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Details

Product Status	Active
Core Processor	12V1
Core Size	16-Bit
Speed	25MHz
Connectivity	CANbus, IrDA, LINbus, SCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	54
Program Memory Size	240KB (240K x 8)
Program Memory Type	FLASH
EEPROM Size	4K x 8
RAM Size	11K x 8
Voltage - Supply (Vcc/Vdd)	3.13V ~ 5.5V
Data Converters	A/D 16x12b; D/A 2x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 105°C (TA)
Mounting Type	Surface Mount
Package / Case	64-LQFP
Supplier Device Package	64-LQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/s9s12ga240f0vlh

The S12GA192 and the S12GA240 contain a Reference Voltage Attenuator (RVA) module. The connection of the ADC's VRH/VRL inputs on these devices is shown in [Figure 1-27](#).

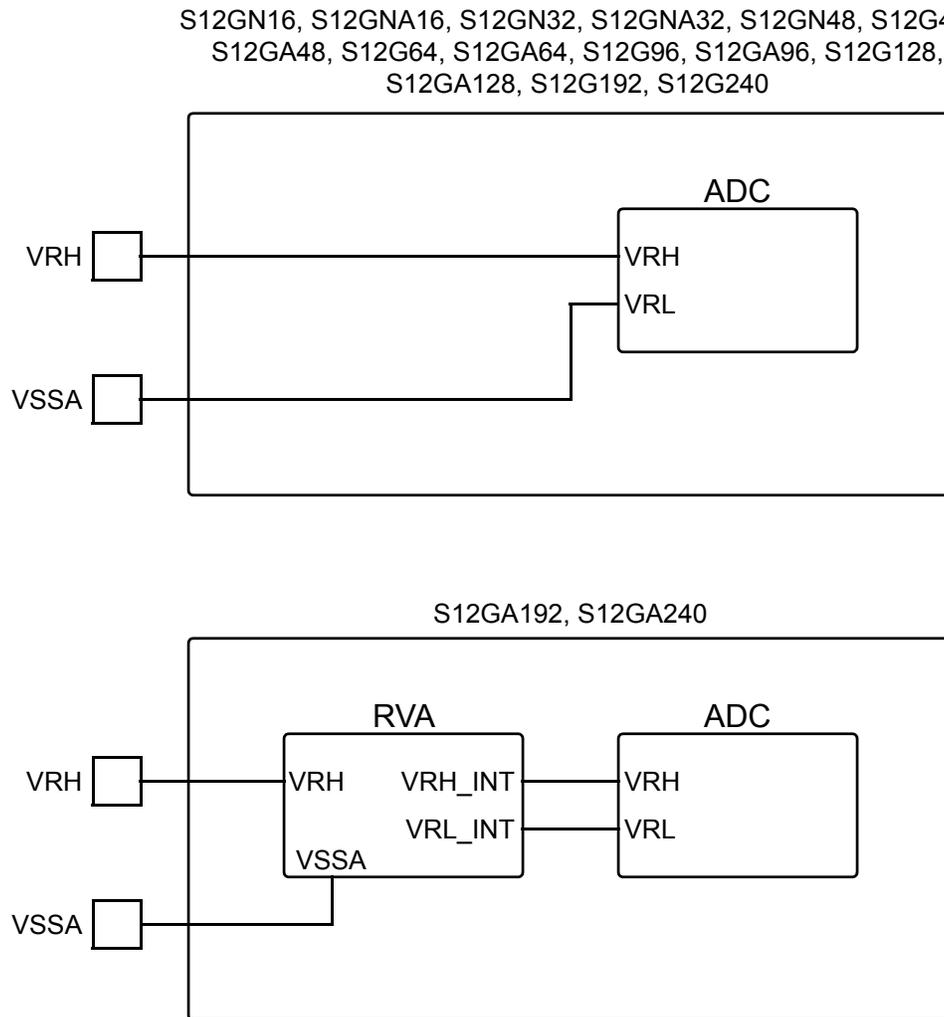


Figure 1-27. ADC VRH/VRL Signal Connection

1.19 BDM Clock Source Connectivity

The BDM clock is mapped to the VCO clock divided by 8.

Address: 0x0021

	7	6	5	4	3	2	1	0
R	TBF	0	0	0	0	SSF2	SSF1	SSF0
W								
Reset	—	0	0	0	0	0	0	0
POR	0	0	0	0	0	0	0	0

= Unimplemented or Reserved

Figure 8-4. Debug Status Register (DBGSR)

Read: Anytime

Write: Never

Table 8-5. DBGSR Field Descriptions

Field	Description
7 TBF	Trace Buffer Full — The TBF bit indicates that the trace buffer has stored 64 or more lines of data since it was last armed. If this bit is set, then all 64 lines will be valid data, regardless of the value of DBGCNT bits. The TBF bit is cleared when ARM in DBGCR1 is written to a one. The TBF is cleared by the power on reset initialization. Other system generated resets have no affect on this bit This bit is also visible at DBGCNT[7]
2–0 SSF[2:0]	State Sequencer Flag Bits — The SSF bits indicate in which state the State Sequencer is currently in. During a debug session on each transition to a new state these bits are updated. If the debug session is ended by software clearing the ARM bit, then these bits retain their value to reflect the last state of the state sequencer before disarming. If a debug session is ended by an internal event, then the state sequencer returns to state0 and these bits are cleared to indicate that state0 was entered during the session. On arming the module the state sequencer enters state1 and these bits are forced to SSF[2:0] = 001. See Table 8-6 .

Table 8-6. SSF[2:0] — State Sequence Flag Bit Encoding

SSF[2:0]	Current State
000	State0 (disarmed)
001	State1
010	State2
011	State3
100	Final State
101,110,111	Reserved

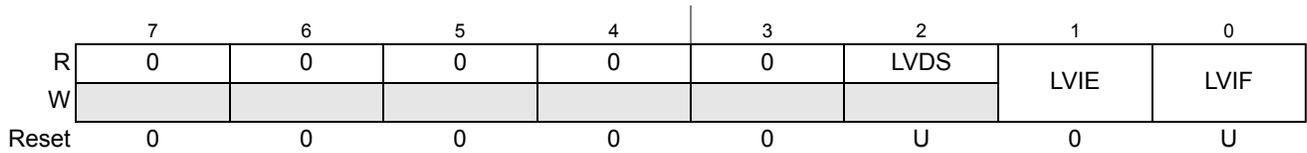
8.3.2.3 Debug Trace Control Register (DBGTCR)

Address: 0x0022

	7	6	5	4	3	2	1	0
R	0	TSOURCE	0	0	TRCMOD		0	TALIGN
W								
Reset	0	0	0	0	0	0	0	0

Figure 8-5. Debug Trace Control Register (DBGTCR)

