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Applications of "<u>Embedded -</u> <u>Microcontrollers</u>"

### Details

Product Status	Active
Core Processor	e200z6
Core Size	32-Bit Single-Core
Speed	144MHz
Connectivity	CANbus, EBI/EMI, Ethernet, SCI, SPI
Peripherals	DMA, POR, PWM, WDT
Number of I/O	256
Program Memory Size	3MB (3M × 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	128K x 8
Voltage - Supply (Vcc/Vdd)	1.35V ~ 1.65V
Data Converters	A/D 40x12b
Oscillator Type	External
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	416-BBGA
Supplier Device Package	416-PBGA (27x27)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/spc5566mzp144r

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong



#### **Ordering Information** 2



Note: Not all options are available on all devices. Refer to Table 1.

Figure 1. MPC5500 Family Part Number Example

Unless noted in this data sheet, all specifications apply from  $T_{L}$  to  $T_{H}$ .

Table	1.	Orderable	Part	Numbers
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Freescale Part Number <sup>1</sup>	Package Description	Spee	Speed (MHz)		Operating Temperature <sup>2</sup>	
	i ackage bescription	Nominal	Max. <sup>3</sup> (f <sub>MAX</sub> )	Min. (T <sub>L</sub> )	Max. (T <sub>H</sub> )	
MPC5566MVR144		144	147		125° C	
MPC5566MVR132	MPC5566 416 package Lead-free (PbFree)	132	135	–40° C		
MPC5566MVR112		112	114			
MPC5566MVR80		80	82			
MPC5566MZP144		144	147			
MPC5566MZP132	MPC5566 416 package	132	135	40° C	125° C	
MPC5566MZP112	Leaded (SnPb)	112	114	-40 C	120 C	
MPC5566MZP80		80	82			

1 All devices are PPC5566, rather than MPC5566 or SPC5566, until product qualifications are complete. Not all configurations are available in the PPC parts.

2 The lowest ambient operating temperature is referenced by T<sub>L</sub>; the highest ambient operating temperature is referenced by T<sub>H</sub>.

3 Speed is the nominal maximum frequency. Max, speed is the maximum speed allowed including frequency modulation (FM). 82 MHz parts allow for 80 MHz system clock + 2% FM; 114 MHz parts allow for 112 MHz system clock + 2% FM; 135 MHz parts allow for 132 MHz system clock + 2% FM; and 147 MHz parts allow for 144 MHz system clock + 2% FM.



### 3.2.1 General Notes for Specifications at Maximum Junction Temperature

An estimation of the device junction temperature,  $T_{\mu}$ , can be obtained from the equation:

$$T_{J} = T_{A} + (R_{\theta JA} \times P_{D})$$

 $T_A$  = ambient temperature for the package (°C)

 $R_{\theta JA}$  = junction to ambient thermal resistance (°C/W)

 $P_D$  = power dissipation in the package (W)

The thermal resistance values used are based on the JEDEC JESD51 series of standards to provide consistent values for estimations and comparisons. The difference between the values determined for the single-layer (1s) board compared to a four-layer board that has two signal layers, a power and a ground plane (2s2p), demonstrate that the effective thermal resistance is not a constant. The thermal resistance depends on the:

- Construction of the application board (number of planes)
- Effective size of the board which cools the component
- Quality of the thermal and electrical connections to the planes
- Power dissipated by adjacent components

Connect all the ground and power balls to the respective planes with one via per ball. Using fewer vias to connect the package to the planes reduces the thermal performance. Thinner planes also reduce the thermal performance. When the clearance between the vias leave the planes virtually disconnected, the thermal performance is also greatly reduced.

As a general rule, the value obtained on a single-layer board is within the normal range for the tightly packed printed circuit board. The value obtained on a board with the internal planes is usually within the normal range if the application board has:

- One oz. (35 micron nominal thickness) internal planes
- Components are well separated
- Overall power dissipation on the board is less than  $0.02 \text{ W/cm}^2$

The thermal performance of any component depends on the power dissipation of the surrounding components. In addition, the ambient temperature varies widely within the application. For many natural convection and especially closed box applications, the board temperature at the perimeter (edge) of the package is approximately the same as the local air temperature near the device. Specifying the local ambient conditions explicitly as the board temperature provides a more precise description of the local ambient conditions that determine the temperature of the device.



At a known board temperature, the junction temperature is estimated using the following equation:

 $T_{J} = T_{B} + (R_{\theta JB} \times P_{D})$ 

where:

 $T_J =$ junction temperature (°C)

 $T_B$  = board temperature at the package perimeter (°C/W)

 $R_{\theta JB}$  = junction-to-board thermal resistance (°C/W) per JESD51-8

 $P_D$  = power dissipation in the package (W)

When the heat loss from the package case to the air does not factor into the calculation, an acceptable value for the junction temperature is predictable. Ensure the application board is similar to the thermal test condition, with the component soldered to a board with internal planes.

The thermal resistance is expressed as the sum of a junction-to-case thermal resistance plus a case-to-ambient thermal resistance:

 $R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$ 

where:

 $R_{\theta JA}$  = junction-to-ambient thermal resistance (°C/W)

 $R_{\theta JC}$  = junction-to-case thermal resistance (°C/W)

 $R_{\theta CA}$  = case-to-ambient thermal resistance (°C/W)

 $R_{\theta JC}$  is device related and is not affected by other factors. The thermal environment can be controlled to change the case-to-ambient thermal resistance,  $R_{\theta CA}$ . For example, change the air flow around the device, add a heat sink, change the mounting arrangement on the printed circuit board, or change the thermal dissipation on the printed circuit board surrounding the device. This description is most useful for packages with heat sinks where 90% of the heat flow is through the case to heat sink to ambient. For most packages, a better model is required.

