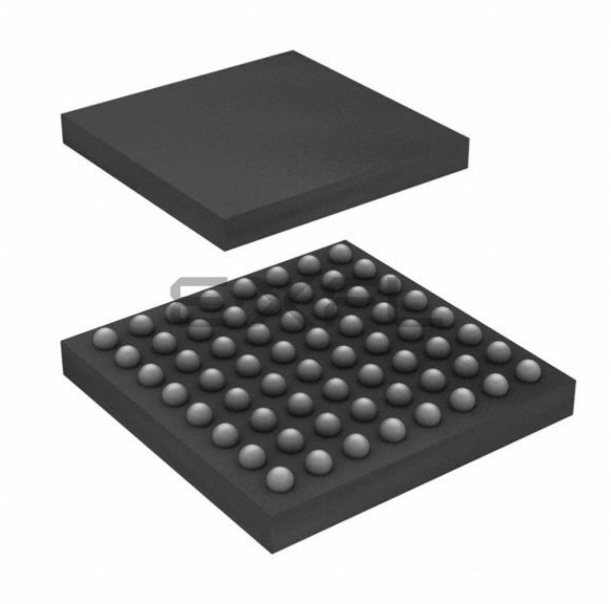

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	ARM7®
Core Size	16/32-Bit
Speed	44MHz
Connectivity	EBI/EMI, I²C, SPI, UART/USART
Peripherals	PLA, PWM, PSM, Temp Sensor, WDT
Number of I/O	30
Program Memory Size	62KB (31K x16)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	2K x 32
Voltage - Supply (Vcc/Vdd)	2.7V ~ 3.6V
Data Converters	A/D 8x12b; D/A 4x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	64-LFBGA, CSPBGA
Supplier Device Package	64-CSPBGA (6x6)
Purchase URL	https://www.e-xfl.com/product-detail/analog-devices/aduc7028bbcz62

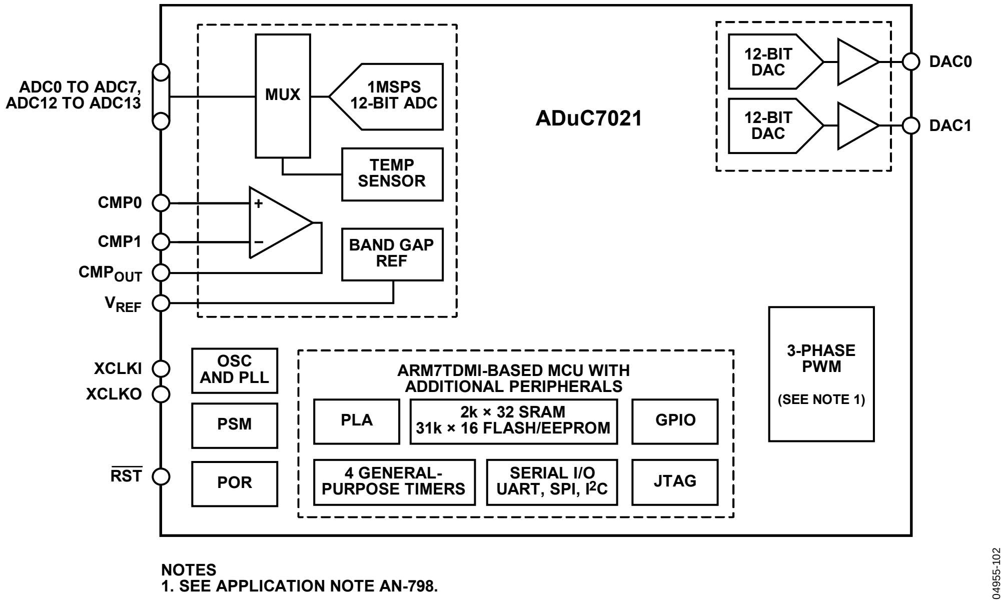

Figure 3.

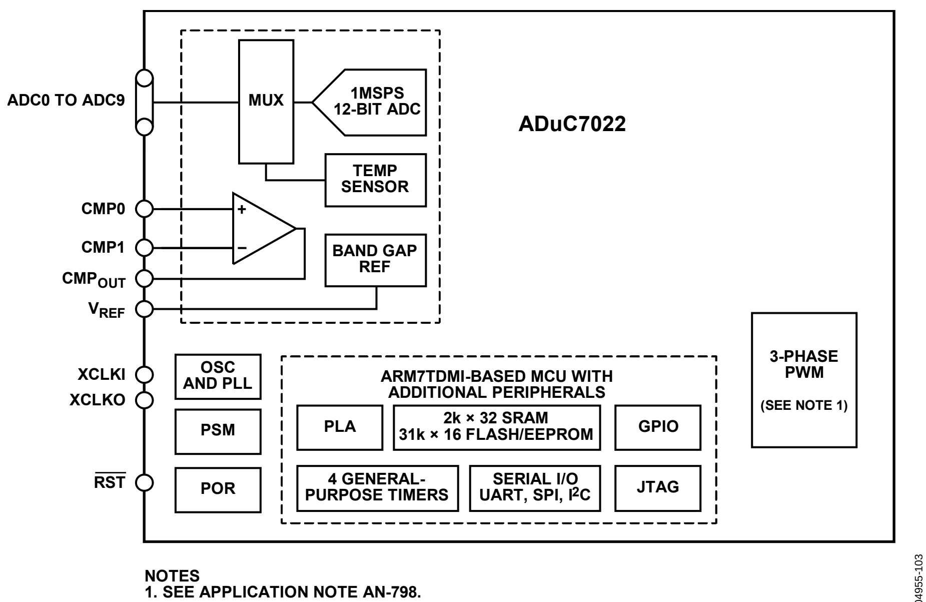

Figure 4.

DETAILED BLOCK DIAGRAM

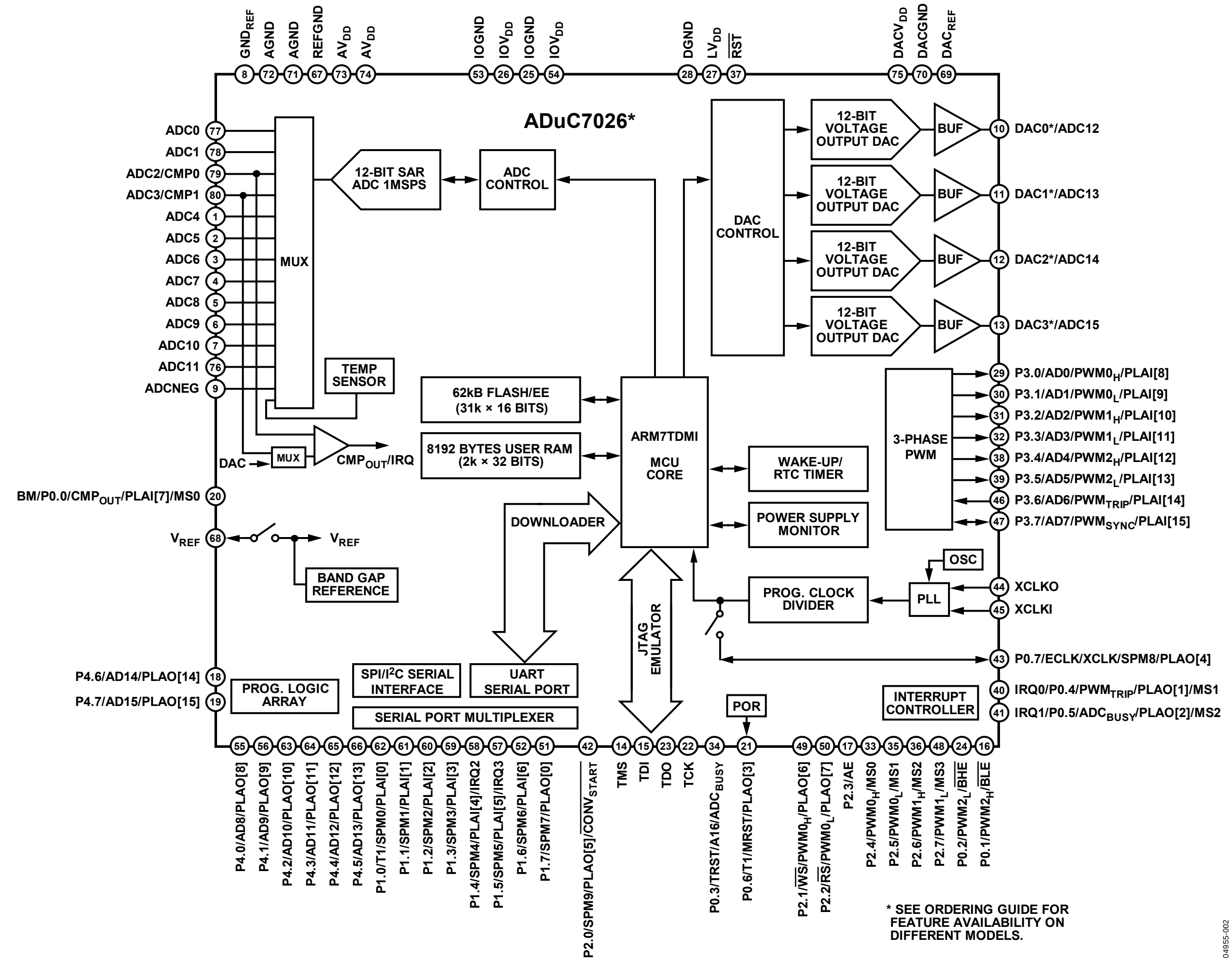

Figure 11.

Table 3. External Memory Read Cycle

Parameter	Min	Typ	Max	Unit
CLK[1]	1/MD clock	ns typ × (POWCON[2:0] + 1)		
$t_{MS\_AFTER\_CLKH}$	4		8	ns
$t_{ADDR\_AFTER\_CLKH}$	4		16	ns
$t_{AE\_H\_AFTER\_MS}$		½ CLK		
t_{AE}		(XMxPAR[14:12] + 1) × CLK		
$t_{HOLD\_ADDR\_AFTER\_AE\_L}$		½ CLK + (! XMxPAR[10]) × CLK		
$t_{RD\_L\_AFTER\_AE\_L}$		½ CLK + (! XMxPAR[10]+ ! XMxPAR[9]) × CLK		
$t_{RD\_H\_AFTER\_CLKH}$	0		4	
t_{RD}		(XMxPAR[3:0] + 1) × CLK		
$t_{DATA\_BEFORE\_RD\_H}$	16			ns
$t_{DATA\_AFTER\_RD\_H}$	8	+ (! XMxPAR[9]) × CLK		
$t_{RELEASE\_MS\_AFTER\_RD\_H}$		1 × CLK		

[1] See Table 78.

