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"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Not For New Designs
Core Processor	8051
Core Size	8-Bit
Speed	25MHz
Connectivity	SMBus (2-Wire/I <sup>2</sup> C), SPI, UART/USART
Peripherals	POR, PWM, Temp Sensor, WDT
Number of I/O	25
Program Memory Size	8KB (8K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	1.25K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 3.6V
Data Converters	A/D 17x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	28-VFQFN Exposed Pad
Supplier Device Package	28-QFN (5x5)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/silicon-labs/c8051f313-gm">https://www.e-xfl.com/product-detail/silicon-labs/c8051f313-gm</a>

# C8051F310/1/2/3/4/5/6/7

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SFR Definition 13.5. P0MDOUT: Port0 Output Mode .....	137
SFR Definition 13.6. P0SKIP: Port0 Skip .....	137
SFR Definition 13.7. P1: Port1 .....	138
SFR Definition 13.8. P1MDIN: Port1 Input Mode .....	138
SFR Definition 13.9. P1MDOUT: Port1 Output Mode .....	139
SFR Definition 13.10. P1SKIP: Port1 Skip .....	139
SFR Definition 13.11. P2: Port2 .....	140
SFR Definition 13.12. P2MDIN: Port2 Input Mode .....	140
SFR Definition 13.13. P2MDOUT: Port2 Output Mode .....	141
SFR Definition 13.14. P2SKIP: Port2 Skip .....	141
SFR Definition 13.15. P3: Port3 .....	142
SFR Definition 13.16. P3MDIN: Port3 Input Mode .....	142
SFR Definition 13.17. P3MDOUT: Port3 Output Mode .....	143
SFR Definition 14.1. SMB0CF: SMBus Clock/Configuration .....	152
SFR Definition 14.2. SMB0CN: SMBus Control .....	154
SFR Definition 14.3. SMB0DAT: SMBus Data .....	156
SFR Definition 15.1. SCON0: Serial Port 0 Control .....	168
SFR Definition 15.2. SBUF0: Serial (UART0) Port Data Buffer .....	169
SFR Definition 16.1. SPI0CFG: SPI0 Configuration .....	180
SFR Definition 16.2. SPI0CN: SPI0 Control .....	181
SFR Definition 16.3. SPI0CKR: SPI0 Clock Rate .....	182
SFR Definition 16.4. SPI0DAT: SPI0 Data .....	182
SFR Definition 17.1. TCON: Timer Control .....	191
SFR Definition 17.2. TMOD: Timer Mode .....	192
SFR Definition 17.3. CKCON: Clock Control .....	193
SFR Definition 17.4. TL0: Timer 0 Low Byte .....	194
SFR Definition 17.5. TL1: Timer 1 Low Byte .....	194
SFR Definition 17.6. TH0: Timer 0 High Byte .....	194
SFR Definition 17.7. TH1: Timer 1 High Byte .....	194
SFR Definition 17.8. TMR2CN: Timer 2 Control .....	197
SFR Definition 17.9. TMR2RLL: Timer 2 Reload Register Low Byte .....	198
SFR Definition 17.10. TMR2RLH: Timer 2 Reload Register High Byte .....	198
SFR Definition 17.11. TMR2L: Timer 2 Low Byte .....	198
SFR Definition 17.12. TMR2H: Timer 2 High Byte .....	198
SFR Definition 17.13. TMR3CN: Timer 3 Control .....	201
SFR Definition 17.14. TMR3RLL: Timer 3 Reload Register Low Byte .....	202
SFR Definition 17.15. TMR3RLH: Timer 3 Reload Register High Byte .....	202
SFR Definition 17.16. TMR3L: Timer 3 Low Byte .....	202
SFR Definition 17.17. TMR3H: Timer 3 High Byte .....	202
SFR Definition 18.1. PCA0CN: PCA Control .....	215
SFR Definition 18.2. PCA0MD: PCA Mode .....	216
SFR Definition 18.3. PCA0CPMn: PCA Capture/Compare Mode Registers .....	217
SFR Definition 18.4. PCA0L: PCA Counter/Timer Low Byte .....	218
SFR Definition 18.5. PCA0H: PCA Counter/Timer High Byte .....	218
SFR Definition 18.6. PCA0CPLn: PCA Capture Module Low Byte .....	218

