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Details

Product Status	Obsolete
Core Processor	AVR
Core Size	8-Bit
Speed	10MHz
Connectivity	I ² C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	23
Program Memory Size	4KB (2K x 16)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	512 x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 5.5V
Data Converters	A/D 8x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	32-TQFP
Supplier Device Package	32-TQFP (7x7)
Purchase URL	https://www.e-xfl.com/product-detail/atmel/atmega48v-10ai

purpose, it is recommended to write the Sleep Enable (SE) bit to one just before the execution of the SLEEP instruction and to clear it immediately after waking up.

7.1 Idle Mode

When the SM2..0 bits are written to 000, the SLEEP instruction makes the MCU enter Idle mode, stopping the CPU but allowing the SPI, USART, Analog Comparator, ADC, 2-wire Serial Interface, Timer/Counters, Watchdog, and the interrupt system to continue operating. This sleep mode basically halts clk_{CPU} and $\text{clk}_{\text{FLASH}}$, while allowing the other clocks to run.

Idle mode enables the MCU to wake up from external triggered interrupts as well as internal ones like the Timer Overflow and USART Transmit Complete interrupts. If wake-up from the Analog Comparator interrupt is not required, the Analog Comparator can be powered down by setting the ACD bit in the Analog Comparator Control and Status Register – ACSR. This will reduce power consumption in Idle mode. If the ADC is enabled, a conversion starts automatically when this mode is entered.

7.2 ADC Noise Reduction Mode

When the SM2..0 bits are written to 001, the SLEEP instruction makes the MCU enter ADC Noise Reduction mode, stopping the CPU but allowing the ADC, the external interrupts, the 2-wire Serial Interface address watch, Timer/Counter2, and the Watchdog to continue operating (if enabled). This sleep mode basically halts $\text{clk}_{\text{I/O}}$, clk_{CPU} , and $\text{clk}_{\text{FLASH}}$, while allowing the other clocks to run.

This improves the noise environment for the ADC, enabling higher resolution measurements. If the ADC is enabled, a conversion starts automatically when this mode is entered. Apart from the ADC Conversion Complete interrupt, only an External Reset, a Watchdog System Reset, a Watchdog Interrupt, a Brown-out Reset, a 2-wire Serial Interface address match, a Timer/Counter2 interrupt, an SPM/EEPROM ready interrupt, an external level interrupt on INT0 or INT1 or a pin change interrupt can wake up the MCU from ADC Noise Reduction mode.

7.3 Power-down Mode

When the SM2..0 bits are written to 010, the SLEEP instruction makes the MCU enter Power-down mode. In this mode, the external Oscillator is stopped, while the external interrupts, the 2-wire Serial Interface address watch, and the Watchdog continue operating (if enabled). Only an External Reset, a Watchdog System Reset, a Watchdog Interrupt, a Brown-out Reset, a 2-wire Serial Interface address match, an external level interrupt on INT0 or INT1, or a pin change interrupt can wake up the MCU. This sleep mode basically halts all generated clocks, allowing operation of asynchronous modules only.

Note that if a level triggered interrupt is used for wake-up from Power-down mode, the changed level must be held for some time to wake up the MCU. Refer to ["External Interrupts" on page 83](#) for details.

When waking up from Power-down mode, there is a delay from the wake-up condition occurs until the wake-up becomes effective. This allows the clock to restart and become stable after having been stopped. The wake-up period is defined by the same CKSEL Fuses that define the Reset Time-out period, as described in ["Clock Sources" on page 26](#).

7.4 Power-save Mode

When the SM2..0 bits are written to 011, the SLEEP instruction makes the MCU enter Power-save mode. This mode is identical to Power-down, with one exception:

Table 10-7 and Table 10-8 relate the alternate functions of Port C to the overriding signals shown in Figure 10-5 on page 69.

Table 10-7. Overriding Signals for Alternate Functions in PC6..PC4⁽¹⁾

Signal Name	PC6/ $\overline{\text{RESET}}$ /PCINT14	PC5/SCL/ADC5/PCINT13	PC4/SDA/ADC4/PCINT12
PUOE	RSTDISBL	TWEN	TWEN
PUOV	1	PORTC5 • $\overline{\text{PUD}}$	PORTC4 • $\overline{\text{PUD}}$
DDOE	RSTDISBL	TWEN	TWEN
DDOV	0	SCL_OUT	SDA_OUT
PVOE	0	TWEN	TWEN
PVOV	0	0	0
DIEOE	RSTDISBL + PCINT14 • PCIE1	PCINT13 • PCIE1 + ADC5D	PCINT12 • PCIE1 + ADC4D
DIEOV	RSTDISBL	PCINT13 • PCIE1	PCINT12 • PCIE1
DI	PCINT14 INPUT	PCINT13 INPUT	PCINT12 INPUT
AIO	RESET INPUT	ADC5 INPUT / SCL INPUT	ADC4 INPUT / SDA INPUT

Note: 1. When enabled, the 2-wire Serial Interface enables slew-rate controls on the output pins PC4 and PC5. This is not shown in the figure. In addition, spike filters are connected between the AIO outputs shown in the port figure and the digital logic of the TWI module.

Table 10-8. Overriding Signals for Alternate Functions in PC3..PC0

Signal Name	PC3/ADC3/PCINT11	PC2/ADC2/PCINT10	PC1/ADC1/PCINT9	PC0/ADC0/PCINT8
PUOE	0	0	0	0
PUOV	0	0	0	0
DDOE	0	0	0	0
DDOV	0	0	0	0
PVOE	0	0	0	0
PVOV	0	0	0	0
DIEOE	PCINT11 • PCIE1 + ADC3D	PCINT10 • PCIE1 + ADC2D	PCINT9 • PCIE1 + ADC1D	PCINT8 • PCIE1 + ADC0D
DIEOV	PCINT11 • PCIE1	PCINT10 • PCIE1	PCINT9 • PCIE1	PCINT8 • PCIE1
DI	PCINT11 INPUT	PCINT10 INPUT	PCINT9 INPUT	PCINT8 INPUT
AIO	ADC3 INPUT	ADC2 INPUT	ADC1 INPUT	ADC0 INPUT

- **T1/OC0B/PCINT21 – Port D, Bit 5**

T1, Timer/Counter1 counter source.

OC0B, Output Compare Match output: The PD5 pin can serve as an external output for the Timer/Counter0 Compare Match B. The PD5 pin has to be configured as an output (DDD5 set (one)) to serve this function. The OC0B pin is also the output pin for the PWM mode timer function.

PCINT21: Pin Change Interrupt source 21. The PD5 pin can serve as an external interrupt source.

- **XCK/T0/PCINT20 – Port D, Bit 4**

XCK, USART external clock.

T0, Timer/Counter0 counter source.

