Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	32MHz
Connectivity	I²C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	11
Program Memory Size	14KB (8K x 14)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	1K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 8x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	16-UQFN Exposed Pad
Supplier Device Package	16-UQFN (4x4)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16lf1825-e-jq

TABLE 1-2: **PIC16(L)F1825 PINOUT DESCRIPTION (CONTINUED)**

| Name	Function	Input Type	Output Type	Description
RA4/AN3/CPS3/OSC2/ CLKOUT/T1OSO/CLKR/ SDO1[1]/P2B[1]/T1G[1,2]	RA4	TTL	CMOS	General purpose I/O.
	AN3	AN	—	A/D Channel 3 input.
	CPS3	AN	—	Capacitive sensing input 3.
	OSC2	—	CMOS	Comparator C2 output.
	CLKOUT	—	CMOS	F$_{OSC}$/4 output.
	T1OSO	XTAL	XTAL	Timer1 oscillator connection.
	CLKR	—	CMOS	Clock Reference output.
	SDO1	—	CMOS	SPI data output.
	P2B	—	CMOS	PWM output.
	T1G	ST	—	Timer1 Gate input.
RA5/CLKIN/OSC1/T1OSI/ T1CKI/P2A[1]/CCP2[1]	RA5	TTL	CMOS	General purpose I/O.
	CLKIN	CMOS	—	External clock input (EC mode).
	OSC1	XTAL	—	Crystal/Resonator (LP, XT, HS modes).
	T1OSI	XTAL	XTAL	Timer1 oscillator connection.
	T1CKI	ST	—	Timer1 clock input.
	P2A	—	CMOS	PWM output.
	CCP2	ST	CMOS	Capture/Compare/PWM2.
RC0/AN4/CPS4/C2IN+/SCL/ SCK/P1D[1]	RC0	TTL	CMOS	General purpose I/O.
	AN4	AN	—	A/D Channel 4 input.
	CPS4	AN	—	Capacitive sensing input 4.
	C2IN+	AN	—	Comparator C2 positive input.
	SCL	I^{2}C	OD	I^{2}C™ clock.
	SCK	ST	CMOS	SPI clock.
	P1D	—	CMOS	PWM output.
RC1/AN5/CPS5/C12IN1-/SDA/ SDI/P1C[1]/CCP4	RC1	TTL	CMOS	General purpose I/O.
	AN5	AN	—	A/D Channel 5 input.
	CPS5	AN	—	Capacitive sensing input 5.
	C12IN1-	AN	—	Comparator C1 or C2 negative input.
	SDA	I^{2}C	OD	I^{2}C data input/output.
	SDI	CMOS	—	SPI data input.
	P1C	—	CMOS	PWM output.
	CCP4	AN	—	Capture/Compare/PWM4.
RC2/AN6/CPS6/C12IN2-/ P1D[1,2]/P2B[1,2]/SDO1[1,2]/ MDCIN1	RC2	TTL	CMOS	General purpose I/O.
	AN6	AN	—	A/D Channel 6 input.
	CPS6	AN	—	Capacitive sensing input 6.
	C12IN2-	AN	—	Comparator C1 or C2 negative input.
	P1D	—	CMOS	PWM output.
	P2B	—	CMOS	PWM output.
	SDO1	—	CMOS	SPI data output.
	MDCIN1	ST	—	Modulator Carrier Input 1.

Legend: AN = Analog input or output CMOS = CMOS compatible input or output OD = Open Drain
TTL = TTL compatible input ST = Schmitt Trigger input with CMOS levels I^{2}C™ = Schmitt Trigger input with I^{2}C
HV = High Voltage XTAL = Crystal levels
Note 1: Pin functions can be moved using the APFCON0 or APFCON1 register.
2: Default function location.

FIGURE 5-7: **INTERNAL OSCILLATOR SWITCH TIMING**
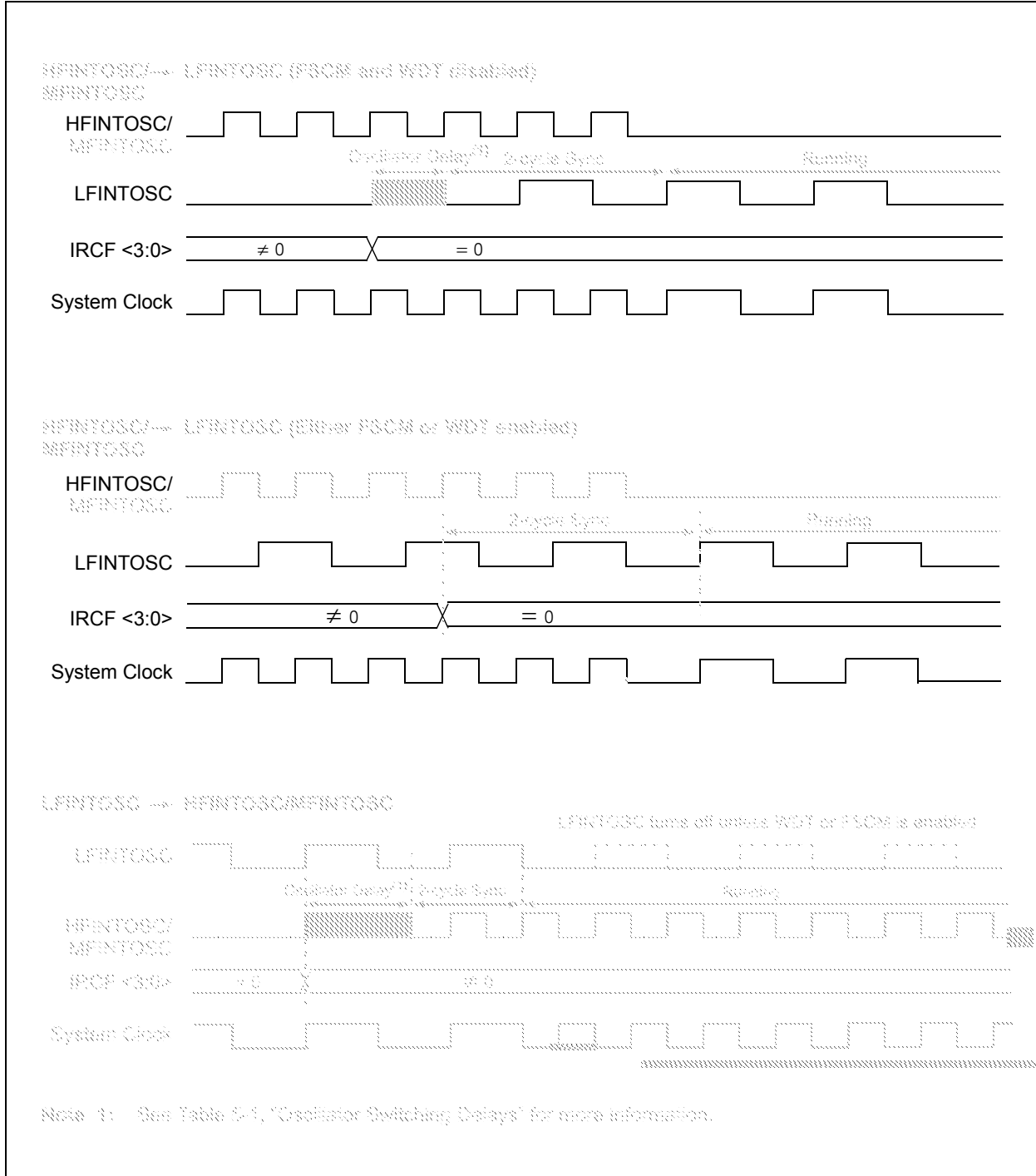
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10.6   Watchdog Control Register

REGISTER 10-1:   WDTCON: WATCHDOG TIMER CONTROL REGISTER

U-0	U-0	R/W-0/0	R/W-1/1	R/W-0/0	R/W-1/1	R/W-1/1	R/W-0/0
—	—	WDTPS<4:0>					SWDTEN
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-m/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-6   **Unimplemented:** Read as '0'

bit 5-1   **WDTPS<4:0>:** Watchdog Timer Period Select bits[1]