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# C8051F310/1/2/3/4/5/6/7

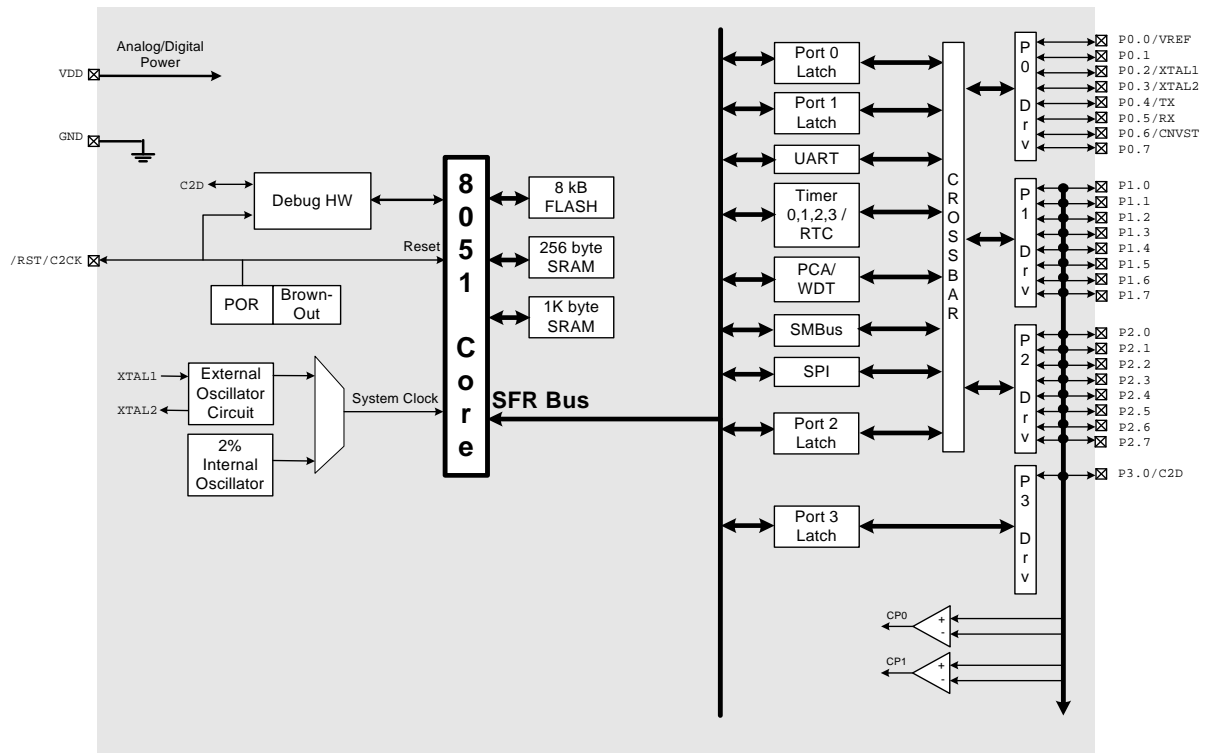
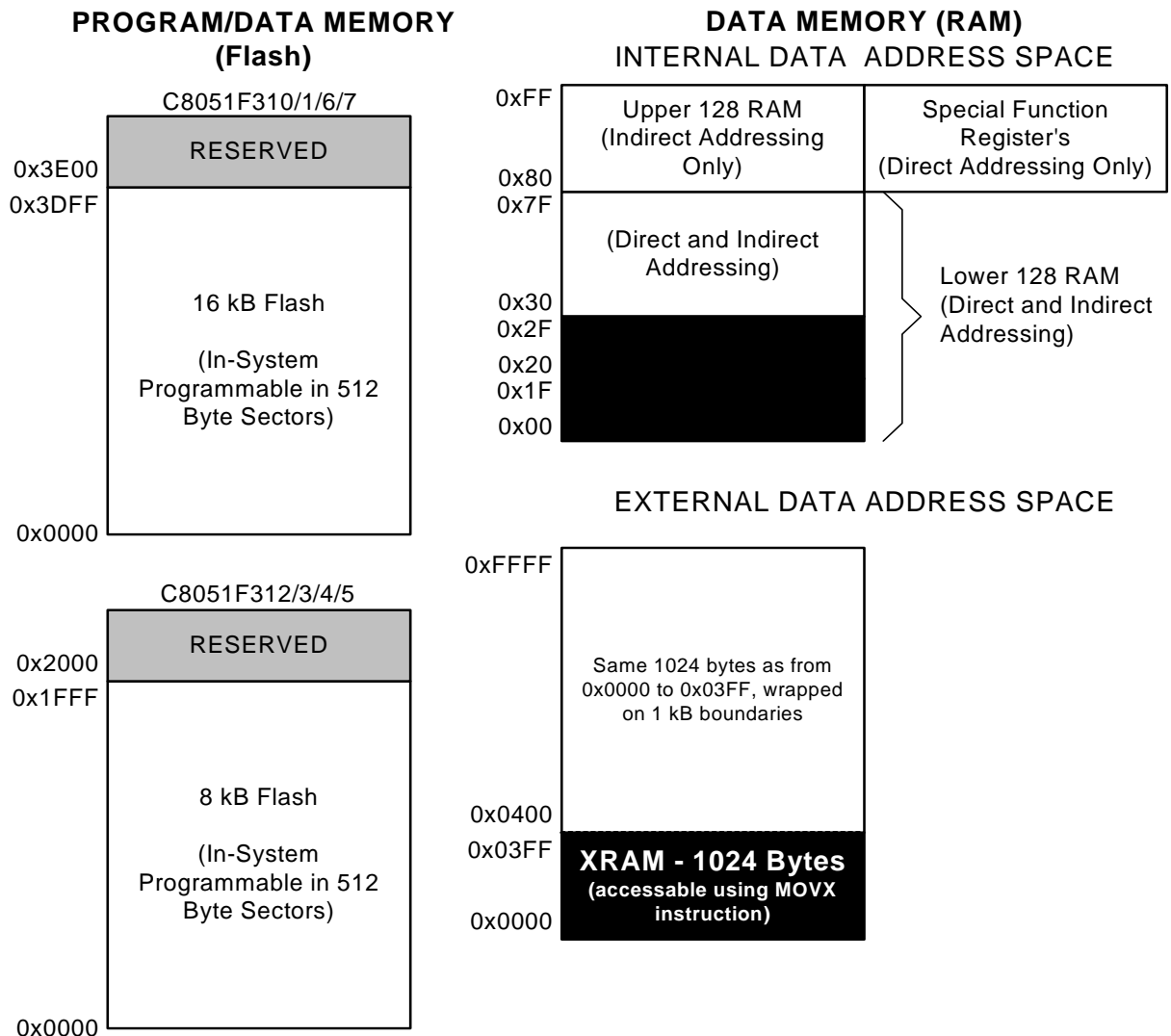


Figure 1.6. C8051F315 Block Diagram

## 1.2. On-Chip Memory

The CIP-51 has a standard 8051 program and data address configuration. It includes 256 bytes of data RAM, with the upper 128 bytes dual-mapped. Indirect addressing accesses the upper 128 bytes of general purpose RAM, and direct addressing accesses the 128 byte SFR address space. The lower 128 bytes of RAM are accessible via direct and indirect addressing. The first 32 bytes are addressable as four banks of general purpose registers, and the next 16 bytes can be byte addressable or bit addressable.

Program memory consists of 8 or 16 kB of Flash. This memory may be reprogrammed in-system in 512 byte sectors, and requires no special off-chip programming voltage. See Figure 1.11 for the MCU system memory map.