A more accurate two-resistor thermal model can be constructed from the junction-to-board thermal resistance and the junction-to-case thermal resistance. The junction-to-case thermal resistance describes when using a heat sink or where a substantial amount of heat is dissipated from the top of the package. The junction-to-board thermal resistance describes the thermal performance when most of the heat is conducted to the printed circuit board. This model can be used to generate simple estimations and for computational fluid dynamics (CFD) thermal models.

To determine the junction temperature of the device in the application on a prototype board, use the thermal characterization parameter ( $\Psi_{JT}$ ) to determine the junction temperature by measuring the temperature at the top center of the package case using the following equation:

 $T_{J} = T_{T} + (\Psi_{JT} \times P_{D})$ where:  $T_{T} = \text{thermocouple temperature on top of the package (°C)}$  $\Psi_{JT} = \text{thermal characterization parameter (°C/W)}$  $P_{D} = \text{power dissipation in the package (W)}$ 



The thermal characterization parameter is measured in compliance with the JESD51-2 specification using a 40-gauge type T thermocouple epoxied to the top center of the package case. Position the thermocouple so that the thermocouple junction rests on the package. Place a small amount of epoxy on the thermocouple junction and approximately 1 mm of wire extending from the junction. Place the thermocouple wire flat against the package case to avoid measurement errors caused by the cooling effects of the thermocouple wire.

### **References:**

Semiconductor Equipment and Materials International 3081 Zanker Rd. San Jose, CA., 95134 (408) 943-6900

MIL-SPEC and EIA/JESD (JEDEC) specifications are available from Global Engineering Documents at 800-854-7179 or 303-397-7956.

JEDEC specifications are available on the web at http://www.jedec.org.

- 1. C.E. Triplett and B. Joiner, "An Experimental Characterization of a 272 PBGA Within an Automotive Engine Controller Module," Proceedings of SemiTherm, San Diego, 1998, pp. 47–54.
- 2. G. Kromann, S. Shidore, and S. Addison, "Thermal Modeling of a PBGA for Air-Cooled Applications," Electronic Packaging and Production, pp. 53–58, March 1998.
- 3. B. Joiner and V. Adams, "Measurement and Simulation of Junction to Board Thermal Resistance and Its Application in Thermal Modeling," Proceedings of SemiTherm, San Diego, 1999, pp. 212–220.

### 3.3 Package

The MPC5566 is available in packaged form. Read the package options in Section 2, "Ordering Information." Refer to Section 4, "Mechanicals," for pinouts and package drawings.

# 3.4 EMI (Electromagnetic Interference) Characteristics

Spec	Characteristic	Minimum	Typical	Maximum	Unit
1	Scan range	0.15	_	1000	MHz
2	Operating frequency	_	_	f <sub>MAX</sub>	MHz
3	V <sub>DD</sub> operating voltages	_	1.5	—	V
4	V <sub>DDSYN</sub> , V <sub>RC33</sub> , V <sub>DD33</sub> , V <sub>FLASH</sub> , V <sub>DDE</sub> operating voltages	_	3.3	—	V
5	V <sub>PP</sub> , V <sub>DDEH</sub> , V <sub>DDA</sub> operating voltages	_	5.0	—	V
6	Maximum amplitude	—	_	14 <sup>2</sup> 32 <sup>3</sup>	dBuV
7	Operating temperature	_	_	25	٥C

Table 4. EMI Testing Specifications <sup>1</sup>

<sup>1</sup> EMI testing and I/O port waveforms per SAE J1752/3 issued 1995-03. Qualification testing was performed on the MPC5554 and applied to the MPC5500 family as generic EMI performance data.

<sup>2</sup> Measured with the single-chip EMI program.

<sup>3</sup> Measured with the expanded EMI program.



# 3.5 ESD (Electromagnetic Static Discharge) Characteristics

Characteristic	Symbol	Value	Unit
ESD for human body model (HBM)		2000	V
	R1	1500	Ω
	С	100	pF
ESD for field induced oberge model (EDCM)		500 (all pins)	
		750 (corner pins)	V
Number of pulses per pin:			
Positive pulses (HBM)		1	—
Negative pulses (HBM)	—	1	_
Interval of pulses	_	1	second

Table 5. ESD Ratings <sup>1, 2</sup>

<sup>1</sup> All ESD testing conforms to CDF-AEC-Q100 Stress Test Qualification for Automotive Grade Integrated Circuits.

<sup>2</sup> Device failure is defined as: 'If after exposure to ESD pulses, the device does not meet the device specification requirements, which includes the complete DC parametric and functional testing at room temperature and hot temperature.

# 3.6 Voltage Regulator Controller (V<sub>RC</sub>) and Power-On Reset (POR) Electrical Specifications

The following table lists the  $V_{RC}$  and POR electrical specifications:

Spec	Charact	Characteristic S		Min.	Max.	Units
1	1.5 V (V <sub>DD</sub> ) POR <sup>1</sup>	Negated (ramp up) Asserted (ramp down)	V <sub>POR15</sub>	1.1 1.1	1.35 1.35	V
2	3.3 V (V <sub>DDSYN</sub> ) POR <sup>1</sup>	Asserted (ramp up) Negated (ramp up) Asserted (ramp down) Negated (ramp down)	V <sub>POR33</sub>	0.0 2.0 2.0 0.0	0.30 2.85 2.85 0.30	V
3	RESET pin supply (V <sub>DDEH6</sub> ) POR <sup>1, 2</sup>	Negated (ramp up) Asserted (ramp down)	V <sub>POR5</sub>	2.0 2.0	2.85 2.85	V
4		Before V <sub>RC</sub> allows the pass transistor to start turning on	V <sub>TRANS_START</sub>	1.0	2.0	V
5	V <sub>RC33</sub> voltage	When $V_{RC}$ allows the pass transistor to completely turn on <sup>3, 4</sup>	V <sub>TRANS_ON</sub>	2.0	2.85	V
6		When the voltage is greater than the voltage at which the $V_{RC}$ keeps the 1.5 V supply in regulation <sup>5, 6</sup>	V <sub>VRC33REG</sub>	3.0	_	V
	Current can be sourced	–40° C		11.0	_	mA
7	by V <sub>RCCTL</sub> at Tj:	25° C	I <sub>VRCCTL</sub> <sup>7</sup>	9.0	—	mA
		150° C		7.5	_	mA
8	Voltage differential during power up such that: $V_{DD33}$ can lag $V_{DDSYN}$ or $V_{DDEH6}$ before $V_{DDSYN}$ and $V_{DDEH6}$ reach the $V_{POR33}$ and $V_{POR5}$ minimums respectively.		V <sub>DD33_LAG</sub>	_	1.0	V

### Table 6. V<sub>RC</sub> and POR Electrical Specifications



1.5 V POR asserts and stops the system clock, causing the voltage on  $V_{DD}$  to rise until the 1.5 V POR negates again. All oscillations stop when  $V_{RC33}$  is powered sufficiently.

