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### What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

## Applications of "<u>Embedded -</u> <u>Microcontrollers</u>"

### Details

Product Status	Obsolete
Core Processor	AVR
Core Size	8-Bit
Speed	16MHz
Connectivity	I <sup>2</sup> C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	23
Program Memory Size	4KB (2K x 16)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	512 x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 8x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	32-VFQFN Exposed Pad
Supplier Device Package	32-QFN (5x5)
Purchase URL	https://www.e-xfl.com/product-detail/atmel/atmega48-15mz

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

The AVR<sup>®</sup> core combines a rich instruction set with 32 general purpose working registers. All the 32 registers are directly connected to the arithmetic logic unit (ALU), allowing two independent registers to be accessed in one single instruction executed in one clock cycle. The resulting architecture is more code efficient while achieving throughputs up to ten times faster than conventional CISC microcontrollers.

The Atmel<sup>®</sup> ATmega48/88/168 provides the following features: 4K/8K/16Kbytes of in-system programmable flash with read-while-write capabilities, 256/512/512 bytes EEPROM, 512/1K/1Kbytes SRAM, 23 general purpose I/O lines, 32 general purpose working registers, three flexible Timer/Counters with compare modes, internal and external interrupts, a serial programmable USART, a byte-oriented 2-wire serial interface, an SPI serial port, a 6-channel 10-bit ADC (8 channels in TQFP and QFN packages), a programmable watchdog timer with internal oscillator, and five software selectable power saving modes. The idle mode stops the CPU while allowing the SRAM, Timer/Counters, USART, 2-wire serial interface, SPI port, and interrupt system to continue functioning. The power-down mode saves the register contents but freezes the oscillator, disabling all other chip functions until the next interrupt or hardware reset. In power-save mode, the asynchronous timer continues to run, allowing the user to maintain a timer base while the rest of the device is sleeping. The ADC noise reduction mode stops the CPU and all I/O modules except asynchronous timer and ADC, to minimize switching noise during ADC conversions. In Standby mode, the crystal/resonator Oscillator is running while the rest of the device is sleeping. This allows very fast start-up combined with low power consumption.

The device is manufactured using Atmel's high density non-volatile memory technology. The on-chip ISP flash allows the program memory to be reprogrammed in-system through an SPI serial interface, by a conventional non-volatile memory programmer, or by an on-chip boot program running on the AVR core. The boot program can use any interface to download the application program in the application flash memory. Software in the boot flash section will continue to run while the application flash section is updated, providing true read-while-write operation. By combining an 8-bit RISC CPU with in-system self-programmable flash on a monolithic chip, the Atmel ATmega48/88/168 is a powerful microcontroller that provides a highly flexible and cost effective solution to many embedded control applications.

The Atmel ATmega48/88/168 AVR is supported with a full suite of program and system development tools including: C compilers, macro assemblers, program debugger/simulators, in-circuit emulators, and evaluation kits.

## 2.2 Automotive Quality Grade

The ATmega48-15AZ, ATmega88-15AZ and ATmega168-15AZ have been developed and manufactured according to the most stringent requirements of the international standard ISO-TS-16949 grade 1. This data sheet contains limit values extracted from the results of extensive characterization (temperature and voltage). The quality and reliability of the ATmega48-15AZ, ATmega88-15AZ and ATmega168-15AZ have been verified during regular product qualification as per AEC-Q100.

As indicated in the ordering information paragraph (see Section 31. "Ordering Information" on page 296), the products are available in three different temperature grades, but with equivalent quality and reliability objectives. Different temperature identifiers have been defined as listed in Table 2-1.

Temperature	Temperature Identifier	Comments
-40; +85	Т	Similar to industrial temperature grade but with automotive quality
-40; +105	T1	Reduced automotive temperature range
-40; +125	Z	Full automotive temperature range

Table 2-1. Temperature Grade Identification for Automotive Products

# 3. About Code Examples

This documentation contains simple code examples that briefly show how to use various parts of the device. These code examples assume that the part specific header file is included before compilation. Be aware that not all C compiler vendors include bit definitions in the header files and interrupt handling in C is compiler dependent. Please confirm with the C compiler documentation for more details.

# 4. AVR CPU Core

## 4.1 Introduction

This section discusses the AVR<sup>®</sup> core architecture in general. The main function of the CPU core is to ensure correct program execution. The CPU must therefore be able to access memories, perform calculations, control peripherals, and handle interrupts.

## 4.2 Architectural Overview







In order to maximize performance and parallelism, the AVR<sup>®</sup> uses a Harvard architecture – with separate memories and buses for program and data. Instructions in the program memory are executed with a single level pipelining. While one instruction is being executed, the next instruction is pre-fetched from the program memory. This concept enables instructions to be executed in every clock cycle. The program memory is in-system reprogrammable flash memory.

