



Welcome to [E-XFL.COM](https://www.e-xfl.com)

Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	64
Number of Logic Elements/Cells	-
Total RAM Bits	14779
Number of I/O	58
Number of Gates	1500
Voltage - Supply	4.25V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	68-LCC (J-Lead)
Supplier Device Package	68-PLCC (24.23x24.23)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xc3120a-3pc68c

Flexible routing allows use of common or individual CLB clocking.

The combinatorial-logic portion of the CLB uses a 32 by 1 look-up table to implement Boolean functions. Variables selected from the five logic inputs and two internal block flip-flops are used as table address inputs. The combinatorial propagation delay through the network is independent of the logic function generated and is spike free for single input variable changes. This technique can generate two independent logic functions of up to four variables each as shown in Figure 6a, or a single function of five variables as shown in Figure 6b, or some functions of seven variables as shown in Figure 6c. Figure 7 shows a modulo-8 binary counter with parallel enable. It uses one CLB of each type. The partial functions of six or seven variables are implemented using the input variable (E) to dynamically select between two functions of four different variables. For the two functions of four variables each, the independent results (F and G) may be used as data inputs to either flip-flop or either logic block output. For the single function of five variables and merged functions of six or seven variables, the F and G outputs are identical. Symmetry of the F and G functions and the flip-flops allows the interchange of CLB outputs to optimize routing efficiencies of the networks interconnecting the CLBs and IOBs.

Programmable Interconnect

Programmable-interconnection resources in the Field Programmable Gate Array provide routing paths to connect inputs and outputs of the IOBs and CLBs into logic networks. Interconnections between blocks are composed of a two-layer grid of metal segments. Specially designed pass transistors, each controlled by a configuration bit, form programmable interconnect points (PIPs) and switching matrices used to implement the necessary connections between selected metal segments and block pins. Figure 8 is an example of a routed net. The development system provides automatic routing of these interconnections. Interactive routing is also available for design optimization. The inputs of the CLBs or IOBs are multiplexers which can be programmed to select an input network from the adjacent interconnect segments. **Since the switch connections to block inputs are unidirectional, as are block outputs, they are usable only for block input connection and not for routing.** Figure 9 illustrates routing access to logic block input variables, control inputs and block outputs. Three types of metal resources are provided to accommodate various network interconnect requirements.

- General Purpose Interconnect
- Direct Connection
- Longlines (multiplexed busses and wide AND gates)

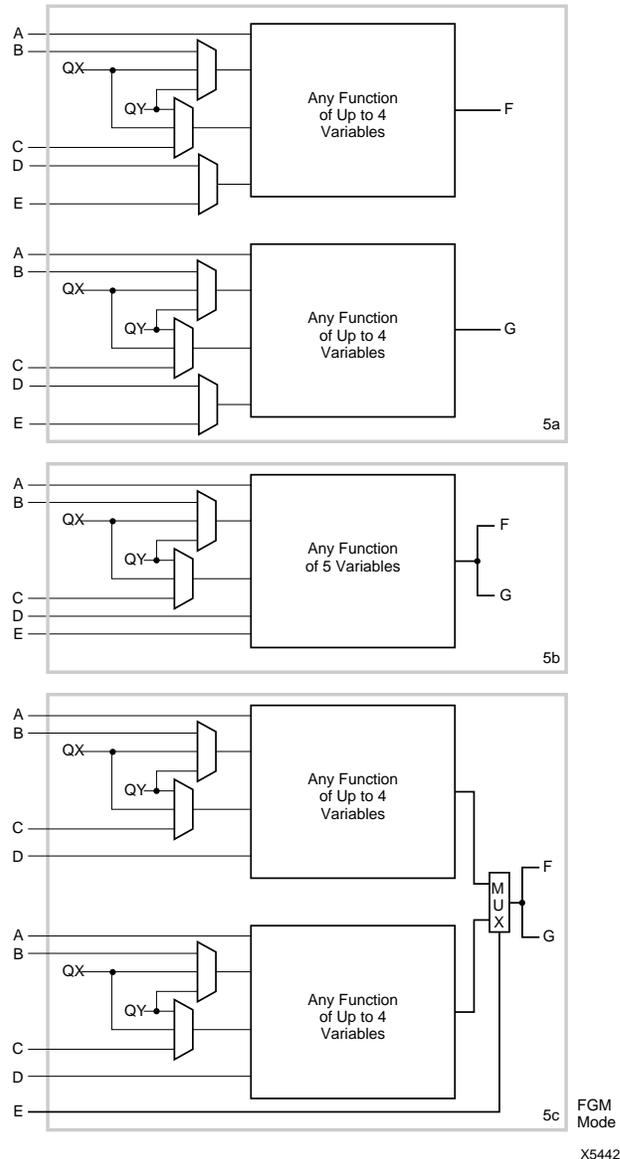


Figure 6: Combinational Logic Options

6a. Combinational Logic Option FG generates two functions of four variables each. One variable, A, must be common to both functions. The second and third variable can be any choice of B, C, QX and QY. The fourth variable can be any choice of D or E.

6b. Combinational Logic Option F generates any function of five variables: A, D, E and two choices out of B, C, QX, QY.

6c. Combinational Logic Option FGM allows variable E to select between two functions of four variables: Both have common inputs A and D and any choice out of B, C, QX and QY for the remaining two variables. Option 3 can then implement some functions of six or seven variables.

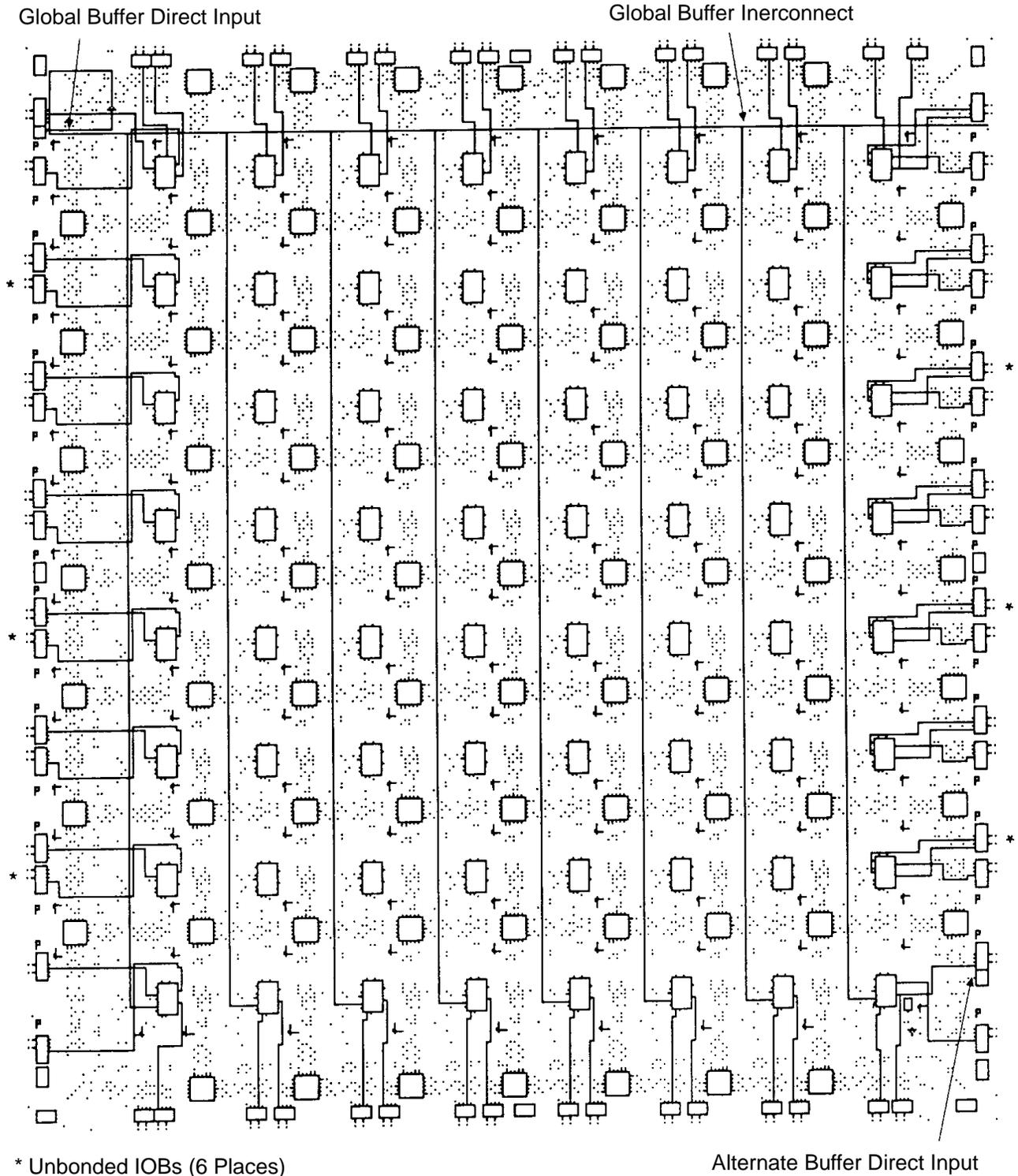


Figure 13: XC3020A Die-Edge IOBs. The XC3020A die-edge IOBs are provided with direct access to adjacent CLBs.

