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#### Details

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Product Status	Active
Core Processor	AVR
Core Size	8-Bit
Speed	20MHz
Connectivity	I <sup>2</sup> C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	23
Program Memory Size	4KB (2K x 16)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	512 x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 8x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	28-VFQFN Exposed Pad
Supplier Device Package	28-VQFN (4x4)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/atmega48-20mmu

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong



The setup of the OC0x should be performed before setting the Data Direction Register for the port pin to output. The easiest way of setting the OC0x value is to use the Force Output Compare (FOC0x) strobe bits in Normal mode. The OC0x Registers keep their values even when changing between Waveform Generation modes.

Be aware that the COM0x1:0 bits are not double buffered together with the compare value. Changing the COM0x1:0 bits will take effect immediately.

#### 12.5 Compare Match Output Unit

The Compare Output mode (COM0x1:0) bits have two functions. The Waveform Generator uses the COM0x1:0 bits for defining the Output Compare (OC0x) state at the next compare match. Also, the COM0x1:0 bits control the OC0x pin output source. Figure 12-4 shows a simplified schematic of the logic affected by the COM0x1:0 bit setting. The I/O Registers, I/O bits, and I/O pins in the figure are shown in bold. Only the parts of the general I/O port control registers (DDR and PORT) that are affected by the COM0x1:0 bits are shown. When referring to the OC0x state, the reference is for the internal OC0x Register, not the OC0x pin. If a system reset occur, the OC0x Register is reset to "0".



Figure 12-4. Compare Match Output Unit, Schematic

The general I/O port function is overridden by the Output Compare (OC0x) from the Waveform Generator if either of the COM0x1:0 bits are set. However, the OC0x pin direction (input or output) is still controlled by the Data Direction Register (DDR) for the port pin. The Data Direction Register bit for the OC0x pin (DDR\_OC0x) must be set as output before the OC0x value is visible on the pin. The port override function is independent of the Waveform Generation mode.

The design of the Output Compare pin logic allows initialization of the OC0x state before the output is enabled. Note that some COM0x1:0 bit settings are reserved for certain modes of operation. See Section "12.8" on page 99.

#### 12.5.1 Compare Output Mode and Waveform Generation

The Waveform Generator uses the COM0x1:0 bits differently in Normal, CTC, and PWM modes. For all modes, setting the COM0x1:0 = 0 tells the Waveform Generator that no action on the OC0x Register is to be performed on the next compare match. For compare output actions in the



The PWM resolution for fast PWM can be fixed to 8-, 9-, or 10-bit, or defined by either ICR1 or OCR1A. The minimum resolution allowed is 2-bit (ICR1 or OCR1A set to 0x0003), and the maximum resolution is 16-bit (ICR1 or OCR1A set to MAX). The PWM resolution in bits can be calculated by using the following equation:

$$R_{FPWM} = \frac{\log(TOP + 1)}{\log(2)}$$

In fast PWM mode the counter is incremented until the counter value matches either one of the fixed values 0x00FF, 0x01FF, or 0x03FF (WGM13:0 = 5, 6, or 7), the value in ICR1 (WGM13:0 = 14), or the value in OCR1A (WGM13:0 = 15). The counter is then cleared at the following timer clock cycle. The timing diagram for the fast PWM mode is shown in Figure 13-7. The figure shows fast PWM mode when OCR1A or ICR1 is used to define TOP. The TCNT1 value is in the timing diagram shown as a histogram for illustrating the single-slope operation. The diagram includes non-inverted and inverted PWM outputs. The small horizontal line marks on the TCNT1 slopes represent compare matches between OCR1x and TCNT1. The OC1x Interrupt Flag will be set when a compare match occurs.

Figure 13-7. Fast PWM Mode, Timing Diagram



The Timer/Counter Overflow Flag (TOV1) is set each time the counter reaches TOP. In addition the OC1A or ICF1 Flag is set at the same timer clock cycle as TOV1 is set when either OCR1A or ICR1 is used for defining the TOP value. If one of the interrupts are enabled, the interrupt handler routine can be used for updating the TOP and compare values.

When changing the TOP value the program must ensure that the new TOP value is higher or equal to the value of all of the Compare Registers. If the TOP value is lower than any of the Compare Registers, a compare match will never occur between the TCNT1 and the OCR1x. Note that when using fixed TOP values the unused bits are masked to zero when any of the OCR1x Registers are written.