0x02F1



The Reset state of LVDS and LVIF depends on the external supplied VDDA level

 = Unimplemented or Reserved

Figure 10-16. Low Voltage Control Register (CPMULVCTL)

Read: Anytime

Write: LVIE and LVIF are write anytime, LVDS is read only

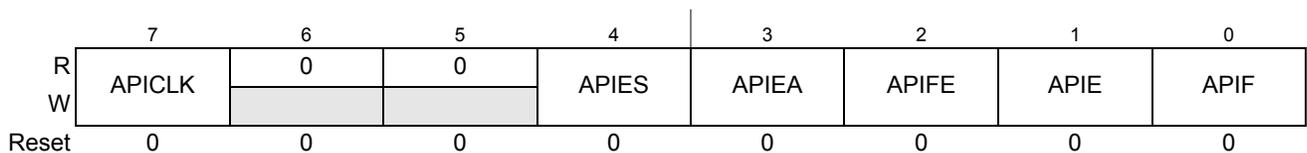
Table 10-15. CPMULVCTL Field Descriptions

Field	Description
2 LVDS	Low-Voltage Detect Status Bit — This read-only status bit reflects the voltage level on VDDA. Writes have no effect. 0 Input voltage VDDA is above level V_{LVID} or RPM. 1 Input voltage VDDA is below level V_{LVIA} and FPM.
1 LVIE	Low-Voltage Interrupt Enable Bit 0 Interrupt request is disabled. 1 Interrupt will be requested whenever LVIF is set.
0 LVIF	Low-Voltage Interrupt Flag — LVIF is set to 1 when LVDS status bit changes. This flag can only be cleared by writing a 1. Writing a 0 has no effect. If enabled (LVIE = 1), LVIF causes an interrupt request. 0 No change in LVDS bit. 1 LVDS bit has changed.

10.3.2.14 Autonomous Periodical Interrupt Control Register (CPMUAPICTL)

The CPMUAPICTL register allows the configuration of the autonomous periodical interrupt features.

0x02F2



 = Unimplemented or Reserved

Figure 10-17. Autonomous Periodical Interrupt Control Register (CPMUAPICTL)

Read: Anytime

12.3.2.12 ATD Conversion Result Registers (ATDDR n)

The A/D conversion results are stored in 8 result registers. Results are always in unsigned data representation. Left and right justification is selected using the DJM control bit in ATDCTL3.

If automatic compare of conversions results is enabled (CMPE[n]=1 in ATDCMPE), these registers must be written with the compare values in left or right justified format depending on the actual value of the DJM bit. In this case, as the ATDDR n register is used to hold the compare value, the result will not be stored there at the end of the conversion but is lost.

Attention, n is the conversion number, NOT the channel number!

Read: Anytime

Write: Anytime

NOTE

For conversions not using automatic compare, results are stored in the result registers after each conversion. In this case avoid writing to ATDDR n except for initial values, because an A/D result might be overwritten.

12.3.2.12.1 Left Justified Result Data (DJM=0)

Module Base +

0x0010 = ATDDR0, 0x0012 = ATDDR1, 0x0014 = ATDDR2, 0x0016 = ATDDR3

0x0018 = ATDDR4, 0x001A = ATDDR5, 0x001C = ATDDR6, 0x001E = ATDDR7

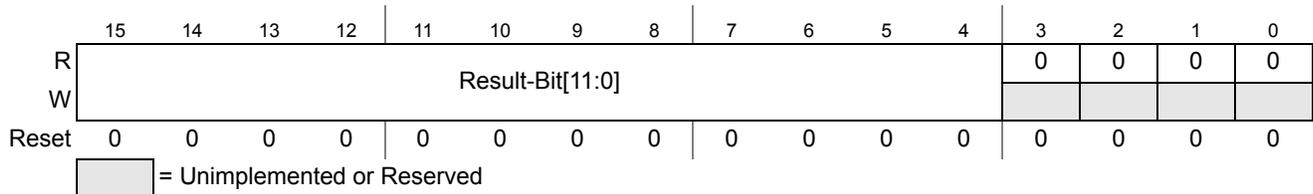


Figure 12-14. Left justified ATD conversion result register (ATDDR n)

Table 12-21 shows how depending on the A/D resolution the conversion result is transferred to the ATD result registers for left justified data. Compare is always done using all 12 bits of both the conversion result and the compare value in ATDDR n .

Table 12-21. Conversion result mapping to ATDDR n

A/D resolution	DJM	conversion result mapping to ATDDR n
8-bit data	0	Result-Bit[11:4] = conversion result, Result-Bit[3:0]=0000
10-bit data	0	Result-Bit[11:2] = conversion result, Result-Bit[1:0]=00
12-bit data	0	Result-Bit[11:0] = result

18.5 Initialization/Application Information

18.5.1 MSCAN initialization

The procedure to initially start up the MSCAN module out of reset is as follows:

1. Assert CANE
2. Write to the configuration registers in initialization mode
3. Clear INTRQ to leave initialization mode

If the configuration of registers which are only writable in initialization mode shall be changed:

1. Bring the module into sleep mode by setting SLPRQ and awaiting SLPK to assert after the CAN bus becomes idle.
2. Enter initialization mode: assert INTRQ and await INITAK
3. Write to the configuration registers in initialization mode
4. Clear INTRQ to leave initialization mode and continue

18.5.2 Bus-Off Recovery

The bus-off recovery is user configurable. The bus-off state can either be left automatically or on user request.

For reasons of backwards compatibility, the MSCAN defaults to automatic recovery after reset. In this case, the MSCAN will become error active again after counting 128 occurrences of 11 consecutive recessive bits on the CAN bus (see the Bosch CAN 2.0 A/B specification for details).

If the MSCAN is configured for user request (BORM set in **MSCAN Control Register 1 (CANCTL1)**), the recovery from bus-off starts after both independent events have become true:

- 128 occurrences of 11 consecutive recessive bits on the CAN bus have been monitored
- BOHOLD in **MSCAN Miscellaneous Register (CANMISC)** has been cleared by the user

These two events may occur in any order.

Table 19-4. PWMCLK Field Descriptions

Note: Bits related to available channels have functional significance. Writing to unavailable bits has no effect. Read from unavailable bits return a zero

Field	Description
7-0 PCLK[7:0]	Pulse Width Channel 7-0 Clock Select 0 Clock A or B is the clock source for PWM channel 7-0, as shown in Table 19-5 and Table 19-6 . 1 Clock SA or SB is the clock source for PWM channel 7-0, as shown in Table 19-5 and Table 19-6 .

The clock source of each PWM channel is determined by PCLKx bits in PWMCLK and PCLKABx bits in PWMCLKAB (see [Section 19.3.2.7, “PWM Clock A/B Select Register \(PWMCLKAB\)”](#)). For Channel 0, 1, 4, 5, the selection is shown in [Table 19-5](#); For Channel 2, 3, 6, 7, the selection is shown in [Table 19-6](#).