Bit Value = Prescale Rate

00000 = 1:32 (Interval 1 ms nominal)
00001 = 1:64 (Interval 2 ms nominal)
00010 = 1:128 (Interval 4 ms nominal)
00011 = 1:256 (Interval 8 ms nominal)
00100 = 1:512 (Interval 16 ms nominal)
00101 = 1:1024 (Interval 32 ms nominal)
00110 = 1:2048 (Interval 64 ms nominal)
00111 = 1:4096 (Interval 128 ms nominal)
01000 = 1:8192 (Interval 256 ms nominal)
01001 = 1:16384 (Interval 512 ms nominal)
01010 = 1:32768 (Interval 1s nominal)
01011 = 1:65536 (Interval 2s nominal) (Reset value)
01100 = 1:131072 (2^{17}) (Interval 4s nominal)
01101 = 1:262144 (2^{18}) (Interval 8s nominal)
01110 = 1:524288 (2^{19}) (Interval 16s nominal)
01111 = 1:1048576 (2^{20}) (Interval 32s nominal)
10000 = 1:2097152 (2^{21}) (Interval 64s nominal)
10001 = 1:4194304 (2^{22}) (Interval 128s nominal)
10010 = 1:8388608 (2^{23}) (Interval 256s nominal)

10011 = Reserved. Results in minimum interval (1:32)
•
•
•
11111 = Reserved. Results in minimum interval (1:32)

bit 0   **SWDTEN:** Software Enable/Disable for Watchdog Timer bit
If WDTE<1:0> = 00:
This bit is ignored.
If WDTE<1:0> = 01:
1 = WDT is turned on
0 = WDT is turned off
If WDTE<1:0> = 1x:
This bit is ignored.

Note 1:   Times are approximate. WDT time is based on 31 kHz LFINTOSC.

11.3.2   ERASING FLASH PROGRAM MEMORY

While executing code, program memory can only be erased by rows. To erase a row:

1. Load the EEADRH:EEADRL register pair with the address of new row to be erased.
2. Clear the CFGS bit of the EECON1 register.
3. Set the EEPGD, FREE, and WREN bits of the EECON1 register.
4. Write 55h, then AAh, to EECON2 (Flash programming unlock sequence).
5. Set control bit WR of the EECON1 register to begin the erase operation.
6. Poll the FREE bit in the EECON1 register to determine when the row erase has completed.

See Example 11-4.

After the "BSF EECON1,WR" instruction, the processor requires two cycles to set up the erase operation. The user must place two NOP instructions after the WR bit is set. The processor will halt internal operations for the typical 2 ms erase time. This is not Sleep mode as the clocks and peripherals will continue to run. After the erase cycle, the processor will resume operation with the third instruction after the EECON1 write instruction.

11.3.3   WRITING TO FLASH PROGRAM MEMORY

Program memory is programmed using the following steps:

1. Load the starting address of the word(s) to be programmed.
2. Load the write latches with data.
3. Initiate a programming operation.
4. Repeat steps 1 through 3 until all data is written.

Before writing to program memory, the word(s) to be written must be erased or previously unwritten. Program memory can only be erased one row at a time. No automatic erase occurs upon the initiation of the write.

Program memory can be written one or more words at a time. The maximum number of words written at one time is equal to the number of write latches. See Figure 11-2 for more details. The write latches are aligned to the address boundary defined by EEADRL as shown in Table 11-1. Write operations do not cross these boundaries. At the completion of a program memory write operation, the write latches are reset to contain 0x3FFF.

The following steps should be completed to load the write latches and program a block of program memory. These steps are divided into two parts. First, all write latches are loaded with data except for the last program memory location. Then, the last write latch is loaded and the programming sequence is initiated. A special unlock sequence is required to load a write latch with data or initiate a Flash programming operation. This unlock sequence should not be interrupted.

1. Set the EEPGD and WREN bits of the EECON1 register.
2. Clear the CFGS bit of the EECON1 register.
3. Set the LWLO bit of the EECON1 register. When the LWLO bit of the EECON1 register is '1', the write sequence will only load the write latches and will not initiate the write to Flash program memory.
4. Load the EEADRH:EEADRL register pair with the address of the location to be written.
5. Load the EEDATH:EEDATL register pair with the program memory data to be written.
6. Write 55h, then AAh, to EECON2, then set the WR bit of the EECON1 register (Flash programming unlock sequence). The write latch is now loaded.
7. Increment the EEADRH:EEADRL register pair to point to the next location.
8. Repeat steps 5 through 7 until all but the last write latch has been loaded.
9. Clear the LWLO bit of the EECON1 register. When the LWLO bit of the EECON1 register is '0', the write sequence will initiate the write to Flash program memory.
10. Load the EEDATH:EEDATL register pair with the program memory data to be written.
11. Write 55h, then AAh, to EECON2, then set the WR bit of the EECON1 register (Flash programming unlock sequence). The entire latch block is now written to Flash program memory.

It is not necessary to load the entire write latch block with user program data. However, the entire write latch block will be written to program memory.

An example of the complete write sequence for eight words is shown in Example 11-5. The initial address is loaded into the EEADRH:EEADRL register pair; the eight words of data are loaded using indirect addressing.

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12.0 I/O PORTS

Depending on the device selected and peripherals enabled, there are up to two ports available. In general, when a peripheral is enabled on a port pin, that pin cannot be used as a general purpose output. However, the pin can still be read.

Each port has three standard registers for its operation. These registers are:

- TRISx registers (data direction)
- PORTx registers (reads the levels on the pins of the device)
- LATx registers (output latch)

Some ports may have one or more of the following additional registers. These registers are:

- ANSELx (analog select)
- WPUx (weak pull-up)
- INLVLx (input level control)

TABLE 12-1: PORT AVAILABILITY PER DEVICE

Device	PORTA	PORTB	PORTC
PIC16(L)F1825	●		●
PIC16(L)F1829	●	●	●

The Data Latch (LATx registers) is useful for read-modify-write operations on the value that the I/O pins are driving.

A write operation to the LATx register has the same effect as a write to the corresponding PORTx register. A read of the LATx register reads of the values held in the I/O PORT latches, while a read of the PORTx register reads the actual I/O pin value.

Ports that support analog inputs have an associated ANSELx register. When an ANSEL bit is set, the digital input buffer associated with that bit is disabled. Disabling the input buffer prevents analog signal levels on the pin between a logic high and low from causing excessive current in the logic input circuitry. A simplified model of a generic I/O port, without the interfaces to other peripherals, is shown in Figure 12-1.

FIGURE 12-1: GENERIC I/O PORT OPERATION
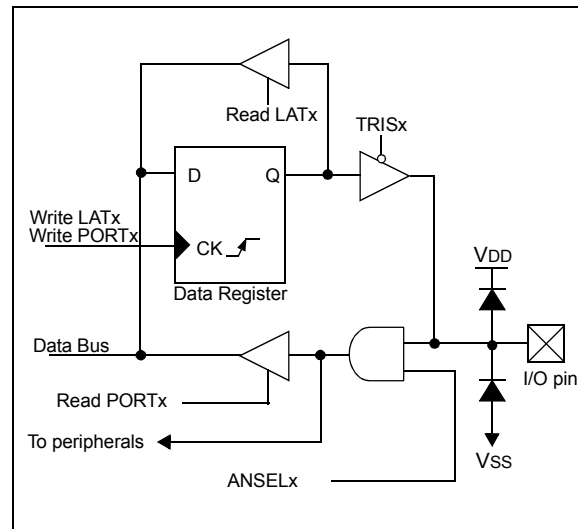
EXAMPLE 12-1: INITIALIZING PORTA

```
; This code example illustrates
; initializing the PORTA register. The
; other ports are initialized in the same
; manner.

BANKSEL  PORTA         ;
CLRF     PORTA         ;Init PORTA
BANKSEL  LATA          ;Data Latch
CLRF     LATA          ;
BANKSEL  ANSELA        ;
CLRF     ANSELA        ;digital I/O
BANKSEL  TRISA         ;
MOVLW    B'00111000'   ;Set RA<5:3> as inputs
MOVWF    TRISA         ;and set RA<2:0> as
                       ;outputs
```

16.2.6 A/D CONVERSION PROCEDURE

This is an example procedure for using the ADC to perform an Analog-to-Digital conversion:

1. Configure Port:
   - Disable pin output driver (Refer to the TRIS register)
   - Configure pin as analog (Refer to the ANSEL register)
2. Configure the ADC module:
   - Select ADC conversion clock
   - Configure voltage reference
   - Select ADC input channel
   - Turn on ADC module
3. Configure ADC interrupt (optional):
   - Clear ADC interrupt flag
   - Enable ADC interrupt
   - Enable peripheral interrupt
   - Enable global interrupt[1]
4. Wait the required acquisition time[2].
5. Start conversion by setting the GO/$\overline{\text{DONE}}$ bit.
6. Wait for ADC conversion to complete by one of the following:
   - Polling the GO/$\overline{\text{DONE}}$ bit
   - Waiting for the ADC interrupt (interrupts enabled)
7. Read ADC Result.
8. Clear the ADC interrupt flag (required if interrupt is enabled).