The procedure for updating ICR1 differs from updating OCR1A when used for defining the TOP value. The ICR1 Register is not double buffered. This means that if ICR1 is changed to a low value when the counter is running with none or a low prescaler value, there is a risk that the new ICR1 value written is lower than the current value of TCNT1. The result will then be that the counter will miss the compare match at the TOP value. The counter will then have to count to the MAX value (0xFFFF) and wrap around starting at 0x0000 before the compare match can occur. The OCR1A Register however, is double buffered. This feature allows the OCR1A I/O location



### 13.10 16-bit Timer/Counter Register Description

#### 13.10.1 Timer/Counter1 Control Register A – TCCR1A

Bit	7	6	5	4	3	2	1	0	_
	COM1A1	COM1A0	COM1B1	COM1B0	-	-	WGM11	WGM10	TCCR1A
Read/Write	R/W	R/W	R/W	R/W	R	R	R/W	R/W	
Initial Value	0	0	0	0	0	0	0	0	

#### • Bit 7:6 – COM1A1:0: Compare Output Mode for Channel A

#### • Bit 5:4 – COM1B1:0: Compare Output Mode for Channel B

The COM1A1:0 and COM1B1:0 control the Output Compare pins (OC1A and OC1B respectively) behavior. If one or both of the COM1A1:0 bits are written to one, the OC1A output overrides the normal port functionality of the I/O pin it is connected to. If one or both of the COM1B1:0 bit are written to one, the OC1B output overrides the normal port functionality of the I/O pin it is connected to. However, note that the *Data Direction Register* (DDR) bit corresponding to the OC1A or OC1B pin must be set in order to enable the output driver.

When the OC1A or OC1B is connected to the pin, the function of the COM1x1:0 bits is dependent of the WGM13:0 bits setting. Table 13-1 shows the COM1x1:0 bit functionality when the WGM13:0 bits are set to a Normal or a CTC mode (non-PWM).

COM1A1/COM1B1	COM1A0/COM1B0	Description
0	0	Normal port operation, OC1A/OC1B disconnected.
0	1	Toggle OC1A/OC1B on Compare Match.
1	0	Clear OC1A/OC1B on Compare Match (Set output to low level).
1	1	Set OC1A/OC1B on Compare Match (Set output to high level).

Table 13-1. Compare Output Mode, non-PWM

Table 13-2 shows the COM1x1:0 bit functionality when the WGM13:0 bits are set to the fast PWM mode.

Table 13-2.	Compare	Output M	lode, Fast	PWM <sup>(1)</sup>
			,	

COM1A1/COM1B1	COM1A0/COM1B0	Description
0	0	Normal port operation, OC1A/OC1B disconnected.
0	1	WGM13:0 = 14 or 15: Toggle OC1A on Compare Match, OC1B disconnected (normal port operation). For all other WGM1 settings, normal port operation, OC1A/OC1B disconnected.
1	0	Clear OC1A/OC1B on Compare Match, set OC1A/OC1B at TOP
1	1	Set OC1A/OC1B on Compare Match, clear OC1A/OC1B at TOP

 A special case occurs when OCR1A/OCR1B equals TOP and COM1A1/COM1B1 is set. In this case the compare match is ignored, but the set or clear is done at TOP. See Section "13.8.3" on page 119. for more details. When the ICR1 is used as TOP value (see description of the WGM13:0 bits located in the TCCR1A and the TCCR1B Register), the ICP1 is disconnected and consequently the Input Capture function is disabled.

#### • Bit 5 – Reserved Bit

This bit is reserved for future use. For ensuring compatibility with future devices, this bit must be written to zero when TCCR1B is written.

#### • Bit 4:3 – WGM13:2: Waveform Generation Mode

See TCCR1A Register description.

#### • Bit 2:0 - CS12:0: Clock Select

The three Clock Select bits select the clock source to be used by the Timer/Counter, see Figure 13-10 and Figure 13-11.

CS12	CS11	CS10	Description
0	0	0	No clock source (Timer/Counter stopped).
0	0	1	clk <sub>I/O</sub> /1 (No prescaling)
0	1	0	clk <sub>I/O</sub> /8 (From prescaler)
0	1	1	clk <sub>I/O</sub> /64 (From prescaler)
1	0	0	clk <sub>I/O</sub> /256 (From prescaler)
1	0	1	clk <sub>I/O</sub> /1024 (From prescaler)
1	1	0	External clock source on T1 pin. Clock on falling edge.
1	1	1	External clock source on T1 pin. Clock on rising edge.

Table 13-5. Clock Select Bit Description

If external pin modes are used for the Timer/Counter1, transitions on the T1 pin will clock the counter even if the pin is configured as an output. This feature allows software control of the counting.

