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### What is "[Embedded - Microcontrollers](#)"?

"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Active
Core Processor	e200z6
Core Size	32-Bit Single-Core
Speed	132MHz
Connectivity	CANbus, EBI/EMI, Ethernet, SCI, SPI
Peripherals	DMA, POR, PWM, WDT
Number of I/O	256
Program Memory Size	3MB (3M x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	128K x 8
Voltage - Supply (Vcc/Vdd)	1.35V ~ 1.65V
Data Converters	A/D 40x12b
Oscillator Type	External
Operating Temperature	-40°C ~ 105°C (TA)
Mounting Type	Surface Mount
Package / Case	416-BBGA
Supplier Device Package	416-PBGA (27x27)
Purchase URL	<a href="https://www.e-xfl.com/pro/item?MUrl=&amp;PartUrl=spc5566mvr132">https://www.e-xfl.com/pro/item?MUrl=&amp;PartUrl=spc5566mvr132</a>

### 3.2.1 General Notes for Specifications at Maximum Junction Temperature

An estimation of the device junction temperature,  $T_J$ , can be obtained from the equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

$T_A$  = ambient temperature for the package ( $^{\circ}\text{C}$ )

$R_{\theta JA}$  = junction to ambient thermal resistance ( $^{\circ}\text{C}/\text{W}$ )

$P_D$  = power dissipation in the package (W)

The thermal resistance values used are based on the JEDEC JESD51 series of standards to provide consistent values for estimations and comparisons. The difference between the values determined for the single-layer (1s) board compared to a four-layer board that has two signal layers, a power and a ground plane (2s2p), demonstrate that the effective thermal resistance is not a constant. The thermal resistance depends on the:

- Construction of the application board (number of planes)
- Effective size of the board which cools the component
- Quality of the thermal and electrical connections to the planes
- Power dissipated by adjacent components

Connect all the ground and power balls to the respective planes with one via per ball. Using fewer vias to connect the package to the planes reduces the thermal performance. Thinner planes also reduce the thermal performance. When the clearance between the vias leave the planes virtually disconnected, the thermal performance is also greatly reduced.

As a general rule, the value obtained on a single-layer board is within the normal range for the tightly packed printed circuit board. The value obtained on a board with the internal planes is usually within the normal range if the application board has:

- One oz. (35 micron nominal thickness) internal planes
- Components are well separated
- Overall power dissipation on the board is less than  $0.02 \text{ W}/\text{cm}^2$

The thermal performance of any component depends on the power dissipation of the surrounding components. In addition, the ambient temperature varies widely within the application. For many natural convection and especially closed box applications, the board temperature at the perimeter (edge) of the package is approximately the same as the local air temperature near the device. Specifying the local ambient conditions explicitly as the board temperature provides a more precise description of the local ambient conditions that determine the temperature of the device.

At a known board temperature, the junction temperature is estimated using the following equation:

$$T_J = T_B + (R_{\theta JB} \times P_D)$$

where:

$T_J$  = junction temperature ( $^{\circ}\text{C}$ )

$T_B$  = board temperature at the package perimeter ( $^{\circ}\text{C/W}$ )

$R_{\theta JB}$  = junction-to-board thermal resistance ( $^{\circ}\text{C/W}$ ) per JESD51-8

$P_D$  = power dissipation in the package (W)

When the heat loss from the package case to the air does not factor into the calculation, an acceptable value for the junction temperature is predictable. Ensure the application board is similar to the thermal test condition, with the component soldered to a board with internal planes.

The thermal resistance is expressed as the sum of a junction-to-case thermal resistance plus a case-to-ambient thermal resistance:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$$

where:

$R_{\theta JA}$  = junction-to-ambient thermal resistance ( $^{\circ}\text{C/W}$ )

$R_{\theta JC}$  = junction-to-case thermal resistance ( $^{\circ}\text{C/W}$ )

$R_{\theta CA}$  = case-to-ambient thermal resistance ( $^{\circ}\text{C/W}$ )

$R_{\theta JC}$  is device related and is not affected by other factors. The thermal environment can be controlled to change the case-to-ambient thermal resistance,  $R_{\theta CA}$ . For example, change the air flow around the device, add a heat sink, change the mounting arrangement on the printed circuit board, or change the thermal dissipation on the printed circuit board surrounding the device. This description is most useful for packages with heat sinks where 90% of the heat flow is through the case to heat sink to ambient. For most packages, a better model is required.

A more accurate two-resistor thermal model can be constructed from the junction-to-board thermal resistance and the junction-to-case thermal resistance. The junction-to-case thermal resistance describes when using a heat sink or where a substantial amount of heat is dissipated from the top of the package. The junction-to-board thermal resistance describes the thermal performance when most of the heat is conducted to the printed circuit board. This model can be used to generate simple estimations and for computational fluid dynamics (CFD) thermal models.

To determine the junction temperature of the device in the application on a prototype board, use the thermal characterization parameter ( $\Psi_{JT}$ ) to determine the junction temperature by measuring the temperature at the top center of the package case using the following equation:

$$T_J = T_T + (\Psi_{JT} \times P_D)$$

where:

$T_T$  = thermocouple temperature on top of the package ( $^{\circ}\text{C}$ )

$\Psi_{JT}$  = thermal characterization parameter ( $^{\circ}\text{C/W}$ )

$P_D$  = power dissipation in the package (W)

**Table 6.  $V_{RC}$  and POR Electrical Specifications (continued)**

Spec	Characteristic	Symbol	Min.	Max.	Units
9	Absolute value of slew rate on power supply pins	—	—	50	V/ms
10	Required gain at Tj: $I_{DD} \div I_{VRCCTL}$ (@ $f_{sys} = f_{MAX}$ ) <sup>6, 7, 8, 9</sup>	— BETA <sup>10</sup>	60 65 85	— — 500	— — —
	– 40° C 25° C 150° C				

<sup>1</sup> The internal POR signals are  $V_{POR15}$ ,  $V_{POR33}$ , and  $V_{POR5}$ . On power up, assert  $\overline{RESET}$  before the internal POR negates.  $\overline{RESET}$  must remain asserted until the power supplies are within the operating conditions as specified in [Table 9](#) DC Electrical Specifications. On power down, assert  $\overline{RESET}$  before any power supplies fall outside the operating conditions and until the internal POR asserts.

<sup>2</sup>  $V_{IL\_S}$  ([Table 9](#), Spec15) is guaranteed to scale with  $V_{DDEH6}$  down to  $V_{POR5}$ .

<sup>3</sup> Supply full operating current for the 1.5 V supply when the 3.3 V supply reaches this range.

<sup>4</sup> It is possible to reach the current limit during ramp up—do not treat this event as short circuit current.

<sup>5</sup> At peak current for device.

<sup>6</sup> Requires compliance with Freescale's recommended board requirements and transistor recommendations. Board signal traces/routing from the  $V_{RCCTL}$  package signal to the base of the external pass transistor and between the emitter of the pass transistor to the  $V_{DD}$  package signals must have a maximum of 100 nH inductance and minimal resistance (less than 1  $\Omega$ ).  $V_{RCCTL}$  must have a nominal 1  $\mu F$  phase compensation capacitor to ground.  $V_{DD}$  must have a 20  $\mu F$  (nominal) bulk capacitor (greater than 4  $\mu F$  over all conditions, including lifetime). Place high-frequency bypass capacitors consisting of eight 0.01  $\mu F$ , two 0.1  $\mu F$ , and one 1  $\mu F$  capacitors around the package on the  $V_{DD}$  supply signals.

<sup>7</sup>  $I_{VRCCTL}$  is measured at the following conditions:  $V_{DD} = 1.35$  V,  $V_{RC33} = 3.1$  V,  $V_{VRCCTL} = 2.2$  V.

<sup>8</sup> Refer to [Table 1](#) for the maximum operating frequency.

<sup>9</sup> Values are based on  $I_{DD}$  from high-use applications as explained in the  $I_{DD}$  Electrical Specification.

<sup>10</sup> BETA represents the worst-case external transistor. It is measured on a per-part basis and calculated as  $(I_{DD} \div I_{VRCCTL})$ .

