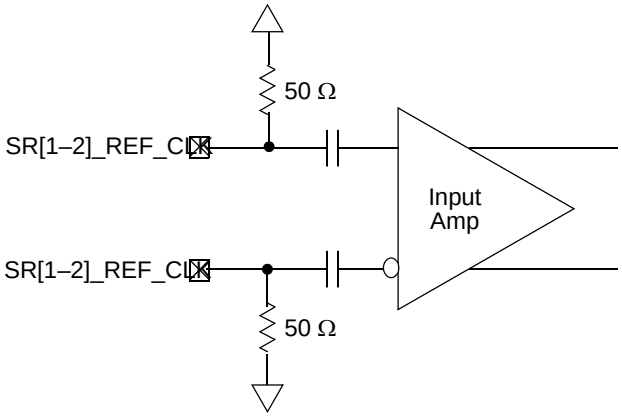


Figure 5. Receiver of SerDes Reference Clocks

The characteristics of the clock signals are as follows:

- The supply voltage requirements for V_{DDSXC} are as specified in **Table 3.**
- The SerDes reference clock receiver reference circuit structure is as follows:
  - The SR[1–2]_REF_CLK and $\overline{\text{SR[1–2]\_REF\_CLK}}$ are internally AC-coupled differential inputs as shown in Figure 5. Each differential clock input (SR[1–2]_REF_CLK or $\overline{\text{SR[1–2]\_REF\_CLK}}$) has on-chip 50-Ω termination to GND_{SXC} followed by on-chip AC-coupling.
  - The external reference clock driver must be able to drive this termination.
  - The SerDes reference clock input can be either differential or single-ended. Refer to the differential mode and single-ended mode descriptions below for detailed requirements.
- The maximum average current requirement also determines the common mode voltage range.
  - When the SerDes reference clock differential inputs are DC coupled externally with the clock driver chip, the maximum average current allowed for each input pin is 8 mA. In this case, the exact common mode input voltage is not critical as long as it is within the range allowed by the maximum average current of 8 mA because the input is AC-coupled on-chip.
  - This current limitation sets the maximum common mode input voltage to be less than 0.4 V (0.4 V / 50 = 8 mA) while the minimum common mode input level is 0.1 V above GND_{SXC}. For example, a clock with a 50/50 duty cycle can be produced by a clock driver with output driven by its current source from 0 mA to 16 mA (0–0.8 V), such that each phase of the differential input has a single-ended swing from 0 V to 800 mV with the common mode voltage at 400 mV.
  - If the device driving the SR[1–2]_REF_CLK and $\overline{\text{SR[1–2]\_REF\_CLK}}$ inputs cannot drive 50 Ω to GND_{SXC} DC or the drive strength of the clock driver chip exceeds the maximum input current limitations, it must be AC-coupled externally.
- The input amplitude requirement is described in detail in the following sections.

Table 14. Serial RapidIO Receiver DC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Notes
Differential input voltage	V_{IN}	200	—	1600	mVp-p	1

Notes: **1.** Measured at receiver.

2.5.3.4 DC-Level Requirements for SGMII Configurations

Note: Specifications are valid at the recommended operating conditions listed in Table 3

Table 15 describes the SGMII SerDes transmitter AC-coupled DC electrical characteristics. Transmitter DC characteristics are measured at the transmitter outputs (SR[1–2]_TX[n] and $\overline{\text{SR[1–2]\_TX[n]}}$) as shown in Figure 10.

Table 15. SGMII DC Transmitter Electrical Characteristics

Parameter	Symbol	Min	Typ	Max	Unit	Notes
Output high voltage	V_{OH}	—	—	$XV_{DD\_SRDS\text{-}Typ}/2 + \lvert V_{OD}\rvert_{\text{-}max}/2$	mV	1
Output low voltage	V_{OL}	$XV_{DD\_SRDS\text{-}Typ}/2 - \lvert V_{OD}\rvert_{\text{-}max}/2$	—	—	mV	1
Output differential voltage ($XV_{DD\text{-}Typ}$ at 1.0 V)	$\lvert V_{OD}\rvert$	323	500	725	mV	2,3,4
		296	459	665		2,3,5
		269	417	604		2,3,6
		243	376	545		2,3,7
		215	333	483		2,3,8
		189	292	424		2,3,9
		162	250	362		2,3,10
Output impedance (single-ended)	R_O	40	50	60	Ω	—

Notes:
1. This does not align to DC-coupled SGMII. $XV_{DD\_SRDS2\text{-}Typ}$ = 1.1 V.
2. The $\lvert V_{OD}\rvert$ value shown in the table assumes full multitude by setting srd_smit_lvl as 000 and the following transmit equalization setting in the XMITEQ**AB** (for lanes A and B) or XMITEQ**EF** (for lanes E and F) bit field of Control Register:
• The MSB (bit 0) of the above bit field is set to zero (selecting the full $V_{DD\text{-}DIFF\text{-}p\text{-}p}$ amplitude which is power up default);
• The LSB (bit [1–3]) of the above bit field is set based on the equalization settings listed in notes 4 through 10.
3. The $\lvert V_{OD}\rvert$ value shown in the Typ column is based on the condition of $XV_{DD\_SRDS2\text{-}Typ}$ = 1.0 V, no common mode offset variation (V_{OS} =500mV), SerDes transmitter is terminated with 100-Ω differential load between
4. Equalization setting: 1.0x: 0000.
5. Equalization setting: 1.09x: 1000.
6. Equalization setting: 1.2x: 0100.
7. Equalization setting: 1.33x: 1100.
8. Equalization setting: 1.5x: 0010.
9. Equalization setting: 1.71x: 1010.
10. Equalization setting: 2.0x: 0110.
11. $\lvert V_{OD}\rvert = \lvert V_{SR[1–2]\_TXn} - V_{\overline{SR[1–2]\_TXn}}\rvert$. $\lvert V_{OD}\rvert$ is also referred to as output differential peak voltage. $V_{TX\text{-}DIFFp\text{-}p} = 2*\lvert V_{OD}\rvert$.

