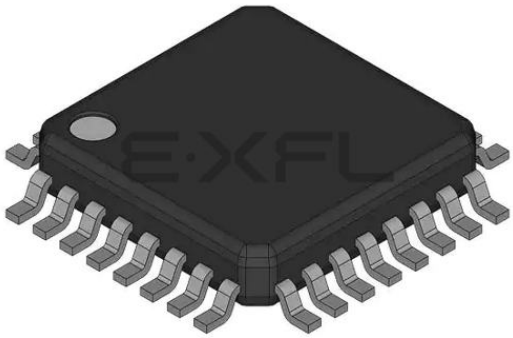

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Number of I/O	25
Program Memory Size	16KB (16K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	1K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	-
Oscillator Type	Internal
Operating Temperature	0°C ~ 70°C (TA)
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MC9S08RC8/16/32/60
MC9S08RD8/16/32/60
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Data Sheet

HCS08
Microcontrollers

MC9S08RG60/D
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Chapter 2
Pins and Connections

Chapter 3
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Table 2-1. Pin Sharing References

Port Pins	Alternate Function	Reference[1]
PTA7–PTA0	KBI1P7–KBI1P0	Chapter 9, “Keyboard Interrupt (S08KBIV1)”
PTB7	TPM1CH1	Chapter 10, “Timer/PWM Module (S08TPMV1)”
PTB6–PTB2	—	Chapter 6, “Parallel Input/Output”
PTB1 PTB0	RxD1 TxD1	Chapter 11, “Serial Communications Interface (S08SCIV1)”
PTC7 PTC6 PTC5 PTC4	SS1 SPSCK1 MISO1 MOSI1	Chapter 13, “Serial Peripheral Interface (S08SPIV3)”
PTC3–PTC0	KBI2P3–KBI2P0	Chapter 9, “Keyboard Interrupt (S08KBIV1)”
PTD6	TPM1CH0	Chapter 10, “Timer/PWM Module (S08TPMV1)”
PTD5 PTD4	ACMP1+ ACMP1–	Chapter 14, “Analog Comparator (S08ACMPV1)”
PTD2	IRQ	Chapter 5, “Resets, Interrupts, and System Configuration”
PTD1	RESET	
PTD0	BKGD/MS	
PTE7–PTE0	—	Chapter 6, “Parallel Input/Output”

1. See this chapter for information about modules that share these pins.

When an on-chip peripheral system is controlling a pin, data direction control bits still determine what is read from port data registers even though the peripheral module controls the pin direction by controlling the enable for the pin’s output buffer. See the Chapter 6, “Parallel Input/Output,” for more details.

Pullup enable bits for each input pin control whether on-chip pullup devices are enabled whenever the pin is acting as an input even if it is being controlled by an on-chip peripheral module. When the PTA7–PTA4 pins are controlled by the KBI module and are configured for rising-edge/high-level sensitivity, the pullup enable control bits enable pulldown devices rather than pullup devices. Similarly, when PTD2 is configured as the IRQ input and is set to detect rising edges, the pullup enable control bit enables a pulldown device rather than a pullup device.

2.3.7 Signal Properties Summary

Table 2-2 summarizes I/O pin characteristics. These characteristics are determined by the way the common pin interfaces are hardwired to internal circuits.