The fast-access register file contains  $32 \times 8$ -bit general purpose working registers with a single clock cycle access time. This allows single-cycle arithmetic logic unit (ALU) operation. In a typical ALU operation, two operands are output from the register file, the operation is executed, and the result is stored back in the register file – in one clock cycle.

Six of the 32 registers can be used as three 16-bit indirect address register pointers for data space addressing – enabling efficient address calculations. One of the these address pointers can also be used as an address pointer for look up tables in flash program memory. These added function registers are the 16-bit X-, Y-, and Z-register, described later in this section.

The ALU supports arithmetic and logic operations between registers or between a constant and a register. Single register operations can also be executed in the ALU. After an arithmetic operation, the status register is updated to reflect information about the result of the operation.

Program flow is provided by conditional and unconditional jump and call instructions, able to directly address the whole address space. Most AVR instructions have a single 16-bit word format. Every program memory address contains a 16- or 32-bit instruction.

Program flash memory space is divided in two sections, the boot program section and the application program section. Both sections have dedicated lock bits for write and read/write protection. The SPM instruction that writes into the application flash memory section must reside in the boot program section.

During interrupts and subroutine calls, the return address program counter (PC) is stored on the stack. The stack is effectively allocated in the general data SRAM, and consequently the stack size is only limited by the total SRAM size and the usage of the SRAM. All user programs must initialize the SP in the reset routine (before subroutines or interrupts are executed). The stack pointer (SP) is read/write accessible in the I/O space. The data SRAM can easily be accessed through the five different addressing modes supported in the AVR architecture.

The memory spaces in the AVR architecture are all linear and regular memory maps.

A flexible interrupt module has its control registers in the I/O space with an additional global interrupt enable bit in the status register. All interrupts have a separate interrupt vector in the interrupt vector table. The interrupts have priority in accordance with their interrupt vector position. The lower the interrupt vector address, the higher the priority.

The I/O memory space contains 64 addresses for CPU peripheral functions as control registers, SPI, and other I/O functions. The I/O memory can be accessed directly, or as the data space locations following those of the register file, 0x20 - 0x5F. In addition, the ATmega48/88/168 has extended I/O space from 0x60 - 0xFF in SRAM where only the ST/STS/STD and LD/LDS/LDD instructions can be used.

## 4.3 ALU – Arithmetic Logic Unit

The high-performance AVR ALU operates in direct connection with all the 32 general purpose working registers. Within a single clock cycle, arithmetic operations between general purpose registers or between a register and an immediate are executed. The ALU operations are divided into three main categories – arithmetic, logical, and bit-functions. Some implementations of the architecture also provide a powerful multiplier supporting both signed/unsigned multiplication and fractional format. See the "Instruction Set" section for a detailed description.

## 8.7 MCU Status Register – MCUSR

The MCU status register provides information on which reset source caused an MCU reset.



#### • Bit 7..4: Res: Reserved Bits

These bits are unused bits in the Atmel® ATmega48/88/168, and will always read as zero.

### • Bit 3 – WDRF: Watchdog System Reset Flag

This bit is set if a watchdog system reset occurs. The bit is reset by a power-on reset, or by writing a logic zero to the flag.

#### • Bit 2 - BORF: Brown-out Reset Flag

This bit is set if a brown-out reset occurs. The bit is reset by a power-on reset, or by writing a logic zero to the flag.

#### Bit 1 – EXTRF: External Reset Flag

This bit is set if an external reset occurs. The bit is reset by a power-on reset, or by writing a logic zero to the flag.

### • Bit 0 – PORF: Power-on Reset Flag

This bit is set if a power-on reset occurs. The bit is reset only by writing a logic zero to the flag.

To make use of the reset flags to identify a reset condition, the user should read and then reset the MCUSR as early as possible in the program. If the register is cleared before another reset occurs, the source of the reset can be found by examining the reset flags.

### 8.8 Internal Voltage Reference

Atmel ATmega48/88/168 features an internal bandgap reference. This reference is used for brown-out detection, and it can be used as an input to the analog comparator or the ADC.

### 8.8.1 Voltage Reference Enable Signals and Start-up Time

The voltage reference has a start-up time that may influence the way it should be used. The start-up time is given in Table 8-4. To save power, the reference is not always turned on. The reference is on during the following situations:

- 1. When the BOD is enabled (by programming the BODLEVEL [2..0] fuses).
- 2. When the bandgap reference is connected to the analog comparator (by setting the ACBG bit in ACSR).
- 3. When the ADC is enabled.

