Welcome to E-XFL.COM

What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	ST7
Core Size	8-Bit
Speed	16MHz
Connectivity	I²C, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	22
Program Memory Size	4KB (4K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	256 x 8
Voltage - Supply (Vcc/Vdd)	3.2V ~ 5.5V
Data Converters	A/D 6x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	28-SOIC (0.295", 7.50mm Width)
Supplier Device Package	28-SO
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/st72c254g1m6

Table of Contents

3 REGISTER & MEMORY MAP

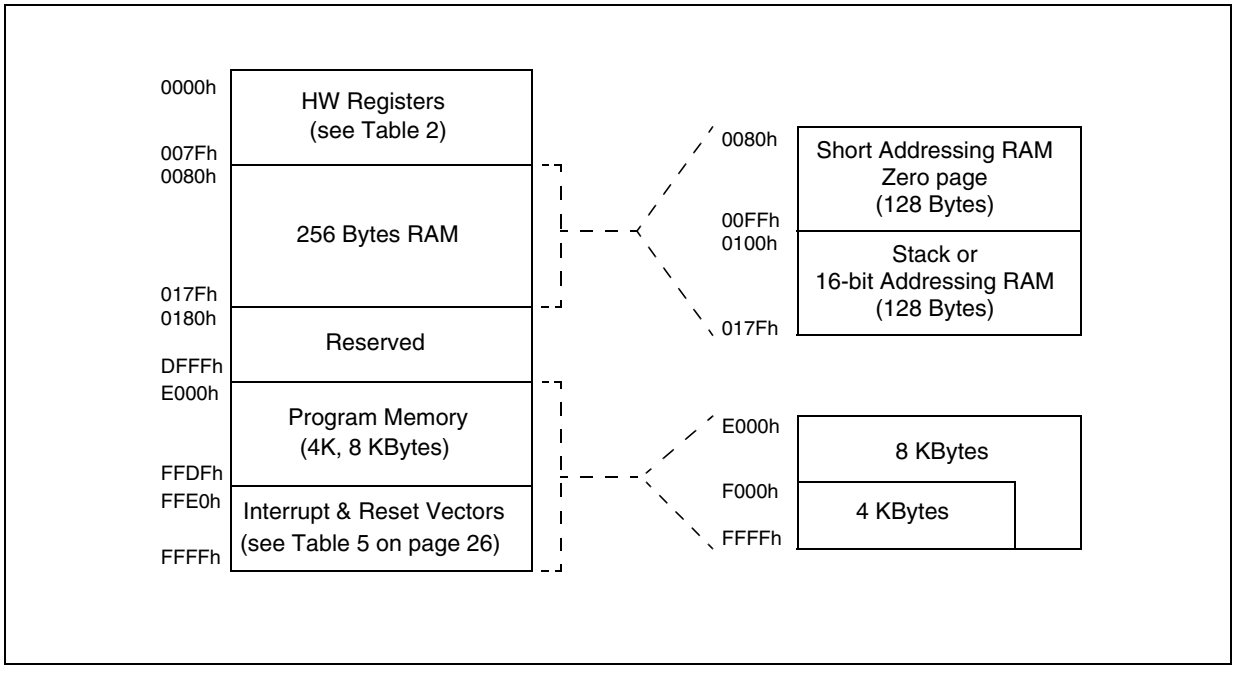
As shown in the Figure 4, the MCU is capable of addressing 64K bytes of memories and I/O registers.

The available memory locations consist of 128 bytes of register location, 256 bytes of RAM and up to 8Kbytes of user program memory. The RAM space includes up to 128 bytes for the stack from 0100h to 017Fh.

The highest address bytes contain the user reset and interrupt vectors.

IMPORTANT: Memory locations marked as "Reserved" must never be accessed. Accessing a reserved area can have unpredictable effects on the device.

Figure 4. Memory Map



4 FLASH PROGRAM MEMORY

4.1 INTRODUCTION

FLASH devices have a single voltage non-volatile FLASH memory that may be programmed in-situ (or plugged in a programming tool) on a byte-by-byte basis.

4.2 MAIN FEATURES

- Remote In-Situ Programming (ISP) mode
- Up to 16 bytes programmed in the same cycle
- MTP memory (Multiple Time Programmable)
- Read-out memory protection against piracy

4.3 STRUCTURAL ORGANISATION

The FLASH program memory is organised in a single 8-bit wide memory block which can be used for storing both code and data constants.

The FLASH program memory is mapped in the upper part of the ST7 addressing space and includes the reset and interrupt user vector area .

4.4 IN-SITU PROGRAMMING (ISP) MODE

The FLASH program memory can be programmed using Remote ISP mode. This ISP mode allows the contents of the ST7 program memory to be updated using a standard ST7 programming tools after the device is mounted on the application board. This feature can be implemented with a minimum number of added components and board area impact.

An example Remote ISP hardware interface to the standard ST7 programming tool is described below. For more details on ISP programming, refer to the ST7 Programming Specification.

Remote ISP Overview

The Remote ISP mode is initiated by a specific sequence on the dedicated ISPSEL pin.

The Remote ISP is performed in three steps:

– Selection of the RAM execution mode
– Download of Remote ISP code in RAM
– Execution of Remote ISP code in RAM to program the user program into the FLASH

Remote ISP hardware configuration

In Remote ISP mode, the ST7 has to be supplied with power (V_{DD} and V_{SS}) and a clock signal (oscillator and application crystal circuit for example).

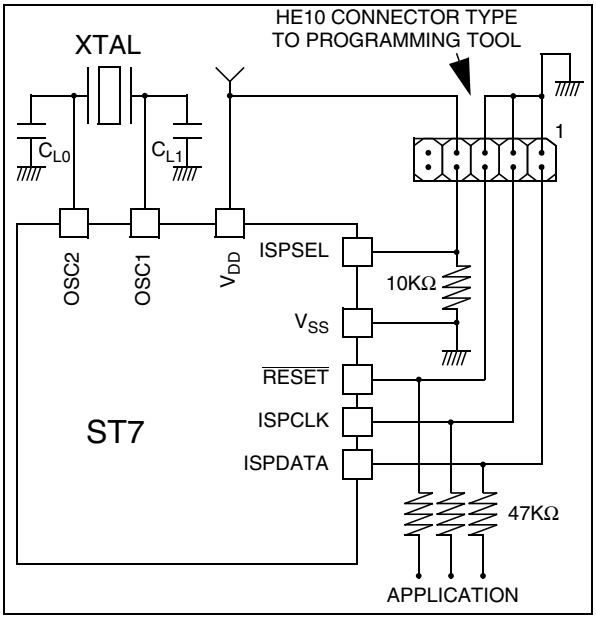
This mode needs five signals (plus the V_{DD} signal if necessary) to be connected to the programming tool. This signals are:

– $\overline{\text{RESET}}$: device reset
– V_{SS}: device ground power supply
– ISPCLK: ISP output serial clock pin
– ISPDATA: ISP input serial data pin
– ISPSEL: Remote ISP mode selection. This pin must be connected to V_{SS} on the application board through a pull-down resistor.

If any of these pins are used for other purposes on the application, a serial resistor has to be implemented to avoid a conflict if the other device forces the signal level.

Figure 5 shows a typical hardware interface to a standard ST7 programming tool. For more details on the pin locations, refer to the device pinout description.

Figure 5. Typical Remote ISP Interface



4.5 MEMORY READ-OUT PROTECTION

The read-out protection is enabled through an option bit.

For FLASH devices, when this option is selected, the program and data stored in the FLASH memory are protected against read-out piracy (including a re-write protection). When this protection option is removed the entire FLASH program memory is first automatically erased. However, the E^2PROM data memory (when available) can be protected only with ROM devices.

RESET SEQUENCE MANAGER (Cont'd)

6.2.2 Asynchronous External $\overline{\text{RESET}}$ pin

The $\overline{\text{RESET}}$ pin is both an input and an open-drain output with integrated R_{ON} weak pull-up resistor. This pull-up has no fixed value but varies in accordance with the input voltage. It can be pulled low by external circuitry to reset the device. See electrical characteristics section for more details.