Table 19-5. PWM Channel 0, 1, 4, 5 Clock Source Selection

PCLKAB[0,1,4,5]	PCLK[0,1,4,5]	Clock Source Selection
0	0	Clock A
0	1	Clock SA
1	0	Clock B
1	1	Clock SB

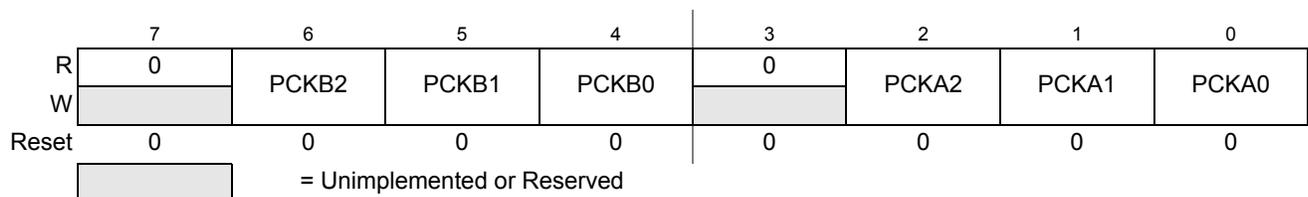
Table 19-6. PWM Channel 2, 3, 6, 7 Clock Source Selection

PCLKAB[2,3,6,7]	PCLK[2,3,6,7]	Clock Source Selection
0	0	Clock B
0	1	Clock SB
1	0	Clock A
1	1	Clock SA

19.3.2.4 PWM Prescale Clock Select Register (PWMPRCLK)

This register selects the prescale clock source for clocks A and B independently.

Module Base + 0x0003

**Figure 19-6. PWM Prescale Clock Select Register (PWMPRCLK)**

Read: Anytime

Write: Anytime

NOTE

PCKB2–0 and PCKA2–0 register bits can be written anytime. If the clock pre-scale is changed while a PWM signal is being generated, a truncated or stretched pulse can occur during the transition.

20.1.4 Block Diagram

Figure 20-1 is a high level block diagram of the SCI module, showing the interaction of various function blocks.

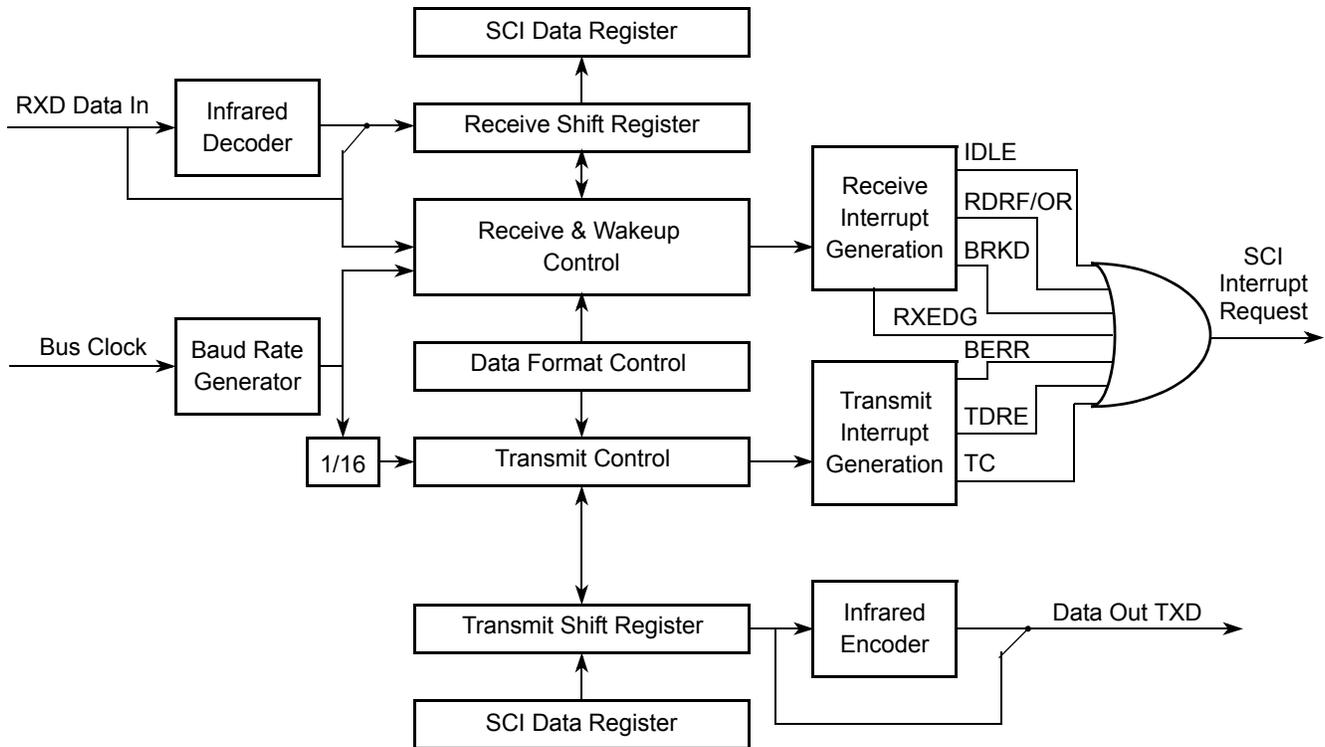


Figure 20-1. SCI Block Diagram

20.2 External Signal Description

The SCI module has a total of two external pins.

20.2.1 TXD — Transmit Pin

The TXD pin transmits SCI (standard or infrared) data. It will idle high in either mode and is high impedance anytime the transmitter is disabled.

20.2.2 RXD — Receive Pin

The RXD pin receives SCI (standard or infrared) data. An idle line is detected as a line high. This input is ignored when the receiver is disabled and should be terminated to a known voltage.

20.3 Memory Map and Register Definition

This section provides a detailed description of all the SCI registers.

Table 20-13. SCIDRH and SCIDRL Field Descriptions

Field	Description
SCIDRH 7 R8	Received Bit 8 — R8 is the ninth data bit received when the SCI is configured for 9-bit data format (M = 1).
SCIDRH 6 T8	Transmit Bit 8 — T8 is the ninth data bit transmitted when the SCI is configured for 9-bit data format (M = 1).
SCIDRL 7:0 R[7:0] T[7:0]	R7:R0 — Received bits seven through zero for 9-bit or 8-bit data formats T7:T0 — Transmit bits seven through zero for 9-bit or 8-bit formats

NOTE

If the value of T8 is the same as in the previous transmission, T8 does not have to be rewritten. The same value is transmitted until T8 is rewritten.

In 8-bit data format, only SCI data register low (SCIDRL) needs to be accessed.