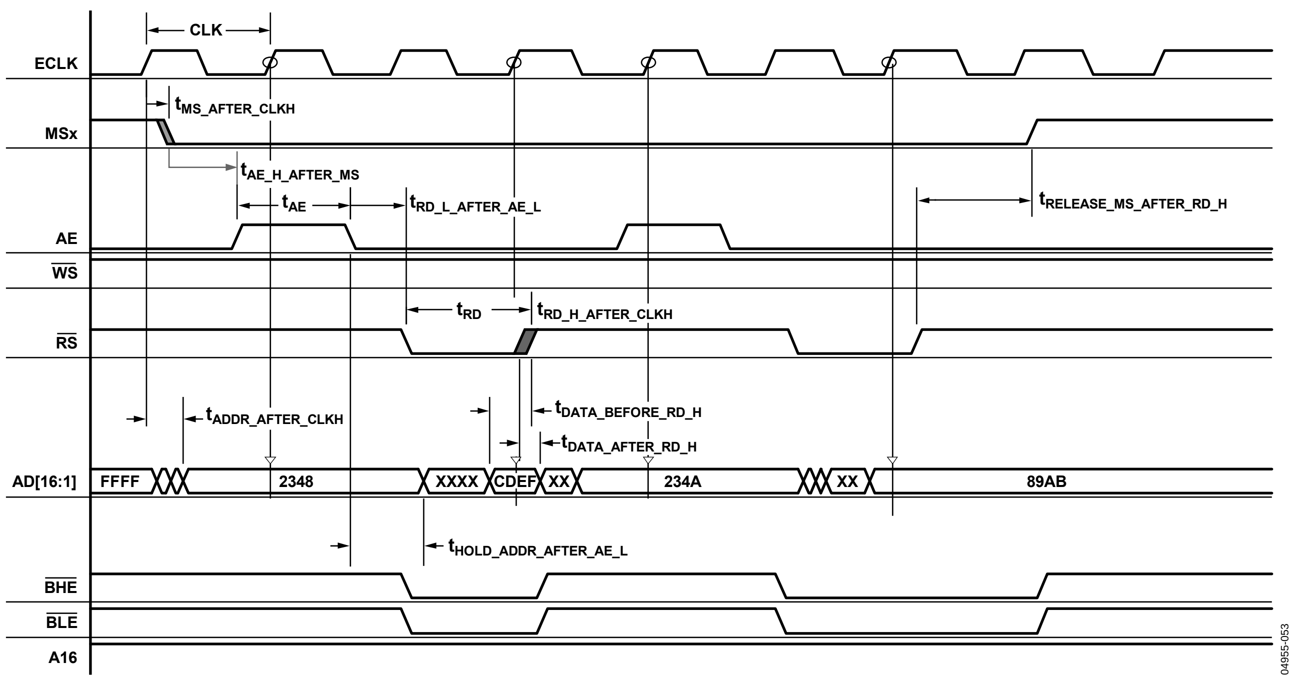

Figure 13. External Memory Read Cycle (See Table 78)

Table 11. Pin Function Descriptions (ADuC7019/ADuC7020/ADuC7021/ADuC7022)

Pin No.				
7019/7020	7021	7022	Mnemonic	Description
38	37	36	ADC0	Single-Ended or Differential Analog Input 0.
39	38	37	ADC1	Single-Ended or Differential Analog Input 1.
40	39	38	ADC2/CMP0	Single-Ended or Differential Analog Input 2/Comparator Positive Input.
1	40	39	ADC3/CMP1	Single-Ended or Differential Analog Input 3 (Buffered Input on ADuC7019)/ Comparator Negative Input.
2	1	40	ADC4	Single-Ended or Differential Analog Input 4.
–	2	1	ADC5	Single-Ended or Differential Analog Input 5.
–	3	2	ADC6	Single-Ended or Differential Analog Input 6.
–	4	3	ADC7	Single-Ended or Differential Analog Input 7.
–	–	4	ADC8	Single-Ended or Differential Analog Input 8.
–	–	5	ADC9	Single-Ended or Differential Analog Input 9.
3	5	6	GND_{REF}	Ground Voltage Reference for the ADC. For optimal performance, the analog power supply should be separated from IOGND and DGND.
4	6	–	DAC0/ADC12	DAC0 Voltage Output/Single-Ended or Differential Analog Input 12.
5	7	–	DAC1/ADC13	DAC1 Voltage Output/Single-Ended or Differential Analog Input 13.
6	–	–	DAC2/ADC14	DAC2 Voltage Output/Single-Ended or Differential Analog Input 14.
7	–	–	DAC3/ADC15	DAC3 Voltage Output on ADuC7020. On the ADuC7019, a 10 nF capacitor must be connected between this pin and AGND/Single-Ended or Differential Analog Input 15 (see Figure 53).
8	8	7	TMS	Test Mode Select, JTAG Test Port Input. Debug and download access. This pin has an internal pull-up resistor to IOV_{DD}. In some cases, an external pull-up resistor (~100K) is also required to ensure that the part does not enter an erroneous state.
9	9	8	TDI	Test Data In, JTAG Test Port Input. Debug and download access.
10	10	9	BM/P0.0/CMP_{OUT}/PLAI[7]	Multifunction I/O Pin. Boot Mode (BM). The ADuC7019/20/21/22 enter serial download mode if BM is low at reset and execute code if BM is pulled high at reset through a 1 kΩ resistor/General-Purpose Input and Output Port 0.0/Voltage Comparator Output/Programmable Logic Array Input Element 7.
11	11	10	P0.6/T1/MRST/PLAO[3]	Multifunction Pin. Driven low after reset. General-Purpose Output Port 0.6/ Timer1 Input/Power-On Reset Output/Programmable Logic Array Output Element 3.
12	12	11	TCK	Test Clock, JTAG Test Port Input. Debug and download access. This pin has an internal pull-up resistor to IOV_{DD}. In some cases an external pull-up resistor (~100K) is also required to ensure that the part does not enter an erroneous state.
13	13	12	TDO	Test Data Out, JTAG Test Port Output. Debug and download access.
14	14	13	IOGND	Ground for GPIO (see Table 78). Typically connected to DGND.
15	15	14	IOV_{DD}	3.3 V Supply for GPIO (see Table 78) and Input of the On-Chip Voltage Regulator.
16	16	15	LV_{DD}	2.6 V Output of the On-Chip Voltage Regulator. This output must be connected to a 0.47 μF capacitor to DGND only.
17	17	16	DGND	Ground for Core Logic.
18	18	17	P0.3/TRST/ADC_{BUSY}	General-Purpose Input and Output Port 0.3/Test Reset, JTAG Test Port Input/ ADC_{BUSY} Signal Output.
19	19	18	$\overline{RST}$	Reset Input, Active Low.
20	20	19	IRQ0/P0.4/PWM_{TRIP}/PLAO[1]	Multifunction I/O Pin. External Interrupt Request 0, Active High/General-Purpose Input and Output Port 0.4/PWM Trip External Input/Programmable Logic Array Output Element 1.
21	21	20	IRQ1/P0.5/ADC_{BUSY}/PLAO[2]	Multifunction I/O Pin. External Interrupt Request 1, Active High/General-Purpose Input and Output Port 0.5/ADC_{BUSY} Signal Output/Programmable Logic Array Output Element 2.

