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What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	32MHz
Connectivity	I²C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	18
Program Memory Size	7KB (4K x 14)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	512 x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 12x10b; D/A 1x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Through Hole
Package / Case	20-DIP (0.300", 7.62mm)
Supplier Device Package	20-PDIP
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16lf1708-i-p

5.4 Low-Power Brown-Out Reset (LPBOR)

The Low-Power Brown-Out Reset (LPBOR) is an essential part of the Reset subsystem. Refer to Figure 5-1 to see how the BOR interacts with other modules.

The LPBOR is used to monitor the external VDD pin. When too low of a voltage is detected, the device is held in Reset. When this occurs, a register bit ($\overline{BOR}$) is changed to indicate that a BOR Reset has occurred. The same bit is set for both the BOR and the LPBOR. Refer to Register 5-2.

5.4.1 ENABLING LPBOR

The LPBOR is controlled by the $\overline{LPBOR}$ bit of Configuration Words. When the device is erased, the LPBOR module defaults to disabled.

5.4.1.1 LPBOR Module Output

The output of the LPBOR module is a signal indicating whether or not a Reset is to be asserted. This signal is OR'd together with the Reset signal of the BOR module to provide the generic $\overline{BOR}$ signal, which goes to the PCON register and to the power control block.

5.5 $\overline{MCLR}$

The $\overline{MCLR}$ is an optional external input that can reset the device. The $\overline{MCLR}$ function is controlled by the MCLRE bit of Configuration Words and the LVP bit of Configuration Words (Table 5-2).

TABLE 5-2: $\overline{MCLR}$ CONFIGURATION

MCLRE	LVP	$\overline{MCLR}$
0	0	Disabled
1	0	Enabled
x	1	Enabled

5.5.1 $\overline{MCLR}$ ENABLED

When $\overline{MCLR}$ is enabled and the pin is held low, the device is held in Reset. The $\overline{MCLR}$ pin is connected to VDD through an internal weak pull-up.

The device has a noise filter in the $\overline{MCLR}$ Reset path. The filter will detect and ignore small pulses.

Note: A Reset does not drive the $\overline{MCLR}$ pin low.

5.5.2 $\overline{MCLR}$ DISABLED

When $\overline{MCLR}$ is disabled, the pin functions as a general purpose input and the internal weak pull-up is under software control. See **Section 11.1 "PORTA Registers"** for more information.

5.6 Watchdog Timer (WDT) Reset

The Watchdog Timer generates a Reset if the firmware does not issue a CLRWDT instruction within the time-out period. The $\overline{TO}$ and $\overline{PD}$ bits in the STATUS register are changed to indicate the WDT Reset. See **Section 9.0 "Watchdog Timer (WDT)"** for more information.

5.7 RESET Instruction

A RESET instruction will cause a device Reset. The $\overline{RI}$ bit in the PCON register will be set to '0'. See Table 5-4 for default conditions after a RESET instruction has occurred.

5.8 Stack Overflow/Underflow Reset

The device can reset when the Stack Overflows or Underflows. The STKOVF or STKUNF bits of the PCON register indicate the Reset condition. These Resets are enabled by setting the STVREN bit in Configuration Words. See **3.6.2 "Overflow/Underflow Reset"** for more information.

5.9 Programming Mode Exit

Upon exit of Programming mode, the device will behave as if a POR had just occurred.

5.10 Power-Up Timer

The Power-up Timer optionally delays device execution after a BOR or POR event. This timer is typically used to allow VDD to stabilize before allowing the device to start running.

The Power-up Timer is controlled by the $\overline{PWRTE}$ bit of Configuration Words.

5.11 Start-up Sequence

Upon the release of a POR or BOR, the following must occur before the device will begin executing:

1. Power-up Timer runs to completion (if enabled).
2. Oscillator start-up timer runs to completion (if required for oscillator source).
3. $\overline{MCLR}$ must be released (if enabled).