Thus, when the BOD is not enabled, after setting the ACBG bit or enabling the ADC, the user must always allow the reference to start up before the output from the analog comparator or ADC is used. To reduce power consumption in power-down mode, the user can avoid the three conditions above to ensure that the reference is turned off before entering power-down mode.

### Table 8-4. Internal Voltage Reference Characteristics<sup>(1)</sup>

Parameter	Condition	Symbol	Min	Тур	Max	Unit
Bandgap reference voltage	TBD	V <sub>BG</sub>	1.0	1.1	1.2	V
Bandgap reference start-up time	TBD	t <sub>BG</sub>		40	70	μs
Bandgap reference current consumption	TBD	I <sub>BG</sub>		10	TBD	μA

Note: 1. Values are guidelines only. Actual values are TBD.

## 10.2 Ports as General Digital I/O

The ports are bi-directional I/O ports with optional internal pull-ups. Figure 10-2 shows a functional description of one I/O-port pin, here generically called Pxn.





Note: 1. WRx, WPx, WDx, RRx, RPx, and RDx are common to all pins within the same port. clk<sub>I/O</sub>, SLEEP, and PUD are common to all ports.

### 10.2.1 Configuring the Pin

Each port pin consists of three register bits: DDxn, PORTxn, and PINxn. As shown in Section 10.4 "Register Description for I/O Ports" on page 71, the DDxn bits are accessed at the DDRx I/O address, the PORTxn bits at the PORTx I/O address, and the PINxn bits at the PINx I/O address.

The DDxn bit in the DDRx register selects the direction of this pin. If DDxn is written logic one, Pxn is configured as an output pin. If DDxn is written logic zero, Pxn is configured as an input pin.

If PORTxn is written logic one when the pin is configured as an input pin, the pull-up resistor is activated. To switch the pull-up resistor off, PORTxn has to be written logic zero or the pin has to be configured as an output pin. The port pins are tri-stated when reset condition becomes active, even if no clocks are running.

If PORTxn is written logic one when the pin is configured as an output pin, the port pin is driven high (one). If PORTxn is written logic zero when the pin is configured as an output pin, the port pin is driven low (zero).



### • SCK/PCINT5 - Port B, Bit 5

SCK: master clock output, slave clock input pin for SPI channel. When the SPI is enabled as a slave, this pin is configured as an input regardless of the setting of DDB5. When the SPI is enabled as a master, the data direction of this pin is controlled by DDB5. When the pin is forced by the SPI to be an input, the pull-up can still be controlled by the PORTB5 bit.

PCINT5: pin change interrupt source 5. The PB5 pin can serve as an external interrupt source.

### • MISO/PCINT4 - Port B, Bit 4

MISO: master data input, slave data output pin for SPI channel. When the SPI is enabled as a master, this pin is configured as an input regardless of the setting of DDB4. When the SPI is enabled as a slave, the data direction of this pin is controlled by DDB4. When the pin is forced by the SPI to be an input, the pull-up can still be controlled by the PORTB4 bit.

PCINT4: pin change interrupt source 4. The PB4 pin can serve as an external interrupt source.

### • MOSI/OC2/PCINT3 – Port B, Bit 3

MOSI: SPI master data output, slave data input for SPI channel. When the SPI is enabled as a slave, this pin is configured as an input regardless of the setting of DDB3. When the SPI is enabled as a master, the data direction of this pin is controlled by DDB3. When the pin is forced by the SPI to be an input, the pull-up can still be controlled by the PORTB3 bit.

OC2, output compare match output: The PB3 pin can serve as an external output for the Timer/Counter2 compare match. The PB3 pin has to be configured as an output (DDB3 set (one)) to serve this function. The OC2 pin is also the output pin for the PWM mode timer function.

PCINT3: pin change interrupt source 3. The PB3 pin can serve as an external interrupt source.

## • SS/OC1B/PCINT2 – Port B, Bit 2

 $\overline{SS}$ : slave select input. When the SPI is enabled as a slave, this pin is configured as an input regardless of the setting of DDB2. As a slave, the SPI is activated when this pin is driven low. When the SPI is enabled as a master, the data direction of this pin is controlled by DDB2. When the pin is forced by the SPI to be an input, the pull-up can still be controlled by the PORTB2 bit.

OC1B, output compare match output: The PB2 pin can serve as an external output for the Timer/Counter1 compare match B. The PB2 pin has to be configured as an output (DDB2 set (one)) to serve this function. The OC1B pin is also the output pin for the PWM mode timer function.

PCINT2: pin change interrupt source 2. The PB2 pin can serve as an external interrupt source.

## • OC1A/PCINT1 - Port B, Bit 1

OC1A, output compare match output: The PB1 pin can serve as an external output for the Timer/Counter1 compare match A. The PB1 pin has to be configured as an output (DDB1 set (one)) to serve this function. The OC1A pin is also the output pin for the PWM mode timer function.