When transmitting in 9-bit data format and using 8-bit write instructions, write first to SCI data register high (SCIDRH), then SCIDRL.

20.4 Functional Description

This section provides a complete functional description of the SCI block, detailing the operation of the design from the end user perspective in a number of subsections.

Figure 20-14 shows the structure of the SCI module. The SCI allows full duplex, asynchronous, serial communication between the CPU and remote devices, including other CPUs. The SCI transmitter and receiver operate independently, although they use the same baud rate generator. The CPU monitors the status of the SCI, writes the data to be transmitted, and processes received data.

20.4.5.5 LIN Transmit Collision Detection

This module allows to check for collisions on the LIN bus.

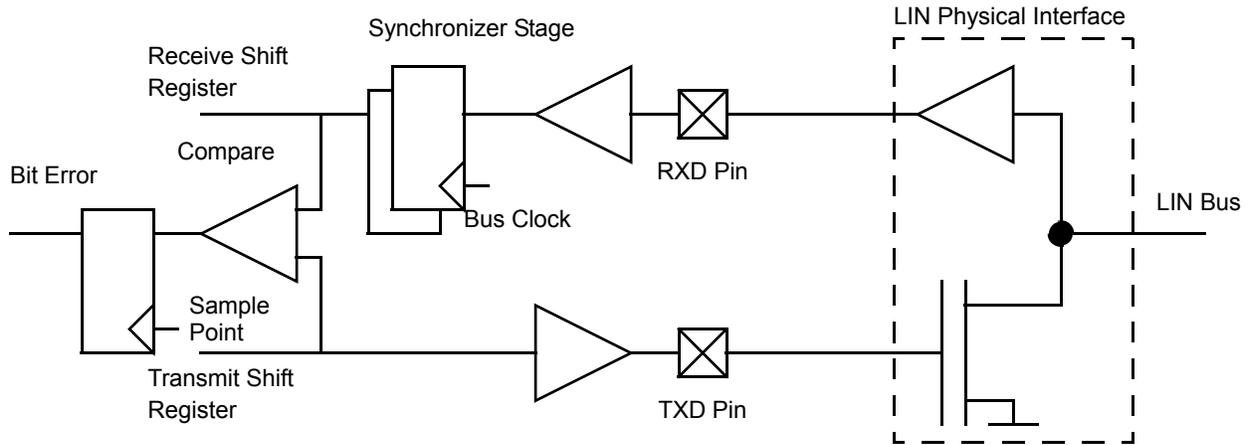


Figure 20-18. Collision Detect Principle

If the bit error circuit is enabled ($BERRM[1:0] = 0:1$ or $= 1:0$), the error detect circuit will compare the transmitted and the received data stream at a point in time and flag any mismatch. The timing checks run when transmitter is active (not idle). As soon as a mismatch between the transmitted data and the received data is detected the following happens:

- The next bit transmitted will have a high level ($TXPOL = 0$) or low level ($TXPOL = 1$)
- The transmission is aborted and the byte in transmit buffer is discarded.
- the transmit data register empty and the transmission complete flag will be set
- The bit error interrupt flag, $BERRIF$, will be set.
- No further transmissions will take place until the $BERRIF$ is cleared.

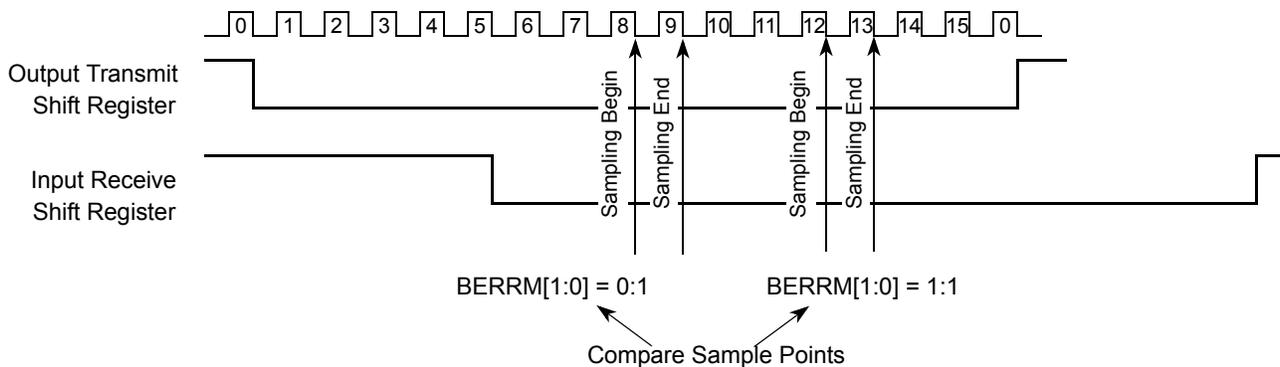


Figure 20-19. Timing Diagram Bit Error Detection

If the bit error detect feature is disabled, the bit error interrupt flag is cleared.

NOTE

The $RXPOL$ and $TXPOL$ bit should be set the same when transmission collision detect feature is enabled, otherwise the bit error interrupt flag may be set incorrectly.

Table 21-6. Example SPI Baud Rate Selection (25 MHz Bus Clock)

SPPR2	SPPR1	SPPR0	SPR2	SPR1	SPR0	Baud Rate Divisor	Baud Rate
0	0	1	0	1	0	16	1.5625 Mbit/s
0	0	1	0	1	1	32	781.25 kbit/s
0	0	1	1	0	0	64	390.63 kbit/s
0	0	1	1	0	1	128	195.31 kbit/s
0	0	1	1	1	0	256	97.66 kbit/s
0	0	1	1	1	1	512	48.83 kbit/s
0	1	0	0	0	0	6	4.16667 Mbit/s
0	1	0	0	0	1	12	2.08333 Mbit/s
0	1	0	0	1	0	24	1.04167 Mbit/s
0	1	0	0	1	1	48	520.83 kbit/s
0	1	0	1	0	0	96	260.42 kbit/s
0	1	0	1	0	1	192	130.21 kbit/s
0	1	0	1	1	0	384	65.10 kbit/s
0	1	0	1	1	1	768	32.55 kbit/s
0	1	1	0	0	0	8	3.125 Mbit/s
0	1	1	0	0	1	16	1.5625 Mbit/s
0	1	1	0	1	0	32	781.25 kbit/s
0	1	1	0	1	1	64	390.63 kbit/s
0	1	1	1	0	0	128	195.31 kbit/s
0	1	1	1	0	1	256	97.66 kbit/s
0	1	1	1	1	0	512	48.83 kbit/s
0	1	1	1	1	1	1024	24.41 kbit/s
1	0	0	0	0	0	10	2.5 Mbit/s
1	0	0	0	0	1	20	1.25 Mbit/s
1	0	0	0	1	0	40	625 kbit/s
1	0	0	0	1	1	80	312.5 kbit/s
1	0	0	1	0	0	160	156.25 kbit/s
1	0	0	1	0	1	320	78.13 kbit/s
1	0	0	1	1	0	640	39.06 kbit/s
1	0	0	1	1	1	1280	19.53 kbit/s
1	0	1	0	0	0	12	2.08333 Mbit/s
1	0	1	0	0	1	24	1.04167 Mbit/s
1	0	1	0	1	0	48	520.83 kbit/s
1	0	1	0	1	1	96	260.42 kbit/s
1	0	1	1	0	0	192	130.21 kbit/s
1	0	1	1	0	1	384	65.10 kbit/s
1	0	1	1	1	0	768	32.55 kbit/s
1	0	1	1	1	1	1536	16.28 kbit/s
1	1	0	0	0	0	14	1.78571 Mbit/s