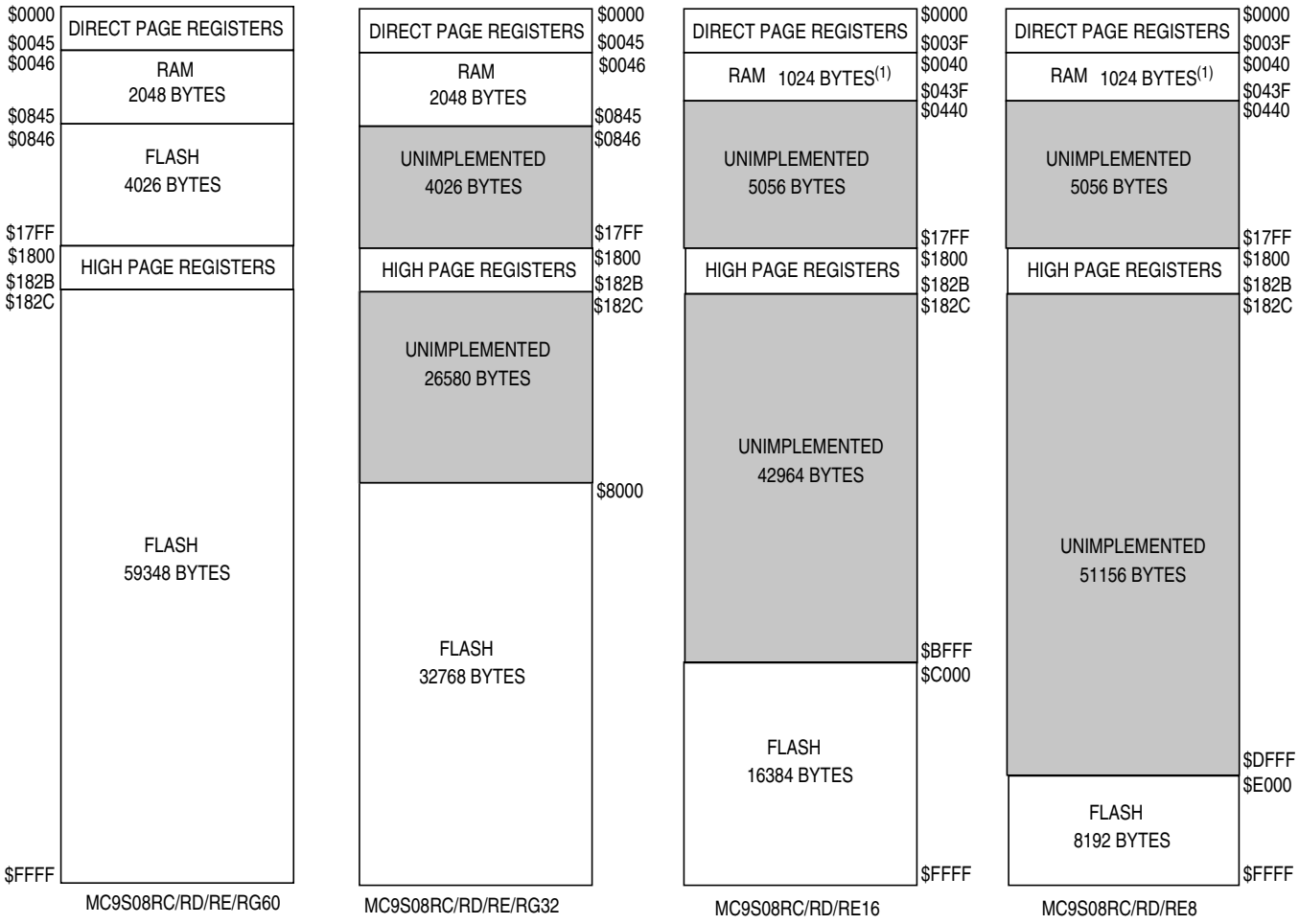


Figure 4-1. MC9S08RC/RD/RE/RG Memory Map

4.1.1 Reset and Interrupt Vector Assignments

Figure 4-2 shows address assignments for reset and interrupt vectors. The vector names shown in this table are the labels used in the Freescale-provided equate file for the MC9S08RC/RD/RE/RG. For more details about resets, interrupts, interrupt priority, and local interrupt mask controls, refer to the Chapter 5, “Resets, Interrupts, and System Configuration."

Figure 4-2. Reset and Interrupt Vectors

Vector Number	Address (High/Low)	Vector	Vector Name
16 through 31	$FFC0:FFC1 ⇕ $FFDE:FFDF	Unused Vector Space (available for user program)	
15	$FFE0:FFE1	SPI[1]	Vspi1
14	$FFE2:FFE3	RTI	Vrti
13	$FFE4:FFE5	KBI2	Vkeyboard2
12	$FFE6:FFE7	KBI1	Vkeyboard1
11	$FFE8:FFE9	ACMP[2]	Vacmp1
10	$FFEA:FFEB	CMT	Vcmt
9	$FFEC:FFED	SCI Transmit[3]	Vsci1tx
8	$FFEE:FFEF	SCI Receive[3]	Vsci1rx
7	$FFF0:FFF1	SCI Error[3]	Vsci1err
6	$FFF2:FFF3	TPM Overflow	Vtpm1ovf
5	$FFF4:FFF5	TPM Channel 1	Vtpm1ch1
4	$FFF6:FFF7	TPM Channel 0	Vtpm1ch0
3	$FFF8:FFF9	IRQ	Virq
2	$FFFA:FFFB	Low Voltage Detect	Vlvd
1	$FFFC:FFFD	SWI	Vswi
0	$FFFE:FFFF	Reset	Vreset

1. The SPI module is not included on the MC9S08RC/RD/RE devices. This vector location is unused for those devices.
2. The analog comparator (ACMP) module is not included on the MC9S08RD devices. This vector location is unused for those devices.
3. The SCI module is not included on the MC9S08RC devices. This vector location is unused for those devices.

4.6.2 FLASH Options Register (FOPT and NVOPT)

During reset, the contents of the nonvolatile location NVOPT are copied from FLASH into FOPT. Bits 5 through 2 are not used and always read 0. This register may be read at any time, but writes have no meaning or effect. To change the value in this register, erase and reprogram the NVOPT location in FLASH memory as usual and then issue a new MCU reset.

	7	6	5	4	3	2	1	0
R	KEYEN	FNORED	0	0	0	0	SEC01	SEC00
W								
Reset	This register is loaded from nonvolatile location NVOPT during reset.							

= Unimplemented or Reserved

Figure 4-6. FLASH Options Register (FOPT)

Table 4-7. FOPT Field Descriptions

Field	Description
7 KEYEN	**Backdoor Key Mechanism Enable** — When this bit is 0, the backdoor key mechanism cannot be used to disengage security. The backdoor key mechanism is accessible only from user (secured) firmware. BDM commands cannot be used to write key comparison values that would unlock the backdoor key. For more detailed information about the backdoor key mechanism, refer to Section 4.5, “Security." 0 No backdoor key access allowed. 1 If user firmware writes an 8-byte value that matches the nonvolatile backdoor key (NVBACKKEY through NVBACKKEY+7 in that order), security is temporarily disengaged until the next MCU reset.
6 FNORED	**Vector Redirection Disable** — When this bit is 1, then vector redirection is disabled. 0 Vector redirection enabled. 1 Vector redirection disabled.
1:0 SEC0[1:0]	**Security State Code** — This 2-bit field determines the security state of the MCU as shown below. When the MCU is secure, the contents of RAM and FLASH memory cannot be accessed by instructions from any unsecured source including the background debug interface. For more detailed information about security, refer to Section 4.5, “Security.” 00 Secure 01 Secure 10 Unsecured 11 Secure SEC01:SEC00 changes to 1:0 after successful backdoor key entry or a successful blank check of FLASH.

