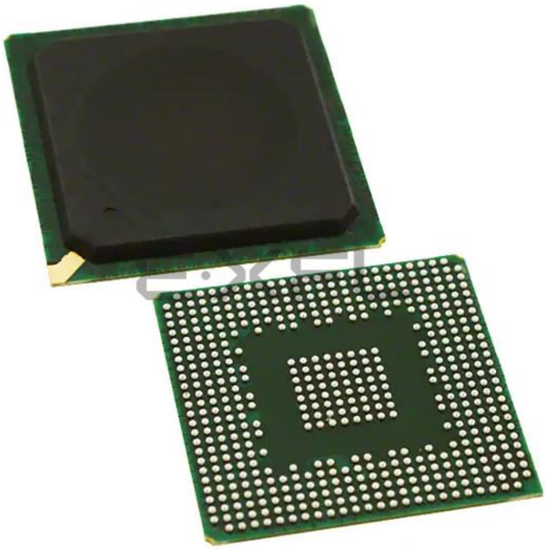

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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e300c3
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	267MHz
Co-Processors/DSP	Security; SEC 2.2
RAM Controllers	DDR, DDR2
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (2)
SATA	-
USB	USB 2.0 + PHY (1)
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	Cryptography
Package / Case	516-BBGA Exposed Pad
Supplier Device Package	516-TEPBGA (27x27)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8313ezqadd

5.2 RESET AC Electrical Characteristics

This table provides the reset initialization AC timing specifications.

Table 10. RESET Initialization Timing Specifications

Parameter/Condition	Min	Max	Unit	Note
Required assertion time of $\overline{\text{HRESET}}$ or $\overline{\text{SRESET}}$ (input) to activate reset flow	32	—	$t_{PCI\_SYNC\_IN}$	1
Required assertion time of $\overline{\text{PORESET}}$ with stable clock and power applied to SYS_CLK_IN when the device is in PCI host mode	32	—	$t_{SYS\_CLK\_IN}$	2
Required assertion time of $\overline{\text{PORESET}}$ with stable clock and power applied to PCI_SYNC_IN when the device is in PCI agent mode	32	—	$t_{PCI\_SYNC\_IN}$	1
$\overline{\text{HRESET}}$ assertion (output)	512	—	$t_{PCI\_SYNC\_IN}$	1
Input setup time for POR configuration signals (CFG_RESET_SOURCE[0:3] and $\overline{\text{CFG\_CLK\_IN\_DIV}}$) with respect to negation of $\overline{\text{PORESET}}$ when the device is in PCI host mode	4	—	$t_{SYS\_CLK\_IN}$	2
Input setup time for POR configuration signals (CFG_RESET_SOURCE[0:2] and $\overline{\text{CFG\_CLKIN\_DIV}}$) with respect to negation of $\overline{\text{PORESET}}$ when the device is in PCI agent mode	4	—	$t_{PCI\_SYNC\_IN}$	1
Input hold time for POR configuration signals with respect to negation of $\overline{\text{HRESET}}$	0	—	ns	—
Time for the device to turn off POR configuration signal drivers with respect to the assertion of $\overline{\text{HRESET}}$	—	4	ns	3
Time for the device to turn on POR configuration signal drivers with respect to the negation of $\overline{\text{HRESET}}$	1	—	$t_{PCI\_SYNC\_IN}$	1, 3

Notes:
1. $t_{PCI\_SYNC\_IN}$ is the clock period of the input clock applied to PCI_SYNC_IN. When the device is In PCI host mode the primary clock is applied to the SYS_CLK_IN input, and PCI_SYNC_IN period depends on the value of $\overline{\text{CFG\_CLKIN\_DIV}}$.
2. $t_{SYS\_CLK\_IN}$ is the clock period of the input clock applied to SYS_CLK_IN. It is only valid when the device is in PCI host mode.
3. POR configuration signals consists of CFG_RESET_SOURCE[0:2] and $\overline{\text{CFG\_CLKIN\_DIV}}$.

This table provides the PLL lock times.

Table 11. PLL Lock Times

Parameter/Condition	Min	Max	Unit	Note
PLL lock times	—	100	µs	—

6 DDR and DDR2 SDRAM

This section describes the DC and AC electrical specifications for the DDR SDRAM interface. Note that DDR SDRAM is GV_{DD}(typ) = 2.5 V and DDR2 SDRAM is GV_{DD}(typ) = 1.8 V.

Table 14. DDR SDRAM DC Electrical Characteristics for GV_{DD}(typ) = 2.5 V (continued)

Parameter/Condition	Symbol	Min	Max	Unit	Note
Output leakage current	I_{OZ}	–9.9	–9.9	μA	4
Output high current (V_{OUT} = 1.95 V)	I_{OH}	–16.2	—	mA	—
Output low current (V_{OUT} = 0.35 V)	I_{OL}	16.2	—	mA	—

Note:
1. GV_{DD} is expected to be within 50 mV of the DRAM GV_{DD} at all times.
2. MV_{REF} is expected to be equal to 0.5 × GV_{DD}, and to track GV_{DD} DC variations as measured at the receiver. Peak-to-peak noise on MV_{REF} may not exceed ±2% of the DC value.
3. V_{TT} is not applied directly to the device. It is the supply to which far end signal termination is made and is expected to be equal to MV_{REF}. This rail should track variations in the DC level of MV_{REF}.
4. Output leakage is measured with all outputs disabled, 0 V ≤ V_{OUT} ≤ GV_{DD}.

This table provides the DDR capacitance when GV_{DD}(typ) = 2.5 V.

Table 15. DDR SDRAM Capacitance for GV_{DD}(typ) = 2.5 V

Parameter/Condition	Symbol	Min	Max	Unit	Note
Input/output capacitance: DQ, DQS	C_{IO}	6	8	pF	1
Delta input/output capacitance: DQ, DQS	C_{DIO}	—	0.5	pF	1

Note:
1. This parameter is sampled. GV_{DD} = 2.5 V ± 0.125 V, f = 1 MHz, T_A = 25°C, V_{OUT} = GV_{DD}/2, V_{OUT} (peak-to-peak) = 0.2 V.

This table provides the current draw characteristics for MV_{REF}.

Table 16. Current Draw Characteristics for MV_{REF}

Parameter/Condition	Symbol	Min	Max	Unit	Note
Current draw for MV_{REF}	I_{MVREF}	—	500	μA	1

Note:
1. The voltage regulator for MV_{REF} must be able to supply up to 500 μA current.

6.2 DDR and DDR2 SDRAM AC Electrical Characteristics

This section provides the AC electrical characteristics for the DDR SDRAM interface.

6.2.1 DDR and DDR2 SDRAM Input AC Timing Specifications

This table provides the input AC timing specifications for the DDR2 SDRAM when GV_{DD}(typ) = 1.8 V.

Table 17. DDR2 SDRAM Input AC Timing Specifications for 1.8-V Interface

At recommended operating conditions with GV_{DD} of 1.8 ± 5%.