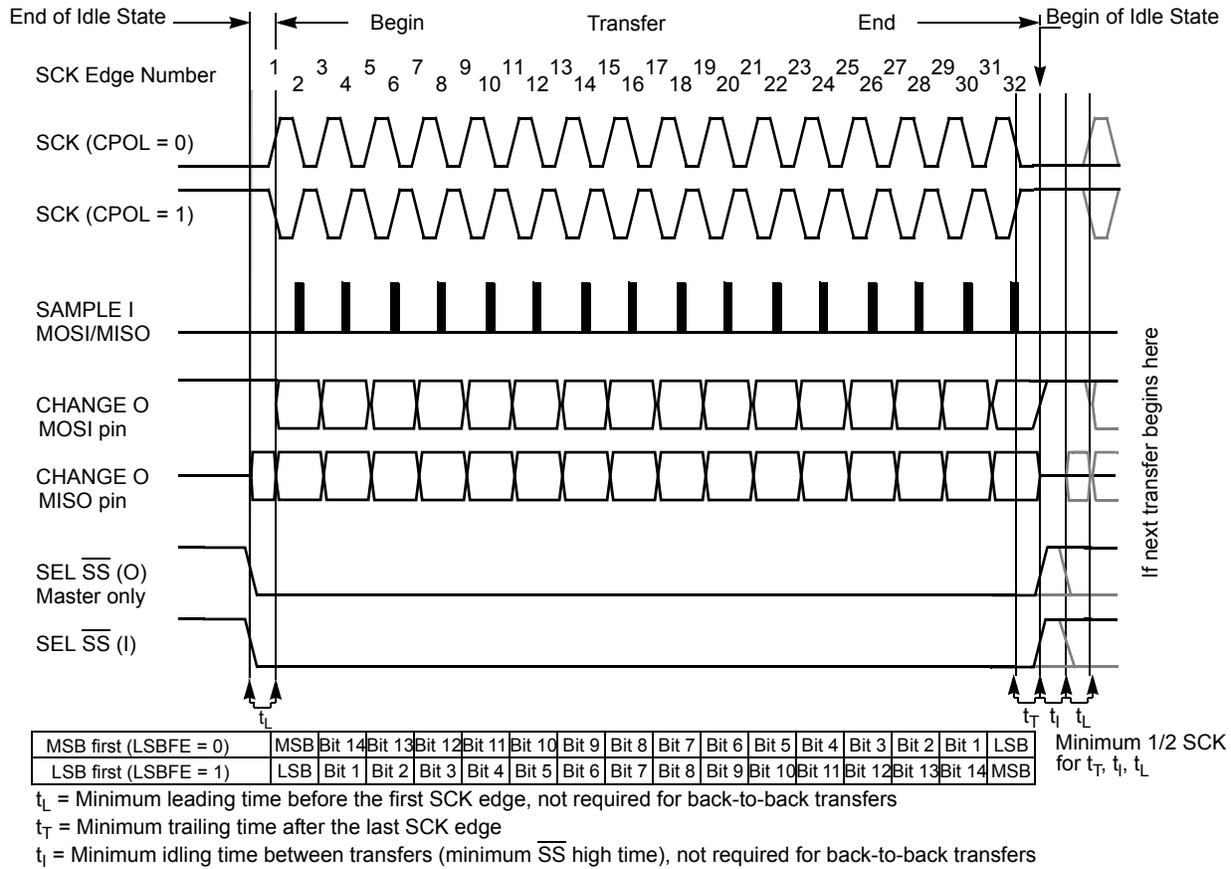


Figure 21-15. SPI Clock Format 1 (CPHA = 1), with 16-Bit Transfer Width selected (XFRW = 1)

The \overline{SS} line can remain active low between successive transfers (can be tied low at all times). This format is sometimes preferred in systems having a single fixed master and a single slave that drive the MISO data line.

- Back-to-back transfers in master mode

In master mode, if a transmission has completed and new data is available in the SPI data register, this data is sent out immediately without a trailing and minimum idle time.

The SPI interrupt request flag (SPIF) is common to both the master and slave modes. SPIF gets set one half SCK cycle after the last SCK edge.

21.4.4 SPI Baud Rate Generation

Baud rate generation consists of a series of divider stages. Six bits in the SPI baud rate register (SPPR2, SPPR1, SPPR0, SPR2, SPR1, and SPR0) determine the divisor to the SPI module clock which results in the SPI baud rate.

The SPI clock rate is determined by the product of the value in the baud rate preselection bits (SPPR2–SPPR0) and the value in the baud rate selection bits (SPR2–SPR0). The module clock divisor equation is shown in [Equation 21-3](#).

¹ The register is available only if corresponding channel exists.

22.3.2.1 Timer Input Capture/Output Compare Select (TIOS)

Module Base + 0x0000

	7	6	5	4	3	2	1	0
R	RESERVED	RESERVED	IOS5	IOS4	IOS3	IOS2	IOS1	IOS0
W								
Reset	0	0	0	0	0	0	0	0

Figure 22-4. Timer Input Capture/Output Compare Select (TIOS)

Read: Anytime

Write: Anytime

Table 22-2. TIOS Field Descriptions

Note: Writing to unavailable bits has no effect. Reading from unavailable bits return a zero.

Field	Description
5:0 IOS[5:0]	Input Capture or Output Compare Channel Configuration 0 The corresponding implemented channel acts as an input capture. 1 The corresponding implemented channel acts as an output compare.

22.3.2.2 Timer Compare Force Register (CFORC)

Module Base + 0x0001

	7	6	5	4	3	2	1	0
R	0	0	0	0	0	0	0	0
W	RESERVED	RESERVED	FOC5	FOC4	FOC3	FOC2	FOC1	FOC0
Reset	0	0	0	0	0	0	0	0

Figure 22-5. Timer Compare Force Register (CFORC)

Read: Anytime but will always return 0x0000 (1 state is transient)

Write: Anytime

Table 22-3. CFORC Field Descriptions

Note: Writing to unavailable bits has no effect. Reading from unavailable bits return a zero.