4.6.3 FLASH Configuration Register (FCNFG)

Bits 7 through 5 may be read or written at any time. Bits 4 through 0 always read 0 and cannot be written.

The MC9S08RC/RD/RE/RG has these sources for reset:

- Power-on reset (POR)
- Low-voltage detect (LVD)
- Computer operating properly (COP) timer
- Illegal opcode detect
- Illegal address (16K and 8K devices only)
- Background debug forced reset
- The reset pin ($\overline{\text{RESET}}$)

Each of these sources, with the exception of the background debug forced reset, has an associated bit in the system reset status register. Whenever the MCU enters reset, the reset pin is driven low for 34 internal bus cycles where the internal bus frequency is one-half the OSC frequency. After the 34 cycles are completed, the pin is released and will be pulled up by the internal pullup resistor, unless it is held low externally. After the pin is released, it is sampled after another 38 cycles to determine whether the reset pin is the cause of the MCU reset.

5.4 Computer Operating Properly (COP) Watchdog

The COP watchdog is intended to force a system reset when the application software fails to execute as expected. To prevent a system reset from the COP timer (when it is enabled), application software must reset the COP timer periodically. If the application program gets lost and fails to reset the COP before it times out, a system reset is generated to force the system back to a known starting point. The COP watchdog is enabled by the COPE bit in SOPT (see Section 5.8.4, “System Options Register (SOPT),” for additional information). The COP timer is reset by writing any value to the address of SRS. This write does not affect the data in the read-only SRS. Instead, the act of writing to this address is decoded and sends a reset signal to the COP timer.

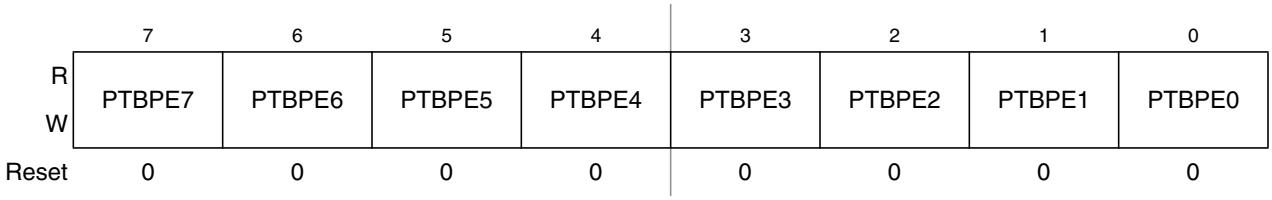
After any reset, the COP timer is enabled. This provides a reliable way to detect code that is not executing as intended. If the COP watchdog is not used in an application, it can be disabled by clearing the COPE bit in the write-once SOPT register. Also, the COPT bit can be used to choose one of two timeout periods (2^{18} or 2^{20} cycles of the bus rate clock). Even if the application will use the reset default settings in COPE and COPT, the user must write to write-once SOPT during reset initialization to lock in the settings. That way, they cannot be changed accidentally if the application program gets lost.

The write to SRS that services (clears) the COP timer must not be placed in an interrupt service routine (ISR) because the ISR could continue to be executed periodically even if the main application program fails.

When the MCU is in active background mode, the COP timer is temporarily disabled.

5.5 Interrupts

Interrupts provide a way to save the current CPU status and registers, execute an interrupt service routine (ISR), and then restore the CPU status so processing resumes where it was before the interrupt. Other than the software interrupt (SWI), which is a program instruction, interrupts are caused by hardware events such

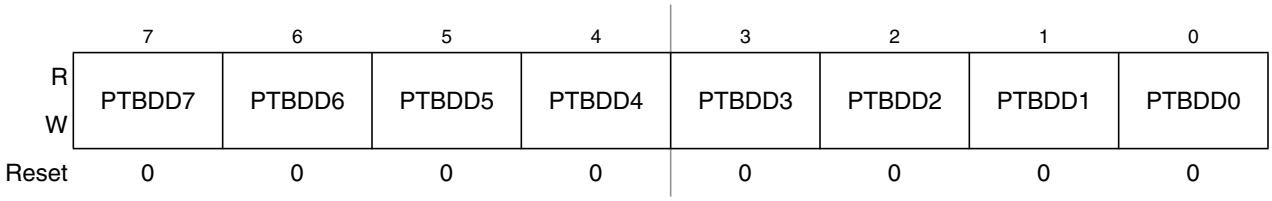


	7	6	5	4	3	2	1	0
R W	PTBPE7	PTBPE6	PTBPE5	PTBPE4	PTBPE3	PTBPE2	PTBPE1	PTBPE0
Reset	0	0	0	0	0	0	0	0

Figure 6-10. Pullup Enable for Port B (PTBPE)

Table 6-5. PTBPE Field Descriptions

Field	Description
7:0 PTBPE[7:0]	**Pullup Enable for Port B Bits** — For port B pins that are inputs, these read/write control bits determine whether internal pullup devices are enabled. For port B pins that are configured as outputs, these bits are ignored and the internal pullup devices are disabled. 0 Internal pullup device disabled. 1 Internal pullup device enabled.