Pin No.	Mnemonic	Description
D7	P1.6/SPM6/PLAI[6]	Serial Port Multiplexed. General-Purpose Input and Output Port 1.6/UART, SPI/Programmable Logic Array Input Element 6.
D8	IOV_{DD}	3.3 V Supply for GPIO (see Table 78) and Input of the On-Chip Voltage Regulator.
E1	DAC3/ADC15	DAC3 Voltage Output/ADC Input 15.
E2	DAC2/ADC14	DAC2 Voltage Output/ADC Input 14.
E3	DAC1/ADC13	DAC1 Voltage Output/ADC Input 13.
E4	P3.0/$PWM0_H$/PLAI[8]	General-Purpose Input and Output Port 3.0/PWM Phase 0 High-Side Output/Programmable Logic Array Input Element 8.
E5	P3.2/$PWM1_H$/PLAI[10]	General-Purpose Input and Output Port 3.2/PWM Phase 1 High-Side Output/Programmable Logic Array Input Element 10.
E6	P1.5/SPM5/PLAI[5]/IRQ3	Serial Port Multiplexed. General-Purpose Input and Output Port 1.5/UART, SPI/Programmable Logic Array Input Element 5/External Interrupt Request 3, Active High.
E7	P3.7/PWM_{SYNC}/PLAI[15]	General-Purpose Input and Output Port 3.7/PWM Synchronization/Programmable Logic Array Input Element 15.
E8	XCLKI	Input to the Crystal Oscillator Inverter and Input to the Internal Clock Generator Circuits.
F1	P4.6/PLAO[14]	General-Purpose Input and Output Port 4.6/Programmable Logic Array Output Element 14.
F2	TDI	JTAG Test Port Input, Test Data In. Debug and download access.
F3	DAC0/ADC12	DAC0 Voltage Output/ADC Input 12.
F4	P3.1/$PWM0_L$/PLAI[9]	General-Purpose Input and Output Port 3.1/PWM Phase 0 Low-Side Output/Programmable Logic Array Input Element 9.
F5	P3.3/$PWM1_L$/PLAI[11]	General-Purpose Input and Output Port 3.3/PWM Phase 1 Low-Side Output/Programmable Logic Array Input Element 11.
F6	$\overline{RST}$	Reset Input, Active Low.
F7	P0.7/ECLK/XCLK/SPM8/PLAO[4]	Serial Port Multiplexed. General-Purpose Input and Output Port 0.7/Output for External Clock Signal/Input to the Internal Clock Generator Circuits/UART/Programmable Logic Array Output Element 4.
F8	XCLKO	Output from the Crystal Oscillator Inverter.
G1	BM/P0.0/CMP_{OUT}/PLAI[7]	Multifunction I/O Pin. Boot mode. The ADuC7028 enters UART download mode if BM is low at reset and executes code if BM is pulled high at reset through a 1 kΩ resistor/General-Purpose Input and Output Port 0.0/Voltage Comparator Output/Programmable Logic Array Input Element 7.
G2	P4.7/PLAO[15]	General-Purpose Input and Output Port 4.7/Programmable Logic Array Output Element 15.
G3	TMS	JTAG Test Port Input, Test Mode Select. Debug and download access.
G4	TDO	JTAG Test Port Output, Test Data Out. Debug and download access.
G5	P0.3/TRST/ADC_{BUSY}	General-Purpose Input and Output Port 0.3/JTAG Test Port Input, Test Reset/ADC_{BUSY} Signal Output.
G6	P3.4/$PWM2_H$/PLAI[12]	General-Purpose Input and Output Port 3.4/PWM Phase 2 High-Side Output/Programmable Logic Array Input 12.
G7	P3.5/$PWM2_L$/PLAI[13]	General-Purpose Input and Output Port 3.5/PWM Phase 2 Low-Side Output/Programmable Logic Array Input Element 13.
G8	P2.0/SPM9/PLAO[5]/$\overline{CONV_{START}}$	Serial Port Multiplexed. General-Purpose Input and Output Port 2.0/UART/Programmable Logic Array Output Element 5/Start Conversion Input Signal for ADC.
H1	P0.6/T1/MRST/PLAO[3]	Multifunction Pin, Driven Low After Reset. General-Purpose Output Port 0.6/Timer1 Input/Power-On Reset Output/Programmable Logic Array Output Element 3.
H2	TCK	JTAG Test Port Input, Test Clock. Debug and download access.
H3	IOGND	Ground for GPIO (see Table 78). Typically connected to DGND.
H4	IOV_{DD}	3.3 V Supply for GPIO (see Table 78) and Input of the On-Chip Voltage Regulator.
H5	LV_{DD}	2.6 V Output of the On-Chip Voltage Regulator. This output must be connected to a 0.47 µF capacitor to DGND only.
H6	DGND	Ground for Core Logic.
H7	IRQ0/P0.4/PWM_{TRIP}/PLAO[1]	Multifunction I/O Pin. External Interrupt Request 0, Active High/General-Purpose Input and Output Port 0.4/PWM Trip External Input/Programmable Logic Array Output Element 1.
H8	IRQ1/P0.5/ADC_{BUSY}/PLAO[2]	Multifunction I/O Pin. External Interrupt Request 1, Active High/General-Purpose Input and Output Port 0.5/ADC_{BUSY} Signal Output/Programmable Logic Array Output Element 2.

TYPICAL PERFORMANCE CHARACTERISTICS

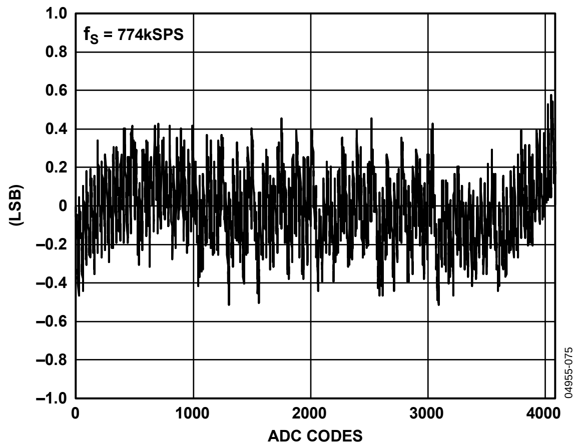

Figure 28. Typical INL Error, f_S = 774 kSPS

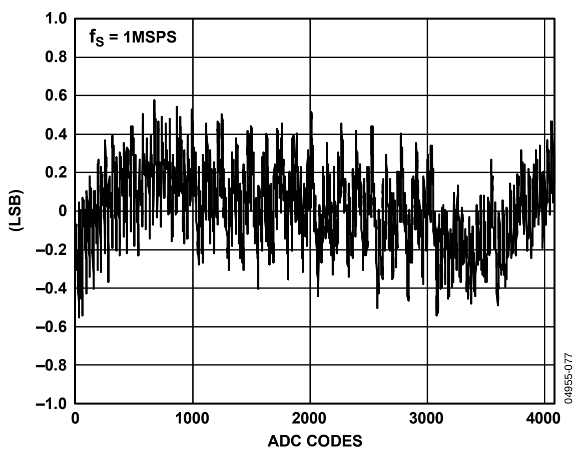

Figure 29. Typical INL Error, f_S = 1 MSPS

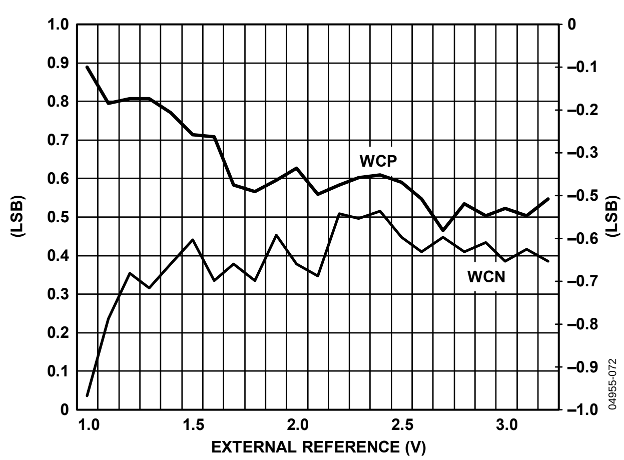

Figure 30. Typical Worst-Case (Positive (WCP) and Negative (WCN)) INL Error vs. V_{REF}, f_S = 774 kSPS

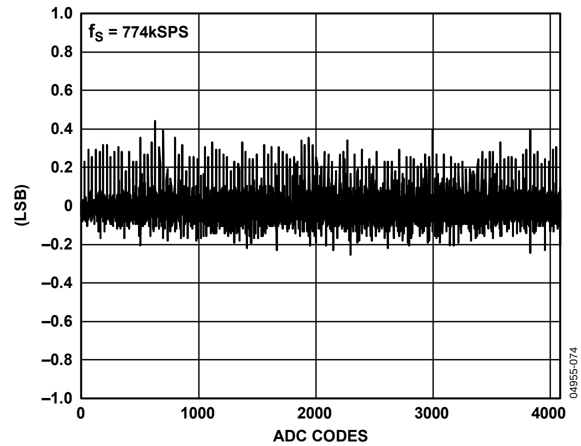

Figure 31. Typical DNL Error, f_S = 774 kSPS

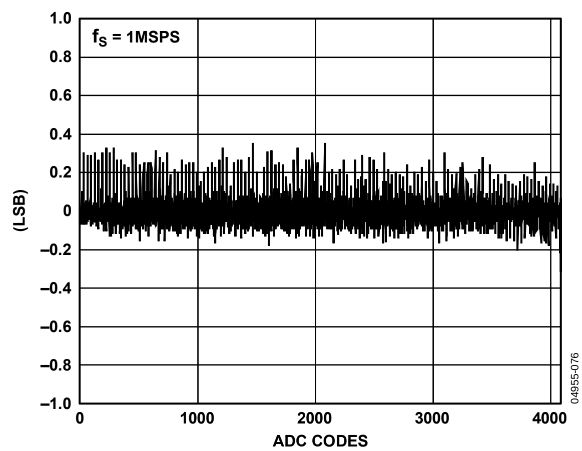

Figure 32. Typical DNL Error, f_S = 1 MSPS

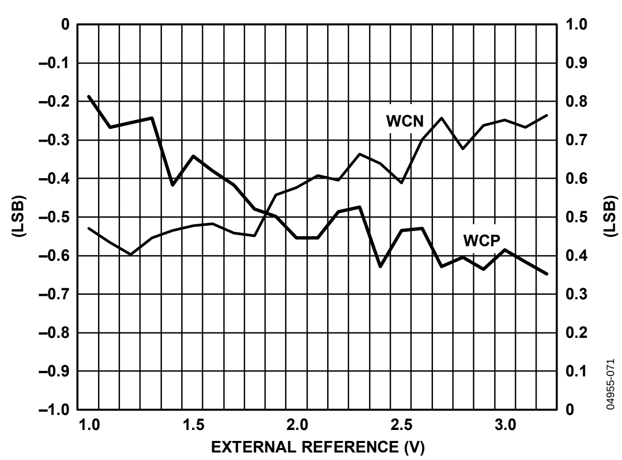

Figure 33. Typical Worst-Case (Positive (WCP)and Negative (WCN)) DNL Error vs. V_{REF}, f_S = 774 kSPS

More information relative to the programmer's model and the ARM7TDMI core architecture can be found in the following materials from ARM:

- DDI0029G, *ARM7TDMI Technical Reference Manual*
- DDI-0100, *ARM Architecture Reference Manual*

INTERRUPT LATENCY

The worst-case latency for a fast interrupt request (FIQ) consists of the following:

- The longest time the request can take to pass through the synchronizer
- The time for the longest instruction to complete (the longest instruction is an LDM) that loads all the registers including the PC
- The time for the data abort entry
- The time for FIQ entry

At the end of this time, the ARM7TDMI executes the instruction at 0x1C (FIQ interrupt vector address). The maximum total time is 50 processor cycles, which is just under 1.2 µs in a system using a continuous 41.78 MHz processor clock.

The maximum interrupt request (IRQ) latency calculation is similar but must allow for the fact that FIQ has higher priority and may delay entry into the IRQ handling routine for an arbitrary length of time. This time can be reduced to 42 cycles if the LDM command is not used. Some compilers have an option to compile without using this command. Another option is to run the part in thumb mode where the time is reduced to 22 cycles.

The minimum latency for FIQ or IRQ interrupts is a total of five cycles, which consist of the shortest time the request can take through the synchronizer plus the time to enter the exception mode.

Note that the ARM7TDMI always runs in ARM (32-bit) mode when in privileged modes, for example, when executing interrupt service routines.

TYPICAL OPERATION

Once configured via the ADC control and channel selection registers, the ADC converts the analog input and provides a 12-bit result in the ADC data register.