2.5.4 RGMII and Other Interface DC Electrical Characteristics

Table 17 describes the DC electrical characteristics for the following interfaces:

- RGMII Ethernet
- SPI
- TDM
- GPIO
- UART
- TIMER
- EE
- I^2C
- Interrupts ($\overline{\text{IRQn}}$, $\overline{\text{NMI\_OUT}}$, $\overline{\text{INT\_OUT}}$)
- Clock and resets (CLKIN, $\overline{\text{PORESET}}$, $\overline{\text{HRESET}}$, $\overline{\text{SRESET}}$)
- DMA External Request
- JTAG signals

Table 17. 2.5 V I/O DC Electrical Characteristics

Characteristic	Symbol	Min	Max	Unit	Notes
Input high voltage	V_{IH}	1.7	—	V	1
Input low voltage	V_{IL}	—	0.7	V	1
Input high current ($V_{IN} = V_{DDIO}$)	I_{IN}	—	30	μA	2
Output high voltage (V_{DDIO} = min, I_{OH} = –1.0 mA)	V_{OH}	2.0	VDDIO + 0.3	V	1
Output low voltage (V_{DDIO} = min, I_{OL}= 1.0 mA)	V_{OL}	GND – 0.3	0.40	V	1

Notes: **1.** The min V_{IL} and max V_{IH} values are based on the respective min and max V_{IN} values listed in Table 3.
2. The symbol V_{IN} represents the input voltage of the supply. It is referenced in Table 3.

Figure 13 shows the DDR SDRAM output timing diagram.

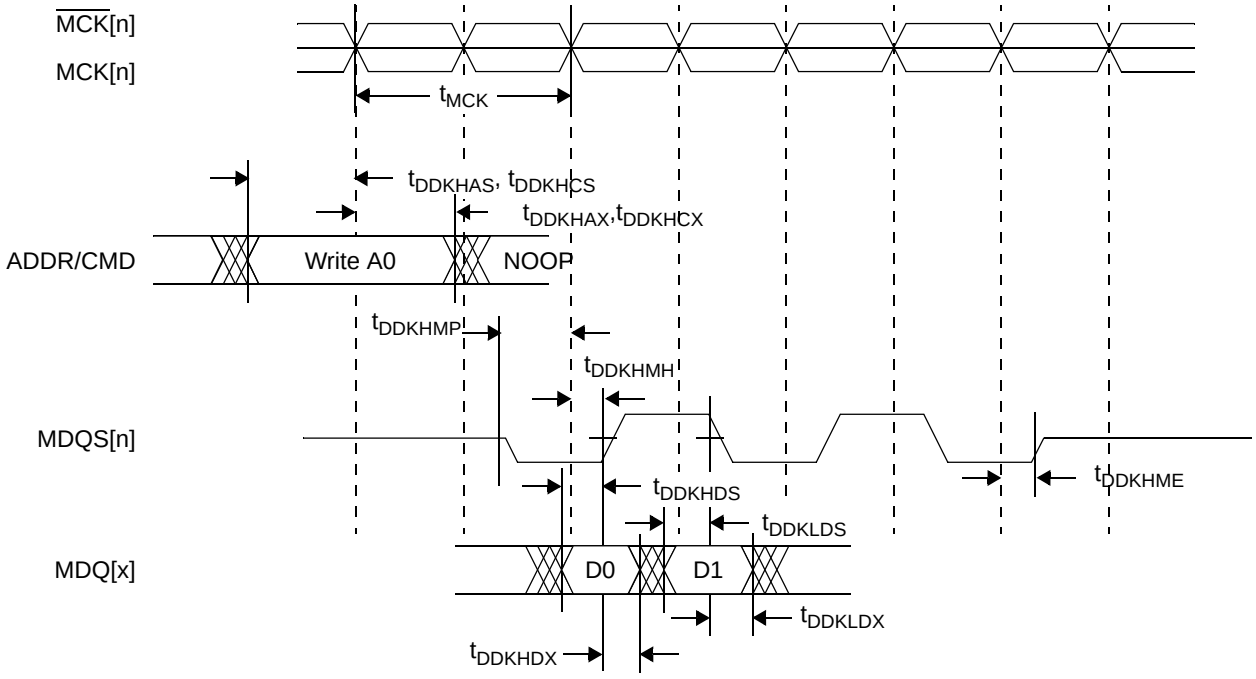


Figure 13. DDR SDRAM Output Timing

Figure 14 provides the AC test load for the DDR2 and DDR3 controller bus.

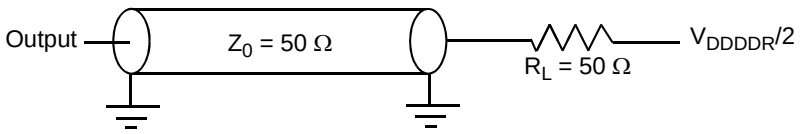


Figure 14. DDR2 and DDR3 Controller Bus AC Test Load

2.6.1.3 DDR2 and DDR3 SDRAM Differential Timing Specifications

This section describes the DC and AC differential timing specifications for the DDR2 and DDR3 SDRAM controller interface. Figure 15 shows the differential timing specification.