A RESET signal originating from an external source must have a duration of at least $t_{h(RSTL)in}$ in order to be recognized. This detection is asynchronous and therefore the MCU can enter reset state even in HALT mode.

The $\overline{\text{RESET}}$ pin is an asynchronous signal which plays a major role in EMS performance. In a noisy environment, it is recommended to follow the guidelines mentioned in the electrical characteristics section.

Two RESET sequences can be associated with this RESET source: short or long external reset pulse (see Figure 12).

Starting from the external RESET pulse recognition, the device $\overline{\text{RESET}}$ pin acts as an output that is pulled low during at least $t_{w(RSTL)out}$.

6.2.3 Internal Low Voltage Detection RESET

Two different RESET sequences caused by the internal LVD circuitry can be distinguished:

- Power-On RESET
- Voltage Drop RESET

The device $\overline{\text{RESET}}$ pin acts as an output that is pulled low when $V_{DD}<V_{IT+}$ (rising edge) or $V_{DD}<V_{IT-}$ (falling edge) as shown in Figure 12.

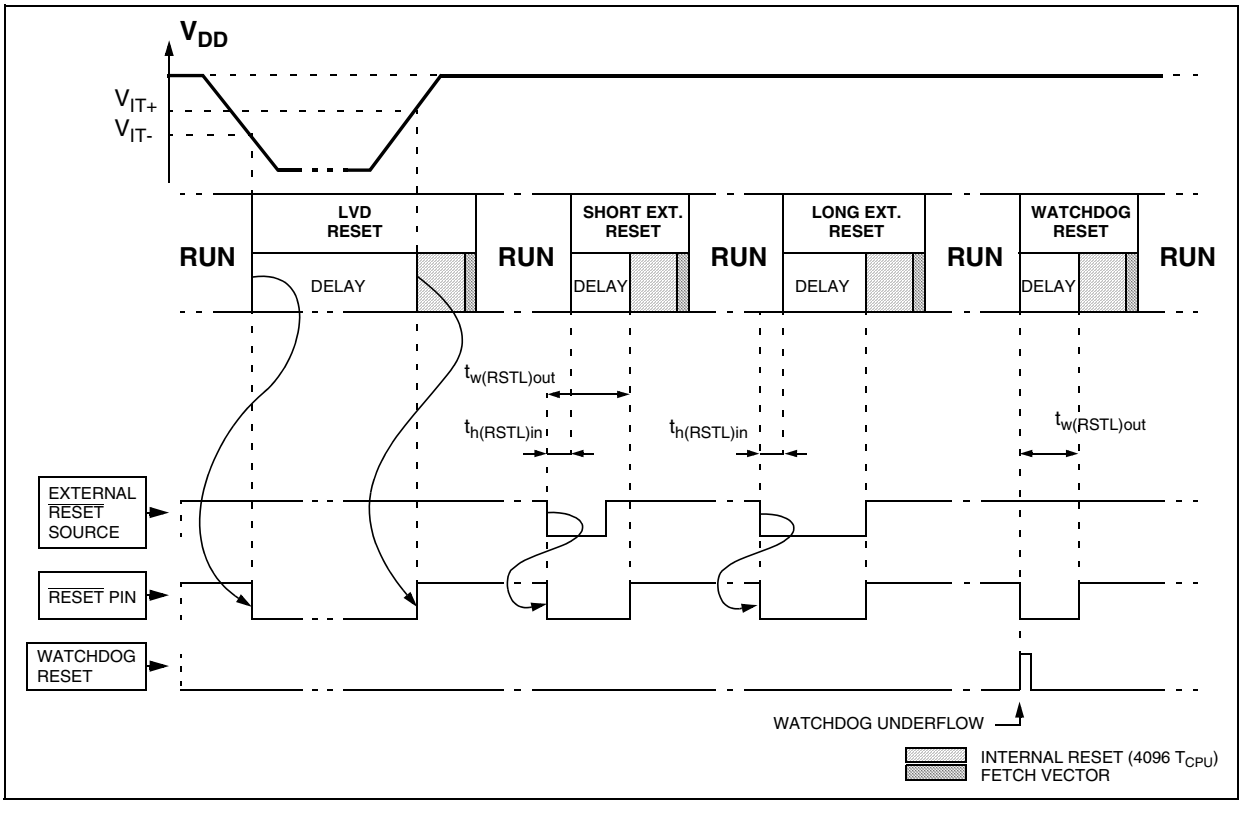
The LVD filters spikes on V_{DD} larger than $t_{g(VDD)}$ to avoid parasitic resets.

6.2.4 Internal Watchdog RESET

The RESET sequence generated by a internal Watchdog counter overflow is shown in Figure 12.

Starting from the Watchdog counter underflow, the device $\overline{\text{RESET}}$ pin acts as an output that is pulled low during at least $t_{w(RSTL)out}$.

Figure 12. RESET Sequences



MISCELLANEOUS REGISTERS (Cont'd)

10.3 MISCELLANEOUS REGISTER DESCRIPTION

MISCELLANEOUS REGISTER 1 (MISCR1)

Read/Write

Reset Value: 0000 0000 (00h)

7							0
IS11	IS10	MCO	IS01	IS00	CP1	CP0	SMS

Bit 7:6 = **IS1[1:0]** *ei1 sensitivity*
The interrupt sensitivity, defined using the IS1[1:0] bits, is applied to the ei1 external interrupts. These two bits can be written only when the I bit of the CC register is set to 1 (interrupt masked).

ei1: Port B (C optional)

External Interrupt Sensitivity	IS11	IS10
Falling edge & low level	0	0
Rising edge only	0	1
Falling edge only	1	0
Rising and falling edge	1	1

Bit 5 = **MCO** *Main clock out selection*
This bit enables the MCO alternate function on the PC2 I/O port. It is set and cleared by software.
0: MCO alternate function disabled (I/O pin free for general-purpose I/O)
1: MCO alternate function enabled (f_{CPU} on I/O port)

Bit 4:3 = **IS0[1:0]** *ei0 sensitivity*
The interrupt sensitivity, defined using the IS0[1:0] bits, is applied to the ei0 external interrupts. These two bits can be written only when the I bit of the CC register is set to 1 (interrupt masked).

ei0: Port A (C optional)

External Interrupt Sensitivity	IS01	IS00
Falling edge & low level	0	0
Rising edge only	0	1
Falling edge only	1	0
Rising and falling edge	1	1

Bit 2:1 = **CP[1:0]** *CPU clock prescaler*
These bits select the CPU clock prescaler which is applied in the different slow modes. Their action is conditioned by the setting of the SMS bit. These two bits are set and cleared by software

f_{CPU} in SLOW mode	CP1	CP0
f_{OSC} / 4	0	0
f_{OSC} / 8	1	0
f_{OSC} / 16	0	1
f_{OSC} / 32	1	1

Bit 0 = **SMS** *Slow mode select*
This bit is set and cleared by software.
0: Normal mode. $f_{CPU} = f_{OSC} / 2$
1: Slow mode. f_{CPU} is given by CP1, CP0
See low power consumption mode and MCC chapters for more details.

11 ON-CHIP PERIPHERALS

11.1 WATCHDOG TIMER (WDG)

11.1.1 Introduction

The Watchdog timer is used to detect the occurrence of a software fault, usually generated by external interference or by unforeseen logical conditions, which causes the application program to abandon its normal sequence. The Watchdog circuit generates an MCU reset on expiry of a programmed time period, unless the program refreshes the counter's contents before the T6 bit becomes cleared.

11.1.2 Main Features

- Programmable timer (64 increments of 12288 CPU cycles)
- Programmable reset
- Reset (if watchdog activated) when the T6 bit reaches zero
- Optional reset on HALT instruction (configurable by option byte)
- Hardware Watchdog selectable by option byte.

11.1.3 Functional Description

The counter value stored in the CR register (bits T6:T0), is decremented every 12,288 machine cycles, and the length of the timeout period can be programmed by the user in 64 increments.

If the watchdog is activated (the WDGA bit is set) and when the 7-bit timer (bits T6:T0) rolls over from 40h to 3Fh (T6 becomes cleared), it initiates a reset cycle pulling low the reset pin for typically 500ns.