PCINT1: pin change interrupt source 1. The PB1 pin can serve as an external interrupt source.

### • ICP1/CLKO/PCINT0 - Port B, Bit 0

ICP1, input capture pin: The PB0 pin can act as an input capture pin for Timer/Counter1.

CLKO, divided system clock: The divided system clock can be output on the PB0 pin. The divided system clock will be output if the CKOUT fuse is programmed, regardless of the PORTB0 and DDB0 settings. It will also be output during reset.

PCINT0: pin change interrupt source 0. The PB0 pin can serve as an external interrupt source.

Table 10-4 and Table 10-5 relate the alternate functions of port B to the overriding signals shown in Figure 10-5 on page 62. SPI MSTR INPUT and SPI SLAVE OUTPUT constitute the MISO signal, while MOSI is divided into SPI MSTR OUTPUT and SPI SLAVE INPUT.

Signal Name	PB7/XTAL2/TOSC2/PCINT7 <sup>(1)</sup>	PB6/XTAL1/TOSC1/PCINT6 <sup>(1)</sup>	PB5/SCK/PCINT5	PB4/MISO/PCINT4
PUOE	$\overline{\text{INTRC}} \times \overline{\text{EXTCK}} + \text{AS2}$	INTRC + AS2	$SPE \times \overline{MSTR}$	$SPE \times MSTR$
PUOV	0	0	$PORTB5 \times \overline{PUD}$	$PORTB4 \times \overline{PUD}$
DDOE	$\overline{\text{INTRC}} \times \overline{\text{EXTCK}} + \text{AS2}$	INTRC + AS2	$SPE \times \overline{MSTR}$	$SPE \times MSTR$
DDOV	0	0	0	0
PVOE	0	0	SPE × MSTR	$SPE \times \overline{MSTR}$
PVOV	0	0	SCK OUTPUT	SPI SLAVE OUTPUT
DIEOE	$\overline{\text{INTRC}} \times \overline{\text{EXTCK}} + \text{AS2} + \text{PCINT7} \times \text{PCIE0}$	INTRC + AS2 + PCINT6 × PCIE0	PCINT5 × PCIE0	$PCINT4 \times PCIE0$
DIEOV	(INTRC + EXTCK) $\times$ AS2	$INTRC \times \overline{AS2}$	1	1
DI	PCINT7 INPUT	PCINT6 INPUT	PCINT5 INPUT SCK INPUT	PCINT4 INPUT SPI MSTR INPUT
AIO	Oscillator output	Oscillator/clock input	-	-

Table 10-4. Overholding Signals for Alternate Functions in PB7PD4	Table 10-4.	Overriding	Signals for	Alternate	Functions	in PB7PB4
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Note: 1. INTRC means that one of the internal RC oscillators are selected (by the CKSEL fuses), EXTCK means that external clock is selected (by the CKSEL fuses).

### Table 10-5. Overriding Signals for Alternate Functions in PB3..PB0

Signal Name	PB3/MOSI/OC2/PCINT3	PB2/SS/OC1B/PCINT2	PB1/OC1A/PCINT1	PB0/ICP1/PCINT0
PUOE	$SPE \times \overline{MSTR}$	$SPE \times \overline{MSTR}$	0	0
PUOV	$PORTB3 \times \overline{PUD}$	$PORTB2 \times \overline{PUD}$	0	0
DDOE	$SPE \times \overline{MSTR}$	$SPE \times \overline{MSTR}$	0	0
DDOV	0	0	0	0
PVOE	SPE × MSTR + OC2A ENABLE	OC1B ENABLE	OC1A ENABLE	0
PVOV	SPI MSTR OUTPUT + OC2A	OC1B	OC1A	0
DIEOE	PCINT3 × PCIE0	$PCINT2 \times PCIE0$	PCINT1 × PCIE0	PCINT0 × PCIE0
DIEOV	1	1	1	1
DI	PCINT3 INPUT SPI SLAVE INPUT	PCINT2 INPUT SPI SS	PCINT1 INPUT	PCINT0 INPUT ICP1 INPUT
AIO	-	-	-	-

## 10.3.4 Alternate Functions of Port D

The port D pins with alternate functions are shown in Table 10-9.