### 3.7 Power-Up/Down Sequencing

Power sequencing between the 1.5 V power supply and  $V_{DDSYN}$  or the  $\overline{RESET}$  power supplies is required if using an external 1.5 V power supply with  $V_{RC33}$  tied to ground (GND). To avoid power-sequencing,  $V_{RC33}$  must be powered up within the specified operating range, even if the on-chip voltage regulator controller is not used. Refer to [Section 3.7.2, “Power-Up Sequence \(VRC33 Grounded\),”](#) and [Section 3.7.3, “Power-Down Sequence \(VRC33 Grounded\).”](#)

Power sequencing requires that  $V_{DD33}$  must reach a certain voltage where the values are read as ones before the POR signal negates. Refer to [Section 3.7.1, “Input Value of Pins During POR Dependent on VDD33.”](#)

Although power sequencing is not required between  $V_{RC33}$  and  $V_{DDSYN}$  during power up,  $V_{RC33}$  must not lead  $V_{DDSYN}$  by more than 600 mV or lag by more than 100 mV for the  $V_{RC}$  stage turn-on to operate within specification. Higher spikes in the emitter current of the pass transistor occur if  $V_{RC33}$  leads or lags  $V_{DDSYN}$  by more than these amounts. The value of that higher spike in current depends on the board power supply circuitry and the amount of board level capacitance.

Furthermore, when all of the PORs negate, the system clock starts to toggle, adding another large increase of the current consumed by  $V_{RC33}$ . If  $V_{RC33}$  lags  $V_{DDSYN}$  by more than 100 mV, the increase in current consumed can drop  $V_{DD}$  low enough to assert the 1.5 V POR again. Oscillations are possible when the

# Electrical Characteristics

During initial power ramp-up, when  $V_{stby}$  is 0.6v or above, a typical current of 1-3mA and maximum of 4mA may be seen until  $V_{DD}$  is applied. This current will not reoccur until  $V_{stby}$  is lowered below  $V_{stby}$  min. specification.

Figure 2 shows an approximate interpolation of the  $I_{STBY}$  worst-case specification to estimate values at different voltages and temperatures. The vertical lines shown at 25 °C, 60 °C, and 150 °C in Figure 2 are the actual  $I_{DD\_STBY}$  specifications (27d) listed in Table 9.

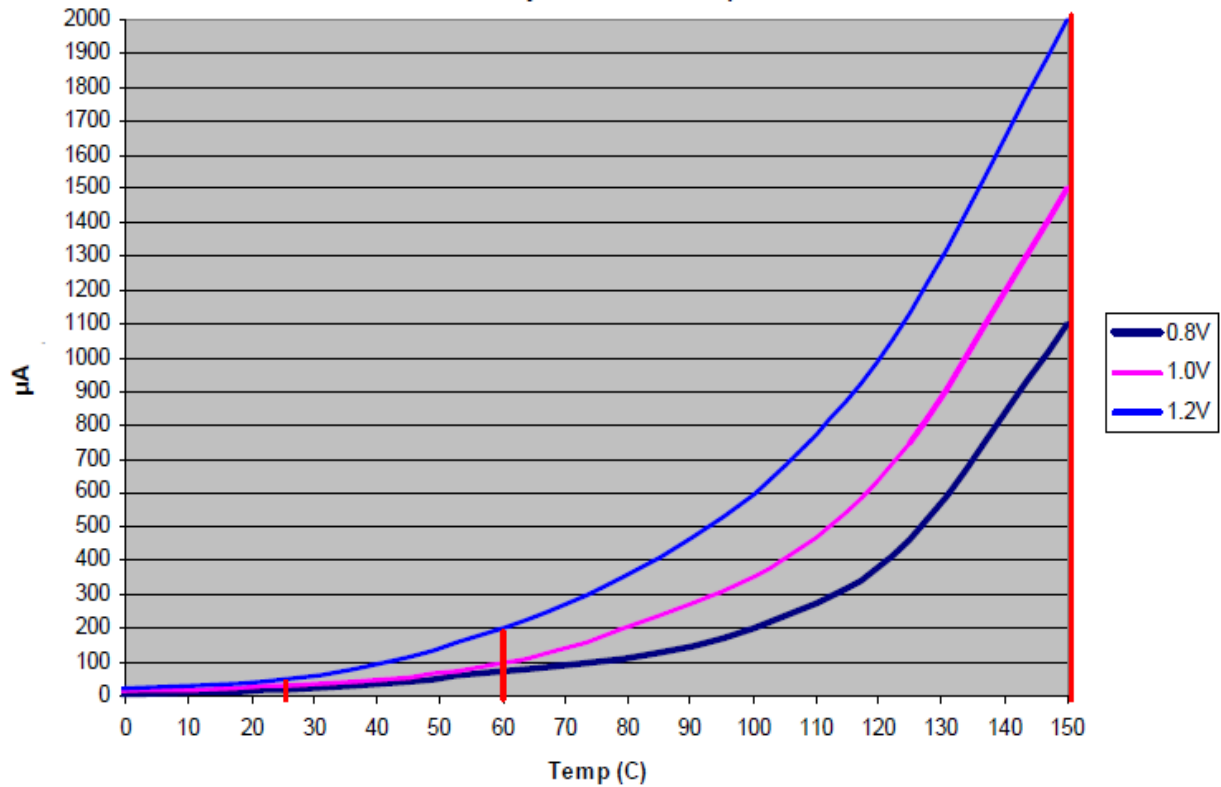


Figure 2.  $I_{STBY}$  Worst-case Specifications

Table 9. DC Electrical Specifications ( $T_A = T_L$  to  $T_H$ ) (continued)

Spec	Characteristic	Symbol	Min	Max.	Unit
41	$V_{SSSYN}$ to $V_{SS}$ differential voltage	$V_{SSSYN} - V_{SS}$	-50	50	mV
42	$V_{RCVSS}$ to $V_{SS}$ differential voltage	$V_{RCVSS} - V_{SS}$	-50	50	mV
43	$V_{DDF}$ to $V_{DD}$ differential voltage	$V_{DDF} - V_{DD}$	-100	100	mV
43a	$V_{RC33}$ to $V_{DDSYN}$ differential voltage	$V_{RC33} - V_{DDSYN}$	-0.1	0.1 <sup>19</sup>	V
44	Analog input differential signal range (with common mode 2.5 V)	$V_{IDIFF}$	-2.5	2.5	V
45	Operating temperature range, ambient (packaged)	$T_A = (T_L \text{ to } T_H)$	$T_L$	$T_H$	°C
46	Slew rate on power-supply pins	—	—	50	V/ms

<sup>1</sup>  $V_{DDE2}$  and  $V_{DDE3}$  are limited to 2.25–3.6 V only if  $SIU\_ECCR[EBTS] = 0$ ;  $V_{DDE2}$  and  $V_{DDE3}$  have a range of 1.6–3.6 V if  $SIU\_ECCR[EBTS] = 1$ .

<sup>2</sup>  $|V_{DDA0} - V_{DDA1}|$  must be  $< 0.1$  V.

<sup>3</sup>  $V_{PP}$  can drop to 3.0 V during read operations.

<sup>4</sup> If standby operation is not required, connect  $V_{STBY}$  to ground.

<sup>5</sup> Applies to CLKOUT, external bus pins, and Nexus pins.

<sup>6</sup> Maximum average RMS DC current.

<sup>7</sup> Eight-way cache enabled ( $L1CSR0[CORG] = 0b0$ ).

<sup>8</sup> Average current measured on automotive benchmark.

<sup>9</sup> Peak currents can be higher on specialized code.

<sup>10</sup> High use current measured while running optimized SPE assembly code with all code and data 100% locked in cache (0% miss rate) with all channels of the eMIOS and eTPU running autonomously, plus the eDMA transferring data continuously from SRAM to SRAM. Higher currents are possible if an 'idle' loop that crosses cache lines is run from cache. Write code to avoid this condition.

<sup>11</sup> Four-way cache enabled ( $L1CSR0[CORG] = 0b1$ ) or ( $L1CSR0[CORG] = 0b0$  with  $L1CSR0[WAM] = 0b1$ ,  $L1CSR0[WD] = 0b1111$ ,  $L1CSR0[WDD] = 0b1111$ ,  $L1CSR0[AWID] = 0b1$ , and  $L1CSR0[AWDD] = 0b1$ ).