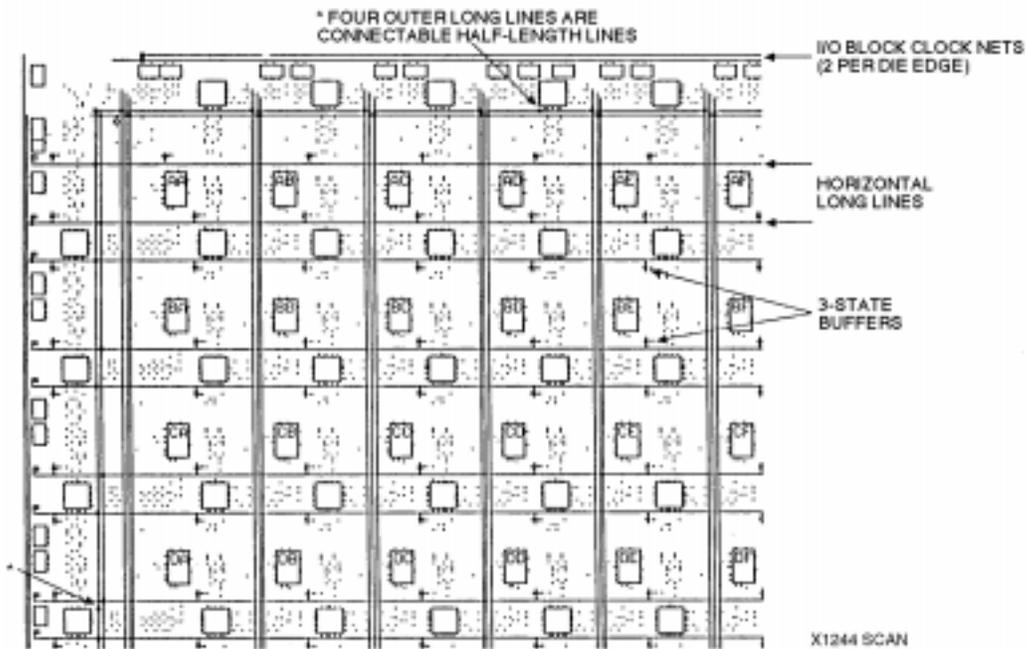


Figure 15: Programmable Interconnection of Longlines. This is provided at the edges of the routing area. Three-state buffers allow the use of horizontal Longlines to form on-chip wired AND and multiplexed buses. The left two non-clock vertical Longlines per column (except XC3020A) and the outer perimeter Longlines may be programmed as connectable half-length lines.

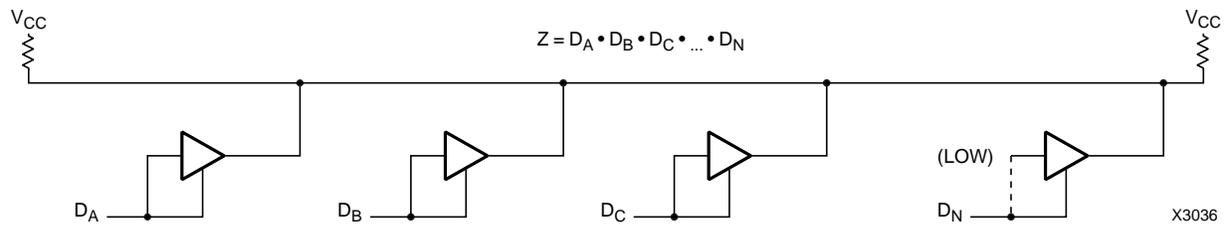


Figure 16: 3-State Buffers Implement a Wired-AND Function. When all the buffer 3-state lines are High, (high impedance), the pull-up resistor(s) provide the High output. The buffer inputs are driven by the control signals or a Low.

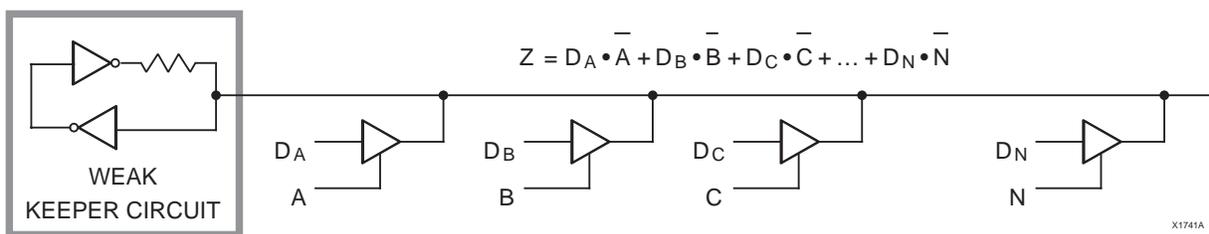


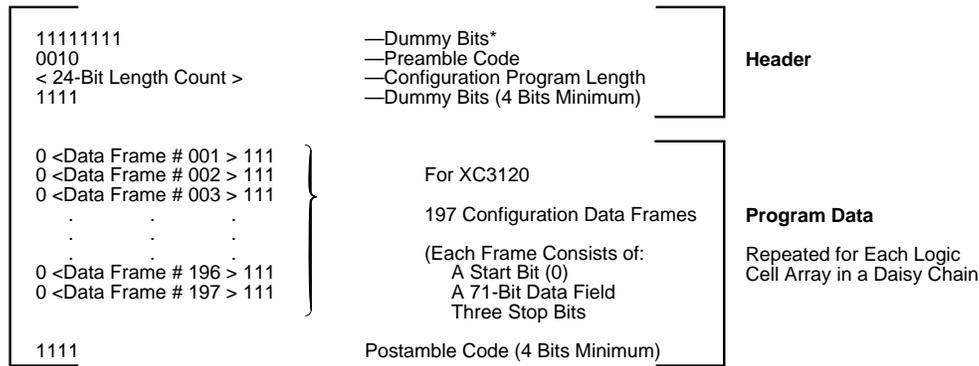
Figure 17: 3-State Buffers Implement a Multiplexer. The selection is accomplished by the buffer 3-state signal.

A re-program is initiated when a configured XC3000 series device senses a High-to-Low transition and subsequent >6 μ s Low level on the DONE/ $\overline{\text{PROG}}$ package pin, or, if this pin is externally held permanently Low, a High-to-Low transition and subsequent >6 μ s Low time on the $\overline{\text{RESET}}$ package pin.

The device returns to the Clear state where the configuration memory is cleared and mode lines re-sampled, as for an aborted configuration. The complete configuration program is cleared and loaded during each configuration program cycle.

Length count control allows a system of multiple Field Programmable Gate Arrays, of assorted sizes, to begin operation in a synchronized fashion. The configuration program

generated by the development system begins with a preamble of 11111110010 followed by a 24-bit length count representing the total number of configuration clocks needed to complete loading of the configuration program(s). The data framing is shown in Figure 21. All FPGAs connected in series read and shift preamble and length count in on positive and out on negative configuration clock edges. A device which has received the preamble and length count then presents a High Data Out until it has intercepted the appropriate number of data frames. When the configuration program memory of an FPGA is full and the length count does not yet compare, the device shifts any additional data through, as it did for preamble and length count. When the FPGA configuration memory is full and the length count compares, the device will execute



*The LCA Device Require Four Dummy Bits Min; Software Generates Eight Dummy Bits

X5300_01

Device	XC3020A XC3020L XC3120A	XC3030A XC3030L XC3130A	XC3042A XC3042L XC3142A XC3142L	XC3064A XC3064L XC3164A	XC3090A XC3090L XC3190A XC3190L	XC3195A
Gates	1,000 to 1,500	1,500 to 2,000	2,000 to 3,000	3,500 to 4,500	5,000 to 6,000	6,500 to 7,500
CLBs	64	100	144	224	320	484
Row x Col	(8 x 8)	(10 x 10)	(12 x 12)	(16 x 14)	(20 x 16)	(22 x 22)
I/Os	64	80	96	120	144	176
Flip-flops	256	360	480	688	928	1,320
Horizontal Longlines	16	20	24	32	40	44
TBUFs/Horizontal LL	9	11	13	15	17	23
Bits per Frame (including 1 start and 3 stop bits)	75	92	108	140	172	188
Frames	197	241	285	329	373	505
Program Data = Bits x Frames + 4 bits (excludes header)	14,779	22,176	30,784	46,064	64,160	94,944
PROM size (bits) = Program Data + 40-bit Header	14,819	22,216	30,824	46,104	64,200	94,984

Figure 21: Internal Configuration Data Structure for an FPGA. This shows the preamble, length count and data frames generated by the Development System.