Port Pin	Alternate Function
PD7	AIN1 (analog comparator negative input) PCINT23 (pin change interrupt 23)
PD6	AIN0 (analog comparator positive input) OC0A (Timer/Counter0 output compare match A output) PCINT22 (pin change interrupt 22)
PD5	T1 (Timer/Counter 1 external counter input) OC0B (Timer/Counter0 output compare match B output) PCINT21 (pin change interrupt 21)
PD4	XCK (USART external clock input/output) T0 (Timer/Counter 0 external counter input) PCINT20 (pin change interrupt 20)
PD3	INT1 (external interrupt 1 input) OC2B (Timer/Counter2 output compare match B output) PCINT19 (pin change interrupt 19)
PD2	INT0 (external interrupt 0 input) PCINT18 (pin change interrupt 18)
PD1	TXD (USART output pin) PCINT17 (pin change interrupt 17)
PD0	RXD (USART input pin) PCINT16 (pin change interrupt 16)

Table 10-9.	Port D Pins	Alternate	Functions

The alternate pin configuration is as follows:

## • AIN1/OC2B/PCINT23 - Port D, Bit 7

AIN1, analog comparator negative input. Configure the port pin as input with the internal pull-up switched off to avoid the digital port function from interfering with the function of the analog comparator.

PCINT23: pin change interrupt source 23. The PD7 pin can serve as an external interrupt source.

## • AIN0/OC0A/PCINT22 - Port D, Bit 6

AINO, analog comparator positive input. Configure the port pin as input with the internal pull-up switched off to avoid the digital port function from interfering with the function of the analog comparator.

OC0A, output compare match output: The PD6 pin can serve as an external output for the Timer/Counter0 compare match A. The PD6 pin has to be configured as an output (DDD6 set (one)) to serve this function. The OC0A pin is also the output pin for the PWM mode timer function.

PCINT22: pin change interrupt source 22. The PD6 pin can serve as an external interrupt source.

## • T1/OC0B/PCINT21 - Port D, Bit 5

T1, Timer/Counter1 counter source.

OC0B, output compare match output: The PD5 pin can serve as an external output for the Timer/Counter0 compare match B. The PD5 pin has to be configured as an output (DDD5 set (one)) to serve this function. The OC0B pin is also the output pin for the PWM mode timer function.

PCINT21: pin change interrupt source 21. The PD5 pin can serve as an external interrupt source.

# 12. 8-bit Timer/Counter0 with PWM

Timer/Counter0 is a general purpose 8-bit Timer/Counter module, with two independent output compare units, and with PWM support. It allows accurate program execution timing (event management) and wave generation. The main features are:

- Two independent output compare units
- Double buffered output compare registers
- Clear timer on compare match (auto reload)
- Glitch free, phase correct pulse width modulator (PWM)
- Variable PWM period
- Frequency generator
- Three independent interrupt sources (TOV0, OCF0A, and OCF0B)

## 12.1 Overview

A simplified block diagram of the 8-bit Timer/Counter is shown in Figure 12-1. For the actual placement of I/O pins, refer to Section 1-1 "Pinout ATmega48/88/168" on page 3. CPU accessible I/O registers, including I/O bits and I/O pins, are shown in bold. The device-specific I/O register and bit locations are listed in Section 12.8 "8-bit Timer/Counter Register Description" on page 87.

The PRTIM0 bit in Section 7.7.1 "Power Reduction Register - PRR" on page 35 must be written to zero to enable Timer/Counter0 module.

## Figure 12-1. 8-bit Timer/Counter Block Diagram



## 12.1.1 Definitions

Many register and bit references in this section are written in general form. A lower case "n" replaces the Timer/Counter number, in this case 0. A lower case "x" replaces the output compare unit, in this case compare unit A or compare unit B. However, when using the register or bit defines in a program, the precise form must be used, i.e., TCNT0 for accessing Timer/Counter0 counter value and so on.

The definitions in the following table are also used extensively throughout the document.

Table 12-1. D	Definitions
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Parameters	Definitions
BOTTOM	The counter reaches the BOTTOM when it becomes 0x00.
MAX	The counter reaches its MAXimum when it becomes 0xFF (decimal 255).
ТОР	The counter reaches the TOP when it becomes equal to the highest value in the count sequence. The TOP value can be assigned to be the fixed value 0xFF (MAX) or the value stored in the OCR0A register. The assignment is dependent on the mode of operation.

### 12.1.2 Registers

The Timer/Counter (TCNT0) and output compare registers (OCR0A and OCR0B) are 8-bit registers. Interrupt request (abbreviated to int.req. in the figure) signals are all visible in the timer interrupt flag register (TIFR0). All interrupts are individually masked with the timer interrupt mask register (TIMSK0). TIFR0 and TIMSK0 are not shown in the figure.

The Timer/Counter can be clocked internally, via the prescaler, or by an external clock source on the T0 pin. The clock select logic block controls which clock source and edge the Timer/Counter uses to increment (or decrement) its value. The Timer/Counter is inactive when no clock source is selected. The output from the clock select logic is referred to as the timer clock ( $clk_{T0}$ ).