Parameter	Symbol	Min	Max	Unit	Note
AC input low voltage	V_{IL}	—	MV_{REF} – 0.25	V	—
AC input high voltage	V_{IH}	MV_{REF} + 0.25	—	V	—

8.1 Enhanced Three-Speed Ethernet Controller (eTSEC) (10/100/1000 Mbps)—MII/RMII/RGMII/SGMII/RTBI Electrical Characteristics

The electrical characteristics specified here apply to all the media independent interface (MII), reduced gigabit media independent interface (RGMII), serial gigabit media independent interface (SGMII), and reduced ten-bit interface (RTBI) signals except management data input/output (MDIO) and management data clock (MDC). The RGMII and RTBI interfaces are defined for 2.5 V, while the MII interface can be operated at 3.3 V. The RMII and SGMII interfaces can be operated at either 3.3 or 2.5 V. The RGMII and RTBI interfaces follow the Hewlett-Packard reduced pin-count interface for *Gigabit Ethernet Physical Layer Device Specification Version 1.2*a (9/22/2000). The electrical characteristics for MDIO and MDC are specified in Section 8.5, "Ethernet Management Interface Electrical Characteristics."

8.1.1 TSEC DC Electrical Characteristics

All RGMII, RMII, and RTBI drivers and receivers comply with the DC parametric attributes specified in Table 24 and Table 25. The RGMII and RTBI signals are based on a 2.5-V CMOS interface voltage as defined by JEDEC EIA/JESD8-5.

NOTE

eTSEC should be interfaced with peripheral operating at same voltage level.

Table 24. MII DC Electrical Characteristics

Parameter	Symbol	Conditions		Min	Max	Unit
Supply voltage 3.3 V	LV_{DDA}/LV_{DDB}	—		2.97	3.63	V
Output high voltage	V_{OH}	I_{OH} = –4.0 mA	LV_{DDA} or LV_{DDB} = Min	2.40	LV_{DDA} + 0.3 or LV_{DDB} + 0.3	V
Output low voltage	V_{OL}	I_{OL} = 4.0 mA	LV_{DDA} or LV_{DDB} = Min	V_{SS}	0.50	V
Input high voltage	V_{IH}	—	—	2.0	LV_{DDA} + 0.3 or LV_{DDB} + 0.3	V
Input low voltage	V_{IL}	—	—	–0.3	0.90	V
Input high current	I_{IH}	V_{IN}[1] = LV_{DDA} or LV_{DDB}		—	40	μA
Input low current	I_{IL}	V_{IN}[1] = VSS		–600	—	μA

Note:

1. The symbol V_{IN}, in this case, represents the LV_{IN} symbol referenced in Table 1 and Table 2.

Table 25. RGMII/RTBI DC Electrical Characteristics

Parameters	Symbol	Conditions	Min	Max	Unit
Supply voltage 2.5 V	LV_{DDA}/LV_{DDB}	—	2.37	2.63	V

This figure shows the MII transmit AC timing diagram.

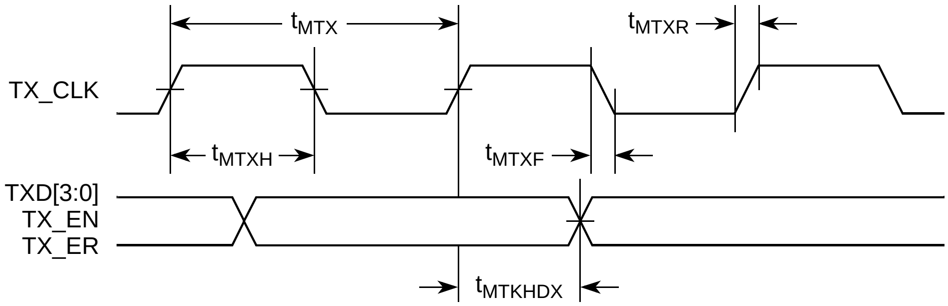


Figure 8. MII Transmit AC Timing Diagram

8.2.1.2 MII Receive AC Timing Specifications

This table provides the MII receive AC timing specifications.

Table 27. MII Receive AC Timing Specifications

At recommended operating conditions with $LV_{DDA}/LV_{DDB}/NV_{DD}$ of 3.3 V ± 0.3 V.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit
RX_CLK clock period 10 Mbps	t_{MRX}	—	400	—	ns
RX_CLK clock period 100 Mbps	t_{MRX}	—	40	—	ns
RX_CLK duty cycle	t_{MRXH}/t_{MRX}	35	—	65	%
RXD[3:0], RX_DV, RX_ER setup time to RX_CLK	t_{MRDVKH}	10.0	—	—	ns
RXD[3:0], RX_DV, RX_ER hold time to RX_CLK	t_{MRDXKH}	10.0	—	—	ns
RX_CLK clock rise V_{IL}(min) to V_{IH}(max)	t_{MRXR}	1.0	—	4.0	ns
RX_CLK clock fall time V_{IH}(max) to V_{IL}(min)	t_{MRXF}	1.0	—	4.0	ns

Note:

1. The symbols used for timing specifications follow the pattern of $t_{(\text{first two letters of functional block})(\text{signal})(\text{state})(\text{reference})(\text{state})}$ for inputs and $t_{(\text{first two letters of functional block})(\text{reference})(\text{state})(\text{signal})(\text{state})}$ for outputs. For example, t_{MRDVKH} symbolizes MII receive timing (MR) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{MRX} clock reference (K) going to the high (H) state or setup time. Also, t_{MRDXKL} symbolizes MII receive timing (GR) with respect to the time data input signals (D) went invalid (X) relative to the t_{MRX} clock reference (K) going to the low (L) state or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{MRX} represents the MII (M) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. The frequency of RX_CLK should not exceed the TX_CLK by more than 300 ppm

This figure provides the AC test load for TSEC.

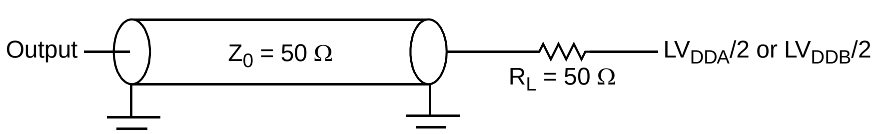


Figure 9. TSEC AC Test Load

8.2.1.4 RMII Receive AC Timing Specifications

This table provides the RMII receive AC timing specifications.

Table 29. RMII Receive AC Timing Specifications

At recommended operating conditions with NV_{DD} of 3.3 V ± 0.3 V.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit
REF_CLK clock period	t_{RMX}	—	20	—	ns
REF_CLK duty cycle	t_{RMXH}/t_{RMX}	35	—	65	%
RXD[1:0], CRS_DV, RX_ER setup time to REF_CLK	$t_{RMRDVKH}$	4.0	—	—	ns
RXD[1:0], CRS_DV, RX_ER hold time to REF_CLK	$t_{RMRDXKH}$	2.0	—	—	ns
REF_CLK clock rise V_{IL}(min) to V_{IH}(max)	t_{RMXR}	1.0	—	4.0	ns
REF_CLK clock fall time V_{IH}(max) to V_{IL}(min)	t_{RMXF}	1.0	—	4.0	ns

Note:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first three letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{RMRDVKH}$ symbolizes RMII receive timing (RMR) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{RMX} clock reference (K) going to the high (H) state or setup time. Also, $t_{RMRDXKL}$ symbolizes RMII receive timing (RMR) with respect to the time data input signals (D) went invalid (X) relative to the t_{RMX} clock reference (K) going to the low (L) state or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{RMX} represents the RMII (RM) reference (X) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).

This table provides the AC test load.