#### 13.10.3 Timer/Counter1 Control Register C – TCCR1C



#### • Bit 7 – FOC1A: Force Output Compare for Channel A

#### • Bit 6 – FOC1B: Force Output Compare for Channel B

The FOC1A/FOC1B bits are only active when the WGM13:0 bits specifies a non-PWM mode. However, for ensuring compatibility with future devices, these bits must be set to zero when TCCR1A is written when operating in a PWM mode. When writing a logical one to the FOC1A/FOC1B bit, an immediate compare match is forced on the Waveform Generation unit. The OC1A/OC1B output is changed according to its COM1x1:0 bits setting. Note that the FOC1A/FOC1B bits are implemented as strobes. Therefore it is the value present in the COM1x1:0 bits that determine the effect of the forced compare.





### 15. 8-bit Timer/Counter2 with PWM and Asynchronous Operation

Timer/Counter2 is a general purpose, single channel, 8-bit Timer/Counter module. The main features are:

- Single Channel Counter
- Clear Timer on Compare Match (Auto Reload)
- Glitch-free, Phase Correct Pulse Width Modulator (PWM)
- Frequency Generator
- 10-bit Clock Prescaler
- Overflow and Compare Match Interrupt Sources (TOV2, OCF2A and OCF2B)
- Allows Clocking from External 32 kHz Watch Crystal Independent of the I/O Clock

#### 15.1 Overview

A simplified block diagram of the 8-bit Timer/Counter is shown in Figure 15-1. For the actual placement of I/O pins, refer to "Pinout ATmega48/88/168" on page 2. CPU accessible I/O Registers, including I/O bits and I/O pins, are shown in bold. The device-specific I/O Register and bit locations are listed in the "8-bit Timer/Counter Register Description" on page 149.

The PRTIM2 bit in "Power Reduction Register - PRR" on page 40 must be written to zero to enable Timer/Counter2 module.









Figure 15-10 shows the setting of OCF2A in all modes except CTC mode.

Figure 15-10. Timer/Counter Timing Diagram, Setting of OCF2A, with Prescaler ( $f_{clk_l/O}/8$ )



Figure 15-11 shows the setting of OCF2A and the clearing of TCNT2 in CTC mode.

Figure 15-11. Timer/Counter Timing Diagram, Clear Timer on Compare Match mode, with Prescaler (f<sub>clk\_l/O</sub>/8)





#### 17.4 USART Initialization

The USART has to be initialized before any communication can take place. The initialization process normally consists of setting the baud rate, setting frame format and enabling the Transmitter or the Receiver depending on the usage. For interrupt driven USART operation, the Global Interrupt Flag should be cleared (and interrupts globally disabled) when doing the initialization.

Before doing a re-initialization with changed baud rate or frame format, be sure that there are no ongoing transmissions during the period the registers are changed. The TXCn Flag can be used to check that the Transmitter has completed all transfers, and the RXC Flag can be used to check that there are no unread data in the receive buffer. Note that the TXCn Flag must be cleared before each transmission (before UDRn is written) if it is used for this purpose.

The following simple USART initialization code examples show one assembly and one C function that are equal in functionality. The examples assume asynchronous operation using polling (no interrupts enabled) and a fixed frame format. The baud rate is given as a function parameter.

# ATmega48/88/168

The following code example shows a simple USART receive function based on polling of the Receive Complete (RXCn) Flag. When using frames with less than eight bits the most significant bits of the data read from the UDRn will be masked to zero. The USART has to be initialized before the function can be used.

```
Assembly Code Example<sup>(1)</sup>
```

```
USART_Receive:

; Wait for data to be received

sbis UCSRNA, RXCn

rjmp USART_Receive

; Get and return received data from buffer

in r16, UDRn

ret
```

C Code Example<sup>(1)</sup>

```
unsigned char USART_Receive( void )
{
    /* Wait for data to be received */
    while ( !(UCSRnA & (1<<RXCn)) )
        ;
    /* Get and return received data from buffer */
    return UDRn;
}</pre>
```

Note: 1. See "About Code Examples" on page 6.

For I/O Registers located in extended I/O map, "IN", "OUT", "SBIS", "SBIC", "CBI", and "SBI" instructions must be replaced with instructions that allow access to extended I/O. Typically "LDS" and "STS" combined with "SBRS", "SBRC", "SBR", and "CBR".

The function simply waits for data to be present in the receive buffer by checking the RXCn Flag, before reading the buffer and returning the value.

#### 17.6.2 Receiving Frames with 9 Data Bits

If 9-bit characters are used (UCSZn=7) the ninth bit must be read from the RXB8n bit in UCS-RnB **before** reading the low bits from the UDRn. This rule applies to the FEn, DORn and UPEn Status Flags as well. Read status from UCSRnA, then data from UDRn. Reading the UDRn I/O location will change the state of the receive buffer FIFO and consequently the TXB8n, FEn, DORn and UPEn bits, which all are stored in the FIFO, will change.