When powering down,  $V_{RC33}$  and  $V_{DDSYN}$  have no delta requirement to each other, because the bypass capacitors internal and external to the device are already charged. When not powering up or down, no delta between  $V_{RC33}$  and  $V_{DDSYN}$  is required for the  $V_{RC}$  to operate within specification.

There are no power up/down sequencing requirements to prevent issues such as latch-up, excessive current spikes, and so on. Therefore, the state of the I/O pins during power up and power down varies depending on which supplies are powered.

Table 7 gives the pin state for the sequence cases for all pins with pad type pad\_fc (fast type).

V <sub>DDE</sub>	V <sub>DD33</sub>	V <sub>DD</sub>	POR	Pin Status for Fast Pad Output Driver pad_fc (fast)
Low	—	_	Asserted	Low
V <sub>DDE</sub>	Low	Low	Asserted	High
V <sub>DDE</sub>	Low	V <sub>DD</sub>	Asserted	High
V <sub>DDE</sub>	V <sub>DD33</sub>	Low	Asserted	High impedance (Hi-Z)
V <sub>DDE</sub>	V <sub>DD33</sub>	V <sub>DD</sub>	Asserted	Hi-Z
V <sub>DDE</sub>	V <sub>DD33</sub>	V <sub>DD</sub>	Negated	Functional

Table 7. Pin Status for Fast Pads During the Power Sequence

Table 8 gives the pin state for the sequence cases for all pins with pad type pad\_mh (medium type) and pad\_sh (slow type).

Table 8. Pin Status for Medium and Slow Pads During the Power Sequence

V <sub>DDEH</sub>	V <sub>DD</sub>	POR	Pin Status for Medium and Slow Pad Output Driver pad_mh (medium) pad_sh (slow)
Low	_	Asserted	Low
V <sub>DDEH</sub>	Low	Asserted	High impedance (Hi-Z)
V <sub>DDEH</sub>	$V_{DD}$	Asserted	Hi-Z
V <sub>DDEH</sub>	V <sub>DD</sub>	Negated	Functional

The values in Table 7 and Table 8 do not include the effect of the weak-pull devices on the output pins during power up.

Before exiting the internal POR state, the voltage on the pins go to a high-impedance state until POR negates. When the internal POR negates, the functional state of the signal during reset applies and the weak-pull devices

(up or down) are enabled as defined in the device reference manual. If  $V_{DD}$  is too low to correctly propagate the logic signals, the weak-pull devices can pull the signals to  $V_{DDE}$  and  $V_{DDEH}$ .

To avoid this condition, minimize the ramp time of the  $V_{DD}$  supply to a time period less than the time required to enable the external circuitry connected to the device outputs.



# 3.7.1 Input Value of Pins During POR Dependent on V<sub>DD33</sub>

When powering up the device,  $V_{DD33}$  must not lag the latest  $V_{DDSYN}$  or RESET power pin ( $V_{DDEH6}$ ) by more than the  $V_{DD33}$  lag specification listed in Table 6, spec 8. This avoids accidentally selecting the bypass clock mode because the internal versions of PLLCFG[0:1] and RSTCFG are not powered and therefore cannot read the default state when POR negates.  $V_{DD33}$  can lag  $V_{DDSYN}$  or the RESET power pin ( $V_{DDEH6}$ ), but cannot lag both by more than the  $V_{DD33}$  lag specification. This  $V_{DD33}$  lag specification applies during power up only.  $V_{DD33}$  has no lead or lag requirements when powering down.

# 3.7.2 Power-Up Sequence (V<sub>RC33</sub> Grounded)

The 1.5 V V<sub>DD</sub> power supply must rise to 1.35 V before the 3.3 V V<sub>DDSYN</sub> power supply and the RESET power supply rises above 2.0 V. This ensures that digital logic in the PLL for the 1.5 V power supply does not begin to operate below the specified operation range lower limit of 1.35 V. Because the internal 1.5 V POR is disabled, the internal 3.3 V POR or the RESET power POR must hold the device in reset. Since they can negate as low as 2.0 V, V<sub>DD</sub> must be within specification before the 3.3 V POR and the RESET POR negate.



Figure 3. Power-Up Sequence (V<sub>RC33</sub> Grounded)

## 3.7.3 Power-Down Sequence (V<sub>RC33</sub> Grounded)

The only requirement for the power-down sequence with  $V_{RC33}$  grounded is if  $V_{DD}$  decreases to less than its operating range,  $V_{DDSYN}$  or the RESET power must decrease to less than 2.0 V before the  $V_{DD}$  power increases to its operating range. This ensures that the digital 1.5 V logic, which is reset only by an ORed POR and can cause the 1.5 V supply to decrease less than its specification value, resets correctly. See Table 6, footnote 1.