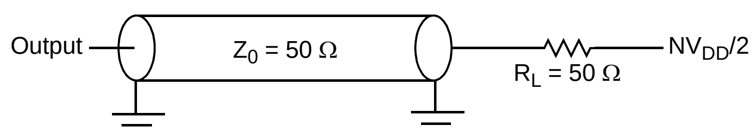


Figure 12. AC Test Load

This table shows the RMII receive AC timing diagram.

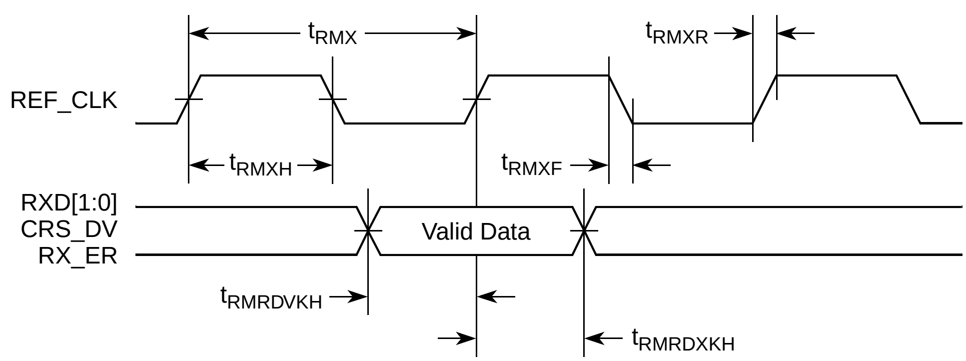


Figure 13. RMII Receive AC Timing Diagram

This figure shows the RGMII and RTBI AC timing and multiplexing diagrams.

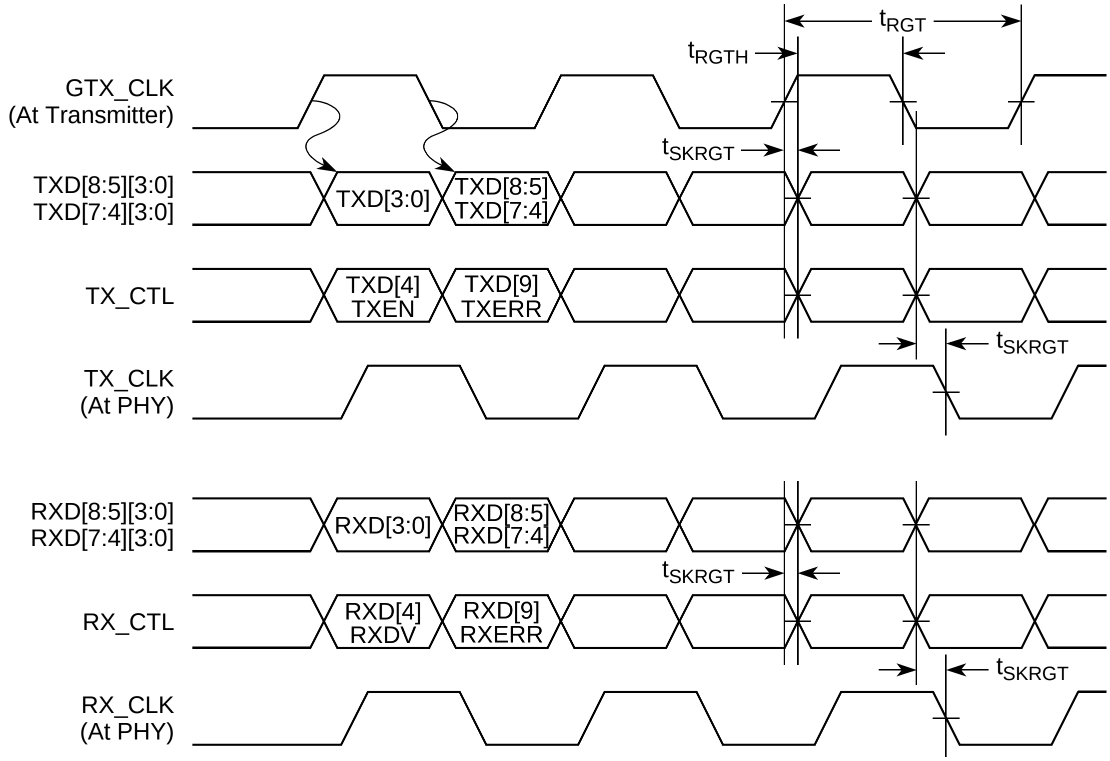


Figure 14. RGMII and RTBI AC Timing and Multiplexing Diagrams

8.3 SGMII Interface Electrical Characteristics

Each SGMII port features a 4-wire AC-coupled serial link from the dedicated SerDes interface of MPC8313E as shown in Figure 15, where C_{TX} is the external (on board) AC-coupled capacitor. Each output pin of the SerDes transmitter differential pair features a 50-Ω output impedance. Each input of the SerDes receiver differential pair features 50-Ω on-die termination to XCOREVSS. The reference circuit of the SerDes transmitter and receiver is shown in Figure 33.

When an eTSEC port is configured to operate in SGMII mode, the parallel interface’s output signals of this eTSEC port can be left floating. The input signals should be terminated based on the guidelines described in Section 22.5, “Connection Recommendations,” as long as such termination does not violate the desired POR configuration requirement on these pins, if applicable.

When operating in SGMII mode, the TSEC_GTX_CLK125 clock is not required for this port. Instead, the SerDes reference clock is required on SD_REF_CLK and $\overline{\text{SD\_REF\_CLK}}$ pins.

8.3.1 DC Requirements for SGMII SD_REF_CLK and $\overline{\text{SD\_REF\_CLK}}$

The characteristics and DC requirements of the separate SerDes reference clock are described in Section 9, “High-Speed Serial Interfaces (HSSI).”

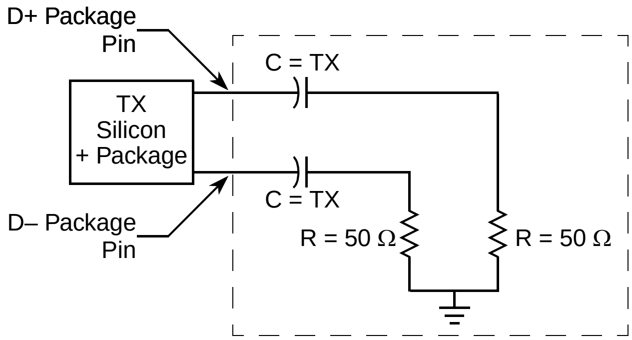
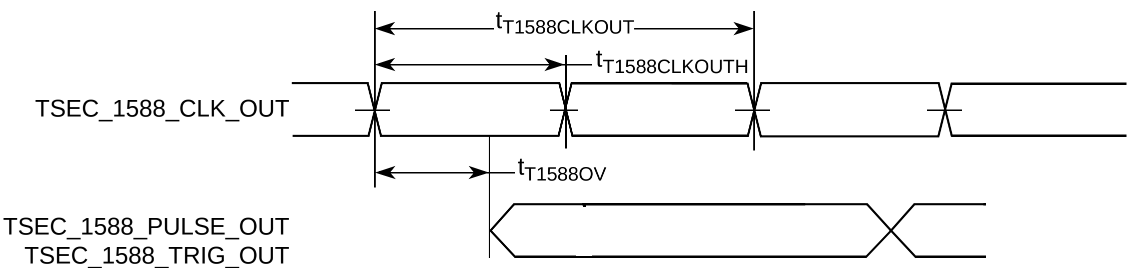


Figure 18. SGMII AC Test/Measurement Load

8.4 eTSEC IEEE 1588 AC Specifications

This figure provides the data and command output timing diagram.