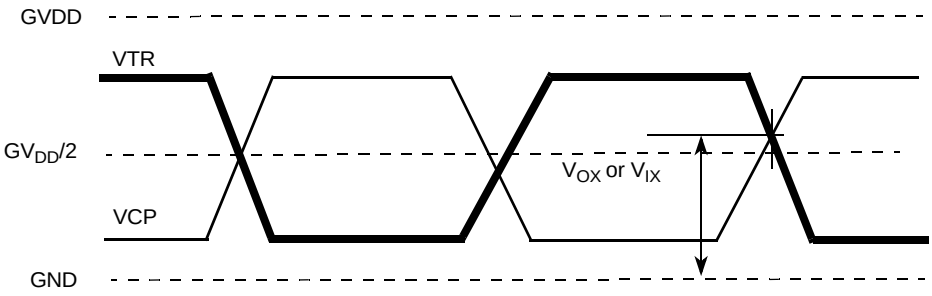


Figure 15. DDR2 and DDR3 SDRAM Differential Timing Specifications

Note: VTR specifies the true input signal (such as MCK or MDQS) and VCP is the complementary input signal (such as $\overline{\text{MCK}}$ or $\overline{\text{MDQS}}$).

Table 24. SR[1–2]_REF_CLK and $\overline{\text{SR[1–2]\_REF\_CLK}}$ Input Clock Requirements (continued)

Parameter	Symbol	Min	Typical	Max	Units	Notes

Notes:
1. Caution: Only 100 and 125 have been tested. Other values will not work correctly with the rest of the system.
2. Limits from PCI Express CEM Rev 1.0a
3. Measured from –200 mV to +200 mV on the differential waveform (derived from SR[1–2]_REF_CLK minus $\overline{\text{SR[1–2]\_REF\_CLK}}$). The signal must be monotonic through the measurement region for rise and fall time. The 400 mV measurement window is centered on the differential zero crossing. See Figure 16.
4. Measurement taken from differential waveform
5. Measurement taken from single-ended waveform
6. Matching applies to rising edge for SR[1–2]_REF_CLK and falling edge rate for $\overline{\text{SR[1–2]\_REF\_CLK}}$. It is measured using a 200 mV window centered on the median cross point where SR[1–2]_REF_CLK rising meets $\overline{\text{SR[1–2]\_REF\_CLK}}$ falling. The median cross point is used to calculate the voltage thresholds that the oscilloscope uses for the edge rate calculations. The rise edge rate of SR[1–2]_REF_CLK should be compared to the fall edge rate of $\overline{\text{SR[1–2]\_REF\_CLK}}$; the maximum allowed difference should not exceed 20% of the slowest edge rate. See Figure 17.

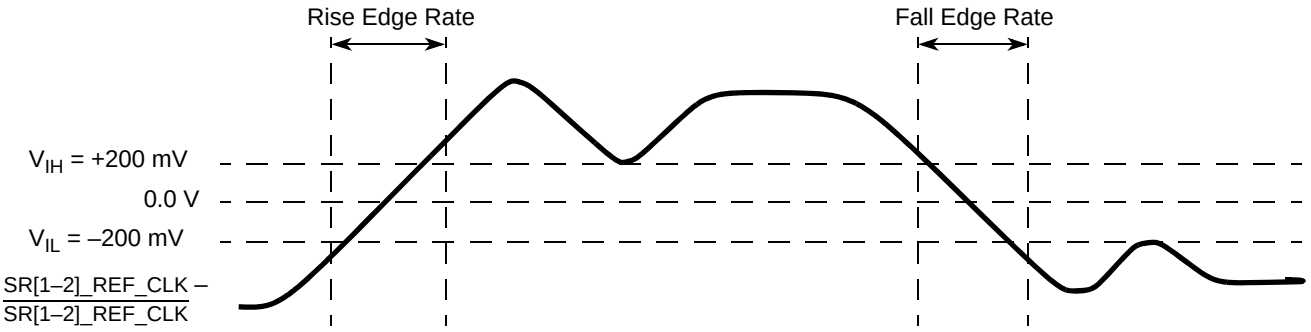


Figure 16. Differential Measurement Points for Rise and Fall Time

Figure 17. Single-Ended Measurement Points for Rise and Fall Time Matching

2.6.2.2 PCI Express AC Physical Layer Specifications

The AC requirements for PCI Express implementations have separate requirements for the Tx and Rx lines. The MSC8154 supports a 2.5 Gbps PCI Express interface defined by the *PCI Express Base Specification, Revision 1.0a*. The transmitter specifications are defined in Table 25 and the receiver specifications are defined in Table 26. The parameters are specified at the component pins. the AC timing specifications do not include REF_CLK jitter.

Note: Specifications are valid at the recommended operating conditions listed in Table 3.