# 3.8 DC Electrical Specifications

### Table 9. DC Electrical Specifications ( $T_A = T_L$ to $T_H$ )

Spec	Characteristic	Symbol	Min	Max.	Unit
1	Core supply voltage (average DC RMS voltage)	V <sub>DD</sub>	1.35	1.65	V
2	Input/output supply voltage (fast input/output) <sup>1</sup>	V <sub>DDE</sub>	1.62	3.6	V
3	Input/output supply voltage (slow and medium input/output)	V <sub>DDEH</sub>	3.0	5.25	V
4	3.3 V input/output buffer voltage	V <sub>DD33</sub>	3.0	3.6	V
5	Voltage regulator control input voltage	V <sub>RC33</sub>	3.0	3.6	V
6	Analog supply voltage <sup>2</sup>	V <sub>DDA</sub>	4.5	5.25	V
8	Flash programming voltage <sup>3</sup>	V <sub>PP</sub>	4.5	5.25	V
9	Flash read voltage	V <sub>FLASH</sub>	3.0	3.6	V
10	SRAM standby voltage <sup>4</sup>	V <sub>STBY</sub>	0.8	1.2	V
11	Clock synthesizer operating voltage	V <sub>DDSYN</sub>	3.0	3.6	V
12	Fast I/O input high voltage	V <sub>IH_F</sub>	$0.65 \times V_{DDE}$	V <sub>DDE</sub> + 0.3	V
13	Fast I/O input low voltage	V <sub>IL_F</sub>	V <sub>SS</sub> – 0.3	$0.35 \times V_{DDE}$	V
14	Medium and slow I/O input high voltage	V <sub>IH_S</sub>	$0.65 \times V_{DDEH}$	V <sub>DDEH</sub> + 0.3	V
15	Medium and slow I/O input low voltage	V <sub>IL_S</sub>	V <sub>SS</sub> – 0.3	$0.35 \times V_{DDEH}$	V
16	Fast input hysteresis	V <sub>HYS_F</sub>	$0.1 \times V_{DDE}$		V
17	Medium and slow I/O input hysteresis	V <sub>HYS_S</sub>	$0.1 \times V_{DDEH}$		V
18	Analog input voltage	V <sub>INDC</sub>	V <sub>SSA</sub> – 0.3	V <sub>DDA</sub> + 0.3	V
19	Fast output high voltage (I <sub>OH_F</sub> = -2.0 mA)	V <sub>OH_F</sub>	$0.8 \times V_{DDE}$	_	V
20	Slow and medium output high voltage $I_{OH_S} = -2.0 \text{ mA}$ $I_{OH_S} = -1.0 \text{ mA}$	V <sub>OH_S</sub>	$\begin{array}{l} 0.80 \times V_{DDEH} \\ 0.85 \times V_{DDEH} \end{array}$	–	V
21	Fast output low voltage (I <sub>OL_F</sub> = 2.0 mA)	V <sub>OL_F</sub>	—	$0.2 \times V_{DDE}$	V
22	Slow and medium output low voltage $I_{OL_S} = 2.0 \text{ mA}$ $I_{OL_S} = 1.0 \text{ mA}$	V <sub>OL_S</sub>	_	$0.20 \times V_{DDEH}$ $0.15 \times V_{DDEH}$	V
23	Load capacitance (fast $I/O$ ) <sup>5</sup> DSC (SIU_PCR[8:9]) = 0b00 = 0b01 = 0b10 = 0b11	CL	 	10 20 30 50	pF pF pF pF
24	Input capacitance (digital pins)	C <sub>IN</sub>	_	7	pF
25	Input capacitance (analog pins)	C <sub>IN_A</sub>	—	10	pF
26	Input capacitance: (Shared digital and analog pins AN[12]_MA[0]_SDS, AN[13]_MA[1]_SDO, AN[14]_MA[2]_SDI, and AN[15]_FCK)	C <sub>IN_M</sub>	_	12	pF



Spec	Characteristic	Symbol	Min	Max.	Unit
28	Operating current 3.3 V supplies @ f <sub>MAX</sub> MHz				
	V <sub>DD33</sub> <sup>13</sup>	I <sub>DD_33</sub>	_	2 + (values derived from procedure of footnote <sup>13</sup> )	mA
	V <sub>FLASH</sub>	I <sub>VFLASH</sub>	—	10	mA
	V <sub>DDSYN</sub>	IDDSYN	—	15	mA
29	Operating current 5.0 V supplies (12 MHz ADCLK): V <sub>DDA</sub> (V <sub>DDA0</sub> + V <sub>DDA1</sub> ) Analog reference supply current (V <sub>RH</sub> , V <sub>RL</sub> ) V <sub>PP</sub>	I <sub>DD_A</sub> I <sub>REF</sub> I <sub>PP</sub>	  	20.0 1.0 25.0	mA mA mA
30	$\begin{array}{c} \text{Operating current } V_{\text{DDE}} \text{ supplies: }^{14} \\ V_{\text{DDE1}} \\ V_{\text{DDE2}} \\ V_{\text{DDE3}} \\ V_{\text{DDE44}} \\ V_{\text{DDE5}} \\ V_{\text{DDE46}} \\ V_{\text{DDE7}} \\ V_{\text{DDE48}} \\ V_{\text{DDEH9}} \end{array}$	I <sub>DD1</sub> I <sub>DD2</sub> I <sub>DD3</sub> I <sub>DD4</sub> I <sub>DD5</sub> I <sub>DD6</sub> I <sub>DD7</sub> I <sub>DD8</sub> I <sub>DD8</sub> I <sub>DD9</sub>		Refer to footnote <sup>14</sup>	mA mA mA mA mA mA mA
31	Fast I/O weak pullup current <sup>15</sup> 1.62–1.98 V 2.25–2.75 V 3.00–3.60 V		10 20 20	110 130 170	μΑ μΑ μΑ
	Fast I/O weak pulldown current <sup>15</sup> 1.62–1.98 V 2.25–2.75 V 3.00–3.60 V	- I <sub>ACT_F</sub>	10 20 20	100 130 170	μΑ μΑ μΑ
32	Slow and medium I/O weak pullup/down current <sup>15</sup> 3.0–3.6 V 4.5–5.5 V	I <sub>ACT_S</sub>	10 20	150 170	μA μA
33	I/O input leakage current <sup>16</sup>	I <sub>INACT_D</sub>	-2.5	2.5	μA
34	DC injection current (per pin)	I <sub>IC</sub>	-2.0	2.0	mA
35	Analog input current, channel off <sup>17</sup>	I <sub>INACT_A</sub>	-150	150	nA
35a	Analog input current, shared analog / digital pins (AN[12], AN[13], AN[14], AN[15])	I <sub>INACT_AD</sub>	-2.5	2.5	μA
36	$V_{SS}$ to $V_{SSA}$ differential voltage <sup>18</sup>	$V_{SS} - V_{SSA}$	-100	100	mV
37	Analog reference low voltage	V <sub>RL</sub>	V <sub>SSA</sub> – 0.1	V <sub>SSA</sub> + 0.1	V
38	V <sub>RL</sub> differential voltage	V <sub>RL</sub> – V <sub>SSA</sub>	-100	100	mV
39	Analog reference high voltage	V <sub>RH</sub>	V <sub>DDA</sub> – 0.1	V <sub>DDA</sub> + 0.1	V
40	V <sub>REF</sub> differential voltage	V <sub>RH</sub> – V <sub>RL</sub>	4.5	5.25	V