The Length Count produced by the program = [(40-bit preamble + sum of program data + 1 per daisy chain device) rounded up to multiple of 8] – (2 ≤ K ≤ 4) where K is a function of DONE and RESET timing selected. An additional 8 is added if roundup increment is less than K. K additional clocks are needed to complete start-up after length count is reached.

Special Configuration Functions

The configuration data includes control over several special functions in addition to the normal user logic functions and interconnect.

- Input thresholds
- Readback disable
- DONE pull-up resistor
- DONE timing
- RESET timing
- Oscillator frequency divided by two

Each of these functions is controlled by configuration data bits which are selected as part of the normal development system bitstream generation process.

Input Thresholds

Prior to the completion of configuration all FPGA input thresholds are TTL compatible. Upon completion of configuration, the input thresholds become either TTL or CMOS compatible as programmed. The use of the TTL threshold option requires some additional supply current for threshold shifting. The exception is the threshold of the PWRDWN input and direct clocks which always have a CMOS input. Prior to the completion of configuration the user I/O pins each have a high impedance pull-up. The configuration program can be used to enable the IOB pull-up resistors in the Operational mode to act either as an input load or to avoid a floating input on an otherwise unused pin.

Readback

The contents of a Field Programmable Gate Array may be read back if it has been programmed with a bitstream in which the Readback option has been enabled. Readback may be used for verification of configuration and as a method of determining the state of internal logic nodes during debugging. There are three options in generating the configuration bitstream.

- “Never” inhibits the Readback capability.
- “One-time,” inhibits Readback after one Readback has been executed to verify the configuration.
- “On-command” allows unrestricted use of Readback.

Readback is accomplished without the use of any of the user I/O pins; only M0, M1 and CCLK are used. The initiation of Readback is produced by a Low to High transition of the M0/RTRIG (Read Trigger) pin. The CCLK input must then be driven by external logic to read back the configuration data. The first three Low-to-High CCLK transitions clock out dummy data. The subsequent Low-to-High CCLK transitions shift the data frame information out on the M1/RDATA (Read Data) pin. Note that the logic polarity is always inverted, a zero in configuration becomes a one in Readback, and vice versa. Note also that each Readback frame has one Start bit (read back as a one) but, unlike in

configuration, each Readback frame has only one Stop bit (read back as a zero). The third leading dummy bit mentioned above can be considered the Start bit of the first frame. All data frames must be read back to complete the process and return the Mode Select and CCLK pins to their normal functions.

Readback data includes the current state of each CLB flip-flop, each input flip-flop or latch, and each device pad. These data are imbedded into unused configuration bit positions during Readback. This state information is used by the development system In-Circuit Verifier to provide visibility into the internal operation of the logic while the system is operating. To readback a uniform time-sample of all storage elements, it may be necessary to inhibit the system clock.

Reprogram

To initiate a re-programming cycle, the dual-function pin DONE/PROG must be given a High-to-Low transition. To reduce sensitivity to noise, the input signal is filtered for two cycles of the FPGA internal timing generator. When reprogram begins, the user-programmable I/O output buffers are disabled and high-impedance pull-ups are provided for the package pins. The device returns to the Clear state and clears the configuration memory before it indicates ‘initialized’. Since this Clear operation uses chip-individual internal timing, the master might complete the Clear operation and then start configuration before the slave has completed the Clear operation. To avoid this problem, the slave INIT pins must be AND-wired and used to force a RESET on the master (see Figure 25). Reprogram control is often implemented using an external open-collector driver which pulls DONE/PROG Low. Once a stable request is recognized, the DONE/PROG pin is held Low until the new configuration has been completed. Even if the re-program request is externally held Low beyond the configuration period, the FPGA will begin operation upon completion of configuration.

DONE Pull-up

DONE/PROG is an open-drain I/O pin that indicates the FPGA is in the operational state. An optional internal pull-up resistor can be enabled by the user of the development system. The DONE/PROG pins of multiple FPGAs in a daisy-chain may be connected together to indicate all are DONE or to direct them all to reprogram.

DONE Timing

The timing of the DONE status signal can be controlled by a selection to occur either a CCLK cycle before, or after, the outputs going active. See Figure 22. This facilitates control of external functions such as a PROM enable or holding a system in a wait state.

RESET Timing

As with DONE timing, the timing of the release of the internal reset can be controlled to occur either a CCLK cycle before, or after, the outputs going active. See [Figure 22](#). This reset keeps all user programmable flip-flops and latches in a zero state during configuration.

Crystal Oscillator Division

A selection allows the user to incorporate a dedicated divide-by-two flip-flop between the crystal oscillator and the alternate clock line. This guarantees a symmetrical clock signal. Although the frequency stability of a crystal oscillator is very good, the symmetry of its waveform can be affected by bias or feedback drive.

Bitstream Error Checking

Bitstream error checking protects against erroneous configuration.

Each Xilinx FPGA bitstream consists of a 40-bit preamble, followed by a device-specific number of data frames. The number of bits per frame is also device-specific; however, each frame ends with three stop bits (111) followed by a start bit for the next frame (0).

All devices in all XC3000 families start reading in a new frame when they find the first 0 after the end of the previous frame. An original XC3000 device does not check for the correct stop bits, but XC3000A, XC3100A, XC3000L, and XC3100L devices check that the last three bits of any frame are actually 111.

Under normal circumstances, all these FPGAs behave the same way; however, if the bitstream is corrupted, an XC3000 device will always start a new frame as soon as it finds the first 0 after the end of the previous frame, even if the data is completely wrong or out-of-sync. Given sufficient zeros in the data stream, the device will also go Done,

but with incorrect configuration and the possibility of internal contention.

An XC3000A/XC3100A/XC3000L/XC3100L device starts any new frame only if the three preceding bits are all ones. If this check fails, it pulls $\overline{\text{INIT}}$ Low and stops the internal configuration, although the Master CCLK keeps running. The user must then start a new configuration by applying a $>6 \mu\text{s}$ Low level on $\overline{\text{RESET}}$.

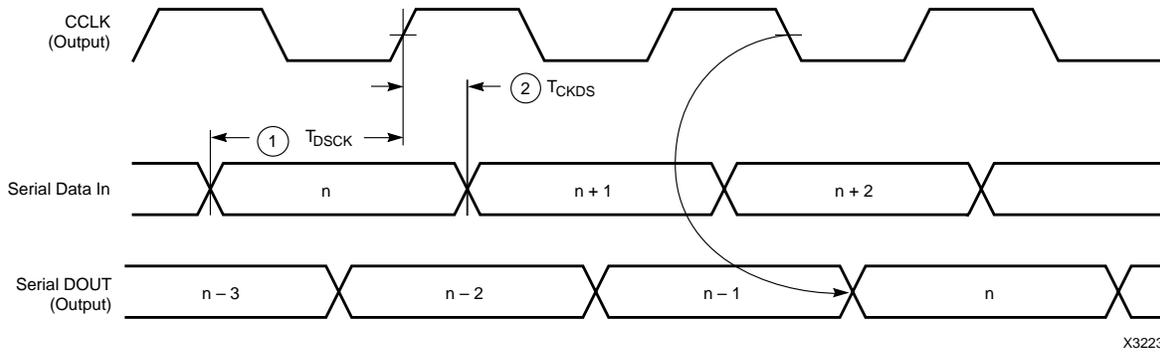
This simple check does not protect against random bit errors, but it offers almost 100 percent protection against erroneous configuration files, defective configuration data sources, synchronization errors between configuration source and FPGA, or PC-board level defects, such as broken lines or solder-bridges.

Reset Spike Protection

A separate modification slows down the $\overline{\text{RESET}}$ input before configuration by using a two-stage shift register driven from the internal clock. It tolerates submicrosecond High spikes on $\overline{\text{RESET}}$ before configuration. The XC3000 master can be connected like an XC4000 master, but with its $\overline{\text{RESET}}$ input used instead of $\overline{\text{INIT}}$. (On XC3000, $\overline{\text{INIT}}$ is output only).

Soft Start-up

After configuration, the outputs of all FPGAs in a daisy-chain become active simultaneously, as a result of the same CCLK edge. In the original XC3000/3100 devices, each output becomes active in either fast or slew-rate limited mode, depending on the way it is configured. This can lead to large ground-bounce signals. In XC3000A, XC3000L, XC3100A, and XC3100L devices, all outputs become active first in slew-rate limited mode, reducing the ground bounce. After this soft start-up, each individual output slew rate is again controlled by the respective configuration bit.

