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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Obsolete
Number of LABs/CLBs	224
Number of Logic Elements/Cells	-
Total RAM Bits	46064
Number of I/O	120
Number of Gates	4500
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xc3064l-8tq144c

Introduction

XC3000-Series Field Programmable Gate Arrays (FPGAs) provide a group of high-performance, high-density, digital integrated circuits. Their regular, extendable, flexible, user-programmable array architecture is composed of a configuration program store plus three types of configurable elements: a perimeter of I/O Blocks (IOBs), a core array of Configurable Logic Bocks (CLBs) and resources for interconnection. The general structure of an FPGA is shown in Figure 2. The development system provides schematic capture and auto place-and-route for design entry. Logic and timing simulation, and in-circuit emulation are available as design verification alternatives. The design editor is used for interactive design optimization, and to compile the data pattern that represents the configuration program.

The FPGA user logic functions and interconnections are determined by the configuration program data stored in internal static memory cells. The program can be loaded in any of several modes to accommodate various system requirements. The program data resides externally in an EEPROM, EPROM or ROM on the application circuit board, or on a floppy disk or hard disk. On-chip initialization logic provides for optional automatic loading of program data at power-up. The companion XC17XX Serial Configuration PROMs provide a very simple serial configuration program storage in a one-time programmable package.

The XC3000 Field Programmable Gate Array families provide a variety of logic capacities, package styles, temperature ranges and speed grades.

XC3000 Series Overview

There are now four distinct family groupings within the XC3000 Series of FPGA devices:

- XC3000A Family
- XC3000L Family
- XC3100A Family
- XC3100L Family

All four families share a common architecture, development software, design and programming methodology, and also common package pin-outs. An extensive Product Description covers these common aspects.

Detailed parametric information for the XC3000A, XC3000L, XC3100A, and XC3100L product families is then provided. (The XC3000 and XC3100 families are not recommended for new designs.)

Here is a simple overview of those XC3000 products currently emphasized:

- **XC3000A Family** — The XC3000A is an enhanced version of the basic XC3000 family, featuring additional interconnect resources and other user-friendly enhancements.
- **XC3000L Family** — The XC3000L is identical in architecture and features to the XC3000A family, but operates at a nominal supply voltage of 3.3 V. The XC3000L is the right solution for battery-operated and low-power applications.
- **XC3100A Family** — The XC3100A is a performance-optimized relative of the XC3000A family. While both families are bitstream and footprint compatible, the XC3100A family extends toggle rates to 370 MHz and in-system performance to over 80 MHz. The XC3100A family also offers one additional array size, the XC3195A.
- **XC3100L Family** — The XC3100L is identical in architectures and features to the XC3100A family, but operates at a nominal supply voltage of 3.3V.

Figure 1 illustrates the relationships between the families. Compared to the original XC3000 family, XC3000A offers additional functionality and increased speed. The XC3000L family offers the same additional functionality, but reduced speed due to its lower supply voltage of 3.3 V. The XC3100A family offers substantially higher speed and higher density with the XC3195A.

New XC3000 Series Compared to Original XC3000 Family

For readers already familiar with the original XC3000 family of FPGAs, the major new features in the XC3000A, XC3000L, XC3100A, and XC3100L families are listed in this section.

All of these new families are upward-compatible extensions of the original XC3000 FPGA architecture. Any bitstream used to configure an XC3000 device will configure the corresponding XC3000A, XC3000L, XC3100A, or XC3100L device exactly the same way.

The XC3100A and XC3100L FPGA architectures are upward-compatible extensions of the XC3000A and XC3000L architectures. Any bitstream used to configure an XC3000A or XC3000L device will configure the corresponding XC3100A or XC3100L device exactly the same way.
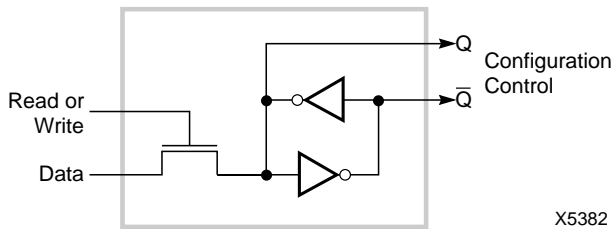
Figure 3: Static Configuration Memory Cell.
It is loaded with one bit of configuration program and controls one program selection in the Field Programmable Gate Array.

The memory cell outputs Q and $\overline{Q}$ use ground and V_{CC} levels and provide continuous, direct control. The additional capacitive load together with the absence of address decoding and sense amplifiers provide high stability to the cell. Due to the structure of the configuration memory cells, they are not affected by extreme power-supply excursions or very high levels of alpha particle radiation. In reliability testing, no soft errors have been observed even in the presence of very high doses of alpha radiation.

The method of loading the configuration data is selectable. Two methods use serial data, while three use byte-wide data. The internal configuration logic utilizes framing information, embedded in the program data by the development system, to direct memory-cell loading. The serial-data framing and length-count preamble provide programming compatibility for mixes of various FPGA device devices in a synchronous, serial, daisy-chain fashion.

I/O Block

Each user-configurable IOB shown in Figure 4, provides an interface between the external package pin of the device and the internal user logic. Each IOB includes both registered and direct input paths. Each IOB provides a programmable 3-state output buffer, which may be driven by a registered or direct output signal. Configuration options allow each IOB an inversion, a controlled slew rate and a high impedance pull-up. Each input circuit also provides input clamping diodes to provide electrostatic protection, and circuits to inhibit latch-up produced by input currents.
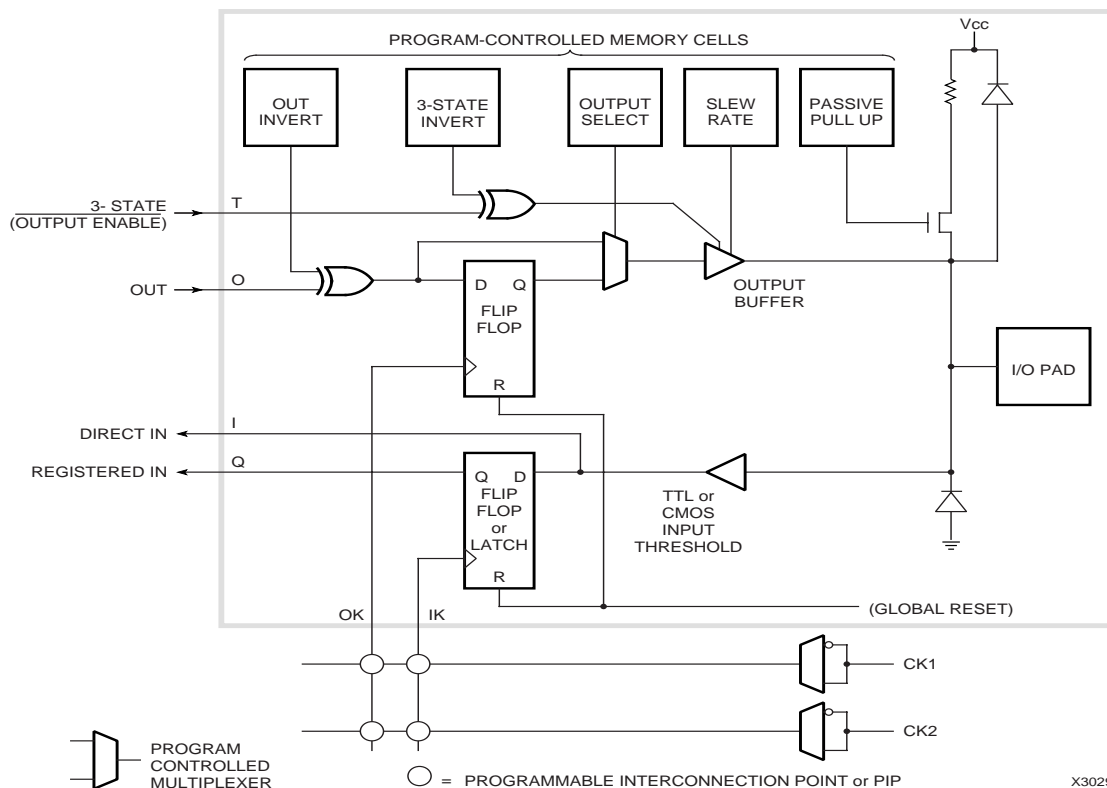
7

Figure 4: Input/Output Block.
Each IOB includes input and output storage elements and I/O options selected by configuration memory cells. A choice of two clocks is available on each die edge. The polarity of each clock line (not each flip-flop or latch) is programmable. A clock line that triggers the flip-flop on the rising edge is an active Low Latch Enable (Latch transparent) signal and vice versa. Passive pull-up can only be enabled on inputs, not on outputs. All user inputs are programmed for TTL or CMOS thresholds.

The input-buffer portion of each IOB provides threshold detection to translate external signals applied to the package pin to internal logic levels. The global input-buffer threshold of the IOBs can be programmed to be compatible with either TTL or CMOS levels. The buffered input signal drives the data input of a storage element, which may be configured as either a flip-flop or a latch. The clocking polarity (rising/falling edge-triggered flip-flop, High/Low transparent latch) is programmable for each of the two clock lines on each of the four die edges. Note that a clock line driving a *rising* edge-triggered flip-flop makes any latch driven by the same line on the same edge Low-level transparent and vice versa (*falling* edge, *High* transparent). All Xilinx primitives in the supported schematic-entry packages, however, are positive edge-triggered flip-flops or High transparent latches. When one clock line must drive flip-flops as well as latches, it is necessary to compensate for the difference in clocking polarities with an additional inverter either in the flip-flop clock input or the latch-enable input. I/O storage elements are reset during configuration or by the active-Low chip RESET input. Both direct input (from IOB pin I) and registered input (from IOB pin Q) signals are available for interconnect.