Note: The output delay is count starting rising edge if $t_{T1588CLKOUT}$ is non-inverting. Otherwise, it is count starting falling edge.

Figure 19. eTSEC IEEE 1588 Output AC Timing

This figure provides the data and command input timing diagram.

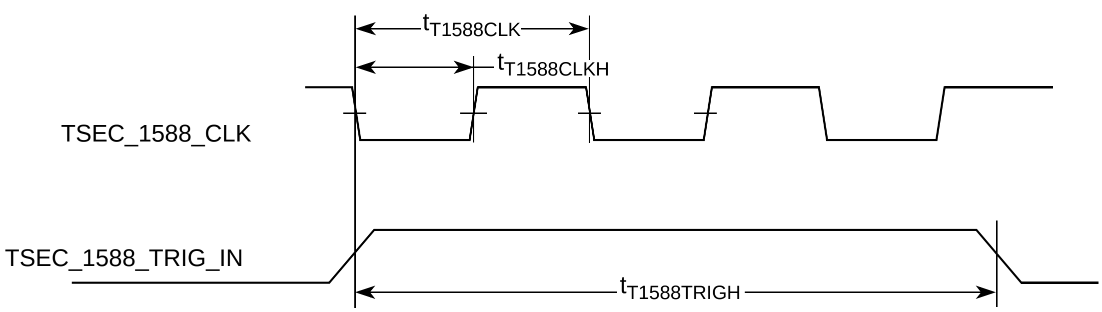


Figure 20. eTSEC IEEE 1588 Input AC Timing

This table lists the IEEE 1588 AC timing specifications.

Table 36. eTSEC IEEE 1588 AC Timing Specifications

At recommended operating conditions with L/TV_{DD} of 3.3 V ± 5%.

Parameter/Condition	Symbol	Min	Typ	Max	Unit	Note
TSEC_1588_CLK clock period	$t_{T1588CLK}$	3.8	—	$T_{RX\_CLK} \times 9$	ns	1, 3
TSEC_1588_CLK duty cycle	$t_{T1588CLKH}/t_{T1588CLK}$	40	50	60	%	

Note:

1. Note that the symbol V_{IN}, in this case, represents the NV_{IN} symbol referenced in Table 1 and Table 2.

8.5.2 MII Management AC Electrical Specifications

This table provides the MII management AC timing specifications.

Table 38. MII Management AC Timing Specifications

At recommended operating conditions with NV_{DD} is 3.3 V ± 0.3V

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit	Note
MDC frequency	f_{MDC}	—	2.5	—	MHz	2
MDC period	t_{MDC}	—	400	—	ns	
MDC clock pulse width high	t_{MDCH}	32	—	—	ns	
MDC to MDIO delay	t_{MDKHDX}	10	—	170	ns	
MDIO to MDC setup time	t_{MDDVKH}	5	—	—	ns	
MDIO to MDC hold time	t_{MDDXKH}	0	—	—	ns	
MDC rise time	t_{MDCR}	—	—	10	ns	
MDC fall time	t_{MDHF}	—	—	10	ns	

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{MDKHDX} symbolizes management data timing (MD) for the time t_{MDC} from clock reference (K) high (H) until data outputs (D) are invalid (X) or data hold time. Also, t_{MDDVKH} symbolizes management data timing (MD) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{MDC} clock reference (K) going to the high (H) state or setup time. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. This parameter is dependent on the csb_clk speed. (The MIIMCFG[Mgmt Clock Select] field determines the clock frequency of the Mgmt Clock EC_MDC.)

This figure shows the MII management AC timing diagram.

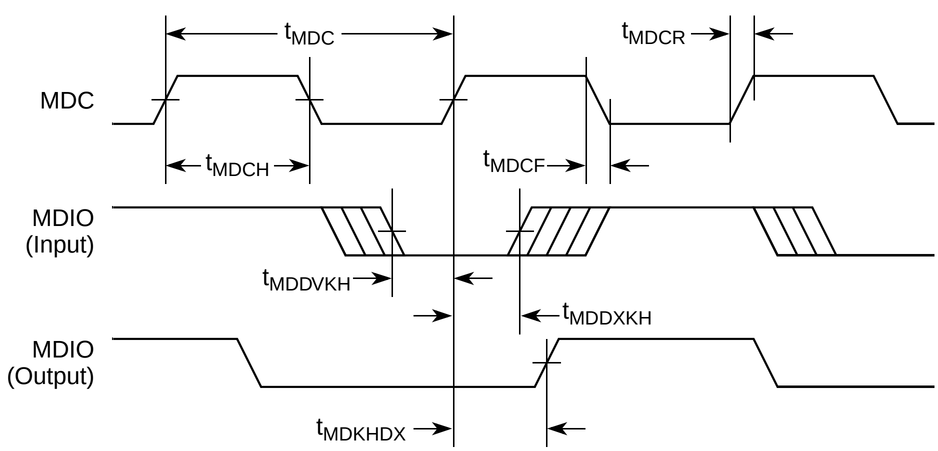


Figure 21. MII Management Interface Timing Diagram

— The SD_REF_CLK and $\overline{\text{SD\_REF\_CLK}}$ are internally AC-coupled differential inputs as shown in Figure 23. Each differential clock input (SD_REF_CLK or $\overline{\text{SD\_REF\_CLK}}$) has a 50-Ω termination to $XCOREV_{SS}$ followed by on-chip AC coupling.

— The external reference clock driver must be able to drive this termination.

— The SerDes reference clock input can be either differential or single-ended. Refer to the differential mode and single-ended mode description below for further detailed requirements.

- The maximum average current requirement that also determines the common mode voltage range:

— When the SerDes reference clock differential inputs are DC coupled externally with the clock driver chip, the maximum average current allowed for each input pin is 8 mA. In this case, the exact common mode input voltage is not critical as long as it is within the range allowed by the maximum average current of 8 mA (refer to the following bullet for more detail), since the input is AC-coupled on-chip.

— This current limitation sets the maximum common mode input voltage to be less than 0.4 V (0.4 V/50 = 8 mA) while the minimum common mode input level is 0.1 V above $XCOREV_{SS}$. For example, a clock with a 50/50 duty cycle can be produced by a clock driver with output driven by its current source from 0 to 16 mA (0–0.8 V), such that each phase of the differential input has a single-ended swing from 0 V to 800 mV with the common mode voltage at 400 mV.

— If the device driving the SD_REF_CLK and $\overline{\text{SD\_REF\_CLK}}$ inputs cannot drive 50 Ω to $XCOREV_{SS}$ DC, or it exceeds the maximum input current limitations, then it must be AC-coupled off-chip.

- The input amplitude requirement. This requirement is described in detail in the following sections.