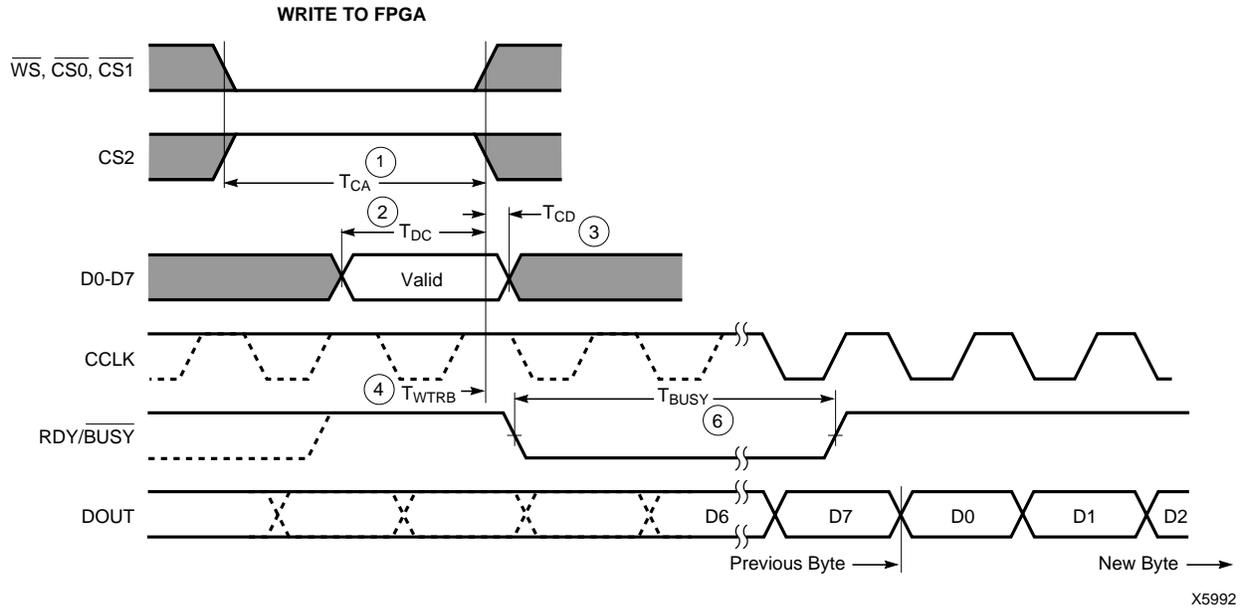


X3223

	Description		Symbol	Min	Max	Units
CCLK	Data In setup	1	T_{DSCK}	60		ns
	Data In hold	2	T_{CKDS}	0		ns

- Notes:
1. At power-up, V_{CC} must rise from 2.0 V to V_{CC} min in less than 25 ms. If this is not possible, configuration can be delayed by holding RESET Low until V_{CC} has reached 4.0 V (2.5 V for the XC3000L). A very long V_{CC} rise time of >100 ms, or a non-monotonically rising V_{CC} may require >6- μ s High level on RESET, followed by a >6- μ s Low level on RESET and D/\bar{P} after V_{CC} has reached 4.0 V (2.5 V for the XC3000L).
 2. Configuration can be controlled by holding RESET Low with or until after the \overline{INIT} of all daisy-chain slave-mode devices is High.
 3. Master-serial-mode timing is based on slave-mode testing.

Figure 24: Master Serial Mode Programming Switching Characteristics



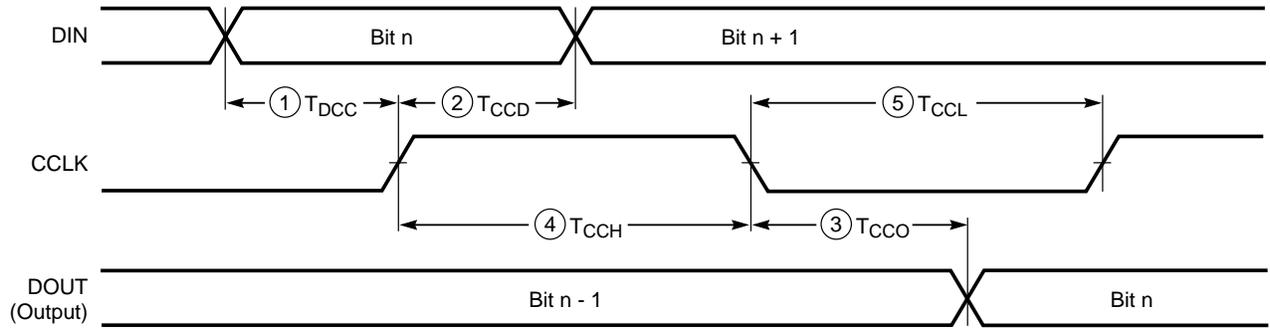
X5992

	Description	Symbol	Min	Max	Units
WRITE	Effective Write time required (Assertion of CS0, CS1, CS2, WS)	1 T_{CA}	100		ns
	DIN Setup time required	2 T_{DC}	60		ns
	DIN Hold time required	3 T_{CD}	0		ns
	RDY/BUSY delay after end of WS	4 T_{WTRB}		60	ns
RDY	Earliest next WS after end of BUSY	5 T_{RBWT}	0		ns
	BUSY Low time generated	6 T_{BUSY}	2.5	9	CCLK periods

- Notes:
- At power-up, V_{CC} must rise from 2.0 V to V_{CC} min in less than 25 ms. If this is not possible, configuration can be delayed by holding \overline{RESET} Low until V_{CC} has reached 4.0 V (2.5 V for the XC3000L). A very long V_{CC} rise time of >100 ms, or a non-monotonically rising V_{CC} may require a >6- μ s High level on \overline{RESET} , followed by a >6- μ s Low level on \overline{RESET} and D/\overline{P} after V_{CC} has reached 4.0 V (2.5 V for the XC3000L).
 - Configuration must be delayed until the INIT of all FPGAs is High.
 - Time from end of WS to CCLK cycle for the new byte of data depends on completion of previous byte processing and the phase of the internal timing generator for CCLK.
 - CCLK and DOUT timing is tested in slave mode.
 - T_{BUSY} indicates that the double-buffered parallel-to-serial converter is not yet ready to receive new data. The shortest T_{BUSY} occurs when a byte is loaded into an empty parallel-to-serial converter. The longest T_{BUSY} occurs when a new word is loaded into the input register before the second-level buffer has started shifting out data.

Note: This timing diagram shows very relaxed requirements: Data need not be held beyond the rising edge of \overline{WS} . \overline{BUSY} will go active within 60 ns after the end of \overline{WS} . \overline{BUSY} will stay active for several microseconds. \overline{WS} may be asserted immediately after the end of \overline{BUSY} .

Figure 28: Peripheral Mode Programming Switching Characteristics



X5379

	Description	Symbol	Min	Max	Units
CCLK	To DOUT	3 T_{CCO}		100	ns
	DIN setup	1 T_{DCC}	60		ns
	DIN hold	2 T_{CCD}	0		ns
	High time	4 T_{CCH}	0.05		μ s
	Low time (Note 1)	5 T_{CCL}	0.05	5.0	μ s
	Frequency	F_{CC}		10	MHz

- Notes:
1. The max limit of CCLK Low time is caused by dynamic circuitry inside the FPGA.
 2. Configuration must be delayed until the INIT of all FPGAs is High.
 3. At power-up, V_{CC} must rise from 2.0 V to V_{CC} min in less than 25 ms. If this is not possible, configuration can be delayed by holding RESET Low until VCC has reached 4.0 V (2.5 V for the XC3000L). A very long V_{CC} rise time of >100 ms, or a non-monotonically rising V_{CC} may require a >6- μ s High level on RESET, followed by a >6- μ s Low level on RESET and D/P after V_{CC} has reached 4.0 V (2.5 V for the XC3000L).

Figure 30: Slave Serial Mode Programming Switching Characteristics

Device Performance

The XC3000 families of FPGAs can achieve very high performance. This is the result of

- A sub-micron manufacturing process, developed and continuously being enhanced for the production of state-of-the-art CMOS SRAMs.
- Careful optimization of transistor geometries, circuit design, and lay-out, based on years of experience with the XC3000 family.
- A look-up table based, coarse-grained architecture that can collapse multiple-layer combinatorial logic into a single function generator. One CLB can implement up to four layers of conventional logic in as little as 1.5 ns.

Actual system performance is determined by the timing of critical paths, including the delay through the combinatorial and sequential logic elements within CLBs and IOBs, plus the delay in the interconnect routing. The AC-timing specifications state the worst-case timing parameters for the various logic resources available in the XC3000-families architecture. **Figure 31** shows a variety of elements involved in determining system performance.

Logic block performance is expressed as the propagation time from the interconnect point at the input to the block to the output of the block in the interconnect area. Since combinatorial logic is implemented with a memory lookup table within a CLB, the combinatorial delay through the CLB, called T_{ILO} , is always the same, regardless of the function being implemented. For the combinatorial logic function driving the data input of the storage element, the critical timing is data set-up relative to the clock edge provided to the flip-flop element. The delay from the clock source to the output of the logic block is critical in the timing signals pro-

duced by storage elements. Loading of a logic-block output is limited only by the resulting propagation delay of the larger interconnect network. Speed performance of the logic block is a function of supply voltage and temperature. See **Figure 32**.