PCINT20: Pin Change Interrupt source 20. The PD4 pin can serve as an external interrupt source.

- **INT1/OC2B/PCINT19 – Port D, Bit 3**

INT1, External Interrupt source 1: The PD3 pin can serve as an external interrupt source.

OC2B, Output Compare Match output: The PD3 pin can serve as an external output for the Timer/Counter0 Compare Match B. The PD3 pin has to be configured as an output (DDD3 set (one)) to serve this function. The OC2B pin is also the output pin for the PWM mode timer function.

PCINT19: Pin Change Interrupt source 19. The PD3 pin can serve as an external interrupt source.

- **INT0/PCINT18 – Port D, Bit 2**

INT0, External Interrupt source 0: The PD2 pin can serve as an external interrupt source.

PCINT18: Pin Change Interrupt source 18. The PD2 pin can serve as an external interrupt source.

- **TXD/PCINT17 – Port D, Bit 1**

TXD, Transmit Data (Data output pin for the USART). When the USART Transmitter is enabled, this pin is configured as an output regardless of the value of DDD1.

PCINT17: Pin Change Interrupt source 17. The PD1 pin can serve as an external interrupt source.

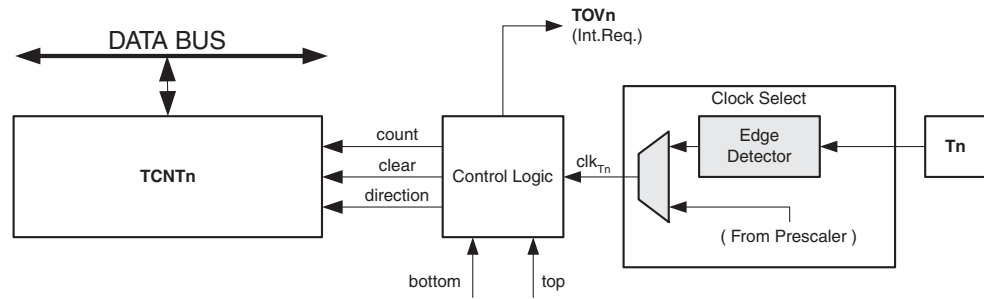
- **RXD/PCINT16 – Port D, Bit 0**

RXD, Receive Data (Data input pin for the USART). When the USART Receiver is enabled this pin is configured as an input regardless of the value of DDD0. When the USART forces this pin to be an input, the pull-up can still be controlled by the PORTD0 bit.

PCINT16: Pin Change Interrupt source 16. The PD0 pin can serve as an external interrupt source.

[Table 10-10](#) and [Table 10-11](#) relate the alternate functions of Port D to the overriding signals shown in [Figure 10-5](#) on page 69.

Figure 12-2. Counter Unit Block Diagram



Signal description (internal signals):

count	Increment or decrement TCNT0 by 1.
direction	Select between increment and decrement.
clear	Clear TCNT0 (set all bits to zero).
clk_{Tn}	Timer/Counter clock, referred to as clk _{T0} in the following.
top	Signalize that TCNT0 has reached maximum value.
bottom	Signalize that TCNT0 has reached minimum value (zero).

Depending of the mode of operation used, the counter is cleared, incremented, or decremented at each timer clock (clk_{T0}). clk_{T0} can be generated from an external or internal clock source, selected by the Clock Select bits (CS02:0). When no clock source is selected (CS02:0 = 0) the timer is stopped. However, the TCNT0 value can be accessed by the CPU, regardless of whether clk_{T0} is present or not. A CPU write overrides (has priority over) all counter clear or count operations.

The counting sequence is determined by the setting of the WGM01 and WGM00 bits located in the Timer/Counter Control Register (TCCR0A) and the WGM02 bit located in the Timer/Counter Control Register B (TCCR0B). There are close connections between how the counter behaves (counts) and how waveforms are generated on the Output Compare outputs OC0A and OC0B. For more details about advanced counting sequences and waveform generation, see ["Modes of Operation" on page 93](#).

The Timer/Counter Overflow Flag (TOV0) is set according to the mode of operation selected by the WGM02:0 bits. TOV0 can be used for generating a CPU interrupt.

12.4 Output Compare Unit

The 8-bit comparator continuously compares TCNT0 with the Output Compare Registers (OCR0A and OCR0B). Whenever TCNT0 equals OCR0A or OCR0B, the comparator signals a match. A match will set the Output Compare Flag (OCF0A or OCF0B) at the next timer clock cycle. If the corresponding interrupt is enabled, the Output Compare Flag generates an Output Compare interrupt. The Output Compare Flag is automatically cleared when the interrupt is executed. Alternatively, the flag can be cleared by software by writing a logical one to its I/O bit location. The Waveform Generator uses the match signal to generate an output according to operating mode set by the WGM02:0 bits and Compare Output mode (COM0x1:0) bits. The max and bottom signals are used by the Waveform Generator for handling the special cases of the extreme values in some modes of operation (["Modes of Operation" on page 93](#)).

[Figure 12-3](#) shows a block diagram of the Output Compare unit.

Assembly Code Example⁽¹⁾

```
TIM16_WriteTCNT1:
    ; Save global interrupt flag
    in r18,SREG
    ; Disable interrupts
    cli
    ; Set TCNT1 to r17:r16
    out TCNT1H,r17
    out TCNT1L,r16
    ; Restore global interrupt flag
    out SREG,r18
    ret
```

C Code Example⁽¹⁾

```
void TIM16_WriteTCNT1( unsigned int i )
{
    unsigned char sreg;
    unsigned int i;
    /* Save global interrupt flag */
    sreg = SREG;
    /* Disable interrupts */
    _CLI();
    /* Set TCNT1 to i */
    TCNT1 = i;
    /* Restore global interrupt flag */
    SREG = sreg;
}
```

Note: 1. See "About Code Examples" on page 6.
For I/O Registers located in extended I/O map, "IN", "OUT", "SBIS", "SBIC", "CBI", and "SBI" instructions must be replaced with instructions that allow access to extended I/O. Typically "LDS" and "STS" combined with "SBR", "SBRC", "SBR", and "CBR".

The assembly code example requires that the r17:r16 register pair contains the value to be written to TCNT1.

13.2.1 Reusing the Temporary High Byte Register

If writing to more than one 16-bit register where the high byte is the same for all registers written, then the high byte only needs to be written once. However, note that the same rule of atomic operation described previously also applies in this case.

13.3 Timer/Counter Clock Sources

The Timer/Counter can be clocked by an internal or an external clock source. The clock source is selected by the Clock Select logic which is controlled by the *Clock Select* (CS12:0) bits located in the *Timer/Counter control Register B* (TCCR1B). For details on clock sources and prescaler, see "Timer/Counter0 and Timer/Counter1 Prescalers" on page 135.