# Table 9. DC Electrical Specifications ( $T_A = T_L \text{ to } T_H$ ) (continued)



## 3.8.2 I/O Pad V<sub>DD33</sub> Current Specifications

The power consumption of the  $V_{DD33}$  supply dependents on the usage of the pins on all I/O segments. The power consumption is the sum of all input and output pin  $V_{DD33}$  currents for all I/O segments. The output pin  $V_{DD33}$  current can be calculated from Table 11 based on the voltage, frequency, and load on all fast (pad\_fc) pins. The input pin  $V_{DD33}$  current can be calculated from Table 11 based on the voltage, frequency, and load on all pad\_sh and pad\_mh pins. Use linear scaling to calculate pin currents for voltage, frequency, and load parameters that fall outside the values given in Table 11.

Spec	Pad Type	Symbol	Frequency (MHz)	Load <sup>2</sup> (pF)	V <sub>DD33</sub> (V)	V <sub>DDE</sub> (V)	Drive Select	Current (mA)		
Inputs										
1	Slow	I <sub>33_SH</sub>	66	0.5	3.6	5.5	NA	0.003		
2	Medium	I <sub>33_MH</sub>	66	0.5	3.6	5.5	NA	0.003		
	Outputs									
3			66	10	3.6	3.6	00	0.35		
4			66	20	3.6	3.6	01	0.53		
5			66	30	3.6	3.6	10	0.62		
6			66	50	3.6	3.6	11	0.79		
7			66	10	3.6	1.98	00	0.35		
8			66	20	3.6	1.98	01	0.44		
9			66	30	3.6	1.98	10	0.53		
10			66	50	3.6	1.98	11	0.70		
11			56	10	3.6	3.6	00	0.30		
12			56	20	3.6	3.6	01	0.45		
13			56	30	3.6	3.6	10	0.52		
14	Feet	Fast I <sub>33_FC</sub>	56	50	3.6	3.6	11	0.67		
15	Γαδι		56	10	3.6	1.98	00	0.30		
16			56	20	3.6	1.98	01	0.37		
17			56	30	3.6	1.98	10	0.45		
18			56	50	3.6	1.98	11	0.60		
19			40	10	3.6	3.6	00	0.21		
20			40	20	3.6	3.6	01	0.31		
21			40	30	3.6	3.6	10	0.37		
22			40	50	3.6	3.6	11	0.48		
23			40	10	3.6	1.98	00	0.21		
24			40	20	3.6	1.98	01	0.27		
25			40	30	3.6	1.98	10	0.32		
26			40	50	3.6	1.98	11	0.42		

Table 11. V <sub>DD3</sub>	3 Pad Averag	e DC Current	(T <sub>A</sub> =	T <sub>L</sub> to T <sub>H</sub> ) <sup>1</sup>
----------------------------	--------------	--------------	-------------------	---

<sup>1</sup> These values are estimated from simulation and not tested. Currents apply to output pins for the fast pads only and to input pins for the slow and medium pads only.

<sup>2</sup> All loads are lumped.



Spec	Pad	SRC/DSC (binary)	Out Delay <sup>2, 3, 4</sup> (ns)	Rise / Fall <sup>3, 5</sup> (ns)	Load Drive (pF)
		00		2.4	10
3	Fast	01	3.2	2.2	20
3	Fasi	10		2.1	30
		11		2.1	50
4	Pullup/down (3.6 V max)	—	—	7500	50
5	Pullup/down (5.5 V max)	—	—	9500	50

### Table 18. Derated Pad AC Specifications ( $V_{DDEH}$ = 3.3 V, $V_{DDE}$ = 3.3 V)<sup>1</sup> (continued)

<sup>1</sup> These are worst-case values that are estimated from simulation (not tested). The values in the table are simulated at:  $V_{DD} = 1.35-1.65 \text{ V}; V_{DDE} = 3.0-3.6 \text{ V}; V_{DDEH} = 3.0-3.6 \text{ V}; V_{DD33} \text{ and } V_{DDSYN} = 3.0-3.6 \text{ V}; \text{ and } T_A = T_L \text{ to } T_H.$ 

<sup>2</sup> This parameter is supplied for reference and guaranteed by design (not tested).

<sup>3</sup> The output delay, and the rise and fall, are calculated to 20% or 80% of the respective signal.

- <sup>4</sup> The output delay is shown in Figure 4. To calculate the output delay with respect to the system clock, add a maximum of one system clock to the output delay.
- <sup>5</sup> This parameter is guaranteed by characterization rather than 100% tested.