The following code example shows a simple USART receive function that handles both nine bit characters and the status bits.





#### Assembly Code Example<sup>(1)</sup>

```
USART_Receive:
     ; Wait for data to be received
     sbis UCSRnA, RXCn
     rjmp USART_Receive
     ; Get status and 9th bit, then data from buffer
     in
          r18, UCSRnA
     in
          r17, UCSRnB
          r16, UDRn
     in
     ; If error, return -1
     andi r18, (1<<FEn) | (1<<DORn) | (1<<UPEn)
     breq USART_ReceiveNoError
     1di r17, HIGH(-1)
     ldi r16, LOW(-1)
   USART_ReceiveNoError:
     ; Filter the 9th bit, then return
     lsr r17
     andi r17, 0x01
     ret
C Code Example<sup>(1)</sup>
   unsigned int USART_Receive( void )
   {
     unsigned char status, resh, resl;
     /* Wait for data to be received */
     while ( !(UCSRnA & (1<<RXCn)) )</pre>
     /* Get status and 9th bit, then data */
     /* from buffer */
     status = UCSRnA;
     resh = UCSRnB;
     resl = UDRn;
     /* If error, return -1 */
     if ( status & (1<<FEn) | (1<<DORn) | (1<<UPEn) )
       return -1;
     /* Filter the 9th bit, then return */
     resh = (resh >> 1) & 0x01;
     return ((resh << 8) | resl);</pre>
   }
```

Note: 1. See "About Code Examples" on page 6.

For I/O Registers located in extended I/O map, "IN", "OUT", "SBIS", "SBIC", "CBI", and "SBI" instructions must be replaced with instructions that allow access to extended I/O. Typically "LDS" and "STS" combined with "SBRS", "SBRC", "SBR", and "CBR".

The receive function example reads all the I/O Registers into the Register File before any computation is done. This gives an optimal receive buffer utilization since the buffer location read will be free to accept new data as early as possible.



		$f_{osc} = 8.0$	0000 MHz			f <sub>osc</sub> = 11.0592 MHz			f <sub>osc</sub> = 14.7456 MHz			
Baud	U2X	n = 0	U2X	n = 1	U2X	n = 0	U2X	n = 1	U2X	n = 0	U2X	n = 1
Rate (bps)	UBRR n	Error	UBRR n	Error	UBRR n	Error	UBRR n	Error	UBRR n	Error	UBRR n	Error
2400	207	0.2%	416	-0.1%	287	0.0%	575	0.0%	383	0.0%	767	0.0%
4800	103	0.2%	207	0.2%	143	0.0%	287	0.0%	191	0.0%	383	0.0%
9600	51	0.2%	103	0.2%	71	0.0%	143	0.0%	95	0.0%	191	0.0%
14.4k	34	-0.8%	68	0.6%	47	0.0%	95	0.0%	63	0.0%	127	0.0%
19.2k	25	0.2%	51	0.2%	35	0.0%	71	0.0%	47	0.0%	95	0.0%
28.8k	16	2.1%	34	-0.8%	23	0.0%	47	0.0%	31	0.0%	63	0.0%
38.4k	12	0.2%	25	0.2%	17	0.0%	35	0.0%	23	0.0%	47	0.0%
57.6k	8	-3.5%	16	2.1%	11	0.0%	23	0.0%	15	0.0%	31	0.0%
76.8k	6	-7.0%	12	0.2%	8	0.0%	17	0.0%	11	0.0%	23	0.0%
115.2k	3	8.5%	8	-3.5%	5	0.0%	11	0.0%	7	0.0%	15	0.0%
230.4k	1	8.5%	3	8.5%	2	0.0%	5	0.0%	3	0.0%	7	0.0%
250k	1	0.0%	3	0.0%	2	-7.8%	5	-7.8%	3	-7.8%	6	5.3%
0.5M	0	0.0%	1	0.0%	-	_	2	-7.8%	1	-7.8%	3	-7.8%
1M	-	-	0	0.0%	-	—	-	-	0	-7.8%	1	-7.8%
Max. (1)	0.5 M	Vbps	1 N	lbps	691.2	2 kbps	1.3824	1 Mbps	921.6	6 kbps	1.8432	2 Mbps

Table 17-11.	Examples of UBRRn	Settings for Commonly	Used Oscillator Frequencies	(Continued)
				(Contantaoa)

