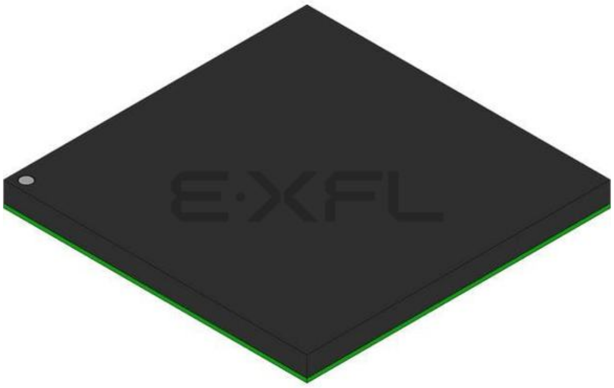
Freescale Semiconductor - MPC8313CZQAFFB Datasheet



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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e300c3
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	333MHz
Co-Processors/DSP	-
RAM Controllers	DDR, DDR2
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (2)
SATA	-
USB	USB 2.0 + PHY (1)
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	-40°C ~ 105°C (TA)
Security Features	-
Package / Case	516-BBGA Exposed Pad
Supplier Device Package	516-TEPBGA (27x27)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=mpc8313czqaffb

- Full and half-duplex Ethernet support (1000 Mbps supports only full-duplex):
  - IEEE 802.3 full-duplex flow control (automatic PAUSE frame generation or software-programmed PAUSE frame generation and recognition)
  - Programmable maximum frame length supports jumbo frames (up to 9.6 Kbytes) and IEEE 802.1 virtual local area network (VLAN) tags and priority
  - VLAN insertion and deletion
    - Per-frame VLAN control word or default VLAN for each eTSEC
    - Extracted VLAN control word passed to software separately
  - Retransmission following a collision
  - CRC generation and verification of inbound/outbound packets
  - Programmable Ethernet preamble insertion and extraction of up to 7 bytes
- MAC address recognition:
  - Exact match on primary and virtual 48-bit unicast addresses
    - VRRP and HSRP support for seamless router fail-over
  - Up to 16 exact-match MAC addresses supported
  - Broadcast address (accept/reject)
  - Hash table match on up to 512 multicast addresses
  - Promiscuous mode
- Buffer descriptors backward compatible with MPC8260 and MPC860T 10/100 Ethernet programming models
- RMON statistics support
- 10-Kbyte internal transmit and 2-Kbyte receive FIFOs
- MII management interface for control and status

1.8 Programmable Interrupt Controller (PIC)

The programmable interrupt controller (PIC) implements the necessary functions to provide a flexible solution for general-purpose interrupt control. The PIC programming model supports 5 external and 34 internal discrete interrupt sources. Interrupts can also be redirected to an external interrupt controller.

1.9 Power Management Controller (PMC)

The MPC8313E power management controller includes the following features:

- Provides power management when the device is used in both host and agent modes
- Supports PCI power management 1.2 D0, D1, D2, D3hot, and D3cold states
- On-chip split power supply controlled through external power switch for minimum standby power
- Support for PME generation in PCI agent mode, PME detection in PCI host mode
- Supports wake-up from Ethernet (Magic Packet), USB, GPIO, and PCI (PME input as host)

5.2 RESET AC Electrical Characteristics

This table provides the reset initialization AC timing specifications.

Table 10. RESET Initialization Timing Specifications

Parameter/Condition	Min	Max	Unit	Note
Required assertion time of $\overline{\text{HRESET}}$ or $\overline{\text{SRESET}}$ (input) to activate reset flow	32	—	$t_{\text{PCI\_SYNC\_IN}}$	1
Required assertion time of $\overline{\text{PORESET}}$ with stable clock and power applied to SYS_CLK_IN when the device is in PCI host mode	32	—	$t_{\text{SYS\_CLK\_IN}}$	2
Required assertion time of $\overline{\text{PORESET}}$ with stable clock and power applied to PCI_SYNC_IN when the device is in PCI agent mode	32	—	$t_{\text{PCI\_SYNC\_IN}}$	1
$\overline{\text{HRESET}}$ assertion (output)	512	—	$t_{\text{PCI\_SYNC\_IN}}$	1
Input setup time for POR configuration signals (CFG_RESET_SOURCE[0:3] and $\overline{\text{CFG\_CLK\_IN\_DIV}}$) with respect to negation of $\overline{\text{PORESET}}$ when the device is in PCI host mode	4	—	$t_{\text{SYS\_CLK\_IN}}$	2
Input setup time for POR configuration signals (CFG_RESET_SOURCE[0:2] and $\overline{\text{CFG\_CLKIN\_DIV}}$) with respect to negation of $\overline{\text{PORESET}}$ when the device is in PCI agent mode	4	—	$t_{\text{PCI\_SYNC\_IN}}$	1
Input hold time for POR configuration signals with respect to negation of $\overline{\text{HRESET}}$	0	—	ns	—
Time for the device to turn off POR configuration signal drivers with respect to the assertion of $\overline{\text{HRESET}}$	—	4	ns	3
Time for the device to turn on POR configuration signal drivers with respect to the negation of $\overline{\text{HRESET}}$	1	—	$t_{\text{PCI\_SYNC\_IN}}$	1, 3

Notes:
1. $t_{\text{PCI\_SYNC\_IN}}$ is the clock period of the input clock applied to PCI_SYNC_IN. When the device is In PCI host mode the primary clock is applied to the SYS_CLK_IN input, and PCI_SYNC_IN period depends on the value of $\overline{\text{CFG\_CLKIN\_DIV}}$.
2. $t_{\text{SYS\_CLK\_IN}}$ is the clock period of the input clock applied to SYS_CLK_IN. It is only valid when the device is in PCI host mode.
3. POR configuration signals consists of CFG_RESET_SOURCE[0:2] and $\overline{\text{CFG\_CLKIN\_DIV}}$.

This table provides the PLL lock times.

Table 11. PLL Lock Times

Parameter/Condition	Min	Max	Unit	Note
PLL lock times	—	100	µs	—

6 DDR and DDR2 SDRAM

This section describes the DC and AC electrical specifications for the DDR SDRAM interface. Note that DDR SDRAM is GV_{DD}(typ) = 2.5 V and DDR2 SDRAM is GV_{DD}(typ) = 1.8 V.

6.1 DDR and DDR2 SDRAM DC Electrical Characteristics

This table provides the recommended operating conditions for the DDR2 SDRAM component(s) when GV_{DD}(typ) = 1.8 V.