Interconnect performance depends on the routing resources used to implement the signal path. Direct interconnects to the neighboring CLB provide an extremely fast path. Local interconnects go through switch matrices (magic boxes) and suffer an RC delay, equal to the resistance of the pass transistor multiplied by the capacitance of the driven metal line. Longlines carry the signal across the length or breadth of the chip with only one access delay. Generous on-chip signal buffering makes performance relatively insensitive to signal fan-out; increasing fan-out from 1 to 8 changes the CLB delay by only 10%. Clocks can be distributed with two low-skew clock distribution networks.

The tools in the Development System used to place and route a design in an XC3000 FPGA automatically calculate the actual maximum worst-case delays along each signal path. This timing information can be back-annotated to the design's netlist for use in timing simulation or examined with, a static timing analyzer.

Actual system performance is applications dependent. The maximum clock rate that can be used in a system is determined by the critical path delays within that system. These delays are combinations of incremental logic and routing delays, and vary from design to design. In a synchronous system, the maximum clock rate depends on the number of combinatorial logic layers between re-synchronizing flip-flops. **Figure 33** shows the achievable clock rate as a function of the number of CLB layers.

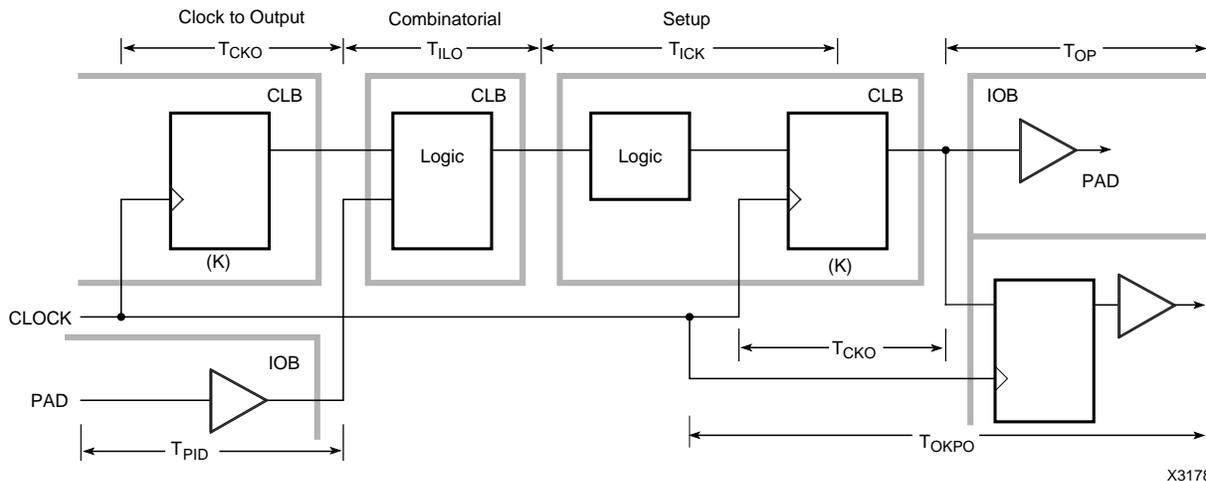


Figure 31: Primary Block Speed Factors. Actual timing is a function of various block factors combined with routing factors. Overall performance can be evaluated with the timing calculator or by an optional simulation.

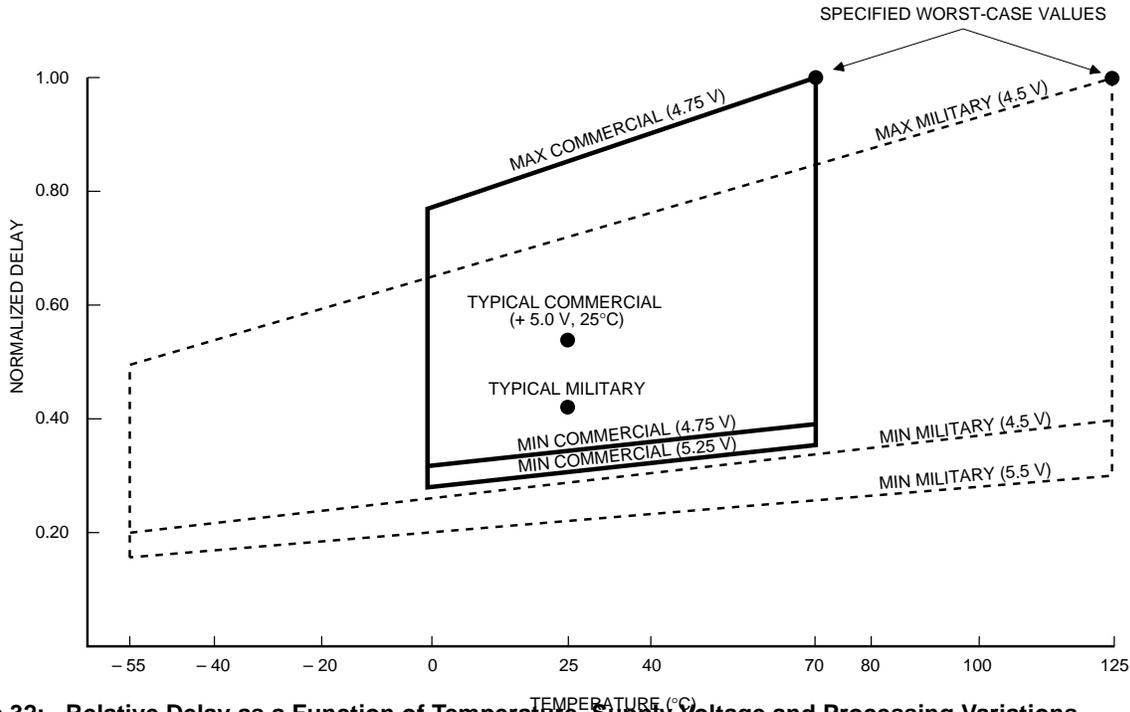


Figure 32: Relative Delay as a Function of Temperature, Supply Voltage and Processing Variations X6094

Power

Power Distribution

Power for the FPGA is distributed through a grid to achieve high noise immunity and isolation between logic and I/O. Inside the FPGA, a dedicated V_{CC} and ground ring surrounding the logic array provides power to the I/O drivers. An independent matrix of V_{CC} and groundlines supplies the interior logic of the device. This power distribution grid provides a stable supply and ground for all internal logic, providing the external package power pins are all connected and appropriately decoupled. Typically a 0.1- μ F capacitor connected near the V_{CC} and ground pins will provide adequate decoupling.

Output buffers capable of driving the specified 4- or 8-mA loads under worst-case conditions may be capable of driving as much as 25 to 30 times that current in a best case. Noise can be reduced by minimizing external load capacitance and reducing simultaneous output transitions in the same direction. It may also be beneficial to locate heavily loaded output buffers near the ground pads. The I/O Block output buffers have a slew-limited mode which should be used where output rise and fall times are not speed critical. Slew-limited outputs maintain their dc drive capability, but generate less external reflections and internal noise.

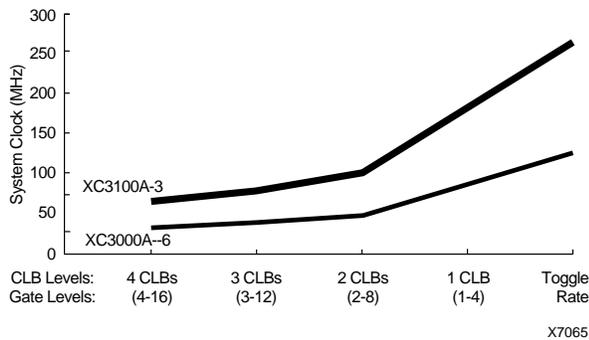


Figure 33: Clock Rate as a Function of Logic Complexity (Number of Combinational Levels between Flip-Flops)

Dynamic Power Consumption

	XC3042A	XC3042L	XC3142A	
One CLB driving three local interconnects	0.25	0.17	0.25	mW per MHz
One global clock buffer and clock line	2.25	1.40	1.70	mW per MHz
One device output with a 50 pF load	1.25	1.25	1.25	mW per MHz

Power Consumption

The Field Programmable Gate Array exhibits the low power consumption characteristic of CMOS ICs. For any design, the configuration option of TTL chip input threshold requires power for the threshold reference. The power required by the static memory cells that hold the configuration data is very low and may be maintained in a power-down mode.

Typically, most of power dissipation is produced by external capacitive loads on the output buffers. This load and frequency dependent power is 25 μ W/pF/MHz per output. Another component of I/O power is the external dc loading on all output pins.