Note 1: The global interrupt can be disabled if the user is attempting to wake-up from Sleep and resume in-line code execution.
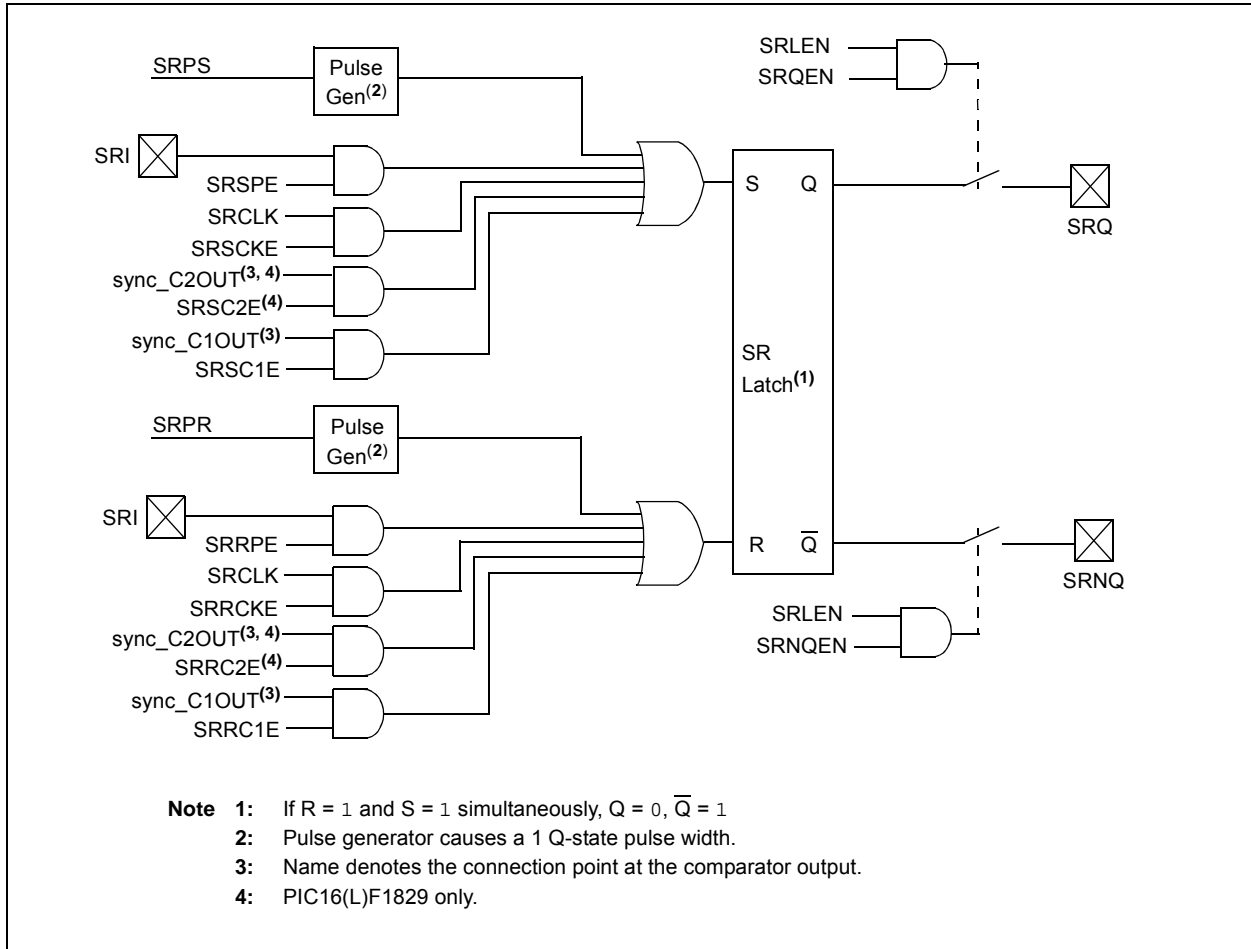
2: Refer to **Section 16.3 "A/D Acquisition Requirements"**.

EXAMPLE 16-1: A/D CONVERSION

```
;This code block configures the ADC
;for polling, Vdd and Vss references, Frc
;clock and AN0 input.
;
;Conversion start & polling for completion
; are included.
;
BANKSEL   ADCON1       ;
MOVLW     B'11110000'  ;Right justify, Frc
                       ;clock
MOVWF     ADCON1       ;Vdd and Vss Vref
BANKSEL   TRISA        ;
BSF       TRISA,0      ;Set RA0 to input
BANKSEL   ANSEL        ;
BSF       ANSEL,0      ;Set RA0 to analog
BANKSEL   ADCON0       ;
MOVLW     B'00000001'  ;Select channel AN0
MOVWF     ADCON0       ;Turn ADC On
CALL      SampleTime   ;Acquisiton delay
BSF       ADCON0,ADGO  ;Start conversion
BTFSC     ADCON0,ADGO  ;Is conversion done?
GOTO      $-1          ;No, test again
BANKSEL   ADRESH       ;
MOVF      ADRESH,W     ;Read upper 2 bits
MOVWF     RESULTHI     ;store in GPR space
BANKSEL   ADRESL       ;
MOVF      ADRESL,W     ;Read lower 8 bits
MOVWF     RESULTLO     ;Store in GPR space
```

FIGURE 18-1: SR LATCH SIMPLIFIED BLOCK DIAGRAM

Note 1: If R = 1 and S = 1 simultaneously, Q = 0, Q̄ = 1
2: Pulse generator causes a 1 Q-state pulse width.
3: Name denotes the connection point at the comparator output.
4: PIC16(L)F1829 only.

REGISTER 18-2: **SRCON1: SR LATCH CONTROL 1 REGISTER**

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
SRSPE	SRSCKE	SRSC2E[1]	SRSC1E	SRRPE	SRRCKE	SRRC2E[1]	SRRC1E
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7 **SRSPE:** SR Latch Peripheral Set Enable bit
1 = SR latch is set when the SRI pin is high
0 = SRI pin has no effect on the set input of the SR latch

bit 6 **SRSCKE:** SR Latch Set Clock Enable bit
1 = Set input of SR latch is pulsed with SRCLK
0 = SRCLK has no effect on the set input of the SR latch

bit 5 **SRSC2E:** SR Latch C2 Set Enable bit[1]
1 = SR latch is set when the C2 Comparator output is high
0 = C2 Comparator output has no effect on the set input of the SR latch

bit 4 **SRSC1E:** SR Latch C1 Set Enable bit
1 = SR latch is set when the C1 Comparator output is high
0 = C1 Comparator output has no effect on the set input of the SR latch

bit 3 **SRRPE:** SR Latch Peripheral Reset Enable bit
1 = SR latch is reset when the SRI pin is high
0 = SRI pin has no effect on the Reset input of the SR latch

bit 2 **SRRCKE:** SR Latch Reset Clock Enable bit
1 = Reset input of SR latch is pulsed with SRCLK
0 = SRCLK has no effect on the Reset input of the SR latch

bit 1 **SRRC2E:** SR Latch C2 Reset Enable bit[1]
1 = SR latch is reset when the C2 Comparator output is high
0 = C2 Comparator output has no effect on the Reset input of the SR latch

bit 0 **SRRC1E:** SR Latch C1 Reset Enable bit
1 = SR latch is reset when the C1 Comparator output is high
0 = C1 Comparator output has no effect on the Reset input of the SR latch

Note 1: PIC16(L)F1829 only.

20.1.3 SOFTWARE PROGRAMMABLE PRESCALER

A software programmable prescaler is available for exclusive use with Timer0. The prescaler is enabled by clearing the PSA bit of the OPTION_REG register.

Note:	The Watchdog Timer (WDT) uses its own independent prescaler.