Table 12. DDR2 SDRAM DC Electrical Characteristics for GV_{DD}(typ) = 1.8 V

Parameter/Condition	Symbol	Min	Max	Unit	Note
I/O supply voltage	GV_{DD}	1.7	1.9	V	1
I/O reference voltage	MV_{REF}	$0.49 \times GV_{DD}$	$0.51 \times GV_{DD}$	V	2
I/O termination voltage	V_{TT}	$MV_{REF} - 0.04$	$MV_{REF} + 0.04$	V	3
Input high voltage	V_{IH}	$MV_{REF} + 0.125$	$GV_{DD} + 0.3$	V	—
Input low voltage	V_{IL}	–0.3	$MV_{REF} - 0.125$	V	—
Output leakage current	I_{OZ}	–9.9	9.9	μA	4
Output high current (V_{OUT} = 1.420 V)	I_{OH}	–13.4	—	mA	—
Output low current (V_{OUT} = 0.280 V)	I_{OL}	13.4	—	mA	—

Notes:

1. GV_{DD} is expected to be within 50 mV of the DRAM GV_{DD} at all times.
2. MV_{REF} is expected to be equal to $0.5 \times GV_{DD}$, and to track GV_{DD} DC variations as measured at the receiver. Peak-to-peak noise on MV_{REF} may not exceed ±2% of the DC value.
3. V_{TT} is not applied directly to the device. It is the supply to which far end signal termination is made and is expected to be equal to MV_{REF}. This rail should track variations in the DC level of MV_{REF}.
4. Output leakage is measured with all outputs disabled, $0\text{ V} \le V_{OUT} \le GV_{DD}$.

This table provides the DDR2 capacitance when GV_{DD}(typ) = 1.8 V.

Table 13. DDR2 SDRAM Capacitance for GV_{DD}(typ) = 1.8 V

Parameter/Condition	Symbol	Min	Max	Unit	Note
Input/output capacitance: DQ, DQS, $\overline{DQS}$	C_{IO}	6	8	pF	1
Delta input/output capacitance: DQ, DQS, $\overline{DQS}$	C_{DIO}	—	0.5	pF	1

Note:

1. This parameter is sampled. GV_{DD} = 1.8 V ± 0.090 V, f = 1 MHz, T_A = 25°C, $V_{OUT} = GV_{DD}/2$, V_{OUT} (peak-to-peak) = 0.2 V.

This table provides the recommended operating conditions for the DDR SDRAM component(s) when GV_{DD}(typ) = 2.5 V.

Table 14. DDR SDRAM DC Electrical Characteristics for GV_{DD}(typ) = 2.5 V

Parameter/Condition	Symbol	Min	Max	Unit	Note
I/O supply voltage	GV_{DD}	2.3	2.7	V	1
I/O reference voltage	MV_{REF}	$0.49 \times GV_{DD}$	$0.51 \times GV_{DD}$	V	2
I/O termination voltage	V_{TT}	$MV_{REF} - 0.04$	$MV_{REF} + 0.04$	V	3
Input high voltage	V_{IH}	$MV_{REF} + 0.15$	$GV_{DD} + 0.3$	V	—
Input low voltage	V_{IL}	–0.3	$MV_{REF} - 0.15$	V	—

This table provides the input AC timing specifications for the DDR SDRAM when GV_{DD}(typ) = 2.5 V.

Table 18. DDR SDRAM Input AC Timing Specifications for 2.5-V Interface

At recommended operating conditions with GV_{DD} of 2.5 ± 5%.

Parameter	Symbol	Min	Max	Unit	Note
AC input low voltage	V_{IL}	—	MV_{REF} – 0.31	V	—
AC input high voltage	V_{IH}	MV_{REF} + 0.31	—	V	—

This table provides the input AC timing specifications for the DDR2 SDRAM interface.

Table 19. DDR and DDR2 SDRAM Input AC Timing Specifications

At recommended operating conditions. with GV_{DD} of 2.5 ± 5%.

Parameter	Symbol	Min	Max	Unit	Note
Controller skew for MDQS—MDQ	t_{CISKEW}	—	—	ps	1, 2
333 MHz	—	–750	750	—	—
266 MHz	—	–750	750	—	—

Notes:

1. t_{CISKEW} represents the total amount of skew consumed by the controller between MDQS[*n*] and any corresponding bit that is captured with MDQS[*n*]. This should be subtracted from the total timing budget.
2. The amount of skew that can be tolerated from MDQS to a corresponding MDQ signal is called t_{DISKEW}. This can be determined by the following equation: $t_{DISKEW} = \pm (T/4 - \text{abs}(t_{CISKEW}))$ where T is the clock period and abs(t_{CISKEW}) is the absolute value of t_{CISKEW}.

This figure illustrates the DDR input timing diagram showing the t_{DISKEW} timing parameter.

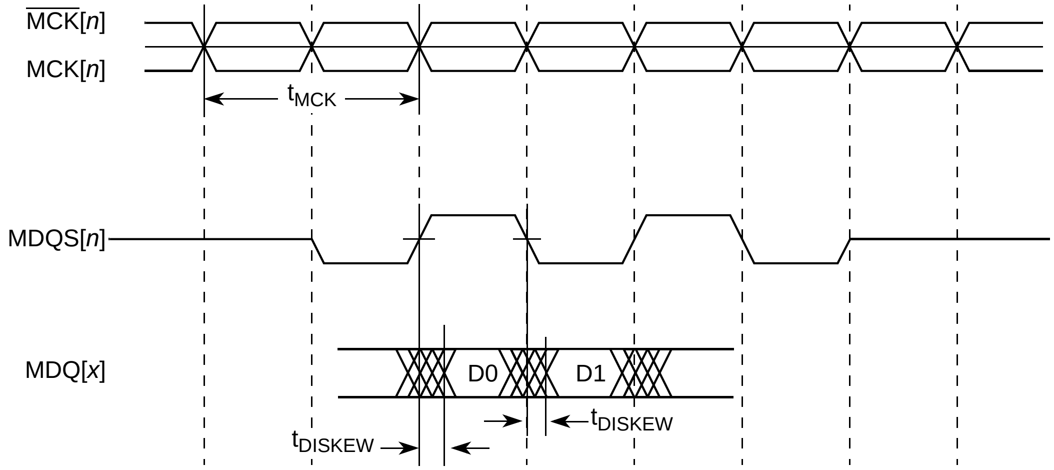


Figure 4. DDR Input Timing Diagram

This figure provides the AC test load for the DDR bus.

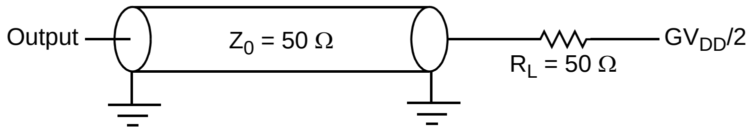


Figure 7. DDR AC Test Load

7 DUART

This section describes the DC and AC electrical specifications for the DUART interface.

7.1 DUART DC Electrical Characteristics

This table provides the DC electrical characteristics for the DUART interface.