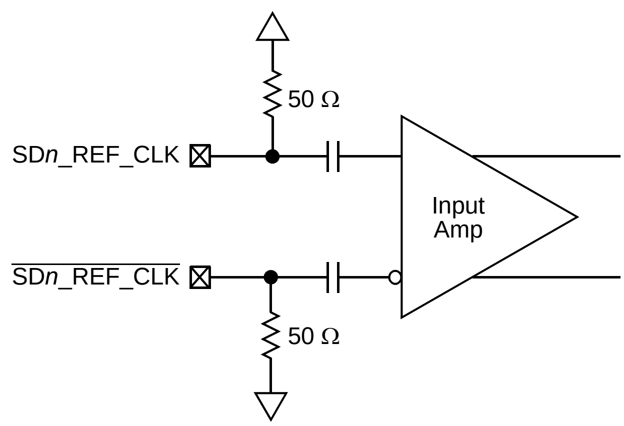


Figure 23. Receiver of SerDes Reference Clocks

9.2.2 DC Level Requirement for SerDes Reference Clocks

The DC level requirement for the MPC8313E SerDes reference clock inputs is different depending on the signaling mode used to connect the clock driver chip and SerDes reference clock inputs as described below.

- Differential mode

— The input amplitude of the differential clock must be between 400 and 1600 mV differential peak-to-peak (or between 200 and 800 mV differential peak). In other words, each signal wire of the differential pair must have a single-ended swing less than 800 mV and greater than 200 mV. This requirement is the same for both external DC-coupled or AC-coupled connection.

— For external DC-coupled connection, as described in Section 9.2.1, "SerDes Reference Clock Receiver Characteristics," the maximum average current requirements sets the requirement for average voltage (common mode voltage) to be between 100 and 400 mV. Figure 24 shows the SerDes reference clock input requirement for the DC-coupled connection scheme.

— For external AC-coupled connection, there is no common mode voltage requirement for the clock driver. Since the external AC-coupling capacitor blocks the DC level, the clock driver and the SerDes reference clock receiver operate in different command mode voltages. The SerDes reference clock receiver in this connection scheme has its common mode voltage set to $XCOREV_{SS}$. Each signal wire of the differential inputs is allowed to swing below and above the command mode voltage ($XCOREV_{SS}$). Figure 25 shows the SerDes reference clock input requirement for AC-coupled connection scheme.

- Single-ended mode
  - — The reference clock can also be single-ended. The SD_REF_CLK input amplitude (single-ended swing) must be between 400 and 800 mV peak-to-peak (from V_{min} to V_{max}) with $\overline{\text{SD\_REF\_CLK}}$ either left unconnected or tied to ground.
  - — The SD_REF_CLK input average voltage must be between 200 and 400 mV. Figure 26 shows the SerDes reference clock input requirement for the single-ended signaling mode.
  - — To meet the input amplitude requirement, the reference clock inputs might need to be DC or AC coupled externally. For the best noise performance, the reference of the clock could be DC or AC coupled into the unused phase ($\overline{\text{SD\_REF\_CLK}}$) through the same source impedance as the clock input (SD_REF_CLK) in use.

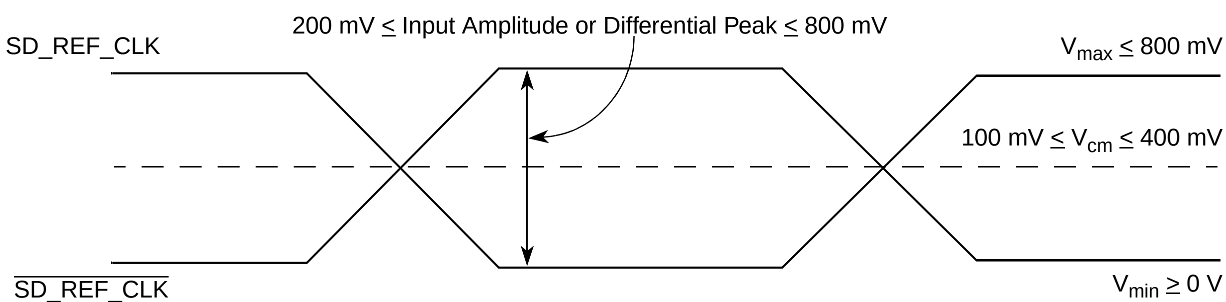


Figure 24. Differential Reference Clock Input DC Requirements (External DC-Coupled)

Table 62. MPC8313E TEPBGAII Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Note
RXB	R1	I		—
$\overline{\mathrm{RXB}}$	P1	I		—
SD_IMP_CAL_RX	V5	I		200 Ω ± 10% to GND
$\overline{\mathrm{SD\_REF\_CLK}}$	T5	I		—
SD_REF_CLK	T4	I		—
SD_PLL_TPD	T2	O		—
SD_IMP_CAL_TX	N5	I		100 Ω ± 10% to GND
SDAVDD	R5	I/O		—
SD_PLL_TPA_ANA	R4	O		—
SDAVSS	R3	I/O		—
USB PHY				
USB_DP	P26	I/O		—
USB_DM	N26	I/O		—
USB_VBUS	P24	I/O		—
USB_TPA	L26	I/O		—
USB_RBIAS	M24	I/O		—
USB_PLL_PWR3	M26	I/O		—
USB_PLL_GND	N24	I/O		—
USB_PLL_PWR1	N25	I/O		—
USB_VSSA_BIAS	M25	I/O		—
USB_VDDA_BIAS	M22	I/O		—
USB_VSSA	N22	I/O		—
USB_VDDA	P22	I/O		—
GTM/USB				
USBDR_DRIVE_VBUS/GTM1_TIN1/GTM2_TIN2/LSRCID0	AD23	I/O	NV_{DD}	—
USBDR_PWRFAULT/$\overline{\mathrm{GTM1\_TGATE1}}$/$\overline{\mathrm{GTM2\_TGATE2}}$/LSRCID1	AE23	I/O	NV_{DD}	—
USBDR_PCTL0/$\overline{\mathrm{GTM1\_TOUT1}}$/LSRCID2	AC22	O	NV_{DD}	—
USBDR_PCTL1/LBC_PM_REF_10/LSRCID3	AB21	O	NV_{DD}	—

Table 62. MPC8313E TEPBGAII Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Note
SPI				
SPIMOSI/GTM1_TIN3/GTM2_TIN4/GPIO28/LSRCID4	H1	I/O	NV_{DD}	—
SPIMISO/$\overline{GTM1\_TGATE3}$/$\overline{GTM2\_TGATE4}$/GPIO29/ LDVAL	H3	I/O	NV_{DD}	—
SPICLK/$\overline{GTM1\_TOUT3}$/GPIO30	G1	I/O	NV_{DD}	—
SPISEL/GPIO31	G3	I/O	NV_{DD}	—
Power and Ground Supplies				
AV_{DD1}	F14	Power for e300 core APLL (1.0 V)	—	—
AV_{DD2}	P21	Power for system APLL (1.0 V)	—	—
GV_{DD}	A2,A3,A4,A24,A25,B3, B4,B5,B12,B13,B20,B21, B24,B25,B26,D1,D2,D8, D9,D16,D17	Power for DDR1 and DDR2 DRAM I/O voltage (1.8/2.5 V)	—	—
LV_{DD}	D24,D25,G23,H23,R23, T23,W25,Y25,AA22,AC23	Power for local bus (3.3 V)	—	—
LV_{DDA}	W2,Y2	Power for eTSEC2 (2.5 V, 3.3 V)	—	—
LV_{DDB}	AC8,AC9,AE4,AE5	Power for eTSEC1/ USB DR (2.5 V, 3.3 V)	—	—
MV_{REF}	C14,D14	Reference voltage signal for DDR	—	—
NV_{DD}	G4,H4,L2,M2,AC16,AC17, AD25,AD26,AE12,AE13, AE20,AE21,AE24,AE25, AE26,AF24,AF25	Standard I/O voltage (3.3 V)	—	—
V_{DD}	K11,K12,K13,K14,K15, K16,L10,L17,M10,M17, N10,N17,U12,U13,	Power for core (1.0 V)	—	—
V_{DDC}	F6,F10,F19,K6,K10,K17, K21,P6,P10,P17,R10,R17, T10,T17,U10,U11,U14, U15,U16,U17,W6,W21, AA6,AA10,AA14,AA19	Internal core logic constant power (1.0 V)	—	—