Table 25. PCI Express (2.5 Gbps) Differential Transmitter (Tx) Output AC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Notes
Unit interval	UI	399.88	400.00	400.12	ps	1
Minimum Tx eye width	T_{TX-EYE}	0.70	—	—	UI	2, 3
Maximum time between the jitter median and maximum deviation from the median.	$T_{TX-EYE-MEDIAN-to-MAX-JITTER}$	—	—	0.15	UI	3, 4
AC coupling capacitor	C_{TX}	75	—	200	nF	5

Notes:
1. Each UI is 400 ps ± 300 ppm. UI does not account for spread spectrum clock dictated variations. No test load is necessarily associated with this value.
2. The maximum transmitter jitter can be derived as $T_{TX-MAX-JITTER} = 1 – T_{TX-EYE} = 0.3$ UI.
3. Specified at the measurement point into a timing and voltage compliance test load as shown in Figure 8 and measured over any 250 consecutive Tx UIs. A $T_{TX-EYE} = 0.70$ UI provides for a total sum of deterministic and random jitter budget of $T_{TX-JITTER-MAX} = 0.30$ UI for the transmitter collected over any 250 consecutive Tx UIs. The $T_{TX-EYE-MEDIAN-to-MAX-JITTER}$ median is less than half of the total Tx jitter budget collected over any 250 consecutive Tx UIs. It should be noted that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value. Jitter is defined as the measurement variation of the crossing points ($V_{TX-DIFFp-p} = 0$ V) in relation to a recovered Tx UI. A recovered Tx UI is calculated over 3500 consecutive unit intervals of sample data.
4. Jitter is measured using all edges of the 250 consecutive UI in the center of the 3500 UI used for calculating the Tx UI.
5. All transmitters shall be AC-coupled. The AC coupling is required either within the media or within the transmitting component itself. The SerDes transmitter does not have built-in Tx capacitance. An external AC coupling capacitor is required.

Table 26. PCI Express (2.5 Gbps) Differential Receiver (Rx) Input AC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Notes
Unit Interval	UI	399.88	400.00	400.12	ps	1
Minimum receiver eye width	T_{RX-EYE}	0.4	—	—	UI	2, 3, 4
Maximum time between the jitter median and maximum deviation from the median.	$T_{RX-EYE-MEDIAN-to-MAX-JITTER}$	—	—	0.3	UI	3, 4, 5

Notes:
1. Each UI is 400 ps ± 300 ppm. UI does not account for spread spectrum clock dictated variations. No test load is necessarily associated with this value.
2. The maximum interconnect media and transmitter jitter that can be tolerated by the receiver can be derived as $T_{RX-MAX-JITTER} = 1 – T_{RX-EYE} = 0.6$ UI.
3. Specified at the measurement point and measured over any 250 consecutive UIs. The test load in Figure 8 should be used as the Rx device when taking measurements. If the clocks to the Rx and Tx are not derived from the same reference clock, the Tx UI recovered from 3500 consecutive UI must be used as a reference for the eye diagram.
4. A $T_{RX-EYE} = 0.40$ UI provides for a total sum of 0.60 UI deterministic and random jitter budget for the transmitter and interconnect collected any 250 consecutive UIs. The $T_{RX-EYE-MEDIAN-to-MAX-JITTER}$ specification ensures a jitter distribution in which the median and the maximum deviation from the median is less than half of the total. UI jitter budget collected over any 250 consecutive Tx UIs. It should be noted that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value. If the clocks to the Rx and Tx are not derived from the same reference clock, the Tx UI recovered from 3500 consecutive UI must be used as the reference for the eye diagram.
5. Jitter is defined as the measurement variation of the crossing points ($V_{RX-DIFFp-p} = 0$ V) in relation to a recovered Tx UI. A recovered Tx UI is calculated over 3500 consecutive unit intervals of sample data. Jitter is measured using all edges of the 250 consecutive UI in the center of the 3500 UI used for calculating the Tx UI. It is recommended that the recovered Tx UI is calculated using all edges in the 3500 consecutive UI interval with a fit algorithm using a minimization merit function. Least squares and median deviation fits have worked well with experimental and simulated data.

Figure 21 shows the TDM transmit signal timing.

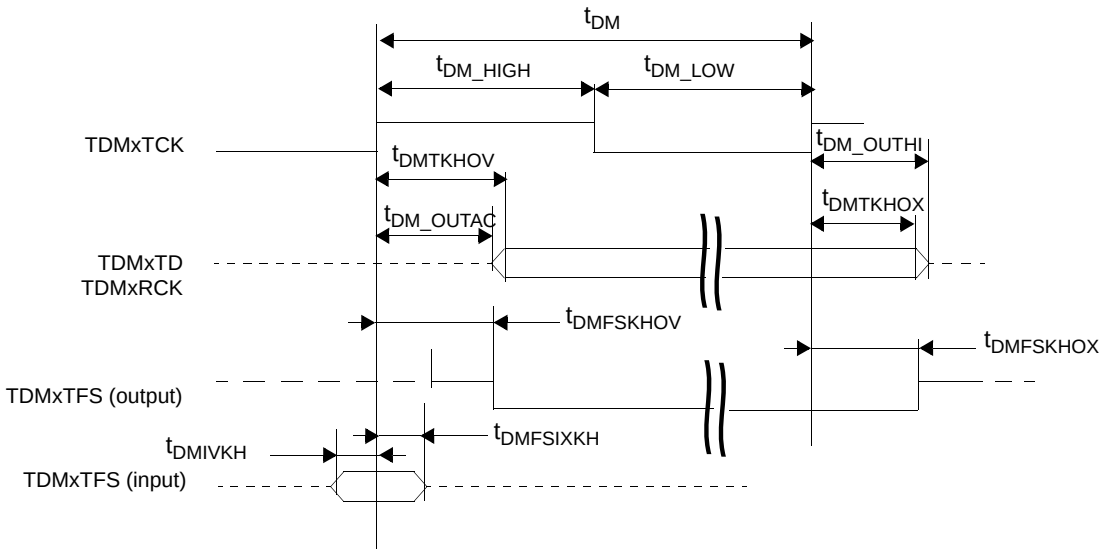


Figure 21. TDM Transmit Signals

Figure 22 provides the AC test load for the TDM/SI.