**Figure 1.11. On-Board Memory Map**

# C8051F310/1/2/3/4/5/6/7

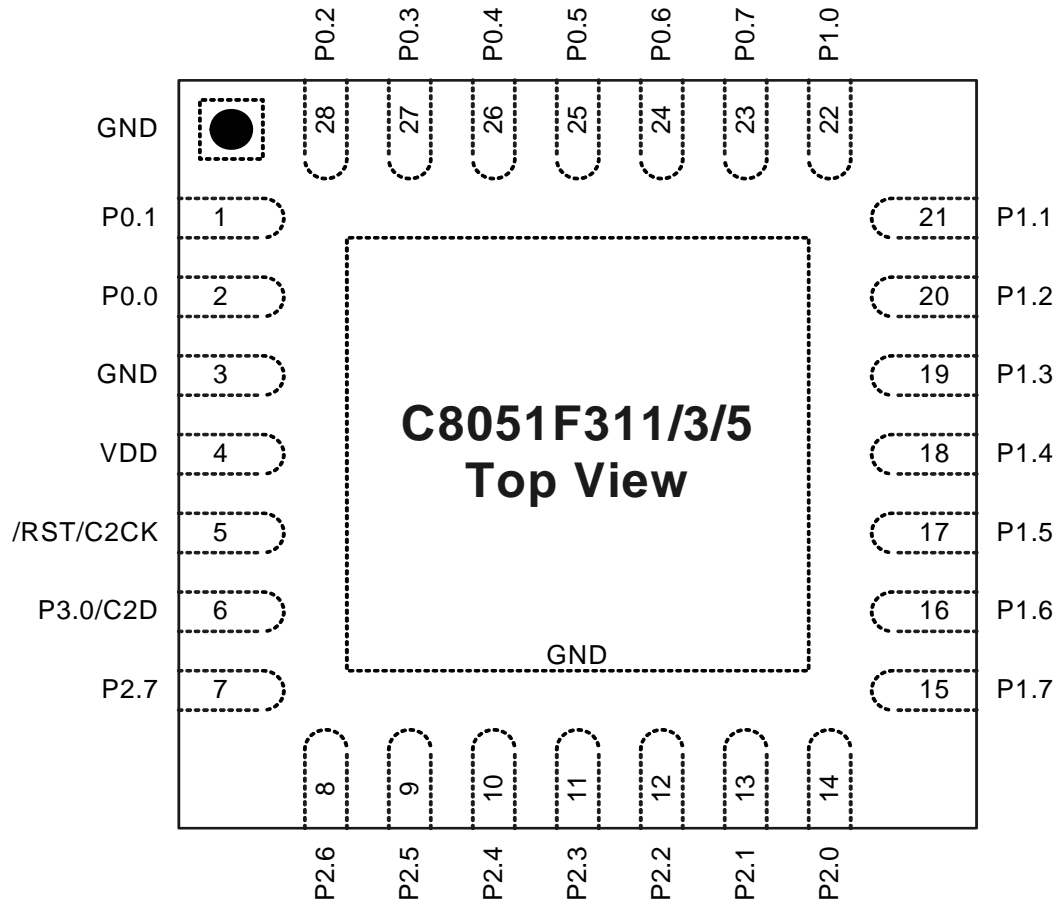


Figure 4.3. QFN-28 Pinout Diagram (Top View)

## SFR Definition 7.1. CPT0CN: Comparator0 Control

R/W	R	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
CP0EN	CP0OUT	CP0RIF	CP0FIF	CP0HYP1	CP0HYP0	CP0HYN1	CP0HYN0	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0x9B
<p>Bit7: CP0EN: Comparator0 Enable Bit. 0: Comparator0 Disabled. 1: Comparator0 Enabled.</p> <p>Bit6: CP0OUT: Comparator0 Output State Flag. 0: Voltage on CP0+ &lt; CP0−. 1: Voltage on CP0+ &gt; CP0−.</p> <p>Bit5: CP0RIF: Comparator0 Rising-Edge Interrupt Flag. 0: No Comparator0 Rising Edge Interrupt has occurred since this flag was last cleared. 1: Comparator0 Rising Edge Interrupt has occurred.</p> <p>Bit4: CP0FIF: Comparator0 Falling-Edge Interrupt Flag. 0: No Comparator0 Falling-Edge Interrupt has occurred since this flag was last cleared. 1: Comparator0 Falling-Edge Interrupt has occurred.</p> <p>Bits3–2: CP0HYP1–0: Comparator0 Positive Hysteresis Control Bits. 00: Positive Hysteresis Disabled. 01: Positive Hysteresis = 5 mV. 10: Positive Hysteresis = 10 mV. 11: Positive Hysteresis = 20 mV.</p> <p>Bits1–0: CP0HYN1–0: Comparator0 Negative Hysteresis Control Bits. 00: Negative Hysteresis Disabled. 01: Negative Hysteresis = 5 mV. 10: Negative Hysteresis = 10 mV. 11: Negative Hysteresis = 20 mV.</p>								

# C8051F310/1/2/3/4/5/6/7

## SFR Definition 8.4. PSW: Program Status Word

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R	Reset Value
CY	AC	F0	RS1	RS0	OV	F1	PARITY	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
						(bit addressable)		0xD0

Bit7: CY: Carry Flag.  
 This bit is set when the last arithmetic operation resulted in a carry (addition) or a borrow (subtraction). It is cleared to logic 0 by all other arithmetic operations.

Bit6: AC: Auxiliary Carry Flag  
 This bit is set when the last arithmetic operation resulted in a carry into (addition) or a borrow from (subtraction) the high order nibble. It is cleared to logic 0 by all other arithmetic operations.

Bit5: F0: User Flag 0.  
 This is a bit-addressable, general purpose flag for use under software control.

Bits4–3: RS1–RS0: Register Bank Select.  
 These bits select which register bank is used during register accesses.

RS1	RS0	Register Bank	Address
0	0	0	0x00–0x07
0	1	1	0x08–0x0F
1	0	2	0x10–0x17
1	1	3	0x18–0x1F

Bit2: OV: Overflow Flag.  
 This bit is set to 1 under the following circumstances: an ADD, ADDC, or SUBB instruction causes a sign-change overflow, a MUL instruction results in an overflow (result is greater than 255), or a DIV instruction causes a divide-by-zero condition. The OV bit is cleared to 0 by the ADD, ADDC, SUBB, MUL, and DIV instructions in all other cases.

Bit1: F1: User Flag 1.  
 This is a bit-addressable, general purpose flag for use under software control.

Bit0: PARITY: Parity Flag.  
 This bit is set to logic 1 if the sum of the eight bits in the accumulator is odd and cleared if the sum is even.

## SFR Definition 8.5. ACC: Accumulator

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
ACC.7	ACC.6	ACC.5	ACC.4	ACC.3	ACC.2	ACC.1	ACC.0	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address:
						(bit addressable)		0xE0

Bits7–0: ACC: Accumulator.  
 This register is the accumulator for arithmetic operations.

## SFR Definition 8.11. IT01CF: INT0/INT1 Configuration

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
IN1PL	IN1SL2	IN1SL1	IN1SL0	IN0PL	IN0SL2	IN0SL1	IN0SL0	00000001
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xE4

**Note:** Refer to SFR Definition 17.1 for INT0/1 edge- or level-sensitive interrupt selection.

Bit7: IN1PL: /INT1 Polarity  
 0: /INT1 input is active low.  
 1: /INT1 input is active high.

Bits6–4: IN1SL2–0: /INT1 Port Pin Selection Bits

These bits select which Port pin is assigned to /INT1. Note that this pin assignment is independent of the Crossbar; /INT1 will monitor the assigned Port pin without disturbing the peripheral that has been assigned the Port pin via the Crossbar. The Crossbar will not assign the Port pin to a peripheral if it is configured to skip the selected pin (accomplished by setting to '1' the corresponding bit in register P0SKIP).

IN1SL2–0	/INT1 Port Pin
000	P0.0
001	P0.1
010	P0.2
011	P0.3
100	P0.4
101	P0.5
110	P0.6
111	P0.7

Bit3: IN0PL: /INT0 Polarity  
 0: /INT0 interrupt is active low.  
 1: /INT0 interrupt is active high.

Bits2–0: IN0SL2–0: /INT0 Port Pin Selection Bits

These bits select which Port pin is assigned to /INT0. Note that this pin assignment is independent of the Crossbar; /INT0 will monitor the assigned Port pin without disturbing the peripheral that has been assigned the Port pin via the Crossbar. The Crossbar will not assign the Port pin to a peripheral if it is configured to skip the selected pin (accomplished by setting to '1' the corresponding bit in register P0SKIP).

IN0SL2–0	/INT0 Port Pin
000	P0.0
001	P0.1
010	P0.2
011	P0.3
100	P0.4
101	P0.5
110	P0.6
111	P0.7



## 8.4.2. Stop Mode

Setting the Stop Mode Select bit (PCON.1) causes the CIP-51 to enter Stop mode as soon as the instruction that sets the bit completes execution. In Stop mode the internal oscillator, CPU, and all digital peripherals are stopped; the state of the external oscillator circuit is not affected. Each analog peripheral (including the external oscillator circuit) may be shut down individually prior to entering Stop Mode. Stop mode can only be terminated by an internal or external reset. On reset, the CIP-51 performs the normal reset sequence and begins program execution at address 0x0000.