<sup>12</sup> The current specification relates to average standby operation after SRAM has been loaded with data. For power up current see [Section 3.7, "Power-Up/Down Sequencing"](#), [Figure 2](#).

<sup>13</sup> Power requirements for the  $V_{DD33}$  supply depend on the frequency of operation, load of all I/O pins, and the voltages on the I/O segments. Refer to [Table 11](#) for values to calculate the power dissipation for a specific operation.

<sup>14</sup> Power requirements for each I/O segment are dependent on the frequency of operation and load of the I/O pins on a particular I/O segment, and the voltage of the I/O segment. Refer to [Table 10](#) for values to calculate power dissipation for specific operation. The total power consumption of an I/O segment is the sum of the individual power consumptions for each pin on the segment.

<sup>15</sup> Absolute value of current, measured at  $V_{IL}$  and  $V_{IH}$ .

<sup>16</sup> Weak pullup/down inactive. Measured at  $V_{DDE} = 3.6$  V and  $V_{DDEH} = 5.25$  V. Applies to pad types: pad\_fc, pad\_sh, and pad\_mh.

<sup>17</sup> Maximum leakage occurs at maximum operating temperature. Leakage current decreases by approximately one-half for each 8 °C to 12 °C, in the ambient temperature range of 50 °C to 125 °C. Applies to pad types: pad\_a and pad\_ae.

<sup>18</sup>  $V_{SSA}$  refers to both  $V_{SSA0}$  and  $V_{SSA1}$ .  $|V_{SSA0} - V_{SSA1}|$  must be  $< 0.1$  V.

<sup>19</sup> Up to 0.6 V during power up and power down.

### 3.8.2 I/O Pad $V_{DD33}$ Current Specifications

The power consumption of the  $V_{DD33}$  supply depends on the usage of the pins on all I/O segments. The power consumption is the sum of all input and output pin  $V_{DD33}$  currents for all I/O segments. The output pin  $V_{DD33}$  current can be calculated from Table 11 based on the voltage, frequency, and load on all fast (pad\_fc) pins. The input pin  $V_{DD33}$  current can be calculated from Table 11 based on the voltage, frequency, and load on all pad\_sh and pad\_mh pins. Use linear scaling to calculate pin currents for voltage, frequency, and load parameters that fall outside the values given in Table 11.

**Table 11.  $V_{DD33}$  Pad Average DC Current ( $T_A = T_L$  to  $T_H$ )<sup>1</sup>**

Spec	Pad Type	Symbol	Frequency (MHz)	Load <sup>2</sup> (pF)	$V_{DD33}$ (V)	$V_{DDE}$ (V)	Drive Select	Current (mA)
<b>Inputs</b>								
1	Slow	$I_{33\_SH}$	66	0.5	3.6	5.5	NA	0.003
2	Medium	$I_{33\_MH}$	66	0.5	3.6	5.5	NA	0.003
<b>Outputs</b>								
3	Fast	$I_{33\_FC}$	66	10	3.6	3.6	00	0.35
4			66	20	3.6	3.6	01	0.53
5			66	30	3.6	3.6	10	0.62
6			66	50	3.6	3.6	11	0.79
7			66	10	3.6	1.98	00	0.35
8			66	20	3.6	1.98	01	0.44
9			66	30	3.6	1.98	10	0.53
10			66	50	3.6	1.98	11	0.70
11			56	10	3.6	3.6	00	0.30
12			56	20	3.6	3.6	01	0.45
13			56	30	3.6	3.6	10	0.52
14			56	50	3.6	3.6	11	0.67
15			56	10	3.6	1.98	00	0.30
16			56	20	3.6	1.98	01	0.37
17			56	30	3.6	1.98	10	0.45
18			56	50	3.6	1.98	11	0.60
19			40	10	3.6	3.6	00	0.21
20			40	20	3.6	3.6	01	0.31
21			40	30	3.6	3.6	10	0.37
22			40	50	3.6	3.6	11	0.48
23			40	10	3.6	1.98	00	0.21
24			40	20	3.6	1.98	01	0.27
25			40	30	3.6	1.98	10	0.32
26			40	50	3.6	1.98	11	0.42

<sup>1</sup> These values are estimated from simulation and not tested. Currents apply to output pins for the fast pads only and to input pins for the slow and medium pads only.

<sup>2</sup> All loads are lumped.

### 3.9 Oscillator and FMPLL Electrical Characteristics

**Table 12. FMPLL Electrical Specifications**
 $(V_{DDSYN} = 3.0\text{--}3.6\text{ V}; V_{SS} = V_{SSSYN} = 0.0\text{ V}; T_A = T_L \text{ to } T_H)$ 

Spec	Characteristic	Symbol	Minimum	Maximum	Unit
1	PLL reference frequency range: <sup>1</sup> Crystal reference External reference Dual controller (1:1 mode)	$f_{ref\_crystal}$ $f_{ref\_ext}$ $f_{ref\_1:1}$	8 8 24	20 20 $f_{sys} \div 2$	MHz
2	System frequency <sup>2</sup>	$f_{sys}$	$f_{ICO(MIN)} \div 2^{RFD}$	$f_{MAX}$ <sup>3</sup>	MHz
3	System clock period	$t_{CYC}$	—	$1 \div f_{sys}$	ns
4	Loss of reference frequency <sup>4</sup>	$f_{LOR}$	100	1000	kHz
5	Self-clocked mode (SCM) frequency <sup>5</sup>	$f_{SCM}$	7.4	17.5	MHz
6	EXTAL input high voltage crystal mode <sup>6</sup>	$V_{IHEXT}$	$V_{XTAL} + 0.4\text{ V}$	—	V
	All other modes [dual controller (1:1), bypass, external reference]	$V_{IHEXT}$	$(V_{DDE5} \div 2) + 0.4\text{ V}$	—	V
7	EXTAL input low voltage crystal mode <sup>7</sup>	$V_{ILEXT}$	—	$V_{XTAL} - 0.4\text{ V}$	V
	All other modes [dual controller (1:1), bypass, external reference]	$V_{ILEXT}$	—	$(V_{DDE5} \div 2) - 0.4\text{ V}$	V
8	XTAL current <sup>8</sup>	$I_{XTAL}$	2	6	mA
9	Total on-chip stray capacitance on XTAL	$C_{S\_XTAL}$	—	1.5	pF
10	Total on-chip stray capacitance on EXTAL	$C_{S\_EXTAL}$	—	1.5	pF
11	Crystal manufacturer's recommended capacitive load	$C_L$	Refer to crystal specification	Refer to crystal specification	pF
12	Discrete load capacitance to connect to EXTAL	$C_{L\_EXTAL}$	—	$(2 \times C_L) - C_{S\_EXTAL} - C_{PCB\_EXTAL}$ <sup>9</sup>	pF
13	Discrete load capacitance to connect to XTAL	$C_{L\_XTAL}$	—	$(2 \times C_L) - C_{S\_XTAL} - C_{PCB\_XTAL}$ <sup>9</sup>	pF
14	PLL lock time <sup>10</sup>	$t_{pll}$	—	750	μs
15	Dual controller (1:1) clock skew (between CLKOUT and EXTAL) <sup>11, 12</sup>	$t_{skew}$	−2	2	ns
16	Duty cycle of reference	$t_{DC}$	40	60	%
17	Frequency unLOCK range	$f_{UL}$	−4.0	4.0	% $f_{sys}$
18	Frequency LOCK range	$f_{LCK}$	−2.0	2.0	% $f_{sys}$



**Table 12. FMPLL Electrical Specifications (continued)**
 $(V_{DDSYN} = 3.0\text{--}3.6\text{ V}; V_{SS} = V_{SSSYN} = 0.0\text{ V}; T_A = T_L \text{ to } T_H)$ 

Spec	Characteristic	Symbol	Minimum	Maximum	Unit
19	CLKOUT period jitter, measured at $f_{SYS}$ max: <sup>13, 14</sup> Peak-to-peak jitter (clock edge to clock edge) Long term jitter (averaged over a 2 ms interval)	$C_{JITTER}$	— —	5.0 0.01	% $f_{CLKOUT}$
20	Frequency modulation range limit <sup>15</sup> (do not exceed $f_{SYS}$ maximum)	$C_{MOD}$	0.8	2.4	% $f_{SYS}$
21	ICO frequency $f_{ICO} = [f_{ref\_crystal} \times (MFD + 4)] \div (PREDIV + 1)$ <sup>16</sup> $f_{ICO} = [f_{ref\_ext} \times (MFD + 4)] \div (PREDIV + 1)$	$f_{ICO}$	48	$f_{MAX}$	MHz
22	Predivider output frequency (to PLL)	$f_{PREDIV}$	4	20 <sup>17</sup>	MHz

<sup>1</sup> Nominal crystal and external reference values are worst-case not more than 1%. The device operates correctly if the frequency remains within  $\pm 5\%$  of the specification limit. This tolerance range allows for a slight frequency drift of the crystals over time. The designer must thoroughly understand the drift margin of the source clock.