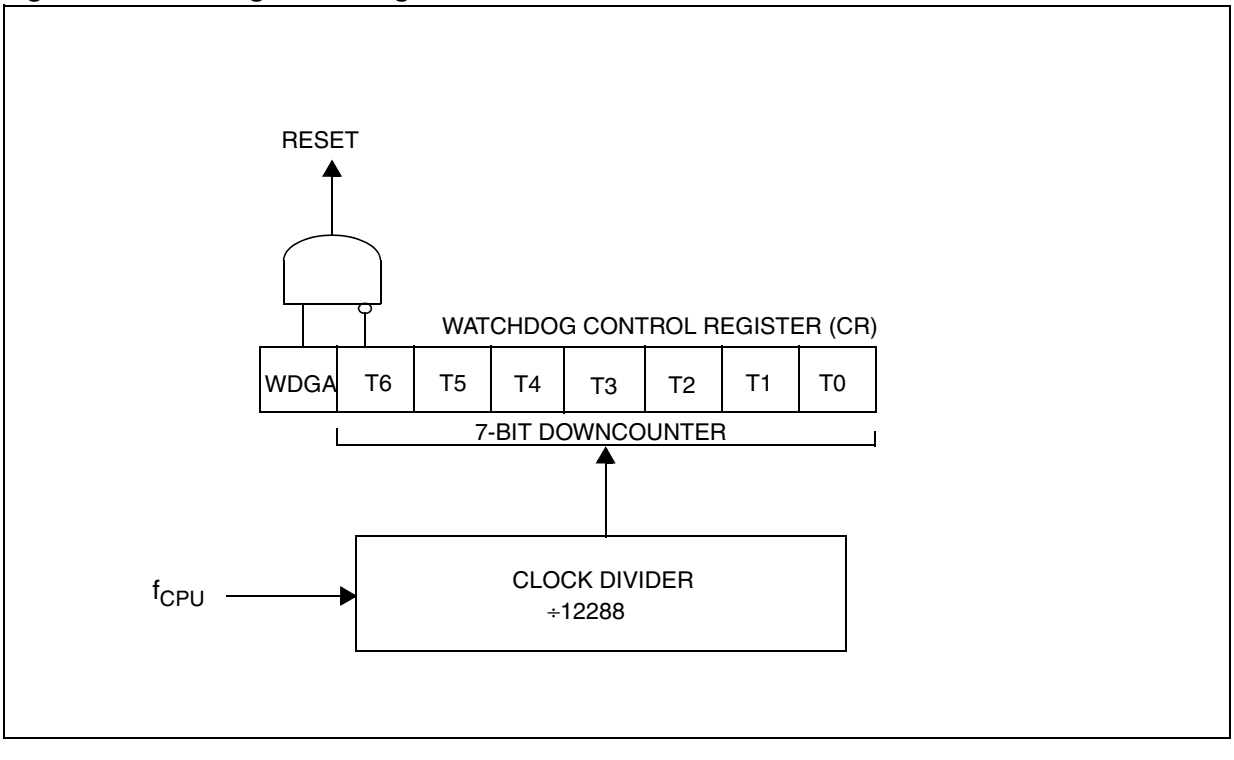
The application program must write in the CR register at regular intervals during normal operation to prevent an MCU reset. The value to be stored in the CR register must be between FFh and C0h (see Table 11 . Watchdog Timing (fCPU = 8 MHz)):

– The WDGA bit is set (watchdog enabled)

– The T6 bit is set to prevent generating an immediate reset

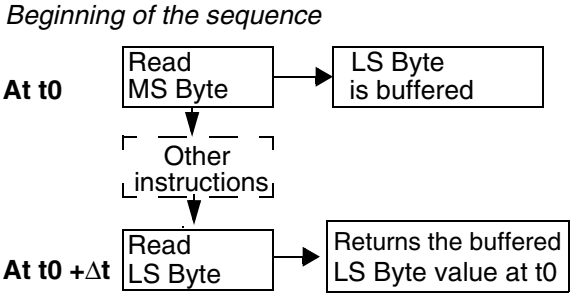
– The T5:T0 bits contain the number of increments which represents the time delay before the watchdog produces a reset.

Figure 25. Watchdog Block Diagram



16-BIT TIMER (Cont'd)

16-bit Read Sequence: (from either the Counter Register or the Alternate Counter Register).



The user must read the MS Byte first, then the LS Byte value is buffered automatically.

This buffered value remains unchanged until the 16-bit read sequence is completed, even if the user reads the MS Byte several times.

After a complete reading sequence, if only the CLR register or ACLR register are read, they return the LS Byte of the count value at the time of the read.

Whatever the timer mode used (input capture, output compare, One Pulse mode or PWM mode) an overflow occurs when the counter rolls over from FFFFh to 0000h then:

– The TOF bit of the SR register is set.

– A timer interrupt is generated if:

  – TOIE bit of the CR1 register is set and

  – I bit of the CC register is cleared.

If one of these conditions is false, the interrupt remains pending to be issued as soon as they are both true.

Clearing the overflow interrupt request is done in two steps:

1. Reading the SR register while the TOF bit is set.
2. An access (read or write) to the CLR register.

Note: The TOF bit is not cleared by accessing the ACLR register. The advantage of accessing the ACLR register rather than the CLR register is that it allows simultaneous use of the overflow function and reading the free running counter at random times (for example, to measure elapsed time) without the risk of clearing the TOF bit erroneously.

The timer is not affected by WAIT mode.

In HALT mode, the counter stops counting until the mode is exited. Counting then resumes from the previous count (MCU awakened by an interrupt) or from the reset count (MCU awakened by a Reset).

11.2.3.2 External Clock

The external clock (where available) is selected if CC0 = 1 and CC1 = 1 in the CR2 register.

The status of the EXEDG bit in the CR2 register determines the type of level transition on the external clock pin EXTCLK that will trigger the free running counter.

The counter is synchronized with the falling edge of the internal CPU clock.

A minimum of four falling edges of the CPU clock must occur between two consecutive active edges of the external clock; thus the external clock frequency must be less than a quarter of the CPU clock frequency.

16-BIT TIMER (Cont'd)

11.2.3.3 Input Capture

In this section, the index, *i*, may be 1 or 2 because there are 2 input capture functions in the 16-bit timer.

The two input capture 16-bit registers (IC1R and IC2R) are used to latch the value of the free running counter after a transition is detected by the ICAP*i* pin (see Figure 5).

	MS Byte	LS Byte
ICiR	IC*i*HR	IC*i*LR

The IC*i*R register is a read-only register.

The active transition is software programmable through the IEDG*i* bit of Control Registers (CR*i*).

Timing resolution is one count of the free running counter: (f_{CPU}/CC[1:0]).

Procedure:

To use the input capture function, select the following in the CR2 register:

– Select the timer clock (CC[1:0]) (see Table 1).

– Select the edge of the active transition on the ICAP2 pin with the IEDG2 bit (the ICAP2 pin must be configured as a floating input or input with pull-up without interrupt if this configuration is available).

And select the following in the CR1 register:

– Set the ICIE bit to generate an interrupt after an input capture coming from either the ICAP1 pin or the ICAP2 pin

– Select the edge of the active transition on the ICAP1 pin with the IEDG1 bit (the ICAP1 pin must be configured as a floating input or input with pull-up without interrupt if this configuration is available).

When an input capture occurs:

– The ICF*i* bit is set.

– The IC*i*R register contains the value of the free running counter on the active transition on the ICAP*i* pin (see Figure 6).

– A timer interrupt is generated if the ICIE bit is set and the I bit is cleared in the CC register. Otherwise, the interrupt remains pending until both conditions become true.

Clearing the Input Capture interrupt request (i.e. clearing the ICF*i* bit) is done in two steps:

1. Reading the SR register while the ICF*i* bit is set.
2. An access (read or write) to the IC*i*LR register.

Notes:

1. After reading the IC*i*HR register, the transfer of input capture data is inhibited and ICF*i* will never be set until the IC*i*LR register is also read.
2. The IC*i*R register contains the free running counter value which corresponds to the most recent input capture.
3. The two input capture functions can be used together even if the timer also uses the two output compare functions.
4. In One Pulse mode and PWM mode only the input capture 2 function can be used.
5. The alternate inputs (ICAP1 and ICAP2) are always directly connected to the timer. So any transitions on these pins activate the input capture function.
   Moreover if one of the ICAP*i* pin is configured as an input and the second one as an output, an interrupt can be generated if the user toggles the output pin and if the ICIE bit is set.
   This can be avoided if the input capture function *i* is disabled by reading the IC*i*HR (see note 1).
6. The TOF bit can be used with an interrupt in order to measure events that exceed the timer range (FFFFh).