Figure 4. Pad Output Delay

# 3.13 AC Timing

### 3.13.1 Reset and Configuration Pin Timing

Table 19. Reset and Configuration Pin Timing <sup>1</sup>

Spec	Characteristic	Symbol	Min.	Max.	Unit
1	RESET pulse width	t <sub>RPW</sub>	10	_	t <sub>CYC</sub>
2	RESET glitch detect pulse width	t <sub>GPW</sub>	2	_	t <sub>CYC</sub>









Figure 9. JTAG Boundary Scan Timing



#### **Nexus Timing** 3.13.3

Spec	Characteristic	Symbol	Min.	Max.	Unit
1	MCKO cycle time	t <sub>MCYC</sub>	1 <sup>2</sup>	8	t <sub>CYC</sub>
2	MCKO duty cycle	t <sub>MDC</sub>	40	60	%
3	MCKO low to MDO data valid <sup>3</sup>	t <sub>MDOV</sub>	-1.5	3.0	ns
4	MCKO low to MSEO data valid <sup>3</sup>	t <sub>MSEOV</sub>	-1.5	3.0	ns
5	MCKO low to EVTO data valid <sup>3</sup>	t <sub>EVTOV</sub>	-1.5	3.0	ns
6	EVTI pulse width	t <sub>EVTIPW</sub>	4.0	—	t <sub>TCYC</sub>
7	EVTO pulse width	t <sub>EVTOPW</sub>	1	—	t <sub>MCYC</sub>
8	TCK cycle time	t <sub>TCYC</sub>	4 <sup>4</sup>	—	t <sub>CYC</sub>
9	TCK duty cycle	t <sub>TDC</sub>	40	60	%
10	TDI, TMS data setup time	t <sub>NTDIS</sub> , t <sub>NTMSS</sub>	8	—	ns
11	TDI, TMS data hold time	t <sub>NTDIH,</sub> t <sub>NTMSH</sub>	5	—	ns
	TCK low to TDO data valid	t <sub>JOV</sub>			
12	V <sub>DDE</sub> = 2.25–3.0 V		0	12	ns
	V <sub>DDE</sub> = 3.0–3.6 V		0	10	ns
13	RDY valid to MCKO <sup>5</sup>	_	_	—	—

Table 21. Nexus Debug Port Timing <sup>1</sup>

1 JTAG specifications apply when used for debug functionality. All Nexus timing relative to MCKO is measured from 50% of MCKO and 50% of the respective signal. Nexus timing specified at  $V_{DD}$  = 1.35–1.65 V,  $V_{DDE}$  = 2.25–3.6 V,

 $V_{DD33}$  and  $V_{DDSYN}$  = 3.0–3.6 V,  $T_A$  =  $T_L$  to  $T_H$ , and CL = 30 pF with DSC = 0b10.

- <sup>2</sup> The Nexus AUX port runs up to 82 MHz. Set NPC\_PCR[MCKO\_DIV] to divide-by-two if the system frequency is greater than 82 MHz.
- <sup>3</sup> MDO, MSEO, and EVTO data is held valid until the next MCKO low cycle occurs.
- <sup>4</sup> Limit the maximum frequency to approximately 16 MHz (V<sub>DDE</sub> = 2.25–3.0 V) or 20 MHz (V<sub>DDE</sub> = 3.0–3.6 V) to meet the timing specification for t<sub>JOV</sub> of [0.2 x t<sub>JCYC</sub>] as outlined in the IEEE-ISTO 5001-2003 specification.
- <sup>5</sup> The RDY pin timing is asynchronous to MCKO and is guaranteed by design to function correctly.



Figure 10. Nexus Output Timing



# 3.13.4 External Bus Interface (EBI) Timing

Table 22 lists the timing information for the external bus interface (EBI).

	Characteristic		External Bus Frequency <sup>2, 3</sup>									
Spec	and	Symbol	40 N	ЛНz	56 N	ЛНz	67	MHz	72	٨Hz	Unit	Notes
	Description		Min	Max	Min	Max	Min	Max	Min	Max		
1	CLKOUT period	Т <sub>С</sub>	25.0	_	17.9	_	15.2	_	13.3		ns	Signals are measured at 50% V <sub>DDE</sub> .
2	CLKOUT duty cycle	t <sub>CDC</sub>	45%	55%	45%	55%	45%	55%	45%	55%	Т <sub>С</sub>	
3	CLKOUT rise time	t <sub>CRT</sub>		4		4	_	4		4	ns	
4	CLKOUT fall time	t <sub>CFT</sub>	_	<sup>4</sup>	_	<sup>4</sup>		<sup>4</sup>	_	4	ns	
5	CLKOUT positive edge to output signal <i>invalid</i> or Hi-Z (hold time) External bus interface CS[0:3] ADDR[8:31] DATA[0:31] BDIP BG <sup>5</sup> BR <sup>7</sup> BB OE RD_WR TA TEA TS TSIZ[0:1] WE/BE[0:3]	t <sub>сон</sub>	1.0 <sup>6</sup> 1.5		1.0 <sup>6</sup> 1.5		1.0 <sup>6</sup> 1.5		1.0 <sup>6</sup> 1.5		ns	EBTS = 0 EBTS = 1 Hold time selectable via SIU_ECCR [EBTS] bit.
	CLKOUT positive edge to output signal <i>invalid</i> or Hi-Z (hold time) Calibration bus interface CAL_CS[0:3] CAL_ADDR[9:30] CAL_DATA[0:15] CAL_OE CAL_RD_WR CAL_TS CAL_TS CAL_WE/BE[0:1]	t <sub>ссон</sub>	1.0 <sup>6</sup> 1.5	_	1.0 <sup>6</sup> 1.5	_	1.0 <sup>6</sup> 1.5	_	1.0 <sup>6</sup> 1.5	_	ns	EBTS = 0 EBTS = 1 Hold time selectable via SIU_ECCR [EBTS] bit.