Internal power dissipation is a function of the number and size of the nodes, and the frequency at which they change. In an FPGA, the fraction of nodes changing on a given clock is typically low (10-20%). For example, in a long binary counter, the total activity of all counter flip-flops is equivalent to that of only two CLB outputs toggling at the clock frequency. Typical global clock-buffer power is between 2.0 mW/MHz for the XC3020A and 3.5 mW/MHz for the XC3090A. The internal capacitive load is more a function of interconnect than fan-out. With a typical load of three general interconnect segments, each CLB output requires about 0.25 mW per MHz of its output frequency.

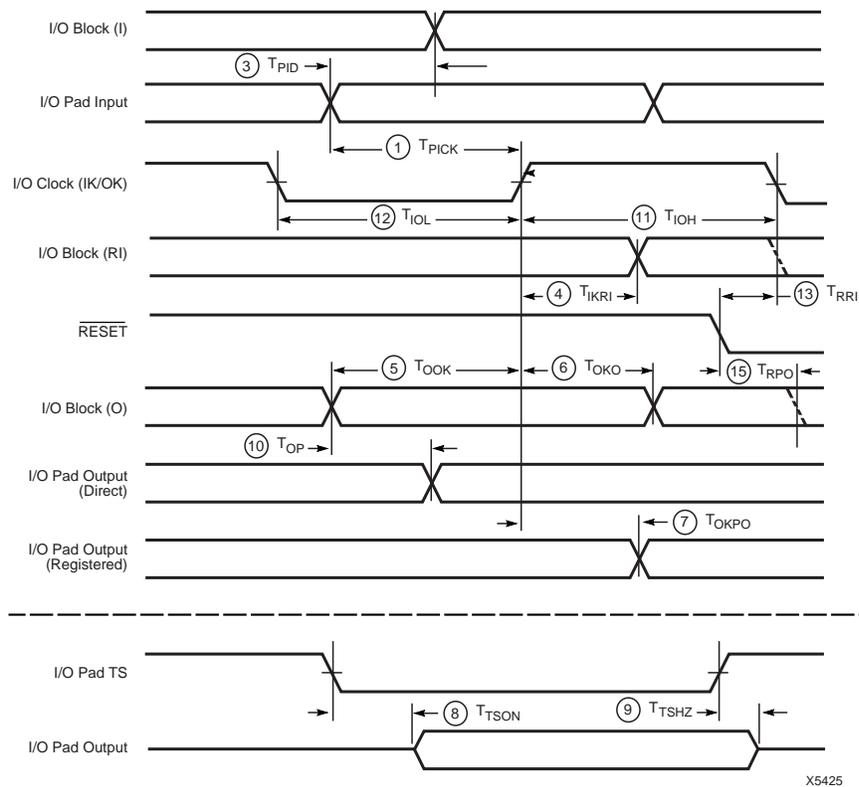
Because the control storage of the FPGA is CMOS static memory, its cells require a very low standby current for data retention. In some systems, this low data retention current characteristic can be used as a method of preserving configurations in the event of a primary power loss. The FPGA

has built in powerdown logic which, when activated, will disable normal operation of the device and retain only the configuration data. All internal operation is suspended and output buffers are placed in their high-impedance state with no pull-ups. Different from the XC3000 family which can be powered down to a current consumption of a few microamps, the XC3100A draws 5 mA, even in power-down. This makes power-down operation less meaningful. In contrast, I_{CCPD} for the XC3000L is only 10 μ A.

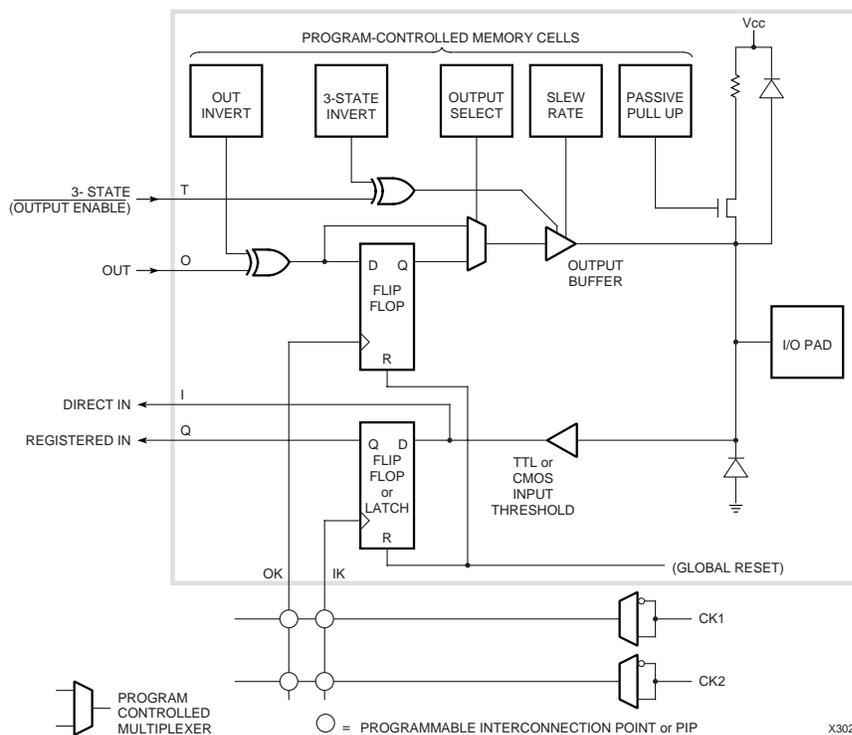
To force the FPGA into the Powerdown state, the user must pull the PWRDWN pin Low and continue to supply a retention voltage to the V_{CC} pins. When normal power is restored, V_{CC} is elevated to its normal operating voltage and PWRDWN is returned to a High. The FPGA resumes operation with the same internal sequence that occurs at the conclusion of configuration. Internal-I/O and logic-block storage elements will be reset, the outputs will become enabled and the DONE/ \overline{PROG} pin will be released.

When V_{CC} is shut down or disconnected, some power might unintentionally be supplied from an incoming signal driving an I/O pin. The conventional electrostatic input protection is implemented with diodes to the supply and ground. A positive voltage applied to an input (or output) will cause the positive protection diode to conduct and drive the V_{CC} connection. This condition can produce invalid power conditions and should be avoided. A large series resistor might be used to limit the current or a bipolar buffer may be used to isolate the input signal.

XC3000A IOB Switching Characteristics Guidelines (continued)



X5425



X3029

XC3000L Absolute Maximum Ratings

Symbol	Description		Units
V_{CC}	Supply voltage relative to GND	-0.5 to +7.0	V
V_{IN}	Input voltage with respect to GND	-0.5 to $V_{CC} + 0.5$	V
V_{TS}	Voltage applied to 3-state output	-0.5 to $V_{CC} + 0.5$	V
T_{STG}	Storage temperature (ambient)	-65 to +150	°C
T_{SOL}	Maximum soldering temperature (10 s @ 1/16 in.)	+260	°C
T_J	Junction temperature plastic	+125	°C
	Junction temperature ceramic	+150	°C

Note: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions is not implied. Exposure to Absolute Maximum Ratings conditions for extended periods of time may affect device reliability.

XC3000L Global Buffer Switching Characteristics Guidelines

Description	Speed Grade	-8	Units
	Symbol	Max	
Global and Alternate Clock Distribution ¹ Either: Normal IOB input pad through clock buffer to any CLB or IOB clock input Or: Fast (CMOS only) input pad through clock buffer to any CLB or IOB clock input	T_{PID}	9.0	ns
	T_{PIDC}	7.0	ns
TBUF driving a Horizontal Longline (L.L.) ¹ I to L.L. while T is Low (buffer active) T↓ to L.L. active and valid with single pull-up resistor T↑ to L.L. High with single pull-up resistor	T_{IO}	5.0	ns
	T_{ON}	12.0	ns
	T_{PUS}	24.0	ns
BIDI Bidirectional buffer delay	T_{BIDI}	2.0	ns

Notes: 1. Timing is based on the XC3042A, for other devices see timing calculator.
2. The use of two pull-up resistors per Longline, available on other XC3000 devices, is not a valid option for XC3000L devices.