There are eight prescaler options for the Timer0 module ranging from 1:2 to 1:256. The prescale values are selectable via the PS<2:0> bits of the OPTION_REG register. In order to have a 1:1 prescaler value for the Timer0 module, the prescaler must be disabled by setting the PSA bit of the OPTION_REG register.

The prescaler is not readable or writable. All instructions writing to the TMR0 register will clear the prescaler.

20.1.4 TIMER0 INTERRUPT

Timer0 will generate an interrupt when the TMR0 register overflows from FFh to 00h. The TMR0IF interrupt flag bit of the INTCON register is set every time the TMR0 register overflows, regardless of whether or not the Timer0 interrupt is enabled. The TMR0IF bit can only be cleared in software. The Timer0 interrupt enable is the TMR0IE bit of the INTCON register.

Note:	The Timer0 interrupt cannot wake the processor from Sleep since the timer is frozen during Sleep.

20.1.5 8-BIT COUNTER MODE SYNCHRONIZATION

When in 8-Bit Counter mode, the incrementing edge on the T0CKI pin must be synchronized to the instruction clock. Synchronization can be accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the instruction clock. The high and low periods of the external clocking source must meet the timing requirements as shown in **Section 30.0 "Electrical Specifications"**.

20.1.6 OPERATION DURING SLEEP

Timer0 cannot operate while the processor is in Sleep mode. The contents of the TMR0 register will remain unchanged while the processor is in Sleep mode.

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21.6.2 TIMER1 GATE SOURCE SELECTION

The Timer1 gate source can be selected from one of four different sources. Source selection is controlled by the T1GSS bits of the T1GCON register. The polarity for each available source is also selectable. Polarity selection is controlled by the T1GPOL bit of the T1GCON register.

TABLE 21-4: TIMER1 GATE SOURCES

T1GSS	Timer1 Gate Source
00	Timer1 Gate Pin
01	Overflow of Timer0 (TMR0 increments from FFh to 00h)
10	Comparator 1 Output sync_C1OUT (optionally Timer1 synchronized output)
11	Comparator 2 Output sync_C2OUT (optionally Timer1 synchronized output)

21.6.2.1 T1G Pin Gate Operation

The T1G pin is one source for Timer1 gate control. It can be used to supply an external source to the Timer1 gate circuitry.

21.6.2.2 Timer0 Overflow Gate Operation

When Timer0 increments from FFh to 00h, a low-to-high pulse will automatically be generated and internally supplied to the Timer1 gate circuitry.

21.6.2.3 Comparator C1 Gate Operation

The output resulting from a Comparator 1 operation can be selected as a source for Timer1 gate control. The Comparator 1 output (sync_C1OUT) can be synchronized to the Timer1 clock or left asynchronous. For more information see **Section 19.4.1 "Comparator Output Synchronization"**.

21.6.2.4 Comparator C2 Gate Operation

The output resulting from a Comparator 2 operation can be selected as a source for Timer1 Gate Control. The Comparator 2 output (sync_C2OUT) can be synchronized to the Timer1 clock or left asynchronous. For more information see **Section 19.4.1 "Comparator Output Synchronization"**.

21.6.3 TIMER1 GATE TOGGLE MODE

When Timer1 Gate Toggle mode is enabled, it is possible to measure the full-cycle length of a Timer1 gate signal, as opposed to the duration of a single level pulse.

The Timer1 gate source is routed through a flip-flop that changes state on every incrementing edge of the signal. See Figure 21-4 for timing details.

Timer1 Gate Toggle mode is enabled by setting the T1GTM bit of the T1GCON register. When the T1GTM bit is cleared, the flip-flop is cleared and held clear. This is necessary in order to control which edge is measured.

> **Note:** Enabling Toggle mode at the same time as changing the gate polarity may result in indeterminate operation.

21.6.4 TIMER1 GATE SINGLE-PULSE MODE

When Timer1 Gate Single-Pulse mode is enabled, it is possible to capture a single pulse gate event. Timer1 Gate Single-Pulse mode is first enabled by setting the T1GSPM bit in the T1GCON register. Next, the T1GGO/DONE bit in the T1GCON register must be set. The Timer1 will be fully enabled on the next incrementing edge. On the next trailing edge of the pulse, the T1GGO/DONE bit will automatically be cleared. No other gate events will be allowed to increment Timer1 until the T1GGO/DONE bit is once again set in software. See Figure 21-5 for timing details.

If the Single Pulse Gate mode is disabled by clearing the T1GSPM bit in the T1GCON register, the T1GGO/DONE bit should also be cleared.

Enabling the Toggle mode and the Single-Pulse mode simultaneously will permit both sections to work together. This allows the cycle times on the Timer1 Gate source to be measured. See Figure 21-6 for timing details.

21.6.5 TIMER1 GATE VALUE STATUS

When Timer1 gate value status is utilized, it is possible to read the most current level of the gate control value. The value is stored in the T1GVAL bit in the T1GCON register. The T1GVAL bit is valid even when the Timer1 gate is not enabled (TMR1GE bit is cleared).

21.6.6 TIMER1 GATE EVENT INTERRUPT

When Timer1 gate event interrupt is enabled, it is possible to generate an interrupt upon the completion of a gate event. When the falling edge of T1GVAL occurs, the TMR1GIF flag bit in the PIR1 register will be set. If the TMR1GIE bit in the PIE1 register is set, then an interrupt will be recognized.

The TMR1GIF flag bit operates even when the Timer1 gate is not enabled (TMR1GE bit is cleared).

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FIGURE 21-6: TIMER1 GATE SINGLE-PULSE AND TOGGLE COMBINED MODE

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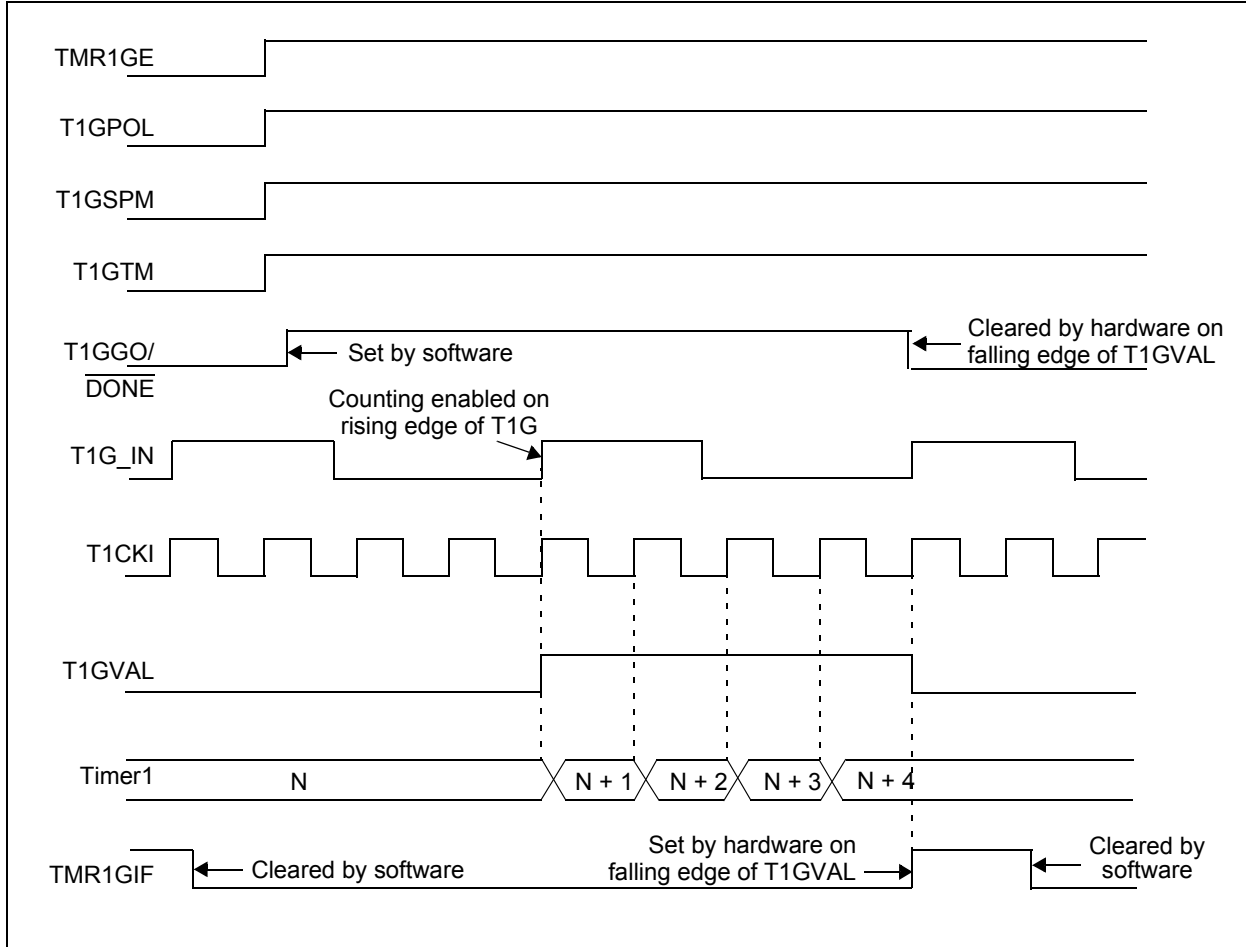
25.2.3 SPI MASTER MODE

The master can initiate the data transfer at any time because it controls the SCKx line. The master determines when the slave (Processor 2, Figure 25-5) is to broadcast data by the software protocol.