For reliable operation, inputs should have transition times of less than 100 ns and should not be left floating. Floating CMOS input-pin circuits might be at threshold and produce oscillations. This can produce additional power dissipation and system noise. A typical hysteresis of about 300 mV reduces sensitivity to input noise. Each user IOB includes a programmable high-impedance pull-up resistor, which may be selected by the program to provide a constant High for otherwise undriven package pins. Although the Field Programmable Gate Array provides circuitry to provide input protection for electrostatic discharge, normal CMOS handling precautions should be observed.

Flip-flop loop delays for the IOB and logic-block flip-flops are short, providing good performance under asynchronous clock and data conditions. Short loop delays minimize the probability of a metastable condition that can result from assertion of the clock during data transitions. Because of the short-loop-delay characteristic in the Field Programmable Gate Array, the IOB flip-flops can be used to synchronize external signals applied to the device. Once synchronized in the IOB, the signals can be used internally without further consideration of their clock relative timing, except as it applies to the internal logic and routing-path delays.

IOB output buffers provide CMOS-compatible 4-mA source-or-sink drive for high fan-out CMOS or TTL- compatible signal levels (8 mA in the XC3100A family). The network driving IOB pin O becomes the registered or direct data source for the output buffer. The 3-state control signal (IOB) pin T can control output activity. An open-drain output may be obtained by using the same signal for driving the output and 3-state signal nets so that the buffer output is enabled only for a Low.

Configuration program bits for each IOB control features such as optional output register, logic signal inversion, and 3-state and slew-rate control of the output.

The program-controlled memory cells of Figure 4 control the following options.

- Logic inversion of the output is controlled by one configuration program bit per IOB.
- Logic 3-state control of each IOB output buffer is determined by the states of configuration program bits that turn the buffer on, or off, or select the output buffer 3-state control interconnection (IOB pin T). When this IOB output control signal is High, a logic one, the buffer is disabled and the package pin is high impedance. When this IOB output control signal is Low, a logic zero, the buffer is enabled and the package pin is active. Inversion of the buffer 3-state control-logic sense (output enable) is controlled by an additional configuration program bit.
- Direct or registered output is selectable for each IOB. The register uses a positive-edge, clocked flip-flop. The clock source may be supplied (IOB pin OK) by either of two metal lines available along each die edge. Each of these lines is driven by an invertible buffer.
- Increased output transition speed can be selected to improve critical timing. Slower transitions reduce capacitive-load peak currents of non-critical outputs and minimize system noise.
- An internal high-impedance pull-up resistor (active by default) prevents unconnected inputs from floating.

Unlike the original XC3000 series, the XC3000A, XC3000L, XC3100A, and XC3100L families include the Soft Startup feature. When the configuration process is finished and the device starts up in user mode, the first activation of the outputs is automatically slew-rate limited. This feature avoids potential ground bounce when all outputs are turned on simultaneously. After start-up, the slew rate of the individual outputs is determined by the individual configuration option.

Summary of I/O Options

- Inputs
  - Direct
  - Flip-flop/latch
  - CMOS/TTL threshold (chip inputs)
  - Pull-up resistor/open circuit
- Outputs
  - Direct/registered
  - Inverted/not
  - 3-state/on/off
  - Full speed/slew limited
  - 3-state/output enable (inverse)
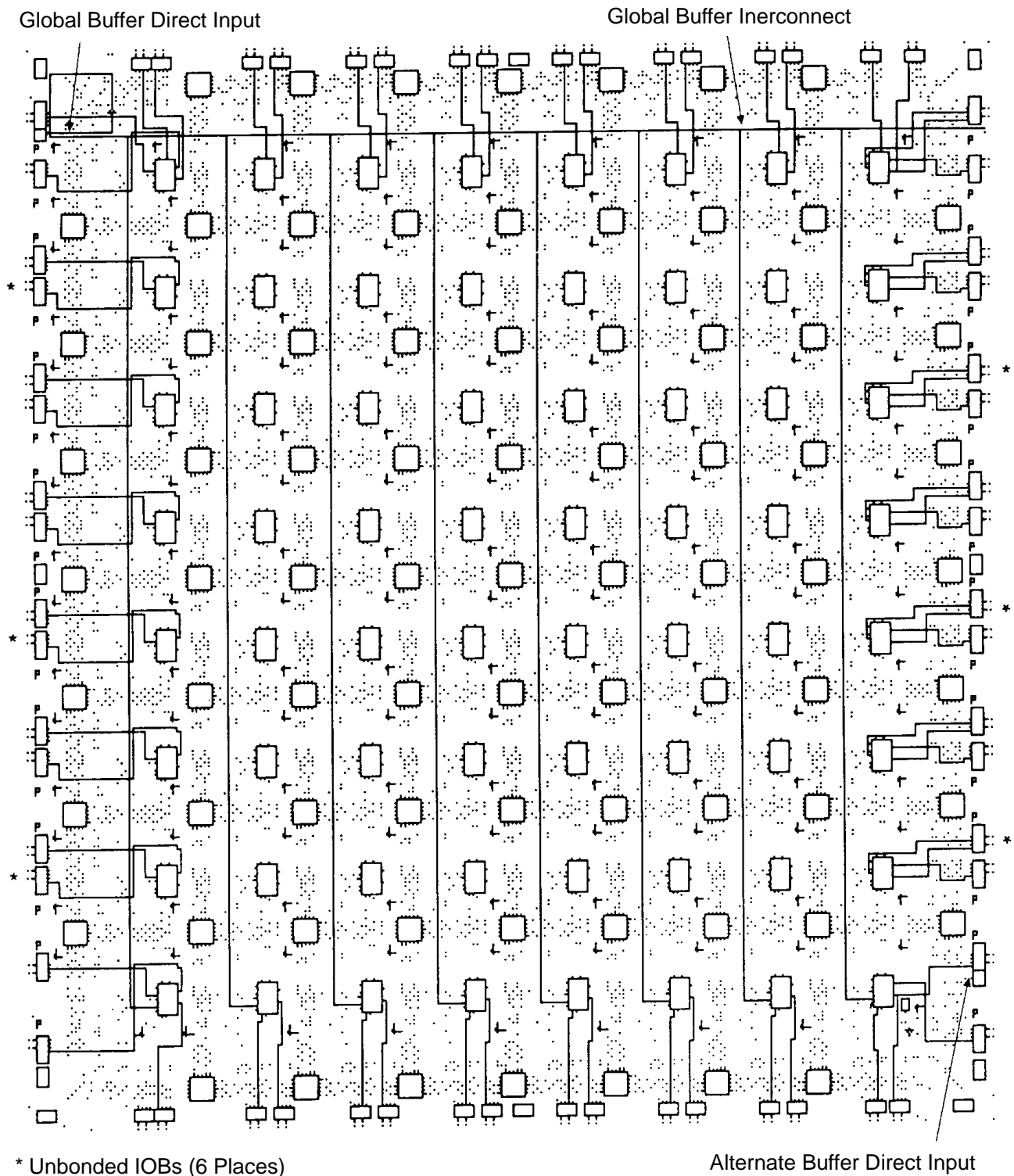
Global Buffer Direct Input

Global Buffer Inerconnect

* Unbonded IOBs (6 Places)

Alternate Buffer Direct Input

Figure 13:   XC3020A Die-Edge IOBs. The XC3020A die-edge IOBs are provided with direct access to adjacent CLBs.
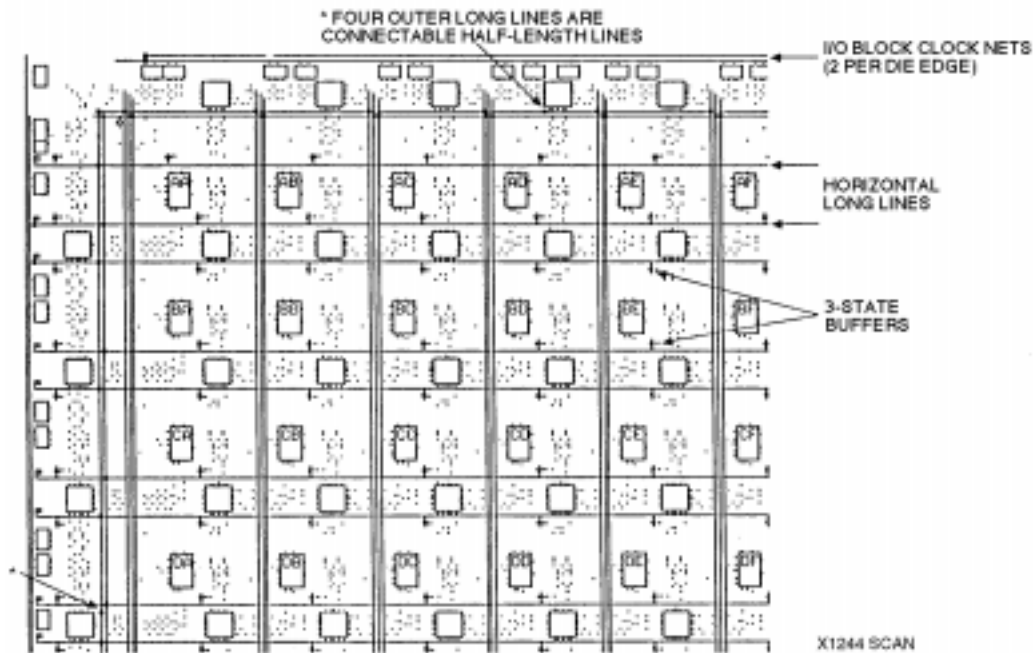
Figure 15: Programmable Interconnection of Longlines. This is provided at the edges of the routing area. Three-state buffers allow the use of horizontal Longlines to form on-chip wired AND and multiplexed buses. The left two non-clock vertical Longlines per column (except XC3020A) and the outer perimeter Longlines may be programmed as connectable half-length lines.

$$Z = D_A \bullet D_B \bullet D_C \bullet_{...} \bullet D_N$$
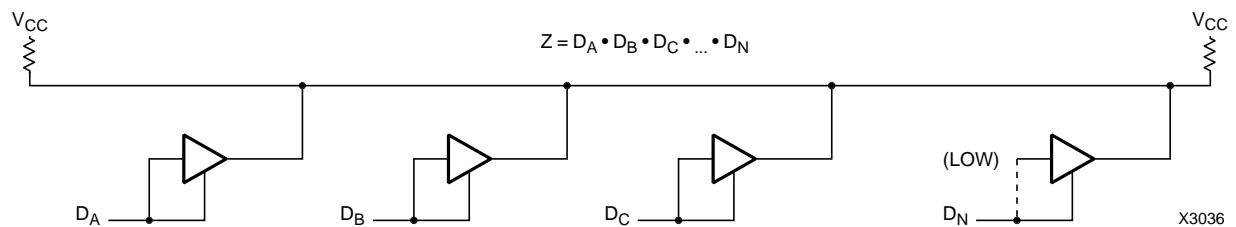
Figure 16: 3-State Buffers Implement a Wired-AND Function. When all the buffer 3-state lines are High, (high impedance), the pull-up resistor(s) provide the High output. The buffer inputs are driven by the control signals or a Low.

$$Z = D_A \bullet \overline{A} + D_B \bullet \overline{B} + D_C \bullet \overline{C} + ... + D_N \bullet \overline{N}$$

Figure 17: 3-State Buffers Implement a Multiplexer. The selection is accomplished by the buffer 3-state signal.