16-BIT TIMER (Cont'd)

Notes:

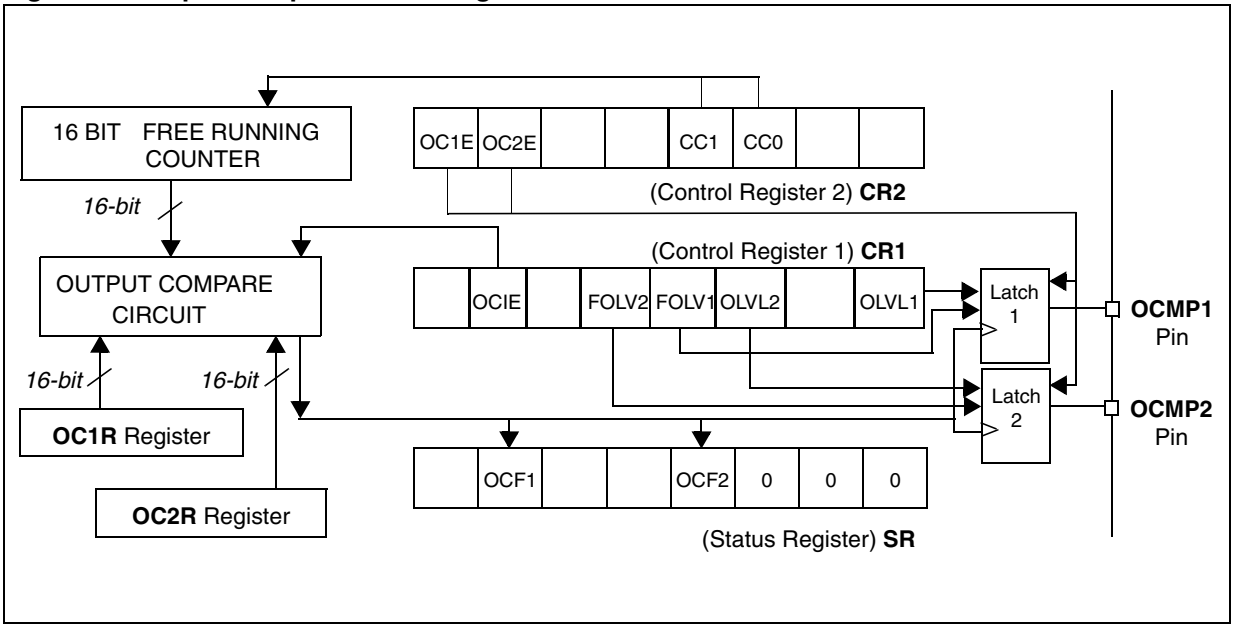
1. After a processor write cycle to the OC*i*HR register, the output compare function is inhibited until the OC*i*LR register is also written.
2. If the OC*i*E bit is not set, the OCMP*i* pin is a general I/O port and the OLVL*i* bit will not appear when a match is found but an interrupt could be generated if the OCIE bit is set.
3. When the timer clock is $f_{CPU}/2$, OCF*i* and OCMP*i* are set while the counter value equals the OC*i*R register value (see Figure 8). This behavior is the same in OPM or PWM mode. When the timer clock is $f_{CPU}/4$, $f_{CPU}/8$ or in external clock mode, OCF*i* and OCMP*i* are set while the counter value equals the OC*i*R register value plus 1 (see Figure 9).
4. The output compare functions can be used both for generating external events on the OCMP*i* pins even if the input capture mode is also used.
5. The value in the 16-bit OC*i*R register and the OLV*i* bit should be changed after each successful comparison in order to control an output waveform or establish a new elapsed timeout.

Forced Compare Output capability

When the FOLV*i* bit is set by software, the OLVL*i* bit is copied to the OCMP*i* pin. The OLV*i* bit has to be toggled in order to toggle the OCMP*i* pin when it is enabled (OC*i*E bit = 1). The OCF*i* bit is then not set by hardware, and thus no interrupt request is generated.

FOLVL*i* bits have no effect in either One-Pulse mode or PWM mode.

Figure 32. Output Compare Block Diagram



16-BIT TIMER (Cont'd)

11.2.3.6 Pulse Width Modulation Mode

Pulse Width Modulation (PWM) mode enables the generation of a signal with a frequency and pulse length determined by the value of the OC1R and OC2R registers.

The Pulse Width Modulation mode uses the complete Output Compare 1 function plus the OC2R register, and so these functions cannot be used when the PWM mode is activated.

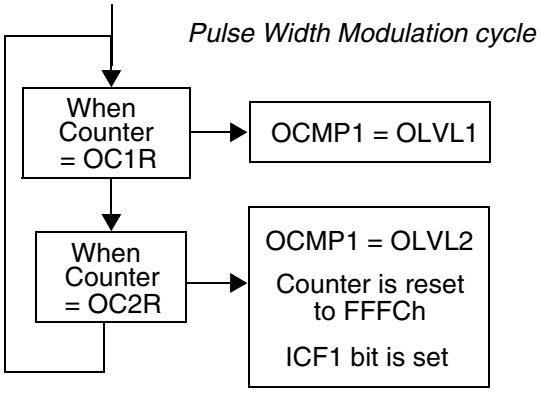
Procedure

To use Pulse Width Modulation mode:

1. Load the OC2R register with the value corresponding to the period of the signal using the formula in the opposite column.
2. Load the OC1R register with the value corresponding to the period of the pulse if OLVL1 = 0 and OLVL2 = 1, using the formula in the opposite column.
3. Select the following in the CR1 register:
   - Using the OLVL1 bit, select the level to be applied to the OCMP1 pin after a successful comparison with OC1R register.
   - Using the OLVL2 bit, select the level to be applied to the OCMP1 pin after a successful comparison with OC2R register.
4. Select the following in the CR2 register:
   - Set OC1E bit: the OCMP1 pin is then dedicated to the output compare 1 function.
   - Set the PWM bit.
   - Select the timer clock (CC[1:0]) (see Table 1).

If OLVL1 = 1 and OLVL2 = 0, the length of the positive pulse is the difference between the OC2R and OC1R registers.

If OLVL1 = OLVL2 a continuous signal will be seen on the OCMP1 pin.



The OC*i*R register value required for a specific timing application can be calculated using the following formula:

$$\text{OC}i\text{R Value} = \frac{t * f_{CPU}}{\text{PRESC}} - 5$$

Where:

t = Signal or pulse period (in seconds)

f_{CPU} = CPU clock frequency (in hertz)

PRESC = Timer prescaler factor (2, 4 or 8 depending on CC[1:0] bits, see Table 1)

If the timer clock is an external clock the formula is:

$$\text{OC}i\text{R} = t * f_{EXT} - 5$$

Where:

t = Signal or pulse period (in seconds)

f_{EXT} = External timer clock frequency (in hertz)

The Output Compare 2 event causes the counter to be initialized to FFFCh (See Figure 11)

Notes:

1. After a write instruction to the OC*i*HR register, the output compare function is inhibited until the OC*i*LR register is also written.
2. The OCF1 and OCF2 bits cannot be set by hardware in PWM mode, therefore the Output Compare interrupt is inhibited.
3. The ICF1 bit is set by hardware when the counter reaches the OC2R value and can produce a timer interrupt if the ICIE bit is set and the I bit is cleared.
4. In PWM mode the ICAP1 pin can not be used to perform input capture because it is disconnected from the timer. The ICAP2 pin can be used to perform input capture (ICF2 can be set and IC2R can be loaded) but the user must take care that the counter is reset after each period and ICF1 can also generate an interrupt if ICIE is set.
5. When the Pulse Width Modulation (PWM) and One Pulse mode (OPM) bits are both set, the PWM mode is the only active one.