Product Obsolete or Under Obsolescence



XC3000L IOB Switching Characteristics Guidelines

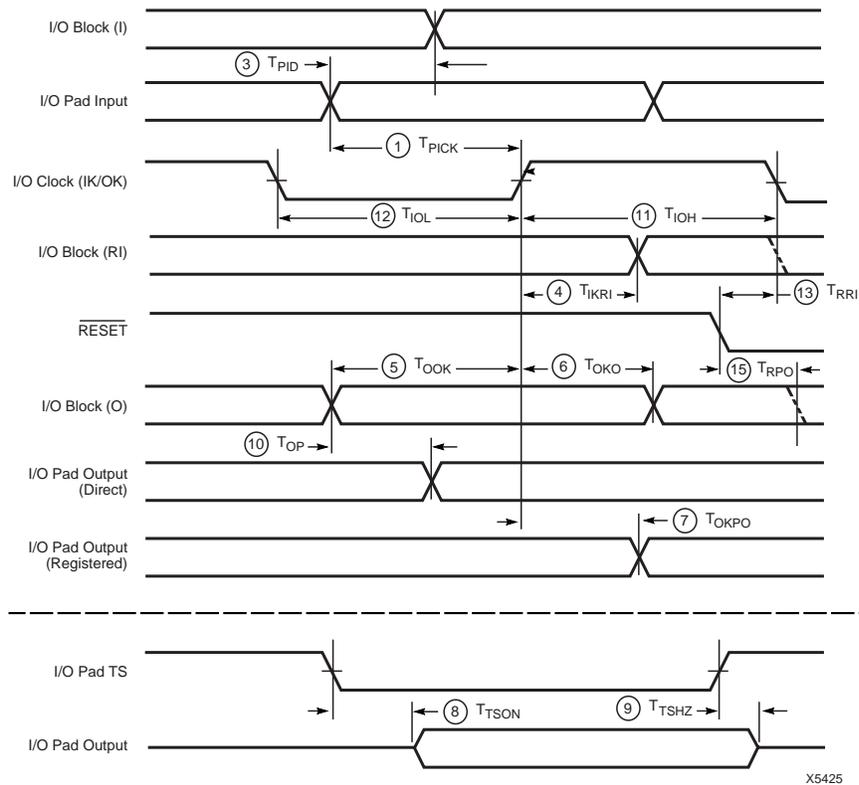
Testing of the switching parameters is modeled after testing methods specified by MIL-M-38510/605. All devices are 100% functionally tested. Since many internal timing parameters cannot be measured directly, they are derived from benchmark timing patterns. The following guidelines reflect worst-case values over the recommended operating conditions. For more detailed, more precise, and more up-to-date timing information, use the values provided by the timing calculator and used in the simulator.

Description	Speed Grade		-8		Units
	Symbol		Min	Max	
Propagation Delays (Input)					
Pad to Direct In (I)	3	T_{PID}		5.0	ns
Pad to Registered In (Q) with latch transparent		T_{PTG}		24.0	ns
Clock (IK) to Registered In (Q)	4	T_{IKRI}		6.0	ns
Set-up Time (Input)					
Pad to Clock (IK) set-up time	1	T_{PICK}	22.0		ns
Propagation Delays (Output)					
Clock (OK) to Pad (fast)	7	T_{OKPO}		12.0	ns
same (slew rate limited)	7	T_{OKPO}		28.0	ns
Output (O) to Pad (fast)	10	T_{OPF}		9.0	ns
same (slew-rate limited)	10	T_{OPS}		25.0	ns
3-state to Pad begin hi-Z (fast)	9	T_{TSHZ}		12.0	ns
same (slew-rate limited)	9	T_{TSHZ}		28.0	ns
3-state to Pad active and valid (fast)	8	T_{TSON}		16.0	ns
same (slew -rate limited)	8	T_{TSON}		32.0	ns
Set-up and Hold Times (Output)					
Output (O) to clock (OK) set-up time	5	T_{OOK}	12.0		ns
Output (O) to clock (OK) hold time	6	T_{OKO}	0		ns
Clock					
Clock High time	11	T_{IOH}	5.0		ns
Clock Low time	12	T_{IOL}	5.0		ns
Max. flip-flop toggle rate		F_{CLK}	80.0		MHz
Global Reset Delays (based on XC3042L)					
\overline{RESET} Pad to Registered In (Q)	13	T_{RRI}		25.0	ns
\overline{RESET} Pad to output pad (fast)	15	T_{RPO}		35.0	ns
\overline{RESET} Pad to output pad (slew-rate limited)	15	T_{RPO}		51.0	ns

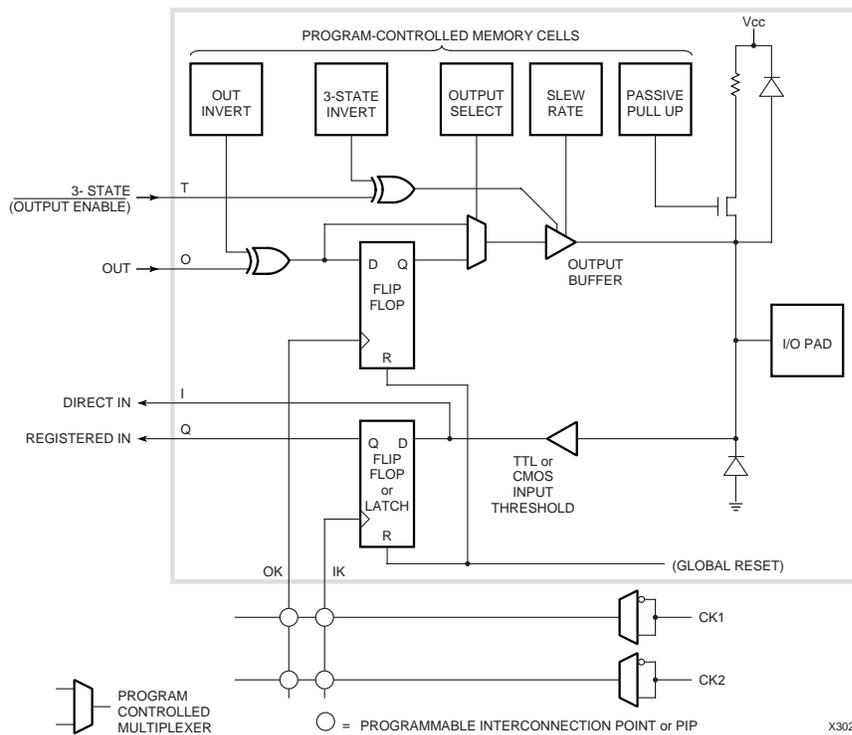
7

- Notes:**
1. Timing is measured at pin threshold, with 50 pF external capacitive loads (incl. test fixture). Typical slew rate limited output rise/fall times are approximately four times longer.
 2. Voltage levels of unused (bonded and unbonded) pads must be valid logic levels. Each can be configured with the internal pull-up resistor or alternatively configured as a driven output or driven from an external source.
 3. Input pad set-up time is specified with respect to the internal clock (ik). In order to calculate system set-up time, subtract clock delay (pad to ik) from the input pad set-up time value. Input pad holdtime with respect to the internal clock (ik) is negative. This means that pad level changes immediately before the internal clock edge (ik) will not be recognized.
 4. T_{PID} , T_{PTG} , and T_{PICK} are 3 ns higher for XTL2 when the pin is configured as a user input.

XC3000L IOB Switching Characteristics Guidelines (continued)



X5425



X3029

XC3100L Absolute Maximum Ratings

Symbol	Description		Units
V_{CC}	Supply voltage relative to GND	-0.5 to +7.0	V
V_{IN}	Input voltage with respect to GND	-0.5 to $V_{CC} + 0.5$	V
V_{TS}	Voltage applied to 3-state output	-0.5 to $V_{CC} + 0.5$	V
T_{STG}	Storage temperature (ambient)	-65 to +150	°C
T_{SOL}	Maximum soldering temperature (10 s @ 1/16 in.)	+260	°C
T_J	Junction temperature plastic	+125	°C
	Junction temperature ceramic	+150	°C

Note: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions is not implied. Exposure to Absolute Maximum Ratings conditions for extended periods of time may affect device reliability.

XC3100L Global Buffer Switching Characteristics Guidelines

Description	Symbol	Speed Grade		Units
		-3 Max	-2 Max	
Global and Alternate Clock Distribution ¹ Either: Normal IOB input pad through clock buffer to any CLB or IOB clock input Or: Fast (CMOS only) input pad through clock buffer to any CLB or IOB clock input	T_{PID}	5.6	4.7	ns
	T_{PIDC}	4.3	3.7	ns
TBUF driving a Horizontal Longline (L.L.) ¹ I to L.L. while T is Low (buffer active) $T \downarrow$ to L.L. active and valid with single pull-up resistor $T \uparrow$ to L.L. High with single pull-up resistor	T_{IO}	3.1	3.1	ns
	T_{ON}	4.2	4.2	ns
	T_{PUS}	11.4	11.4	ns
BIDI Bidirectional buffer delay	T_{BIDI}	1.0	0.9	ns
		Advance		

Notes: 1. Timing is based on the XC3142L, for other devices see timing calculator.
2. The use of two pull-up resistors per longline, available on other XC3000 devices, is not a valid option for XC3100L devices.

Product Obsolete or Under Obsolescence



XC3100L IOB Switching Characteristics Guidelines

Testing of the switching parameters is modeled after testing methods specified by MIL-M-38510/605. All devices are 100% functionally tested. Since many internal timing parameters cannot be measured directly, they are derived from benchmark timing patterns. The following guidelines reflect worst-case values over the recommended operating conditions. For more detailed, more precise, and more up-to-date timing information, use the values provided by the timing calculator and used in the simulator.