	7	6	5	4	3	2	1	0
R W	PTBDD7	PTBDD6	PTBDD5	PTBDD4	PTBDD3	PTBDD2	PTBDD1	PTBDD0
Reset	0	0	0	0	0	0	0	0

Figure 6-11. Data Direction for Port B (PTBDD)

Table 6-6. PTBDD Field Descriptions

Field	Description
7:0 PTBDD[7:0]	**Data Direction for Port B Bits** — These read/write bits control the direction of port B pins and what is read for PTBD reads. 0 Input (output driver disabled) and reads return the pin value. 1 Output driver enabled for port B bit n and PTBD reads return the contents of PTBDn.

8.5 Functional Description

The CMT module consists of a carrier generator, a modulator, a transmitter output, and control registers. The block diagram is shown in Figure 8-2. When operating in time mode, the user independently defines the high and low times of the carrier signal to determine both period and duty cycle. The carrier generator resolution is 125 ns when operating with an 8 MHz internal bus frequency and the CMTDIV1 and CMTDIV0 bits in the CMTMSC register are both equal to 0. The carrier generator can generate signals with periods between 250 ns (4 MHz) and 127.5 µs (7.84 kHz) in steps of 125 ns. See Table 8-1.

Table 8-1. Clock Divide

Bus Clock (MHz)	CMTDIV1:CMTDIV0	Carrier Generator Resolution (µs)	Min Carrier Generator Period (µs)	Min Modulator Period (µs)
8	0:0	0.125	0.25	1.0
8	0:1	0.25	0.5	2.0
8	1:0	0.5	1.0	4.0
8	1:1	1.0	2.0	8.0

The possible duty cycle options will depend upon the number of counts required to complete the carrier period. For example, a 1.6 MHz signal has a period of 625 ns and will therefore require 5 × 125 ns counts to generate. These counts may be split between high and low times, so the duty cycles available will be 20 percent (one high, four low), 40 percent (two high, three low), 60 percent (three high, two low) and 80 percent (four high, one low).

For lower frequency signals with larger periods, higher resolution (as a percentage of the total period) duty cycles are possible.

When the BASE bit in the CMT modulator status and control register (CMTMSC) is set, the carrier output (f_{CG}) to the modulator is held high continuously to allow for the generation of baseband protocols.

A third mode allows the carrier generator to alternate between two sets of high and low times. When operating in FSK mode, the generator will toggle between the two sets when instructed by the modulator, allowing the user to dynamically switch between two carrier frequencies without CPU intervention.

The modulator provides a simple method to control protocol timing. The modulator has a minimum resolution of 1.0 µs with an 8 MHz internal bus clock. It can count bus clocks (to provide real-time control) or it can count carrier clocks (for self-clocked protocols). See Section 8.5.2, “Modulator," for more details.

The transmitter output block controls the state of the infrared out pin (IRO). The modulator output is gated on to the IRO pin when the modulator/carrier generator is enabled.

A summary of the possible modes is shown in Table 8-2.

Table 8-2. CMT Modes of Operation

Mode	MCGEN Bit[1]	BASE Bit[2]	FSK Bit[2]	EXSPC Bit	Comment
Time	1	0	0	0	f_{CG} controlled by primary high and low registers. f_{CG} transmitted to IRO pin when modulator gate is open.
Baseband	1	1	x	0	f_{CG} is always high. IRO pin high when modulator gate is open.
FSK	1	0	1	0	f_{CG} control alternates between primary high/low registers and secondary high/low registers. f_{CG} transmitted to IRO pin when modulator gate is open.
Extended Space	1	x	x	1	Setting the EXSPC bit causes subsequent modulator cycles to be spaces (modulator out not asserted) for the duration of the modulator period (mark and space times).
IRO Latch	0	x	x	x	IROL bit controls state of IRO pin.

1. To prevent spurious operation, initialize all data and control registers before beginning a transmission (MCGEN=1).
2. These bits are not double buffered and should not be changed during a transmission (while MCGEN=1).

8.5.1 Carrier Generator

The carrier signal is generated by counting a register-selected number of input clocks (125 ns for an 8 MHz bus) for both the carrier high time and the carrier low time. The period is determined by the total number of clocks counted. The duty cycle is determined by the ratio of high time clocks to total clocks counted. The high and low time values are user programmable and are held in two registers.

An alternate set of high/low count values is held in another set of registers to allow the generation of dual frequency FSK (frequency shift keying) protocols without CPU intervention.

NOTE

Only non-zero data values are allowed. The carrier generator will not work if any of the count values are equal to zero.

The MCGEN bit in the CMTMSC register must be set and the BASE bit must be cleared to enable carrier generator clocks. When the BASE bit is set, the carrier output to the modulator is held high continuously. The block diagram is shown in Figure 8-3.