SERIAL PERIPHERAL INTERFACE (Cont'd)

11.3.4 Functional Description

Figure 1 shows the serial peripheral interface (SPI) block diagram.

This interface contains three dedicated registers:

- A Control Register (CR)
- A Status Register (SR)
- A Data Register (DR)

Refer to the CR, SR and DR registers in Section 0.1.7 for the bit definitions.

11.3.4.1 Master Configuration

In a master configuration, the serial clock is generated on the SCK pin.

Procedure

- Select the SPR0 & SPR1 bits to define the serial clock baud rate (see CR register).
- Select the CPOL and CPHA bits to define one of the four relationships between the data transfer and the serial clock (see Figure 4).
- The $\overline{SS}$ pin must be connected to a high level signal during the complete byte transmit sequence.
- The MSTR and SPE bits must be set (they remain set only if the $\overline{SS}$ pin is connected to a high level signal).

In this configuration the MOSI pin is a data output and to the MISO pin is a data input.

Transmit sequence

The transmit sequence begins when a byte is written the DR register.

The data byte is parallel loaded into the 8-bit shift register (from the internal bus) during a write cycle and then shifted out serially to the MOSI pin most significant bit first.

When data transfer is complete:

- The SPIF bit is set by hardware
- An interrupt is generated if the SPIE bit is set and the I bit in the CCR register is cleared.

During the last clock cycle the SPIF bit is set, a copy of the data byte received in the shift register is moved to a buffer. When the DR register is read, the SPI peripheral returns this buffered value.

Clearing the SPIF bit is performed by the following software sequence:

1. An access to the SR register while the SPIF bit is set
2. A read to the DR register.

Note: While the SPIF bit is set, all writes to the DR register are inhibited until the SR register is read.

SERIAL PERIPHERAL INTERFACE (Cont'd)

STATUS REGISTER (SR)

Read Only

Reset Value: 0000 0000 (00h)

7							0
SPIF	WCOL	-	MODF	-	-	-	-

Bit 7 = **SPIF** *Serial Peripheral data transfer flag.*
This bit is set by hardware when a transfer has been completed. An interrupt is generated if SPIE=1 in the CR register. It is cleared by a software sequence (an access to the SR register followed by a read or write to the DR register).
0: Data transfer is in progress or has been approved by a clearing sequence.
1: Data transfer between the device and an external device has been completed.

Note: While the SPIF bit is set, all writes to the DR register are inhibited.

Bit 6 = **WCOL** *Write Collision status.*

This bit is set by hardware when a write to the DR register is done during a transmit sequence. It is cleared by a software sequence (see Figure 5).
0: No write collision occurred
1: A write collision has been detected

Bit 5 = Unused.

Bit 4 = **MODF** *Mode Fault flag.*
This bit is set by hardware when the $\overline{SS}$ pin is pulled low in master mode (see Section 0.1.4.5 Master Mode Fault). An SPI interrupt can be generated if SPIE=1 in the CR register. This bit is cleared by a software sequence (An access to the SR register while MODF=1 followed by a write to the CR register).
0: No master mode fault detected
1: A fault in master mode has been detected

Bits 3-0 = Unused.

DATA I/O REGISTER (DR)

Read/Write

Reset Value: Undefined

7							0
D7	D6	D5	D4	D3	D2	D1	D0

The DR register is used to transmit and receive data on the serial bus. In the master device only a write to this register will initiate transmission/reception of another byte.

Notes: During the last clock cycle the SPIF bit is set, a copy of the received data byte in the shift register is moved to a buffer. When the user reads the serial peripheral data I/O register, the buffer is actually being read.

Warning:

A write to the DR register places data directly into the shift register for transmission.

A read to the DR register returns the value located in the buffer and not the contents of the shift register (See Figure 2).

to correctly handle a second interrupt during the 9th pulse of a transmitted byte.

Note: In both cases, SCL line is not held low; however, the SDA line can remain low if the last bits transmitted are all 0. It is then necessary to release both lines by software. The SCL line is not held low while AF=1 but by other flags (SB or BTF) that are set at the same time.

How to release the SDA / SCL lines

Set and subsequently clear the STOP bit while BTF is set. The SDA/SCL lines are released after the transfer of the current byte.

SMBus Compatibility

ST7 I^2C is compatible with SMBus V1.1 protocol. It supports all SMBus adressing modes, SMBus bus protocols and CRC-8 packet error checking. Refer to AN1713: SMBus Slave Driver For ST7 I^2C Peripheral.

11.4.4.2 Master Mode

To switch from default Slave mode to Master mode a Start condition generation is needed.

Start condition

Setting the START bit while the BUSY bit is cleared causes the interface to switch to Master mode (M/SL bit set) and generates a Start condition.

Once the Start condition is sent:

– The EVF and SB bits are set by hardware with an interrupt if the ITE bit is set.

Then the master waits for a read of the SR1 register followed by a write in the DR register with the Slave address, **holding the SCL line low** (see Figure 3 Transfer sequencing EV5).

Slave address transmission

Then the slave address is sent to the SDA line via the internal shift register.

In 7-bit addressing mode, one address byte is sent.

In 10-bit addressing mode, sending the first byte including the header sequence causes the following event:

– The EVF bit is set by hardware with interrupt generation if the ITE bit is set.

Then the master waits for a read of the SR1 register followed by a write in the DR register, **holding the SCL line low** (see Figure 3 Transfer sequencing EV9).
Then the second address byte is sent by the interface.

After completion of this transfer (and acknowledge from the slave if the ACK bit is set):

– The EVF bit is set by hardware with interrupt generation if the ITE bit is set.

Then the master waits for a read of the SR1 register followed by a write in the CR register (for example set PE bit), **holding the SCL line low** (see Figure 3 Transfer sequencing EV6).

Next the master must enter Receiver or Transmitter mode.

Note: In 10-bit addressing mode, to switch the master to Receiver mode, software must generate a repeated Start condition and resend the header sequence with the least significant bit set (11110xx1).

Master Receiver

Following the address transmission and after SR1 and CR registers have been accessed, the master receives bytes from the SDA line into the DR register via the internal shift register. After each byte the interface generates in sequence:

– Acknowledge pulse if the ACK bit is set

– EVF and BTF bits are set by hardware with an interrupt if the ITE bit is set.

Then the interface waits for a read of the SR1 register followed by a read of the DR register, **holding the SCL line low** (see Figure 3 Transfer sequencing EV7).

To close the communication: before reading the last byte from the DR register, set the STOP bit to generate the Stop condition. The interface goes automatically back to slave mode (M/SL bit cleared).

Note: In order to generate the non-acknowledge pulse after the last received data byte, the ACK bit must be cleared just before reading the second last data byte.

FUNCTIONAL OPERATING CONDITIONS (Cont'd)

Figure 56. High LVD Threshold Versus V_{DD} and f_{OSC} for ROM devices 2)

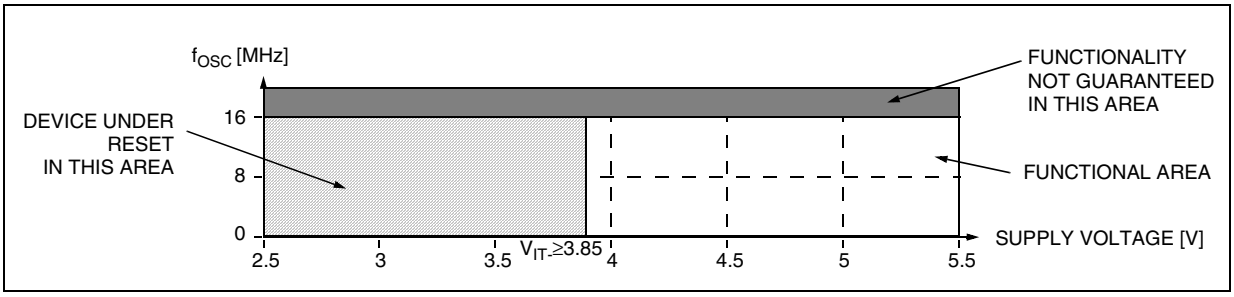


Figure 57. Medium LVD Threshold Versus V_{DD} and f_{OSC} for ROM devices 2)