Table 22. DUART DC Electrical Characteristics

Parameter	Symbol	Min	Max	Unit
High-level input voltage	V_{IH}	2.0	$NV_{DD} + 0.3$	V
Low-level input voltage NV_{DD}	V_{IL}	–0.3	0.8	V
High-level output voltage, $I_{OH} = -100\ \mu A$	V_{OH}	$NV_{DD} - 0.2$	—	V
Low-level output voltage, $I_{OL} = 100\ \mu A$	V_{OL}	—	0.2	V
Input current ($0\ V \leq V_{IN} \leq NV_{DD}$)	I_{IN}	—	±5	μA

7.2 DUART AC Electrical Specifications

This table provides the AC timing parameters for the DUART interface.

Table 23. DUART AC Timing Specifications

Parameter	Value	Unit	Note
Minimum baud rate	256	baud	—
Maximum baud rate	> 1,000,000	baud	1
Oversample rate	16	—	2

Notes:

1. Actual attainable baud rate is limited by the latency of interrupt processing.
2. The middle of a start bit is detected as the 8th sampled 0 after the 1-to-0 transition of the start bit. Subsequent bit values are sampled each 16th sample.

8 Ethernet: Three-Speed Ethernet, MII Management

This section provides the AC and DC electrical characteristics for three-speed, 10/100/1000, and MII management.

This figure shows the MII transmit AC timing diagram.

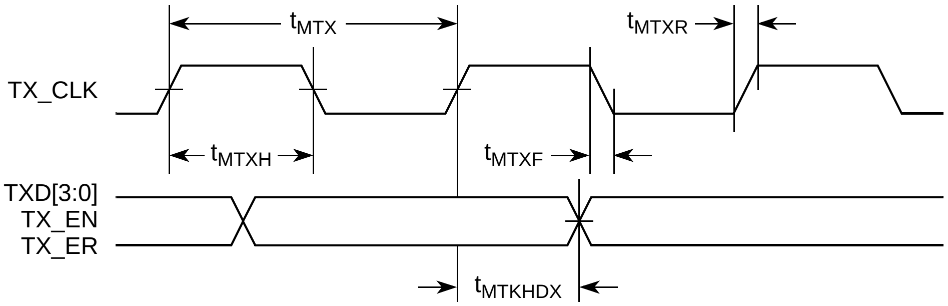


Figure 8. MII Transmit AC Timing Diagram

8.2.1.2 MII Receive AC Timing Specifications

This table provides the MII receive AC timing specifications.

Table 27. MII Receive AC Timing Specifications

At recommended operating conditions with $LV_{DDA}/LV_{DDB}/NV_{DD}$ of 3.3 V ± 0.3 V.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit
RX_CLK clock period 10 Mbps	t_{MRX}	—	400	—	ns
RX_CLK clock period 100 Mbps	t_{MRX}	—	40	—	ns
RX_CLK duty cycle	t_{MRXH}/t_{MRX}	35	—	65	%
RXD[3:0], RX_DV, RX_ER setup time to RX_CLK	t_{MRDVKH}	10.0	—	—	ns
RXD[3:0], RX_DV, RX_ER hold time to RX_CLK	t_{MRDXKH}	10.0	—	—	ns
RX_CLK clock rise V_{IL}(min) to V_{IH}(max)	t_{MRXR}	1.0	—	4.0	ns
RX_CLK clock fall time V_{IH}(max) to V_{IL}(min)	t_{MRXF}	1.0	—	4.0	ns

Note:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{MRDVKH} symbolizes MII receive timing (MR) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{MRX} clock reference (K) going to the high (H) state or setup time. Also, t_{MRDXKL} symbolizes MII receive timing (GR) with respect to the time data input signals (D) went invalid (X) relative to the t_{MRX} clock reference (K) going to the low (L) state or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{MRX} represents the MII (M) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. The frequency of RX_CLK should not exceed the TX_CLK by more than 300 ppm

This figure provides the AC test load for TSEC.