1. UBRRn = 0, Error = 0.0%

```
Assembly Code Example<sup>(1)</sup>
   USART_MSPIM_Transfer:
     ; Wait for empty transmit buffer
     sbis UCSRnA, UDREn
     rjmp USART_MSPIM_Transfer
     ; Put data (r16) into buffer, sends the data
     out UDRn,r16
     ; Wait for data to be received
   USART_MSPIM_Wait_RXCn:
     sbis UCSRnA, RXCn
     rjmp USART_MSPIM_Wait_RXCn
     ; Get and return received data from buffer
     in r16, UDRn
     ret
C Code Example<sup>(1)</sup>
   unsigned char USART_Receive( void )
   {
     /* Wait for empty transmit buffer */
     while ( !( UCSRnA & (1<<UDREn)) );</pre>
     /* Put data into buffer, sends the data */
     UDRn = data;
     /* Wait for data to be received */
     while ( !(UCSRnA & (1<<RXCn)) );</pre>
     /* Get and return received data from buffer */
     return UDRn;
   }
```

Note: 1. See "About Code Examples" on page 6.

#### 18.5.1 Transmitter and Receiver Flags and Interrupts

The RXCn, TXCn, and UDREn flags and corresponding interrupts in USART in MSPIM mode are identical in function to the normal USART operation. However, the receiver error status flags (FE, DOR, and PE) are not in use and is always read as zero.

#### 18.5.2 Disabling the Transmitter or Receiver

The disabling of the transmitter or receiver in USART in MSPIM mode is identical in function to the normal USART operation.

#### 18.6 USART MSPIM Register Description

The following section describes the registers used for SPI operation using the USART.

#### 18.6.1 USART MSPIM I/O Data Register - UDRn

The function and bit description of the USART data register (UDRn) in MSPI mode is identical to normal USART operation. See "USART I/O Data Register n– UDRn" on page 187.

#### 18.6.2 USART MSPIM Control and Status Register n A - UCSRnA

Bit 7 6 5 4 3 2 1 0





#### 19.3.4 Data Packet Format

All data packets transmitted on the TWI bus are nine bits long, consisting of one data byte and an acknowledge bit. During a data transfer, the Master generates the clock and the START and STOP conditions, while the Receiver is responsible for acknowledging the reception. An Acknowledge (ACK) is signalled by the Receiver pulling the SDA line low during the ninth SCL cycle. If the Receiver leaves the SDA line high, a NACK is signalled. When the Receiver has received the last byte, or for some reason cannot receive any more bytes, it should inform the Transmitter by sending a NACK after the final byte. The MSB of the data byte is transmitted first.

#### Figure 19-5. Data Packet Format



#### 19.3.5 Combining Address and Data Packets into a Transmission

A transmission basically consists of a START condition, a SLA+R/W, one or more data packets and a STOP condition. An empty message, consisting of a START followed by a STOP condition, is illegal. Note that the Wired-ANDing of the SCL line can be used to implement handshaking between the Master and the Slave. The Slave can extend the SCL low period by pulling the SCL line low. This is useful if the clock speed set up by the Master is too fast for the Slave, or the Slave needs extra time for processing between the data transmissions. The Slave extending the SCL low period will not affect the SCL high period, which is determined by the Master. As a consequence, the Slave can reduce the TWI data transfer speed by prolonging the SCL duty cycle.

Figure 19-6 shows a typical data transmission. Note that several data bytes can be transmitted between the SLA+R/W and the STOP condition, depending on the software protocol implemented by the application software.