In Master mode, the data is transmitted/received as soon as the SSPxBUF register is written to. If the SPI is only going to receive, the SDOx output could be disabled (programmed as an input). The SSPxSR register will continue to shift in the signal present on the SDIx pin at the programmed clock rate. As each byte is received, it will be loaded into the SSPxBUF register as if a normal received byte (interrupts and Status bits appropriately set).

The clock polarity is selected by appropriately programming the CKP bit of the SSPxCON1 register and the CKE bit of the SSPxSTAT register. This then, would give waveforms for SPI communication as shown in Figure 25-6, Figure 25-8, Figure 25-9 and Figure 25-10, where the MSB is transmitted first. In Master mode, the SPI clock rate (bit rate) is user programmable to be one of the following:

• FOSC/4 (or TCY)
• FOSC/16 (or 4 * TCY)
• FOSC/64 (or 16 * TCY)
• Timer2 output/2
• FOSC/(4 * (SSPxADD + 1))

Figure 25-6 shows the waveforms for Master mode.

When the CKE bit is set, the SDOx data is valid before there is a clock edge on SCKx. The change of the input sample is shown based on the state of the SMP bit. The time when the SSPxBUF is loaded with the received data is shown.

FIGURE 25-6: SPI MODE WAVEFORM (MASTER MODE)
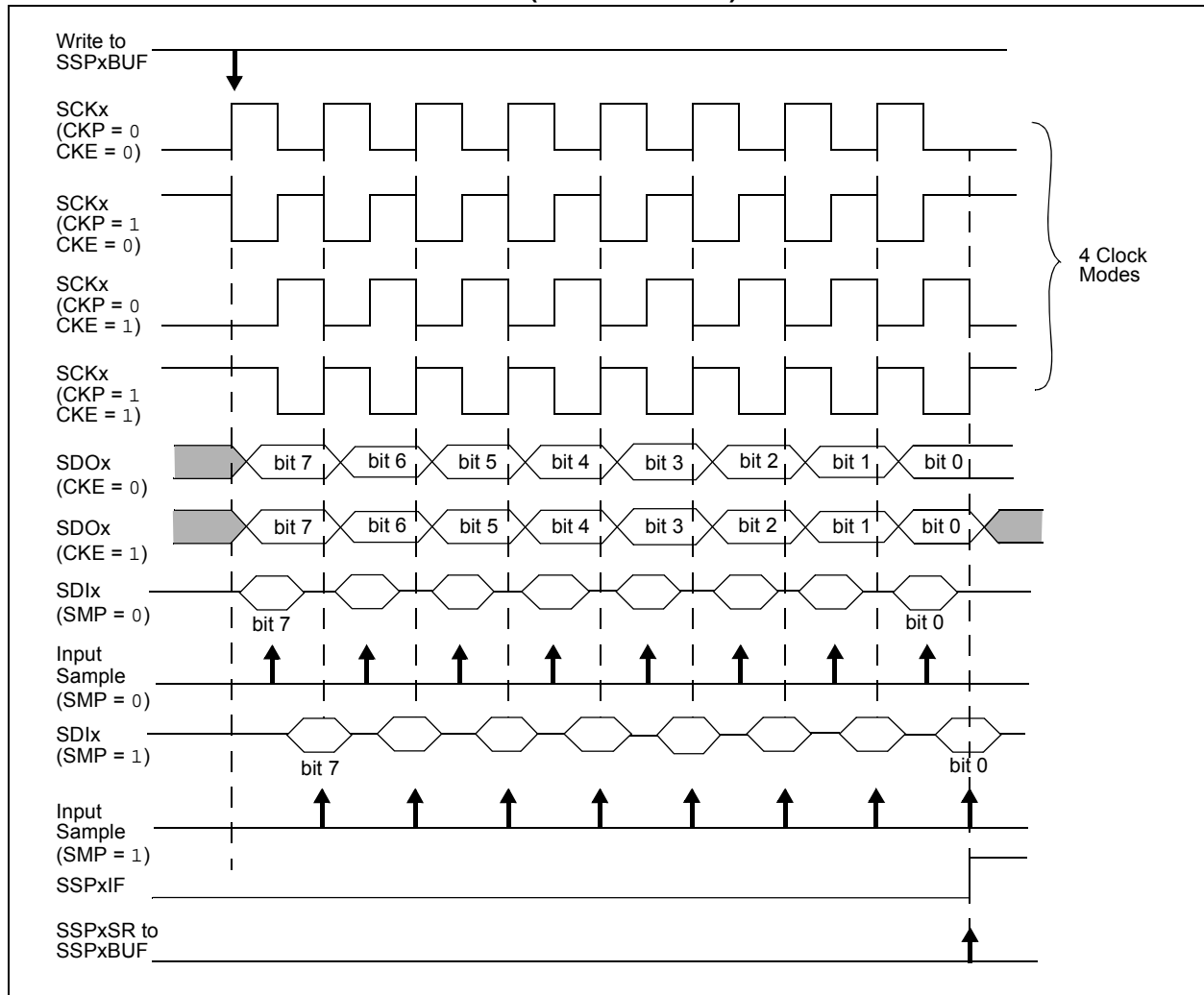
PIC16(L)F1825/9
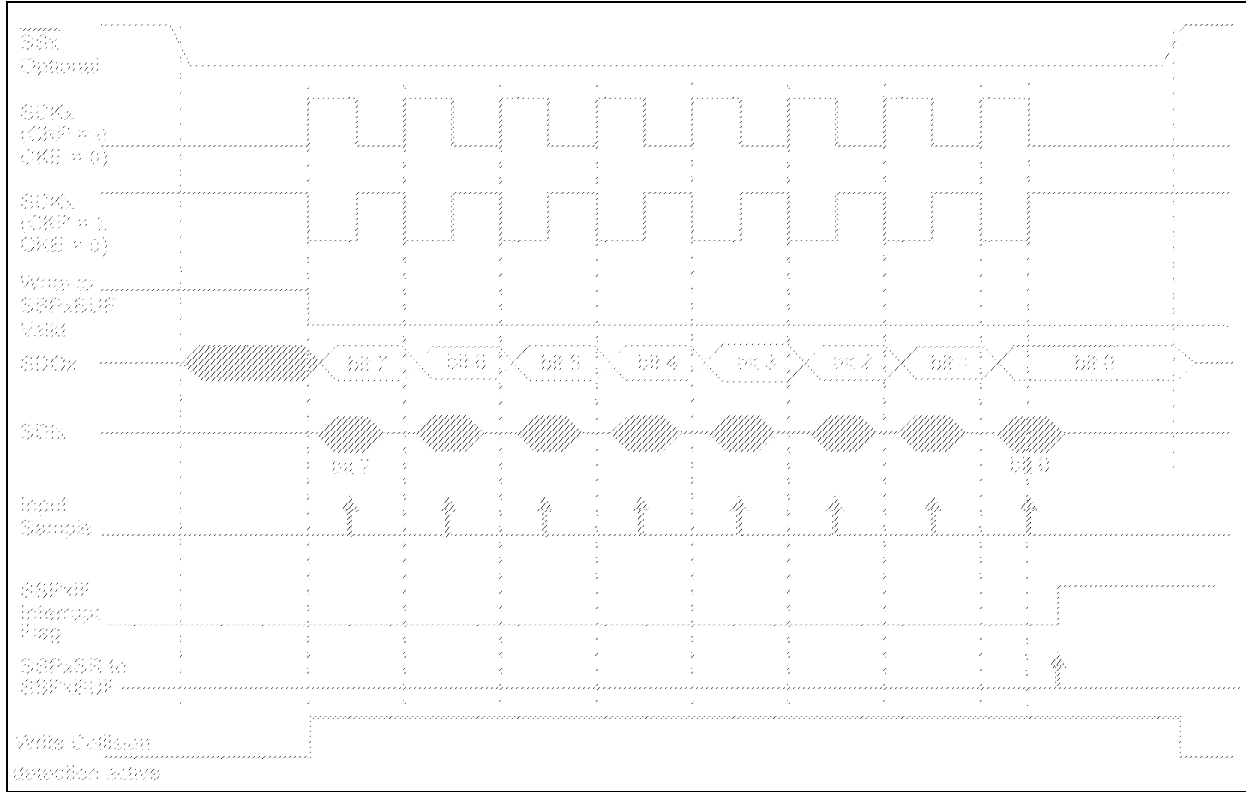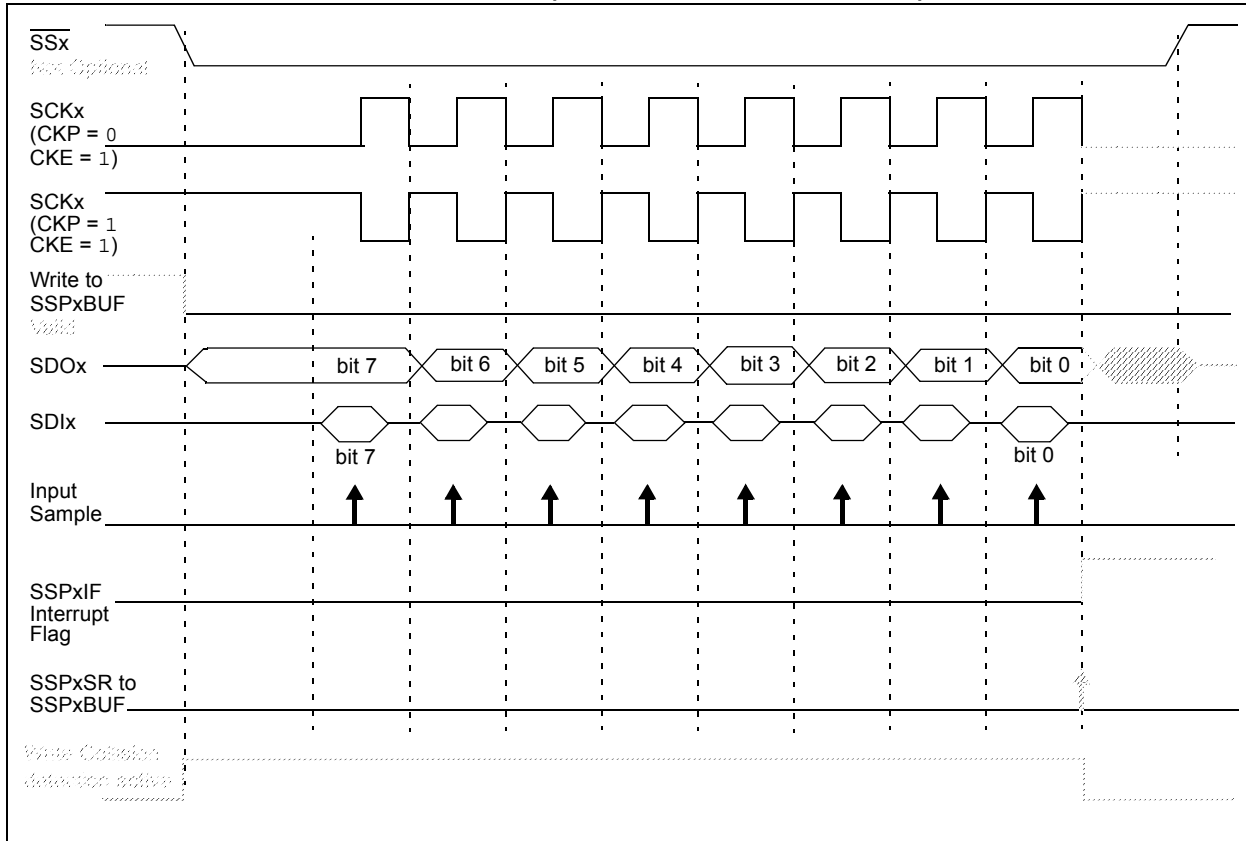
FIGURE 25-9: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 0)