A buffer in the upper left corner of the FPGA chip drives a global net which is available to all K inputs of logic blocks. Using the global buffer for a clock signal provides a skew-free, high fan-out, synchronized clock for use at any or all of the IOBs and CLBs. Configuration bits for the K input to each logic block can select this global line or another routing resource as the clock source for its flip-flops. This net may also be pr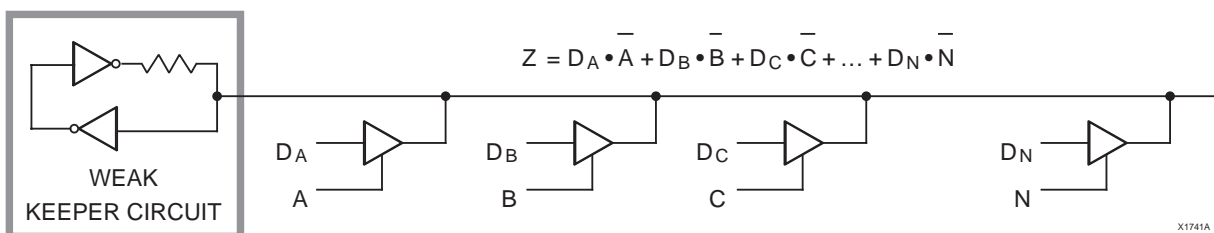ogrammed to drive the die edge clock lines for IOB use. An enhanced speed, CMOS threshold, direct access to this buffer is available at the second pad from the top of the left die edge.

A buffer in the lower right corner of the array drives a horizontal Longline that can drive programmed connections to a vertical Longline in each interconnection column. This alternate buffer also has low skew and high fan-out. The network formed by this alternate buffer's Longlines can be selected to drive the K inputs of the CLBs. CMOS threshold, high speed access to this buffer is available from the third pad from the bottom of the right die edge.

Internal Busses

A pair of 3-state buffers, located adjacent to each CLB, permits logic to drive the horizontal Longlines. Logic operation of the 3-state buffer controls allows them to implement wide multiplexing functions. Any 3-state buffer input can be selected as drive for the horizontal long-line bus by applying a Low logic level on its 3-state control line. See Figure 16. The user is required to avoid contention which can result from multiple drivers with opposing logic levels. Control of the 3-state input by the same signal that drives the buffer input, creates an open-drain wired-AND function. A logic High on both buffer inputs creates a high impedance, which represents no contention. A logic Low enables the buffer to drive the Longline Low. See Figure 17. Pull-up resistors are available at each end of the Longline to provide a High output when all connected buffers are non-conducting. This forms fast, wide gating functions. When data drives the inputs, and separate signals drive the 3-state control lines, these buffers form multiplexers (3-state busses). In this case, care must be used to prevent contention through multiple active buffers of conflicting levels on a common line. Each horizontal Longline is also driven by a weak keeper circuit that prevents undefined floating levels by maintaining the previous logic level when the line is not driven by an active buffer or a pull-up resistor. Figure 18 shows 3-state buffers, Longlines and pull-up resistors.
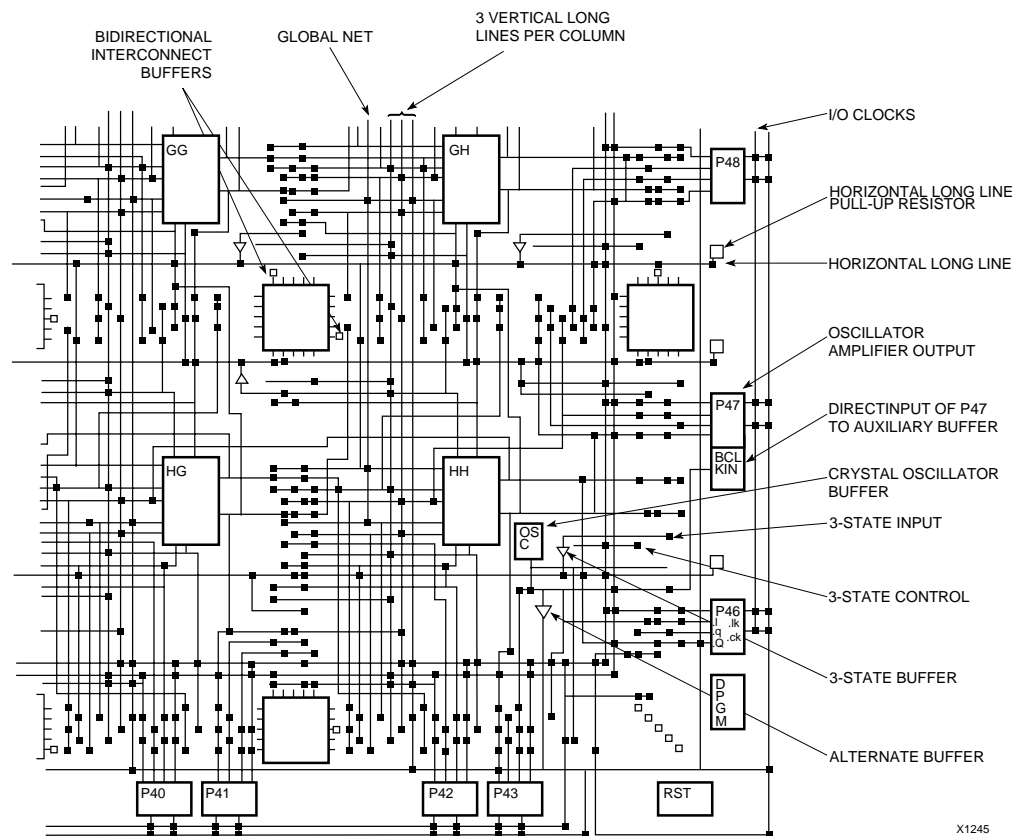

Figure 18: Design Editor.
An extra large view of possible interconnections in the lower right corner of the XC3020A.

Configuration

Initialization Phase

An internal power-on-reset circuit is triggered when power is applied. When V_{CC} reaches the voltage at which portions of the FPGA device begin to operate (nominally 2.5 to 3 V), the programmable I/O output buffers are 3-stated and a high-impedance pull-up resistor is provided for the user I/O pins. A time-out delay is initiated to allow the power supply voltage to stabilize. During this time the power-down mode is inhibited. The Initialization state time-out (about 11 to 33 ms) is determined by a 14-bit counter driven by a self-generated internal timer. This nominal 1-MHz timer is subject to variations with process, temperature and power supply. As shown in Table 1, five configuration mode choices are available as determined by the input levels of three mode pins; M0, M1 and M2.

Table 1: Configuration Mode Choices

M0	M1	M2	CCLK	Mode	Data
0	0	0	output	Master	Bit Serial
0	0	1	output	Master	Byte Wide Addr. = 0000 up
0	1	0	—	reserved	—
0	1	1	output	Master	Byte Wide Addr. = FFFF down
1	0	0	—	reserved	—
1	0	1	output	Peripheral	Byte Wide
1	1	0	—	reserved	—
1	1	1	input	Slave	Bit Serial

In Master configuration modes, the device becomes the source of the Configuration Clock (CCLK). The beginning of configuration of devices using Peripheral or Slave modes must be delayed long enough for their initialization to be completed. An FPGA with mode lines selecting a Master configuration mode extends its initialization state using four times the delay (43 to 130 ms) to assure that all daisy-chained slave devices, which it may be driving, will be ready even if the master is very fast, and the slave(s) very slow. Figure 20 shows the state sequences. At the end of Initialization, the device enters the Clear state where it clears the configuration memory. The active Low, open-drain initialization signal INIT indicates when the Initialization and Clear states are complete. The FPGA tests for the absence of an external active Low RESET before it makes a final sample of the mode lines and enters the Configuration state. An external wired-AND of one or more INIT pins can be used to control configuration by the assertion of the active-Low RESET of a master mode device or to signal a processor that the FPGAs are not yet initialized.

If a configuration has begun, a re-assertion of RESET for a minimum of three internal timer cycles will be recognized and the FPGA will initiate an abort, returning to the Clear state to clear the partially loaded configuration memory words. The FPGA will then resample RESET and the mode lines before re-entering the Configuration state.