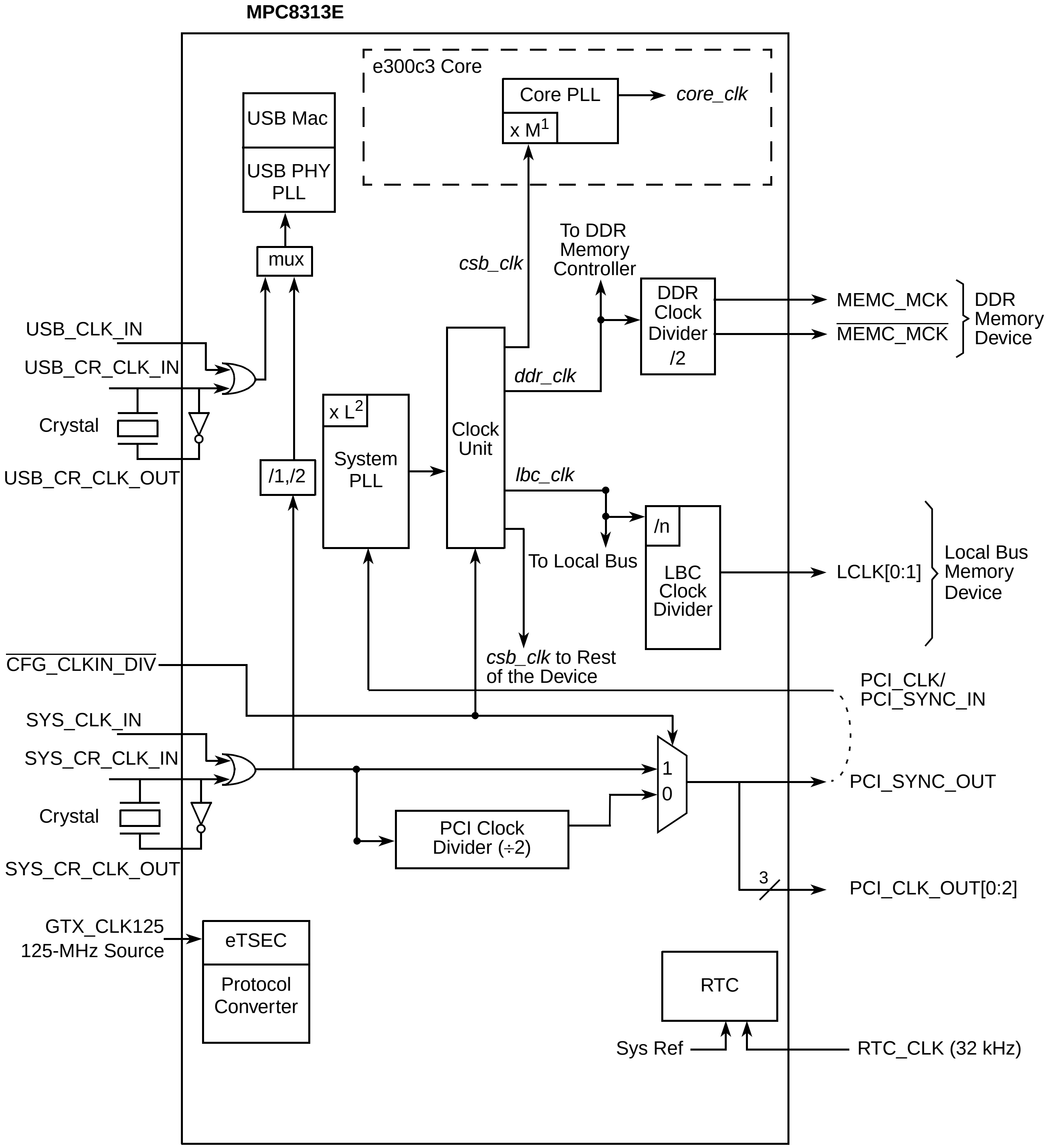
Table 62. MPC8313E TEPBGAII Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Note
V_{SS}	B1,B2,B8,B9,B16,B17,C1, C2,C3,C4,C5,C24,C25, C26,D3,D4,D12,D13,D20, D21,F8,F11,F13,F16,F17, F21,G2,G25,H2,H6,H21, H25,L4,L6,L11,L12,L13, L14,L15,L16,L21,L23,M4, M11,M12,M13,M14,M15, M16,M23,N6,N11,N12, N13,N14,N15,N16, N21,N23,P11,P12,P13, P14,P15,P16,P23,P25, R11,R12,R13,R14,R15, R16,R25,T6,T11,T12,T13, T14,T15,T16,T21,T25,U5, U6,U21,W4,W23,Y4,Y23, AA8,AA11,AA13,AA16, AA17,AA21,AC4,AC5, AC12,AC13,AC20,AC21, AD1,AE2,AE8,AE9,AE16, AE17,AF2	—	—	—
$XCOREV_{DD}$	T1,U2,V2	Core power for SerDes transceivers (1.0 V)	—	—
$XCOREV_{SS}$	P2,R2,T3	—	—	—
$XPADV_{DD}$	P5,U4	Pad power for SerDes transceivers (1.0 V)	—	—
$XPADV_{SS}$	P3,V4	—	—	—

Notes:
1. This pin is an open drain signal. A weak pull-up resistor (1 kΩ) should be placed on this pin to NV_{DD}.
2. This pin is an open drain signal. A weak pull-up resistor (2–10 kΩ) should be placed on this pin to NV_{DD}.
3. This output is actively driven during reset rather than being three-stated during reset.
4. These JTAG pins have weak internal pull-up P-FETs that are always enabled.
5. This pin should have a weak pull up if the chip is in PCI host mode. Follow PCI specifications recommendation.
6. This pin must always be tied to V_{SS}.
7. Internal thermally sensitive resistor, resistor value varies linearly with temperature. Useful for determining the junction temperature.
8. 1588 signals are available on these pins only in MPC8313 Rev 2.x or later.
9. LB_POR_CFG_BOOT_ECC_DIS is available only in MPC8313 Rev 2.x or later.
10. This pin has an internal pull-up.
11. This pin has an internal pull-down.
12. In MII mode, GTX_CLK should be pulled down by 300Ω to V_{SS}.

20 Clocking

This figure shows the internal distribution of clocks within the MPC8313E.



[1] Multiplication factor M = 1, 1.5, 2, 2.5, and 3. Value is decided by RCWLR[COREPLL].

[2] Multiplication factor L = 2, 3, 4, 5, and 6. Value is decided by RCWLR[SPMF].

Figure 57. MPC8313E Clock Subsystem

20.3 Example Clock Frequency Combinations

This table shows several possible frequency combinations that can be selected based on the indicated input reference frequencies, with RCWLR[LBCM] = 0 and RCWLR[DDRCM] =1, such that the LBC operates with a frequency equal to the frequency of *csb_clk* and the DDR controller operates at twice the frequency of *csb_clk*.