Field	Description
5:0 FOC[5:0]	Note: Force Output Compare Action for Channel 5:0 — A write to this register with the corresponding data bit(s) set causes the action which is programmed for output compare “x” to occur immediately. The action taken is the same as if a successful comparison had just taken place with the TCx register except the interrupt flag does not get set. If forced output compare on any channel occurs at the same time as the successful output compare then forced output compare action will take precedence and interrupt flag won't get set.

24.3.2.10 EEPROM Protection Register (EEPROT)

The EEPROT register defines which EEPROM sectors are protected against program and erase operations.

Offset Module Base + 0x0009

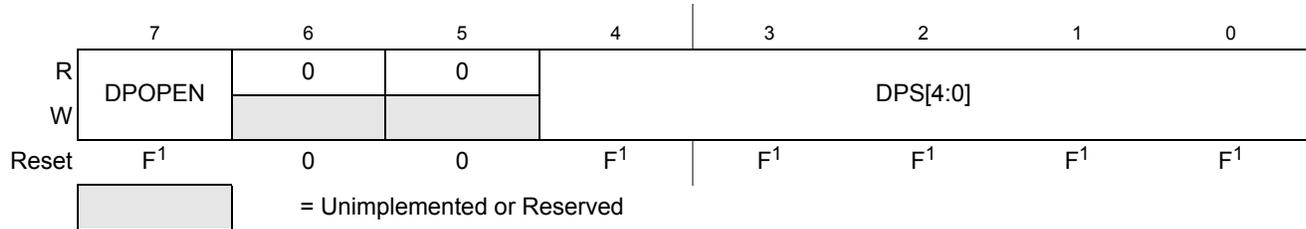


Figure 24-14. EEPROM Protection Register (EEPROT)

¹ Loaded from IFR Flash configuration field, during reset sequence.

The (unreserved) bits of the EEPROT register are writable with the restriction that protection can be added but not removed. Writes must increase the DPS value and the DPOPEN bit can only be written from 1 (protection disabled) to 0 (protection enabled). If the DPOPEN bit is set, the state of the DPS bits is irrelevant.

During the reset sequence, fields DPOPEN and DPS of the EEPROT register are loaded with the contents of the EEPROM protection byte in the Flash configuration field at global address 0x3_FF0D located in P-Flash memory (see [Table 24-4](#)) as indicated by reset condition F in [Table 24-21](#). To change the EEPROM protection that will be loaded during the reset sequence, the P-Flash sector containing the EEPROM protection byte must be unprotected, then the EEPROM protection byte must be programmed. If a double bit fault is detected while reading the P-Flash phrase containing the EEPROM protection byte during the reset sequence, the DPOPEN bit will be cleared and DPS bits will be set to leave the EEPROM memory fully protected.

Trying to alter data in any protected area in the EEPROM memory will result in a protection violation error and the FPVIOL bit will be set in the FSTAT register. Block erase of the EEPROM memory is not possible if any of the EEPROM sectors are protected.

Table 24-20. EEPROT Field Descriptions

Field	Description
7 DPOPEN	EEPROM Protection Control 0 Enables EEPROM memory protection from program and erase with protected address range defined by DPS bits 1 Disables EEPROM memory protection from program and erase
4–0 DPS[4:0]	EEPROM Protection Size — The DPS[4:0] bits determine the size of the protected area in the EEPROM memory as shown in Table 24-21 .

A summary of the Flash module registers is given in Figure 26-4 with detailed descriptions in the following subsections.

Address & Name		7	6	5	4	3	2	1	0
0x0000 FCLKDIV	R	FDIVLD	FDIVLCK	FDIV5	FDIV4	FDIV3	FDIV2	FDIV1	FDIV0
	W								
0x0001 FSEC	R	KEYEN1	KEYEN0	RNV5	RNV4	RNV3	RNV2	SEC1	SEC0
	W								
0x0002 FCCOBIX	R	0	0	0	0	0	CCOBIX2	CCOBIX1	CCOBIX0
	W								
0x0003 FRSV0	R	0	0	0	0	0	0	0	0
	W								
0x0004 FCNFG	R	CCIE	0	0	IGNSF	0	0	DFDF	FSFD
	W								
0x0005 FERCNFG	R	0	0	0	0	0	0	DFDIE	SFDIE
	W								
0x0006 FSTAT	R	CCIF	0	ACCERR	FPVIOL	MGBUSY	RSVD	MGSTAT1	MGSTAT0
	W								
0x0007 FERSTAT	R	0	0	0	0	0	0	DFDIF	SFDIF
	W								
0x0008 FPROT	R	FPOPEN	RNV6	FPHDIS	FPHS1	FPHS0	FPLDIS	FPLS1	FPLS0
	W								
0x0009 EEPROT	R	DPOPEN	0	DPS5	DPS4	DPS3	DPS2	DPS1	DPS0
	W								
0x000A FCCOBHI	R	CCOB15	CCOB14	CCOB13	CCOB12	CCOB11	CCOB10	CCOB9	CCOB8
	W								
0x000B FCCOBLO	R	CCOB7	CCOB6	CCOB5	CCOB4	CCOB3	CCOB2	CCOB1	CCOB0
	W								
0x000C FRSV1	R	0	0	0	0	0	0	0	0
	W								

Figure 26-4. FTMRG48K1 Register Summary

29.4.6.13 Set Field Margin Level Command

The Set Field Margin Level command, valid in special modes only, causes the Memory Controller to set the margin level specified for future read operations of the P-Flash or EEPROM block.

Upon clearing CCIF to launch the Set Field Margin Level command, the Memory Controller will set the **Table 29-57. Set Field Margin Level Command FCCOB Requirements**

CCOBIX[2:0]	FCCOB Parameters	
000	0x0E	Flash block selection code [1:0]. See Table 29-34
001	Margin level setting.	

field margin level for the targeted block and then set the CCIF flag.

NOTE

When the EEPROM block is targeted, the EEPROM field margin levels are applied only to the EEPROM reads. However, when the P-Flash block is targeted, the P-Flash field margin levels are applied to both P-Flash and EEPROM reads. It is not possible to apply field margin levels to the P-Flash block only.

Valid margin level settings for the Set Field Margin Level command are defined in [Table 29-58](#).

Table 29-58. Valid Set Field Margin Level Settings

CCOB (CCOBIX=001)	Level Description
0x0000	Return to Normal Level
0x0001	User Margin-1 Level ¹
0x0002	User Margin-0 Level ²
0x0003	Field Margin-1 Level ¹
0x0004	Field Margin-0 Level ²

¹ Read margin to the erased state

² Read margin to the programmed state

30.3.2.11 Flash Common Command Object Register (FCCOB)

The FCCOB is an array of six words addressed via the CCOBIX index found in the FCCOBIX register. Byte wide reads and writes are allowed to the FCCOB register.