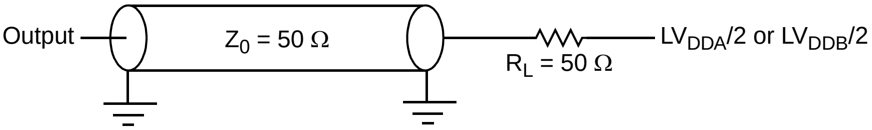


Figure 9. TSEC AC Test Load

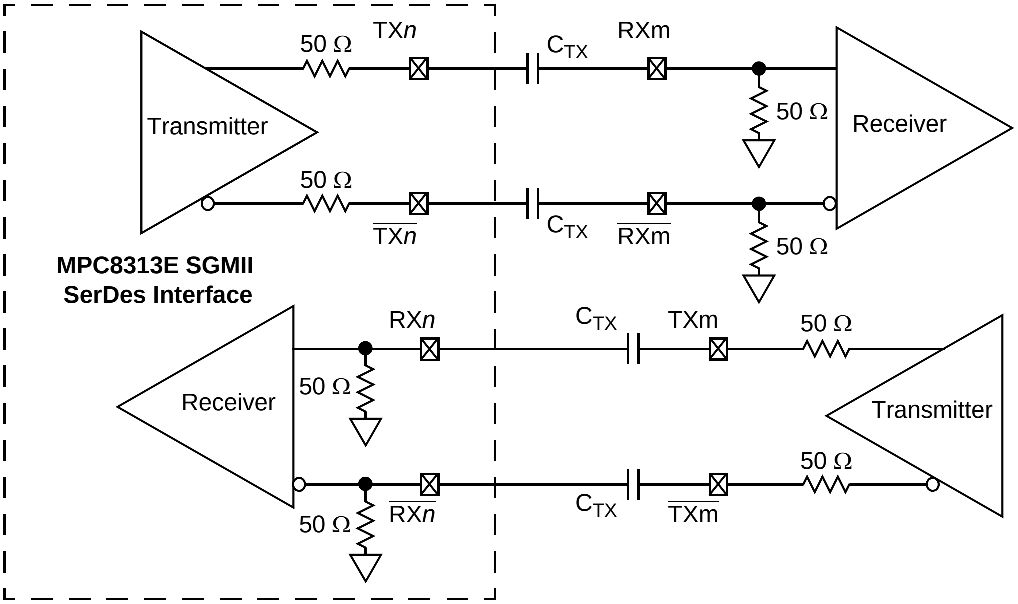


Figure 15. 4-Wire AC-Coupled SGMII Serial Link Connection Example

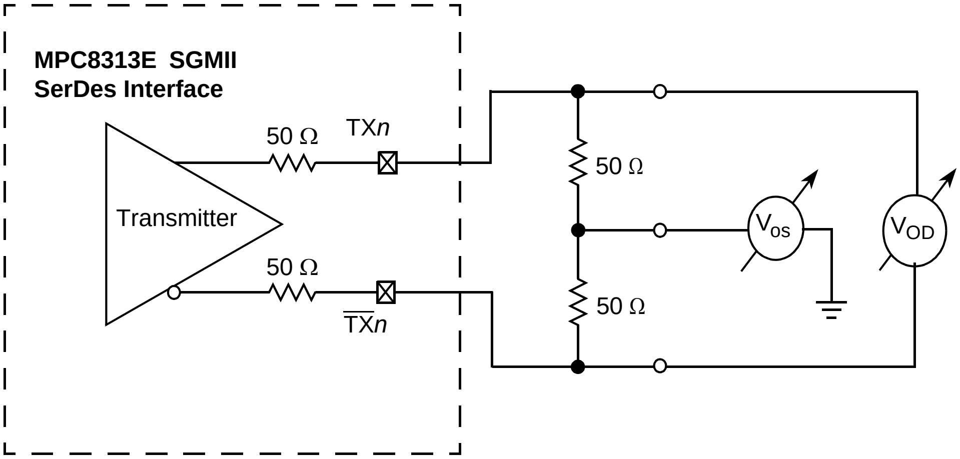


Figure 16. SGMII Transmitter DC Measurement Circuit

Table 33. SGMII DC Receiver Electrical Characteristics

Parameter	Symbol	Min	Typ	Max	Unit	Note
Supply voltage	$XCOREV_{DD}$	0.95	1.0	1.05	V	
DC Input voltage range			N/A			1
Input differential voltage	$V_{RX\_DIFFp-p}$	100	—	1200	mV	2
Loss of signal threshold	VL_{OS}	30	—	100	mV	
Input AC common mode voltage	V_{CM_ACp-p}	—	—	100	mV	3
Receiver differential input impedance	Z_{RX_DIFF}	80	100	120	Ω	
Receiver common mode input impedance	Z_{RX_CM}	20	—	35	Ω	

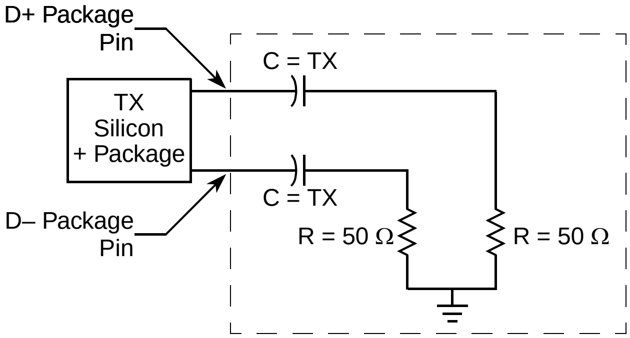
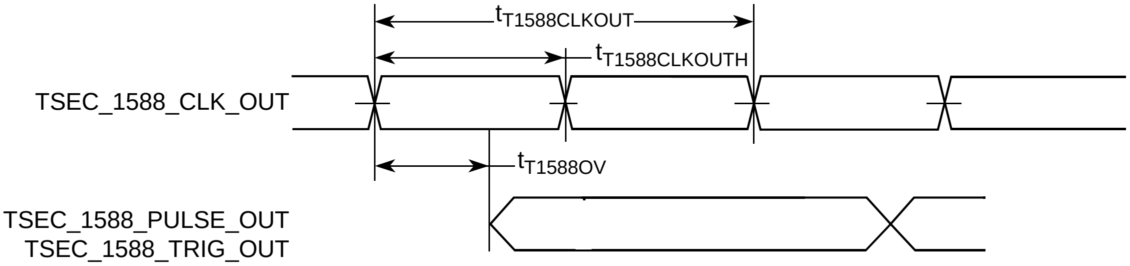


Figure 18. SGMII AC Test/Measurement Load

8.4 eTSEC IEEE 1588 AC Specifications

This figure provides the data and command output timing diagram.



Note: The output delay is count starting rising edge if $t_{T1588CLKOUT}$ is non-inverting. Otherwise, it is count starting falling edge.

Figure 19. eTSEC IEEE 1588 Output AC Timing

This figure provides the data and command input timing diagram.

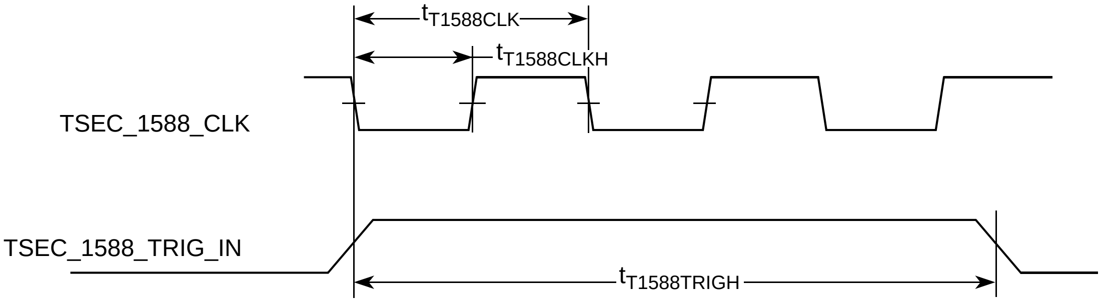


Figure 20. eTSEC IEEE 1588 Input AC Timing

This table lists the IEEE 1588 AC timing specifications.

Table 36. eTSEC IEEE 1588 AC Timing Specifications

At recommended operating conditions with L/TV_{DD} of 3.3 V ± 5%.

Parameter/Condition	Symbol	Min	Typ	Max	Unit	Note
TSEC_1588_CLK clock period	$t_{T1588CLK}$	3.8	—	$T_{RX\_CLK} \times 9$	ns	1, 3
TSEC_1588_CLK duty cycle	$t_{T1588CLKH}/t_{T1588CLK}$	40	50	60	%	

The common mode voltage is equal to one half of the sum of the voltages between each conductor of a balanced interchange circuit and ground. In this example, for SerDes output, $V_{cm\_out} = (V_{TXn} + V_{\overline{TXn}})/2 = (A + B)/2$, which is the arithmetic mean of the two complimentary output voltages within a differential pair. In a system, the common mode voltage may often differ from one component’s output to the other’s input. Sometimes, it may be even different between the receiver input and driver output circuits within the same component. It’s also referred as the DC offset in some occasion.