FIGURE 25-10: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 1)

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TABLE 25-3: **SUMMARY OF REGISTERS ASSOCIATED WITH I²C™ OPERATION**

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page
INLVLA	—	—	INLVLA5[1]	INLVLA4	INLVLA3[2]	INLVLA2	INLVLA1	INLVLA0	124
INLVLB[1]	INLVLB7	INLVLB6	INLVLB5	INLVLB4	—	—	—	—	129
INLVLC	INLVLC7[1]	INLVLC6[1]	INLVLC5	INLVLC4	INLVLC3[2]	INLVLC2[2]	INLVLC1	INLVLC0	135
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	87
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	CCP1IE	TMR2IE	TMR1IE	88
PIE2	OSFIE	C2IE	C1IE	EEIE	BCL1IE	—	—	CCP2IE	89
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	92
PIR2	OSFIF	C2IF	C1IF	EEIF	BCL1IF	—	—	CCP2IF	93
SSP1ADD	ADD7	ADD6	ADD5	ADD4	ADD3	ADD2	ADD1	ADD0	280
SSP1BUF	Synchronous Serial Port Receive Buffer/Transmit Register								233*
SSP1CON1	WCOL	SSPOV	SSPEN	CKP	SSPM<3:0>				277
SSP1CON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	278
SSP1CON3	ACKTIM	PCIE	SCIE	BOEN	SDAHT	SBCDE	AHEN	DHEN	279
SSP1MSK	MSK7	MSK6	MSK5	MSK4	MSK3	MSK2	MSK1	MSK0	280
SSP1STAT	SMP	CKE	D/$\overline{\text{A}}$	P	S	R/$\overline{\text{W}}$	UA	BF	276
TRISA	—	—	TRISA5[1]	TRISA4	TRISA3[2]	TRISA2	TRISA1	TRISA0	122
TRISB[1]	TRISB7	TRISB6	TRISB5	TRISB4	—	—	—	—	128
TRISC	TRISC7[1]	TRISC6[1]	TRISC5	TRISC4	TRISC3[2]	TRISC2[2]	TRISC1	TRISC0	133

Legend: — Unimplemented location, read as '0'. Shaded cells are not used by the MSSP module in I²C™ mode.
    * Page provides register information.
Note 1: PIC16(L)F1829 only.
       2: PIC16(L)F1825 only.

REGISTER 25-4: SSPxCON3: SSPx CONTROL REGISTER 3

R-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
ACKTIM	PCIE	SCIE	BOEN	SDAHT	SBCDE	AHEN	DHEN
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7 **ACKTIM:** Acknowledge Time Status bit (I^{2}C mode only)**[3]**
1 = Indicates the I^{2}C bus is in an Acknowledge sequence, set on 8TH falling edge of SCLx clock
0 = Not an Acknowledge sequence, cleared on 9TH rising edge of SCLx clock

bit 6 **PCIE**: Stop Condition Interrupt Enable bit (I^{2}C mode only)
1 = Enable interrupt on detection of Stop condition
0 = Stop detection interrupts are disabled**[2]**

bit 5 **SCIE**: Start Condition Interrupt Enable bit (I^{2}C mode only)
1 = Enable interrupt on detection of Start or Restart conditions
0 = Start detection interrupts are disabled**[2]**

bit 4 **BOEN:** Buffer Overwrite Enable bit
In SPI Slave mode:**[1]**
  1 = SSPxBUF updates every time that a new data byte is shifted in ignoring the BF bit
  0 = If new byte is received with BF bit of the SSPxSTAT register already set, SSPOV bit of the SSPxCON1 register is set, and the buffer is not updated
In I^{2}C Master mode and SPI Master mode:
  This bit is ignored.
In I^{2}C Slave mode:
  1 = SSPxBUF is updated and $\overline{ACK}$ is generated for a received address/data byte, ignoring the state of the SSPOV bit only if the BF bit = 0.
  0 = SSPxBUF is only updated when SSPOV is clear