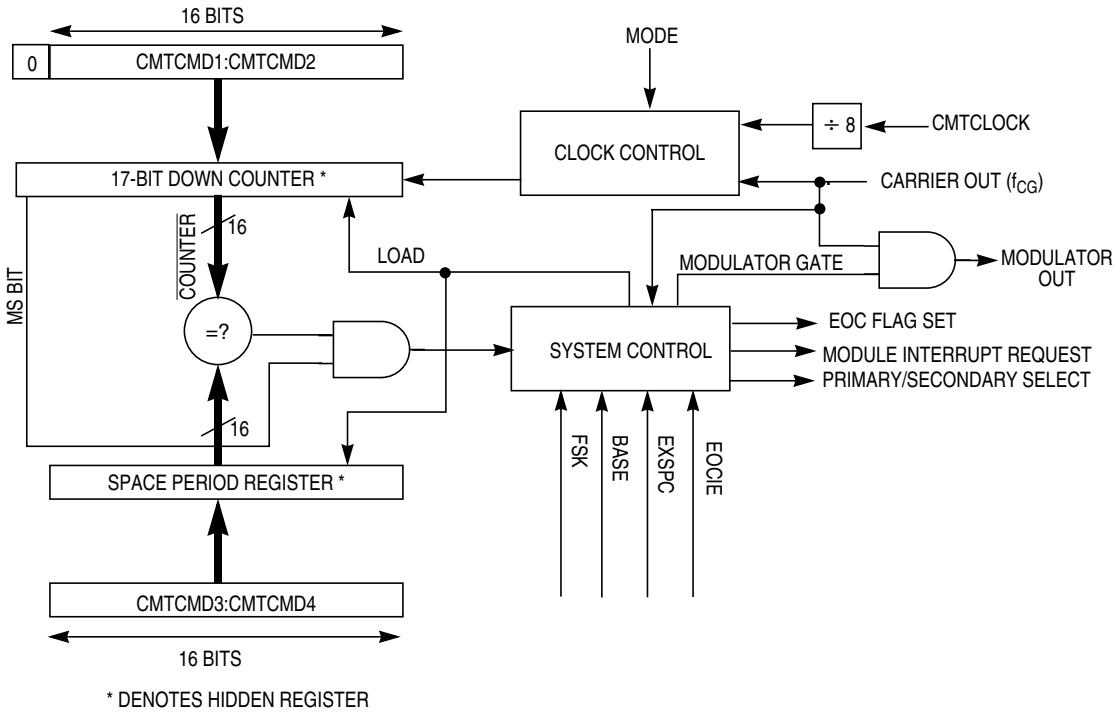


Figure 8-4. Modulator Block Diagram

8.5.2.1 Time Mode

When the modulator operates in time mode (MCGEN bit is set, BASE bit is clear, and FSK bit is clear), the modulation mark period consists of an integer number of CMTCLK ÷ 8 clock periods. The modulation space period consists of zero or an integer number of CMTCLK ÷ 8 clock periods. With an 8 MHz bus and CMTDIV1:CMTDIV0 = 00, the modulator resolution is 1 µs and has a maximum mark and space period of about 65.535 ms each. See Figure 8-5 for an example of the time mode and baseband mode outputs.

The mark and space time equations for time and baseband mode are:

$$t_{mark} = (CMTCMD1{:}CMTCMD2 + 1) \div (f_{CMTCLK} \div 8) \qquad \textit{Eqn. 8-5}$$

$$t_{space} = CMTCMD3{:}CMTCMD4 \div (f_{CMTCLK} \div 8) \qquad \textit{Eqn. 8-6}$$

where CMTCMD1:CMTCMD2 and CMTCMD3:CMTCMD4 are the decimal values of the concatenated registers.

NOTE

If the modulator is disabled while the t_{mark} time is less than the programmed carrier high time ($t_{mark} < CMTCGH1/f_{CMTCLK}$), the modulator can enter into an illegal state and end the curent cycle before the programmed value. Make sure to program t_{mark} greater than the carrier high time to avoid this illegal state.

output compare, and edge-aligned PWM functions. The timer counter modulo registers, TPM1MODH:TPM1MODL, control the modulo value of the counter. (The values $0000 or $FFFF effectively make the counter free running.) Software can read the counter value at any time without affecting the counting sequence. Any write to either byte of the TPM1CNT counter resets the counter regardless of the data value written.

All TPM channels are programmable independently as input capture, output compare, or buffered edge-aligned PWM channels.

10.4 Pin Descriptions

Table 10-2 shows the MCU pins related to the TPM module. When TPM1CH0 is used as an external clock input, the associated TPM channel 0 can not use the pin. (Channel 0 can still be used in output compare mode as a software timer.) When any of the pins associated with the timer is configured as a timer input, a passive pullup can be enabled. After reset, the TPM modules are disabled and all pins default to general-purpose inputs with the passive pullups disabled.

10.4.1 External TPM Clock Sources

When control bits CLKSB:CLKSA in the timer status and control register are set to 1:1, the prescaler and consequently the 16-bit counter for TPM1 are driven by an external clock source connected to the TPM1CH0 pin. A synchronizer is needed between the external clock and the rest of the TPM. This synchronizer is clocked by the bus clock so the frequency of the external source must be less than one-half the frequency of the bus rate clock. The upper frequency limit for this external clock source is specified to be one-fourth the bus frequency to conservatively accommodate duty cycle and phase-locked loop (PLL) or frequency-locked loop (FLL) frequency jitter effects.

When the TPM is using the channel 0 pin for an external clock, the corresponding ELS0B:ELS0A control bits should be set to 0:0 so channel 0 is not trying to use the same pin.

10.4.2 TPM1CHn — TPM1 Channel n I/O Pins

Each TPM channel is associated with an I/O pin on the MCU. The function of this pin depends on the configuration of the channel. In some cases, no pin function is needed so the pin reverts to being controlled by general-purpose I/O controls. When a timer has control of a port pin, the port data and data direction registers do not affect the related pin(s). See the Pins and Connections chapter for additional information about shared pin functions.

10.5 Functional Description

All TPM functions are associated with a main 16-bit counter that allows flexible selection of the clock source and prescale divisor. A 16-bit modulo register also is associated with the main 16-bit counter in the TPM. Each TPM channel is optionally associated with an MCU pin and a maskable interrupt function.