The top four bits are the sign bits. The 12-bit result is placed from Bit 16 to Bit 27, as shown in Figure 51. Again, it should be noted that, in fully differential mode, the result is represented in twos complement format. In pseudo differential and single-ended modes, the result is represented in straight binary format.

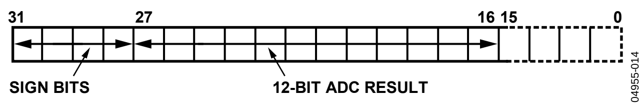

Figure 51. ADC Result Format

The same format is used in DACxDAT, simplifying the software.

Current Consumption

The ADC in standby mode, that is, powered up but not converting, typically consumes 640 µA. The internal reference adds 140 µA. During conversion, the extra current is 0.3 µA multiplied by the sampling frequency (in kilohertz (kHz)). Figure 43 shows the current consumption vs. the sampling frequency of the ADC.

Timing

Figure 52 gives details of the ADC timing. Users control the ADC clock speed and the number of acquisition clocks in the ADCCON MMR. By default, the acquisition time is eight clocks and the clock divider is 2. The number of extra clocks (such as bit trial or write) is set to 19, which gives a sampling rate of 774 kSPS. For conversion on the temperature sensor, the ADC acquisition time is automatically set to 16 clocks, and the ADC clock divider is set to 32. When using multiple channels, including the temperature sensor, the timing settings revert to the user-defined settings after reading the temperature sensor channel.

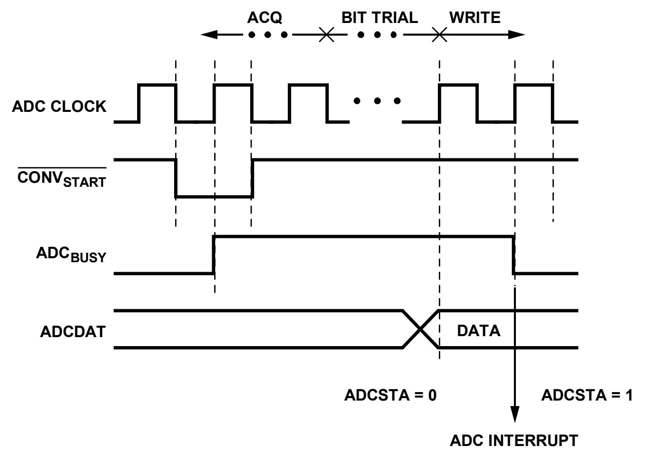

Figure 52. ADC Timing

ADuC7019

The ADuC7019 is identical to the ADuC7020 except for one buffered ADC channel, ADC3, and it has only three DACs. The output buffer of the fourth DAC is internally connected to the ADC3 channel as shown in Figure 53.

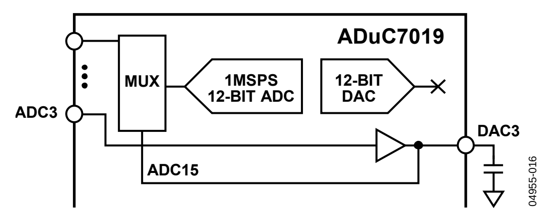

Figure 53. ADC3 Buffered Input

Note that the DAC3 output pin must be connected to a 10 nF capacitor to AGND. This channel should be used to measure dc voltages only. ADC calibration may be necessary on this channel.

MMRS INTERFACE

The ADC is controlled and configured via the eight MMRs described in this section.

Table 17. ADCCON Register

Name	Address	Default Value	Access
ADCCON	0xFFFF0500	0x0600	R/W

ADCCON is an ADC control register that allows the programmer to enable the ADC peripheral, select the mode of operation of the ADC (in single-ended mode, pseudo differential mode, or fully differential mode), and select the conversion type. This MMR is described in Table 18.

SECURITY

The 62 kB of Flash/EE memory available to the user can be read and write protected.

Bit 31 of the FEEPRO/FEEHIDE MMR (see Table 42) protects the 62 kB from being read through JTAG programming mode. The other 31 bits of this register protect writing to the flash memory. Each bit protects four pages, that is, 2 kB. Write protection is activated for all types of access.

Three Levels of Protection

- Protection can be set and removed by writing directly into FEEHIDE MMR. This protection does not remain after reset.
- Protection can be set by writing into the FEEPRO MMR. It takes effect only after a save protection command (0x0C) and a reset. The FEEPRO MMR is protected by a key to avoid direct access. The key is saved once and must be entered again to modify FEEPRO. A mass erase sets the key back to 0xFFFF but also erases all the user code.
- Flash can be permanently protected by using the FEEPRO MMR and a particular value of key: 0xDEADDEAD. Entering the key again to modify the FEEPRO register is not allowed.

Sequence to Write the Key

1. Write the bit in FEEPRO corresponding to the page to be protected.
2. Enable key protection by setting Bit 6 of FEEMOD (Bit 5 must equal 0).
3. Write a 32-bit key in FEEADR and FEEDAT.
4. Run the write key command 0x0C in FEECON; wait for the read to be successful by monitoring FEESTA.
5. Reset the part.

To remove or modify the protection, the same sequence is used with a modified value of FEEPRO. If the key chosen is the value 0xDEAD, the memory protection cannot be removed. Only a mass erase unprotects the part, but it also erases all user code.

The sequence to write the key is illustrated in the following example (this protects writing Page 4 to Page 7 of the Flash):

```
FEEPRO=0xFFFFFFFD;      //Protect pages 4 to 7
FEEMOD=0x48;            //Write key enable
FEEADR=0x1234;          //16 bit key value
FEEDAT=0x5678;          //16 bit key value
FEECON= 0x0C;           // Write key command
```

The same sequence should be followed to protect the part permanently with FEEADR = 0xDEAD and FEEDAT = 0xDEAD.

FLASH/EE CONTROL INTERFACE

Serial and JTAG programming use the Flash/EE control interface, which includes the eight MMRs outlined in this section.

Table 31. FEESTA Register

Name	Address	Default Value	Access
FEESTA	0xFFFFF800	0x20	R

FEESTA is a read-only register that reflects the status of the flash control interface as described in Table 32.

Table 32. FEESTA MMR Bit Designations

Bit	Description
15:6	Reserved.
5	Reserved.
4	Reserved.
3	Flash interrupt status bit. Set automatically when an interrupt occurs, that is, when a command is complete and the Flash/EE interrupt enable bit in the FEEMOD register is set. Cleared when reading the FEESTA register.
2	Flash/EE controller busy. Set automatically when the controller is busy. Cleared automatically when the controller is not busy.
1	Command fail. Set automatically when a command completes unsuccessfully. Cleared automatically when reading the FEESTA register.
0	Command pass. Set by the MicroConverter when a command completes successfully. Cleared automatically when reading the FEESTA register.

Table 33. FEEMOD Register

Name	Address	Default Value	Access
FEEMOD	0xFFFFF804	0x0000	R/W

FEEMOD sets the operating mode of the flash control interface. Table 34 shows FEEMOD MMR bit designations.

Table 34. FEEMOD MMR Bit Designations

Bit	Description
15:9	Reserved.
8	Reserved. This bit should always be set to 0.
7:5	Reserved. These bits should always be set to 0 except when writing keys. See the Sequence to Write the Key section.
4	Flash/EE interrupt enable. Set by user to enable the Flash/EE interrupt. The interrupt occurs when a command is complete. Cleared by user to disable the Flash/EE interrupt.
3	Erase/write command protection. Set by user to enable the erase and write commands. Cleared to protect the Flash against the erase/write command.
2:0	Reserved. These bits should always be set to 0.

OTHER ANALOG PERIPHERALS

DAC

The ADuC7019/20/21/22/24/25/26/27/28/29 incorporate two, three, or four 12-bit voltage output DACs on-chip, depending on the model. Each DAC has a rail-to-rail voltage output buffer capable of driving 5 kΩ/100 pF.

Each DAC has three selectable ranges: 0 V to V_{REF} (internal band gap 2.5 V reference), 0 V to DAC_{REF}, and 0 V to AV_{DD}. DAC_{REF} is equivalent to an external reference for the DAC. The signal range is 0 V to AV_{DD}.

MMRs Interface

Each DAC is independently configurable through a control register and a data register. These two registers are identical for the four DACs. Only DAC0CON (see Table 50) and DAC0DAT (see Table 52) are described in detail in this section.

Table 49. DACxCON Registers

Name	Address	Default Value	Access
DAC0CON	0xFFFF0600	0x00	R/W
DAC1CON	0xFFFF0608	0x00	R/W
DAC2CON	0xFFFF0610	0x00	R/W
DAC3CON	0xFFFF0618	0x00	R/W

Table 50. DAC0CON MMR Bit Designations

Bit	Name	Value	Description
7:6			Reserved.
5	DACCLK		DAC update rate. Set by user to update the DAC using Timer1. Cleared by user to update the DAC using HCLK (core clock).
4	DACCLR		DAC clear bit. Set by user to enable normal DAC operation. Cleared by user to reset data register of the DAC to 0.
3			Reserved. This bit should be left at 0.
2			Reserved. This bit should be left at 0.
1:0			DAC range bits.
		00	Power-down mode. The DAC output is in three-state.
		01	0 V to DAC_{REF} range.
		10	0 V to V_{REF} (2.5 V) range.
		11	0 V to AV_{DD} range.

Table 51. DACxDAT Registers

Name	Address	Default Value	Access
DAC0DAT	0xFFFF0604	0x00000000	R/W
DAC1DAT	0xFFFF060C	0x00000000	R/W
DAC2DAT	0xFFFF0614	0x00000000	R/W
DAC3DAT	0xFFFF061C	0x00000000	R/W

Table 52. DAC0DAT MMR Bit Designations

Bit	Description
31:28	Reserved.
27:16	12-bit data for DAC0.
15:0	Reserved.

Using the DACs

The on-chip DAC architecture consists of a resistor string DAC followed by an output buffer amplifier. The functional equivalent is shown in Figure 63.