<sup>2</sup> All internal registers retain data at 0 Hz.

<sup>3</sup> Up to the maximum frequency rating of the device (refer to [Table 1](#)).

<sup>4</sup> Loss of reference frequency is defined as the reference frequency detected internally, which transitions the PLL into self-clocked mode.

<sup>5</sup> The PLL operates at self-clocked mode (SCM) frequency when the reference frequency falls below  $f_{LOR}$ . SCM frequency is measured on the CLKOUT ball with the divider set to divide-by-two of the system clock.  
NOTE: In SCM, the MFD and PREDIV have no effect and the RFD is bypassed.

<sup>6</sup> Use the EXTAL input high voltage parameter when using the FlexCAN oscillator in crystal mode (no quartz crystals or resonators).  $(V_{extal} - V_{xtal})$  must be  $\geq 400\text{ mV}$  for the oscillator's comparator to produce the output clock.

<sup>7</sup> Use the EXTAL input low voltage parameter when using the FlexCAN oscillator in crystal mode (no quartz crystals or resonators).  $(V_{xtal} - V_{extal})$  must be  $\geq 400\text{ mV}$  for the oscillator's comparator to produce the output clock.

<sup>8</sup>  $I_{xtal}$  is the oscillator bias current out of the XTAL pin with both EXTAL and XTAL pins grounded.

<sup>9</sup>  $C_{PCB\_EXTAL}$  and  $C_{PCB\_XTAL}$  are the measured PCB stray capacitances on EXTAL and XTAL, respectively.

<sup>10</sup> This specification applies to the period required for the PLL to relock after changing the MFD frequency control bits in the synthesizer control register (SYNCR). From power up with crystal oscillator reference, the lock time also includes the crystal startup time.

<sup>11</sup> PLL is operating in 1:1 PLL mode.

<sup>12</sup>  $V_{DDE} = 3.0\text{--}3.6\text{ V}$ .

<sup>13</sup> Jitter is the average deviation from the programmed frequency measured over the specified interval at maximum  $f_{SYS}$ . Measurements are made with the device powered by filtered supplies and clocked by a stable external clock signal. Noise injected into the PLL circuitry via  $V_{DDSYN}$  and  $V_{SSSYN}$  and variation in crystal oscillator frequency increase the jitter percentage for a given interval. CLKOUT divider is set to divide-by-two.

<sup>14</sup> Values are with frequency modulation disabled. If frequency modulation is enabled, jitter is the sum of (jitter + Cmod).

<sup>15</sup> Modulation depth selected must not result in  $f_{SYS}$  value greater than the  $f_{SYS}$  maximum specified value.

<sup>16</sup>  $f_{SYS} = f_{ICO} \div (2^{RFD})$ .

<sup>17</sup> Maximum value for dual controller (1:1) mode is  $(f_{MAX} \div 2)$  with the predivider set to 1 (FMPLL\_SYNCR[PREDIV] = 0b001).

### 3.10 eQADC Electrical Characteristics

Table 13. eQADC Conversion Specifications ( $T_A = T_L$  to  $T_H$ )

Spec	Characteristic	Symbol	Minimum	Maximum	Unit
1	ADC clock (ADCLK) frequency <sup>1</sup>	$F_{ADCLK}$	1	12	MHz
2	Conversion cycles Differential Single ended	CC	13 + 2 (15) 14 + 2 (16)	13 + 128 (141) 14 + 128 (142)	ADCLK cycles
3	Stop mode recovery time <sup>2</sup>	$T_{SR}$	10	—	$\mu s$
4	Resolution <sup>3</sup>	—	1.25	—	mV
5	INL: 6 MHz ADC clock	INL6	−4	4	Counts <sup>3</sup>
6	INL: 12 MHz ADC clock	INL12	−8	8	Counts
7	DNL: 6 MHz ADC clock	DNL6	−3 <sup>4</sup>	3 <sup>4</sup>	Counts
8	DNL: 12 MHz ADC clock	DNL12	−6 <sup>4</sup>	6 <sup>4</sup>	Counts
9	Offset error with calibration	OFFWC	−4 <sup>5</sup>	4 <sup>5</sup>	Counts
10	Full-scale gain error with calibration	GAINWC	−8 <sup>6</sup>	8 <sup>6</sup>	Counts
11	Disruptive input injection current <sup>7, 8, 9, 10</sup>	$I_{INJ}$	−1	1	mA
12	Incremental error due to injection current. All channels are 10 k $\Omega$ < $R_s$ < 100 k $\Omega$ Channel under test has $R_s = 10$ k $\Omega$ , $I_{INJ} = I_{INJMAX}, I_{INJMIN}$	$E_{INJ}$	−4	4	Counts
13	Total unadjusted error (TUE) for single ended conversions with calibration <sup>11, 12, 13, 14, 15</sup>	TUE	−4	4	Counts

<sup>1</sup> Conversion characteristics vary with  $F_{ADCLK}$  rate. Reduced conversion accuracy occurs at maximum  $F_{ADCLK}$  rate. The maximum value is based on 800 KS/s and the minimum value is based on 20 MHz oscillator clock frequency divided by a maximum 16 factor.

<sup>2</sup> Stop mode recovery time begins when the ADC control register enable bits are set until the ADC is ready to perform conversions.

<sup>3</sup> At  $V_{RH} - V_{RL} = 5.12$  V, one least significant bit (LSB) = 1.25, mV = one count.

<sup>4</sup> Guaranteed 10-bit mono tonicity.

<sup>5</sup> The absolute value of the offset error without calibration  $\leq 100$  counts.

<sup>6</sup> The absolute value of the full scale gain error without calibration  $\leq 120$  counts.

<sup>7</sup> Below disruptive current conditions, the channel being stressed has conversion values of: 0x3FF for analog inputs greater than  $V_{RH}$ , and 0x000 for values less than  $V_{RL}$ . This assumes that  $V_{RH} \leq V_{DDA}$  and  $V_{RL} \geq V_{SSA}$  due to the presence of the sample amplifier. Other channels are not affected by non-disruptive conditions.

<sup>8</sup> Exceeding the limit can cause a conversion error on both stressed and unstressed channels. Transitions within the limit do not affect device reliability or cause permanent damage.

<sup>9</sup> Input must be current limited to the value specified. To determine the value of the required current-limiting resistor, calculate resistance values using  $V_{POSCLAMP} = V_{DDA} + 0.5$  V and  $V_{NEGCLAMP} = -0.3$  V, then use the larger of the calculated values.

<sup>10</sup> This condition applies to two adjacent pads on the internal pad.

<sup>11</sup> The TUE specification is always less than the sum of the INL, DNL, offset, and gain errors due to canceling errors.

<sup>12</sup> TUE does not apply to differential conversions.

<sup>13</sup> Measured at 6 MHz ADC clock. TUE with a 12 MHz ADC clock is: −16 counts < TUE < 16 counts.

<sup>14</sup> TUE includes all internal device errors such as internal reference variation (75% Ref, 25% Ref).

<sup>15</sup> Depending on the input impedance, the analog input leakage current (Table 9. DC Electrical Specifications, spec 35a) can affect the actual TUE measured on analog channels AN[12], AN[13], AN[14], AN[15].