bit 3 **SDAHT:** SDAx Hold Time Selection bit (I^{2}C mode only)
1 = Minimum of 300 ns hold time on SDAx after the falling edge of SCLx
0 = Minimum of 100 ns hold time on SDAx after the falling edge of SCLx

bit 2 **SBCDE:** Slave Mode Bus Collision Detect Enable bit (I^{2}C Slave mode only)
If on the rising edge of SCLx, SDAx is sampled low when the module is outputting a high state, the BCLxIF bit of the PIR2 register is set, and bus goes Idle
1 = Enable slave bus collision interrupts
0 = Slave bus collision interrupts are disabled

bit 1 **AHEN:** Address Hold Enable bit (I^{2}C Slave mode only)
1 = Following the 8th falling edge of SCLx for a matching received address byte; CKP bit of the SSPxCON1 register will be cleared and the SCLx will be held low.
0 = Address holding is disabled

bit 0 **DHEN:** Data Hold Enable bit (I^{2}C Slave mode only)
1 = Following the 8th falling edge of SCLx for a received data byte; slave hardware clears the CKP bit of the SSPxCON1 register and SCLx is held low.
0 = Data holding is disabled

Note 1: For daisy-chained SPI operation; allows the user to ignore all but the last received byte. SSPOV is still set when a new byte is received and BF = 1, but hardware continues to write the most recent byte to SSPxBUF.
2: This bit has no effect in Slave modes that Start and Stop condition detection is explicitly listed as enabled.
3: The ACKTIM Status bit is only active when the AHEN bit or DHEN bit is set.

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TABLE 26-5: BAUD RATES FOR ASYNCHRONOUS MODES (CONTINUED)

BAUD RATE	SYNC = 0, BRGH = 1, BRG16 = 0											
	FOSC = 8.000 MHz			FOSC = 4.000 MHz			FOSC = 3.6864 MHz			FOSC = 1.000 MHz		
	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	—	—	—	—	—	—	—	—	—	300	0.16	207
1200	—	—	—	1202	0.16	207	1200	0.00	191	1202	0.16	51
2400	2404	0.16	207	2404	0.16	103	2400	0.00	95	2404	0.16	25
9600	9615	0.16	51	9615	0.16	25	9600	0.00	23	—	—	—
10417	10417	0.00	47	10417	0.00	23	10473	0.53	21	10417	0.00	5
19.2k	19231	0.16	25	19.23k	0.16	12	19.2k	0.00	11	—	—	—
57.6k	55556	-3.55	8	—	—	—	57.60k	0.00	3	—	—	—
115.2k	—	—	—	—	—	—	115.2k	0.00	1	—	—	—

BAUD RATE	SYNC = 0, BRGH = 0, BRG16 = 1											
	FOSC = 32.000 MHz			FOSC = 20.000 MHz			FOSC = 18.432 MHz			FOSC = 11.0592 MHz		
	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	300.0	0.00	6666	300.0	-0.01	4166	300.0	0.00	3839	300.0	0.00	2303
1200	1200	-0.02	3332	1200	-0.03	1041	1200	0.00	959	1200	0.00	575
2400	2401	-0.04	832	2399	-0.03	520	2400	0.00	479	2400	0.00	287
9600	9615	0.16	207	9615	0.16	129	9600	0.00	119	9600	0.00	71
10417	10417	0.00	191	10417	0.00	119	10378	-0.37	110	10473	0.53	65
19.2k	19.23k	0.16	103	19.23k	0.16	64	19.20k	0.00	59	19.20k	0.00	35
57.6k	57.14k	-0.79	34	56.818	-1.36	21	57.60k	0.00	19	57.60k	0.00	11
115.2k	117.6k	2.12	16	113.636	-1.36	10	115.2k	0.00	9	115.2k	0.00	5

BAUD RATE	SYNC = 0, BRGH = 0, BRG16 = 1											
	FOSC = 8.000 MHz			FOSC = 4.000 MHz			FOSC = 3.6864 MHz			FOSC = 1.000 MHz		
	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	299.9	-0.02	1666	300.1	0.04	832	300.0	0.00	767	300.5	0.16	207
1200	1199	-0.08	416	1202	0.16	207	1200	0.00	191	1202	0.16	51
2400	2404	0.16	207	2404	0.16	103	2400	0.00	95	2404	0.16	25
9600	9615	0.16	51	9615	0.16	25	9600	0.00	23	—	—	—
10417	10417	0.00	47	10417	0.00	23	10473	0.53	21	10417	0.00	5
19.2k	19.23k	0.16	25	19.23k	0.16	12	19.20k	0.00	11	—	—	—
57.6k	55556	-3.55	8	—	—	—	57.60k	0.00	3	—	—	—
115.2k	—	—	—	—	—	—	115.2k	0.00	1	—	—	—

MOVIW	Move INDFn to W
Syntax:	[*label*] MOVIW ++FSRn [*label*] MOVIW --FSRn [*label*] MOVIW FSRn++ [*label*] MOVIW FSRn-- [*label*] MOVIW k[FSRn]
Operands:	$n \in [0,1]$ $mm \in [00,01,10,11]$ $-32 \le k \le 31$
Operation:	INDFn $\rightarrow$ W Effective address is determined by • FSR + 1 (preincrement) • FSR - 1 (predecrement) • FSR + k (relative offset) After the Move, the FSR value will be either: • FSR + 1 (all increments) • FSR - 1 (all decrements) • Unchanged
Status Affected:	Z

Mode	Syntax	mm
Preincrement	++FSRn	00
Predecrement	--FSRn	01
Postincrement	FSRn++	10
Postdecrement	FSRn--	11

Description:	This instruction is used to move data between W and one of the indirect registers (INDFn). Before/after this move, the pointer (FSRn) is updated by pre/post incrementing/decrementing it. Note: The INDFn registers are not physical registers. Any instruction that accesses an INDFn register actually accesses the register at the address specified by the FSRn. FSRn is limited to the range 0000h - FFFFh. Incrementing/decrementing it beyond these bounds will cause it to wrap-around.

MOVLB	Move literal to BSR
Syntax:	[*label*] MOVLB k
Operands:	$0 \le k \le 15$
Operation:	k $\rightarrow$ BSR
Status Affected:	None
Description:	The 5-bit literal 'k' is loaded into the Bank Select Register (BSR).

MOVLP	Move literal to PCLATH
Syntax:	[*label*] MOVLP k
Operands:	$0 \le k \le 127$
Operation:	k $\rightarrow$ PCLATH
Status Affected:	None
Description:	The 7-bit literal 'k' is loaded into the PCLATH register.

MOVLW	Move literal to W
Syntax:	[*label*] MOVLW k
Operands:	$0 \le k \le 255$
Operation:	k $\rightarrow$ (W)
Status Affected:	None
Description:	The 8-bit literal 'k' is loaded into W register. The "don't cares" will assemble as '0's.
Words:	1
Cycles:	1
Example:	MOVLW 0x5A

After Instruction
        W   =   0x5A

MOVWF	Move W to f
Syntax:	[*label*] MOVWF f
Operands:	$0 \le f \le 127$
Operation:	(W) $\rightarrow$ (f)
Status Affected:	None
Description:	Move data from W register to register 'f'.
Words:	1
Cycles:	1
Example:	MOVWF OPTION_REG

Before Instruction
    OPTION_REG = 0xFF
          W = 0x4F
After Instruction
    OPTION_REG = 0x4F
          W = 0x4F

TABLE 30-2: OSCILLATOR PARAMETERS

			Freq. Tolerance					
Param No.	**Sym.**	**Characteristic**		**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**
OS08	HF$_{OSC}$	Internal Calibrated HFINTOSC Frequency[1]	±2%	—	16.0	—	MHz	0°C ≤ TA ≤ +60°C, VDD ≥ 2.5V
			±3%	—	16.0	—	MHz	60°C ≤ TA ≤ +85°C, VDD ≥ 2.5V
			±5%	—	16.0	—	MHz	-40°C ≤ TA ≤ +125°C
OS08A	MF$_{OSC}$	Internal Calibrated MFINTOSC Frequency[1]	±2%	—	500	—	kHz	0°C ≤ TA ≤ +60°C, VDD ≥ 2.5V
			±3%	—	500	—	kHz	60°C ≤ TA ≤ +85°C, VDD ≥ 2.5V
			±5%	—	500	—	kHz	-40°C ≤ TA ≤ +125°C
OS09	LF$_{OSC}$	Internal LFINTOSC Frequency	±25%	—	31	—	kHz	-40°C ≤ TA ≤ +125°C
OS10*	T$_{IOSC\ ST}$	HFINTOSC Wake-up from Sleep Start-up Time	—	—	5	8	μs	
		MFINTOSC Wake-up from Sleep Start-up Time	—	—	20	30	μs	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: To ensure these oscillator frequency tolerances, VDD and VSS must be capacitively decoupled as close to the device as possible. 0.1 μF and 0.01 μF values in parallel are recommended.