Offset Module Base + 0x000A

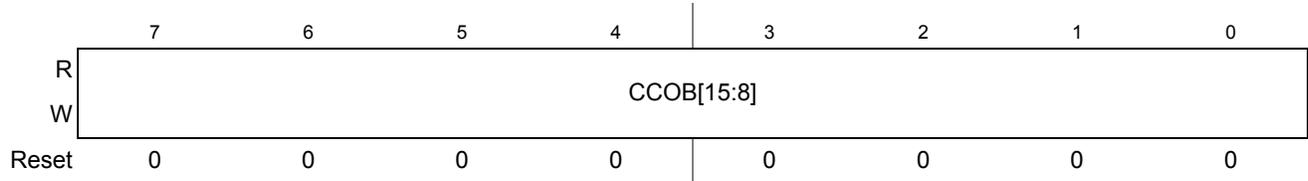


Figure 30-16. Flash Common Command Object High Register (FCCOBHI)

Offset Module Base + 0x000B

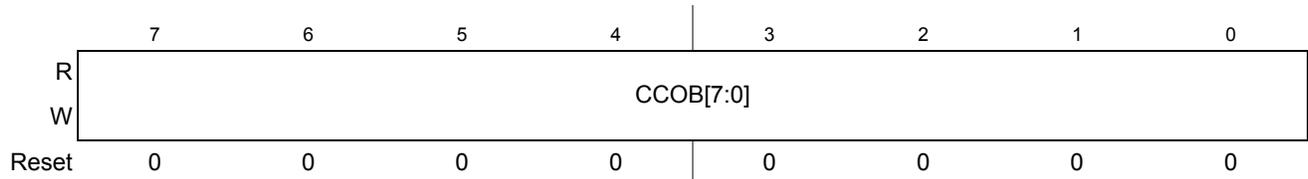


Figure 30-17. Flash Common Command Object Low Register (FCCOBLO)

30.3.2.11.1 FCCOB - NVM Command Mode

NVM command mode uses the indexed FCCOB register to provide a command code and its relevant parameters to the Memory Controller. The user first sets up all required FCCOB fields and then initiates the command's execution by writing a 1 to the CCIF bit in the FSTAT register (a 1 written by the user clears the CCIF command completion flag to 0). When the user clears the CCIF bit in the FSTAT register all FCCOB parameter fields are locked and cannot be changed by the user until the command completes (as evidenced by the Memory Controller returning CCIF to 1). Some commands return information to the FCCOB register array.

The generic format for the FCCOB parameter fields in NVM command mode is shown in [Table 30-24](#). The return values are available for reading after the CCIF flag in the FSTAT register has been returned to 1 by the Memory Controller. Writes to the unimplemented parameter fields (CCOBIX = 110 and CCOBIX = 111) are ignored with reads from these fields returning 0x0000.

[Table 30-24](#) shows the generic Flash command format. The high byte of the first word in the CCOB array contains the command code, followed by the parameters for this specific Flash command. For details on the FCCOB settings required by each command, see the Flash command descriptions in [Section 30.4.6](#).

Table 30-24. FCCOB - NVM Command Mode (Typical Usage)

CCOBIX[2:0]	Byte	FCCOB Parameter Fields (NVM Command Mode)
000	HI	FCMD[7:0] defining Flash command
	LO	6'h0, Global address [17:16]
001	HI	Global address [15:8]
	LO	Global address [7:0]

Table A-8. 5-V I/O Characteristics (Junction Temperature From –40°C To +150°C)

Conditions are 4.5 V < V _{DD35} < 5.5 V junction temperature from –40°C to +150°C, unless otherwise noted I/O Characteristics for all I/O pins except EXTAL, XTAL, TEST and supply pins.							
Num	C	Rating	Symbol	Min	Typ	Max	Unit
1	P	Input high voltage	V _{IH}	0.65*V _{DD35}	—	—	V
2	T	Input high voltage	V _{IH}	—	—	V _{DD35} +0.3	V
3	P	Input low voltage	V _{IL}	—	—	0.35*V _{DD35}	V
4	T	Input low voltage	V _{IL}	V _{SSRX} –0.3	—	—	V
5	C	Input hysteresis	V _{HYS}	0.06*V _{DD35}	—	0.3*V _{DD35}	mV
6	P	Input leakage current (pins in high impedance input mode) ¹ V _{IN} = V _{DD35} or V _{SS35} +125°C to < T _J < 150°C +105°C to < T _J < 125° –40°C to < T _J < 105°C	I _{in}	–1 –0.5 –0.4	— — —	1 0.5 0.4	μA
7	P	Output high voltage (pins in output mode) I _{OH} = –4 mA	V _{OH}	V _{DD35} – 0.8	—	—	V
8	P	Output low voltage (pins in output mode) I _{OL} = +4mA	V _{OL}	—	—	0.8	V
9	P	Internal pull up current V _{IH} min > input voltage > V _{IL} max	I _{PUL}	–10	—	–130	μA
10	P	Internal pull down current V _{IH} min > input voltage > V _{IL} max	I _{PDH}	10	—	130	μA
11	D	Input capacitance	C _{in}	—	7	—	pF
12	T	Injection current ² Single pin limit Total device Limit, sum of all injected currents	I _{ICS} I _{ICP}	–2.5 –25	—	2.5 25	mA

¹ Maximum leakage current occurs at maximum operating temperature. Current decreases by approximately one-half for each 8°C to 12°C in the temperature range from 50°C to 125°C.

² Refer to [Section A.1.4, “Current Injection”](#) for more details

Table A-12. CPMU Configuration for Pseudo Stop Current Measurement

CPMU REGISTER	Bit settings/Conditions
CPMUOSC	OSCE=1, External Square wave on EXTAL $f_{EXTAL}=4MHz$, $V_{IH}=1.8V$, $V_{IL}=0V$
CPMURTI	RTDEC=0, RTR[6:4]=111, RTR[3:0]=1111;
CPMUCOP	WCOP=1, CR[2:0]=111

Table A-13. CPMU Configuration for Run/Wait and Full Stop Current Measurement

CPMU REGISTER	Bit settings/Conditions
CPMUSYNR	VCOFRQ[1:0]=01, SYNDIV[5:0] = 24
CPMUPOSTDIV	POSTDIV[4:0]=0
CPMUCLKS	PLLSEL=1
CPMUOSC	OSCE=0, Reference clock for PLL is $f_{ref}=f_{irc1m}$ trimmed to 1MHz
API settings for STOP current measurement	
CPMUAPICTL	APIEA=0, APIFE=1, APIE=0
CPMUAPITR	trimmed to 10Khz
CPMUAPIRH/RL	set to \$FFFF

Table A-14. Peripheral Configurations for Run & Wait Current Measurement

Peripheral	Configuration
MSCAN	Configured to loop-back mode using a bit rate of 1Mbit/s
SPI	Configured to master mode, continuously transmit data (0x55 or 0xAA) at 1Mbit/s
SCI	Configured into loop mode, continuously transmit data (0x55) at speed of 57600 baud
PWM	Configured to toggle its pins at the rate of 40kHz
ADC	The peripheral is configured to operate at its maximum specified frequency and to continuously convert voltages on all input channels in sequence.