When the ICR1 is used as TOP value (see description of the WGM13:0 bits located in the TCCR1A and the TCCR1B Register), the ICP1 is disconnected and consequently the Input Capture function is disabled.

- **Bit 5 – Reserved Bit**

This bit is reserved for future use. For ensuring compatibility with future devices, this bit must be written to zero when TCCR1B is written.

- **Bit 4:3 – WGM13:2: Waveform Generation Mode**

See TCCR1A Register description.

- **Bit 2:0 – CS12:0: Clock Select**

The three Clock Select bits select the clock source to be used by the Timer/Counter, see [Figure 13-10](#) and [Figure 13-11](#).

Table 13-5. Clock Select Bit Description

CS12	CS11	CS10	Description
0	0	0	No clock source (Timer/Counter stopped).
0	0	1	$clk_{I/O}/1$ (No prescaling)
0	1	0	$clk_{I/O}/8$ (From prescaler)
0	1	1	$clk_{I/O}/64$ (From prescaler)
1	0	0	$clk_{I/O}/256$ (From prescaler)
1	0	1	$clk_{I/O}/1024$ (From prescaler)
1	1	0	External clock source on T1 pin. Clock on falling edge.
1	1	1	External clock source on T1 pin. Clock on rising edge.

If external pin modes are used for the Timer/Counter1, transitions on the T1 pin will clock the counter even if the pin is configured as an output. This feature allows software control of the counting.

13.10.3 Timer/Counter1 Control Register C – TCCR1C

Bit	7	6	5	4	3	2	1	0	
	FOC1A	FOC1B	–	–	–	–	–	–	TCCR1C
Read/Write	R/W	R/W	R	R	R	R	R	R	
Initial Value	0	0	0	0	0	0	0	0	

- **Bit 7 – FOC1A: Force Output Compare for Channel A**

- **Bit 6 – FOC1B: Force Output Compare for Channel B**

The FOC1A/FOC1B bits are only active when the WGM13:0 bits specifies a non-PWM mode. However, for ensuring compatibility with future devices, these bits must be set to zero when TCCR1A is written when operating in a PWM mode. When writing a logical one to the FOC1A/FOC1B bit, an immediate compare match is forced on the Waveform Generation unit. The OC1A/OC1B output is changed according to its COM1x1:0 bits setting. Note that the FOC1A/FOC1B bits are implemented as strobes. Therefore it is the value present in the COM1x1:0 bits that determine the effect of the forced compare.

The following code example shows a simple USART receive function based on polling of the Receive Complete (RXCn) Flag. When using frames with less than eight bits the most significant bits of the data read from the UDRn will be masked to zero. The USART has to be initialized before the function can be used.

Assembly Code Example⁽¹⁾

```
USART_Receive:
    ; Wait for data to be received
    sbis UCSRnA, RXCn
    rjmp USART_Receive
    ; Get and return received data from buffer
    in    r16, UDRn
    ret
```

C Code Example⁽¹⁾

```
unsigned char USART_Receive( void )
{
    /* Wait for data to be received */
    while ( !(UCSRnA & (1<<RXCn)) )
        ;
    /* Get and return received data from buffer */
    return UDRn;
}
```

Note: 1. See ["About Code Examples" on page 6](#).

For I/O Registers located in extended I/O map, "IN", "OUT", "SBIS", "SBIC", "CBI", and "SBI" instructions must be replaced with instructions that allow access to extended I/O. Typically "LDS" and "STS" combined with "SBR", "SBR", "SBR", and "CBR".

The function simply waits for data to be present in the receive buffer by checking the RXCn Flag, before reading the buffer and returning the value.

17.6.2 Receiving Frames with 9 Data Bits

If 9-bit characters are used (UCSZn=7) the ninth bit must be read from the RXB8n bit in UCSRnB **before** reading the low bits from the UDRn. This rule applies to the FEn, DORn and UPEn Status Flags as well. Read status from UCSRnA, then data from UDRn. Reading the UDRn I/O location will change the state of the receive buffer FIFO and consequently the TXB8n, FEn, DORn and UPEn bits, which all are stored in the FIFO, will change.

The following code example shows a simple USART receive function that handles both nine bit characters and the status bits.

UDREN is set after a reset to indicate that the Transmitter is ready.

- **Bit 4 – FEn: Frame Error**

This bit is set if the next character in the receive buffer had a Frame Error when received. I.e., when the first stop bit of the next character in the receive buffer is zero. This bit is valid until the receive buffer (UDRn) is read. The FEn bit is zero when the stop bit of received data is one. Always set this bit to zero when writing to UCSRnA.

- **Bit 3 – DORn: Data OverRun**

This bit is set if a Data OverRun condition is detected. A Data OverRun occurs when the receive buffer is full (two characters), it is a new character waiting in the Receive Shift Register, and a new start bit is detected. This bit is valid until the receive buffer (UDRn) is read. Always set this bit to zero when writing to UCSRnA.

- **Bit 2 – UPEn: USART Parity Error**

This bit is set if the next character in the receive buffer had a Parity Error when received and the Parity Checking was enabled at that point (UPMn1 = 1). This bit is valid until the receive buffer (UDRn) is read. Always set this bit to zero when writing to UCSRnA.

- **Bit 1 – U2Xn: Double the USART Transmission Speed**

This bit only has effect for the asynchronous operation. Write this bit to zero when using synchronous operation.

Writing this bit to one will reduce the divisor of the baud rate divider from 16 to 8 effectively doubling the transfer rate for asynchronous communication.

- **Bit 0 – MPCMn: Multi-processor Communication Mode**

This bit enables the Multi-processor Communication mode. When the MPCMn bit is written to one, all the incoming frames received by the USART Receiver that do not contain address information will be ignored. The Transmitter is unaffected by the MPCMn setting. For more detailed information see ["Multi-processor Communication Mode" on page 185](#).

17.9.3 USART Control and Status Register n B – UCSRnB

Bit	7	6	5	4	3	2	1	0	
	RXCIE _n	TXCIE _n	UDRIE _n	RXEN _n	TXEN _n	UCSZ _{n2}	RXB8 _n	TXB8 _n	UCSRnB
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R	R/W	
Initial Value	0	0	0	0	0	0	0	0	

- **Bit 7 – RXCIE_n: RX Complete Interrupt Enable n**

Writing this bit to one enables interrupt on the RXC_n Flag. A USART Receive Complete interrupt will be generated only if the RXCIE_n bit is written to one, the Global Interrupt Flag in SREG is written to one and the RXC_n bit in UCSRnA is set.

- **Bit 6 – TXCIE_n: TX Complete Interrupt Enable n**

Writing this bit to one enables interrupt on the TXC_n Flag. A USART Transmit Complete interrupt will be generated only if the TXCIE_n bit is written to one, the Global Interrupt Flag in SREG is written to one and the TXC_n bit in UCSRnA is set.