The TPM has center-aligned PWM capabilities controlled by the CPWMS control bit in TPM1SC. When CPWMS is set to 1, timer counter TPM1CNT changes to an up-/down-counter and all channels in the associated TPM act as center-aligned PWM channels. When CPWMS = 0, each channel can

12.2 Register Definition

The SCI has eight 8-bit registers to control baud rate, select SCI options, report SCI status, and for transmit/receive data.

Refer to the direct-page register summary in the Memory chapter of this data sheet for the absolute address assignments for all SCI registers. This section refers to registers and control bits only by their names. A Freescale-provided equate or header file is used to translate these names into the appropriate absolute addresses.

12.2.1 SCI Baud Rate Registers (SCI1BDH, SCI1BHL)

This pair of registers controls the prescale divisor for SCI baud rate generation. To update the 13-bit baud rate setting [SBR12:SBR0], first write to SCI1BDH to buffer the high half of the new value and then write to SCI1BDL. The working value in SCI1BDH does not change until SCI1BDL is written.

SCI1BDL is reset to a non-zero value, so after reset the baud rate generator remains disabled until the first time the receiver or transmitter is enabled (RE or TE bits in SCI1C2 are written to 1).

	7	6	5	4	3	2	1	0
R	0	0	0	SBR12	SBR11	SBR10	SBR9	SBR8
W								
Reset	0	0	0	0	0	0	0	0

= Unimplemented or Reserved

Figure 12-3. SCI Baud Rate Register (SCI1BDH)

Table 12-1. SCI1BDH Register Field Descriptions

Field	Description
4:0 SBR[12:8]	**Baud Rate Modulo Divisor** — These 13 bits are referred to collectively as BR, and they set the modulo divide rate for the SCI baud rate generator. When BR = 0, the SCI baud rate generator is disabled to reduce supply current. When BR = 1 to 8191, the SCI baud rate = BUSCLK/(16×BR). See also BR bits in Table 12-2.

	7	6	5	4	3	2	1	0
R	SBR7	SBR6	SBR5	SBR4	SBR3	SBR2	SBR1	SBR0
W								
Reset	0	0	0	0	0	1	0	0

Figure 12-4. SCI Baud Rate Register (SCI1BDL)

Table 12-2. SCI1BDL Register Field Descriptions

Field	Description
7:0 SBR[7:0]	**Baud Rate Modulo Divisor** — These 13 bits are referred to collectively as BR, and they set the modulo divide rate for the SCI baud rate generator. When BR = 0, the SCI baud rate generator is disabled to reduce supply current. When BR = 1 to 8191, the SCI baud rate = BUSCLK/(16×BR). See also BR bits in Table 12-1.

Table 12-5. SCI1S1 Register Field Descriptions (continued)

Field	Description
5 RDRF	**Receive Data Register Full Flag** — RDRF becomes set when a character transfers from the receive shifter into the receive data register (SCI1D). To clear RDRF, read SCI1S1 with RDRF = 1 and then read the SCI data register (SCI1D). 0 Receive data register empty. 1 Receive data register full.
4 IDLE	**Idle Line Flag** — IDLE is set when the SCI receive line becomes idle for a full character time after a period of activity. When ILT = 0, the receiver starts counting idle bit times after the start bit. So if the receive character is all 1s, these bit times and the stop bit time count toward the full character time of logic high (10 or 11 bit times depending on the M control bit) needed for the receiver to detect an idle line. When ILT = 1, the receiver doesn't start counting idle bit times until after the stop bit. So the stop bit and any logic high bit times at the end of the previous character do not count toward the full character time of logic high needed for the receiver to detect an idle line. To clear IDLE, read SCI1S1 with IDLE = 1 and then read the SCI data register (SCI1D). After IDLE has been cleared, it cannot become set again until after a new character has been received and RDRF has been set. IDLE will get set only once even if the receive line remains idle for an extended period. 0 No idle line detected. 1 Idle line was detected.
3 OR	**Receiver Overrun Flag** — OR is set when a new serial character is ready to be transferred to the receive data register (buffer), but the previously received character has not been read from SCI1D yet. In this case, the new character (and all associated error information) is lost because there is no room to move it into SCI1D. To clear OR, read SCI1S1 with OR = 1 and then read the SCI data register (SCI1D). 0 No overrun. 1 Receive overrun (new SCI data lost).
2 NF	**Noise Flag** — The advanced sampling technique used in the receiver takes seven samples during the start bit and three samples in each data bit and the stop bit. If any of these samples disagrees with the rest of the samples within any bit time in the frame, the flag NF will be set at the same time as the flag RDRF gets set for the character. To clear NF, read SCI1S1 and then read the SCI data register (SCI1D). 0 No noise detected. 1 Noise detected in the received character in SCI1D.
1 FE	**Framing Error Flag** — FE is set at the same time as RDRF when the receiver detects a logic 0 where the stop bit was expected. This suggests the receiver was not properly aligned to a character frame. To clear FE, read SCI1S1 with FE = 1 and then read the SCI data register (SCI1D). 0 No framing error detected. This does not guarantee the framing is correct. 1 Framing error.
0 PF	**Parity Error Flag** — PF is set at the same time as RDRF when parity is enabled (PE = 1) and the parity bit in the received character does not agree with the expected parity value. To clear PF, read SCI1S1 and then read the SCI data register (SCI1D). 0 No parity error. 1 Parity error.

The most common uses of the SPI system include connecting simple shift registers for adding input or output ports or connecting small peripheral devices such as serial A/D or D/A converters. Although Figure 13-2 shows a system where data is exchanged between two MCUs, many practical systems involve simpler connections where data is unidirectionally transferred from the master MCU to a slave or from a slave to the master MCU.