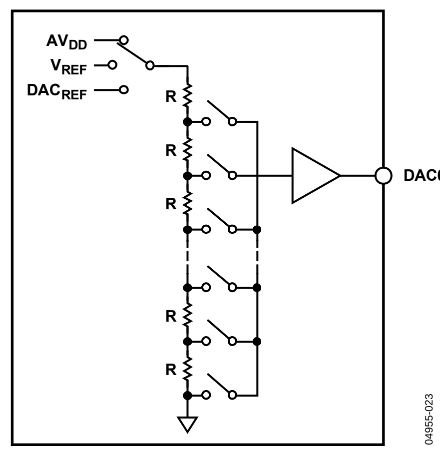

Figure 63. DAC Structure

As illustrated in Figure 63, the reference source for each DAC is user-selectable in software. It can be AV_{DD}, V_{REF}, or DAC_{REF}. In 0-to-AV_{DD} mode, the DAC output transfer function spans from 0 V to the voltage at the AV_{DD} pin. In 0-to-DAC_{REF} mode, the DAC output transfer function spans from 0 V to the voltage at the DAC_{REF} pin. In 0-to-V_{REF} mode, the DAC output transfer function spans from 0 V to the internal 2.5 V reference, V_{REF}.

The DAC output buffer amplifier features a true, rail-to-rail output stage implementation. This means that when unloaded, each output is capable of swinging to within less than 5 mV of both AV_{DD} and ground. Moreover, the DAC's linearity specification (when driving a 5 kΩ resistive load to ground) is guaranteed through the full transfer function, except Code 0 to Code 100, and, in 0-to-AV_{DD} mode only, Code 3995 to Code 4095.

DIGITAL PERIPHERALS

3-PHASE PWM

Each ADuC7019/20/21/22/24/25/26/27/28/29 provides a flexible and programmable, 3-phase pulse-width modulation (PWM) waveform generator. It can be programmed to generate the required switching patterns to drive a 3-phase voltage source inverter for ac induction motor control (ACIM). Note that only active high patterns can be produced.

The PWM generator produces three pairs of PWM signals on the six PWM output pins ($PWM0_H$, $PWM0_L$, $PWM1_H$, $PWM1_L$, $PWM2_H$, and $PWM2_L$). The six PWM output signals consist of three high-side drive signals and three low-side drive signals.

The switching frequency and dead time of the generated PWM patterns are programmable using the PWMDAT0 and PWMDAT1 MMRs. In addition, three duty-cycle control registers (PWMCH0, PWMCH1, and PWMCH2) directly control the duty cycles of the three pairs of PWM signals.

Each of the six PWM output signals can be enabled or disabled by separate output enable bits of the PWMEN register. In addition, three control bits of the PWMEN register permit crossover of the two signals of a PWM pair. In crossover mode, the PWM signal destined for the high-side switch is diverted to the complementary low-side output. The signal destined for the low-side switch is diverted to the corresponding high-side output signal.

In many applications, there is a need to provide an isolation barrier in the gate-drive circuits that turn on the inverter power devices. In general, there are two common isolation techniques: optical isolation using optocouplers and transformer isolation using pulse transformers. The PWM controller permits mixing of the output PWM signals with a high frequency chopping signal to permit easy interface to such pulse transformers. The features of this gate-drive chopping mode can be controlled by the PWMCFG register. An 8-bit value within the PWMCFG register directly controls the chopping frequency. High frequency chopping can be independently enabled for the high-side and low-side outputs using separate control bits in the PWMCFG register.

The PWM generator can operate in one of two distinct modes: single update mode or double update mode. In single update mode, the duty cycle values are programmable only once per PWM period so that the resulting PWM patterns are symmetrical about the midpoint of the PWM period. In the double update mode, a second updating of the PWM duty cycle values is implemented at the midpoint of the PWM period.

In double update mode, it is also possible to produce asymmetrical PWM patterns that produce lower harmonic distortion in 3-phase PWM inverters. This technique permits closed-loop controllers to change the average voltage applied to the machine windings at a faster rate. As a result, faster closed-loop bandwidths are achieved. The operating mode of the PWM block is selected by a control bit in the PWMCON register. In single update mode, an internal synchronization pulse, PWMSYNC, is produced at the start of each PWM period. In double update mode, an additional PWMSYNC pulse is produced at the midpoint of each PWM period.

The PWM block can also provide an internal synchronization pulse on the PWM_{SYNC} pin that is synchronized to the PWM switching frequency. In single update mode, a pulse is produced at the start of each PWM period. In double update mode, an additional pulse is produced at the mid-point of each PWM period. The width of the pulse is programmable through the PWMDAT2 register. The PWM block can also accept an external synchronization pulse on the PWM_{SYNC} pin. The selection of external synchronization or internal synchronization is in the PWMCON register. The SYNC input timing can be synchronized to the internal peripheral clock, which is selected in the PWMCON register. If the external synchronization pulse from the chip pin is asynchronous to the internal peripheral clock (typical case), the external PWMSYNC is considered asynchronous and should be synchronized. The synchronization logic adds latency and jitter from the external pulse to the actual PWM outputs. The size of the pulse on the PWM_{SYNC} pin must be greater than two core clock periods.

The PWM signals produced by the ADuC7019/20/21/22/24/25/26/27/28/29 can be shut off via a dedicated asynchronous PWM shutdown pin, PWM_{TRIP}. When brought low, PWM_{TRIP} instantaneously places all six PWM outputs in the off state (high). This hardware shutdown mechanism is asynchronous so that the associated PWM disable circuitry does not go through any clocked logic. This ensures correct PWM shutdown even in the event of a core clock loss.

Status information about the PWM system is available to the user in the PWMSTA register. In particular, the state of the PWM_{TRIP} pin is available, as well as a status bit that indicates whether operation is in the first half or the second half of the PWM period.

40-Pin Package Devices

On the 40-pin package devices, the PWM outputs are not directly accessible, as described in the General-Purpose Input/Output section. One channel can be brought out on a GPIO (see Table 78) via the PLA as shown in the following example:

```
PWMCON = 0x1;            // enables PWM o/p
PWMDAT0 = 0x055F;        // PWM switching freq

// Configure Port Pins
GP4CON = 0x300;          // P4.2 as PLA output
GP3CON = 0x1;            // P3.0 configured as
                         // output of PWM0
                         //(internally)

// PWM0 onto P4.2
PLAELM8 = 0x0035;        // P3.0 (PWM output)
                         // input of element 8
PLAELM10 = 0x0059;       // PWM from element 8
```

Output Control Unit

The operation of the output control unit is controlled by the 9-bit read/write PWMEN register. This register controls two distinct features of the output control unit that are directly useful in the control of electronic counter measures (ECM) or binary decimal counter measures (BDCM). The PWMEN register contains three crossover bits, one for each pair of PWM outputs. Setting Bit 8 of the PWMEN register enables the crossover mode for the 0H/0L pair of PWM signals, setting Bit 7 enables crossover on the 1H/1L pair of PWM signals, and setting Bit 6 enables crossover on the 2H/2L pair of PWM signals. If crossover mode is enabled for any pair of PWM signals, the high-side PWM signal from the timing unit (0H, for example) is diverted to the associated low-side output of the output control unit so that the signal ultimately appears at the $PWM0_L$ pin. Of course, the corresponding low-side output of the timing unit is also diverted to the complementary high-side output of the output control unit so that the signal appears at the $PWM0_H$ pin. Following a reset, the three crossover bits are cleared, and the crossover mode is disabled on all three pairs of PWM signals. The PWMEN register also contains six bits (Bit 0 to Bit 5) that can be used to individually enable or disable each of the six PWM outputs. If the associated bit of the PWMEN register is set, the corresponding PWM output is disabled regardless of the corresponding value of the duty cycle register. This PWM output signal remains in the off state as long as the corresponding enable/disable bit of the PWMEN register is set. The implementation of this output enable function is implemented after the crossover function.

Following a reset, all six enable bits of the PWMEN register are cleared, and all PWM outputs are enabled by default. In a manner identical to the duty cycle registers, the PWMEN is latched on the rising edge of the PWMSYNC signal. As a result, changes to this register become effective only at the start of each PWM cycle in single update mode. In double update mode, the PWMEN register can also be updated at the midpoint of the PWM cycle.

In the control of an ECM, only two inverter legs are switched at any time, and often the high-side device in one leg must be switched on at the same time as the low-side driver in a second leg. Therefore, by programming identical duty cycle values for two PWM channels (for example, PWMCH0 = PWMCH1) and setting Bit 7 of the PWMEN register to cross over the 1H/1L pair of PWM signals, it is possible to turn on the high-side switch of Phase A and the low-side switch of Phase B at the same time. In the control of ECM, it is usual for the third inverter leg (Phase C in this example) to be disabled for a number of PWM cycles. This function is implemented by disabling both the 2H and 2L PWM outputs by setting Bit 0 and Bit 1 of the PWMEN register.

This situation is illustrated in Figure 71, where it can be seen that both the 0H and 1L signals are identical because PWMCH0 = PWMCH1 and the crossover bit for Phase B is set.

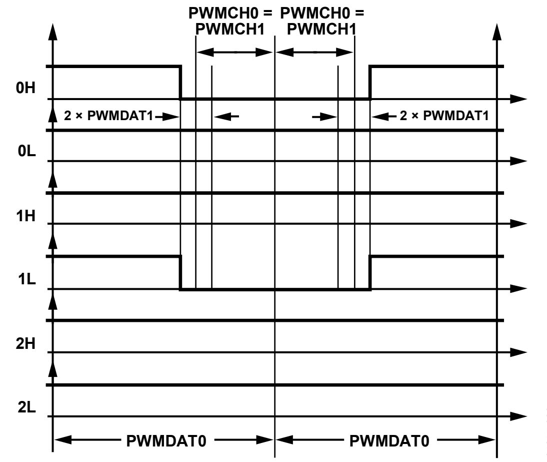

Figure 71. Active Low PWM Signals Suitable for ECM Control, PWMCH0 = PWMCH1, Crossover 1H/1L Pair and Disable 0L, 1H, 2H, and 2L Outputs in Single Update Mode.

In addition, the other four signals (0L, 1H, 2H, and 2L) have been disabled by setting the appropriate enable/disable bits of the PWMEN register. In Figure 71, the appropriate value for the PWMEN register is 0x00A7. In normal ECM operation, each inverter leg is disabled for certain periods of time to change the PWMEN register based on the position of the rotor shaft (motor commutation).