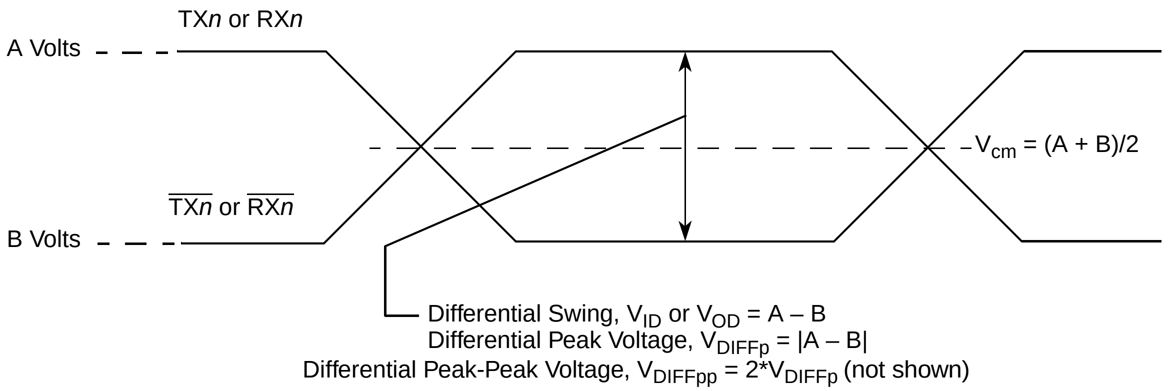


Figure 22. Differential Voltage Definitions for Transmitter or Receiver

To illustrate these definitions using real values, consider the case of a CML (current mode logic) transmitter that has a common mode voltage of 2.25 V and each of its outputs, TD and $\overline{\text{TD}}$, has a swing that goes between 2.5 and 2.0 V. Using these values, the peak-to-peak voltage swing of each signal (TD or $\overline{\text{TD}}$) is 500 mV p-p, which is referred as the single-ended swing for each signal. In this example, since the differential signaling environment is fully symmetrical, the transmitter output’s differential swing (V_{OD}) has the same amplitude as each signal’s single-ended swing. The differential output signal ranges between 500 and –500 mV, in other words, V_{OD} is 500 mV in one phase and –500 mV in the other phase. The peak differential voltage (V_{DIFFp}) is 500 mV. The peak-to-peak differential voltage ($V_{DIFFp\text{-}p}$) is 1000 mV p-p.

9.2 SerDes Reference Clocks

The SerDes reference clock inputs are applied to an internal PLL whose output creates the clock used by the corresponding SerDes lanes. The SerDes reference clocks input is SD_REF_CLK and $\overline{\text{SD\_REF\_CLK}}$ for SGMII interface.

The following sections describe the SerDes reference clock requirements and some application information.

9.2.1 SerDes Reference Clock Receiver Characteristics

Figure 23 shows a receiver reference diagram of the SerDes reference clocks.

- The supply voltage requirements for $XCOREV_{DD}$ are specified in Table 1 and Table 2.
- SerDes reference clock receiver reference circuit structure:

10 USB

10.1 USB Dual-Role Controllers

This section provides the AC and DC electrical specifications for the USB interface.

10.1.1 USB DC Electrical Characteristics

This table provides the DC electrical characteristics for the USB interface.

Table 40. USB DC Electrical Characteristics

Parameter	Symbol	Min	Max	Unit
High-level input voltage	V_{IH}	2.0	LV_{DDB} + 0.3	V
Low-level input voltage	V_{IL}	–0.3	0.8	V
Input current	I_{IN}	—	±5	μA
High-level output voltage, I_{OH} = –100 μA	V_{OH}	LV_{DDB} – 0.2	—	V
Low-level output voltage, I_{OL} = 100 μA	V_{OL}	—	0.2	V

10.1.2 USB AC Electrical Specifications

This table describes the general timing parameters of the USB interface.

Table 41. USB General Timing Parameters (ULPI Mode Only)

Parameter	Symbol[1]	Min	Max	Unit	Note
USB clock cycle time	t_{USCK}	15	—	ns	
Input setup to USB clock—all inputs	t_{USIVKH}	4	—	ns	
input hold to USB clock—all inputs	t_{USIXKH}	1	—	ns	
USB clock to output valid—all outputs	t_{USKHOV}	—	7	ns	
Output hold from USB clock—all outputs	t_{USKHOX}	2	—	ns	

Note:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{USIXKH} symbolizes USB timing (USB) for the input (I) to go invalid (X) with respect to the time the USB clock reference (K) goes high (H). Also, t_{USKHOX} symbolizes us timing (USB) for the USB clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.

The following two figures provide the AC test load and signals for the USB, respectively.

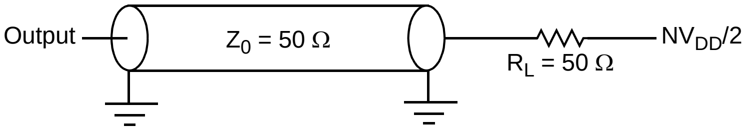


Figure 34. USB AC Test Load

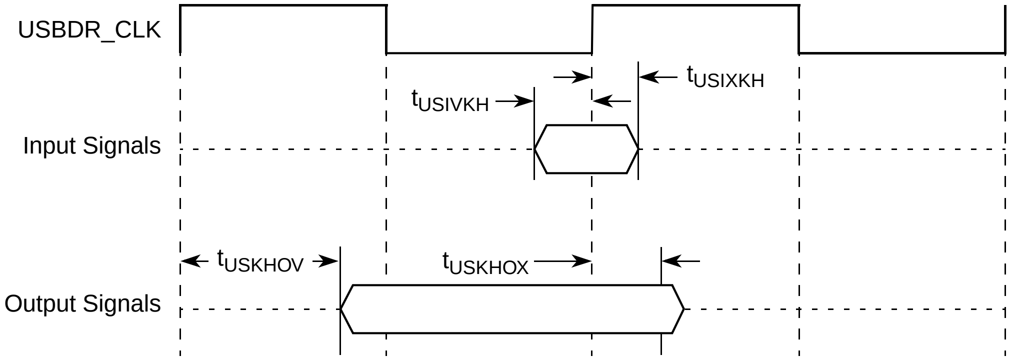


Figure 35. USB Signals

10.2 On-Chip USB PHY

This section describes the DC and AC electrical specifications for the on-chip USB PHY of the MPC8313E. See Chapter 7 in the *USB Specifications Rev. 2*, for more information.

This table provides the USB clock input (USB_CLK_IN) DC timing specifications.

Table 42. USB_CLK_IN DC Electrical Characteristics

Parameter	Symbol	Min	Max	Unit
Input high voltage	V_{IH}	2.7	NV_{DD} + 0.3	V
Input low voltage	V_{IL}	–0.3	0.4	V

This table provides the USB clock input (USB_CLK_IN) AC timing specifications.

Table 43. USB_CLK_IN AC Timing Specifications

Parameter/Condition	Conditions	Symbol	Min	Typ	Max	Unit
Frequency range	—	$f_{USB\_CLK\_IN}$	—	24	48	MHz
Clock frequency tolerance	—	t_{CLK_TOL}	–0.005	0	0.005	%
Reference clock duty cycle	Measured at 1.6 V	t_{CLK_DUTY}	40	50	60	%
Total input jitter/time interval error	Peak-to-peak value measured with a second order high-pass filter of 500 kHz bandwidth	t_{CLK_PJ}	—	—	200	ps

This figure shows the AC timing diagram for the I^2C bus.