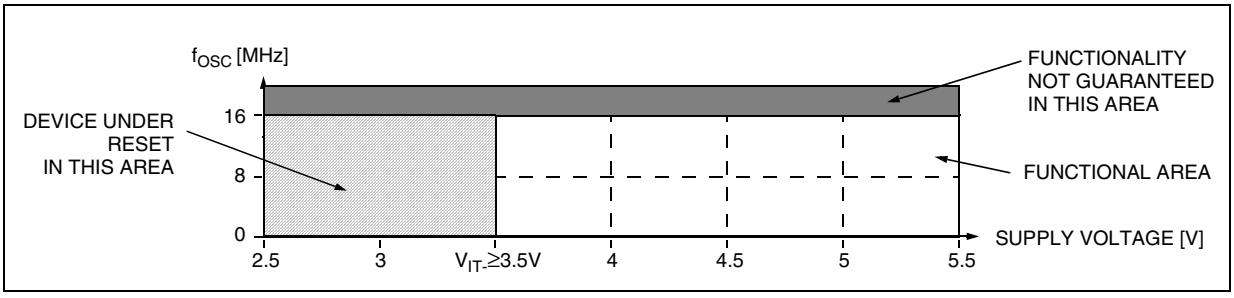
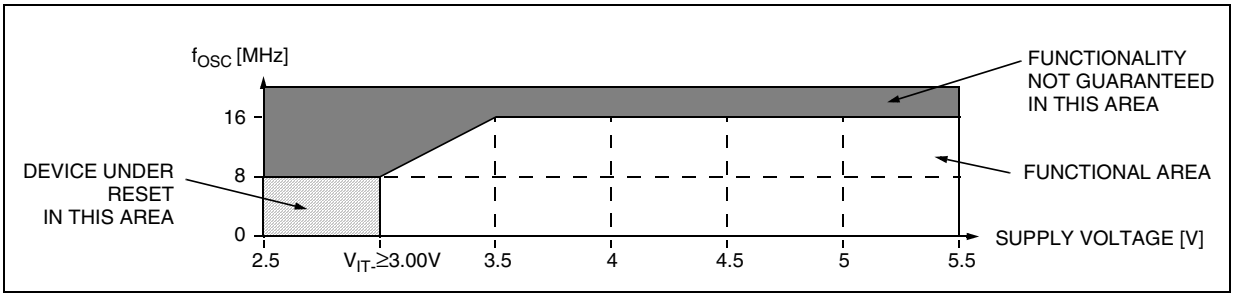


Figure 58. Low LVD Threshold Versus V_{DD} and f_{OSC} for ROM devices 2)3)



Notes:

1. LVD typical data are based on T_A=25°C. They are given only as design guidelines and are not tested.
2. The minimum V_{DD} rise time rate is needed to insure a correct device power-on and LVD reset. Not tested in production.
3. If the low LVD threshold is selected, when V_{DD} falls below 3.2V, the device is guaranteed to be either functioning or under reset.

SUPPLY CURRENT CHARACTERISTICS (Cont'd)

13.4.3 HALT Mode

Symbol	Parameter	Conditions		Typ [1]	Max	Unit
I_{DD}	Supply current in HALT mode [2]	V_{DD}=5.5V	-40°C≤T_A≤+85°C	-	10	µA
			-40°C≤T_A≤+125°C		150	
		V_{DD}=3.6V	-40°C≤T_A≤+85°C		6	
			-40°C≤T_A≤+125°C		100	

13.4.4 Supply and Clock Managers

The previous current consumption specified for the ST7 functional operating modes over temperature range does not take into account the clock source current consumption. To get the total device consumption, the two current values must be added (except for HALT mode).

Symbol	Parameter	Conditions	Typ [1]	Max [3]	Unit
$I_{DD(CK)}$	Supply current of internal RC oscillator		500	750	µA
	Supply current of external RC oscillator [4]		525	750	
	Supply current of resonator oscillator [4] & [5]	LP: Low power oscillator	200	400	
		MP: Medium power oscillator	300	550	
		MS: Medium speed oscillator	450	750	
		HS: High speed oscillator	700	1000	
	Clock security system supply current		150	350	
$I_{DD(LVD)}$	LVD supply current	HALT mode	100	150	

13.4.5 On-Chip Peripherals

Symbol	Parameter	Conditions		Typ	Unit
$I_{DD(TIM)}$	16-bit Timer supply current [6]	f_{CPU}=8MHz	V_{DD}=3.4V	50	µA
			V_{DD}=5.0V	150	
$I_{DD(SPI)}$	SPI supply current [7]	f_{CPU}=8MHz	V_{DD}=3.4V	250	
			V_{DD}=5.0V	350	
$I_{DD(I2C)}$	I^2C supply current [8]	f_{CPU}=8MHz	V_{DD}=3.4V	250	
			V_{DD}=5.0V	350	
$I_{DD(ADC)}$	ADC supply current when converting [9]	f_{ADC}=4MHz	V_{DD}=3.4V	800	
			V_{DD}=5.0V	1100	

Notes:

1. Typical data are based on T_A=25°C.
2. All I/O pins in input mode with a static value at V_{DD} or V_{SS} (no load), CSS and LVD disabled. Data based on characterization results, tested in production at V_{DD} max. and f_{CPU} max.
3. Data based on characterization results, not tested in production.
4. Data based on characterization results done with the external components specified in Section 13.5.3 and Section 13.5.4, not tested in production.
5. As the oscillator is based on a current source, the consumption does not depend on the voltage.
6. Data based on a differential I_{DD} measurement between reset configuration (timer counter running at f_{CPU}/4) and timer counter stopped (selecting external clock capability). Data valid for one timer.
7. Data based on a differential I_{DD} measurement between reset configuration and a permanent SPI master communication (data sent equal to 55h).
8. Data based on a differential I_{DD} measurement between reset configuration and I2C peripheral enabled (PE bit set).
9. Data based on a differential I_{DD} measurement between reset configuration and continuous A/D conversions.

13.6 MEMORY CHARACTERISTICS

Subject to general operating conditions for V_{DD}, f_{OSC}, and T_A unless otherwise specified.

13.6.1 RAM and Hardware Registers

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{RM}	Data retention mode [1]	HALT mode (or RESET)	1.6			V

13.6.2 FLASH Program Memory

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$T_{A(prog)}$	Programming temperature range [2]		0	25	70	°C
t_{prog}	Programming time for 1~16 bytes [3]	T_A=+25°C		8	25	ms
	Programming time for 4 or 8kBytes	T_A=+25°C		2.1	6.4	sec
t_{ret}	Data retention [5]	T_A=+55°C [4]	20			years
N_{RW}	Write erase cycles [5]	T_A=+25°C	100			cycles

Notes:

1. Minimum V_{DD} supply voltage without losing data stored in RAM (in HALT mode or under RESET) or in hardware registers (only in HALT mode). Guaranteed by construction, not tested in production.
2. Data based on characterization results, tested in production at T_A=25°C.
3. Up to 16 bytes can be programmed at a time for a 4kBytes FLASH block (then up to 32 bytes at a time for an 8k device)
4. The data retention time increases when the T_A decreases.
5. Data based on reliability test results and monitored in production.