13.2.2 SPI Module Block Diagram

Figure 13-3 is a block diagram of the SPI module. The central element of the SPI is the SPI shift register. Data is written to the double-buffered transmitter (write to SPI1D) and gets transferred to the SPI shift register at the start of a data transfer. After shifting in a byte of data, the data is transferred into the double-buffered receiver where it can be read (read from SPI1D). Pin multiplexing logic controls connections between MCU pins and the SPI module.

When the SPI is configured as a master, the clock output is routed to the SPSCK1 pin, the shifter output is routed to MOSI1, and the shifter input is routed from the MISO1 pin.

When the SPI is configured as a slave, the SPSCK1 pin is routed to the clock input of the SPI, the shifter output is routed to MISO1, and the shifter input is routed from the MOSI1 pin.

In the external SPI system, simply connect all SPSCK pins to each other, all MISO pins together, and all MOSI pins together. Peripheral devices often use slightly different names for these pins.

13.3.3 SPI Interrupts

There are three flag bits, two interrupt mask bits, and one interrupt vector associated with the SPI system. The SPI interrupt enable mask (SPIE) enables interrupts from the SPI receiver full flag (SPRF) and mode fault flag (MODF). The SPI transmit interrupt enable mask (SPTIE) enables interrupts from the SPI transmit buffer empty flag (SPTEF). When one of the flag bits is set, and the associated interrupt mask bit is set, a hardware interrupt request is sent to the CPU. If the interrupt mask bits are cleared, software can poll the associated flag bits instead of using interrupts. The SPI interrupt service routine (ISR) should check the flag bits to determine what event caused the interrupt. The service routine should also clear the flag bit(s) before returning from the ISR (usually near the beginning of the ISR).

13.3.4 Mode Fault Detection

A mode fault occurs and the mode fault flag (MODF) becomes set when a master SPI device detects an error on the $\overline{SS1}$ pin (provided the $\overline{SS1}$ pin is configured as the mode fault input signal). The $\overline{SS1}$ pin is configured to be the mode fault input signal when MSTR = 1, mode fault enable is set (MODFEN = 1), and slave select output enable is clear (SSOE = 0).

The mode fault detection feature can be used in a system where more than one SPI device might become a master at the same time. The error is detected when a master's $\overline{SS1}$ pin is low, indicating that some other SPI device is trying to address this master as if it were a slave. This could indicate a harmful output driver conflict, so the mode fault logic is designed to disable all SPI output drivers when such an error is detected.

When a mode fault is detected, MODF is set and MSTR is cleared to change the SPI configuration back to slave mode. The output drivers on the SPSCK1, MOSI1, and MISO1 (if not bidirectional mode) are disabled.

MODF is cleared by reading it while it is set, then writing to the SPI control register 1 (SPI1C1). User software should verify the error condition has been corrected before changing the SPI back to master mode.

13.4 SPI Registers and Control Bits

The SPI has five 8-bit registers to select SPI options, control baud rate, report SPI status, and for transmit/receive data.

Refer to the direct-page register summary in the Memory chapter of this data sheet for the absolute address assignments for all SPI registers. This section refers to registers and control bits only by their names, and a Freescale-provided equate or header file is used to translate these names into the appropriate absolute addresses.

Appendix A
Electrical Characteristics

A.1 Introduction

This section contains electrical and timing specifications.

A.2 Absolute Maximum Ratings

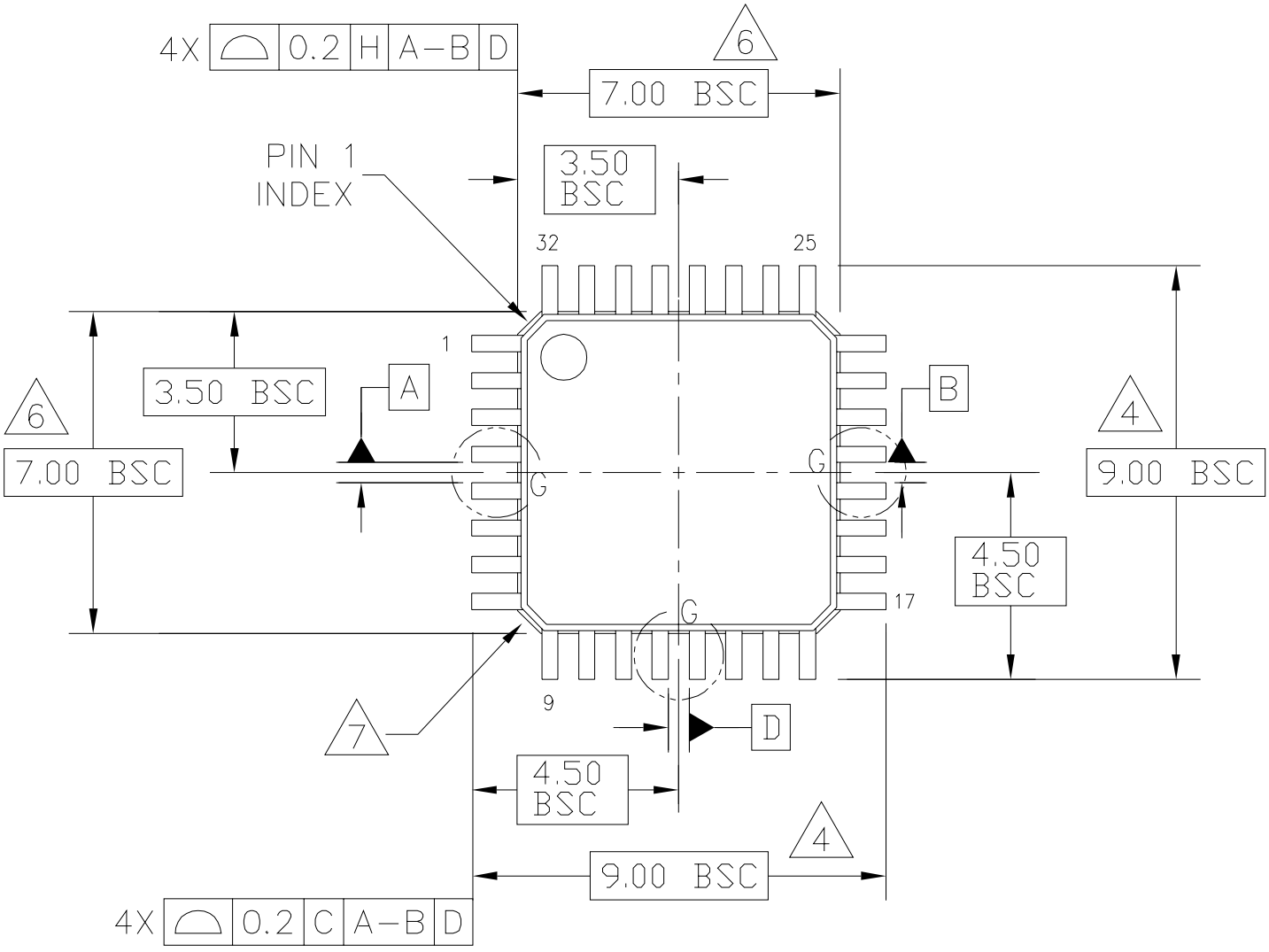
Absolute maximum ratings are stress ratings only, and functional operation at the maxima is not guaranteed. Stress beyond the limits specified in Table A-1 may affect device reliability or cause permanent damage to the device. For functional operating conditions, refer to the remaining tables in this section.