If enabled, the Missing Clock Detector will cause an internal reset and thereby terminate the Stop mode. The Missing Clock Detector should be disabled if the CPU is to be put in STOP mode for longer than the MCD timeout of 100  $\mu$ sec.

### SFR Definition 8.12. PCON: Power Control

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
GF5	GF4	GF3	GF2	GF1	GF0	STOP	IDLE	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0x87

Bits7–2: GF5–GF0: General Purpose Flags 5–0.  
 These are general purpose flags for use under software control.

Bit1: STOP: Stop Mode Select.  
 Setting this bit will place the CIP-51 in Stop mode. This bit will always be read as 0.  
 1: CPU goes into Stop mode (internal oscillator stopped).

Bit0: IDLE: Idle Mode Select.  
 Setting this bit will place the CIP-51 in Idle mode. This bit will always be read as 0.  
 1: CPU goes into Idle mode. (Shuts off clock to CPU, but clock to Timers, Interrupts, Serial Ports, and Analog Peripherals are still active.)

## SFR Definition 12.4. OSCXCN: External Oscillator Control

R	R/W	R/W	R/W	R	R/W	R/W	R/W	Reset Value
XTLVLD	XOSCND2	XOSCND1	XOSCND0		XFCN2	XFCN1	XFCN0	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xB1

- Bit7: XLVLD: Crystal Oscillator Valid Flag.  
(Read only when XOSCND = 11x.)  
0: Crystal Oscillator is unused or not yet stable.  
1: Crystal Oscillator is running and stable.
- Bits6–4: XOSCND2-0: External Oscillator Mode Bits.  
00x: External Oscillator circuit off.  
010: External CMOS Clock Mode.  
011: External CMOS Clock Mode with divide by 2 stage.  
100: RC Oscillator Mode.  
101: Capacitor Oscillator Mode.  
110: Crystal Oscillator Mode.  
111: Crystal Oscillator Mode with divide by 2 stage.
- Bit3: RESERVED. Read = 0, Write = don't care.
- Bits2–0: XFCN2-0: External Oscillator Frequency Control Bits.  
000-111: See table below:

XFCN	Crystal (XOSCND = 11x)	RC (XOSCND = 10x)	C (XOSCND = 10x)
000	$f \leq 32 \text{ kHz}$	$f \leq 25 \text{ kHz}$	K Factor = 0.87
001	$32 \text{ kHz} < f \leq 84 \text{ kHz}$	$25 \text{ kHz} < f \leq 50 \text{ kHz}$	K Factor = 2.6
010	$84 \text{ kHz} < f \leq 225 \text{ kHz}$	$50 \text{ kHz} < f \leq 100 \text{ kHz}$	K Factor = 7.7
011	$225 \text{ kHz} < f \leq 590 \text{ kHz}$	$100 \text{ kHz} < f \leq 200 \text{ kHz}$	K Factor = 22
100	$590 \text{ kHz} < f \leq 1.5 \text{ MHz}$	$200 \text{ kHz} < f \leq 400 \text{ kHz}$	K Factor = 65
101	$1.5 \text{ MHz} < f \leq 4 \text{ MHz}$	$400 \text{ kHz} < f \leq 800 \text{ kHz}$	K Factor = 180
110	$4 \text{ MHz} < f \leq 10 \text{ MHz}$	$800 \text{ kHz} < f \leq 1.6 \text{ MHz}$	K Factor = 664
111	$10 \text{ MHz} < f \leq 30 \text{ MHz}$	$1.6 \text{ MHz} < f \leq 3.2 \text{ MHz}$	K Factor = 1590

**CRYSTAL MODE** (Circuit from Figure 12.1, Option 1; XOSCND = 11x)  
Choose XFCN value to match crystal frequency.

**RC MODE** (Circuit from Figure 12.1, Option 2; XOSCND = 10x)  
Choose XFCN value to match frequency range:  
 $f = 1.23(10^3) / (R \times C)$ , where  
f = frequency of clock in MHz  
C = capacitor value in pF  
R = Pullup resistor value in k $\Omega$

**C MODE** (Circuit from Figure 12.1, Option 3; XOSCND = 10x)  
Choose K Factor (KF) for the oscillation frequency desired:  
 $f = KF / (C \times V_{DD})$ , where  
f = frequency of clock in MHz  
C = capacitor value the XTAL2 pin in pF  
 $V_{DD}$  = Power Supply on MCU in volts

## SFR Definition 13.1. XBR0: Port I/O Crossbar Register 0

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
CP1AE	CP1E	CP0AE	CP0E	SYSCKE	SMB0E	SPI0E	URT0E	00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SFR Address: 0xE1
<p>Bit7: CP1AE: Comparator1 Asynchronous Output Enable 0: Asynchronous CP1 unavailable at Port pin. 1: Asynchronous CP1 routed to Port pin.</p> <p>Bit6: CP1E: Comparator1 Output Enable 0: CP1 unavailable at Port pin. 1: CP1 routed to Port pin.</p> <p>Bit5: CP0AE: Comparator0 Asynchronous Output Enable 0: Asynchronous CP0 unavailable at Port pin. 1: Asynchronous CP0 routed to Port pin.</p> <p>Bit4: CP0E: Comparator0 Output Enable 0: CP0 unavailable at Port pin. 1: CP0 routed to Port pin.</p> <p>Bit3: SYSCKE: /SYSCLK Output Enable 0: /SYSCLK unavailable at Port pin. 1: /SYSCLK output routed to Port pin.</p> <p>Bit2: SMB0E: SMBus I/O Enable 0: SMBus I/O unavailable at Port pins. 1: SMBus I/O routed to Port pins.</p> <p>Bit1: SPI0E: SPI I/O Enable 0: SPI I/O unavailable at Port pins. 1: SPI I/O routed to Port pins.</p> <p>Bit0: URT0E: UART I/O Output Enable 0: UART I/O unavailable at Port pin. 1: UART TX0, RX0 routed to Port pins P0.4 and P0.5.</p>								