I/O PORT PIN CHARACTERISTICS (Cont'd)

Figure 83. Typical V_{OL} vs. V_{DD} (standard I/Os)

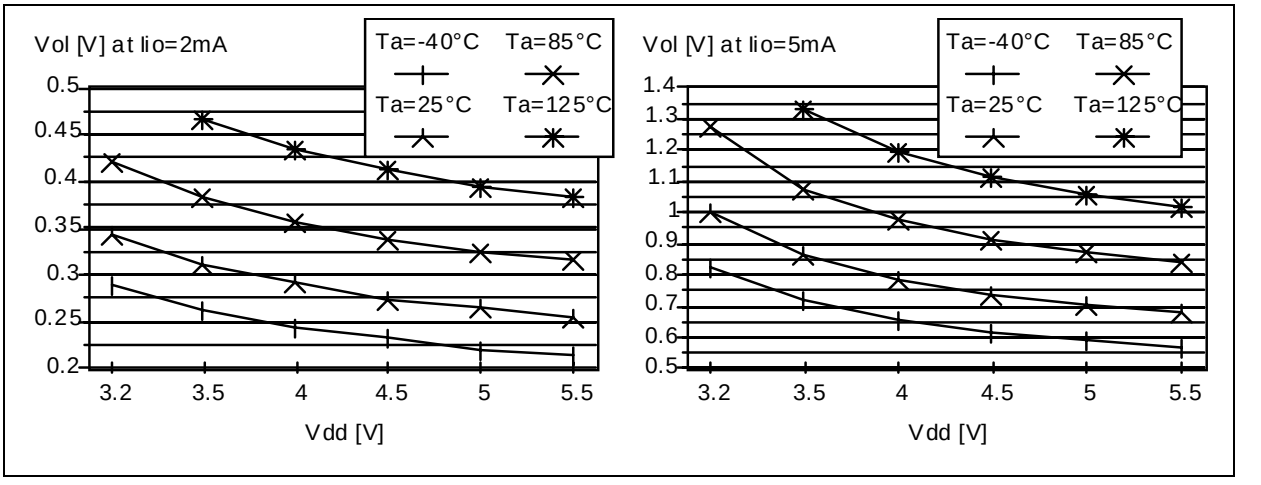


Figure 84. Typical V_{OL} vs. V_{DD} (high-sink I/Os)

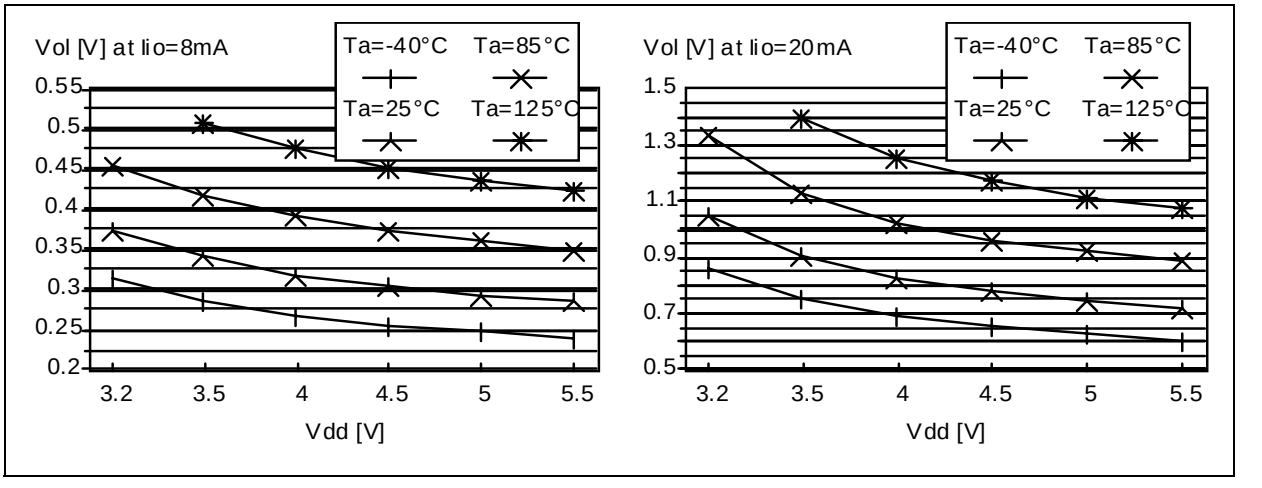
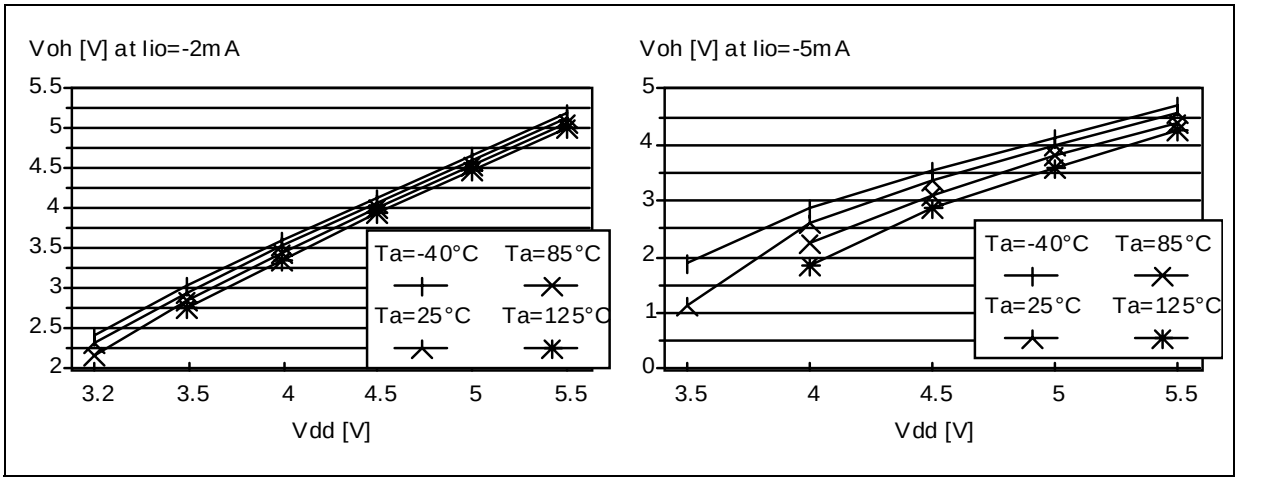


Figure 85. Typical V_{OH} vs. V_{DD}



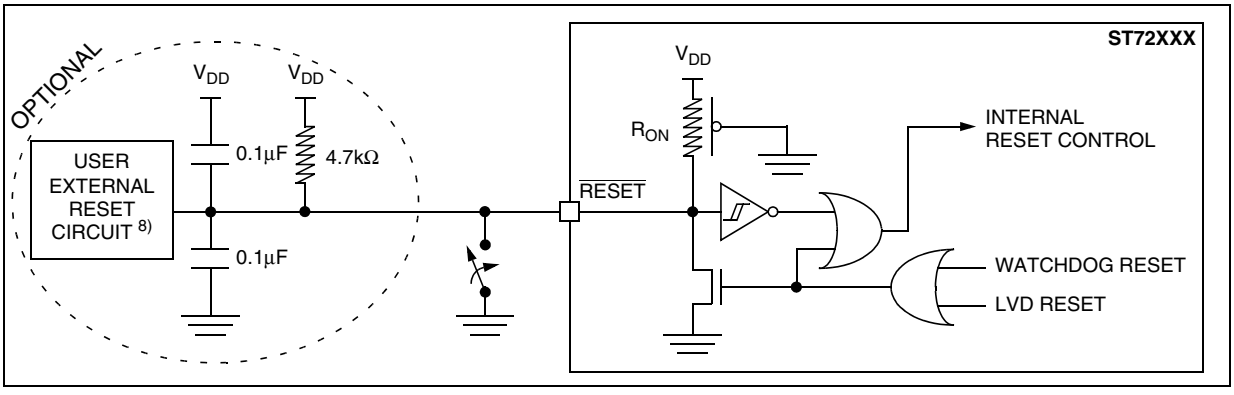
13.9 CONTROL PIN CHARACTERISTICS

13.9.1 Asynchronous $\overline{\text{RESET}}$ Pin

Subject to general operating conditions for V_{DD}, f_{OSC}, and T_A unless otherwise specified.