### Table 22. Bus Operation Timing <sup>1</sup>





Figure 15. External Interrupt Timing

### 3.13.6 eTPU Timing

Table 24. eTPU Timing <sup>1</sup>

Spec	Characteristic	Symbol	Min.	Мах	Unit
1	eTPU input channel pulse width	t <sub>ICPW</sub>	4	_	t <sub>CYC</sub>
2	eTPU output channel pulse width	t <sub>OCPW</sub>	2 <sup>2</sup>		t <sub>CYC</sub>

<sup>1</sup> eTPU timing specified at:  $V_{DDEH}$  = 3.0–5.25 V and  $T_A$  =  $T_L$  to  $T_H$ .

<sup>2</sup> This specification does not include the rise and fall times. When calculating the minimum eTPU pulse width, include the rise and fall times defined in the slew rate control fields (SRC) of the pad configuration registers (PCR).



Figure 16. eTPU Timing



Mechanicals

	1	2	3	4	5	6	7	8	9	10	11	12	13
Α	VSS	VSTBY	AN37	AN11	VDDA1	AN16	AN1	AN5	VRH	AN23	AN27	AN28	AN35
В	VDD	VSS	AN36	AN39	AN19	AN20	AN0	AN4	REF BYPC	AN22	AN26	AN31	AN32
С	VDD33	VDD	VSS	AN8	AN17	VSSA1	AN21	AN3	AN7	VRL	AN25	AN30	AN33
D	ETPUA 30	ETPUA 31	VDD	VSS	AN38	AN9	AN10	AN18	AN2	AN6	AN24	AN29	AN34
Е	ETPUA 28	ETPUA 29	VDDEH 1	VDD									
F	ETPUA 24	ETPUA 27	ETPUA 26	VDDEH 1									
G	ETPUA 23	ETPUA 22	ETPUA 25	ETPUA 21									
н	ETPUA 20	ETPUA 19	ETPUA 18	ETPUA 17									
J	ETPUA 16	ETPUA 15	ETPUA 14	ETPUA 13									
к	ETPUA 12	ETPUA 11	ETPUA 10	ETPUA 9						VSS	VSS	VSS	VSS
L	ETPUA 8	ETPUA 7	ETPUA 6	ETPUA 5						VSS	VSS	VSS	VSS
М	ETPUA 4	ETPUA 3	ETPUA 2	ETPUA 1						VDDE2	VDDE2	VSS	VSS
Ν	BDIP	TEA	ETPUA 0	TCRCLK A						VDDE2	VDDE2	VSS	VSS
Ρ	CS3	CS2	CS1	CS0						VDDE2	VDDE2	VSS	VSS
R	WE3	WE2	WE1	WE0						VDDE2	VDDE2	VSS	VSS
т	VDDE2	TSIZ0	RD_WR	VDDE2						VDDE2	VSS	VDDE2	VDDE2
U	ADDR 16	TSIZ1	TA	VDD33						VSS	VDDE2	VDDE2	VDDE2
V	ADDR 18	ADDR 17	TS	ADDR 8									
w	ADDR 20	ADDR 19	ADDR 9	ADDR 10									
Y	ADDR 22	ADDR 21	ADDR 11	VDDE2									
AA	ADDR 24	ADDR 23	ADDR 13	ADDR 12									
AB	VDDE2	ADDR 25	ADDR 15	ADDR 14									
AC	ADDR 26	ADDR 27	ADDR 31	VSS	VDD	DATA 26	DATA 28	VDDE2	DATA 30	DATA 31	DATA 8	DATA 10	VDDE2
AD	ADDR 28	ADDR 30	VSS	VDD	DATA 24	DATA 25	DATA 27	DATA 29	VDD33	GPIO 207	DATA 9	DATA 11	DATA 13
AE	ADDR 29	VSS	VDD	DATA 17	DATA 19	DATA 21	DATA 23	DATA 0	DATA 2	DATA 4	DATA 6	OE	BR
AF	VSS	VDD	DATA 16	DATA 18	VDDE2	DATA 20	DATA 22	GPIO 206	DATA 1	DATA 3	VDDE2	DATA 5	DATA 7
	1	2	3	4	5	6	7	8	9	10	11	12	13
			Figure	e 33. M	PC556	6 416 F	ackag	ge Left	Side (	view 1	ot 2)		



# 4.2 MPC5566 416-Pin Package Dimensions

The package drawings of the MPC5566 416 pin TEPBGA package are shown in Figure 36.



© FREE	SCALE SEMICONDUCTOR, INC. ALL RIGHTS RESERVED.	MECHANICA	L OUTLINE	PRINT VERSION NO	T TO SCALE
TITLE:	416 I/O, PBGA		DOCUMENT NO	): 98ARE10523D	REV: A
	27 X 27 PKG,		CASE NUMBER	₹: 1494–01	13 JUL 2005
	1 MM PITCH (OMPAC	)	STANDARD: JE	DEC MS-034 AAL-1	





#### **Revision History for the MPC5566 Data Sheet**

Location	Description of Changes							
Section 3.7	Section 3.7.1, "Input Value of Pins During POR Dependent on VDD33:"							
	Added the following text directly before this section and after Table 8 <i>Pin Status for Medium / Slow Pads During the Power-on Sequence</i> : 'The values in Table 7 and Table 8 do not include the effect of the weak pull devices on the output pins during power up.							
	Before exiting the internal POR state, the voltage on the pins goes to high-impedance until POR negates. When the internal POR negates, the functional state of the signal during reset applies and the weak pull devices (up or down) are enabled as defined in the device <i>Reference Manual</i> . If $V_{DD}$ is too low to correctly propagate the logic signals, the weak-pull devices can pull the signals to $V_{DDE}$ and $V_{DDEH}$ .							
	To avoid this condition, minimize the ramp time of the V <sub>DD</sub> supply to a time period less than the time required to enable the external circuitry connected to the device outputs.'							
Section 3.7	7.3, "Power-Down Sequence (VRC33 Grounded)" Deleted the underscore in ORed POR to become ORed POR.							

#### Table 34. Global and Text Changes Between Rev. 0.0 and 1.0 (continued)

The following table lists the information that changed in the figures or tables between Rev. 0.0 and 1.0.