## 3.11 H7Fa Flash Memory Electrical Characteristics

Table 14. Flash Program and Erase Specifications ( $T_A = T_L$  to  $T_H$ )

Spec	Flash Program Characteristic	Symbol	Min.	Typical <sup>1</sup>	Initial Max. <sup>2</sup>	Max. <sup>3</sup>	Unit
3	Doubleword (64 bits) program time <sup>4</sup>	$T_{dwprogram}$	—	10	—	500	$\mu s$
4	Page program time <sup>4</sup>	$T_{pprogram}$	—	22	44 <sup>5</sup>	500	$\mu s$
7	16 KB block pre-program and erase time	$T_{16kpperase}$	—	265	400	5000	ms
9	48 KB block pre-program and erase time	$T_{48kpperase}$	—	345	400	5000	ms
10	64 KB block pre-program and erase time	$T_{64kpperase}$	—	415	500	5000	ms
8	128 KB block pre-program and erase time	$T_{128kpperase}$	—	500	1250	7500	ms
11	Minimum operating frequency for program and erase operations <sup>6</sup>	—	25	—	—	—	MHz

<sup>1</sup> Typical program and erase times are calculated at 25 °C operating temperature using nominal supply values.

<sup>2</sup> Initial factory condition:  $\leq 100$  program/erase cycles, 25 °C, using a typical supply voltage measured at a minimum system frequency of 80 MHz.

<sup>3</sup> The maximum erase time occurs after the specified number of program/erase cycles. This maximum value is characterized but not guaranteed.

<sup>4</sup> Actual hardware programming times. This does not include software overhead.

<sup>5</sup> Page size is 256 bits (8 words).

<sup>6</sup> The read frequency of the flash can range up to the maximum operating frequency. There is no minimum read frequency condition.

Table 15. Flash EEPROM Module Life ( $T_A = T_L$  to  $T_H$ )

Spec	Characteristic	Symbol	Min.	Typical <sup>1</sup>	Unit
1a	Number of program/erase cycles per block for 16 KB, 48 KB, and 64 KB blocks over the operating temperature range ( $T_J$ )	P/E	100,000	—	cycles
1b	Number of program/erase cycles per block for 128 KB blocks over the operating temperature range ( $T_J$ )	P/E	1000	100,000	cycles
2	Data retention Blocks with 0–1,000 P/E cycles Blocks with 1,001–100,000 P/E cycles	Retention	20 5	— —	years

<sup>1</sup> Typical endurance is evaluated at 25° C. Product qualification is performed to the minimum specification. For additional information on the Freescale definition of typical endurance, refer to engineering bulletin EB619 Typical Endurance for Nonvolatile Memory.

Table 16 shows the FLASH\_BIU settings versus frequency of operation. Refer to the device reference manual for definitions of these bit fields.

**Table 16. FLASH\_BIU Settings vs. Frequency of Operation <sup>1</sup>**

Maximum Frequency (MHz)	APC	RWSC	WWSC	DPFEN <sup>2</sup>	IPFEN <sup>2</sup>	PFLIM <sup>3</sup>	BFEN <sup>4</sup>
Up to and including 82 MHz <sup>5</sup>	0b001	0b001	0b01	0b00 0b01 0b11	0b00 0b01 0b11	0b000 to 0b110	0b0 0b1
Up to and including 102 MHz <sup>6</sup>	0b001	0b010	0b01	0b00 0b01 0b11	0b00 0b01 0b11	0b000 to 0b110	0b0 0b1
Up to and including 135 MHz <sup>7</sup>	0b010	0b011	0b01	0b00 0b01 0b11	0b00 0b01 0b11	0b000 to 0b110	0b0 0b1
Up to and including 147 MHz <sup>8</sup>	0b011	0b100	0b01	0b00 0b01 0b11	0b00 0b01 0b11	0b000 to 0b110	0b0 0b1
Default setting after reset	0b111	0b111	0b11	0b00	0b00	0b000	0b0

<sup>1</sup> Illegal combinations exist. Use entries from the same row in this table.

<sup>2</sup> For maximum flash performance, set to 0b11.

<sup>3</sup> For maximum flash performance, set to 0b110.

<sup>4</sup> For maximum flash performance, set to 0b1.

<sup>5</sup> 82 MHz parts allow for 80 MHz system clock + 2% frequency modulation (FM).

<sup>6</sup> 102 MHz parts allow for 100 MHz system clock + 2% FM.

<sup>7</sup> 135 MHz parts allow for 132 MHz system clock + 2% FM.

<sup>8</sup> 147 MHz parts allow for 144 MHz system clock + 2% FM.

## 3.12 AC Specifications

### 3.12.1 Pad AC Specifications

**Table 17. Pad AC Specifications ( $V_{DDEH} = 5.0\text{ V}$ ,  $V_{DDE} = 1.8\text{ V}$ ) <sup>1</sup>**

Spec	Pad	SRC / DSC (binary)	Out Delay <sup>2, 3, 4</sup> (ns)	Rise / Fall <sup>4, 5</sup> (ns)	Load Drive (pF)
1	Slow high voltage (SH)	11	26	15	50
			82	60	200
		01	75	40	50
			137	80	200
		00	377	200	50
			476	260	200

**Table 17. Pad AC Specifications ( $V_{DDEH} = 5.0\text{ V}$ ,  $V_{DDE} = 1.8\text{ V}$ )<sup>1</sup> (continued)**

Spec	Pad	SRC / DSC (binary)	Out Delay <sup>2, 3, 4</sup> (ns)	Rise / Fall <sup>4, 5</sup> (ns)	Load Drive (pF)
2	Medium high voltage (MH)	11	16	8	50
			43	30	200
		01	34	15	50
			61	35	200
		00	192	100	50
			239	125	200
3	Fast	00	3.1	2.7	10
		01		2.5	20
		10		2.4	30
		11		2.3	50
4	Pullup/down (3.6 V max)	—	—	7500	50
5	Pullup/down (5.5 V max)	—	—	9000	50

<sup>1</sup> These are worst-case values that are estimated from simulation (not tested). The values in the table are simulated at:  $V_{DD} = 1.35\text{--}1.65\text{ V}$ ;  $V_{DDE} = 1.62\text{--}1.98\text{ V}$ ;  $V_{DDEH} = 4.5\text{--}5.25\text{ V}$ ;  $V_{DD33}$  and  $V_{DDSYN} = 3.0\text{--}3.6\text{ V}$ ; and  $T_A = T_L$  to  $T_H$ .

<sup>2</sup> This parameter is supplied for reference and is guaranteed by design (not tested).

<sup>3</sup> The output delay is shown in Figure 4. To calculate the output delay with respect to the system clock, add a maximum of one system clock to the output delay.

<sup>4</sup> The output delay and rise and fall are measured to 20% or 80% of the respective signal.

<sup>5</sup> This parameter is guaranteed by characterization rather than 100% tested.

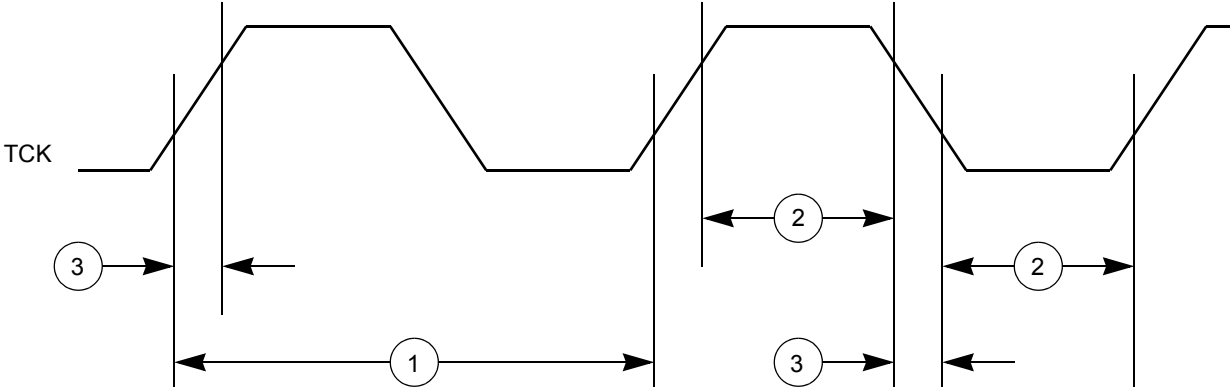
**Table 18. Derated Pad AC Specifications ( $V_{DDEH} = 3.3\text{ V}$ ,  $V_{DDE} = 3.3\text{ V}$ )<sup>1</sup>**