During configuration, the XC3000A, XC3000L, XC3100A, and XC3100L devices check the bit-stream format for stop bits in the appropriate positions. Any error terminates the configuration and pulls INIT Low.
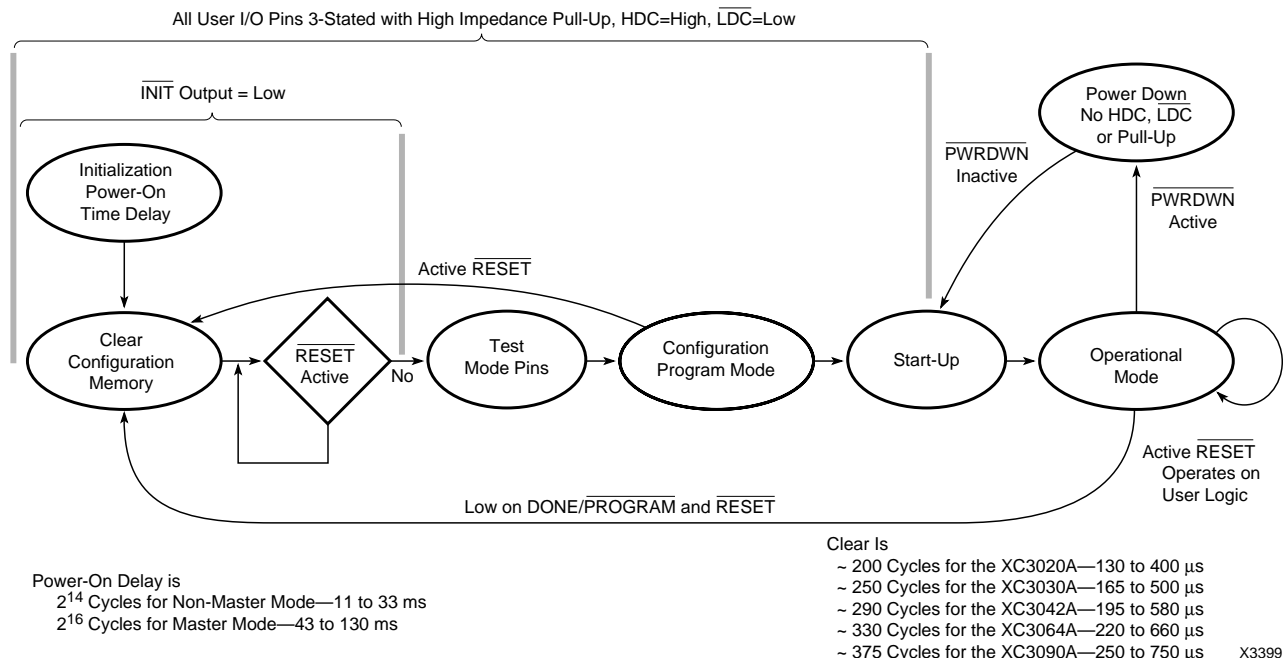

Figure 20: A State Diagram of the Configuration Process for Power-up and Reprogram.

RESET Timing

As with DONE timing, the timing of the release of the internal reset can be controlled to occur either a CCLK cycle before, or after, the outputs going active. See Figure 22. This reset keeps all user programmable flip-flops and latches in a zero state during configuration.

Crystal Oscillator Division

A selection allows the user to incorporate a dedicated divide-by-two flip-flop between the crystal oscillator and the alternate clock line. This guarantees a symmetrical clock signal. Although the frequency stability of a crystal oscillator is very good, the symmetry of its waveform can be affected by bias or feedback drive.

Bitstream Error Checking

Bitstream error checking protects against erroneous configuration.

Each Xilinx FPGA bitstream consists of a 40-bit preamble, followed by a device-specific number of data frames. The number of bits per frame is also device-specific; however, each frame ends with three stop bits (111) followed by a start bit for the next frame (0).

All devices in all XC3000 families start reading in a new frame when they find the first 0 after the end of the previous frame. An original XC3000 device does not check for the correct stop bits, but XC3000A, XC3100A, XC3000L, and XC3100L devices check that the last three bits of any frame are actually 111.

Under normal circumstances, all these FPGAs behave the same way; however, if the bitstream is corrupted, an XC3000 device will always start a new frame as soon as it finds the first 0 after the end of the previous frame, even if the data is completely wrong or out-of-sync. Given sufficient zeros in the data stream, the device will also go Done, but with incorrect configuration and the possibility of internal contention.

An XC3000A/XC3100A/XC3000L/XC3100L device starts any new frame only if the three preceding bits are all ones. If this check fails, it pulls INIT Low and stops the internal configuration, although the Master CCLK keeps running. The user must then start a new configuration by applying a >6 µs Low level on RESET.

This simple check does not protect against random bit errors, but it offers almost 100 percent protection against erroneous configuration files, defective configuration data sources, synchronization errors between configuration source and FPGA, or PC-board level defects, such as broken lines or solder-bridges.

Reset Spike Protection

A separate modification slows down the RESET input before configuration by using a two-stage shift register driven from the internal clock. It tolerates submicrosecond High spikes on RESET before configuration. The XC3000 master can be connected like an XC4000 master, but with its RESET input used instead of INIT. (On XC3000, INIT is output only).

Soft Start-up

After configuration, the outputs of all FPGAs in a daisy-chain become active simultaneously, as a result of the same CCLK edge. In the original XC3000/3100 devices, each output becomes active in either fast or slew-rate limited mode, depending on the way it is configured. This can lead to large ground-bounce signals. In XC3000A, XC3000L, XC3100A, and XC3100L devices, all outputs become active first in slew-rate limited mode, reducing the ground bounce. After this soft start-up, each individual output slew rate is again controlled by the respective configuration bit.
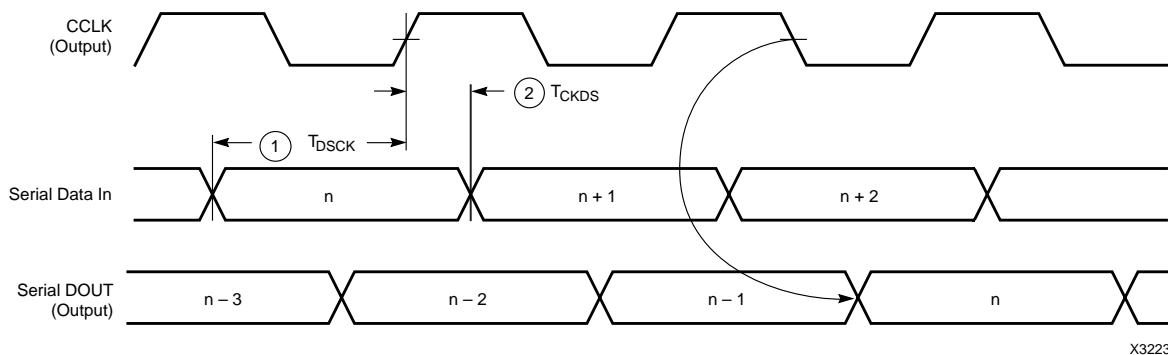
X3223

	Description		Symbol	Min	Max	Units
CCLK	Data In setup	1	T_{DSCK}	60		ns
	Data In hold	2	C_{KDS}	0		ns

Notes: 1. At power-up, V_{CC} must rise from 2.0 V to V_{CC} min in less than 25 ms. If this is not possible, configuration can be delayed by holding RESET Low until V_{CC} has reached 4.0 V (2.5 V for the XC3000L). A very long V_{CC} rise time of >100 ms, or a non-monotonically rising V_{CC} may require >6-μs High level on RESET, followed by a >6-μs Low level on $\overline{RESET}$ and D/$\overline{P}$ after VCC has reached 4.0 V (2.5 V for the XC3000L).
2. Configuration can be controlled by holding $\overline{RESET}$ Low with or until after the $\overline{INIT}$ of all daisy-chain slave-mode devices is High.
3. Master-serial-mode timing is based on slave-mode testing.

Figure 24: Master Serial Mode Programming Switching Characteristics

WRITE TO FPGA

X5992
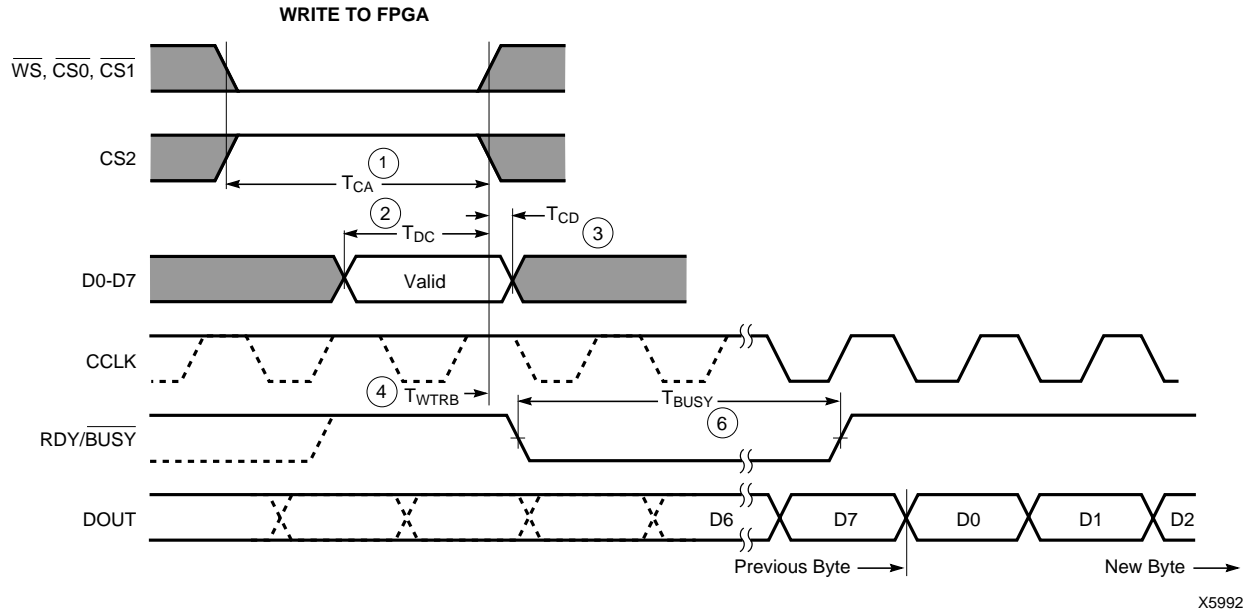
	Description		Symbol	Min	Max	Units
WRITE	Effective Write time required (Assertion of $\overline{CS0}$, $\overline{CS1}$, $CS2$, $\overline{WS}$)	1	T_{CA}	100		ns
	DIN Setup time required	2	T_{DC}	60		ns
	DIN Hold time required	3	T_{CD}	0		ns
	RDY/BUSY delay after end of $\overline{WS}$	4	T_{WTRB}		60	ns
RDY	Earliest next $\overline{WS}$ after end of $\overline{BUSY}$	5	T_{RBWT}	0		ns
	$\overline{BUSY}$ Low time generated	6	T_{BUSY}	2.5	9	CCLK periods