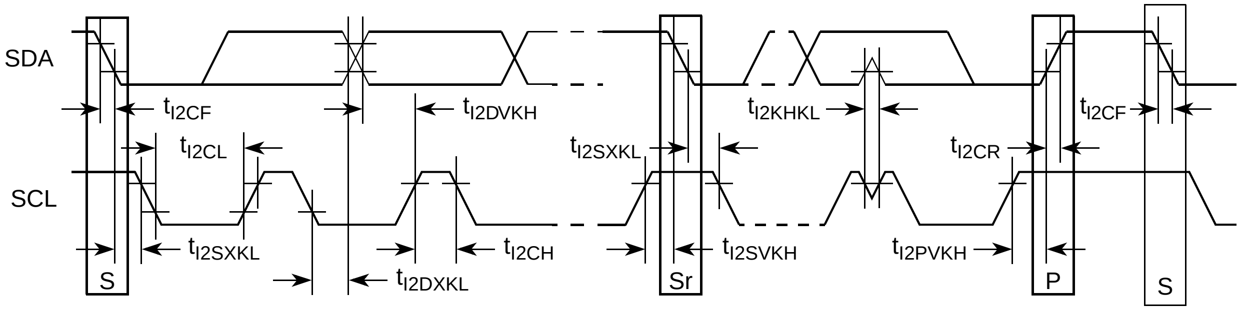


Figure 47. I^2C Bus AC Timing Diagram

14 PCI

This section describes the DC and AC electrical specifications for the PCI bus.

14.1 PCI DC Electrical Characteristics

This table provides the DC electrical characteristics for the PCI interface.

Table 50. PCI DC Electrical Characteristics[1]

Parameter	Symbol	Test Condition	Min	Max	Unit
High-level input voltage	V_{IH}	$V_{OUT} \geq V_{OH}$ (min) or $V_{OUT} \leq V_{OL}$ (max)	$0.5 \times NV_{DD}$	$NV_{DD} + 0.3$	V
Low-level input voltage	V_{IL}		–0.5	$0.3 \times NV_{DD}$	V
High-level output voltage	V_{OH}	NV_{DD} = min, I_{OH} = –100 μA	$0.9 \times NV_{DD}$	—	V
Low-level output voltage	V_{OL}	NV_{DD} = min, I_{OL} = 100 μA	—	$0.1 \times NV_{DD}$	V
Input current	I_{IN}	$0 \text{ V} \leq V_{IN} \leq NV_{DD}$	—	±5	μA

Note:

1. Note that the symbol V_{IN}, in this case, represents the NV_{IN} symbol referenced in Table 1 and Table 2.

14.2 PCI AC Electrical Specifications

This section describes the general AC timing parameters of the PCI bus. Note that the PCI_CLK or PCI_SYNC_IN signal is used as the PCI input clock depending on whether the MPC8313E is configured as a host or agent device.

This table shows the PCI AC timing specifications at 66 MHz.

Table 51. PCI AC Timing Specifications at 66 MHz

Parameter	Symbol[1]	Min	Max	Unit	Note
Clock to output valid	t_{PCKHOV}	—	6.0	ns	2
Output hold from clock	t_{PCKHOX}	1	—	ns	2

Table 62. MPC8313E TEPBGAII Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Note
$\overline{\text{MEMC\_MCS0}}$	D10	O	GV_{DD}	—
$\overline{\text{MEMC\_MCS1}}$	A10	O	GV_{DD}	—
MEMC_MCKE	B14	O	GV_{DD}	3
MEMC_MCK	A13	O	GV_{DD}	—
$\overline{\text{MEMC\_MCK}}$	A14	O	GV_{DD}	—
MEMC_MODT0	B23	O	GV_{DD}	—
MEMC_MODT1	C23	O	GV_{DD}	—
Local Bus Controller Interface				
LAD0	K25	I/O	LV_{DD}	11
LAD1	K24	I/O	LV_{DD}	11
LAD2	K23	I/O	LV_{DD}	11
LAD3	K22	I/O	LV_{DD}	11
LAD4	J25	I/O	LV_{DD}	11
LAD5	J24	I/O	LV_{DD}	11
LAD6	J23	I/O	LV_{DD}	11
LAD7	J22	I/O	LV_{DD}	11
LAD8	H24	I/O	LV_{DD}	11
LAD9	F26	I/O	LV_{DD}	11
LAD10	G24	I/O	LV_{DD}	11
LAD11	F25	I/O	LV_{DD}	11
LAD12	E25	I/O	LV_{DD}	11
LAD13	F24	I/O	LV_{DD}	11
LAD14	G22	I/O	LV_{DD}	11
LAD15	F23	I/O	LV_{DD}	11
LA16	AC25	O	LV_{DD}	11
LA17	AC26	O	LV_{DD}	11
LA18	AB22	O	LV_{DD}	11
LA19	AB23	O	LV_{DD}	11
LA20	AB24	O	LV_{DD}	11
LA21	AB25	O	LV_{DD}	11
LA22	AB26	O	LV_{DD}	11
LA23	E22	O	LV_{DD}	11

The primary clock source for the MPC8313E can be one of two inputs, SYS_CLK_IN or PCI_CLK, depending on whether the device is configured in PCI host or PCI agent mode. When the device is configured as a PCI host device, SYS_CLK_IN is its primary input clock. SYS_CLK_IN feeds the PCI clock divider (÷2) and the multiplexors for PCI_SYNC_OUT and PCI_CLK_OUT. The CFG_CLKIN_DIV configuration input selects whether SYS_CLK_IN or SYS_CLK_IN/2 is driven out on the PCI_SYNC_OUT signal. The OCCR[PCICOE*n*] parameters select whether the PCI_SYNC_OUT is driven out on the PCI_CLK_OUT*n* signals.

PCI_SYNC_OUT is connected externally to PCI_SYNC_IN to allow the internal clock subsystem to synchronize to the system PCI clocks. PCI_SYNC_OUT must be connected properly to PCI_SYNC_IN, with equal delay to all PCI agent devices in the system, to allow the device to function. When the device is configured as a PCI agent device, PCI_CLK is the primary input clock. When the device is configured as a PCI agent device the SYS_CLK_IN signal should be tied to VSS.

As shown in Figure 57, the primary clock input (frequency) is multiplied up by the system phase-locked loop (PLL) and the clock unit to create the coherent system bus clock (*csb_clk*), the internal clock for the DDR controller (*ddr_clk*), and the internal clock for the local bus interface unit (*lbc_clk*).

The *csb_clk* frequency is derived from a complex set of factors that can be simplified into the following equation:

$$\mathit{csb\_clk} = \{\text{PCI\_SYNC\_IN} \times (1 + {\sim}\overline{\text{CFG\_CLKIN\_DIV}})\} \times \text{SPMF}$$

In PCI host mode, $\text{PCI\_SYNC\_IN} \times (1 + {\sim}\overline{\text{CFG\_CLKIN\_DIV}})$ is the SYS_CLK_IN frequency.

The *csb_clk* serves as the clock input to the e300 core. A second PLL inside the e300 core multiplies up the *csb_clk* frequency to create the internal clock for the e300 core (*core_clk*). The system and core PLL multipliers are selected by the SPMF and COREPLL fields in the reset configuration word low (RCWL) which is loaded at power-on reset or by one of the hard-coded reset options. See Chapter 4, “Reset, Clocking, and Initialization,” in the *MPC8313E PowerQUICC II Pro Integrated Processor Family Reference Manual,* for more information on the clock subsystem.