Table 68. System Clock Frequencies

						LBC(lbc_clk)				e300 Core(core_clk)				
SYS_ CLK_IN/ PCI_CLK	SPMF[1]	VCOD[2]	VCO[3]	CSB (*csb_clk*)[4]	DDR (ddr_clk)	/2	/4	/8	USB ref[5]	× 1	× 1.5	× 2	× 2.5	× 3
25.0	6	2	600.0	150.0	300.0	—	37.5	18.8	Note[6]	150.0	225	300	375	—
25.0	5	2	500.0	125.0	250.0	62.5	31.25	15.6	Note 6	125.0	188	250	313	375
33.3	5	2	666.0	166.5	333.0	—	41.63	20.8	Note 6	166.5	250	333	—	—
33.3	4	2	532.8	133.2	266.4	66.6	33.3	16.7	Note 6	133.2	200	266	333	400
48.0	3	2	576.0	144.0	288.0	—	36	18.0	48.0	144.0	216	288	360	—
66.7	2	2	533.4	133.3	266.7	66.7	33.34	16.7	Note 6	133.3	200	267	333	400

Note:

1. System PLL multiplication factor.
2. System PLL VCO divider.
3. When considering operating frequencies, the valid core VCO operating range of 400–800 MHz must not be violated.
4. Due to erratum eTSEC40, *csb_clk* frequencies of less than 133 MHz do not support gigabit Ethernet data rates. The core frequency must be 333 MHz for gigabit Ethernet operation. This erratum will be fixed in revision 2 silicon.
5. Frequency of USB PLL input reference.
6. USB reference clock must be supplied from a separate source as it must be 24 or 48 MHz, the USB reference must be supplied from a separate external source using USB_CLK_IN.

21 Thermal

This section describes the thermal specifications of the MPC8313E.

21.1 Thermal Characteristics

This table provides the package thermal characteristics for the 516, 27 × 27 mm TEPBGAII.

Table 69. Package Thermal Characteristics for TEPBGAII

Characteristic	Board Type	Symbol	TEPBGA II	Unit	Note
Junction-to-ambient natural convection	Single layer board (1s)	$R_{\theta JA}$	25	°C/W	1, 2
Junction-to-ambient natural convection	Four layer board (2s2p)	$R_{\theta JA}$	18	°C/W	1, 2, 3
Junction-to-ambient (@200 ft/min)	Single layer board (1s)	$R_{\theta JMA}$	20	°C/W	1, 3
Junction-to-ambient (@200 ft/min)	Four layer board (2s2p)	$R_{\theta JMA}$	15	°C/W	1, 3
Junction-to-board	—	$R_{\theta JB}$	10	°C/W	4

21.3 Heat Sink Attachment

When attaching heat sinks to these devices, an interface material is required. The best method is to use thermal grease and a spring clip. The spring clip should connect to the printed-circuit board, either to the board itself, to hooks soldered to the board, or to a plastic stiffener. Avoid attachment forces which would lift the edge of the package or peel the package from the board. Such peeling forces reduce the solder joint lifetime of the package. Recommended maximum force on the top of the package is 10 lb (4.5 kg) force. If an adhesive attachment is planned, the adhesive should be intended for attachment to painted or plastic surfaces and its performance verified under the application requirements.

21.3.1 Experimental Determination of the Junction Temperature with a Heat Sink

When heat sink is used, the junction temperature is determined from a thermocouple inserted at the interface between the case of the package and the interface material. A clearance slot or hole is normally required in the heat sink. Minimizing the size of the clearance is important to minimize the change in thermal performance caused by removing part of the thermal interface to the heat sink. Because of the experimental difficulties with this technique, many engineers measure the heat sink temperature and then back calculate the case temperature using a separate measurement of the thermal resistance of the interface. From this case temperature, the junction temperature is determined from the junction to case thermal resistance.

$$T_J = T_C + (R_{\theta JC} \times P_D)$$

where:

T_J = junction temperature (°C)
T_C = case temperature of the package
$R_{\theta JC}$ = junction-to-case thermal resistance
P_D = power dissipation

22 System Design Information

This section provides electrical and thermal design recommendations for successful application of the MPC8313E SYS_CLK_IN

22.1 System Clocking

The MPC8313E includes three PLLs.

1. The platform PLL (AV_{DD2}) generates the platform clock from the externally supplied SYS_CLK_IN input in PCI host mode or SYS_CLK_IN/PCI_SYNC_IN in PCI agent mode. The frequency ratio between the platform and SYS_CLK_IN is selected using the platform PLL ratio configuration bits as described in Section 20.1, "System PLL Configuration."
2. The e300 core PLL (AV_{DD1}) generates the core clock as a slave to the platform clock. The frequency ratio between the e300 core clock and the platform clock is selected using the e300 PLL ratio configuration bits as described in Section 20.2, "Core PLL Configuration."
3. There is a PLL for the SerDes block.

22.2 PLL Power Supply Filtering

Each of the PLLs listed above is provided with power through independent power supply pins (AV_{DD1}, AV_{DD2}, and $SDAV_{DD}$, respectively). The AV_{DD} level should always be equivalent to V_{DD}, and preferably these voltages are derived directly from V_{DD} through a low frequency filter scheme such as the following.

There are a number of ways to reliably provide power to the PLLs, but the recommended solution is to provide independent filter circuits as illustrated in Figure 58, one to each of the five AV_{DD} pins. By providing independent filters to each PLL the opportunity to cause noise injection from one PLL to the other is reduced.

This circuit is intended to filter noise in the PLLs resonant frequency range from a 500 kHz to 10 MHz range. It should be built with surface mount capacitors with minimum effective series inductance (ESL). Consistent with the recommendations of Dr. Howard Johnson in *High Speed Digital Design: A Handbook of Black Magic* (Prentice Hall, 1993), multiple small capacitors of equal value are recommended over a single large value capacitor.

Each circuit should be placed as close as possible to the specific AV_{DD} pin being supplied to minimize noise coupled from nearby circuits. It should be possible to route directly from the capacitors to the AV_{DD} pin, which is on the periphery of package, without the inductance of vias.

This figure shows the PLL power supply filter circuits.

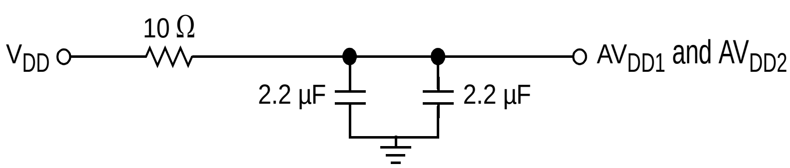
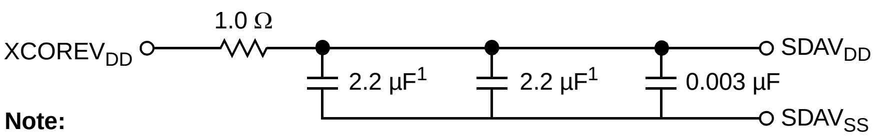


Figure 58. PLL Power Supply Filter Circuit

The $SDAV_{DD}$ signal provides power for the analog portions of the SerDes PLL. To ensure stability of the internal clock, the power supplied to the PLL is filtered using a circuit like the one shown in Figure 59. For maximum effectiveness, the filter circuit should be placed as closely as possible to the $SDAV_{DD}$ ball to ensure it filters out as much noise as possible. The ground connection should be near the $SDAV_{DD}$ ball. The 0.003-µF capacitor is closest to the ball, followed by the two 2.2-µF capacitors, and finally the 1-Ω resistor to the board supply plane. The capacitors are connected from traces from $SDAV_{DD}$ to the ground plane. Use ceramic chip capacitors with the highest possible self-resonant frequency. All traces should be kept short, wide, and direct.