Table A-33. Static Electrical Characteristics

Characteristics noted under conditions 3.13V <= VDDA <= 5.5V>, -40°C < Tj < 150°C >, VRH=VDDA, VRL=VSSA unless otherwise noted. Typical values noted reflect the approximate parameter mean at T _A = 25°C under nominal conditions unless otherwise noted.							
Num	C	Ratings	Symbol	Min	Typ	Max	Unit
8	C	Output Voltage unbuffered range A or B (load >= 50MΩ)	V _{out}	full DAC Range A or B			V
9	P	Output Voltage (DRIVE bit = 0) ¹ buffered range A (load >= 100KΩ to VSSA) buffered range A (load >= 100KΩ to VDDA)	V _{out}	0 0.15	- -	VDDA-0.15 VDDA	V
		buffered range B (load >= 100KΩ to VSSA) buffered range B (load >= 100KΩ to VDDA)		full DAC Range B			
10	P	Output Voltage (DRIVE bit = 1) ² buffered range B with 6.4KΩ load into resistor divider of 800Ω /6.56KΩ between VDDA and VSSA. (equivalent load is >= 65KΩ to VSSA) or (equivalent load is >= 7.5KΩ to VDDA)	V _{out}	full DAC Range B			V
11	D	Buffer Output Capacitive load	C _{load}	0	-	100	pF
12	P	Buffer Output Offset	V _{offset}	-30	-	+30	mV
13	P	Settling time	t _{delay}	-	3	5	μs
14	D	Reverence voltage high	V _{refh}	VDDA-0.1V	VDDA	VDDA+0.1V	V

¹ DRIVE bit = 1 is not recommended in this case.

² DRIVE bit = 0 is not allowed with this high load.

A.7 NVM

A.7.1 Timing Parameters

The time base for all NVM program or erase operations is derived from the bus clock using the FCLKDIV register. The frequency of this derived clock must be set within the limits specified as f_{NVMOP}. The NVM module does not have any means to monitor the frequency and will not prevent program or erase operation at frequencies above or below the specified minimum. When attempting to program or erase the NVM module at a lower frequency, a full program or erase transition is not assured.

All timing parameters are a function of the bus clock frequency, f_{NVMBUS}. All program and erase times are also a function of the NVM operating frequency, f_{NVMOP}.

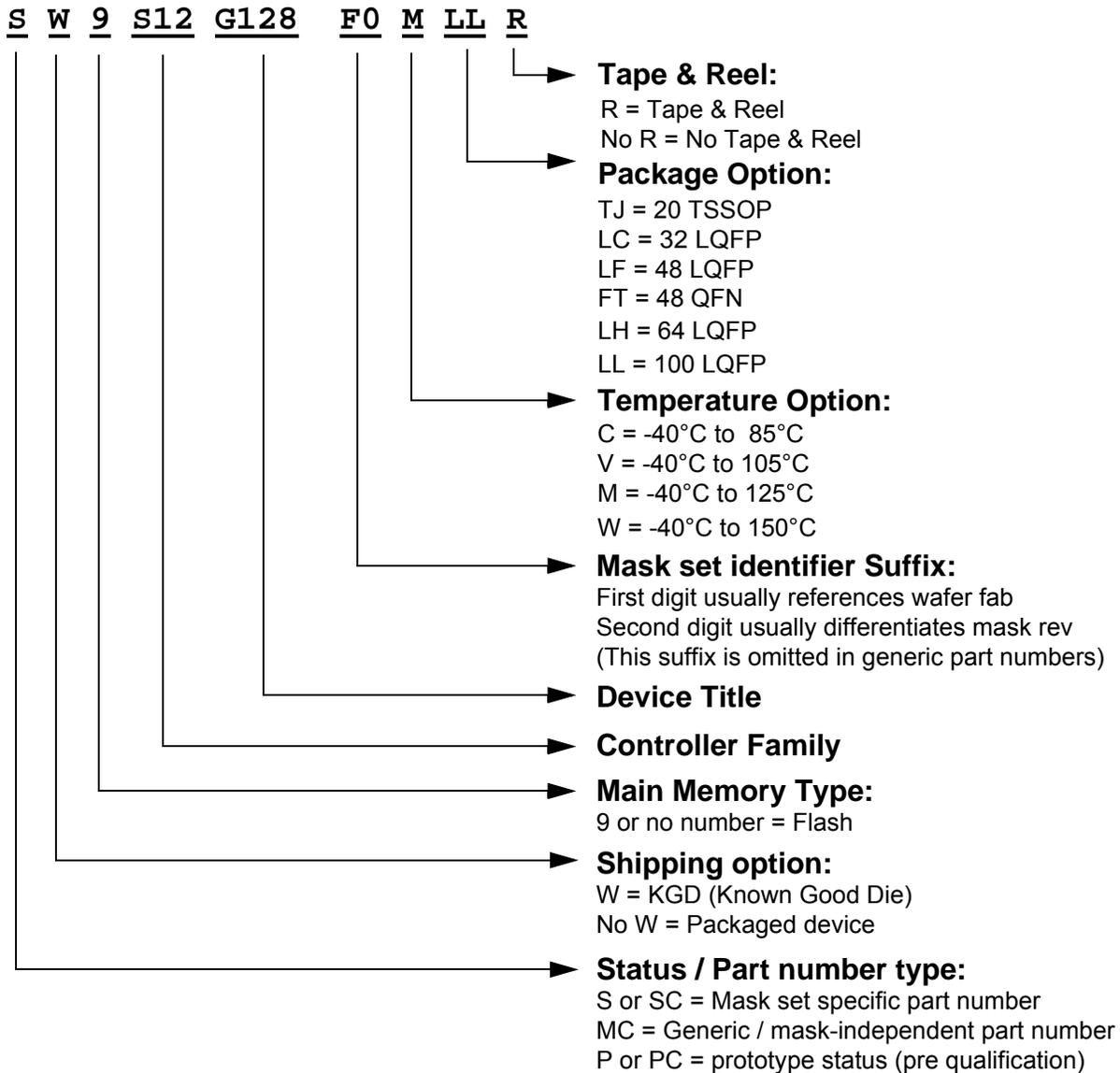


Figure C-1. Order Part Number Example