Gate Drive Unit

The gate drive unit of the PWM controller adds features that simplify the design of isolated gate-drive circuits for PWM inverters. If a transformer-coupled, power device, gate-drive amplifier is used, the active PWM signal must be chopped at a high frequency. The 16-bit read/write PWMCFG register programs this high frequency chopping mode. The chopped active PWM signals can be required for the high-side drivers only, the low-side drivers only, or both the high-side and low-side switches. Therefore, independent control of this mode for both high-side and low-side switches is included with two separate control bits in the PWMCFG register.

Typical PWM output signals with high frequency chopping enabled on both high-side and low-side signals are shown in Figure 72. Chopping of the high-side PWM outputs (0H, 1H, and 2H) is enabled by setting Bit 8 of the PWMCFG register. Chopping of the low-side PWM outputs (0L, 1L, and 2L) is enabled by setting Bit 9 of the PWMCFG register. The high chopping frequency is controlled by the 8-bit word (GDCLK) placed in Bit 0 to Bit 7 of the PWMCFG register. The period of this high frequency carrier is

$$t_{CHOP} = (4 \times (GDCLK + 1)) \times t_{CORE}$$

The chopping frequency is, therefore, an integral subdivision of the MicroConverter core frequency

$$f_{CHOP} = f_{CORE}/(4 \times (GDCLK + 1))$$

The GDCLK value can range from 0 to 255, corresponding to a programmable chopping frequency rate of 40.8 kHz to 10.44 MHz for a 41.78 MHz core frequency. The gate drive features must be programmed before operation of the PWM controller and are typically not changed during normal operation of the PWM controller. Following a reset, all bits of the PWMCFG register are cleared so that high frequency chopping is disabled, by default.

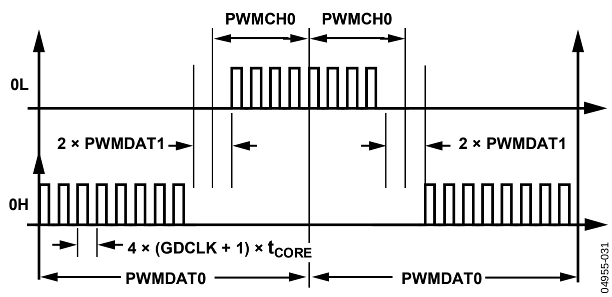

Figure 72. Typical PWM Signals with High Frequency Gate Chopping Enabled on Both High-Side and Low-Side Switches

PWM Shutdown

In the event of external fault conditions, it is essential that the PWM system be instantaneously shut down in a safe fashion. A low level on the PWM_{TRIP} pin provides an instantaneous, asynchronous (independent of the MicroConverter core clock) shutdown of the PWM controller. All six PWM outputs are placed in the off state, that is, in low state. In addition, the PWMSYNC pulse is disabled. The PWM_{TRIP} pin has an internal pull-down resistor to disable the PWM if the pin becomes disconnected. The state of the PWM_{TRIP} pin can be read from Bit 3 of the PWMSTA register.

If a PWM shutdown command occurs, a PWMTRIP interrupt is generated, and internal timing of the 3-phase timing unit of the PWM controller is stopped. Following a PWM shutdown, the PWM can be reenabled (in a PWMTRIP interrupt service routine, for example) only by writing to all of the PWMDAT0, PWMCH0, PWMCH1, and PWMCH2 registers. Provided that the external fault is cleared and the PWMTRIP is returned to a high level, the internal timing of the 3-phase timing unit resumes, and new duty-cycle values are latched on the next PWMSYNC boundary.

Note that the PWMTRIP interrupt is available in IRQ only, and the PWMSYNC interrupt is available in FIQ only. Both interrupts share the same bit in the interrupt controller. Therefore, only one of the interrupts can be used at a time. See the Interrupt System section for further details.

PWM MMRs Interface

The PWM block is controlled via the MMRs described in this section.

Table 66. PWMCON Register

Name	Address	Default Value	Access
PWMCON	0xFFFFFC00	0x0000	R/W

PWMCON is a control register that enables the PWM and chooses the update rate.

Table 67. PWMCON MMR Bit Descriptions

Bit	Name	Description
7:5		Reserved.
4	PWM_SYNCSEL	External sync select. Set to use external sync. Cleared to use internal sync.
3	PWM_EXTSYNC	External sync select. Set to select external synchronous sync signal. Cleared for asynchronous sync signal.
2	PWMDBL	Double update mode. Set to 1 by user to enable double update mode. Cleared to 0 by the user to enable single update mode.
1	PWM_SYNC_EN	PWM synchronization enable. Set by user to enable synchronization. Cleared by user to disable synchronization.
0	PWMEN	PWM enable bit. Set to 1 by user to enable the PWM. Cleared to 0 by user to disable the PWM. Also cleared automatically with PWMTRIP (PWMSTA MMR).

Table 68. PWMSTA Register

Name	Address	Default Value	Access
PWMSTA	0xFFFFFC04	0x0000	R/W

PWMSTA reflects the status of the PWM.

Table 69. PWMSTA MMR Bit Descriptions

Bit	Name	Description
15:10		Reserved.
9	PWMSYNCINT	PWM sync interrupt bit. Writing a 1 to this bit clears this interrupt.
8	PWMTRIPINT	PWM trip interrupt bit. Writing a 1 to this bit clears this interrupt.
3	PWMTRIP	Raw signal from the PWM_{TRIP} pin.
2:1		Reserved.
0	PWMPHASE	PWM phase bit. Set to 1 by the Micro-Converter when the timer is counting down (first half). Cleared to 0 by the MicroConverter when the timer is counting up (second half).

Table 85. GPxDAT Registers

Name	Address	Default Value[1]	Access
GP0DAT	0xFFFFF420	0x000000XX	R/W
GP1DAT	0xFFFFF430	0x000000XX	R/W
GP2DAT	0xFFFFF440	0x000000XX	R/W
GP3DAT	0xFFFFF450	0x000000XX	R/W
GP4DAT	0xFFFFF460	0x000000XX	R/W

[1] X = 0, 1, 2, or 3.

GPxDAT are Port x configuration and data registers. They configure the direction of the GPIO pins of Port x, set the output value for the pins configured as output, and store the input value of the pins configured as input.

Table 86. GPxDAT MMR Bit Descriptions

Bit	Description
31:24	Direction of the data. Set to 1 by user to configure the GPIO pin as an output. Cleared to 0 by user to configure the GPIO pin as an input.
23:16	Port x data output.
15:8	Reflect the state of Port x pins at reset (read only).
7:0	Port x data input (read only).

Table 87. GPxSET Registers

Name	Address	Default Value[1]	Access
GP0SET	0xFFFFF424	0x000000XX	W
GP1SET	0xFFFFF434	0x000000XX	W
GP2SET	0xFFFFF444	0x000000XX	W
GP3SET	0xFFFFF454	0x000000XX	W
GP4SET	0xFFFFF464	0x000000XX	W

[1] X = 0, 1, 2, or 3.

GPxSET are data set Port x registers.

Table 88. GPxSET MMR Bit Descriptions

Bit	Description
31:24	Reserved.
23:16	Data Port x set bit. Set to 1 by user to set bit on Port x; also sets the corresponding bit in the GPxDAT MMR. Cleared to 0 by user; does not affect the data out.
15:0	Reserved.

Table 89. GPxCLR Registers

Name	Address	Default Value[1]	Access
GP0CLR	0xFFFFF428	0x000000XX	W
GP1CLR	0xFFFFF438	0x000000XX	W
GP2CLR	0xFFFFF448	0x000000XX	W
GP3CLR	0xFFFFF458	0x000000XX	W
GP4CLR	0xFFFFF468	0x000000XX	W

[1] X = 0, 1, 2, or 3.

GPxCLR are data clear Port x registers.

Table 90. GPxCLR MMR Bit Descriptions

Bit	Description
31:24	Reserved.
23:16	Data Port x clear bit. Set to 1 by user to clear bit on Port x; also clears the corresponding bit in the GPxDAT MMR. Cleared to 0 by user; does not affect the data out.
15:0	Reserved.

SERIAL PORT MUX

The serial port mux multiplexes the serial port peripherals (an SPI, UART, and two I^2Cs) and the programmable logic array (PLA) to a set of 10 GPIO pins. Each pin must be configured to one of its specific I/O functions as described in Table 91.

Table 91. SPM Configuration

SPMMUX	GPIO (00)	UART (01)	UART/I^{2}C/SPI (10)	PLA (11)
SPM0	P1.0	SIN	I2C0SCL	PLAI[0]
SPM1	P1.1	SOUT	I2C0SDA	PLAI[1]
SPM2	P1.2	RTS	I2C1SCL	PLAI[2]
SPM3	P1.3	CTS	I2C1SDA	PLAI[3]
SPM4	P1.4	RI	SCLK	PLAI[4]
SPM5	P1.5	DCD	MISO	PLAI[5]
SPM6	P1.6	DSR	MOSI	PLAI[6]
SPM7	P1.7	DTR	$\overline{CS}$	PLAO[0]
SPM8	P0.7	ECLK/XCLK	SIN	PLAO[4]
SPM9	P2.0	CONV	SOUT	PLAO[5]

Table 91 also details the mode for each of the SPMMUX pins. This configuration must be done via the GP0CON, GP1CON, and GP2CON MMRs. By default, these 10 pins are configured as GPIOs.

UART SERIAL INTERFACE

The UART peripheral is a full-duplex, universal, asynchronous receiver/transmitter. It is fully compatible with the 16,450 serial port standard. The UART performs serial-to-parallel conversions on data characters received from a peripheral device or modem, and parallel-to-serial conversions on data characters received from the CPU. The UART includes a fractional divider for baud rate generation and has a network addressable mode. The UART function is made available on the 10 pins of the ADuC7019/20/21/22/24/25/26/27/28/29 (see Table 92).

Table 92. UART Signal Description

Pin	Signal	Description
SPM0 (Mode 1)	SIN	Serial receive data.
SPM1 (Mode 1)	SOUT	Serial transmit data.
SPM2 (Mode 1)	RTS	Request to send.
SPM3 (Mode 1)	CTS	Clear to send.
SPM4 (Mode 1)	RI	Ring indicator.
SPM5 (Mode 1)	DCD	Data carrier detect.
SPM6 (Mode 1)	DSR	Data set ready.
SPM7 (Mode 1)	DTR	Data terminal ready.
SPM8 (Mode 2)	SIN	Serial receive data.
SPM9 (Mode 2)	SOUT	Serial transmit data.