Notes: 1. At power-up, V_{CC} must rise from 2.0 V to V_{CC} min in less than 25 ms. If this is not possible, configuration can be delayed by holding RESET Low until V_{CC} has reached 4.0 V (2.5 V for the XC3000L). A very long V_{CC} rise time of >100 ms, or a non-monotonically rising V_{CC} may require a >6-µs High level on $\overline{RESET}$, followed by a >6-µs Low level on $\overline{RESET}$ and D/$\overline{P}$ after V_{CC} has reached 4.0 V (2.5 V for the XC3000L).
2. Configuration must be delayed until the $\overline{INIT}$ of all FPGAs is High.
3. Time from end of $\overline{WS}$ to CCLK cycle for the new byte of data depends on completion of previous byte processing and the phase of the internal timing generator for CCLK.
4. CCLK and DOUT timing is tested in slave mode.
5. T_{BUSY} indicates that the double-buffered parallel-to-serial converter is not yet ready to receive new data. The shortest T_{BUSY} occurs when a byte is loaded into an empty parallel-to-serial converter. The longest TBUSY occurs when a new word is loaded into the input register before the second-level buffer has started shifting out data.

Note: *This timing diagram shows very relaxed requirements: Data need not be held beyond the rising edge of $\overline{WS}$. $\overline{BUSY}$ will go active within 60 ns after the end of $\overline{WS}$. BUSY will stay active for several microseconds. WS may be asserted immediately after the end of $\overline{BUSY}$.*

Figure 28: Peripheral Mode Programming Switching Characteristics

SPECIFIED WORST-CASE VALUES
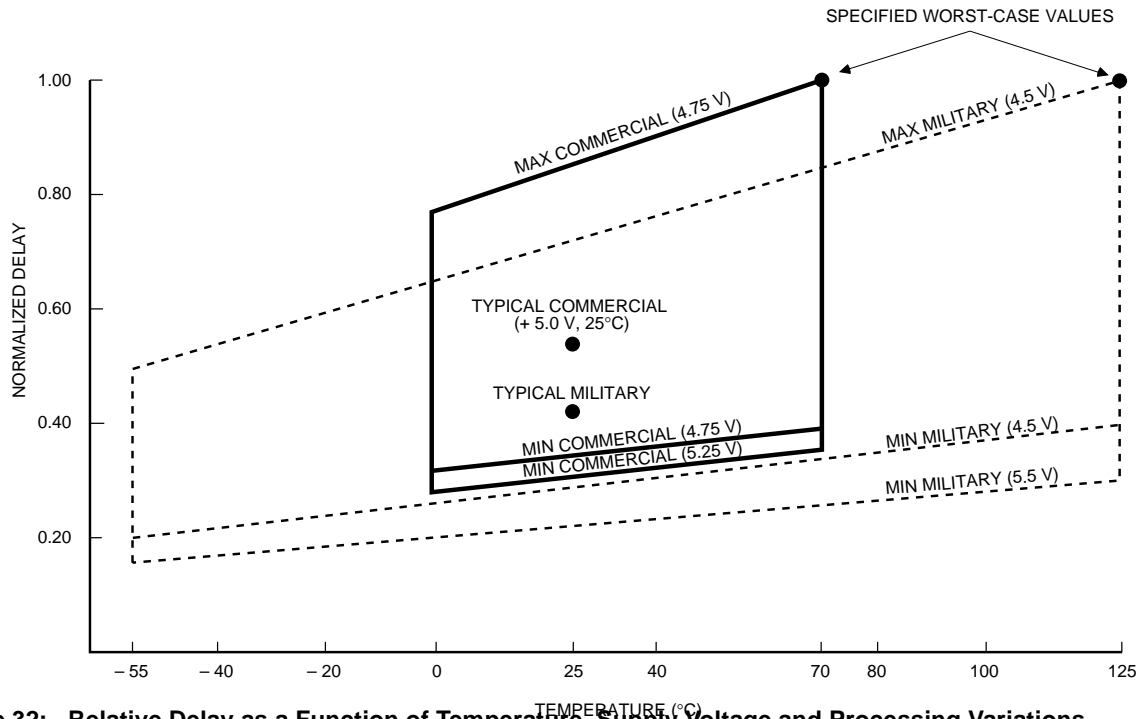
Figure 32:   Relative Delay as a Function of Temperature, Supply Voltage and Processing Variations   X6094

Figure 33:   Clock Rate as a Function of Logic Complexity (Number of Combinational Levels between Flip-Flops)

Power

Power Distribution

Power for the FPGA is distributed through a grid to achieve high noise immunity and isolation between logic and I/O. Inside the FPGA, a dedicated V_{CC} and ground ring surrounding the logic array provides power to the I/O drivers. An independent matrix of V_{CC} and groundlines supplies the interior logic of the device. This power distribution grid provides a stable supply and ground for all internal logic, providing the external package power pins are all connected and appropriately decoupled. Typically a 0.1-µF capacitor connected near the V_{CC} and ground pins will provide adequate decoupling.

Output buffers capable of driving the specified 4- or 8-mA loads under worst-case conditions may be capable of driving as much as 25 to 30 times that current in a best case. Noise can be reduced by minimizing external load capacitance and reducing simultaneous output transitions in the same direction. It may also be beneficial to locate heavily loaded output buffers near the ground pads. The I/O Block output buffers have a slew-limited mode which should be used where output rise and fall times are not speed critical. Slew-limited outputs maintain their dc drive capability, but generate less external reflections and internal noise.

Pin Descriptions

Permanently Dedicated Pins

V_{CC}

Two to eight (depending on package type) connections to the positive V supply voltage. All must be connected.

GND

Two to eight (depending on package type) connections to ground. All must be connected.
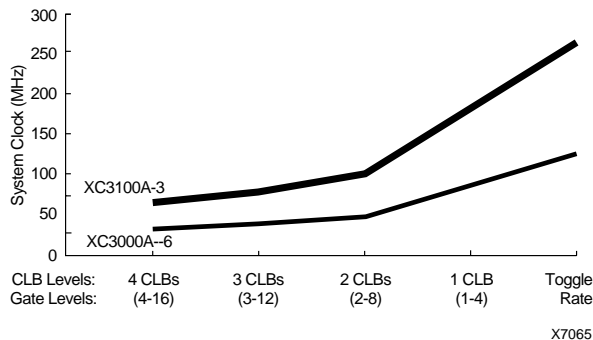
PWRDWN

A Low on this CMOS-compatible input stops all internal activity, but retains configuration. All flip-flops and latches are reset, all outputs are 3-stated, and all inputs are interpreted as High, independent of their actual level. When PWDWN returns High, the FPGA becomes operational with DONE Low for two cycles of the internal 1-MHz clock. Before and during configuration, PWRDWN must be High. If not used, PWRDWN must be tied to V_{CC}.

RESET

This is an active Low input which has three functions.

Prior to the start of configuration, a Low input will delay the start of the configuration process. An internal circuit senses the application of power and begins a minimal time-out cycle. When the time-out and RESET are complete, the levels of the M lines are sampled and configuration begins.

If RESET is asserted during a configuration, the FPGA is re-initialized and restarts the configuration at the termination of RESET.

If RESET is asserted after configuration is complete, it provides a global asynchronous RESET of all IOB and CLB storage elements of the FPGA.

CCLK

During configuration, Configuration Clock is an output of an FPGA in Master mode or Peripheral mode, but an input in Slave mode. During Readback, CCLK is a clock input for shifting configuration data out of the FPGA.

CCLK drives dynamic circuitry inside the FPGA. The Low time may, therefore, not exceed a few microseconds. When used as an input, CCLK must be "parked High". An internal pull-up resistor maintains High when the pin is not being driven.

DONE/PROG (D/P)

DONE is an open-drain output, configurable with or without an internal pull-up resistor of 2 to 8 k Ω. At the completion of configuration, the FPGA circuitry becomes active in a synchronous order; DONE is programmed to go active High one cycle either before or after the outputs go active.

Once configuration is done, a High-to-Low transition of this pin will cause an initialization of the FPGA and start a reconfiguration.

M0/RTRIG

As Mode 0, this input is sampled on power-on to determine the power-on delay (2^{14} cycles if M0 is High, 2^{16} cycles if M0 is Low). Before the start of configuration, this input is again sampled together with M1, M2 to determine the configuration mode to be used.

A Low-to-High input transition, after configuration is complete, acts as a Read Trigger and initiates a Readback of configuration and storage-element data clocked by CCLK. By selecting the appropriate Readback option when generating the bitstream, this operation may be limited to a single Readback, or be inhibited altogether.

M1/RDATA

As Mode 1, this input and M0, M2 are sampled before the start of configuration to establish the configuration mode to be used. If Readback is never used, M1 can be tied directly to ground or V_{CC}. If Readback is ever used, M1 must use a 5-kΩ resistor to ground or V_{CC}, to accommodate the RDATA output.