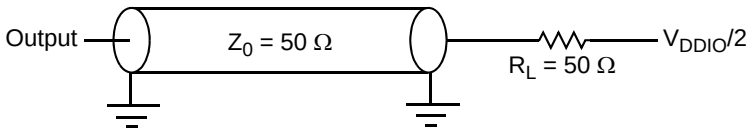


Figure 22. TDM AC Test Load

2.6.4 Timers AC Timing Specifications

Table 32 lists the timer input AC timing specifications.

Table 32. Timers Input AC Timing Specifications

Characteristics	Symbol	Minimum	Unit	Notes
Timers inputs—minimum pulse width	T_{TIWID}	8	ns	1, 2

Notes:
1. The maximum allowed frequency of timer outputs is 125 MHz. Configure the timer modules appropriately.
2. Timer inputs and outputs are asynchronous to any visible clock. Timer outputs should be synchronized before use by any external synchronous logic. Timer inputs are required to be valid for at least t_{TIWID} ns to ensure proper operation.

Note: For recommended operating conditions, see Table 3.

Figure 23 shows the AC test load for the timers.

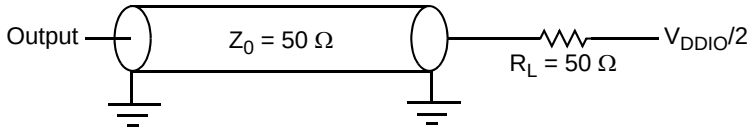


Figure 23. Timer AC Test Load

2.6.5.2 RGMII AC Timing Specifications

Table 34 presents the RGMII AC timing specifications for applications requiring an on-board delayed clock.

Table 34. RGMII at 1 Gbps[2] with On-Board Delay[3] AC Timing Specifications

Parameter/Condition	Symbol	Min	Typ	Max	Unit
Data to clock output skew (at transmitter)[4]	t_{SKEWT}	–-0.5	—	0.5	ns
Data to clock input skew (at receiver) [4]	t_{SKEWR}	1	—	2.6	ns

Notes:
1. At recommended operating conditions with V_{DDIO} of 2.5 V ± 5%.
2. RGMII at 100 Mbps support is guaranteed by design.
3. Program GCR4 as 0x00000000.
4. This implies that PC board design requires clocks to be routed such that an additional trace delay of greater than 1.5 ns and less than 2.0 ns is added to the associated clock signal.

Table 35 presents the RGMII AC timing specification for applications required non-delayed clock on board.

Table 35. RGMII at 1 Gbps[2] with No On-Board Delay[3] AC Timing Specifications

Parameter/Condition	Symbol	Min	Typ	Max	Unit
Data to clock output skew (at transmitter)[4]	t_{SKEWT}	–2.6	—	–1.0	ns
Data to clock input skew (at receiver)[4]	t_{SKEWR}	–0.5	—	0.5	ns

Notes:
1. At recommended operating conditions with V_{DDIO} of 2.5 V ± 5%.
2. RGMII at 100 Mbps support is guaranteed by design.
3. GCR4 should be programmed as 0x000CC330.
4. This implies that PC board design requires clocks to be routed with no additional trace delay

Figure 25 shows the RGMII AC timing and multiplexing diagrams.

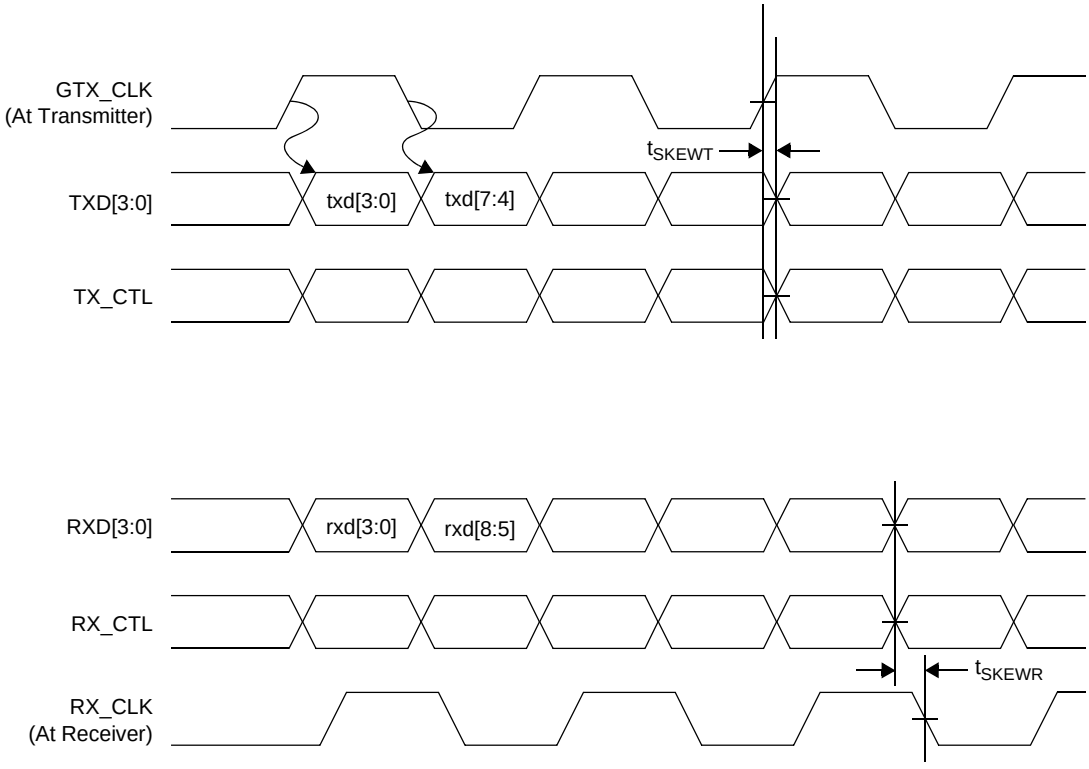


Figure 25. RGMII AC Timing and Multiplexing

2.6.6 SPI Timing

Table 36 lists the SPI input and output AC timing specifications.