Note:

1. An 0805 sized capacitor is recommended for system initial bring-up.

Figure 59. SerDes PLL Power Supply Filter Circuit

Note the following:

- $SDAV_{DD}$ should be a filtered version of $XCOREV_{DD}$.

The COP function of these processors allows a remote computer system (typically, a PC with dedicated hardware and debugging software) to access and control the internal operations of the processor. The COP interface connects primarily through the JTAG port of the processor, with some additional status monitoring signals. The COP port requires the ability to independently assert $\overline{\text{TRST}}$ without causing $\overline{\text{PORESET}}$. If the target system has independent reset sources, such as voltage monitors, watchdog timers, power supply failures, or push-button switches, then the COP reset signals must be merged into these signals with logic.

The arrangement shown in Figure 61 allows the COP to independently assert $\overline{\text{HRESET}}$ or $\overline{\text{TRST}}$, while ensuring that the target can drive $\overline{\text{HRESET}}$ as well. If the JTAG interface and COP header are not used, $\overline{\text{TRST}}$ should be tied to $\overline{\text{PORESET}}$ so that it is asserted when the system reset signal ($\overline{\text{PORESET}}$) is asserted.

The COP header shown in Figure 61 adds many benefits—breakpoints, watchpoints, register and memory examination/modification, and other standard debugger features are possible through this interface—and can be as inexpensive as an unpopulated footprint for a header to be added when needed.

The COP interface has a standard header for connection to the target system, based on the 0.025" square-post, 0.100" centered header assembly (often called a Berg header).

There is no standardized way to number the COP header shown in Figure 61; consequently, many different pin numbers have been observed from emulator vendors. Some are numbered top-to-bottom then left-to-right, while others use left-to-right then top-to-bottom, while still others number the pins counter clockwise from pin 1 (as with an IC). Regardless of the numbering, the signal placement recommended in Figure 61 is common to all known emulators.

23.1 Part Numbers Fully Addressed by this Document

This table provides the Freescale part numbering nomenclature for the MPC8313E. Note that the individual part numbers correspond to a maximum processor core frequency. For available frequencies, contact your local Freescale sales office. In addition to the processor frequency, the part numbering scheme also includes an application modifier which may specify special application conditions. Each part number also contains a revision code which refers to the die mask revision number.

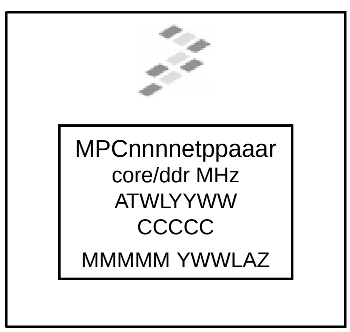
Table 72. Part Numbering Nomenclature

MPC	*nnnn*	*e*	*t*	*pp*	*aa*	*a*	*x*
Product Code	**Part Identifier**	**Encryption Acceleration**	**Temperature Range [3]**	**Package [1, 4]**	**e300 core Frequency [2]**	**DDR Frequency**	**Revision Level**
MPC	8313	Blank = Not included E = included	Blank = 0° to 105°C C= –40° to 105°C	ZQ = PB TEPBGAII VR = PB free TEPBGAII	AD = 266 MHz AF = 333 MHz AG = 400 MHz	D = 266 MHz F = 333 MHz	Blank = 1.0 A = 2.0 B = 2.1 C = 2.2

Note:
1. See Section 19, “Package and Pin Listings,” for more information on available package types.
2. Processor core frequencies supported by parts addressed by this specification only. Not all parts described in this specification support all core frequencies. Additionally, parts addressed by Part Number Specifications may support other maximum core frequencies.
3. Contact local Freescale office on availability of parts with °C temperature range.
4. ZQ package was available for Rev 1.0. For Rev 2.x, only VR package is available.

23.2 Part Marking

Parts are marked as shown in this figure.



TePBGA

Notes:
MPCnnnnetppaar is the orderable part number.
ATWLYYWW is the standard assembly, test, year, and work week codes.
CCCCC is the country code.
MMMMM is the mask number.

Figure 62. Part Marking for TEPBGAII Device

24 Revision History

This table summarizes a revision history for this document.

Table 73. Document Revision History

Rev. Number	Date	Substantive Change(s)
4	11/2011	• In Table 2, added following notes: – Note 3: Min temperature is specified with T_A; Max temperature is specified with T_J – Note 4: All Power rails must be connected and power applied to the MPC8313 even if the IP interfaces are not used. – Note 5: All I/O pins should be interfaced with peripherals operating at same voltage level. – Note 6: This voltage is the input to the filter discussed in Section 22.2, “PLL Power Supply Filtering.” and not necessarily the voltage at the AVDD pin, which may be reduced from VDD by the filter • Decoupled PCI_CLK and SYS_CLK_IN rise and fall times in Table 8. Relaxed maximum rise/fall time of SYS_CLK_IN to 4ns. • Added a note in Table 27 stating “The frequency of RX_CLK should not exceed the TX_CLK by more than 300 ppm." • In Table 30: – Changed max value of t_{skrgt} in “Data to clock input skew (at receiver)” row from 2.8 to 2.6. – Added Note 7, stating that, “The frequency of RX_CLK should not exceed the GTX_CLK125 by more than 300 ppm.” • Added a note stating “eTSEC should be interfaced with peripheral operating at same voltage level” in Section 8.1.1, “TSEC DC Electrical Characteristics.” • TSEC1_MDC and TSEC_MDIO are powered at 3.3V by NVDD. Replaced LVDDA/LVDDB with NVDD and removed instances of 2.5V at several places in Section 8.5, “Ethernet Management Interface Electrical Characteristics.” • In Table 43, changed min/max values of t_{CLK_TOL} from 0.05 to 0.005. • In Table 62: – Added Note 2 for LGPL4 in showing LGPL4 as open-drain. – Removed Note 2 from TSEC1_MDIO. – Added Note 10: This pin has an internal pull-up. – Added Note 11: This pin has an internal pull-down. – Added Note 12: “In MII mode, GTX_CLK should be pulled down by 300 Ω to V_{SS}” to TSEC1_GTX_CLK and TSEC2_GTX_CLK. • In Section 19.1, “Package Parameters for the MPC8313E TEPBGAII,” replaced "5.5 Sn/0.5 Cu/4 Ag" with "Sn/3.5 Ag." • Added foot note 3 in Table 65 stating “The VCO divider needs to be set properly so that the System PLL VCO frequency is in the range of 450–750 MHz.” • In Table 72: – Added AD = 266 and D = 266. – Added “C = 2.2” in “Revision level” column. – Added Note 4. • Changed resitor from 1.0 Ω to 10 Ω in Figure 58. • Replaced LCCR with LCRR throughout. • Added high-speed to USB Phy description.
3	01/2009	• Table 72, in column aa, changed to AG = 400 MHz.
2.2	12/2008	• Made cross-references active for sections, figures, and tables.
2.1	12/2008	• Added Figure 2, after Table 2 and renumbered the following figures.