As an active-Low Read Data, after configuration is complete, this pin is the output of the Readback data.

User I/O Pins That Can Have Special Functions

M2

During configuration, this input has a weak pull-up resistor. Together with M0 and M1, it is sampled before the start of configuration to establish the configuration mode to be used. After configuration, this pin is a user-programmable I/O pin.

HDC

During configuration, this output is held at a High level to indicate that configuration is not yet complete. After configuration, this pin is a user-programmable I/O pin.

LDC

During Configuration, this output is held at a Low level to indicate that the configuration is not yet complete. After configuration, this pin is a user-programmable I/O pin. LDC is particularly useful in Master mode as a Low enable for an EPROM, but it must then be programmed as a High after configuration.

INIT

This is an active Low open-drain output with a weak pull-up and is held Low during the power stabilization and internal clearing of the configuration memory. It can be used to indicate status to a configuring microprocessor or, as a wired

XC3000A Absolute Maximum Ratings

Symbol	Description		Units
V_{CC}	Supply voltage relative to GND	−0.5 to +7.0	V
V_{IN}	Input voltage with respect to GND	−0.5 to V_{CC} +0.5	V
V_{TS}	Voltage applied to 3-state output	−0.5 to V_{CC} +0.5	V
T_{STG}	Storage temperature (ambient)	−65 to +150	°C
T_{SOL}	Maximum soldering temperature (10 s @ 1/16 in.)	+260	°C
T_J	Junction temperature plastic	+125	°C
	Junction temperature ceramic	+150	°C

Note: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions is not implied. Exposure to Absolute Maximum Ratings conditions for extended periods of time may affect device reliability.

XC3000A Global Buffer Switching Characteristics Guidelines

	Speed Grade	-7	-6	
Description	Symbol	Max	Max	Units
Global and Alternate Clock Distribution[1] Either: **Normal** IOB input pad through clock buffer to any CLB or IOB clock input	T_{PID}	7.5	7.0	ns
Or: **Fast** (CMOS only) input pad through clock buffer to any CLB or IOB clock input	T_{PIDC}	6.0	5.7	ns
TBUF driving a Horizontal Longline (L.L.)[1] I to L.L. while T is Low (buffer active)	T_{IO}	4.5	4.0	ns
T↓ to L.L. active and valid with single pull-up resistor	T_{ON}	9.0	8.0	ns
T↓ to L.L. active and valid with pair of pull-up resistors	T_{ON}	11.0	10.0	ns
T↑ to L.L. High with single pull-up resistor	T_{PUS}	16.0	14.0	ns
T↑ to L.L. High with pair of pull-up resistors	T_{PUF}	10.0	8.0	ns
BIDI Bidirectional buffer delay	T_{BIDI}	1.7	1.5	ns

Note: 1. Timing is based on the XC3042A, for other devices see timing calculator.

XC3000A IOB Switching Characteristics Guidelines (continued)
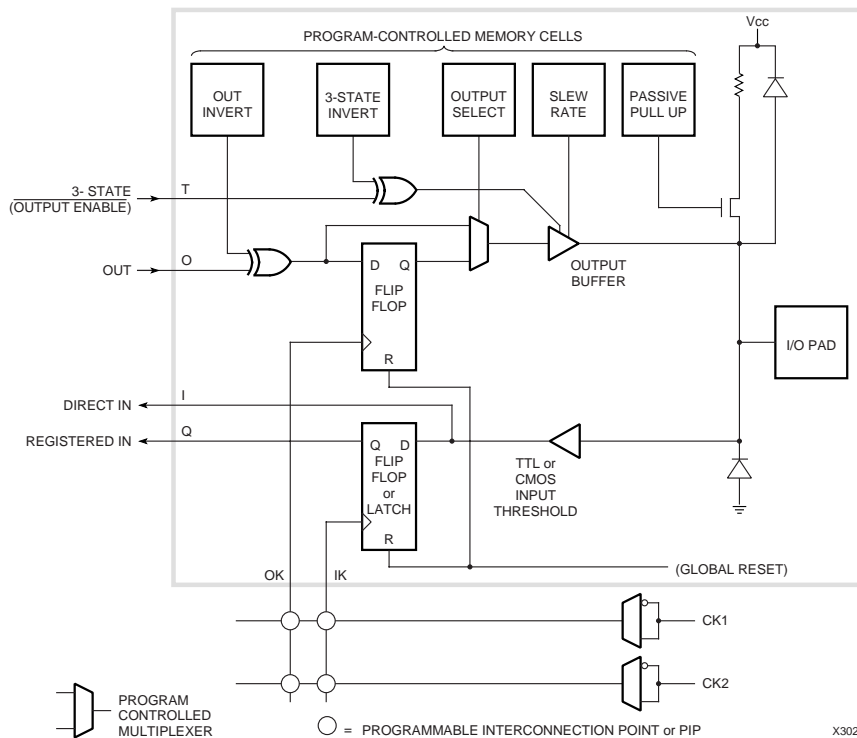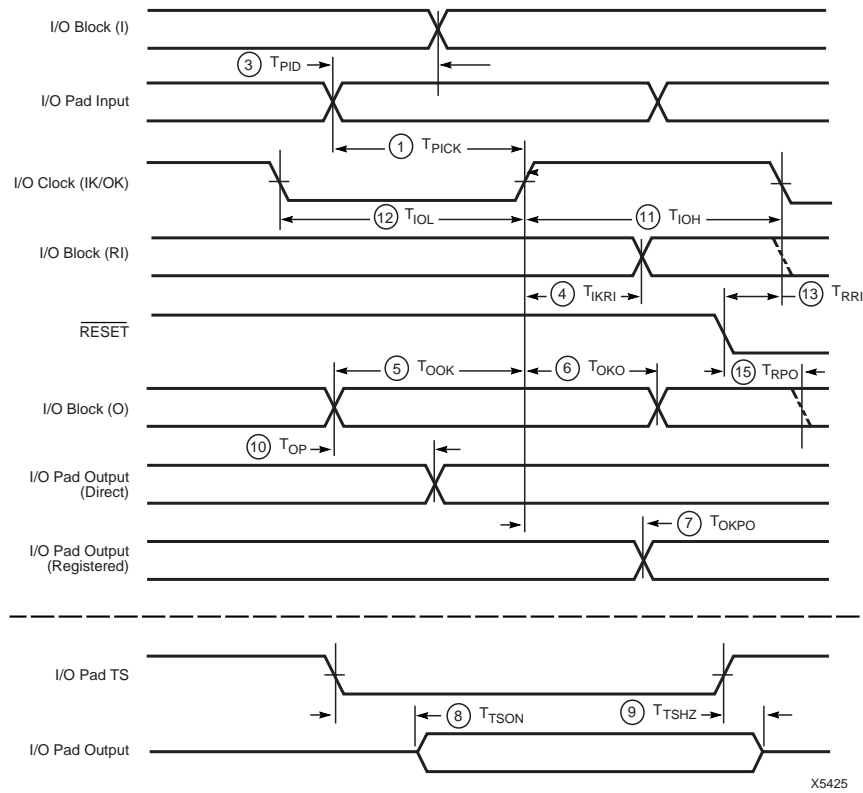
X5425

X3029

XC3000L CLB Switching Characteristics Guidelines (continued)
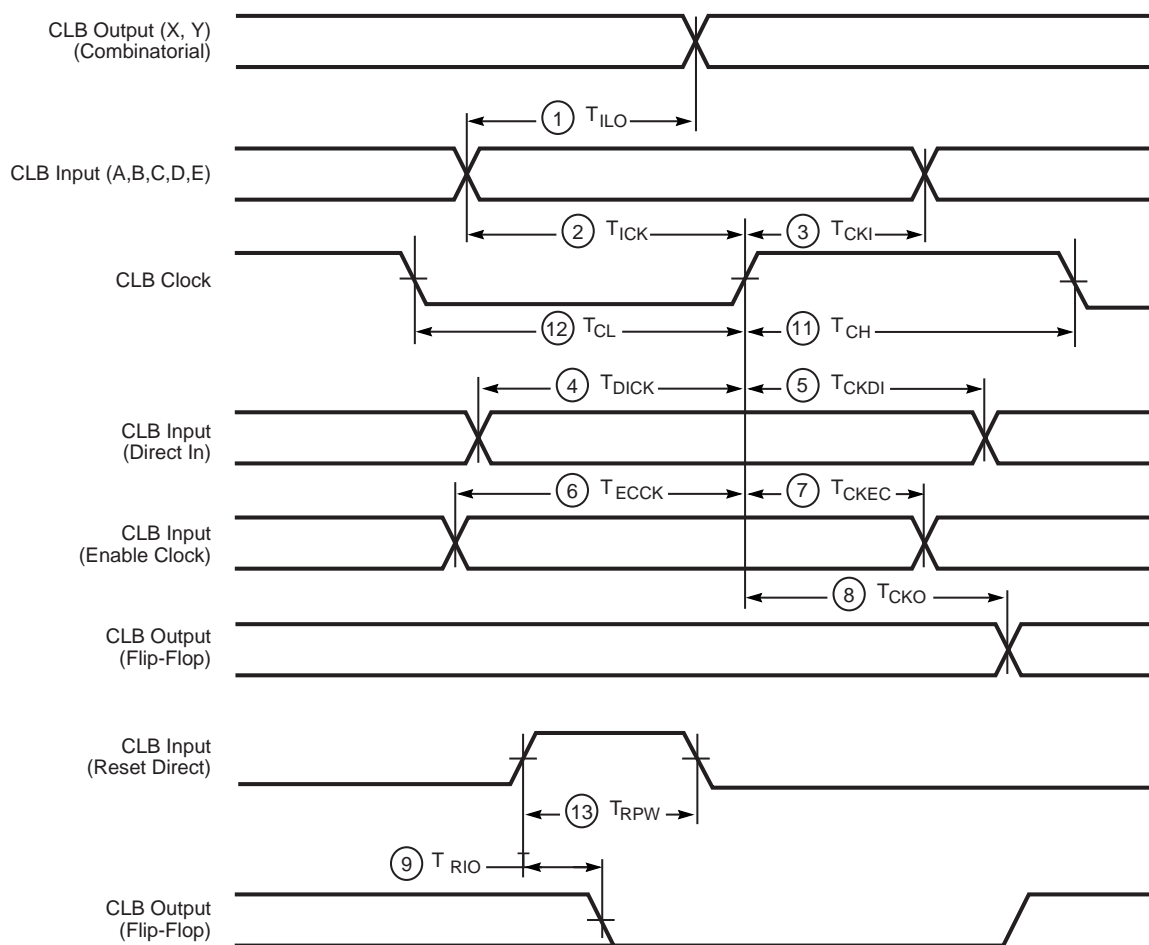
X5424