Table 36. SPI AC Timing Specifications

Parameter	Symbol [1]	Min	Max	Unit	Note
SPI outputs valid—Master mode (internal clock) delay	t_{NIKHOV}	—	6	ns	2
SPI outputs hold—Master mode (internal clock) delay	t_{NIKHOX}	0.5	—	ns	2
SPI outputs valid—Slave mode (external clock) delay	t_{NEKHOV}	—	12	ns	2
SPI outputs hold—Slave mode (external clock) delay	t_{NEKHOX}	2	—	ns	2
SPI inputs—Master mode (internal clock) input setup time	t_{NIIVKH}	12	—	ns	—
SPI inputs—Master mode (internal clock) input hold time	t_{NIIXKH}	0	—	ns	—
SPI inputs—Slave mode (external clock) input setup time	t_{NEIVKH}	4	—	ns	—
SPI inputs—Slave mode (external clock) input hold time	t_{NEIXKH}	2	—	ns	—

Notes:
1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{NIKHOX} symbolizes the internal timing (NI) for the time SPICLK clock reference (K) goes to the high state (H) until outputs (O) are invalid (X).
2. Output specifications are measured from the 50% level of the rising edge of SPICLK to the 50% level of the signal. Timings are measured at the pin.

Figure 26 provides the AC test load for the SPI.

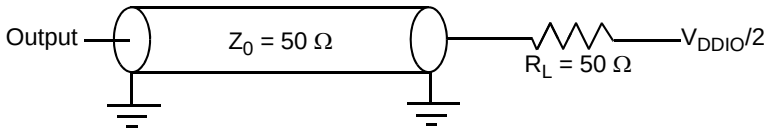
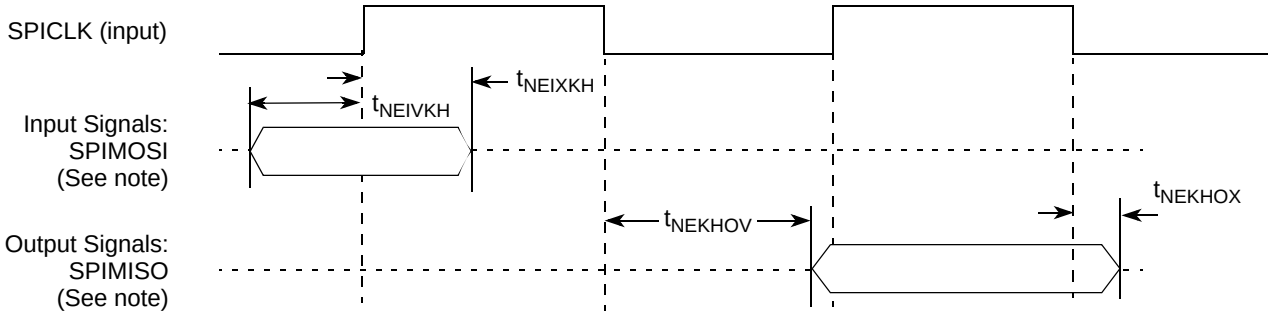


Figure 26. SPI AC Test Load

Figure 27 and Figure 28 represent the AC timings from Table 36. Note that although the specifications generally reference the rising edge of the clock, these AC timing diagrams also apply when the falling edge is the active edge.

Figure 27 shows the SPI timings in slave mode (external clock).



Note: measured with SPMODE[CI] = 0, SPMODE[CP] = 0

Figure 27. SPI AC Timing in Slave Mode (External Clock)

Figure 28 shows the SPI timings in master mode (internal clock).

3.2 PLL Power Supply Design Considerations

Each global PLL power supply must have an external RC filter for the PLLn_AVDD input (see Figure 37) in which the following components are defined as listed:

- R = 5 Ω ± 5%
- C1 = 10 µF ± 10%, 0603, X5R, with ESL ≤ 0.5 nH, low ESL Surface Mount Capacitor.
- C2 = 1.0 µF ± 10%, 0402, X5R, with ESL ≤ 0.5 nH, low ESL Surface Mount Capacitor.

Note: A higher capacitance value for C2 may be used to improve the filter as long as the other C2 parameters do not change.

All three PLLs can connect to a single supply voltage source (such as a voltage regulator) as long as the external RC filter is applied to each PLL separately. For optimal noise filtering, place the circuit as close as possible to its PLLn_AVDD inputs.