Spec	Pad	SRC/DSC (binary)	Out Delay <sup>2, 3, 4</sup> (ns)	Rise / Fall <sup>3, 5</sup> (ns)	Load Drive (pF)
1	Slow high voltage (SH)	11	39	23	50
			120	87	200
		01	101	52	50
			188	111	200
		00	507	248	50
			597	312	200
2	Medium high voltage (MH)	11	23	12	50
			64	44	200
		01	50	22	50
			90	50	200
		00	261	123	50
			305	156	200

**Table 20. JTAG Pin AC Electrical Characteristics <sup>1</sup> (continued)**

Spec	Characteristic	Symbol	Min.	Max.	Unit
12	TCK falling-edge to output valid out of high impedance	$t_{\text{BSDVZ}}$	—	50	ns
13	TCK falling-edge to output high impedance (Hi-Z)	$t_{\text{BSDHZ}}$	—	50	ns
14	Boundary scan input valid to TCK rising-edge	$t_{\text{BSDST}}$	50	—	ns
15	TCK rising-edge to boundary scan input invalid	$t_{\text{BSDHT}}$	50	—	ns

<sup>1</sup> These specifications apply to JTAG boundary scan only. JTAG timing specified at:  $V_{\text{DDE}} = 3.0\text{--}3.6\text{ V}$  and  $T_{\text{A}} = T_{\text{L}}$  to  $T_{\text{H}}$ . Refer to [Table 21](#) for Nexus specifications.



**Figure 6. JTAG Test Clock Input Timing**

### 3.13.4 External Bus Interface (EBI) Timing

Table 22 lists the timing information for the external bus interface (EBI).

Table 22. Bus Operation Timing <sup>1</sup>

Spec	Characteristic and Description	Symbol	External Bus Frequency <sup>2, 3</sup>								Unit	Notes
			40 MHz		56 MHz		67 MHz		72 MHz			
			Min	Max	Min	Max	Min	Max	Min	Max		
1	CLKOUT period	T <sub>C</sub>	25.0	—	17.9	—	15.2	—	13.3	—	ns	Signals are measured at 50% V <sub>DDE</sub> .
2	CLKOUT duty cycle	t <sub>CDC</sub>	45%	55%	45%	55%	45%	55%	45%	55%	T <sub>C</sub>	
3	CLKOUT rise time	t <sub>CRT</sub>	—	— <sup>4</sup>	—	— <sup>4</sup>	—	— <sup>4</sup>	—	— <sup>4</sup>	ns	
4	CLKOUT fall time	t <sub>CFT</sub>	—	— <sup>4</sup>	—	— <sup>4</sup>	—	— <sup>4</sup>	—	— <sup>4</sup>	ns	
5	CLKOUT positive edge to output signal <i>invalid</i> or Hi-Z (hold time)	t <sub>COH</sub>	1.0 <sup>6</sup> 1.5	—	1.0 <sup>6</sup> 1.5	—	1.0 <sup>6</sup> 1.5	—	1.0 <sup>6</sup> 1.5	—	ns	EBTS = 0  EBTS = 1  Hold time selectable via SIU_ECCR [EBTS] bit.
	External bus interface CS[0:3] ADDR[8:31] DATA[0:31] BDIP BG <sup>5</sup> BR <sup>7</sup> BB OE RD_WR TA TEA TS TSIZ[0:1] WE/BE[0:3]											
	CLKOUT positive edge to output signal <i>invalid</i> or Hi-Z (hold time)	t <sub>CCOH</sub>	1.0 <sup>6</sup> 1.5	—	1.0 <sup>6</sup> 1.5	—	1.0 <sup>6</sup> 1.5	—	1.0 <sup>6</sup> 1.5	—	ns	EBTS = 0  EBTS = 1  Hold time selectable via SIU_ECCR [EBTS] bit.
	Calibration bus interface CAL_CS[0:3] CAL_ADDR[9:30] CAL_DATA[0:15] CAL_OE CAL_RD_WR CAL_TS CAL_WE/BE[0:1]											

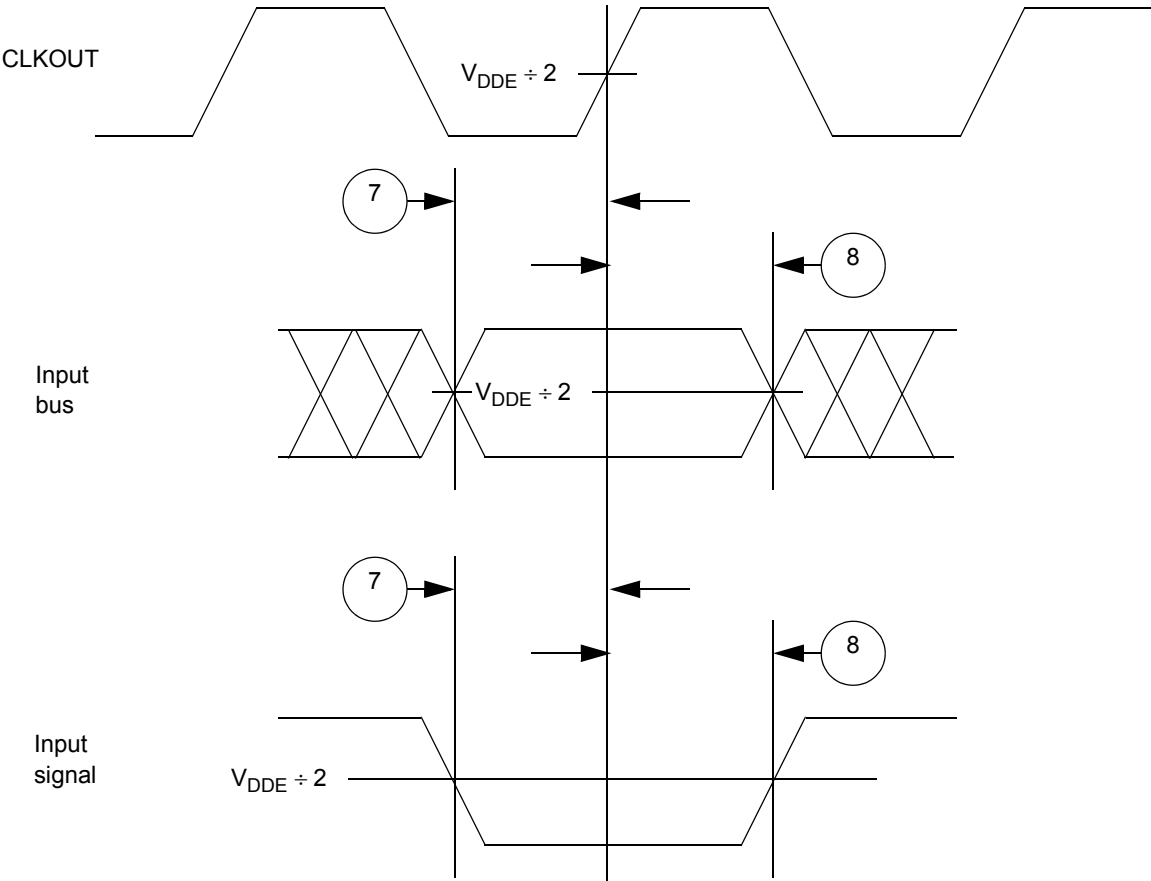


Figure 14. Synchronous Input Timing

### 3.13.5 External Interrupt Timing (IRQ Signals)

Table 23. External Interrupt Timing <sup>1</sup>

Spec	Characteristic	Symbol	Min.	Max.	Unit
1	IRQ pulse-width low	$t_{IPWL}$	3	—	$t_{CYC}$
2	IRQ pulse-width high	$T_{IPWH}$	3	—	$t_{CYC}$
3	IRQ edge-to-edge time <sup>2</sup>	$t_{ICYC}$	6	—	$t_{CYC}$

<sup>1</sup> IRQ timing specified at:  $V_{DDEH} = 3.0\text{--}5.25\text{ V}$  and  $T_A = T_L$  to  $T_H$ .