The internal *ddr_clk* frequency is determined by the following equation:

$$\mathit{ddr\_clk} = \mathit{csb\_clk} \times (1 + \text{RCWL[DDRCM]})$$

Note that *ddr_clk* is not the external memory bus frequency; *ddr_clk* passes through the DDR clock divider (÷2) to create the differential DDR memory bus clock outputs (MCK and $\overline{\text{MCK}}$). However, the data rate is the same frequency as *ddr_clk*.

The internal *lbc_clk* frequency is determined by the following equation:

$$\mathit{lbc\_clk} = \mathit{csb\_clk} \times (1 + \text{RCWL[LBCM]})$$

Note that *lbc_clk* is not the external local bus frequency; *lbc_clk* passes through the a LBC clock divider to create the external local bus clock outputs (LCLK[0:1]). The LBC clock divider ratio is controlled by LCRR[CLKDIV].

In addition, some of the internal units may be required to be shut off or operate at lower frequency than the *csb_clk* frequency. Those units have a default clock ratio that can be configured by a memory mapped register after the device comes out of reset. Table 63 specifies which units have a configurable clock frequency.

20.3 Example Clock Frequency Combinations

This table shows several possible frequency combinations that can be selected based on the indicated input reference frequencies, with RCWLR[LBCM] = 0 and RCWLR[DDRCM] =1, such that the LBC operates with a frequency equal to the frequency of *csb_clk* and the DDR controller operates at twice the frequency of *csb_clk*.

Table 68. System Clock Frequencies

						LBC(lbc_clk)				**e300 Core(core_clk)**				
SYS_ CLK_IN/ PCI_CLK	**SPMF**[1]	**VCOD**[2]	**VCO**[3]	**CSB (*csb_clk*)**[4]	**DDR (ddr_clk)**	**/2**	**/4**	**/8**	**USB ref**[5]	**× 1**	**× 1.5**	**× 2**	**× 2.5**	**× 3**
25.0	6	2	600.0	150.0	300.0	—	37.5	18.8	Note[6]	150.0	225	300	375	—
25.0	5	2	500.0	125.0	250.0	62.5	31.25	15.6	Note 6	125.0	188	250	313	375
33.3	5	2	666.0	166.5	333.0	—	41.63	20.8	Note 6	166.5	250	333	—	—
33.3	4	2	532.8	133.2	266.4	66.6	33.3	16.7	Note 6	133.2	200	266	333	400
48.0	3	2	576.0	144.0	288.0	—	36	18.0	48.0	144.0	216	288	360	—
66.7	2	2	533.4	133.3	266.7	66.7	33.34	16.7	Note 6	133.3	200	267	333	400

Note:

1. System PLL multiplication factor.
2. System PLL VCO divider.
3. When considering operating frequencies, the valid core VCO operating range of 400–800 MHz must not be violated.
4. Due to erratum eTSEC40, *csb_clk* frequencies of less than 133 MHz do not support gigabit Ethernet data rates. The core frequency must be 333 MHz for gigabit Ethernet operation. This erratum will be fixed in revision 2 silicon.
5. Frequency of USB PLL input reference.
6. USB reference clock must be supplied from a separate source as it must be 24 or 48 MHz, the USB reference must be supplied from a separate external source using USB_CLK_IN.

21 Thermal

This section describes the thermal specifications of the MPC8313E.

21.1 Thermal Characteristics

This table provides the package thermal characteristics for the 516, 27 × 27 mm TEPBGAII.

Table 69. Package Thermal Characteristics for TEPBGAII

Characteristic	Board Type	Symbol	TEPBGA II	Unit	Note
Junction-to-ambient natural convection	Single layer board (1s)	$R_{\theta JA}$	25	°C/W	1, 2
Junction-to-ambient natural convection	Four layer board (2s2p)	$R_{\theta JA}$	18	°C/W	1, 2, 3
Junction-to-ambient (@200 ft/min)	Single layer board (1s)	$R_{\theta JMA}$	20	°C/W	1, 3
Junction-to-ambient (@200 ft/min)	Four layer board (2s2p)	$R_{\theta JMA}$	15	°C/W	1, 3
Junction-to-board	—	$R_{\theta JB}$	10	°C/W	4

Table 69. Package Thermal Characteristics for TEPBGAII (continued)

Characteristic	Board Type	Symbol	TEPBGA II	Unit	Note
Junction-to-case	—	$R_{\theta JC}$	8	°C/W	5
Junction-to-package top	Natural convection	Ψ_{JT}	7	°C/W	6

Note:

1. Junction temperature is a function of die size, on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, airflow, power dissipation of other components on the board, and board thermal resistance.
2. Per JEDEC JESD51-2 with the single layer board horizontal. Board meets JESD51-9 specification.
3. Per JEDEC JESD51-6 with the board horizontal.
4. Thermal resistance between the die and the printed-circuit board per JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.
5. Thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1).
6. Thermal characterization parameter indicating the temperature difference between package top and the junction temperature per JEDEC JESD51-2. When Greek letters are not available, the thermal characterization parameter is written as Psi-JT.

21.2 Thermal Management Information

For the following sections, $P_D = (V_{DD} \times I_{DD}) + P_{I/O}$, where $P_{I/O}$ is the power dissipation of the I/O drivers.

21.2.1 Estimation of Junction Temperature with Junction-to-Ambient Thermal Resistance

An estimation of the chip junction temperature, T_J, can be obtained from the equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

T_J = junction temperature (°C)
T_A = ambient temperature for the package (°C)
$R_{\theta JA}$ = junction-to-ambient thermal resistance (°C/W)
P_D = power dissipation in the package (W)

The junction-to-ambient thermal resistance is an industry standard value that provides a quick and easy estimation of thermal performance. As a general statement, the value obtained on a single layer board is appropriate for a tightly packed printed-circuit board. The value obtained on the board with the internal planes is usually appropriate if the board has low power dissipation and the components are well separated. Test cases have demonstrated that errors of a factor of two (in the quantity $T_J - T_A$) are possible.

21.2.2 Estimation of Junction Temperature with Junction-to-Board Thermal Resistance

The thermal performance of a device cannot be adequately predicted from the junction-to-ambient thermal resistance. The thermal performance of any component is strongly dependent on the power dissipation of surrounding components. In addition, the ambient temperature varies widely within the application. For many natural convection and especially closed box applications, the board temperature at the perimeter (edge) of the package is approximately the same as the local air temperature near the device. Specifying the local ambient conditions explicitly as the board temperature provides a more precise description of the local ambient conditions that determine the temperature of the device.

At a known board temperature, the junction temperature is estimated using the following equation:

$$T_J = T_B + (R_{\theta JB} \times P_D)$$

where:

T_J = junction temperature (°C)
T_B = board temperature at the package perimeter (°C)
$R_{\theta JB}$ = junction-to-board thermal resistance (°C/W) per JESD51–8
P_D = power dissipation in the package (W)

When the heat loss from the package case to the air can be ignored, acceptable predictions of junction temperature can be made. The application board should be similar to the thermal test condition: the component is soldered to a board with internal planes.