	Assembly Code Example	C Example	Comments
	<b>ldi</b> r16,	TWCR = (1< <twint) (1<<twsta)="" th=""  =""  <=""><th></th></twint)>	
1	(1 << TWINT)   (1 << TWSTA)	(1< <twen)< th=""><th>Send START condition</th></twen)<>	Send START condition
'	(1< <twen)< th=""><th></th><th>Send STATT Condition</th></twen)<>		Send STATT Condition
	out TWCR, r16		
	wait1:	<pre>while (!(TWCR &amp; (1&lt;<twint)))< pre=""></twint)))<></pre>	
2	in r16,TWCR	;	indicates that the START
	sbrs r16,TWINT		condition has been transmitted
	<b>rjmp</b> wait1		
	in r16,TWSR	<b>if</b> ((TWSR & 0xF8) != START)	Check value of TWI Status
	<b>andi</b> r16, 0xF8	ERROR();	Register. Mask prescaler bits. If
	<b>cpi</b> r16, START		status different from START go to
	brne ERROR		ERROR
3	<b>ldi</b> r16, SLA_W	TWDR = SLA_W;	
	out TWDR, r16	TWCR = (1< <twint)< th=""><th>Load SLA_W Into TWDR Begister Clear TWINT bit in</th></twint)<>	Load SLA_W Into TWDR Begister Clear TWINT bit in
	ldi r16, (1< <twint)< th=""><th>(1&lt;<twen);< th=""><th>TWCR to start transmission of</th></twen);<></th></twint)<>	(1< <twen);< th=""><th>TWCR to start transmission of</th></twen);<>	TWCR to start transmission of
	(1< <twen)< th=""><th></th><th>address</th></twen)<>		address
	out TWCR, r16		
	wait2:	<pre>while (!(TWCR &amp; (1&lt;<twint)))< pre=""></twint)))<></pre>	Wait for TWINT Flag set. This
4	in r16,TWCR	;	indicates that the SLA+W has
	sbrs r16, TWINT		ACK/NACK has been received
	rjmp wait2		
	in r16,TWSR	if ((TWSR & 0xF8) !=	Check value of TWI Status
	andi r16, 0xF8		Register. Mask prescaler bits. If
	<b>cpi</b> r16, MT_SLA_ACK	ERROR ();	MT SLA ACK go to FBROR
_	brne ERROR		
5	ldi r16, DATA	TWDR = DATA;	
	out TWDR, r16	TWCR = (1 < TWINT)	Load DATA into TWDR Register.
	ldi r16, (1< <twint) th=""  <=""><th>(1~~1WEN),</th><th>start transmission of data</th></twint)>	(1~~1WEN),	start transmission of data
	wait3:	<b>while</b> (!(TWCR & (1< <twint)))< th=""><th></th></twint)))<>	
	in r16.TWCR	:	Wait for I WIN I Flag set. This
6	sbrs r16, TWINT	, ,	transmitted, and ACK/NACK has
	rimp wait3		been received.
	in r16, TWSR	<b>if</b> ((TWSR & 0xF8) !=	Check value of TMI Status
	<b>andi</b> r16, 0xF8	MT_DATA_ACK)	Begister Mask prescaler hits If
	<b>cpi</b> r16, MT DATA ACK	ERROR();	status different from
_	brne ERROR		MT_DATA_ACK go to ERROR
/	<b>ldi</b> r16,	TWCR = (1< <twint) (1<<twen)="" th=""  =""  <=""><th></th></twint)>	
	(1< <twint) (1<<twen)="" th=""  =""  <=""><th>(1&lt;<twsto);< th=""><th></th></twsto);<></th></twint)>	(1< <twsto);< th=""><th></th></twsto);<>	
	(1< <twsto)< th=""><th></th><th>Transmit STOP condition</th></twsto)<>		Transmit STOP condition
	out TWCR, r16		







#### 19.8.2 Master Receiver Mode

In the Master Receiver mode, a number of data bytes are received from a Slave Transmitter (Slave see Figure 19-14). In order to enter a Master mode, a START condition must be transmitted. The format of the following address packet determines whether Master Transmitter or Master Receiver mode is to be entered. If SLA+W is transmitted, MT mode is entered, if SLA+R is transmitted, MR mode is entered. All the status codes mentioned in this section assume that the prescaler bits are zero or are masked to zero.

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Figure 21-12. Integral Non-linearity (INL)



• Differential Non-linearity (DNL): The maximum deviation of the actual code width (the interval between two adjacent transitions) from the ideal code width (1 LSB). Ideal value: 0 LSB.

Figure 21-13. Differential Non-linearity (DNL)



- Quantization Error: Due to the quantization of the input voltage into a finite number of codes, a range of input voltages (1 LSB wide) will code to the same value. Always ±0.5 LSB.
- Absolute accuracy: The maximum deviation of an actual (unadjusted) transition compared to an ideal transition for any code. This is the compound effect of offset, gain error, differential error, non-linearity, and quantization error. Ideal value: ±0.5 LSB.



shown below. See Table 25-5 on page 282 for detailed description and mapping of the Extended Fuse byte.

Bit 7 6 5 3 2 0 4 1 Rd FHB7 FHB6 FHB5 FHB4 FHB3 FHB2 FHB1 FHB0

Fuse and Lock bits that are programmed, will be read as zero. Fuse and Lock bits that are unprogrammed, will be read as one.

#### 23.1.4 Preventing Flash Corruption

During periods of low  $V_{CC}$ , the Flash program can be corrupted because the supply voltage is too low for the CPU and the Flash to operate properly. These issues are the same as for board level systems using the Flash, and the same design solutions should be applied.

A Flash program corruption can be caused by two situations when the voltage is too low. First, a regular write sequence to the Flash requires a minimum voltage to operate correctly. Secondly, the CPU itself can execute instructions incorrectly, if the supply voltage for executing instructions is too low.