Table 17-9. Examples of UBRRn Settings for Commonly Used Oscillator Frequencies

Baud Rate (bps)	$f_{osc} = 1.0000 \text{ MHz}$				$f_{osc} = 1.8432 \text{ MHz}$				$f_{osc} = 2.0000 \text{ MHz}$			
	U2Xn = 0		U2Xn = 1		U2Xn = 0		U2Xn = 1		U2Xn = 0		U2Xn = 1	
	UBRR n	Error	UBRR n	Error	UBRR n	Error	UBRR n	Error	UBRR n	Error	UBRR n	Error
2400	25	0.2%	51	0.2%	47	0.0%	95	0.0%	51	0.2%	103	0.2%
4800	12	0.2%	25	0.2%	23	0.0%	47	0.0%	25	0.2%	51	0.2%
9600	6	-7.0%	12	0.2%	11	0.0%	23	0.0%	12	0.2%	25	0.2%
14.4k	3	8.5%	8	-3.5%	7	0.0%	15	0.0%	8	-3.5%	16	2.1%
19.2k	2	8.5%	6	-7.0%	5	0.0%	11	0.0%	6	-7.0%	12	0.2%
28.8k	1	8.5%	3	8.5%	3	0.0%	7	0.0%	3	8.5%	8	-3.5%
38.4k	1	-18.6%	2	8.5%	2	0.0%	5	0.0%	2	8.5%	6	-7.0%
57.6k	0	8.5%	1	8.5%	1	0.0%	3	0.0%	1	8.5%	3	8.5%
76.8k	—	—	1	-18.6%	1	-25.0%	2	0.0%	1	-18.6%	2	8.5%
115.2k	—	—	0	8.5%	0	0.0%	1	0.0%	0	8.5%	1	8.5%
230.4k	—	—	—	—	—	—	0	0.0%	—	—	—	—
250k	—	—	—	—	—	—	—	—	—	—	0	0.0%
Max. ⁽¹⁾	62.5 kbps		125 kbps		115.2 kbps		230.4 kbps		125 kbps		250 kbps	

Note: 1. UBRRn = 0, Error = 0.0%

19. 2-wire Serial Interface

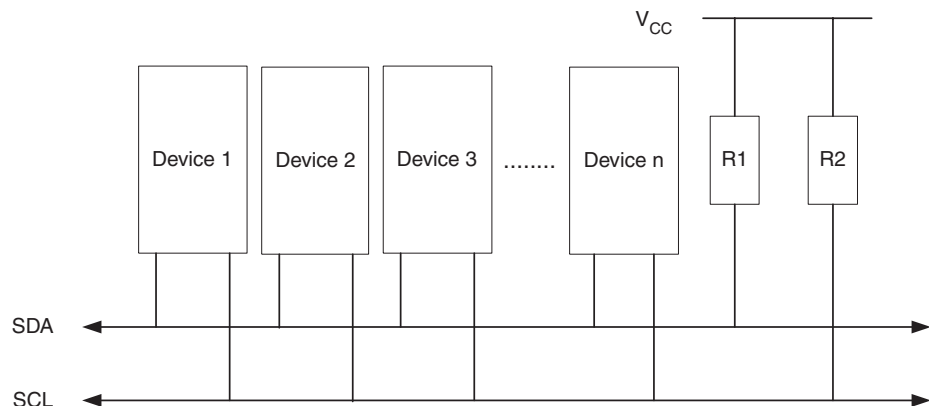
19.1 Features

- Simple Yet Powerful and Flexible Communication Interface, only two Bus Lines Needed
- Both Master and Slave Operation Supported
- Device can Operate as Transmitter or Receiver
- 7-bit Address Space Allows up to 128 Different Slave Addresses
- Multi-master Arbitration Support
- Up to 400 kHz Data Transfer Speed
- Slew-rate Limited Output Drivers
- Noise Suppression Circuitry Rejects Spikes on Bus Lines
- Fully Programmable Slave Address with General Call Support
- Address Recognition Causes Wake-up When AVR is in Sleep Mode

19.2 2-wire Serial Interface Bus Definition

The 2-wire Serial Interface (TWI) is ideally suited for typical microcontroller applications. The TWI protocol allows the systems designer to interconnect up to 128 different devices using only two bi-directional bus lines, one for clock (SCL) and one for data (SDA). The only external hardware needed to implement the bus is a single pull-up resistor for each of the TWI bus lines. All devices connected to the bus have individual addresses, and mechanisms for resolving bus contention are inherent in the TWI protocol.

Figure 19-1. TWI Bus Interconnection



19.2.1 TWI Terminology

The following definitions are frequently encountered in this section.

Table 19-1. TWI Terminology

Term	Description
Master	The device that initiates and terminates a transmission. The Master also generates the SCL clock.
Slave	The device addressed by a Master.
Transmitter	The device placing data on the bus.
Receiver	The device reading data from the bus.

21.4.1 ADC Input Channels

When changing channel selections, the user should observe the following guidelines to ensure that the correct channel is selected:

In Single Conversion mode, always select the channel before starting the conversion. The channel selection may be changed one ADC clock cycle after writing one to ADSC. However, the simplest method is to wait for the conversion to complete before changing the channel selection.

In Free Running mode, always select the channel before starting the first conversion. The channel selection may be changed one ADC clock cycle after writing one to ADSC. However, the simplest method is to wait for the first conversion to complete, and then change the channel selection. Since the next conversion has already started automatically, the next result will reflect the previous channel selection. Subsequent conversions will reflect the new channel selection.

21.4.2 ADC Voltage Reference

The reference voltage for the ADC (V_{REF}) indicates the conversion range for the ADC. Single ended channels that exceed V_{REF} will result in codes close to 0x3FF. V_{REF} can be selected as either AV_{CC} , internal 1.1V reference, or external AREF pin.

AV_{CC} is connected to the ADC through a passive switch. The internal 1.1V reference is generated from the internal bandgap reference (V_{BG}) through an internal amplifier. In either case, the external AREF pin is directly connected to the ADC, and the reference voltage can be made more immune to noise by connecting a capacitor between the AREF pin and ground. V_{REF} can also be measured at the AREF pin with a high impedant voltmeter. Note that V_{REF} is a high impedant source, and only a capacitive load should be connected in a system.

If the user has a fixed voltage source connected to the AREF pin, the user may not use the other reference voltage options in the application, as they will be shorted to the external voltage. If no external voltage is applied to the AREF pin, the user may switch between AV_{CC} and 1.1V as reference selection. The first ADC conversion result after switching reference voltage source may be inaccurate, and the user is advised to discard this result.