TABLE 30-3: PLL CLOCK TIMING SPECIFICATIONS (VDD = 2.7V TO 5.5V)

Param No.	**Sym.**	**Characteristic**	**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**
F10	F$_{OSC}$	Oscillator Frequency Range	4	—	8	MHz	
F11	F$_{SYS}$	On-Chip VCO System Frequency	16	—	32	MHz	
F12	T$_{RC}$	PLL Start-up Time (Lock Time)	—	—	2	ms	
F13*	ΔCLK	CLKOUT Stability (Jitter)	-0.25%	—	+0.25%	%	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

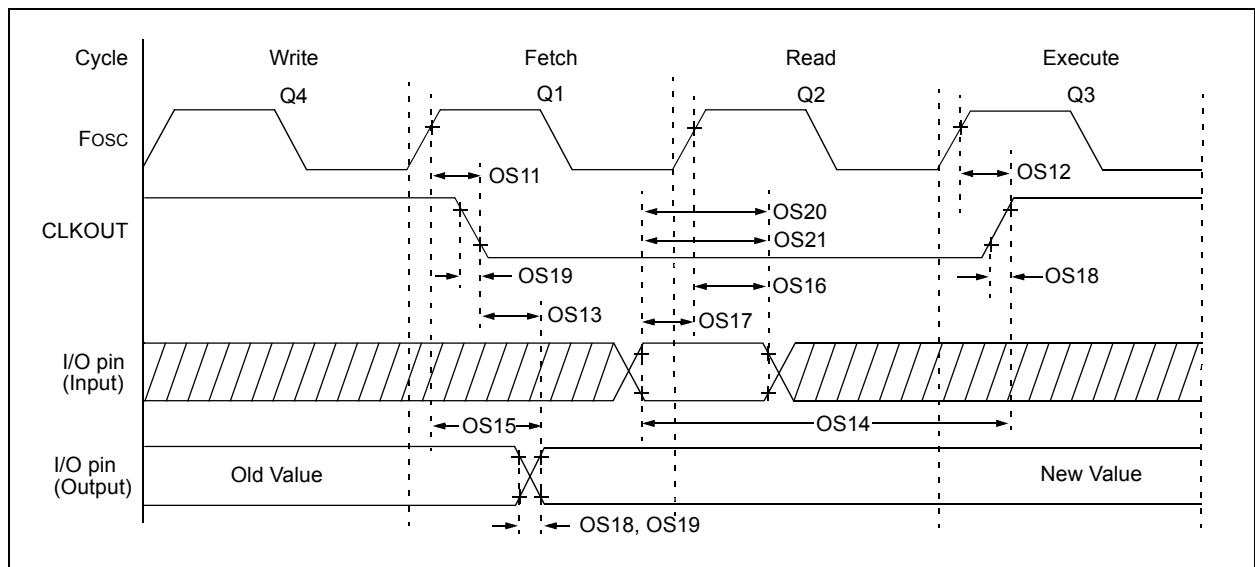## FIGURE 30-7: CLKOUT AND I/O TIMING

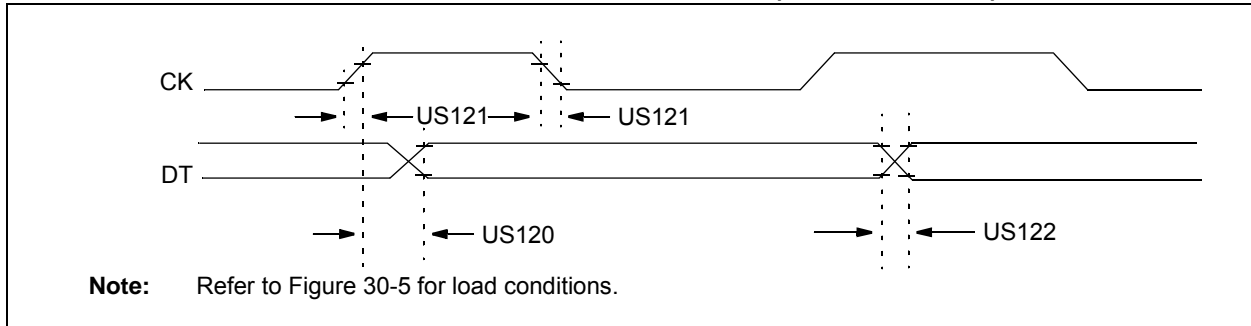**FIGURE 30-14:** **USART SYNCHRONOUS TRANSMISSION (MASTER/SLAVE) TIMING**

Note: Refer to Figure 30-5 for load conditions.

TABLE 30-13: **USART SYNCHRONOUS TRANSMISSION REQUIREMENTS**

Param. No.	Symbol	Characteristic		Min.	Max.	Units	Conditions
US120	TCKH2DTV	SYNC XMIT (Master and Slave) Clock high to data-out valid	3.0-5.5V	—	80	ns	
			1.8-5.5V	—	100	ns	
US121	TCKRF	Clock out rise time and fall time (Master mode)	3.0-5.5V	—	45	ns	
			1.8-5.5V	—	50	ns	
US122	TDTRF	Data-out rise time and fall time	3.0-5.5V	—	45	ns	
			1.8-5.5V	—	50	ns	

FIGURE 30-15: **USART SYNCHRONOUS RECEIVE (MASTER/SLAVE) TIMING**

Note: Refer to Figure 30-5 for load conditions.
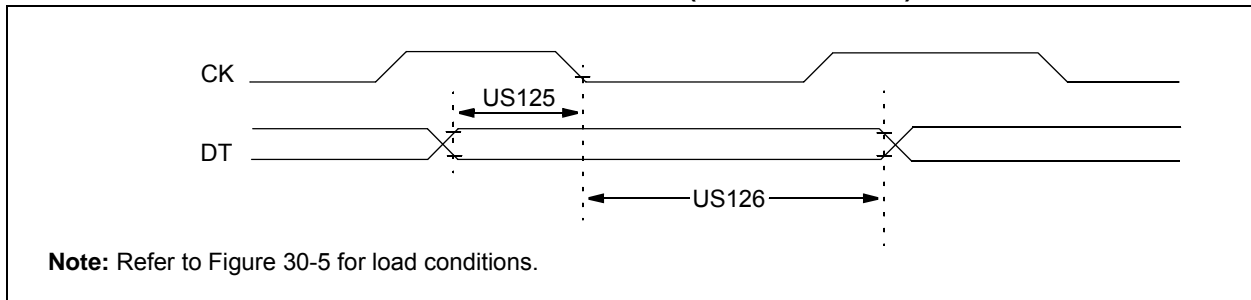
TABLE 30-14: **USART SYNCHRONOUS RECEIVE REQUIREMENTS**

Param. No.	Symbol	Characteristic	Min.	Max.	Units	Conditions
US125	TDTV2CKL	SYNC RCV (Master and Slave) Data-hold before CK ↓ (DT hold time)	10	—	ns	
US126	TCKL2DTL	Data-hold after CK ↓ (DT hold time)	15	—	ns	

31.0 DC AND AC CHARACTERISTICS GRAPHS AND CHARTS

The graphs and tables provided in this section are for **design guidance** and are **not tested**.

In some graphs or tables, the data presented are **outside specified operating range** (i.e., outside specified V_{DD} range). This is for **information only** and devices are ensured to operate properly only within the specified range.

> **Note:** The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore, outside the warranted range.

"Typical" represents the mean of the distribution at 25°C. "MAXIMUM", "Max.", "MINIMUM" or "Min." represents (mean + 3σ) or (mean - 3σ) respectively, where σ is a standard deviation, over each temperature range.