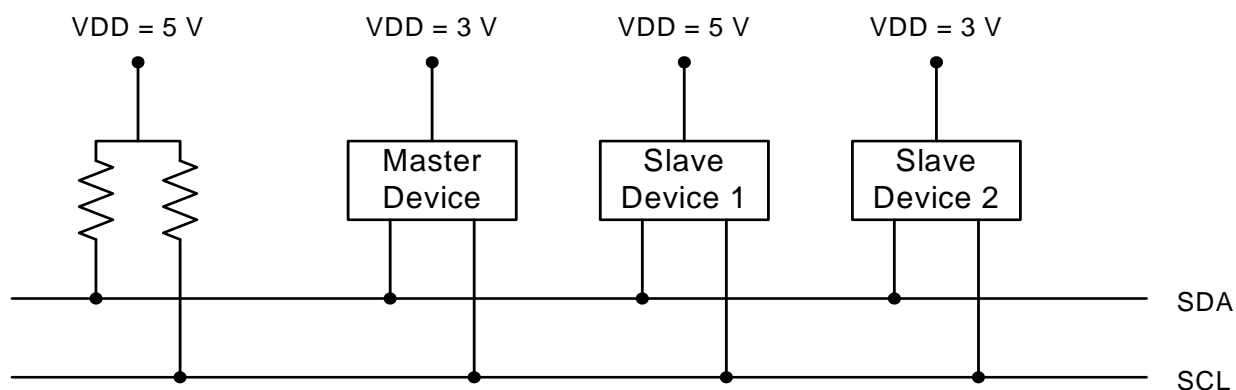
## 14.1. Supporting Documents

It is assumed the reader is familiar with or has access to the following supporting documents:

- The I2C-Bus and How to Use It (including specifications), Philips Semiconductor.
- The I2C-Bus Specification—Version 2.0, Philips Semiconductor.
- System Management Bus Specification—Version 1.1, SBS Implementers Forum.

## 14.2. SMBus Configuration

Figure 14.2 shows a typical SMBus configuration. The SMBus specification allows any recessive voltage between 3.0 V and 5.0 V; different devices on the bus may operate at different voltage levels. The bi-directional SCL (serial clock) and SDA (serial data) lines must be connected to a positive power supply voltage through a pullup resistor or similar circuit. Every device connected to the bus must have an open-drain or open-collector output for both the SCL and SDA lines, so that both are pulled high (recessive state) when the bus is free. The maximum number of devices on the bus is limited only by the requirement that the rise and fall times on the bus not exceed 300 ns and 1000 ns, respectively.



**Figure 14.2. Typical SMBus Configuration**

## 14.3. SMBus Operation

Two types of data transfers are possible: data transfers from a master transmitter to an addressed slave receiver (WRITE), and data transfers from an addressed slave transmitter to a master receiver (READ). The master device initiates both types of data transfers and provides the serial clock pulses on SCL. The SMBus interface may operate as a master or a slave, and multiple master devices on the same bus are supported. If two or more masters attempt to initiate a data transfer simultaneously, an arbitration scheme is employed with a single master always winning the arbitration. Note that it is not necessary to specify one device as the Master in a system; any device who transmits a START and a slave address becomes the master for the duration of that transfer.

A typical SMBus transaction consists of a START condition followed by an address byte (Bits7–1: 7-bit slave address; Bit0: R/W direction bit), one or more bytes of data, and a STOP condition. Each byte that is received (by a master or slave) must be acknowledged (ACK) with a low SDA during a high SCL (see Figure 14.3). If the receiving device does not ACK, the transmitting device will read a NACK (not acknowledge), which is a high SDA during a high SCL.

## 14.3.2. Clock Low Extension

SMBus provides a clock synchronization mechanism, similar to I2C, which allows devices with different speed capabilities to coexist on the bus. A clock-low extension is used during a transfer in order to allow slower slave devices to communicate with faster masters. The slave may temporarily hold the SCL line LOW to extend the clock low period, effectively decreasing the serial clock frequency.

## 14.3.3. SCL Low Timeout

If the SCL line is held low by a slave device on the bus, no further communication is possible. Furthermore, the master cannot force the SCL line high to correct the error condition. To solve this problem, the SMBus protocol specifies that devices participating in a transfer must detect any clock cycle held low longer than 25 ms as a “timeout” condition. Devices that have detected the timeout condition must reset the communication no later than 10 ms after detecting the timeout condition.

When the SMBTOE bit in SMB0CF is set, Timer 3 is used to detect SCL low timeouts. Timer 3 is forced to reload when SCL is high, and allowed to count when SCL is low. With Timer 3 enabled and configured to overflow after 25 ms (and SMBTOE set), the Timer 3 interrupt service routine can be used to reset (disable and re-enable) the SMBus in the event of an SCL low timeout.

## 14.3.4. SCL High (SMBus Free) Timeout

The SMBus specification stipulates that if the SCL and SDA lines remain high for more than 50  $\mu$ s, the bus is designated as free. When the SMBFTE bit in SMB0CF is set, the bus will be considered free if SCL and SDA remain high for more than 10 SMBus clock source periods. If the SMBus is waiting to generate a Master START, the START will be generated following this timeout. Note that a clock source is required for free timeout detection, even in a slave-only implementation.

## SFR Definition 15.2. SBUF0: Serial (UART0) Port Data Buffer

R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	Reset Value
								00000000
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	

SFR Address: 0x99

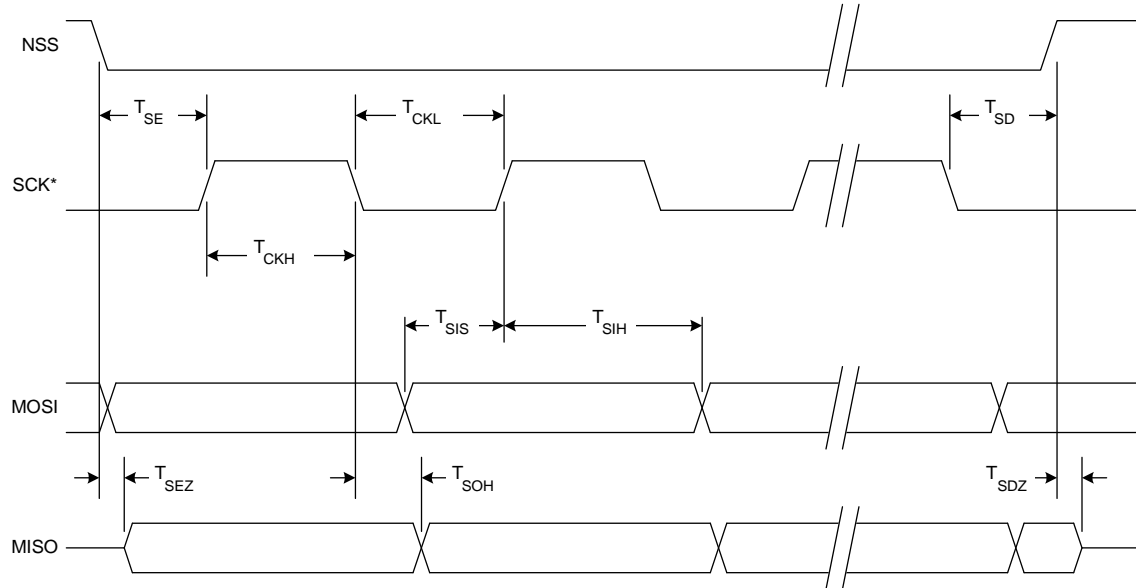
Bits7–0: SBUF0[7:0]: Serial Data Buffer Bits 7–0 (MSB–LSB)  
 This SFR accesses two registers; a transmit shift register and a receive latch register. When data is written to SBUF0, it goes to the transmit shift register and is held for serial transmission. Writing a byte to SBUF0 initiates the transmission. A read of SBUF0 returns the contents of the receive latch.