21.5 ADC Noise Canceler

The ADC features a noise canceler that enables conversion during sleep mode to reduce noise induced from the CPU core and other I/O peripherals. The noise canceler can be used with ADC Noise Reduction and Idle mode. To make use of this feature, the following procedure should be used:

- a. Make sure that the ADC is enabled and is not busy converting. Single Conversion mode must be selected and the ADC conversion complete interrupt must be enabled.
- b. Enter ADC Noise Reduction mode (or Idle mode). The ADC will start a conversion once the CPU has been halted.
- c. If no other interrupts occur before the ADC conversion completes, the ADC interrupt will wake up the CPU and execute the ADC Conversion Complete interrupt routine. If another interrupt wakes up the CPU before the ADC conversion is complete, that interrupt will be executed, and an ADC Conversion Complete interrupt request will be generated when the ADC conversion completes. The CPU will remain in active mode until a new sleep command is executed.

Note that the ADC will not be automatically turned off when entering other sleep modes than Idle mode and ADC Noise Reduction mode. The user is advised to write zero to ADEN before entering such sleep modes to avoid excessive power consumption.

```

    ; return to RWW section
    ; verify that RWW section is safe to read
Return:
    in    temp1, SPMCSR
    sbrs temp1, RWWSB      ; If RWWSB is set, the RWW section is not ready yet
    ret
    ; re-enable the RWW section
    ldi   spmcrval, (1<<RWWSRE) | (1<<SELFPRGEN)
    rcall Do_spm
    rjmp  Return

Do_spm:
    ; check for previous SPM complete
Wait_spm:
    in    temp1, SPMCSR
    sbrc temp1, SELFPRGEN
    rjmp Wait_spm
    ; input: spmcrval determines SPM action
    ; disable interrupts if enabled, store status
    in    temp2, SREG
    cli
    ; check that no EEPROM write access is present
Wait_ee:
    sbic EECR, EEPE
    rjmp Wait_ee
    ; SPM timed sequence
    out   SPMCSR, spmcrval
    spm
    ; restore SREG (to enable interrupts if originally enabled)
    out   SREG, temp2
    ret

```

the RWWSB by writing the RWWSRE. See ["Simple Assembly Code Example for a Boot Loader" on page 275](#) for an example.

24.7.7 Setting the Boot Loader Lock Bits by SPM

To set the Boot Loader Lock bits, write the desired data to R0, write "X0001001" to SPMCSR and execute SPM within four clock cycles after writing SPMCSR. The only accessible Lock bits are the Boot Lock bits that may prevent the Application and Boot Loader section from any software update by the MCU.

Bit	7	6	5	4	3	2	1	0
R0	1	1	BLB12	BLB11	BLB02	BLB01	1	1

See [Table 24-2](#) and [Table 24-3](#) for how the different settings of the Boot Loader bits affect the Flash access.

If bits 5..2 in R0 are cleared (zero), the corresponding Boot Lock bit will be programmed if an SPM instruction is executed within four cycles after BLBSET and SELFPRGEN are set in SPMCSR. The Z-pointer is don't care during this operation, but for future compatibility it is recommended to load the Z-pointer with 0x0001 (same as used for reading the IO_{ck} bits). For future compatibility it is also recommended to set bits 7, 6, 1, and 0 in R0 to "1" when writing the Lock bits. When programming the Lock bits the entire Flash can be read during the operation.

24.7.8 EEPROM Write Prevents Writing to SPMCSR

Note that an EEPROM write operation will block all software programming to Flash. Reading the Fuses and Lock bits from software will also be prevented during the EEPROM write operation. It is recommended that the user checks the status bit (EEPE) in the EECR Register and verifies that the bit is cleared before writing to the SPMCSR Register.

24.7.9 Reading the Fuse and Lock Bits from Software

It is possible to read both the Fuse and Lock bits from software. To read the Lock bits, load the Z-pointer with 0x0001 and set the BLBSET and SELFPRGEN bits in SPMCSR. When an LPM instruction is executed within three CPU cycles after the BLBSET and SELFPRGEN bits are set in SPMCSR, the value of the Lock bits will be loaded in the destination register. The BLBSET and SELFPRGEN bits will auto-clear upon completion of reading the Lock bits or if no LPM instruction is executed within three CPU cycles or no SPM instruction is executed within four CPU cycles. When BLBSET and SELFPRGEN are cleared, LPM will work as described in the Instruction set Manual.

Bit	7	6	5	4	3	2	1	0
Rd	–	–	BLB12	BLB11	BLB02	BLB01	LB2	LB1

The algorithm for reading the Fuse Low byte is similar to the one described above for reading the Lock bits. To read the Fuse Low byte, load the Z-pointer with 0x0000 and set the BLBSET and SELFPRGEN bits in SPMCSR. When an LPM instruction is executed within three cycles after the BLBSET and SELFPRGEN bits are set in the SPMCSR, the value of the Fuse Low byte (FLB) will be loaded in the destination register as shown below. Refer to [Table 25-5 on page 282](#) for a detailed description and mapping of the Fuse Low byte.

Bit	7	6	5	4	3	2	1	0
Rd	FLB7	FLB6	FLB5	FLB4	FLB3	FLB2	FLB1	FLB0

Similarly, when reading the Fuse High byte, load 0x0003 in the Z-pointer. When an LPM instruction is executed within three cycles after the BLBSET and SELFPRGEN bits are set in the SPMCSR, the value of the Fuse High byte (FHB) will be loaded in the destination register as

Table 25-17. Serial Programming Instruction Set (Continued)