The double buffered output compare registers (OCR0A and OCR0B) are compared with the Timer/Counter value at all times. The result of the compare can be used by the waveform generator to generate a PWM or variable frequency output on the output compare pins (OC0A and OC0B). See Section 14.6.3 "Using the Output Compare Unit" on page 104 for details. The compare match event will also set the compare flag (OCF0A or OCF0B) which can be used to generate an output compare interrupt request.

## 12.2 Timer/Counter Clock Sources

The Timer/Counter can be clocked by an internal or an external clock source. The clock source is selected by the clock select logic which is controlled by the clock select (CS02:0) bits located in the Timer/Counter control register (TCCR0B). For details on clock sources and prescaler, see Section 13. "Timer/Counter0 and Timer/Counter1 Prescalers" on page 92.

## 12.3 Counter Unit

The main part of the 8-bit Timer/Counter is the programmable bi-directional counter unit. Figure 12-2 shows a block diagram of the counter and its surroundings.



### Figure 12-2. Counter Unit Block Diagram



## 14.8.5 Phase and Frequency Correct PWM Mode

The phase and frequency correct pulse width modulation, or phase and frequency correct PWM mode (WGM13:0 = 8 or 9) provides a high resolution phase and frequency correct PWM waveform generation option. The phase and frequency correct PWM mode is, like the phase correct PWM mode, based on a dual-slope operation. The counter counts repeatedly from BOTTOM (0x0000) to TOP and then from TOP to BOTTOM. In non-inverting compare output mode, the output compare (OC1x) is cleared on the compare match between TCNT1 and OCR1x while upcounting, and set on the compare match while downcounting. In inverting compare output mode, the operation is inverted. The dual-slope operation gives a lower maximum operation frequency compared to the single-slope operation. However, due to the symmetric feature of the dual-slope PWM modes, these modes are preferred for motor control applications.

The main difference between the phase correct, and the phase and frequency correct PWM mode is the time the OCR1x register is updated by the OCR1x buffer register, (see Figure 14-8 on page 109 and Figure 14-9).

The PWM resolution for the phase and frequency correct PWM mode can be defined by either ICR1 or OCR1A. The minimum resolution allowed is 2-bit (ICR1 or OCR1A set to 0x0003), and the maximum resolution is 16-bit (ICR1 or OCR1A set to MAX). The PWM resolution in bits can be calculated using the following equation:

$$R_{PFCPWM} = \frac{\log(TOP + 1)}{\log(2)}$$

In phase and frequency correct PWM mode the counter is incremented until the counter value matches either the value in ICR1 (WGM13:0 = 8), or the value in OCR1A (WGM13:0 = 9). The counter has then reached the TOP and changes the count direction. The TCNT1 value will be equal to TOP for one timer clock cycle. The timing diagram for the phase correct and frequency correct PWM mode is shown on Figure 14-9. The figure shows phase and frequency correct PWM mode when OCR1A or ICR1 is used to define TOP. The TCNT1 value is in the timing diagram shown as a histogram for illustrating the dual-slope operation. The diagram includes non-inverted and inverted PWM outputs. The small horizontal line marks on the TCNT1 slopes represent compare matches between OCR1x and TCNT1. The OC1x interrupt flag will be set when a compare match occurs.





The Timer/Counter overflow flag (TOV1) is set at the same timer clock cycle as the OCR1x registers are updated with the double buffer value (at BOTTOM). When either OCR1A or ICR1 is used for defining the TOP value, the OC1A or ICF1 flag set when TCNT1 has reached TOP. The interrupt flags can then be used to generate an interrupt each time the counter reaches the TOP or BOTTOM value.

# 19. 2-wire Serial Interface

## 19.1 Features

- Simple yet powerful and flexible communication interface, only two bus lines needed
- Both master and slave operation supported
- Device can operate as transmitter or receiver
- 7-bit address space allows up to 128 different slave addresses
- Multi-master arbitration support
- Up to 400kHz data transfer speed
- Slew-rate limited output drivers
- Noise suppression circuitry rejects spikes on bus lines
- Fully programmable slave address with general call support
- Address recognition causes wake-up when AVR® is in sleep mode

## 19.2 2-wire Serial Interface Bus Definition

The 2-wire serial interface (TWI) is ideally suited for typical microcontroller applications. The TWI protocol allows the systems designer to interconnect up to 128 different devices using only two bi-directional bus lines, one for clock (SCL) and one for data (SDA). The only external hardware needed to implement the bus is a single pull-up resistor for each of the TWI bus lines. All devices connected to the bus have individual addresses, and mechanisms for resolving bus contention are inherent in the TWI protocol.

### Figure 19-1. TWI Bus Interconnection



## 19.2.1 TWI Terminology

The following definitions are frequently encountered in this section.