Flash corruption can easily be avoided by following these design recommendations (one is sufficient):

- Keep the AVR RESET active (low) during periods of insufficient power supply voltage. This can be done by enabling the internal Brown-out Detector (BOD) if the operating voltage matches the detection level. If not, an external low V<sub>CC</sub> reset protection circuit can be used. If a reset occurs while a write operation is in progress, the write operation will be completed provided that the power supply voltage is sufficient.
- Keep the AVR core in Power-down sleep mode during periods of low V<sub>CC</sub>. This will prevent the CPU from attempting to decode and execute instructions, effectively protecting the SPMCSR Register and thus the Flash from unintentional writes.

#### 23.1.5 Programming Time for Flash when Using SPM

The calibrated RC Oscillator is used to time Flash accesses. Table 24-5 shows the typical programming time for Flash accesses from the CPU.

#### Table 23-1. SPM Programming Time

Symbol	Min Programming Time	Max Programming Time
Flash write (Page Erase, Page Write, and write Lock bits by SPM)	3.7 ms	4.5 ms

#### 23.1.6 Simple Assembly Code Example for a Boot Loader

Note that the RWWSB bit will always be read as zero in ATmega48. Nevertheless, it is recommended to check this bit as shown in the code example, to ensure compatibility with devices supporting Read-While-Write.





Figure 27-8. Idle Supply Current vs. Frequency (1 - 24 MHz)









Figure 27-23. I/O Pin Source Current vs. Output Voltage (V<sub>CC</sub> = 2.7V)



I/O PIN SOURCE CURRENT vs. OUTPUT VOLTAGE  $V_{\rm CC}$  = 2.7V







Figure 27-29. I/O Pin Input Threshold Voltage vs. V<sub>CC</sub> (VIL, I/O Pin Read As '0')