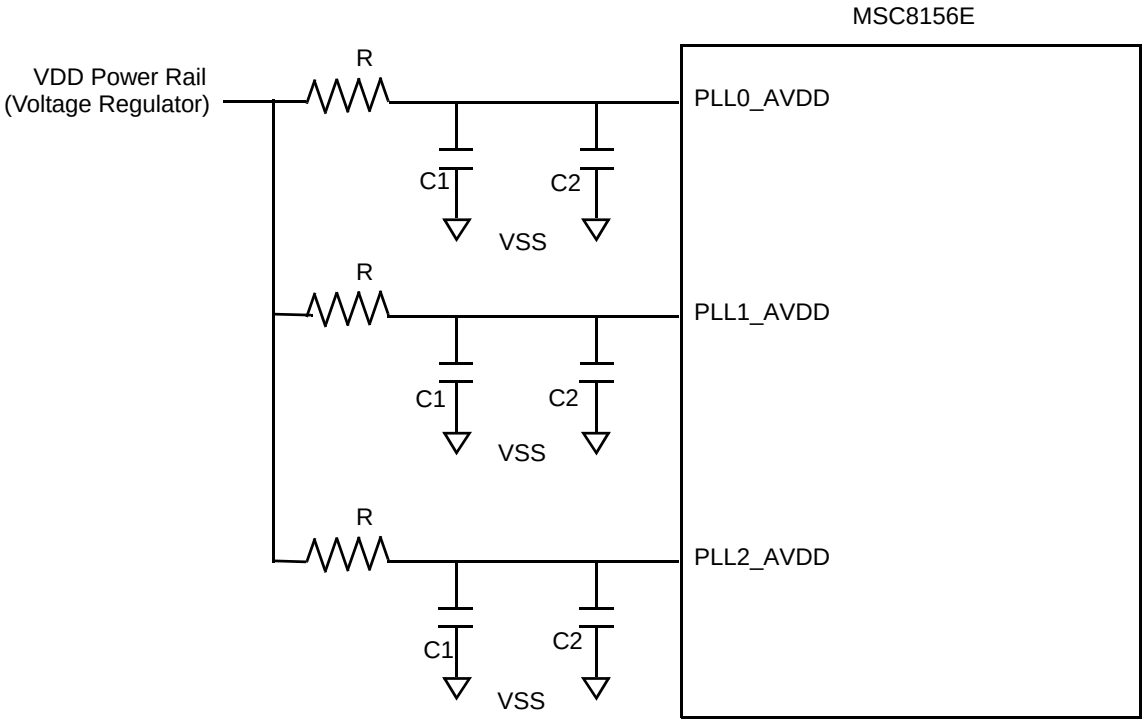


Figure 37. PLL Supplies

Each SerDes PLL power supply must be filtered using a circuit similar to the one shown in Figure 38, to ensure stability of the internal clock. For maximum effectiveness, the filter circuit should be placed as closely as possible to the SRn_PLL_AVDD ball to ensure it filters out as much noise as possible. The ground connection should be near the SRn_PLL_AVDD ball. The 0.003 µF capacitor is closest to the ball, followed by the two 2.2 µF capacitors, and finally the 1 Ω resistor to the board supply plane. The capacitors are connected from SRn_PLL_AVDD to the ground plane. Use ceramic chip capacitors with the highest possible self-resonant frequency. All trances should be kept short, wide, and direct.

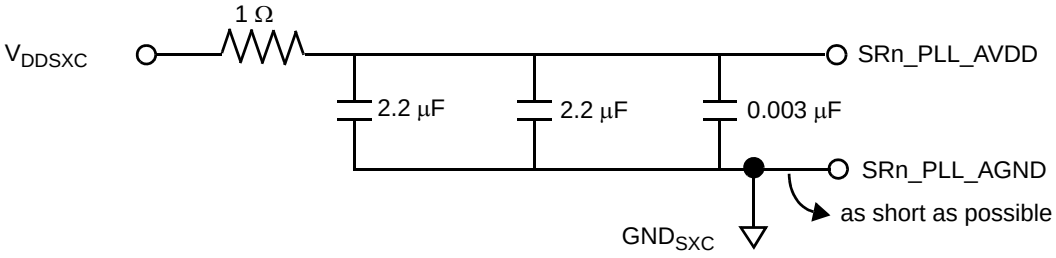


Figure 38. SerDes PLL Supplies

3.3 Clock and Timing Signal Board Layout Considerations

When laying out the system board, use the following guidelines:

- Keep clock and timing signal paths as short as possible and route with 50 Ω impedance.
- Use a serial termination resistor placed close to the clock buffer to minimize signal reflection. Use the following equation to compute the resistor value:

  Rterm = Rim – Rbuf

  where Rim = trace characteristic impedance

  Rbuf = clock buffer internal impedance.

3.4 SGMII AC-Coupled Serial Link Connection Example

Figure 39 shows an example of a 4-wire AC-coupled serial link connection. For additional layout suggestions, see *AN3556 MSC815x High Speed Serial Interface Hardware Design Considerations*, available on the Freescale website or from your local sales office or representative.