Table 108. COMSTA1 Register

Name	Address	Default Value	Access
COMSTA1	0xFFFF0718	0x00	R

COMSTA1 is a modem status register.

Table 109. COMSTA1 MMR Bit Descriptions

Bit	Name	Description
7	DCD	Data carrier detect.
6	RI	Ring indicator.
5	DSR	Data set ready.
4	CTS	Clear to send.
3	DDCD	Delta DCD. Set automatically if DCD changed state since last COMSTA1 read. Cleared automatically by reading COMSTA1.
2	TERI	Trailing edge RI. Set if RI changed from 0 to 1 since COMSTA1 was last read. Cleared automatically by reading COMSTA1.
1	DDSR	Delta DSR. Set automatically if DSR changed state since COMSTA1 was last read. Cleared automatically by reading COMSTA1.
0	DCTS	Delta CTS. Set automatically if CTS changed state since COMSTA1 was last read. Cleared automatically by reading COMSTA1.

Table 110. COMSCR Register

Name	Address	Default Value	Access
COMSCR	0xFFFF071C	0x00	R/W

COMSCR is an 8-bit scratch register used for temporary storage. It is also used in network addressable UART mode.

Table 111. COMDIV2 Register

Name	Address	Default Value	Access
COMDIV2	0xFFFF072C	0x0000	R/W

COMDIV2 is a 16-bit fractional baud divide register.

Table 112. COMDIV2 MMR Bit Descriptions

Bit	Name	Description
15	FBEN	Fractional baud rate generator enable bit. Set by user to enable the fractional baud rate generator. Cleared by user to generate baud rate using the standard 450 UART baud rate generator.
14:13		Reserved.
12:11	FBM[1:0]	M if FBM = 0, M = 4 (see the Fractional Divider section).
10:0	FBN[10:0]	N (see the Fractional Divider section).

Network Addressable UART Mode

This mode connects the MicroConverter to a 256-node serial network, either as a hardware single master or via software in a multimaster network. Bit 7 (ENAM) of the COMIEN1 register must be set to enable UART in network addressable mode (see Table 114). Note that there is no parity check in this mode.

Network Addressable UART Register Definitions

Four additional registers, COMIEN0, COMIEN1, COMIID1, and COMADR are used in network addressable UART mode only.

In network address mode, the least significant bit of the COMIEN1 register is the transmitted network address control bit. If set to 1, the device is transmitting an address. If cleared to 0, the device is transmitting data. For example, the following master-based code transmits the slave's address followed by the data:

```
  COMIEN1  = 0xE7;       //Setting ENAM,
E9BT, E9BR, ETD, NABP
  COMTX = 0xA0;  // Slave address is 0xA0
  while(!(0x020==(COMSTA0 & 0x020))){} //
wait for adr tx to finish.
  COMIEN1  = 0xE6;       //  Clear NAB bit
to indicate Data is coming
   COMTX = 0x55; // Tx data to slave: 0x55
```

Table 113. COMIEN1 Register

Name	Address	Default Value	Access
COMIEN1	0xFFFF0720	0x04	R/W

COMIEN1 is an 8-bit network enable register.

Table 114. COMIEN1 MMR Bit Descriptions

Bit	Name	Description
7	ENAM	Network address mode enable bit. Set by user to enable network address mode. Cleared by user to disable network address mode.
6	E9BT	9-bit transmit enable bit. Set by user to enable 9-bit transmit. ENAM must be set. Cleared by user to disable 9-bit transmit.
5	E9BR	9-bit receive enable bit. Set by user to enable 9-bit receive. ENAM must be set. Cleared by user to disable 9-bit receive.
4	ENI	Network interrupt enable bit.
3	E9BD	Word length. Set for 9-bit data. E9BT has to be cleared. Cleared for 8-bit data.
2	ETD	Transmitter pin driver enable bit. Set by user to enable SOUT pin as an output in slave mode or multimaster mode. Cleared by user; SOUT is three-state.
1	NABP	Network address bit. Interrupt polarity bit.
0	NAB	Network address bit (if NABP = 1). Set by user to transmit the slave address. Cleared by user to transmit data.

Table 115. COMIID1 Register

Name	Address	Default Value	Access
COMIID1	0xFFFF0724	0x01	R

COMIID1 is an 8-bit network interrupt register. Bit 7 to Bit 4 are reserved (see Table 116).

I²C-COMPATIBLE INTERFACES

The ADuC7019/20/21/22/24/25/26/27/28/29 support two licensed I²C interfaces. The I²C interfaces are both implemented as a hard-ware master and a full slave interface. Because the two I²C inter-faces are identical, this data sheet describes only I2C0 in detail. Note that the two masters and one of the slaves have individual interrupts (see the Interrupt System section).

Note that when configured as an I²C master device, the ADuC7019/20/21/22/24/25/26/27/28/29 cannot generate a repeated start condition.

The two GPIO pins used for data transfer, SDAx and SCLx, are configured in a wired-AND format that allows arbitration in a multimaster system. These pins require external pull-up resistors. Typical pull-up values are 10 kΩ.

The I²C bus peripheral address in the I²C bus system is pro-grammed by the user. This ID can be modified any time a transfer is not in progress. The user can configure the interface to respond to four slave addresses.

The transfer sequence of an I²C system consists of a master device initiating a transfer by generating a start condition while the bus is idle. The master transmits the slave device address and the direction of the data transfer during the initial address transfer. If the master does not lose arbitration and the slave acknowledges, the data transfer is initiated. This continues until the master issues a stop condition and the bus becomes idle.

The I²C peripheral can be configured only as a master or slave at any given time. The same I²C channel cannot simultaneously support master and slave modes.

Serial Clock Generation

The I²C master in the system generates the serial clock for a transfer. The master channel can be configured to operate in fast mode (400 kHz) or standard mode (100 kHz).

The bit rate is defined in the I2C0DIV MMR as follows:

$$f_{SERIAL\ CLOCK} = \frac{f_{UCLK}}{(2 + DIVH) + (2 + DIVL)}$$

where:
f_{UCLK} = clock before the clock divider.
$DIVH$ = the high period of the clock.
$DIVL$ = the low period of the clock.

Thus, for 100 kHz operation,

$$DIVH = DIVL = 0xCF$$

and for 400 kHz,

$$DIVH = 0x28, DIVL = 0x3C$$

The I2CxDIV registers correspond to DIVH:DIVL.

Slave Addresses

The registers I2C0ID0, I2C0ID1, I2C0ID2, and I2C0ID3 contain the device IDs. The device compares the four I2C0IDx registers to the address byte. To be correctly addressed, the seven MSBs of either ID register must be identical to that of the seven MSBs of the first received address byte. The LSB of the ID registers (the transfer direction bit) is ignored in the process of address recognition.

I²C Registers

The I²C peripheral interface consists of 18 MMRs, which are discussed in this section.

Table 126. I2CxMSTA Registers

Name	Address	Default Value	Access
I2C0MSTA	0xFFFF0800	0x00	R/W
I2C1MSTA	0xFFFF0900	0x00	R/W

I2CxMSTA are status registers for the master channel.

Table 127. I2C0MSTA MMR Bit Descriptions

Bit	Access Type	Description
7	R/W	Master transmit FIFO flush. Set by user to flush the master Tx FIFO. Cleared automatically after the master Tx FIFO is flushed. This bit also flushes the slave receive FIFO.
6	R	Master busy. Set automatically if the master is busy. Cleared automatically.
5	R	Arbitration loss. Set in multimaster mode if another master has the bus. Cleared when the bus becomes available.
4	R	No ACK. Set automatically if there is no acknowledge of the address by the slave device. Cleared automatically by reading the I2C0MSTA register.
3	R	Master receive IRQ. Set after receiving data. Cleared automatically by reading the I2C0MRX register.
2	R	Master transmit IRQ. Set at the end of a transmission. Cleared automatically by writing to the I2C0MTX register.
1	R	Master transmit FIFO underflow. Set automatically if the master transmit FIFO is underflowing. Cleared automatically by writing to the I2C0MTX register.
0	R	Master TX FIFO not full. Set automatically if the slave transmit FIFO is not full. Cleared automati-cally by writing twice to the I2C0STX register.

Table 128. I2CxSSTA Registers

Name	Address	Default Value	Access
I2C0SSTA	0xFFFF0804	0x01	R
I2C1SSTA	0xFFFF0904	0x01	R

I2CxSSTA are status registers for the slave channel.

Table 137. I2CxALT Registers

Name	Address	Default Value	Access
I2C0ALT	0xFFFF0828	0x00	R/W
I2C1ALT	0xFFFF0928	0x00	R/W

I2CxALT are hardware general call ID registers used in slave mode.

Table 138. I2CxCFG Registers

Name	Address	Default Value	Access
I2C0CFG	0xFFFF082C	0x00	R/W
I2C1CFG	0xFFFF092C	0x00	R/W

I2CxCFG are configuration registers.