## 16.6. SPI Special Function Registers

SPI0 is accessed and controlled through four special function registers in the system controller: SPI0CN Control Register, SPI0DAT Data Register, SPI0CFG Configuration Register, and SPI0CKR Clock Rate Register. The four special function registers related to the operation of the SPI0 Bus are described in the following register definitions.

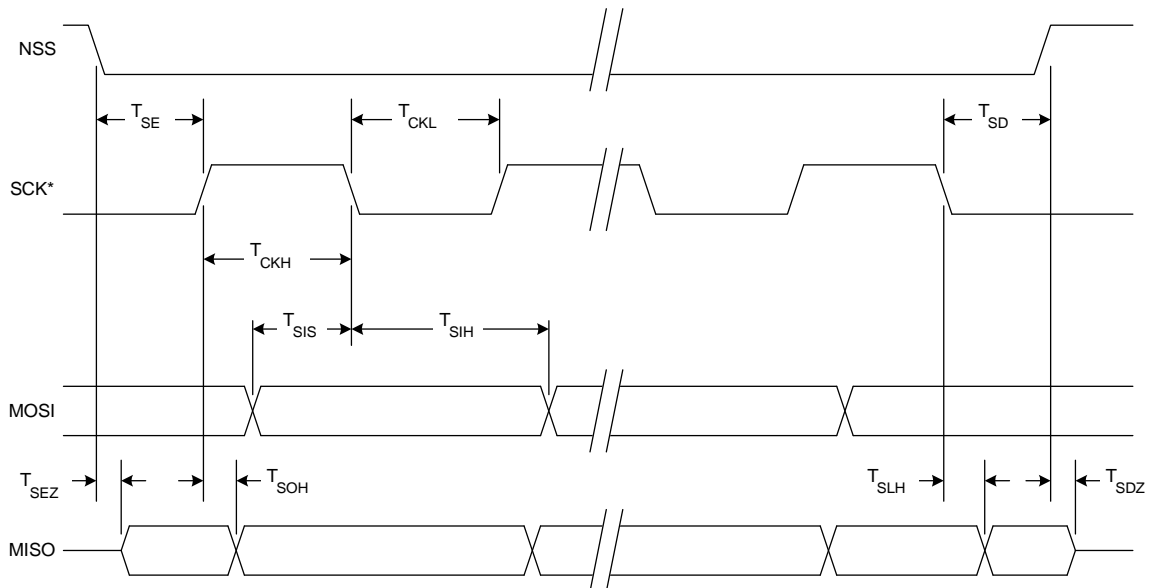
### SFR Definition 16.1. SPI0CFG: SPI0 Configuration

R	R/W	R/W	R/W	R	R	R	R	Reset Value
SPIBSY	MSTEN	CKPHA	CKPOL	SLVSEL	NSSIN	SRMT	RXBMT	00000111
Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
SFR Address: 0xA1								
Bit 7:	SPIBSY: SPI Busy (read only). This bit is set to logic 1 when a SPI transfer is in progress (Master or slave Mode).							
Bit 6:	MSTEN: Master Mode Enable. 0: Disable master mode. Operate in slave mode. 1: Enable master mode. Operate as a master.							
Bit 5:	CKPHA: SPI0 Clock Phase. This bit controls the SPI0 clock phase. 0: Data centered on first edge of SCK period.* 1: Data centered on second edge of SCK period.*							
Bit 4:	CKPOL: SPI0 Clock Polarity. This bit controls the SPI0 clock polarity. 0: SCK line low in idle state. 1: SCK line high in idle state.							
Bit 3:	SLVSEL: Slave Selected Flag (read only). This bit is set to logic 1 whenever the NSS pin is low indicating SPI0 is the selected slave. It is cleared to logic 0 when NSS is high (slave not selected). This bit does not indicate the instantaneous value at the NSS pin, but rather a de-glitched version of the pin input.							
Bit 2:	NSSIN: NSS Instantaneous Pin Input (read only). This bit mimics the instantaneous value that is present on the NSS port pin at the time that the register is read. This input is not de-glitched.							
Bit 1:	SRMT: Shift Register Empty (Valid in Slave Mode, read only). This bit will be set to logic 1 when all data has been transferred in/out of the shift register, and there is no new information available to read from the transmit buffer or write to the receive buffer. It returns to logic 0 when a data byte is transferred to the shift register from the transmit buffer or by a transition on SCK. NOTE: SRMT = 1 when in Master Mode.							
Bit 0:	RXBMT: Receive Buffer Empty (Valid in Slave Mode, read only). This bit will be set to logic 1 when the receive buffer has been read and contains no new information. If there is new information available in the receive buffer that has not been read, this bit will return to logic 0. NOTE: RXBMT = 1 when in Master Mode.							
*Note:	In slave mode, data on MOSI is sampled in the center of each data bit. In master mode, data on MISO is sampled one SYSCLOCK before the end of each data bit, to provide maximum settling time for the slave device. See Table 16.1 for timing parameters.							



\* SCK is shown for CKPOL = 0. SCK is the opposite polarity for CKPOL = 1.

**Figure 16.10. SPI Slave Timing (CKPHA = 0)**



\* SCK is shown for CKPOL = 0. SCK is the opposite polarity for CKPOL = 1.

**Figure 16.11. SPI Slave Timing (CKPHA = 1)**



**Table 16.1. SPI Slave Timing Parameters**

Parameter	Description	Min	Max	Units
<b>Master Mode Timing*</b> (See Figure 16.8 and Figure 16.9)				
$T_{MCKH}$	SCK High Time	$1 \times T_{SYSCLK}$	—	ns
$T_{MCKL}$	SCK Low Time	$1 \times T_{SYSCLK}$	—	ns
$T_{MIS}$	MISO Valid to SCK Shift Edge	$1 \times T_{SYSCLK} + 20$	—	ns
$T_{MIH}$	SCK Shift Edge to MISO Change	0	—	ns
<b>Slave Mode Timing*</b> (See Figure 16.10 and Figure 16.11)				
$T_{SE}$	NSS Falling to First SCK Edge	$2 \times T_{SYSCLK}$	—	ns
$T_{SD}$	Last SCK Edge to NSS Rising	$2 \times T_{SYSCLK}$	—	ns
$T_{SEZ}$	NSS Falling to MISO Valid	—	$4 \times T_{SYSCLK}$	ns
$T_{SDZ}$	NSS Rising to MISO High-Z	—	$4 \times T_{SYSCLK}$	ns
$T_{CKH}$	SCK High Time	$5 \times T_{SYSCLK}$	—	ns
$T_{CKL}$	SCK Low Time	$5 \times T_{SYSCLK}$	—	ns
$T_{SIS}$	MOSI Valid to SCK Sample Edge	$2 \times T_{SYSCLK}$	—	ns
$T_{SIH}$	SCK Sample Edge to MOSI Change	$2 \times T_{SYSCLK}$	—	ns
$T_{SOH}$	SCK Shift Edge to MISO Change	—	$4 \times T_{SYSCLK}$	ns
$T_{SLH}$	Last SCK Edge to MISO Change (CKPHA = 1 ONLY)	$6 \times T_{SYSCLK}$	$8 \times T_{SYSCLK}$	ns
<b>*Note:</b> $T_{SYSCLK}$ is equal to one period of the device system clock (SYSCLK).				