<sup>2</sup> Applies when IRQ signals are configured for rising-edge or falling-edge events, but not both.



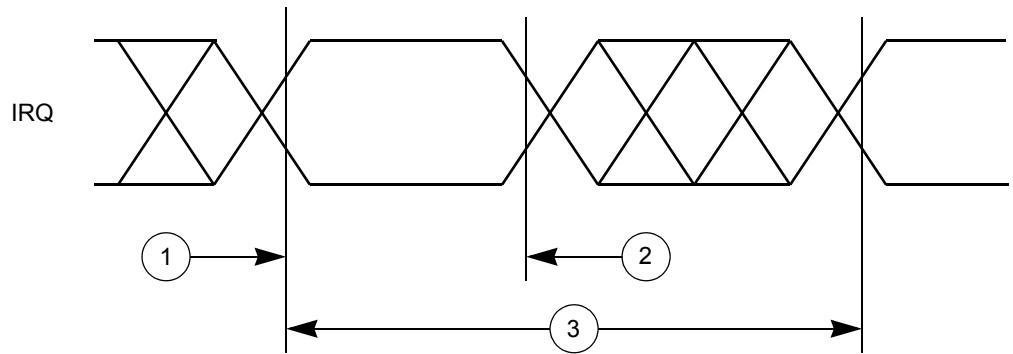


Figure 15. External Interrupt Timing

### 3.13.6 eTPU Timing

Table 24. eTPU Timing <sup>1</sup>

Spec	Characteristic	Symbol	Min.	Max	Unit
1	eTPU input channel pulse width	$t_{ICPW}$	4	—	$t_{CYC}$
2	eTPU output channel pulse width	$t_{OCPW}$	2 <sup>2</sup>	—	$t_{CYC}$

<sup>1</sup> eTPU timing specified at:  $V_{DDEH} = 3.0\text{--}5.25\text{ V}$  and  $T_A = T_L$  to  $T_H$ .  
<sup>2</sup> This specification does not include the rise and fall times. When calculating the minimum eTPU pulse width, include the rise and fall times defined in the slew rate control fields (SRC) of the pad configuration registers (PCR).

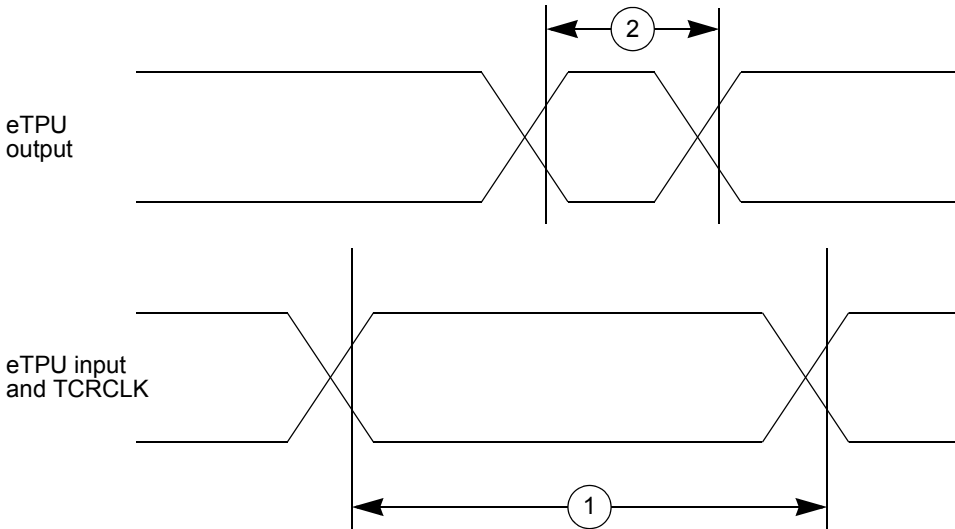


Figure 16. eTPU Timing

# 4 Mechanicals

## 4.1 MPC5566 416 PBGA Pinout

Figure 32, Figure 33, and Figure 34 show the pinout for the MPC5566 416 PBGA package. The alternate Fast Ethernet Controller (FEC) signals are multiplexed with the data calibration bus signals.

### NOTE

The MPC5500 devices are pin compatible for software portability and use the primary function names to label the pins in the BGA diagram. Although some devices do not support all the primary functions shown in the BGA diagram, the muxed and GPIO signals on those pins remain available. See the signals chapter in the device reference manual for the signal muxing.

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23	24	25	26	
A	VSS	VSTBY	AN37	AN11	VDDA1	AN16	AN1	AN5	VRH	AN23	AN27	AN28	AN35	VSSA0	AN15	ETRIG 1	ETPUB 18	ETPUB 20	ETPUB 24	ETPUB 27	GPIO 205	MDO11	MDO8	VDD	VDD33	VSS	A
B	VDD	VSS	AN36	AN39	AN19	AN20	AN0	AN4	REF BYPC	AN22	AN26	AN31	AN32	VSSA0	AN14	ETRIG 0	ETPUB 21	ETPUB 25	ETPUB 28	ETPUB 31	MDO10	MDO7	MDO4	MDO0	VSS	VDD7	B
C	VDD33	VDD	VSS	AN8	AN17	VSSA1	AN21	AN3	AN7	VRL	AN25	AN30	AN33	VDDA0	AN13	ETPUB 19	ETPUB 22	ETPUB 26	ETPUB 30	MDO9	MDO6	MDO3	MDO1	VSS	VDD7	VDD	C
D	ETPUA 30	ETPUA 31	VDD	VSS	AN38	AN9	AN10	AN18	AN2	AN6	AN24	AN29	AN34	VDD7	AN12	ETPUB 16	ETPUB 17	ETPUB 23	ETPUB 29	MDO5	MDO2	VDD7	VSS	VDD7	TCK	TDI	D
E	ETPUA 28	ETPUA 29	VDD7	VSS																			VDD7	TMS	TDO	TEST	E
F	ETPUA 24	ETPUA 27	VDD7	VSS																			MSE00	JCOMP	EVTI	EVTO	F
G	ETPUA 23	ETPUA 22	ETPUA 25	ETPUA 21																			MSE01	MCKO	GPIO 204	ETPUB 15	G
H	ETPUA 20	ETPUA 19	ETPUA 18	ETPUA 17																			RDY	GPIO 203	ETPUB 14	ETPUB 13	H
J	ETPUA 16	ETPUA 15	ETPUA 14	ETPUA 13																			VDD7	ETPUB 12	ETPUB 11	ETPUB 9	J
K	ETPUA 12	ETPUA 11	ETPUA 10	ETPUA 9																			ETPUB 10	ETPUB 8	ETPUB 7	ETPUB 5	K
L	ETPUA 8	ETPUA 7	ETPUA 6	ETPUA 5																			ETPUB 6	ETPUB 4	ETPUB 3	ETPUB 2	L
M	ETPUA 4	ETPUA 3	ETPUA 2	ETPUA 1																			TCRCLK B	ETPUB 1	ETPUB 0	SINB	M
N	BDIP	TEA	ETPUA 0	TCRCLK A																			SOUTB	PCSB3	PCSB0	PCSB1	N
P	CS3	CS2	CS1	CS0																			PCSA3	PCSB4	SCKB	PCSB2	P
R	WE3	WE2	WE1	WE0																			PCSB5	SOUTA	SINA	SCKA	R
T	VDD7	TSIZ0	RD_WR	VDD7																			PCSA1	PCSA0	PCSA2	VPP	T
U	ADDR 16	TSIZ1	TA	VDD33																			PCSA4	TXDA	PCSA5	VFLASH	U
V	ADDR 18	ADDR 17	TS	ADDR 8																			CNTXC	RXDA	RSTOUT	RST CFG	V
W	ADDR 20	ADDR 19	ADDR 9	ADDR 10																			RXDB	CNRXC	TXDB	RESET	W
Y	ADDR 22	ADDR 21	ADDR 11	VDD7																			WKP CFG	BOOT CFG1	VRC VSS	VSS SYN	Y
AA	ADDR 24	ADDR 23	ADDR 13	ADDR 12																			VDD7	PLL CFG1	BOOT CFG0	EXTAL	AA
AB	VDD7	ADDR 25	ADDR 15	ADDR 14																			VDD	VRC CTL	PLL CFG0	XTAL	AB
AC	ADDR 26	ADDR 27	ADDR 31	VSS	VDD	DATA 26	DATA 28	VDD7	DATA 30	DATA 31	DATA 8	DATA 10	VDD7	DATA 12	DATA 14	EMIOS 2	EMIOS 8	EMIOS 12	EMIOS 21	VDD7	VDD5	NC	VSS	VDD	VRC33	VDD SYN	AC
AD	ADDR 28	ADDR 30	VSS	VDD	DATA 24	DATA 25	DATA 27	DATA 29	VDD33	GPIO 207	DATA 9	DATA 11	DATA 13	DATA 15	EMIOS 3	EMIOS 6	EMIOS 10	EMIOS 15	EMIOS 17	EMIOS 22	CNTXA	VDD5	NC	VSS	VDD	VDD33	AD
AE	ADDR 29	VSS	VDD	DATA 17	DATA 19	DATA 21	DATA 23	DATA 0	DATA 2	DATA 4	DATA 6	OE	BR	BG	EMIOS 1	EMIOS 5	EMIOS 9	EMIOS 13	EMIOS 16	EMIOS 19	EMIOS 23	CNRXA	VDD5	CLKOUT	VSS	VDD	AE
AF	VSS	VDD	DATA 16	DATA 18	VDD7	DATA 20	DATA 22	GPIO 206	DATA 1	DATA 3	VDD7	DATA 5	DATA 7	BB	EMIOS 0	EMIOS 4	EMIOS 7	EMIOS 11	EMIOS 14	EMIOS 18	EMIOS 20	CNTXB	CNRXB	VDD5	ENG CLK	VSS	AF