Instruction	Instruction Format				Operation
	Byte 1	Byte 2	Byte 3	Byte4	
Load EEPROM Memory Page (page access)	1100 0001	0000 0000	0000 00 bb	iiii iiii	Load data i to EEPROM memory page buffer. After data is loaded, program EEPROM page.
Write EEPROM Memory Page (page access)	1100 0010	00 xx xxaa	bbbb bb00	xxxx xxxx	Write EEPROM page at address a:b .
Read Lock bits	0101 1000	0000 0000	xxxx xxxx	xx00 0000	Read Lock bits. “0” = programmed, “1” = unprogrammed. See Table 25-1 on page 280 for details.
Write Lock bits	1010 1100	111 x xxxx	xxxx xxxx	11ii iiii	Write Lock bits. Set bits = “0” to program Lock bits. See Table 25-1 on page 280 for details.
Read Signature Byte	0011 0000	000 x xxxx	xxxx xxbb	0000 0000	Read Signature Byte o at address b .
Write Fuse bits	1010 1100	1010 0000	xxxx xxxx	iiii iiii	Set bits = “0” to program, “1” to unprogram. See Table XXX on page XXX for details.
Write Fuse High bits	1010 1100	1010 1000	xxxx xxxx	iiii iiii	Set bits = “0” to program, “1” to unprogram. See Table 21-1 on page 244 for details.
Write Extended Fuse Bits	1010 1100	1010 0100	xxxx xxxx	xxxx xxii	Set bits = “0” to program, “1” to unprogram. See Table 25-4 on page 281 for details.
Read Fuse bits	0101 0000	0000 0000	xxxx xxxx	0000 0000	Read Fuse bits. “0” = programmed, “1” = unprogrammed. See Table XXX on page XXX for details.
Read Fuse High bits	0101 1000	0000 1000	xxxx xxxx	0000 0000	Read Fuse High bits. “0” = programmed, “1” = unprogrammed. See Table 21-1 on page 244 for details.
Read Extended Fuse Bits	0101 0000	0000 1000	xxxx xxxx	0000 0000	Read Extended Fuse bits. “0” = programmed, “1” = unprogrammed. See Table 25-4 on page 281 for details.
Read Calibration Byte	0011 1000	000 x xxxx	0000 0000	0000 0000	Read Calibration Byte
Poll RDY/ $\overline{\text{BSY}}$	1111 0000	0000 0000	xxxx xxxx	xxxx xxxo	If o = “1”, a programming operation is still busy. Wait until this bit returns to “0” before applying another command.

Note: **a** = address high bits, **b** = address low bits, **H** = 0 - Low byte, 1 - High Byte, **o** = data out, **i** = data in, **x** = don't care

25.9.2 SPI Serial Programming Characteristics

For characteristics of the SPI module see “SPI Timing Characteristics” on page 304.

26. Electrical Characteristics

26.1 Absolute Maximum Ratings*

Operating Temperature	-55°C to +125°C
Storage Temperature	-65°C to +150°C
Voltage on any Pin except $\overline{\text{RESET}}$ with respect to Ground	-0.5V to $V_{CC}+0.5V$
Voltage on $\overline{\text{RESET}}$ with respect to Ground	-0.5V to +13.0V
Maximum Operating Voltage	6.0V
DC Current per I/O Pin	40.0 mA
DC Current V_{CC} and GND Pins	200.0 mA

*NOTICE: Stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or other conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

26.2 DC Characteristics

$T_A = -40^\circ\text{C}$ to 85°C , $V_{CC} = 1.8V$ to $5.5V$ (unless otherwise noted)

Symbol	Parameter	Condition	Min. ⁽⁵⁾	Typ.	Max. ⁽⁵⁾	Units
V_{IL}	Input Low Voltage, except XTAL1 and $\overline{\text{RESET}}$ pin	$V_{CC} = 1.8V - 2.4V$ $V_{CC} = 2.4V - 5.5V$	-0.5 -0.5		$0.2V_{CC}^{(1)}$ $0.3V_{CC}^{(1)}$	V
V_{IH}	Input High Voltage, except XTAL1 and $\overline{\text{RESET}}$ pins	$V_{CC} = 1.8V - 2.4V$ $V_{CC} = 2.4V - 5.5V$	$0.7V_{CC}^{(2)}$ $0.6V_{CC}^{(2)}$		$V_{CC} + 0.5$ $V_{CC} + 0.5$	V
V_{IL1}	Input Low Voltage, XTAL1 pin	$V_{CC} = 1.8V - 5.5V$	-0.5		$0.1V_{CC}^{(1)}$	V
V_{IH1}	Input High Voltage, XTAL1 pin	$V_{CC} = 1.8V - 2.4V$ $V_{CC} = 2.4V - 5.5V$	$0.8V_{CC}^{(2)}$ $0.7V_{CC}^{(2)}$		$V_{CC} + 0.5$ $V_{CC} + 0.5$	V
V_{IL2}	Input Low Voltage, $\overline{\text{RESET}}$ pin	$V_{CC} = 1.8V - 5.5V$	-0.5		$0.2V_{CC}^{(1)}$	V
V_{IH2}	Input High Voltage, $\overline{\text{RESET}}$ pin	$V_{CC} = 1.8V - 5.5V$	$0.9V_{CC}^{(2)}$		$V_{CC} + 0.5$	V
V_{IL3}	Input Low Voltage, $\overline{\text{RESET}}$ pin as I/O	$V_{CC} = 1.8V - 2.4V$ $V_{CC} = 2.4V - 5.5V$	-0.5 -0.5		$0.2V_{CC}^{(1)}$ $0.3V_{CC}^{(1)}$	V
V_{IH3}	Input High Voltage, $\overline{\text{RESET}}$ pin as I/O	$V_{CC} = 1.8V - 2.4V$ $V_{CC} = 2.4V - 5.5V$	$0.7V_{CC}^{(2)}$ $0.6V_{CC}^{(2)}$		$V_{CC} + 0.5$ $V_{CC} + 0.5$	V
V_{OL}	Output Low Voltage ⁽³⁾ , $\overline{\text{RESET}}$ pin as I/O	$I_{OL} = 20mA$, $V_{CC} = 5V$ $I_{OL} = 6mA$, $V_{CC} = 3V$			0.7 0.5	V
V_{OH}	Output High Voltage ⁽⁴⁾ , $\overline{\text{RESET}}$ pin as I/O	$I_{OH} = -20mA$, $V_{CC} = 5V$ $I_{OH} = -10mA$, $V_{CC} = 3V$	4.2 2.3			V
V_{OL3}	Output Low Voltage ⁽³⁾ , $\overline{\text{RESET}}$ pin as I/O	TBD			TBD	V
V_{OH3}	Output High Voltage ⁽⁴⁾ , $\overline{\text{RESET}}$ pin as I/O	TBD	TBD			V
I_{IL}	Input Leakage Current I/O Pin	$V_{CC} = 5.5V$, pin low (absolute value)			1	μA
I_{IH}	Input Leakage Current I/O Pin	$V_{CC} = 5.5V$, pin high (absolute value)			1	μA

$T_A = -40^{\circ}\text{C}$ to 85°C , $V_{CC} = 1.8\text{V}$ to 5.5V (unless otherwise noted) (Continued)