Description	Speed Grade		-3		-2		Units
	Symbol		Min	Max	Min	Max	
Propagation Delays (Input)							
Pad to Direct In (I)	3	T_{PID}		2.2		2.0	ns
Pad to Registered In (Q) with latch (XC3100L) transparent		T_{PTG}		11.0		11.0	ns
Clock (IK) to Registered In (Q)	4	T_{IKRI}		2.2		1.9	ns
Set-up Time (Input)							
Pad to Clock (IK) set-up time	1	T_{PICK}					
XC3142L			9.5		9.0		ns
XC3190L			9.9		9.4		ns
Propagation Delays (Output)							
Clock (OK) to Pad (fast)	7	T_{OKPO}		4.4		4.0	ns
same (slew rate limited)	7	T_{PO}		10.0		9.7	ns
Output (O) to Pad (fast)	10	T_{OPF}		3.3		3.0	ns
same (slew-rate limited)(XC3100L)	10	T_{OPF}		9.0		8.7	ns
3-state to Pad begin hi-Z (fast)	9	T_{TSHZ}		5.5		5.0	ns
same (slew-rate limited)	9	T_{TSHZ}		5.5		5.0	ns
3-state to Pad active and valid (fast)(XC3100L)	8	T_{TSON}		9.0		8.5	ns
same (slew -rate limited)	8	T_{TSON}		15.0		14.2	ns
Set-up and Hold Times (Output)							
Output (O) to clock (OK) set-up time (XC3100L)	5	T_{OOK}	4.0		3.6		ns
Output (O) to clock (OK) hold time	6	T_{OKO}	0		0		ns
Clock							
Clock High time	11	T_{IOH}	1.6		1.3		ns
Clock Low time	12	T_{IOL}	1.6		1.3		ns
Export Control Maximum flip-flop toggle rate		F_{TOG}	270		325		MHz
Global Reset Delays							
RESET Pad to Registered In (Q)							
(XC3142L)	13	T_{RRI}		16.0		16.0	ns
(XC3190L)				21.0		21.0	ns
RESET Pad to output pad (fast)	15	T_{RPO}		17.0		17.0	ns
(slew-rate limited)	15	T_{RPO}		23.0		23.0	ns
Advance							

7

- Notes:**
1. Timing is measured at pin threshold, with 50 pF external capacitive loads (incl. test fixture). Typical slew rate limited output rise/fall times are approximately four times longer.
 2. Voltage levels of unused (bonded and unbonded) pads must be valid logic levels. Each can be configured with the internal pull-up resistor or alternatively configured as a driven output or driven from an external source.
 3. Input pad set-up time is specified with respect to the internal clock (IK). In order to calculate system set-up time, subtract clock delay (pad to ik) from the input pad set-up time value. Input pad holdtime with respect to the internal clock (IK) is negative. This means that pad level changes immediately before the internal clock edge (IK) will not be recognized.

Product Obsolete or Under Obsolescence

XC3000 Series Field Programmable Gate Arrays



XC3064A/XC3090A/XC3195A 84-Pin PLCC Pinouts

XC3000A, XC3000L, XC3100A, and XC3100L families have identical pinouts

PLCC Pin Number	XC3064A, XC3090A, XC3195A
12	PWRDN
13	TCLKIN-I/O
14	I/O
15	I/O
16	I/O
17	I/O
18	I/O
19	I/O
20	I/O
21	GND*
22	VCC
23	I/O
24	I/O
25	I/O
26	I/O
27	I/O
28	I/O
29	I/O
30	I/O
31	M1-RDATA
32	M0-RTRIG
33	M2-I/O
34	HDC-I/O
35	I/O
36	LDC-I/O
37	I/O
38	I/O
39	I/O
40	I/O
41	INIT/I/O*
42	VCC*
43	GND
44	I/O
45	I/O
46	I/O
47	I/O
48	I/O
49	I/O
50	I/O
51	I/O
52	I/O
53	XTL2(IN)-I/O

PLCC Pin Number	XC3064A, XC3090A, XC3195A
54	RESET
55	DONE-PG
56	D7-I/O
57	XTL1(OUT)-BCLKIN-I/O
58	D6-I/O
59	I/O
60	D5-I/O
61	CS0-I/O
62	D4-I/O
63	I/O
64	VCC
65	GND*
66	D3-I/O*
67	CS1-I/O*
68	D2-I/O*
69	I/O
70	D1-I/O
71	RDY/BUSY-RCLK-I/O
72	D0-DIN-I/O
73	DOUT-I/O
74	CCLK
75	A0-WS-I/O
76	A1-CS2-I/O
77	A2-I/O
78	A3-I/O
79	I/O
80	I/O
81	A15-I/O
82	A4-I/O
83	A14-I/O
84	A5-I/O
1	GND
2	VCC*
3	A13-I/O*
4	A6-I/O*
5	A12-I/O*
6	A7-I/O*
7	I/O
8	A11-I/O
9	A8-I/O
10	A10-I/O
11	A9-I/O

Unprogrammed IOBs have a default pull-up. This prevents an undefined pad level for unbonded or unused IOBs. Programmed outputs are default slew-rate limited.

* In the PC84 package, XC3064A, XC3090A and XC3195A have additional VCC and GND pins and thus a different pin definition than XC3020A/XC3030A/XC3042A.

Product Obsolete or Under Obsolescence

XC3000 Series Field Programmable Gate Arrays



XC3000 Series 176-Pin TQFP Pinouts

XC3000A, XC3000L, XC3100A, and XC3100L families have identical pinouts

Pin Number	XC3090A	Pin Number	XC3090A	Pin Number	XC3090A	Pin Number	XC3090A
1	PWRDWN	45	M1-RDATA	89	DONE-PG	133	VCC
2	TCLKIN-I/O	46	GND	90	D7-I/O	134	GND
3	I/O	47	M0-RTRIG	91	XTAL1(OUT)-BCLKIN-I/O	135	A0-WS-I/O
4	I/O	48	VCC	92	I/O	136	A1-CS2-I/O
5	I/O	49	M2-I/O	93	I/O	137	–
6	I/O	50	HDC-I/O	94	I/O	138	I/O
7	I/O	51	I/O	95	I/O	139	I/O
8	I/O	52	I/O	96	D6-I/O	140	A2-I/O
9	I/O	53	I/O	97	I/O	141	A3-I/O
10	I/O	54	LDC-I/O	98	I/O	142	–
11	I/O	55	–	99	I/O	143	–
12	I/O	56	I/O	100	I/O	144	I/O
13	I/O	57	I/O	101	I/O	145	I/O
14	I/O	58	I/O	102	D5-I/O	146	A15-I/O
15	I/O	59	I/O	103	CS0-I/O	147	A4-I/O
16	I/O	60	I/O	104	I/O	148	I/O
17	I/O	61	I/O	105	I/O	149	I/O
18	I/O	62	I/O	106	I/O	150	A14-I/O
19	I/O	63	I/O	107	I/O	151	A5-I/O
20	I/O	64	I/O	108	D4-I/O	152	I/O
21	I/O	65	INIT-I/O	109	I/O	153	I/O
22	GND	66	VCC	110	VCC	154	GND
23	VCC	67	GND	111	GND	155	VCC
24	I/O	68	I/O	112	D3-I/O	156	A13-I/O
25	I/O	69	I/O	113	CS1-I/O	157	A6-I/O
26	I/O	70	I/O	114	I/O	158	I/O
27	I/O	71	I/O	115	I/O	159	I/O
28	I/O	72	I/O	116	I/O	160	–
29	I/O	73	I/O	117	I/O	161	–
30	I/O	74	I/O	118	D2-I/O	162	I/O
31	I/O	75	I/O	119	I/O	163	I/O
32	I/O	76	I/O	120	I/O	164	A12-I/O
33	I/O	77	I/O	121	I/O	165	A7-I/O
34	I/O	78	I/O	122	I/O	166	I/O
35	I/O	79	I/O	123	I/O	167	I/O
36	I/O	80	I/O	124	D1-I/O	168	–
37	I/O	81	I/O	125	RDY/BUSY-RCLK-I/O	169	A11-I/O
38	I/O	82	–	126	I/O	170	A8-I/O
39	I/O	83	–	127	I/O	171	I/O
40	I/O	84	I/O	128	I/O	172	I/O
41	I/O	85	XTAL2(IN)-I/O	129	I/O	173	A10-I/O
42	I/O	86	GND	130	D0-DIN-I/O	174	A9-I/O
43	I/O	87	RESET	131	DOOUT-I/O	175	VCC
44	–	88	VCC	132	CCLK	176	GND

Unprogrammed IOBs have a default pull-up. This prevents an undefined pad level for unbonded or unused IOBs. Programmed outputs are default slew-rate limited.