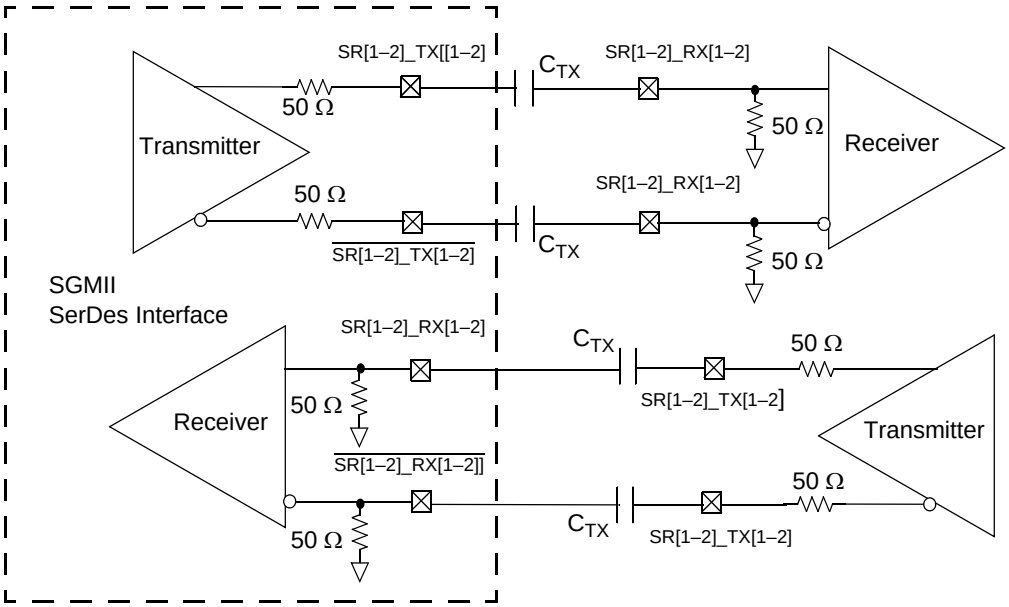


Figure 39. 4-Wire AC-Coupled SGMII Serial Link Connection Example

Table 45. Connectivity of HSSI Related Pins When Specific Lane Is Not Used (continued)

Signal Name	Pin Connection
SR[1–2]_RXD***n***	SXCVSS
$\overline{\text{SR[1–2]\_RXD}\textit{n}}$	SXCVSS
$\overline{\text{SR[1–2]\_TXD}\textit{n}}$	NC
SR[1–2]_TXD***n***	NC
SR[1–2]_PLL_AVDD	in use
SR[1–2]_PLL_AGND	in use
SXPVSS	in use
SXCVSS	in use
SXPVDD	in use
SXCVDD	in use
Note: The ***n*** indicates the lane number {0,1,2,3} for all unused lanes.	

3.5.3 RGMII Ethernet Related Pins

Note: Table 46 and Table 47 assume that the alternate function of the specified pin is not used. If the alternate function is used, connect the pin as required to support that function.

Table 46. Connectivity of RGMII Related Pins When the RGMII Interface Is Not Used

Signal Name	Pin Connection
GE1_RX_CTL	GND
GE2_TX_CTL	NC
Note: Assuming GE1 and GE2 are disabled in the reset configuration word.	

GE_MDC and GE_MDIO pins should be connected as required by the specified protocol. If neither GE1 nor GE2 is used, Table 47 lists the recommended management pin connections.

Table 47. Connectivity of GE Management Pins When GE1 and GE2 Are Not Used

Signal Name	Pin Connection
GE_MDC	NC
GE_MDIO	NC

3.5.4 TDM Interface Related Pins

Table 48 lists the board connections of the TDM pins when an entire specific TDM is not used. For multiplexing options that select a subset of a TDM interface, use the connections described in Table 48 for those signals that are not selected. Table 48 assumes that the alternate function of the specified pin is not used. If the alternate function is used, connect that pin as required to support the selected function.

Table 48. Connectivity of TDM Related Pins When TDM Interface Is Not Used

Signal Name	Pin Connection
TDM***n***RCLK	GND
TDM***n***RDAT	GND
TDM***n***RSYN	GND

Table 48. Connectivity of TDM Related Pins When TDM Interface Is Not Used

Signal Name	Pin Connection
TDM***n***TCLK	GND
TDMT***n***DAT	GND
TDM***n***TSYN	GND
V_{DDIO}	2.5 V

Notes: 1. ***n*** = {0, 1, 2,3}
2. In case of subset of TDM interface usage please make sure to disable unused TDM modules. See *TDM* chapter in the ***MSC8154 Reference Manual*** for details.

3.5.5 Miscellaneous Pins

Table 49 lists the board connections for the pins not required by the system design. Table 49 assumes that the alternate function of the specified pin is not used. If the alternate function is used, connect that pin as required to support the selected function.

Table 49. Connectivity of Individual Pins When They Are Not Required

Signal Name	Pin Connection
CLKOUT	NC
EE0	GND
EE1	NC
GPIO[31–0]	NC
SCL	See the GPIO connectivity guidelines in this table.
SDA	See the GPIO connectivity guidelines in this table.
$\overline{\text{INT\_OUT}}$	NC
$\overline{\text{IRQ}}$[15–0]	See the GPIO connectivity guidelines in this table.
$\overline{\text{NMI}}$	V_{DDIO}
$\overline{\text{NMI\_OUT}}$	NC
RC[21–0]	GND
STOP_BS	GND
TCK	GND
TDI	GND
TDO	NC
TMR[4–0]	See the GPIO connectivity guidelines in this table.
TMS	GND
$\overline{\text{TRST}}$	See **Section 3.1** for guidelines.
URXD	See the GPIO connectivity guidelines in this table.
UTXD	See the GPIO connectivity guidelines in this table.
DDN[1–0]	See the GPIO connectivity guidelines in this table.
DRQ[1–0]	See the GPIO connectivity guidelines in this table.
$\overline{\text{RCW\_LSEL\_0}}$	GND
$\overline{\text{RCW\_LSEL\_1}}$	GND
$\overline{\text{RCW\_LSEL\_2}}$	GND
$\overline{\text{RCW\_LSEL\_3}}$	GND
V_{DDIO}	2.5 V

Note: For details on configuration, see the *MSC8154 Reference Manual*. For additional information, refer to the *MSC815x and MSC825x DSP Family Design Checklist*.