Symbol	Parameter	Condition	Min. ⁽⁵⁾	Typ.	Max. ⁽⁵⁾	Units
R_{RST}	Reset Pull-up Resistor		30		60	$k\Omega$
R_{PU}	I/O Pin Pull-up Resistor		20		50	$k\Omega$
I_{CC}	Power Supply Current ⁽⁶⁾	Active 1MHz, $V_{CC} = 2\text{V}$ (ATmega48/88/168V)			0.55	mA
		Active 4MHz, $V_{CC} = 3\text{V}$ (ATmega48/88/168L)			3.5	mA
		Active 8MHz, $V_{CC} = 5\text{V}$ (ATmega48/88/168)			12	mA
		Idle 1MHz, $V_{CC} = 2\text{V}$ (ATmega48/88/168V)		0.25	0.5	mA
		Idle 4MHz, $V_{CC} = 3\text{V}$ (ATmega48/88/168L)			1.5	mA
		Idle 8MHz, $V_{CC} = 5\text{V}$ (ATmega48/88/168)			5.5	mA
	Power-down mode	WDT enabled, $V_{CC} = 3\text{V}$		<8	15	μA
		WDT disabled, $V_{CC} = 3\text{V}$		<1	2	μA
V_{ACIO}	Analog Comparator Input Offset Voltage	$V_{CC} = 5\text{V}$ $V_{in} = V_{CC}/2$		<10	40	mV
I_{ACLK}	Analog Comparator Input Leakage Current	$V_{CC} = 5\text{V}$ $V_{in} = V_{CC}/2$	-50		50	nA
t_{ACID}	Analog Comparator Propagation Delay	$V_{CC} = 2.7\text{V}$ $V_{CC} = 4.0\text{V}$		750 500		ns

- Notes:
1. "Max" means the highest value where the pin is guaranteed to be read as low
 2. "Min" means the lowest value where the pin is guaranteed to be read as high
 3. Although each I/O port can sink more than the test conditions (20mA at $V_{CC} = 5\text{V}$, 10mA at $V_{CC} = 3\text{V}$) under steady state conditions (non-transient), the following must be observed:
 ATmega48:
 1] The sum of all IOL, for ports C0 - C5, should not exceed 100 mA.
 2] The sum of all IOL, for ports C6, D0 - D4, should not exceed 100 mA.
 3] The sum of all IOL, for ports B0 - B7, D5 - D7, should not exceed 100 mA.
 ATmega88/168:
 1] The sum of all IOL, for ports C0 - C5, should not exceed 100 mA.
 2] The sum of all IOL, for ports C6, D0 - D4, should not exceed 100 mA.
 3] The sum of all IOL, for ports B0 - B7, D5 - D7, should not exceed 100 mA.
 If IOL exceeds the test condition, VOL may exceed the related specification. Pins are not guaranteed to sink current greater than the listed test condition.
 4. Although each I/O port can source more than the test conditions (20mA at $V_{CC} = 5\text{V}$, 10mA at $V_{CC} = 3\text{V}$) under steady state conditions (non-transient), the following must be observed:
 ATmega48:
 1] The sum of all IOH, for ports C0 - C5, should not exceed 100 mA.
 2] The sum of all IOH, for ports C6, D0 - D4, should not exceed 100 mA.
 3] The sum of all IOH, for ports B0 - B7, D5 - D7, should not exceed 100 mA.
 ATmega88/168:
 1] The sum of all IOH, for ports C0 - C5, should not exceed 100 mA.
 2] The sum of all IOH, for ports C6, D0 - D4, should not exceed 100 mA.
 3] The sum of all IOH, for ports B0 - B7, D5 - D7, should not exceed 100 mA.

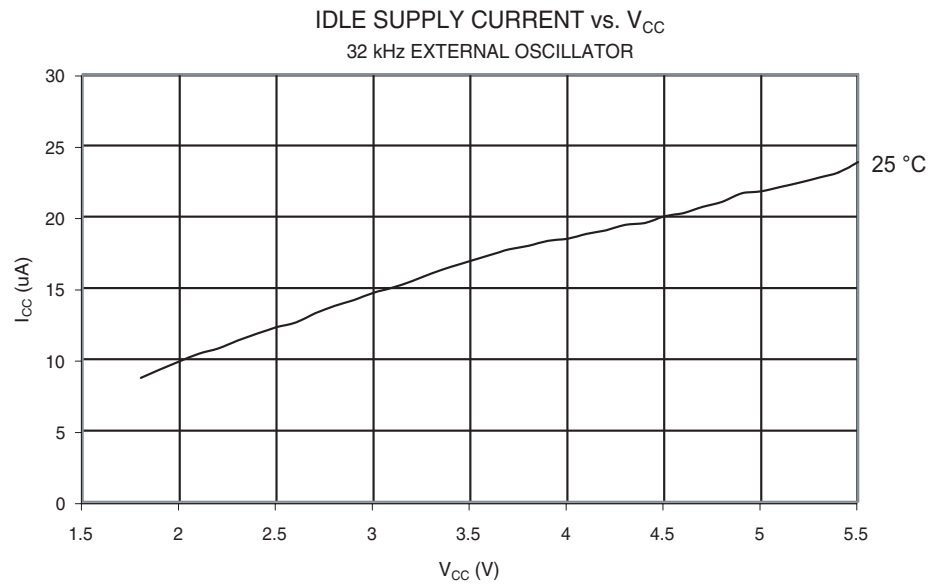
26.8 ADC Characteristics – Preliminary Data

Table 26-4. ADC Characteristics

Symbol	Parameter	Condition	Min	Typ	Max	Units
	Resolution			10		Bits
	Absolute accuracy (Including INL, DNL, quantization error, gain and offset error)	$V_{REF} = 4V, V_{CC} = 4V,$ ADC clock = 200 kHz		2	2.5	LSB
		$V_{REF} = 4V, V_{CC} = 4V,$ ADC clock = 1 MHz		4.5		LSB
		$V_{REF} = 4V, V_{CC} = 4V,$ ADC clock = 200 kHz Noise Reduction Mode		2		LSB
		$V_{REF} = 4V, V_{CC} = 4V,$ ADC clock = 1 MHz Noise Reduction Mode		4.5		LSB
	Integral Non-Linearity (INL)	$V_{REF} = 4V, V_{CC} = 4V,$ ADC clock = 200 kHz		0.5		LSB
	Differential Non-Linearity (DNL)	$V_{REF} = 4V, V_{CC} = 4V,$ ADC clock = 200 kHz		0.25		LSB
	Gain Error	$V_{REF} = 4V, V_{CC} = 4V,$ ADC clock = 200 kHz		2		LSB
	Offset Error	$V_{REF} = 4V, V_{CC} = 4V,$ ADC clock = 200 kHz		2		LSB
	Conversion Time	Free Running Conversion	13		260	μs
	Clock Frequency		50		1000	kHz
$AV_{CC}^{(1)}$	Analog Supply Voltage		$V_{CC} - 0.3$		$V_{CC} + 0.3$	V
V_{REF}	Reference Voltage		1.0		AV_{CC}	V
V_{IN}	Input Voltage		GND		V_{REF}	V
	Input Bandwidth			38.5		kHz
V_{INT}	Internal Voltage Reference		1.0	1.1	1.2	V
R_{REF}	Reference Input Resistance			32		$k\Omega$
R_{AIN}	Analog Input Resistance			100		$M\Omega$