Note: NC No connect. AC22 & AD23 reserved

Figure 32. MPC5566 416 Package

	1	2	3	4	5	6	7	8	9	10	11	12	13
A	VSS	VSTBY	AN37	AN11	VDDA1	AN16	AN1	AN5	VRH	AN23	AN27	AN28	AN35
B	VDD	VSS	AN36	AN39	AN19	AN20	AN0	AN4	REF BYPC	AN22	AN26	AN31	AN32
C	VDD33	VDD	VSS	AN8	AN17	VSSA1	AN21	AN3	AN7	VRL	AN25	AN30	AN33
D	ETPUA 30	ETPUA 31	VDD	VSS	AN38	AN9	AN10	AN18	AN2	AN6	AN24	AN29	AN34
E	ETPUA 28	ETPUA 29	VDDEH 1	VDD									
F	ETPUA 24	ETPUA 27	ETPUA 26	VDDEH 1									
G	ETPUA 23	ETPUA 22	ETPUA 25	ETPUA 21									
H	ETPUA 20	ETPUA 19	ETPUA 18	ETPUA 17									
J	ETPUA 16	ETPUA 15	ETPUA 14	ETPUA 13									
K	ETPUA 12	ETPUA 11	ETPUA 10	ETPUA 9						VSS	VSS	VSS	VSS
L	ETPUA 8	ETPUA 7	ETPUA 6	ETPUA 5						VSS	VSS	VSS	VSS
M	ETPUA 4	ETPUA 3	ETPUA 2	ETPUA 1						VDDE2	VDDE2	VSS	VSS
N	BDIP	TEA	ETPUA 0	TCRCLK A						VDDE2	VDDE2	VSS	VSS
P	CS3	CS2	CS1	CS0						VDDE2	VDDE2	VSS	VSS
R	WE3	WE2	WE1	WE0						VDDE2	VDDE2	VSS	VSS
T	VDDE2	TSIZ0	RD_WR	VDDE2						VDDE2	VSS	VDDE2	VDDE2
U	ADDR 16	TSIZ1	TA	VDD33						VSS	VDDE2	VDDE2	VDDE2
V	ADDR 18	ADDR 17	TS	ADDR 8									
W	ADDR 20	ADDR 19	ADDR 9	ADDR 10									
Y	ADDR 22	ADDR 21	ADDR 11	VDDE2									
AA	ADDR 24	ADDR 23	ADDR 13	ADDR 12									
AB	VDDE2	ADDR 25	ADDR 15	ADDR 14									
AC	ADDR 26	ADDR 27	ADDR 31	VSS	VDD	DATA 26	DATA 28	VDDE2	DATA 30	DATA 31	DATA 8	DATA 10	VDDE2
AD	ADDR 28	ADDR 30	VSS	VDD	DATA 24	DATA 25	DATA 27	DATA 29	VDD33	GPIO 207	DATA 9	DATA 11	DATA 13
AE	ADDR 29	VSS	VDD	DATA 17	DATA 19	DATA 21	DATA 23	DATA 0	DATA 2	DATA 4	DATA 6	OE	BR
AF	VSS	VDD	DATA 16	DATA 18	VDDE2	DATA 20	DATA 22	GPIO 206	DATA 1	DATA 3	VDDE2	DATA 5	DATA 7
	1	2	3	4	5	6	7	8	9	10	11	12	13

**Figure 33. MPC5566 416 Package Left Side (view 1 of 2)**

# 5 Revision History for the MPC5566 Data Sheet

The history of revisions made to this data sheet are listed and described in this section. The information that has changed from a previous revision of this document to the current revision is listed for each revision and are grouped in the following categories:

- Global and text changes
- Table and figure changes

Within each category, the information that has changed is listed in sequential order.

## 5.1 Information Changed Between Revisions 2.0 and 3.0

The following table lists the information that changed in the tables between Rev. 2.0 and 3.0. Click the links to see the change.

**Table 32. Changes Between Rev. 2.0 and 3.0**

Location	Description of Changes
<a href="#">Section 3.7, “Power-Up/Down Sequencing”</a>	Added the following paragraph in <a href="#">Section 3.7, “Power-Up/Down Sequencing”</a> “During initial power ramp-up, when Vstby is 0.6v or above. a typical current of 1-3mA and maximum of 4mA may be seen until VDD is applied. This current will not reoccur until Vstby is lowered below Vstby min. specification”.
	Moved <a href="#">Figure 2 (fISTBY Worst-case Specifications)</a> to <a href="#">Section 3.7, “Power-Up/Down Sequencing”</a> .
<a href="#">Section 3.8, “DC Electrical Specifications</a>	In <a href="#">Table 9 (DC Electrical Specifications (T<sub>A</sub> = T<sub>L</sub> to T<sub>H</sub>))</a> for Spec 27d the Characteristic “Refer to Figure 3 for an interpolation of this data” changed to “RAM standby current”.
	Changed the footnote attached to IDD_STBY to “The current specification relates to average standby operation after SRAM has been loaded with data. For power up current see <a href="#">Section 3.7, “Power-Up/Down Sequencing”</a> , <a href="#">Figure 2 (fISTBY Worst-case Specifications)</a> ).
	Removed the footnote “Figure 3 shows an illustration of the IDD_STBY values interpolated for these temperature values”.

## 5.2 Information Changed Between Revisions 1.0 and 2.0

The following table lists the information that changed in the tables between Rev. 1.0 and 2.0. Click the links to see the change.

**Table 35. Table and Figure Changes Between Rev. 0.0 and Rev. 1.0 (continued)**

Location	Description of Changes
Table 27, EQADC SSI Timing Characteristics:	<ul style="list-style-type: none"> <li>Deleted from table title '(Pads at 3.3 V or 5.0 V)'</li> <li>Deleted 1st line in table 'CLOAD = 25 pF on all outputs. Pad drive strength set to maximum.'</li> <li>Spec 1: FCK frequency -- removed.</li> <li>Combined footnotes 1 and 2, and moved the new footnote to Spec 2. Moved old footnote 3 that is now footnote 2 to Spec 2.</li> <li>Footnote 1, deleted '<math>V_{DD} = 1.35\text{--}1.65\text{ V}</math>' and '<math>V_{DD33}</math> and <math>V_{DDSYN} = 3.0\text{--}3.6\text{ V}</math>.'</li> <li>Changed 'CL = 50 pF' to 'CL = 25 pF.'</li> <li>Footnote 2: added 'cycle' after 'duty' to read: FCK duty cycle is not 50% when . . . .</li> </ul>
Figure 35, MPC5566 416 Package: Deleted the version number and date.	