XC3000L IOB Switching Characteristics Guidelines (continued)

X5425

X3029

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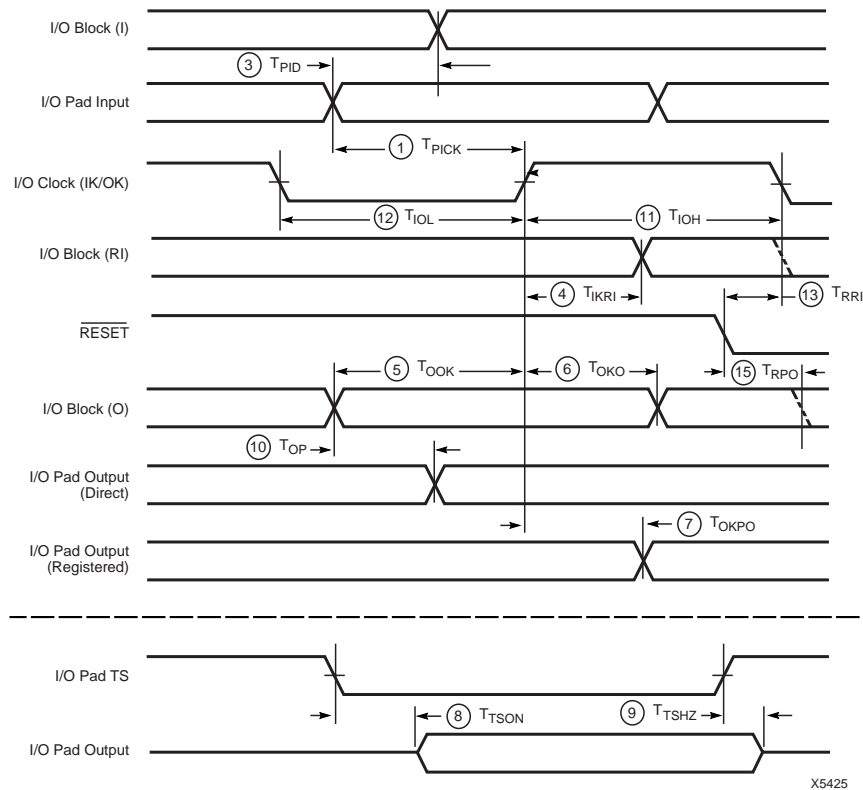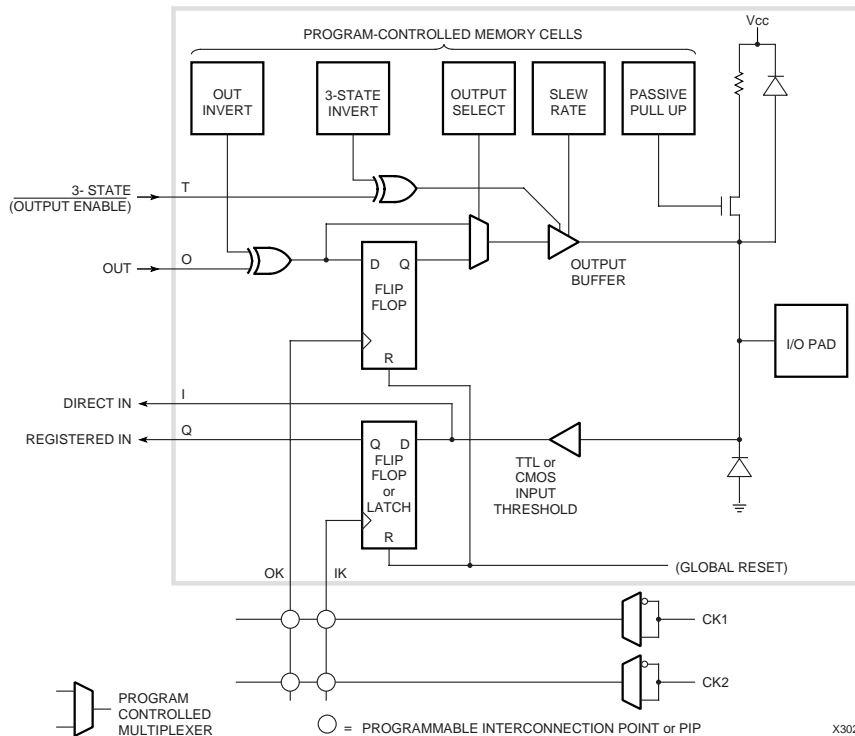
XC3100A IOB Switching Characteristics Guidelines

Testing of the switching parameters is modeled after testing methods specified by MIL-M-38510/605. All devices are 100% functionally tested. Since many internal timing parameters cannot be measured directly, they are derived from benchmark timing patterns. The following guidelines reflect worst-case values over the recommended operating conditions. For more detailed, more precise, and more up-to-date timing information, use the values provided by the timing calculator and used in the simulator.

|	Speed Grade		-4		-3		-2		-1		-09		
Description		Symbol	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Units
Propagation Delays (Input)													
Pad to Direct In (I)	3	T_{PID}		2.5		2.2		2.0		1.7		1.55	ns
Pad to Registered In (Q)													
with latch transparent(XC3100A)Clock (IK)		T_{PTG}		12.0		11.0		11.0		10.0		9.2	ns
to Registered In (Q)	4	T_{IKRI}		2.5		2.2		1.9		1.7		1.55	ns
Set-up Time (Input)													
Pad to Clock (IK) set-up time													
XC3120A, XC3130A	1	T_{PICK}	10.6		9.4		8.9		8.0		7.2		ns
XC3142A			10.7		9.5		9.0		8.1		7.3		ns
XC3164A			11.0		9.7		9.2		8.3		7.5		ns
XC3190A			11.2		9.9		9.4		8.5		7.7		ns
XC3195A			11.6		10.3		9.8		8.9		8.1		ns
Propagation Delays (Output)													
Clock (OK) to Pad (fast)	7	T_{OKPO}		5.0		4.4		3.7		3.4		3.3	ns
same (slew rate limited)	7	T_{OKPO}		12.0		10.0		9.7		8.4		6.9	ns
Output (O) to Pad (fast)	10	T_{OPF}		3.7		3.3		3.0		3.0		2.9	ns
same (slew-rate limited)													ns
(XC3100A)	10	T_{OPS}		11.0		9.0		8.7		8.0		6.5	ns
3-state to Pad													
begin hi-Z (fast)	9	T_{TSHZ}		6.2		5.5		5.0		4.5		4.05	ns
same (slew-rate limited)	9	T_{TSHZ}		6.2		5.5		5.0		4.5		4.05	ns
3-state to Pad													
active and valid (fast) (XC3100A)	8	T_{TSON}		10.0		9.0		8.5		6.5		5.0	ns
same (slew -rate limited)	8	T_{TSON}		17.0		15.0		14.2		11.5		8.6	ns
Set-up and Hold Times (Output)													
Output (O) to clock (OK) set-up time													
(XC3100A)	5	T_{OOK}	4.5				3.6		3.2		2.9		ns
Output (O) to clock (OK) hold time	6	T_{OKO}	0				0		0				ns
Clock													
Clock High time	11	T_{IOH}	2.0		1.6		1.3		1.3		1.3		ns
Clock Low time	12	T_{IOL}	2.0		1.6		1.3		1.3		1.3		ns
Max. flip-flop toggle rate		F_{CLK}	227		270		323		323		370		MHz
Global Reset Delays													
RESET Pad to Registered In (Q)													
(XC3142A)	13	T_{RRI}		15.0		13.0		13.0		13.0		14.4	ns
(XC3190A)				25.5		21.0		21.0		21.0		21.0	ns
RESET Pad to output pad (fast)	15	T_{RPO}		20.0		17.0		17.0		17.0		17.0	ns
(slew-rate limited)	15	T_{RPO}		27.0		23.0		23.0		22.0		21.0	ns

Preliminary

Notes: 1. Timing is measured at pin threshold, with 50 pF external capacitive loads (incl. test fixture). For larger capacitive loads, see XAPP024. Typical slew rate limited output rise/fall times are approximately four times longer.

2. Voltage levels of unused (bonded and unbonded) pads must be valid logic levels. Each can be configured with the internal pull-up resistor or alternatively configured as a driven output or driven from an external source.

3. Input pad set-up time is specified with respect to the internal clock (ik). In order to calculate system set-up time, subtract clock delay (pad to ik) from the input pad set-up time value. Input pad holdtime with respect to the internal clock (ik) is negative. This means that pad level changes immediately before the internal clock edge (ik) will not be recognized.