Note: 1. AV_{CC} absolute min/max: 1.8V/5.5V

Figure 27-12. Idle Supply Current vs. V_{CC} (32 kHz External Oscillator)



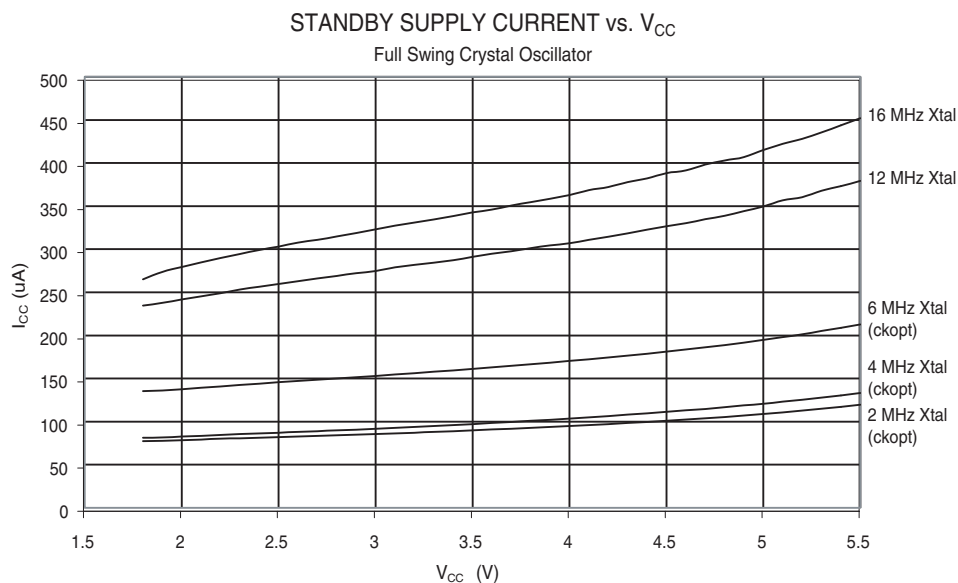
27.3 Supply Current of I/O modules

The tables and formulas below can be used to calculate the additional current consumption for the different I/O modules in Active and Idle mode. The enabling or disabling of the I/O modules are controlled by the Power Reduction Register. See ["Power Reduction Register" on page 39](#) for details.

Table 27-1. Additional Current Consumption for the different I/O modules (absolute values)

PRR bit	Typical numbers		
	$V_{CC} = 2V, F = 1MHz$	$V_{CC} = 3V, F = 4MHz$	$V_{CC} = 5V, F = 8MHz$
PRUSART0	8.0 uA	51 uA	220 uA
PRTWI	12 uA	75 uA	315 uA
PRTIM2	11 uA	72 uA	300 uA
PRTIM1	5.0 uA	32 uA	130 uA
PRTIM0	4.0 uA	24 uA	100 uA
PRSPI	15 uA	95 uA	400 uA
PRADC	12 uA	75 uA	315 uA

Figure 27-17. Standby Supply Current vs. V_{CC} (Full Swing Crystal Oscillator)



27.7 Pin Pull-up

Figure 27-18. I/O Pin Pull-Up Resistor Current vs. Input Voltage ($V_{CC} = 5V$)

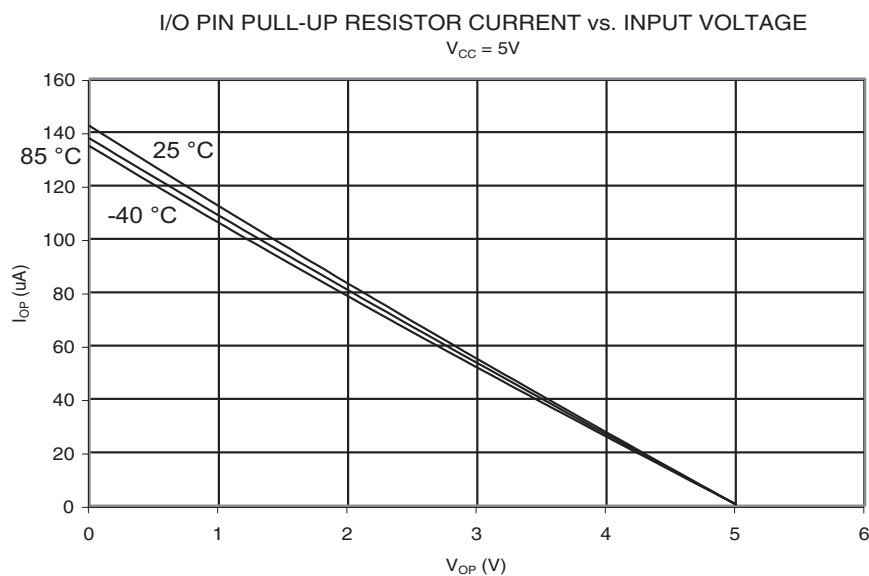


Figure 27-19. I/O Pin Pull-Up Resistor Current vs. Input Voltage ($V_{CC} = 2.7V$)

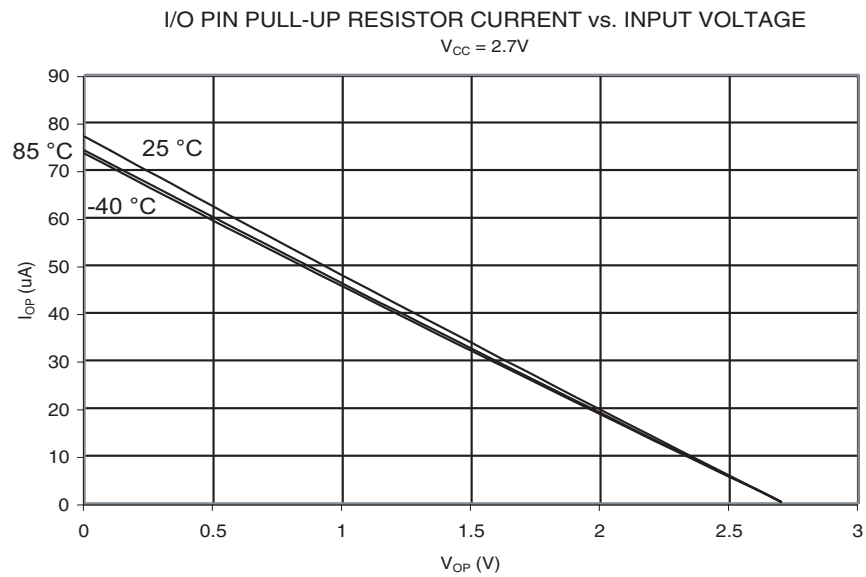


Figure 27-20. Reset Pull-Up Resistor Current vs. Reset Pin Voltage ($V_{CC} = 5V$)