21.2.3 Experimental Determination of Junction Temperature

To determine the junction temperature of the device in the application after prototypes are available, the thermal characterization parameter (Ψ_{JT}) can be used to determine the junction temperature with a measurement of the temperature at the top center of the package case using the following equation:

$$T_J = T_T + (\Psi_{JT} \times P_D)$$

where:

T_J = junction temperature (°C)
T_T = thermocouple temperature on top of package (°C)
Ψ_{JT} = thermal characterization parameter (°C/W)
P_D = power dissipation in the package (W)

The thermal characterization parameter is measured per JESD51-2 specification using a 40 gauge type T thermocouple epoxied to the top center of the package case. The thermocouple should be positioned so that the thermocouple junction rests on the package. A small amount of epoxy is placed over the thermocouple junction and over about 1 mm of wire extending from the junction. The thermocouple wire is placed flat against the package case to avoid measurement errors caused by cooling effects of the thermocouple wire.

21.2.4 Heat Sinks and Junction-to-Case Thermal Resistance

In some application environments, a heat sink is required to provide the necessary thermal management of the device. When a heat sink is used, the thermal resistance is expressed as the sum of a junction to case thermal resistance and a case to ambient thermal resistance:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$$

where:

$R_{\theta JA}$ = junction-to-ambient thermal resistance (°C/W)
$R_{\theta JC}$ = junction-to-case thermal resistance (°C/W)
$R_{\theta CA}$ = case-to- ambient thermal resistance (°C/W)

21.3 Heat Sink Attachment

When attaching heat sinks to these devices, an interface material is required. The best method is to use thermal grease and a spring clip. The spring clip should connect to the printed-circuit board, either to the board itself, to hooks soldered to the board, or to a plastic stiffener. Avoid attachment forces which would lift the edge of the package or peel the package from the board. Such peeling forces reduce the solder joint lifetime of the package. Recommended maximum force on the top of the package is 10 lb (4.5 kg) force. If an adhesive attachment is planned, the adhesive should be intended for attachment to painted or plastic surfaces and its performance verified under the application requirements.

21.3.1 Experimental Determination of the Junction Temperature with a Heat Sink

When heat sink is used, the junction temperature is determined from a thermocouple inserted at the interface between the case of the package and the interface material. A clearance slot or hole is normally required in the heat sink. Minimizing the size of the clearance is important to minimize the change in thermal performance caused by removing part of the thermal interface to the heat sink. Because of the experimental difficulties with this technique, many engineers measure the heat sink temperature and then back calculate the case temperature using a separate measurement of the thermal resistance of the interface. From this case temperature, the junction temperature is determined from the junction to case thermal resistance.

$$T_J = T_C + (R_{\theta JC} \text{ x } P_D)$$

where:

T_J = junction temperature (°C)
T_C = case temperature of the package
$R_{\theta JC}$ = junction-to-case thermal resistance
P_D = power dissipation

22 System Design Information

This section provides electrical and thermal design recommendations for successful application of the MPC8313E SYS_CLK_IN

22.1 System Clocking

The MPC8313E includes three PLLs.

1. The platform PLL (AV_{DD2}) generates the platform clock from the externally supplied SYS_CLK_IN input in PCI host mode or SYS_CLK_IN/PCI_SYNC_IN in PCI agent mode. The frequency ratio between the platform and SYS_CLK_IN is selected using the platform PLL ratio configuration bits as described in Section 20.1, “System PLL Configuration.”
2. The e300 core PLL (AV_{DD1}) generates the core clock as a slave to the platform clock. The frequency ratio between the e300 core clock and the platform clock is selected using the e300 PLL ratio configuration bits as described in Section 20.2, “Core PLL Configuration.”
3. There is a PLL for the SerDes block.

- Third, between the device and any SerDes voltage regulator there should be a 10-µF, low equivalent series resistance (ESR) SMT tantalum chip capacitor and a 100-µF, low ESR SMT tantalum chip capacitor. This should be done for all SerDes supplies.

22.5 Connection Recommendations

To ensure reliable operation, it is highly recommended to connect unused inputs to an appropriate signal level. Unused active low inputs should be tied to NV_{DD}, GV_{DD}, LV_{DD}, LV_{DDA}, or LV_{DDB} as required. Unused active high inputs should be connected to V_{SS}. All NC (no-connect) signals must remain unconnected.

Power and ground connections must be made to all external V_{DD}, NV_{DD}, GV_{DD}, LV_{DD}, LV_{DDA}, LV_{DDB}, and V_{SS} pins of the device.

22.6 Output Buffer DC Impedance

The MPC8313E drivers are characterized over process, voltage, and temperature. For all buses, the driver is a push-pull single-ended driver type (open drain for I^2C).

To measure Z_0 for the single-ended drivers, an external resistor is connected from the chip pad to NV_{DD} or V_{SS}. Then, the value of each resistor is varied until the pad voltage is $NV_{DD}/2$ (see Figure 60). The output impedance is the average of two components, the resistances of the pull-up and pull-down devices. When data is held high, SW1 is closed (SW2 is open), and R_P is trimmed until the voltage at the pad equals $NV_{DD}/2$. R_P then becomes the resistance of the pull-up devices. R_P and R_N are designed to be close to each other in value. Then, $Z_0 = (R_P + R_N)/2$.

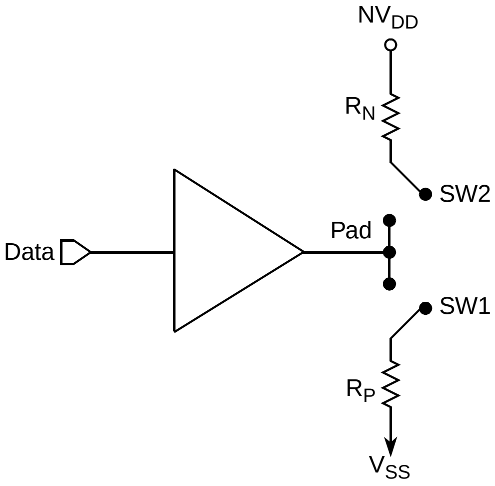


Figure 60. Driver Impedance Measurement

The value of this resistance and the strength of the driver's current source can be found by making two measurements. First, the output voltage is measured while driving logic 1 without an external differential termination resistor. The measured voltage is $V_1 = R_{source} \times I_{source}$. Second, the output voltage is measured while driving logic 1 with an external precision differential termination resistor of value R_{term}. The measured voltage is $V_2 = (1/(1/R_1 + 1/R_2)) \times I_{source}$. Solving for the output impedance gives $R_{source} = R_{term} \times (V_1/V_2 - 1)$. The drive current is then $I_{source} = V_1/R_{source}$.

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