### Table 19-1. TWI Terminology

Term	Description
Master	The device that initiates and terminates a transmission. The Master also generates the SCL clock.
Slave	The device addressed by a Master.
Transmitter	The device placing data on the bus.
Receiver	The device reading data from the bus.

The PRTWI bit in Section 7.7.1 "Power Reduction Register - PRR" on page 35 must be written to zero to enable the 2-wire serial interface.

Figure 19-15. Formats and States in the Master Receiver Mode



#### 19.8.3 Slave Receiver Mode

In the slave receiver mode, a number of data bytes are received from a master transmitter (see Figure 19-16). All the status codes mentioned in this section assume that the prescaler bits are zero or are masked to zero.







## 19.8.6 Combining Several TWI Modes

In some cases, several TWI modes must be combined in order to complete the desired action. Consider for example reading data from a serial EEPROM. Typically, such a transfer involves the following steps:

- 1. The transfer must be initiated.
- 2. The EEPROM must be instructed what location should be read.
- 3. The reading must be performed.
- 4. The transfer must be finished.

Note that data is transmitted both from master to slave and vice versa. The master must instruct the slave what location it wants to read, requiring the use of the MT mode. Subsequently, data must be read from the slave, implying the use of the MR mode. Thus, the transfer direction must be changed. The master must keep control of the bus during all these steps, and the steps should be carried out as an atomical operation. If this principle is violated in a multi master system, another master can alter the data pointer in the EEPROM between steps 2 and 3, and the master will read the wrong data location. Such a change in transfer direction is accomplished by transmitting a REPEATED START between the transmission of the address byte and reception of the data. After a REPEATED START, the master keeps ownership of the bus. The following figure shows the flow in this transfer.

### Figure 19-20. Combining Several TWI Modes to Access a Serial EEPROM



## 19.9 Multi-master Systems and Arbitration

If multiple masters are connected to the same bus, transmissions may be initiated simultaneously by one or more of them. The TWI standard ensures that such situations are handled in such a way that one of the masters will be allowed to proceed with the transfer, and that no data will be lost in the process. An example of an arbitration situation is depicted below, where two masters are trying to transmit data to a slave receiver.







Figure 25-9. Parallel Programming Timing, Reading Sequence (within the Same Page) with Timing Requirements<sup>(1)</sup>

Note: 1. The timing requirements shown in Figure 25-7 on page 254 (i.e., t<sub>DVXH</sub>, t<sub>XHXL</sub>, and t<sub>XLDX</sub>) also apply to reading operation.

Table 25-15.	Parallel Programming	Characteristics,	$V_{\rm CC} = 5V \pm 10\%$
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Parameter	Symbol	Min	Тур	Max	Unit
Programming enable voltage	V <sub>PP</sub>	11.5		12.5	V
Programming enable current	I <sub>PP</sub>			250	μA
Data and control valid before XTAL1 high	t <sub>DVXH</sub>	67			ns
XTAL1 low to XTAL1 high	t <sub>xLXH</sub>	200			ns
XTAL1 pulse width high	t <sub>xHXL</sub>	150			ns
Data and control hold after XTAL1 low	t <sub>xLDX</sub>	67			ns
XTAL1 low to WR low	t <sub>xLWL</sub>	0			ns
XTAL1 low to PAGEL high	t <sub>XLPH</sub>	0			ns
PAGEL low to XTAL1 high	t <sub>PLXH</sub>	150			ns
BS1 valid before PAGEL high	t <sub>BVPH</sub>	67			ns
PAGEL pulse width high	t <sub>PHPL</sub>	150			ns
BS1 hold after PAGEL low	t <sub>PLBX</sub>	67			ns
BS2/1 hold after WR low	t <sub>WLBX</sub>	67			ns
PAGEL low to WR low	t <sub>PLWL</sub>	67			ns
BS1 valid to WR low	t <sub>BVWL</sub>	67			ns
WR pulse width low	t <sub>wLWH</sub>	150			ns
WR low to RDY/BSY low	t <sub>WLRL</sub>	0		1	μs
WR low to RDY/BSY high <sup>(1)</sup>	t <sub>wLRH</sub>	3.7		4.5	ms
WR Low to RDY/BSY high for chip erase <sup>(2)</sup>	t <sub>wLRH_CE</sub>	7.5		9	ms
XTAL1 low to OE low	t <sub>xLOL</sub>	0			ns
BS1 valid to DATA valid	t <sub>BVDV</sub>	0		250	ns
OE low to DATA valid	t <sub>OLDV</sub>			250	ns
OE high to DATA tri-stated	t <sub>OHDZ</sub>			250	ns

Notes: 1. t<sub>WLRH</sub> is valid for the write flash, write EEPROM, write fuse bits and write lock bits commands.