Table 139. I2C0CFG MMR Bit Descriptions

Bit	Description
31:5	Reserved. These bits should be written by the user as 0.
14	Enable stop interrupt. Set by the user to generate an interrupt upon receiving a stop condition and after receiving a valid start condition and matching address. Cleared by the user to disable the generation of an interrupt upon receiving a stop condition.
13	Reserved.
12	Reserved.
11	Enable stretch SCL (holds SCL low). Set by the user to stretch the SCL line. Cleared by the user to disable stretching of the SCL line.
10	Reserved.
9	Slave Tx FIFO request interrupt enable. Set by the user to disable the slave Tx FIFO request interrupt. Cleared by the user to generate an interrupt request just after the negative edge of the clock for the R/W bit. This allows the user to input data into the slave Tx FIFO if it is empty. At 400 ksps and the core clock running at 41.78 MHz, the user has 45 clock cycles to take appropriate action, taking interrupt latency into account.
8	General call status bit clear. Set by the user to clear the general call status bits. Cleared automatically by hardware after the general call status bits are cleared.
7	Master serial clock enable bit. Set by user to enable generation of the serial clock in master mode. Cleared by user to disable serial clock in master mode.
6	Loopback enable bit. Set by user to internally connect the transition to the reception to test user software. Cleared by user to operate in normal mode.
5	Start backoff disable bit. Set by user in multimaster mode. If losing arbitration, the master immediately tries to retransmit. Cleared by user to enable start backoff. After losing arbitration, the master waits before trying to retransmit.
4	Hardware general call enable. When this bit and Bit 3 are set and have received a general call (Address 0x00) and a data byte, the device checks the contents of I2C0ALT against the receive register. If the contents match, the device has received a hardware general call. This is used if a device needs urgent attention from a master device without knowing which master it needs to turn to. This is a "to whom it may concern" call. The ADuC7019/20/21/22/24/25/26/27/28/29 watch for these addresses. The device that requires attention embeds its own address into the message. All masters listen, and the one that can handle the device contacts its slave and acts appropriately. The LSB of the I2C0ALT register should always be written to 1, as indicated in *The I²C-Bus Specification*, January 2000, from NXP.
3	General call enable bit. This bit is set by the user to enable the slave device to acknowledge (ACK) an I²C general call, Address 0x00 (write). The device then recognizes a data bit. If it receives a 0x06 (reset and write programmable part of slave address by hardware) as the data byte, the I²C interface resets as as indicated in *The I²C-Bus Specification,* January 2000, from NXP. This command can be used to reset an entire I²C system. The general call interrupt status bit sets on any general call. The user must take corrective action by setting up the I²C interface after a reset. If it receives a 0x04 (write programmable part of slave address by hardware) as the data byte, the general call interrupt status bit sets on any general call. The user must take corrective action by reprogramming the device address.
2	Reserved.
1	Master enable bit. Set by user to enable the master I²C channel. Cleared by user to disable the master I²C channel.
0	Slave enable bit. Set by user to enable the slave I²C channel. A slave transfer sequence is monitored for the device address in I2C0ID0, I2C0ID1, I2C0ID2, and I2C0ID3. At 400 kSPs, the core clock should run at 41.78 MHz because the interrupt latency could be up to 45 clock cycles alone. After the I²C read bit, the user has 0.5 of an I²C clock cycle to load the Tx FIFO. AT 400 kSPS, this is 1.26 µs, the interrupt latency.

In normal mode, an IRQ is generated each time the value of the counter reaches zero when counting down. It is also generated each time the counter value reaches full scale when counting up. An IRQ can be cleared by writing any value to clear the register of that particular timer (TxCLRI).

When using an asynchronous clock-to-clock timer, the interrupt in the timer block may take more time to clear than the time it takes for the code in the interrupt routine to execute. Ensure that the interrupt signal is cleared before leaving the interrupt service routine. This can be done by checking the IRQSTA MMR.

Hour:Minute:Second:1/128 Format

To use the timer in hour:minute:second:hundredths format, select the 32,768 kHz clock and prescaler of 256. The hundredths field does not represent milliseconds but 1/128 of a second (256/32,768). The bits representing the hour, minute, and second are not consecutive in the register. This arrangement applies to TxLD and TxVAL when using the hour:minute:second:hundredths format as set in TxCON[5:4]. See Table 171 for additional details.

Table 171. Hour:Minnute:Second:Hundredths Format

Bit	Value	Description
31:24	0 to 23 or 0 to 255	Hours
23:22	0	Reserved
21:16	0 to 59	Minutes
15:14	0	Reserved
13.8	0 to 59	Seconds
7	0	Reserved
6:0	0 to 127	1/128 second

Timer0 (RTOS Timer)

Timer0 is a general-purpose, 16-bit timer (count down) with a programmable prescaler (see Figure 77). The prescaler source is the core clock frequency (HCLK) and can be scaled by factors of 1, 16, or 256.

Timer0 can be used to start ADC conversions as shown in the block diagram in Figure 77.

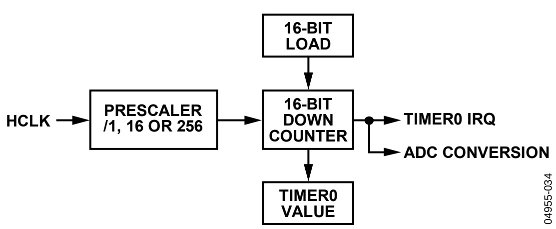

Figure 77. Timer0 Block Diagram

The Timer0 interface consists of four MMRs: T0LD, T0VAL, T0CON, and T0CLRI.

Table 172. T0LD Register

Name	Address	Default Value	Access
T0LD	0xFFFF0300	0x0000	R/W

T0LD is a 16-bit load register.

Table 173. T0VAL Register

Name	Address	Default Value	Access
T0VAL	0xFFFF0304	0xFFFF	R

T0VAL is a 16-bit read-only register representing the current state of the counter.

Table 174. T0CON Register

Name	Address	Default Value	Access
T0CON	0xFFFF0308	0x0000	R/W

T0CON is the configuration MMR described in Table 175.

Table 175. T0CON MMR Bit Descriptions

Bit	Value	Description
15:8		Reserved.
7		Timer0 enable bit. Set by user to enable Timer0. Cleared by user to disable Timer0 by default.
6		Timer0 mode. Set by user to operate in periodic mode. Cleared by user to operate in free-running mode. Default mode.
5:4		Reserved.
3:2		Prescale.
	00	Core Clock/1. Default value.
	01	Core Clock/16.
	10	Core Clock/256.
	11	Undefined. Equivalent to 00.
1:0		Reserved.

Table 176. T0CLRI Register

Name	Address	Default Value	Access
T0CLRI	0xFFFF030C	0xFF	W

T0CLRI is an 8-bit register. Writing any value to this register clears the interrupt.

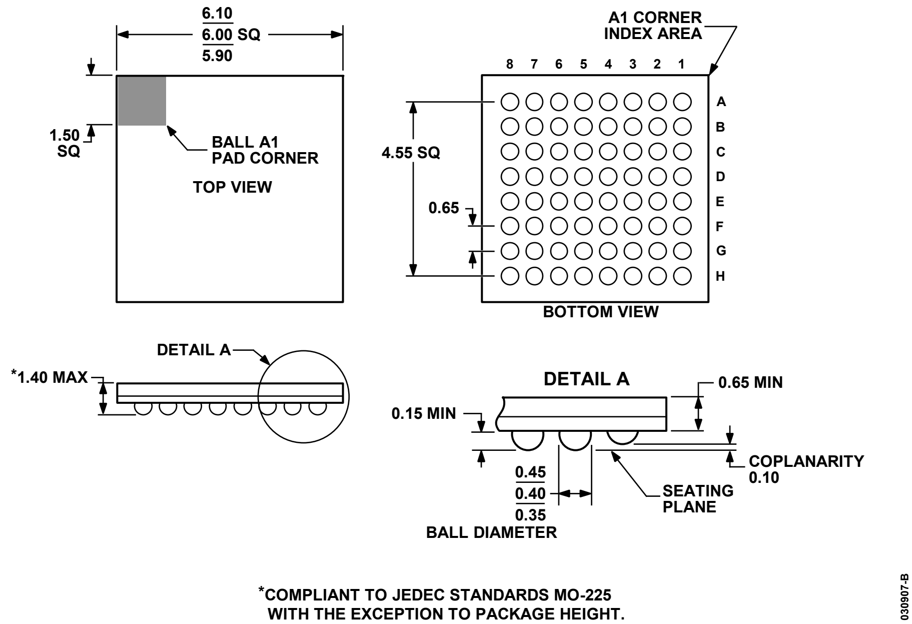

Figure 100. 64-Ball Chip Scale Package Ball Grid Array [CSP_BGA] (BC-64-4) Dimensions shown in millimeters

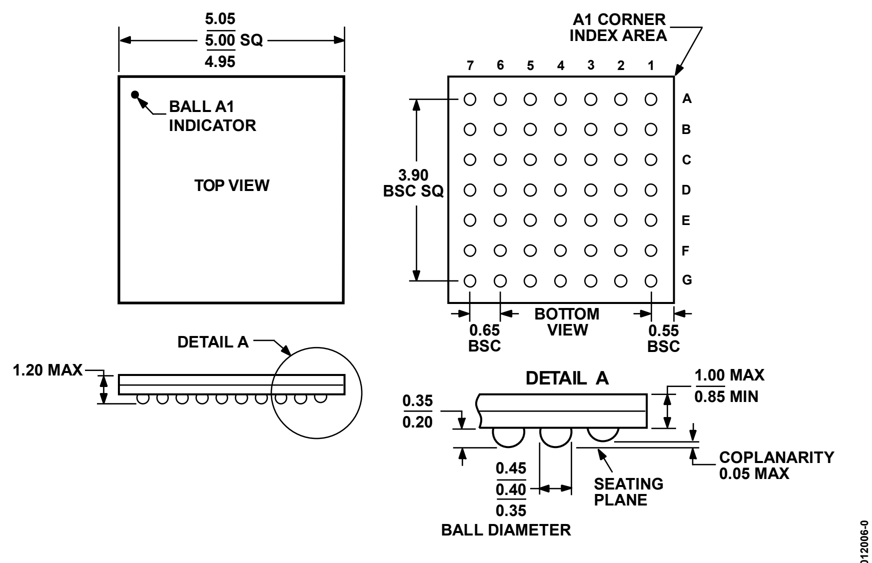

Figure 101. 49-Ball Chip Scale Package Ball Grid Array [CSP_BGA] (BC-49-1) Dimensions shown in millimeters

Model[1, 2]	ADC Channels[3]	DAC Channels	FLASH/ RAM	GPIO	Down-loader	Temperature Range	Package Description	Package Option	Ordering Quantity
EVAL-ADuC7020MKZ							ADuC7020 MiniKit		
EVAL-ADuC7020QSZ							ADuC7020 QuickStart Development System		
EVAL-ADuC7020QSPZ							ADuC7020 QuickStart Development System		
EVAL-ADuC7024QSZ							ADuC7024 QuickStart Development System		
EVAL-ADuC7026QSZ							ADuC7026 QuickStar Development System		
EVAL-ADuC7026QSPZ							ADuC7026 QuickStart Plus Development System		
EVAL-ADuC7028QSZ							ADuC7028 QuickStart Development System		
EVAL-ADUC7029QSZ							ADuC7029 QuickStart Development System		

[1] Z = RoHS Compliant Part.
[2] Models ADuC7026 and ADuC7027 include an external memory interface.
[3] One of the ADC channels is internally buffered for ADuC7019 models.

I^2C refers to a communications protocol originally developed by Phillips Semiconductors (now NXP Semiconductors).

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