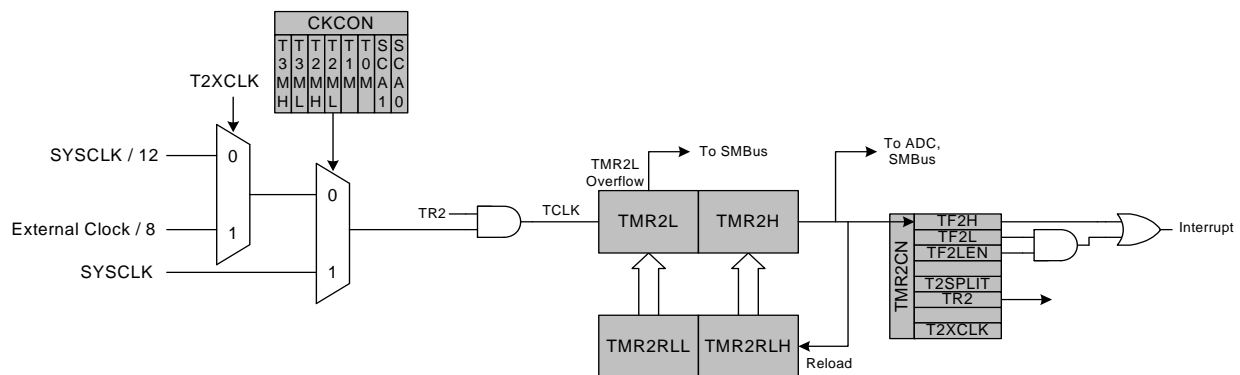
## 17.2. Timer 2

Timer 2 is a 16-bit timer formed by two 8-bit SFRs: TMR2L (low byte) and TMR2H (high byte). Timer 2 may operate in 16-bit auto-reload mode or (split) 8-bit auto-reload mode. The T2SPLIT bit (TMR2CN.3) defines the Timer 2 operation mode.

Timer 2 may be clocked by the system clock, the system clock divided by 12, or the external oscillator source divided by 8. The external clock mode is ideal for real-time clock (RTC) functionality, where the internal oscillator drives the system clock while Timer 2 (and/or the PCA) is clocked by an external precision oscillator. Note that the external oscillator source divided by 8 is synchronized with the system clock.

### 17.2.1. 16-bit Timer with Auto-Reload

When T2SPLIT (TMR2CN.3) is zero, Timer 2 operates as a 16-bit timer with auto-reload. Timer 2 can be clocked by SYSCLK, SYSCLK divided by 12, or the external oscillator clock source divided by 8. As the 16-bit timer register increments and overflows from 0xFFFF to 0x0000, the 16-bit value in the Timer 2 reload registers (TMR2RLH and TMR2RLL) is loaded into the Timer 2 register as shown in Figure 17.4, and the Timer 2 High Byte Overflow Flag (TMR2CN.7) is set. If Timer 2 interrupts are enabled (if IE.5 is set), an interrupt will be generated on each Timer 2 overflow. Additionally, if Timer 2 interrupts are enabled and the TF2LEN bit is set (TMR2CN.5), an interrupt will be generated each time the lower 8 bits (TMR2L) overflow from 0xFF to 0x00.



**Figure 17.4. Timer 2 16-Bit Mode Block Diagram**

## 18. Programmable Counter Array

The Programmable Counter Array (PCA0) provides enhanced timer functionality while requiring less CPU intervention than the standard 8051 counter/timers. The PCA consists of a dedicated 16-bit counter/timer and five 16-bit capture/compare modules. Each capture/compare module has its own associated I/O line (CEXn) which is routed through the Crossbar to Port I/O when enabled (See **Section “13.1. Priority Crossbar Decoder”** on page 131 for details on configuring the Crossbar). The counter/timer is driven by a programmable timebase that can select between six sources: system clock, system clock divided by four, system clock divided by twelve, the external oscillator clock source divided by 8, Timer 0 overflow, or an external clock signal on the ECI input pin. Each capture/compare module may be configured to operate independently in one of six modes: Edge-Triggered Capture, Software Timer, High-Speed Output, Frequency Output, 8-Bit PWM, or 16-Bit PWM (each mode is described in **Section “18.2. Capture/Compare Modules”** on page 205). The external oscillator clock option is ideal for real-time clock (RTC) functionality, allowing the PCA to be clocked by a precision external oscillator while the internal oscillator drives the system clock. The PCA is configured and controlled through the system controller's Special Function Registers. The PCA block diagram is shown in Figure 18.1.

**Important Note:** The PCA Module 4 may be used as a watchdog timer (WDT), and is enabled in this mode following a system reset. Access to certain PCA registers is restricted while WDT mode is enabled. See **Section 18.3** for details.