Symbol	Parameter	Conditions		Min	Typ [1]	Max	Unit
V_{IL}	Input low level voltage [2]					$0.3xV_{DD}$	V
V_{IH}	Input high level voltage [2]			$0.7xV_{DD}$			
V_{hys}	Schmitt trigger voltage hysteresis [3]				400		mV
V_{OL}	Output low level voltage [4] (see Figure 88, Figure 89)	V_{DD}=5V	I_{IO}=+5mA		0.68	0.95	V
			I_{IO}=+2mA		0.28	0.45	
R_{ON}	Weak pull-up equivalent resistor [5]	V_{IN}=V_{SS}	V_{DD}=5V	20	40	60	kΩ
			V_{DD}=3.4V	80	100	120	
$t_{w(RSTL)out}$	Generated reset pulse duration	External pin or internal reset sources			6 30		$1/f_{SFOSC}$ µs
$t_{h(RSTL)in}$	External reset pulse hold time [6]			20			µs
$t_{g(RSTL)in}$	Filtered glitch duration [7]					100	ns

Figure 86. Typical Application with $\overline{\text{RESET}}$ pin [8]



Notes:

1. Unless otherwise specified, typical data are based on T_A=25°C and V_{DD}=5V.
2. Data based on characterization results, not tested in production.
3. Hysteresis voltage between Schmitt trigger switching levels. Based on characterization results, not tested.
4. The I_{IO} current sunk must always respect the absolute maximum rating specified in Section 13.2.2 and the sum of I_{IO} (I/O ports and control pins) must not exceed I_{VSS}.
5. The R_{ON} pull-up equivalent resistor is based on a resistive transistor (corresponding I_{ON} current characteristics described in Figure 87). This data is based on characterization results, not tested in production.
6. To guarantee the reset of the device, a minimum pulse has to be applied to $\overline{\text{RESET}}$ pin. All short pulses applied on $\overline{\text{RESET}}$ pin with a duration below $t_{h(RSTL)in}$ can be ignored.
7. The reset network (the resistor and two capacitors) protects the device against parasitic resets, especially in a noisy environments.
8. The output of the external reset circuit must have an open-drain output to drive the ST7 reset pad. Otherwise the device can be damaged when the ST7 generates an internal reset (LVD or watchdog).

TRANSFER OF CUSTOMER CODE (Cont'd)

MICROCONTROLLER OPTION LIST

Customer . . . . . .
Address . . . . . .
. . . . . .
Contact . . . . . .
Phone No . . . . . .
Reference . . . . . .

Device: [ ] ST72104G1 (4KB) [ ] ST72215G2 (4KB) [ ] ST72254G1 (4KB)
[ ] ST72104G2 (8KB) [ ] ST72216G1 (8KB) [ ] ST72254G2 (8KB)

Package: [ ] SO28 [ ] Tape & Reel [ ] Tube
[ ] SDIP32

Marking: [ ] Standard Marking
[ ] Special Marking SO28 (max. 13 Chars.): _ _ _ _ _ _ _ _ _ _ _ _ _
SDIP32 (max. 15 Chars.): _ _ _ _ _ _ _ _ _ _ _ _ _ _ _ _ _
Authorized characters are letters, digits, '.', '-', '/' and spaces only. Please consult your local STMicroelectronics sales office for other marking details if required.

External Interrupt: [ ] IT0 interrupt vector Port A, IT1 interrupt vector Port B & C
[ ] IT0 interrupt vector Port A & C, IT1 interrupt vector Port B

Temperature Range: [ ] 0°C to + 70°C [ ] - 10°C to + 85°C
[ ] - 40°C to + 85°C [ ] - 40°C to + 105°C [ ] - 40°C to + 125°C

Clock Source Selection: [ ] Resonator: [ ] LP: Low power resonator (1 to 2 MHz)
[ ] MP: Medium power resonator (2 to 4 MHz)
[ ] MS: Medium speed resonator (4 to 8 MHz)
[ ] HS: High speed resonator (8 to 16 MHz)
[ ] RC Network: [ ] Internal
[ ] External
[ ] External Clock

Clock Security System: [ ] Disabled [ ] Enabled

Watchdog Selection: [ ] Software Activation [ ] Hardware Activation
Halt when Watchdog on: [ ] Reset [ ] No reset

Readout Protection: [ ] Disabled [ ] Enabled

LVD Reset [ ] Disabled [ ] Enabled: [ ] Highest threshold
[ ] Medium threshold
[ ] Lowest threshold

Comments : . . . . . .

Supply Operating Range in the application: . . . . . .

Notes: . . . . . .

Signature: . . . . . .

15.4 ST7 APPLICATION NOTES

IDENTIFICATION	DESCRIPTION
EXAMPLE DRIVERS	
AN 969	SCI COMMUNICATION BETWEEN ST7 AND PC
AN 970	SPI COMMUNICATION BETWEEN ST7 AND EEPROM
AN 971	I²C COMMUNICATING BETWEEN ST7 AND M24CXX EEPROM
AN 972	ST7 SOFTWARE SPI MASTER COMMUNICATION
AN 973	SCI SOFTWARE COMMUNICATION WITH A PC USING ST72251 16-BIT TIMER
AN 974	REAL TIME CLOCK WITH ST7 TIMER OUTPUT COMPARE
AN 976	DRIVING A BUZZER THROUGH ST7 TIMER PWM FUNCTION
AN 979	DRIVING AN ANALOG KEYBOARD WITH THE ST7 ADC
AN 980	ST7 KEYPAD DECODING TECHNIQUES, IMPLEMENTING WAKE-UP ON KEYSTROKE
AN1017	USING THE ST7 UNIVERSAL SERIAL BUS MICROCONTROLLER
AN1041	USING ST7 PWM SIGNAL TO GENERATE ANALOG OUTPUT (SINUSOID)
AN1042	ST7 ROUTINE FOR I²C SLAVE MODE MANAGEMENT
AN1044	MULTIPLE INTERRUPT SOURCES MANAGEMENT FOR ST7 MCUS
AN1045	ST7 S/W IMPLEMENTATION OF I²C BUS MASTER
AN1046	UART EMULATION SOFTWARE
AN1047	MANAGING RECEPTION ERRORS WITH THE ST7 SCI PERIPHERALS
AN1048	ST7 SOFTWARE LCD DRIVER
AN1078	PWM DUTY CYCLE SWITCH IMPLEMENTING TRUE 0% & 100% DUTY CYCLE
AN1082	DESCRIPTION OF THE ST72141 MOTOR CONTROL PERIPHERAL REGISTERS
AN1083	ST72141 BLDC MOTOR CONTROL SOFTWARE AND FLOWCHART EXAMPLE
AN1105	ST7 PCAN PERIPHERAL DRIVER
AN1129	PERMANENT MAGNET DC MOTOR DRIVE.
AN1130	AN INTRODUCTION TO SENSORLESS BRUSHLESS DC MOTOR DRIVE APPLICATIONS WITH THE ST72141
AN1148	USING THE ST7263 FOR DESIGNING A USB MOUSE
AN1149	HANDLING SUSPEND MODE ON A USB MOUSE
AN1180	USING THE ST7263 KIT TO IMPLEMENT A USB GAME PAD
AN1276	BLDC MOTOR START ROUTINE FOR THE ST72141 MICROCONTROLLER
AN1321	USING THE ST72141 MOTOR CONTROL MCU IN SENSOR MODE
AN1325	USING THE ST7 USB LOW-SPEED FIRMWARE V4.X
AN1445	USING THE ST7 SPI TO EMULATE A 16-BIT SLAVE
AN1475	DEVELOPING AN ST7265X MASS STORAGE APPLICATION
AN1504	STARTING A PWM SIGNAL DIRECTLY AT HIGH LEVEL USING THE ST7 16-BIT TIMER
PRODUCT EVALUATION	
AN 910	PERFORMANCE BENCHMARKING
AN 990	ST7 BENEFITS VERSUS INDUSTRY STANDARD
AN1077	OVERVIEW OF ENHANCED CAN CONTROLLERS FOR ST7 AND ST9 MCUS
AN1086	U435 CAN-DO SOLUTIONS FOR CAR MULTIPLEXING
AN1150	BENCHMARK ST72 VS PC16
AN1151	PERFORMANCE COMPARISON BETWEEN ST72254 & PC16F876
AN1278	LIN (LOCAL INTERCONNECT NETWORK) SOLUTIONS
PRODUCT MIGRATION	
AN1131	MIGRATING APPLICATIONS FROM ST72511/311/214/124 TO ST72521/321/324
AN1322	MIGRATING AN APPLICATION FROM ST7263 REV.B TO ST7263B
AN1365	GUIDELINES FOR MIGRATING ST72C254 APPLICATION TO ST72F264
PRODUCT OPTIMIZATION	