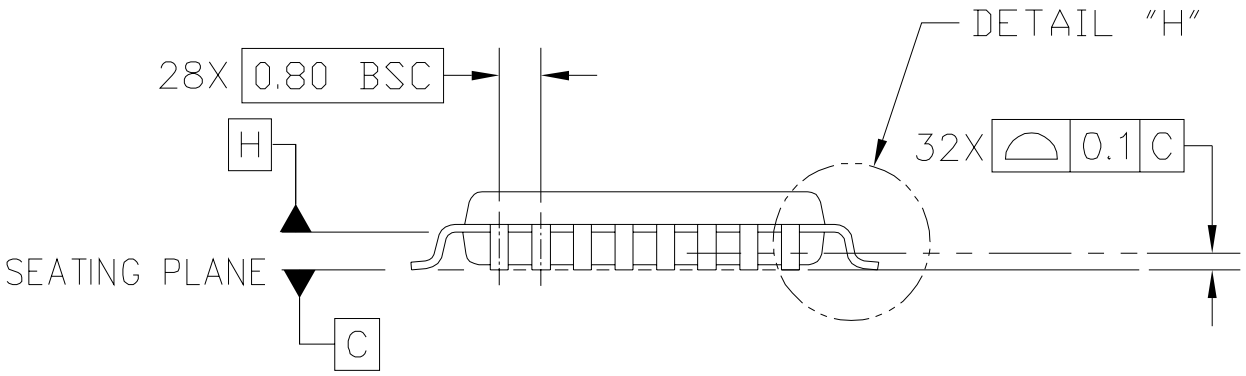
This device contains circuitry protecting against damage due to high static voltage or electrical fields; however, it is advised that normal precautions be taken to avoid application of any voltages higher than maximum-rated voltages to this high-impedance circuit. Reliability of operation is enhanced if unused inputs are tied to an appropriate logic voltage level (for instance, either V_{SS} or V_{DD}) or the programmable pull-up resistor associated with the pin is enabled.

Table A-1. Absolute Maximum Ratings

Rating	Symbol	Value	Unit
Supply voltage	V_{DD}	–0.3 to +3.8	V
Maximum current into V_{DD}	I_{DD}	120	mA
Digital input voltage	V_{In}	–0.3 to V_{DD} + 0.3	V
Instantaneous maximum current Single pin limit (applies to all port pins)(1), (2), (3)	I_D	± 25	mA
Storage temperature range	T_{stg}	–55 to 150	°C

1. Input must be current limited to the value specified. To determine the value of the required current-limiting resistor, calculate resistance values for positive (V_{DD}) and negative (V_{SS}) clamp voltages, then use the larger of the two resistance values.
2. All functional non-supply pins are internally clamped to V_{SS} and V_{DD}.
3. Power supply must maintain regulation within operating V_{DD} range during instantaneous and operating maximum current conditions. If positive injection current ($V_{In} > V_{DD}$) is greater than I_{DD}, the injection current may flow out of V_{DD} and could result in external power supply going out of regulation. Ensure external V_{DD} load will shunt current greater than maximum injection current. This will be the greatest risk when the MCU is not consuming power. Examples are: if no system clock is present, or if the clock rate is very low (which would reduce overall power consumption).





TITLE:

LOW PROFILE QUAD FLAT PACK (LQFP)
32 LEAD, 0.8 PITCH (7 X 7 X 1.4)

CASE NUMBER: 873A-04

STANDARD: JEDEC MS-026 BBA

PACKAGE CODE: 6300