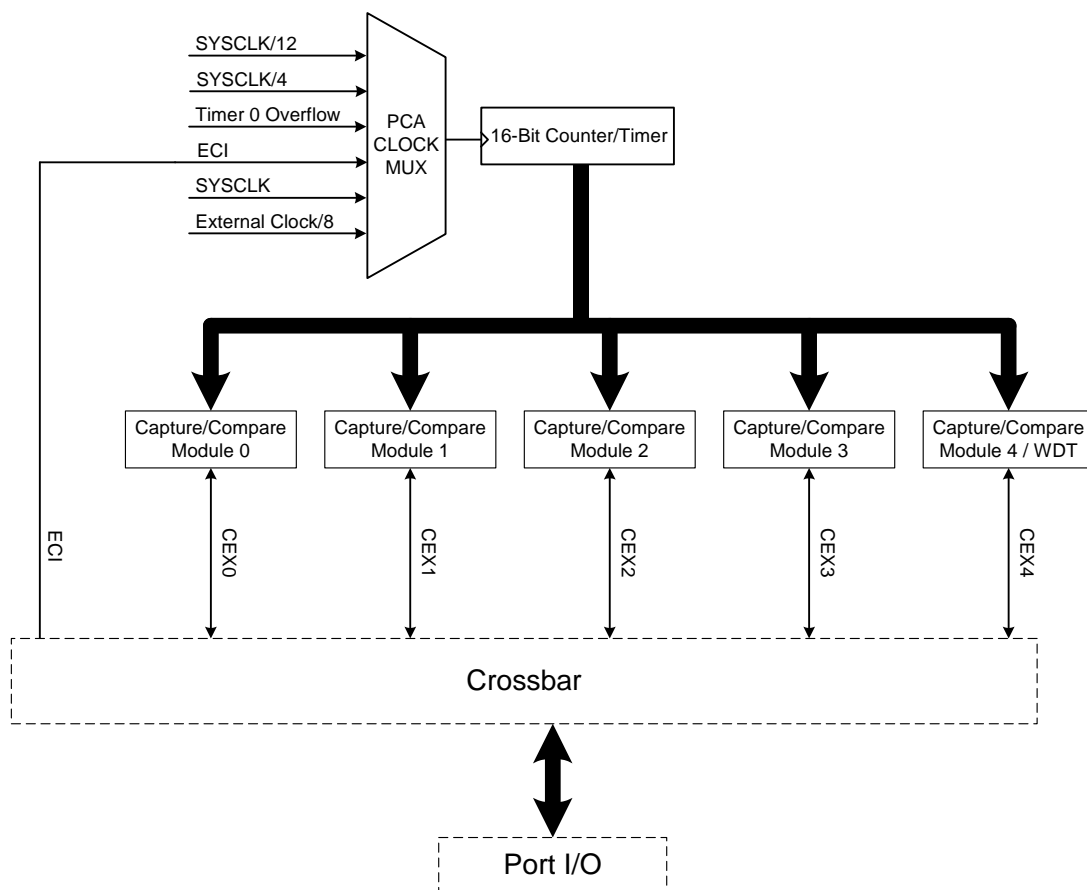


Figure 18.1. PCA Block Diagram

## 18.2.5. 8-Bit Pulse Width Modulator Mode

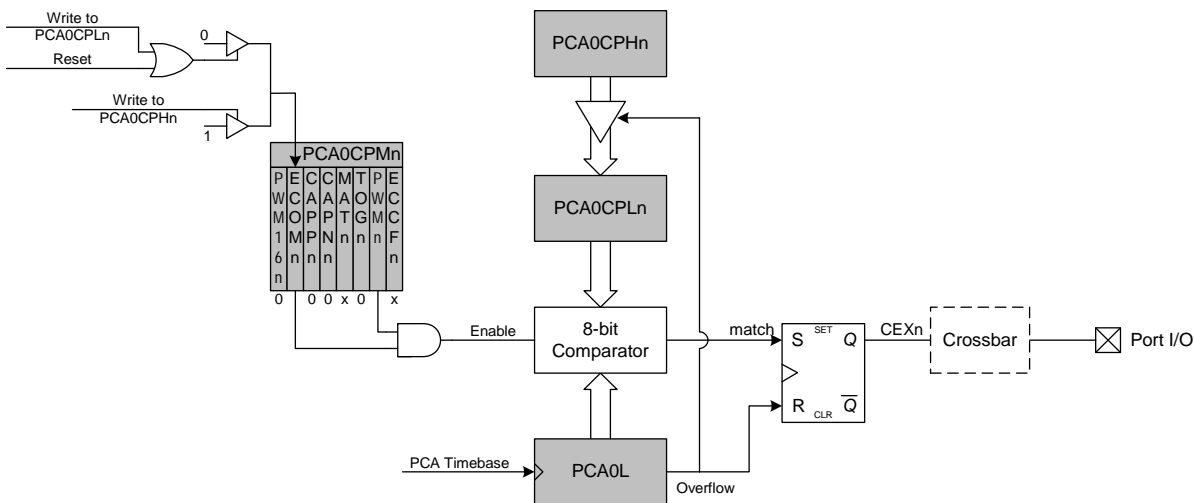
Each module can be used independently to generate a pulse width modulated (PWM) output on its associated CEXn pin. The frequency of the output is dependent on the timebase for the PCA counter/timer. The duty cycle of the PWM output signal is varied using the module's PCA0CPLn capture/compare register. When the value in the low byte of the PCA counter/timer (PCA0L) is equal to the value in PCA0CPLn, the output on the CEXn pin will be set. When the count value in PCA0L overflows, the CEXn output will be reset (see Figure 18.8). Also, when the counter/timer low byte (PCA0L) overflows from 0xFF to 0x00, PCA0CPLn is reloaded automatically with the value stored in the module's capture/compare high byte (PCA0CPHn) without software intervention. Setting the ECOMn and PWMn bits in the PCA0CPMn register enables 8-Bit Pulse Width Modulator mode. The duty cycle for 8-Bit PWM Mode is given by Equation 18.2.

**Important Note About Capture/Compare Registers:** When writing a 16-bit value to the PCA0 Capture/Compare registers, the low byte should always be written first. Writing to PCA0CPLn clears the ECOMn bit to '0'; writing to PCA0CPHn sets ECOMn to '1'.

### Equation 18.2. 8-Bit PWM Duty Cycle

$$DutyCycle = \frac{(256 - PCA0CPHn)}{256}$$

Using Equation 18.2, the largest duty cycle is 100% (PCA0CPHn = 0), and the smallest duty cycle is 0.39% (PCA0CPHn = 0xFF). A 0% duty cycle may be generated by clearing the ECOMn bit to '0'.



**Figure 18.8. PCA 8-Bit PWM Mode Diagram**

Note that the 8-bit offset held in PCA0CPH4 is compared to the upper byte of the 16-bit PCA counter. This offset value is the number of PCA0L overflows before a reset. Up to 256 PCA clocks may pass before the first PCA0L overflow occurs, depending on the value of the PCA0L when the update is performed. The total offset is then given (in PCA clocks) by Equation 18.4, where PCA0L is the value of the PCA0L register at the time of the update.

## Equation 18.4. Watchdog Timer Offset in PCA Clocks

$$\text{Offset} = (256 \times \text{PCA0CPL4}) + (256 - \text{PCA0L})$$

The WDT reset is generated when PCA0L overflows while there is a match between PCA0CPH4 and PCA0H. Software may force a WDT reset by writing a '1' to the CCF4 flag (PCA0CN.4) while the WDT is enabled.

### 18.3.2. Watchdog Timer Usage

To configure the WDT, perform the following tasks:

- Disable the WDT by writing a '0' to the WDTE bit.
- Select the desired PCA clock source (with the CPS2-CPS0 bits).
- Load PCA0CPL4 with the desired WDT update offset value.
- Configure the PCA Idle mode (set CIDL if the WDT should be suspended while the CPU is in Idle mode).
- Enable the WDT by setting the WDTE bit to '1'.
- Write a value to PCA0CPH4 to reload the WDT.

The PCA clock source and Idle mode select cannot be changed while the WDT is enabled. The watchdog timer is enabled by setting the WDTE or WDLCK bits in the PCA0MD register. When WDLCK is set, the WDT cannot be disabled until the next system reset. If WDLCK is not set, the WDT is disabled by clearing the WDTE bit.

The WDT is enabled following any reset. The PCA0 counter clock defaults to the system clock divided by 12, PCA0L defaults to 0x00, and PCA0CPL4 defaults to 0x00. Using Equation 18.4, this results in a WDT timeout interval of 256 system clock cycles. Table 18.3 lists some example timeout intervals for typical system clocks.