The total time-out will vary based on oscillator configuration and Power-up Timer configuration. See **Section 6.0 "Oscillator Module (with Fail-Safe Clock Monitor)"** for more information.

The Power-up Timer and oscillator start-up timer run independently of $\overline{MCLR}$ Reset. If $\overline{MCLR}$ is kept low long enough, the Power-up Timer and oscillator start-up timer will expire. Upon bringing $\overline{MCLR}$ high, the device will begin execution after 10 FOSC cycles (see Figure 5-3). This is useful for testing purposes or to synchronize more than one device operating in parallel.

6.6 Register Definitions: Oscillator Control

REGISTER 6-1: OSCCON: OSCILLATOR CONTROL REGISTER

R/W-0/0	R/W-0/0	R/W-1/1	R/W-1/1	R/W-1/1	U-0	R/W-0/0	R/W-0/0
SPLLEN	IRCF<3:0>				—	SCS<1:0>	
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7 **SPLLEN:** Software PLL Enable bit
If PLLEN in Configuration Words = 1:
SPLLEN bit is ignored. 4x PLL is always enabled (subject to oscillator requirements)
If PLLEN in Configuration Words = 0:
1 = 4x PLL Is enabled
0 = 4x PLL is disabled

bit 6-3 **IRCF<3:0>:** Internal Oscillator Frequency Select bits
1111 = 16 MHz HF
1110 = 8 MHz or 32 MHz HF(2)
1101 = 4 MHz HF
1100 = 2 MHz HF
1011 = 1 MHz HF
1010 = 500 kHz HF(1)
1001 = 250 kHz HF(1)
1000 = 125 kHz HF(1)
0111 = 500 kHz MF (default upon Reset)
0110 = 250 kHz MF
0101 = 125 kHz MF
0100 = 62.5 kHz MF
0011 = 31.25 kHz HF(1)
0010 = 31.25 kHz MF
000x = 31 kHz LF

bit 2 **Unimplemented:** Read as '0'

bit 1-0 **SCS<1:0>:** System Clock Select bits
1x = Internal oscillator block
01 = Secondary oscillator
00 = Clock determined by FOSC<2:0> in Configuration Words

Note 1: Duplicate frequency derived from HFINTOSC.
2: 32 MHz when SPLLEN bit is set. Refer to **Section 6.2.2.6 "32 MHz Internal Oscillator Frequency Selection"**.

REGISTER 7-6: PIR2: PERIPHERAL INTERRUPT REQUEST REGISTER 2

R/W-0/0	R/W-0/0	R/W-0/0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
OSFIF	C2IF	C1IF	—	BCL1IF	TMR6IF	TMR4IF	CCP2IF
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7 **OSFIF:** Oscillator Fail Interrupt Flag bit
1 = Interrupt is pending
0 = Interrupt is not pending

bit 6 **C2IF:** Comparator C2 Interrupt Flag bit
1 = Interrupt is pending
0 = Interrupt is not pending

bit 5 **C1IF:** Comparator C1 Interrupt Flag bit
1 = Interrupt is pending
0 = Interrupt is not pending

bit 4 **Unimplemented:** Read as '0'

bit 3 **BCL1IF:** MSSP Bus Collision Interrupt Flag bit
1 = Interrupt is pending
0 = Interrupt is not pending

bit 2 **TMR6IF:** Timer6 to PR6 Interrupt Flag bit
1 = Interrupt is pending
0 = Interrupt is not pending

bit 1 **TMR4IF:** Timer4 to PR4 Interrupt Flag bit
1 = Interrupt is pending
0 = Interrupt is not pending

bit 0 **CCP2IF:** CCP2 Interrupt Flag bit
1 = Interrupt is pending
0 = Interrupt is not pending

Note: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the Global Enable bit, GIE, of the INTCON register. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

8.2 Low-Power Sleep Mode

The PIC16F1704/8 device contains an internal Low Dropout (LDO) voltage regulator, which allows the device I/O pins to operate at voltages up to 5.5V while the internal device logic operates at a lower voltage. The LDO and its associated reference circuitry must remain active when the device is in Sleep mode. The PIC16F1704/8 allows the user to optimize the operating current in Sleep, depending on the application requirements.