4. T_{PID}, T_{PTG}, and T_{PICK} are 3 ns higher for XTL2 when the pin is configured as a user input.

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XC3000 Series 68-Pin PLCC, 84-Pin PLCC and PGA Pinouts

XC3000A, XC3000L, XC3100A, and XC3100L families have identical pinouts

68 PLCC		XC3020A, XC3030A, XC3042A	84 PLCC	68 PLCC		XC3020A, XC3030A, XC3042A	84 PLCC
XC3030A	XC3020A			XC3030A	XC3020A		
10	10	PWRDN	12	44	44	RESET	54
11	11	TCLKIN-I/O	13	45	45	DONE-PG	55
12	—	I/O*	14	46	46	D7-I/O	56
13	12	I/O	15	47	47	XTL1(OUT)-BCLKIN-I/O	57
14	13	I/O	16	48	48	D6-I/O	58
—	—	I/O	17	—	—	I/O	59
15	14	I/O	18	49	49	D5-I/O	60
16	15	I/O	19	50	50	CS0-I/O	61
—	16	I/O	20	51	51	D4-I/O	62
17	17	I/O	21	—	—	I/O	63
18	18	VCC	22	52	52	VCC	64
19	19	I/O	23	53	53	D3-I/O	65
—	—	I/O	24	54	54	CS1-I/O	66
20	20	I/O	25	55	55	D2-I/O	67
—	21	I/O	26	—	—	I/O	68
21	22	I/O	27	—	—	I/O*	69
22	—	I/O	28	56	56	D1-I/O	70
23	23	I/O	29	57	57	RDY/BUSY-RCLK-I/O	71
24	24	I/O	30	58	58	D0-DIN-I/O	72
25	25	M1-RDATA	31	59	59	DOUT-I/O	73
26	26	M0-RTRIG	32	60	60	CCLK	74
27	27	M2-I/O	33	61	61	A0-WS-I/O	75
28	28	HDC-I/O	34	62	62	A1-CS2-I/O	76
29	29	I/O	35	63	63	A2-I/O	77
30	30	LDC-I/O	36	64	64	A3-I/O	78
—	31	I/O	37	—	—	I/O*	79
—		I/O*	38	—	—	I/O*	80
31	32	I/O	39	65	65	A15-I/O	81
32	33	I/O	40	66	66	A4-I/O	82
33	—	I/O*	41	67	67	A14-I/O	83
34	34	INIT-I/O	42	68	68	A5-I/O	84
35	35	GND	43	1	1	GND	1
36	36	I/O	44	2	2	A13-I/O	2
37	37	I/O	45	3	3	A6-I/O	3
38	38	I/O	46	4	4	A12-I/O	4
39	39	I/O	47	5	5	A7-I/O	5
—	40	I/O	48	—	—	I/O*	6
—	41	I/O	49	—	—	I/O*	7
40		I/O*	50	6	6	A11-I/O	8
41		I/O*	51	7	7	A8-I/O	9
42	42	I/O	52	8	8	A10-I/O	10
43	43	XTL2(IN)-I/O	53	9	9	A9-I/O	11

Unprogrammed IOBs have a default pull-up. This prevents an undefined pad level for unbonded or unused IOBs. Programmed outputs are default slew-rate limited.

This table describes the pinouts of three different chips in three different packages. The pin-description column lists 84 of the 118 pads on the XC3042A (and 84 of the 98 pads on the XC3030A) that are connected to the 84 package pins. Ten pads, indicated by an asterisk, do not exist on the XC3020A, which has 74 pads; therefore the corresponding pins on the 84-pin packages have no connections to an XC3020A. Six pads on the XC3020A and 16 pads on the XC3030A, indicated by a dash (—) in the 68 PLCC column, have no connection to the 68 PLCC, but are connected to the 84-pin packages.

XC3000 Series 144-Pin Plastic TQFP Pinouts

XC3000A, XC3000L, XC3100A, and XC3100L families have identical pinouts

Pin Number	XC3042A XC3064A XC3090A	Pin Number	XC3042A XC3064A XC3090A	Pin Number	XC3042A XC3064A XC3090A
1	$\overline{PWRDN}$	49	I/O	97	I/O
2	I/O-TCLKIN	50	I/O*	98	I/O
3	I/O*	51	I/O	99	I/O*
4	I/O	52	I/O	100	I/O
5	I/O	53	$\overline{INIT}$-I/O	101	I/O*
6	I/O*	54	VCC	102	D1-I/O
7	I/O	55	GND	103	RDY/$\overline{BUSY}$-$\overline{RCLK}$-I/O
8	I/O	56	I/O	104	I/O
9	I/O*	57	I/O	105	I/O
10	I/O	58	I/O	106	D0-DIN-I/O
11	I/O	59	I/O	107	DOUT-I/O
12	I/O	60	I/O	108	CCLK
13	I/O	61	I/O	109	VCC
14	I/O	62	I/O	110	GND
15	I/O*	63	I/O*	111	A0-$\overline{WS}$-I/O
16	I/O	64	I/O*	112	A1-CS2-I/O
17	I/O	65	I/O	113	I/O
18	GND	66	I/O	114	I/O
19	VCC	67	I/O	115	A2-I/O
20	I/O	68	I/O	116	A3-I/O
21	I/O	69	XTL2(IN)-I/O	117	I/O
22	I/O	70	GND	118	I/O
23	I/O	71	$\overline{RESET}$	119	A15-I/O
24	I/O	72	VCC	120	A4-I/O
25	I/O	73	DONE-$\overline{PG}$	121	I/O*
26	I/O	74	D7-I/O	122	I/O*
27	I/O	75	XTL1(OUT)-BCLKIN-I/O	123	A14-I/O
28	I/O*	76	I/O	124	A5-I/O
29	I/O	77	I/O	125	I/O (XC3090 only)
30	I/O	78	D6-I/O	126	GND
31	I/O*	79	I/O	127	VCC
32	I/O*	80	I/O*	128	A13-I/O
33	I/O	81	I/O	129	A6-I/O
34	I/O*	82	I/O	130	I/O*
35	I/O	83	I/O*	131	I/O (XC3090 only)
36	M1-$\overline{RD}$	84	D5-I/O	132	I/O*
37	GND	85	$\overline{CS0}$-I/O	133	A12-I/O
38	M0-RT	86	I/O*	134	A7-I/O
39	VCC	87	I/O*	135	I/O
40	M2-I/O	88	D4-I/O	136	I/O
41	HDC-I/O	89	I/O	137	A11-I/O
42	I/O	90	VCC	138	A8-I/O
43	I/O	91	GND	139	I/O
44	I/O	92	D3-I/O	140	I/O
45	$\overline{LDC}$-I/O	93	$\overline{CS1}$-I/O	141	A10-I/O
46	I/O*	94	I/O*	142	A9-I/O
47	I/O	95	I/O*	143	VCC
48	I/O	96	D2-I/O	144	GND

Unprogrammed IOBs have a default pull-up. This prevents an undefined pad level for unbonded or unused IOBs. Programmed outputs are default slew-rate limited.

* Indicates unconnected package pins (24) for the XC3042A.

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XC3000 Series 176-Pin TQFP Pinouts

XC3000A, XC3000L, XC3100A, and XC3100L families have identical pinouts

Pin Number	XC3090A	Pin Number	XC3090A	Pin Number	XC3090A	Pin Number	XC3090A
1	PWRDWN	45	M1-RDATA	89	DONE-PG	133	VCC
2	TCLKIN-I/O	46	GND	90	D7-I/O	134	GND
3	I/O	47	M0-RTRIG	91	XTAL1(OUT)-BCLKIN-I/O	135	A0-WS-I/O
4	I/O	48	VCC	92	I/O	136	A1-CS2-I/O
5	I/O	49	M2-I/O	93	I/O	137	–
6	I/O	50	HDC-I/O	94	I/O	138	I/O
7	I/O	51	I/O	95	I/O	139	I/O
8	I/O	52	I/O	96	D6-I/O	140	A2-I/O
9	I/O	53	I/O	97	I/O	141	A3-I/O
10	I/O	54	LDC-I/O	98	I/O	142	–
11	I/O	55	–	99	I/O	143	–
12	I/O	56	I/O	100	I/O	144	I/O
13	I/O	57	I/O	101	I/O	145	I/O
14	I/O	58	I/O	102	D5-I/O	146	A15-I/O
15	I/O	59	I/O	103	CS0-I/O	147	A4-I/O
16	I/O	60	I/O	104	I/O	148	I/O
17	I/O	61	I/O	105	I/O	149	I/O
18	I/O	62	I/O	106	I/O	150	A14-I/O
19	I/O	63	I/O	107	I/O	151	A5-I/O
20	I/O	64	I/O	108	D4-I/O	152	I/O
21	I/O	65	INIT-I/O	109	I/O	153	I/O
22	GND	66	VCC	110	VCC	154	GND
23	VCC	67	GND	111	GND	155	VCC
24	I/O	68	I/O	112	D3-I/O	156	A13-I/O
25	I/O	69	I/O	113	CS1-I/O	157	A6-I/O
26	I/O	70	I/O	114	I/O	158	I/O
27	I/O	71	I/O	115	I/O	159	I/O
28	I/O	72	I/O	116	I/O	160	–
29	I/O	73	I/O	117	I/O	161	–
30	I/O	74	I/O	118	D2-I/O	162	I/O
31	I/O	75	I/O	119	I/O	163	I/O
32	I/O	76	I/O	120	I/O	164	A12-I/O
33	I/O	77	I/O	121	I/O	165	A7-I/O
34	I/O	78	I/O	122	I/O	166	I/O
35	I/O	79	I/O	123	I/O	167	I/O
36	I/O	80	I/O	124	D1-I/O	168	–
37	I/O	81	I/O	125	RDY/BUSY-RCLK-I/O	169	A11-I/O
38	I/O	82	–	126	I/O	170	A8-I/O
39	I/O	83	–	127	I/O	171	I/O
40	I/O	84	I/O	128	I/O	172	I/O
41	I/O	85	XTAL2(IN)-I/O	129	I/O	173	A10-I/O
42	I/O	86	GND	130	D0-DIN-I/O	174	A9-I/O
43	I/O	87	RESET	131	DOUT-I/O	175	VCC
44	–	88	VCC	132	CCLK	176	GND

Unprogrammed IOBs have a default pull-up. This prevents an undefined pad level for unbonded or unused IOBs. Programmed outputs are default slew-rate limited.