2.  $t_{WLRH\_CE}$  is valid for the chip erase command.

# 26. Electrical Characteristics

## 26.1 Absolute Maximum Ratings\*

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or other conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Parameter	Min	Тур	Max	Unit
Operating temperature	-55		+125	°C
Storage temperature	-65		+150	°C
Voltage on any pin except RESET with respect to ground	-0.5		V <sub>CC</sub> + 0.5	V
Voltage on RESET with respect to ground	-0.5		+13.0	V
Maximum operating voltage			6.0	V
DC current per I/O pin			40.0	mA
DC current V <sub>CC</sub> and GND pins			200.0	mA

## 26.2 DC Characteristics

 $T_A = -40^{\circ}C$  to +125°C,  $V_{CC} = 2.7V$  to 5.5V (unless otherwise noted)

Parameter	Condition	Symbol	Min. <sup>(5)</sup>	Тур.	Max. <sup>(5)</sup>	Unit
Input low Voltage, except XTAL1 and reset pin	V <sub>CC</sub> = 2.7V - 5.5V	V <sub>IL</sub>	-0.5		0.3V <sub>CC</sub> <sup>(1)</sup>	V
Input low voltage, XTAL1 pin	V <sub>CC</sub> = 2.7V - 5.5V	V <sub>IL1</sub>	-0.5		0.1V <sub>CC</sub> <sup>(1)</sup>	V
Input low voltage, RESET pin	V <sub>CC</sub> = 2.7V - 5.5V	V <sub>IL2</sub>	-0.5		0.1V <sub>CC</sub> <sup>(1)</sup>	V

Notes: 1. "Max" means the highest value where the pin is guaranteed to be read as low

- 2. "Min" means the lowest value where the pin is guaranteed to be read as high
- Although each I/O port can sink more than the test conditions (20mA at V<sub>CC</sub> = 5V, 10mA at V<sub>CC</sub> = 3V) under steady state conditions (non-transient), the following must be observed: Atmel ATmega48:
  - 1] The sum of all IOL, for ports C0 C5, should not exceed 70mA.
  - 2] The sum of all IOL, for ports C6, D0 D4, should not exceed 70mA.
  - 3] The sum of all IOL, for ports B0 B7, D5 D7, should not exceed 70mA.
  - ATmega88/168:
  - 1] The sum of all IOL, for ports C0 C5, should not exceed 100mA.
  - 2] The sum of all IOL, for ports C6, D0 D4, should not exceed 100mA.
  - 3] The sum of all IOL, for ports B0 B7, D5 D7, should not exceed 100mA.

If IOL exceeds the test condition, VOL may exceed the related specification. Pins are not guaranteed to sink current greater than the listed test condition.

- Although each I/O port can source more than the test conditions (20mA at V<sub>CC</sub> = 5V, 10mA at V<sub>CC</sub> = 3V) under steady state conditions (non-transient), the following must be observed: ATmega48:
  - The sum of all IOH, for ports C0 C5, should not exceed 70mA.
    The sum of all IOH, for ports C6, D0 D4, should not exceed 70mA.
    The sum of all IOH, for ports B0 B7, D5 D7, should not exceed 70mA.
    ATmega88/168:
    The sum of all IOH, for ports C0 C5, should not exceed 100mA.
    The sum of all IOH, for ports C0 C5, should not exceed 100mA.
    The sum of all IOH, for ports C6, D0 D4, should not exceed 100mA.
    The sum of all IOH, for ports C6, D0 D4, should not exceed 100mA.
    The sum of all IOH, for ports B0 B7, D5 D7, should not exceed 100mA.
    The sum of all IOH, for ports B0 B7, D5 D7, should not exceed 100mA.
    The sum of all IOH, for ports B0 B7, D5 D7, should not exceed 100mA.
    The sum of all IOH, for ports B0 B7, D5 D7, should not exceed 100mA.
- 5. All DC characteristics contained in this datasheet are based on actual ATmega88 microcontrollers characterization.
- 6. Values with Section 7.7.1 "Power Reduction Register PRR" on page 35 enabled (0xEF).



# 28. ATmega48/88/168 Typical Characteristics

## 28.1 Active Supply Current











## 28.1.4 Pin Thresholds and Hysteresis









Figure 28-17. Reset Input Threshold Voltage versus V<sub>cc</sub> (VIH, Reset Pin Read as '1')



Figure 28-18. Reset Input Threshold Voltage versus V<sub>CC</sub> (VIL, Reset Pin Read as '0')





Figure 28-21. Calibrated 8MHz RC Oscillator Frequency versus V<sub>CC</sub>



Figure 28-22. Calibrated 8MHz RC Oscillator Frequency versus OSCCAL Value (for ATmega48-15AZ and ATmega168-15AZ)