 Table 35. Table and Figure Changes Between Rev. 0.0 and Rev. 1.0

Location	Description of Changes
Figure 1, N	IPC5500 Family Part Numbers:
	<ul> <li>Removed the 2 in the tape and reel designator in both the graphic and in the Tape and Reel Status text.</li> <li>Changed Qualification Status by adding ', general market flow' to the M designator, and added an 'S' designator with the description of 'Fully spec. qualified, automotive flow.</li> </ul>
Table 1, Or	derable Part Numbers:
	<ul> <li>Added a 144 MHz system frequency option for:</li> <li>MPC5566MVR144, Pb-Free (lead free), nominal 144, maximum 147</li> <li>MPC5566MZP144, SnPb (leaded), nominal 144, maximum 147</li> <li>Changed the 132 MHz maximum operating frequency to 135 MHz.</li> </ul>

- Reordered rows to group devices by lead-free package types in descending frequency order, and leaded package types.
- Footnote 1 added that reads: All devices are PPC5566, rather than MPC5566 or SPC5566, until product qualifications are complete. Not all configurations are available in the PPC parts.
- Footnote 2 added that reads: The lowest ambient operating temperature is referenced by T<sub>L</sub>; the highest ambient operating temperature is referenced by T<sub>H</sub>.
- Changed footnote 3 from '132 MHz allows only 128 MHz + 2% FM' to '135 MHz parts allow for 132 MHz systems clock + 2% FM'; and added '147 MHz parts allow for 144 MHz systems clock + 2% FM.



### Revision History for the MPC5566 Data Sheet

Table 35. Table and Figure Changes	Between Rev. 0.0 and Rev. 1.0 (co	ontinued)
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Location	Description of Changes
Table 5, ES	D Characteristics: Added (Electromagnetic Static Discharge) in the table title.
Table 6, VC	R/POR Electrical Specifications:
	<ul> <li>Added footnote 1 to specs 1, 2, and 3 that reads: On power up, assert RESET before V<sub>POR15</sub>, V<sub>POR33</sub>, and V<sub>POR5</sub> negate (internal POR). RESET must remain asserted until the power supplies are within the operating conditions as specified in Table 9 <i>DC Electrical Specifications</i>. On power down, assert RESET before any power supplies fall outside the operating conditions and until the internal POR asserts.</li> <li>Subscript all symbol names that appear after the first underscore character.</li> <li>Specs 7 and 10: added 'at Tj' at the end of the first line in the second column: Characteristic.</li> <li>Spec 10, second column, second line: Added cross-reference to footnote 6: 'I<sub>VRCCTL</sub> is measured at the following conditions: V<sub>DD</sub> = 1.35 V, V<sub>RC33</sub> = 3.1 V, V<sub>VRCCTL</sub> = 2.2 V.' Changed '(@ V<sub>DD</sub> = 1.35 V, f<sub>Sy8</sub> = f<sub>MAX</sub>)' to '(@ f<sub>Sy8</sub> = f<sub>MAX</sub>).'</li> <li>Footnote 10: Deleted 'Preliminary value. Final specification pending characterization."</li> <li>Added to Spec 2: 3.3 V (V<sub>DDSYN</sub>) POR negated (ramp down) Min 0.0 Max 0.30 V</li> <li>3.3 V (V<sub>DDSYN</sub>) POR asserted (ramp up) Min 0.0 Max 0.30 V</li> <li>Added new footnote 1 to both lines in Spec 3: "V<sub>IL_S</sub> (Table 9, Spec 15) is guaranteed to scale with V<sub>DDEH6</sub> down to V<sub>POR5</sub>.</li> <li>Spec 5: Changed old Footnote 1 (now footnote 2): 'User must be able to supply full operating current for the 1.5V supply when the 3.3 V supply reaches this range."</li> <li>Spec 10:</li> <li>Changed old Footnote 3 for both lines: 'It is possible to reach the current limit during ramp updo not treat this event as a short circuit current.'</li> <li>Spec 10:</li> <li>Changed the minimum values of: -40 C = 60; 25 C = 65.</li> <li>Added old footnote 7, 'Refer to Table 1 for the maximum operating frequency.'</li> <li>Rewrote old footnote 7, Refer to Table 1 for the maximum operating frequency.'</li> <li>Rewrote old footnote 7, Refer to Table 1 for the maximum operating frequency.'</li> <li>Rewrote old footnote 5 new footnote 6.</li> <li>Added a new footnote 7, Refer to Ta</li></ul>
Table 7, Po	wer Sequence Pin Status for Fast Pads:
-	<ul> <li>Changed title to <i>Pin Status for Fast Pads During the Power Sequence</i></li> <li>Changed preceding paragraph From: Although there are no power up/down sequencing requirements to prevent issues like latch-up, excessive current spikes, etc., the state of the I/O pins during power up/down varies depending on power. Prior to exiting POR, the pads are in a high impedance state (Hi-Z). To: There are no power up/down sequencing requirements to prevent issues such as latch-up, excessive current spikes, and so on. Therefore, the state of the I/O pins during power up/down varies depending on which supplies are powered.</li> <li>Deleted the 'Comment' column.</li> <li>Added a POR column after the V<sub>DD</sub> column.</li> <li>Added row 2:' V<sub>DDE</sub>, Low, Low, Asserted, High' and row 5: V<sub>DDE</sub>, V<sub>DD33</sub>, V<sub>DD</sub>, Asserted, Hi-Z.</li> </ul>
Table 8, Po	wer Sequence Pin Status for Medium/Slow Pads:
	<ul> <li>Changed title to <i>Pin Status for Medium and Slow Pads During the Power Sequence</i></li> <li>Updated preceding paragraph.</li> <li>Deleted the 'Comment' column.</li> <li>Added a POR column after the V<sub>DD</sub> column.</li> </ul>

• Added row 3:' V<sub>DDEH</sub>, V<sub>DD</sub>, Asserted, Hi-Z.'