## 29. Instruction Set Summary

Mnemonics	Operands	Description	Operation	Flags	#Clocks
ARITHMETIC AND L	OGIC INSTRUCTIONS	S	·	•	•
ADD	Rd, Rr	Add two Registers	$Rd \leftarrow Rd + Rr$	Z,C,N,V,H	1
ADC	Rd, Rr	Add with Carry two Registers	$Rd \leftarrow Rd + Rr + C$	Z,C,N,V,H	1
ADIW	Rdl,K	Add Immediate to Word	Rdh:Rdl ← Rdh:Rdl + K	Z,C,N,V,S	2
SUB	Rd, Rr	Subtract two Registers	$Rd \leftarrow Rd - Rr$	Z,C,N,V,H	1
SUBI	Rd, K	Subtract Constant from Register	$Rd \leftarrow Rd - K$	Z,C,N,V,H	1
SBC	Rd, Rr	Subtract with Carry two Registers	$Rd \gets Rd - Rr - C$	Z,C,N,V,H	1
SBCI	Rd, K	Subtract with Carry Constant from Reg.	$Rd \gets Rd - K - C$	Z,C,N,V,H	1
SBIW	Rdl,K	Subtract Immediate from Word	Rdh:Rdl ← Rdh:Rdl - K	Z,C,N,V,S	2
AND	Rd, Rr	Logical AND Registers	$Rd \leftarrow Rd \bullet Rr$	Z,N,V	1
ANDI	Rd, K	Logical AND Register and Constant	$Rd \leftarrow Rd \bullet K$	Z,N,V	1
OR	Rd, Rr	Logical OR Registers	$Rd \leftarrow Rd v Rr$	Z,N,V	1
ORI	Rd, K	Logical OR Register and Constant	$Rd \leftarrow Rd \lor K$	Z,N,V	1
EOR	Rd, Rr	Exclusive OR Registers	$Rd \leftarrow Rd \oplus Rr$	Z,N,V	1
COM	Rd			Z,C,N,V	1
NEG	Rd	Two's Complement		Z,C,N,V,H	1
SBR	Rd,K	Set Bit(s) in Register		Z,N,V	1
CBR	Ru,r.	Clear Bill(s) in Register	$Rd \leftarrow Rd \bullet (0XFF - K)$	Z,IN,V	1
	Rd	Degrament		Z,IN,V	1
Tet	Rd	Toot for Zoro or Minus			1
	Ru Rd	Clear Register	$Ru \leftarrow Ru \bullet Ru$		1
SEB	Bd	Set Begister		None	1
MUI	Rd Br	Multinly Unsigned	$B1:B0 \leftarrow Bd \times Br$	ZC	2
MULS	Bd Br	Multiply Signed	$B1:B0 \leftarrow Bd \times Br$	7.0	2
MULSU	Bd Br	Multiply Signed with Unsigned	$B1:B0 \leftarrow Bd \times Br$	Z,0	2
FMUL	Rd, Rr	Fractional Multiply Unsigned	$B1:B0 \leftarrow (Bd \times Br) << 1$	Z,C	2
FMULS	Rd. Br	Eractional Multiply Signed	$B1:B0 \leftarrow (Bd \times Br) \le 1$	Z.C	2
FMULSU	Rd. Rr	Fractional Multiply Signed with Unsigned	$B1:B0 \leftarrow (Bd \times Br) << 1$	Z.C	2
BRANCH INSTRUCT	TIONS				•
RJMP	k	Relative Jump	$PC \leftarrow PC + k + 1$	None	2
IJMP		Indirect Jump to (Z)	PC ← Z	None	2
JMP <sup>(1)</sup>	k	Direct Jump	$PC \leftarrow k$	None	3
RCALL	k	Relative Subroutine Call	$PC \leftarrow PC + k + 1$	None	3
ICALL		Indirect Call to (Z)	$PC \leftarrow Z$	None	3
CALL <sup>(1)</sup>	k	Direct Subroutine Call	PC ← k	None	4
RET		Subroutine Return	$PC \leftarrow STACK$	None	4
RETI		Interrupt Return	$PC \leftarrow STACK$	1	4
CPSE	Rd,Rr	Compare, Skip if Equal	if $(Rd = Rr) PC \leftarrow PC + 2 \text{ or } 3$	None	1/2/3
CP	Rd,Rr	Compare	Rd – Rr	Z, N,V,C,H	1
CPC	Rd,Rr	Compare with Carry	Rd – Rr – C	Z, N,V,C,H	1
CPI	Rd,K	Compare Register with Immediate	Rd – K	Z, N,V,C,H	1
SBRC	Rr, b	Skip if Bit in Register Cleared	if (Rr(b)=0) PC ← PC + 2 or 3	None	1/2/3
SBRS	Rr, b	Skip if Bit in Register is Set	if $(\text{Rr}(b)=1) \text{PC} \leftarrow \text{PC} + 2 \text{ or } 3$	None	1/2/3
SBIC	P, b	Skip if Bit in I/O Register Cleared	if (P(b)=0) PC ← PC + 2 or 3	None	1/2/3
SBIS	P, b	Skip it Bit in I/O Register is Set	if $(P(b)=1) PC \leftarrow PC + 2 \text{ or } 3$	None	1/2/3
BRBS	s, k	Branch if Status Flag Set	If $(SHEG(s) = 1)$ then $PC \leftarrow PC + k + 1$	None	1/2
BRBC	S, K	Branch if Status Flag Cleared	If $(SREG(s) = 0)$ then $PC \leftarrow PC + k + 1$	None	1/2
BREQ	ĸ	Branch if Equal	If $(Z = 1)$ then PC $\leftarrow$ PC + k + 1	None	1/2
BRINE	ĸ	Branch if Not Equal	If $(2 = 0)$ then PC $\leftarrow$ PC + k + 1	None	1/2
BRCS	ĸ	Branch if Carry Set	If $(C = 1)$ then $PC \leftarrow PC + k + 1$	None	1/2
BRCC	ĸ	Branch if Carry Cleared	If $(C = 0)$ then $PC \leftarrow PC + k + 1$	None	1/2
BRSH	ĸ	Branch II Same of Higher	If $(C = 0)$ then PC $\leftarrow$ PC + k + 1 if $(C = 1)$ then PC $\leftarrow$ PC + k + 1	None	1/2
BRLU	ĸ	Branch if Minus	if $(N = 1)$ then PC $\leftarrow$ PC + k + 1	None	1/2
BRPI	k	Branch if Plus	if $(N = 0)$ then $PC \leftarrow PC + K + 1$	None	1/2
BRGE	k	Branch if Greater or Found Signed	if $(N \oplus V = 0)$ then PC $\leftarrow$ PC $\pm k \pm 1$	None	1/2
BRIT	k	Branch if Less Than Zero, Signed	if $(N \oplus V = 1)$ then PC $\leftarrow$ PC + k + 1	None	1/2
BBHS	k	Branch if Half Carry Flag Set	if (H = 1) then PC $\leftarrow$ PC + k + 1	None	1/2
BRHC	k	Branch if Half Carry Flag Cleared	if (H = 0) then PC $\leftarrow$ PC + k + 1	None	1/2
BRTS	k	Branch if T Flag Set	if (T = 1) then PC $\leftarrow$ PC + k + 1	None	1/2
BRTC	k	Branch if T Flag Cleared	if $(T = 0)$ then PC $\leftarrow$ PC + k + 1	None	1/2
BRVS	k	Branch if Overflow Flag is Set	if (V = 1) then PC $\leftarrow$ PC + k + 1	None	1/2
BRVC	k	Branch if Overflow Flag is Cleared	if (V = 0) then PC $\leftarrow$ PC + k + 1	None	1/2