A Low-Power Sleep mode can be selected by setting the VREGPM bit of the VREGCON register. With this bit set, the LDO and reference circuitry are placed in a low-power state when the device is in Sleep.

8.2.1 SLEEP CURRENT VS. WAKE-UP TIME

In the default operating mode, the LDO and reference circuitry remain in the normal configuration while in Sleep. The device is able to exit Sleep mode quickly since all circuits remain active. In Low-Power Sleep mode, when waking up from Sleep, an extra delay time is required for these circuits to return to the normal configuration and stabilize.

The Low-Power Sleep mode is beneficial for applications that stay in Sleep mode for long periods of time. The normal mode is beneficial for applications that need to wake from Sleep quickly and frequently.

8.2.2 PERIPHERAL USAGE IN SLEEP

Some peripherals that can operate in Sleep mode will not operate properly with the Low-Power Sleep mode selected. The Low-Power Sleep mode is intended for use only with the following peripherals:

- Brown-Out Reset (BOR)
- Watchdog Timer (WDT)
- External interrupt pin/Interrupt-on-change pins
- Timer1 (with external clock source<100 kHz)

Note:	The PIC16LF1704/8 does not have a configurable Low-Power Sleep mode. PIC16LF1704/8 is an unregulated device and is always in the lowest power state when in Sleep, with no wake-up time penalty. This device has a lower maximum V_{DD} and I/O voltage than the PIC16F1704/8. See **Section 32.0 "Electrical Specifications"** for more information.

11.2 Register Definitions: PORTA

REGISTER 11-1: PORTA: PORTA REGISTER

U-0	U-0	R/W-x/x	R/W-x/x	R-x/x	R/W-x/x	R/W-x/x	R/W-x/x
—	—	RA5	RA4	RA3	RA2	RA1	RA0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-6 **Unimplemented:** Read as '0'

bit 5-0 **RA<5:0>**: PORTA I/O Value bits(1)
1 = Port pin is ≥ VIH
0 = Port pin is ≤ VIL

Note 1: Writes to PORTA are actually written to corresponding LATA register. Reads from PORTA register is return of actual I/O pin values.

REGISTER 11-2: TRISA: PORTA TRI-STATE REGISTER

U-0	U-0	R/W-1/1	R/W-1/1	U-1	R/W-1/1	R/W-1/1	R/W-1/1
—	—	TRISA5	TRISA4	—(1)	TRISA2	TRISA1	TRISA0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-6 **Unimplemented:** Read as '0'

bit 5-4 **TRISA<5:4>:** PORTA Tri-State Control bit
1 = PORTA pin configured as an input (tri-stated)
0 = PORTA pin configured as an output

bit 3 **Unimplemented:** Read as '1'

bit 2-0 **TRISA<2:0>:** PORTA Tri-State Control bit
1 = PORTA pin configured as an input (tri-stated)
0 = PORTA pin configured as an output

Note 1: Unimplemented, read as '1'.

REGISTER 11-20: ANSELC: PORTC ANALOG SELECT REGISTER

R/W-1/1	R/W-1/1	U-0	U-0	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1
ANSC7(2)	ANSC6(2)	ANSC5(3)	ANSC4(3)	ANSC3	ANSC2	ANSC1	ANSC0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **ANSC<7:0>**: Analog Select between Analog or Digital Function on pins RC<7:0>, respectively(1)
0 = Digital I/O. Pin is assigned to port or digital special function.
1 = Analog input. Pin is assigned as analog input(1). Digital input buffer disabled.

Note 1: When setting a pin to an analog input, the corresponding TRIS bit must be set to Input mode in order to allow external control of the voltage on the pin.
2: ANSC<7:6> are available on PIC16(L)F1708 only.
3: ANSC<5:4> are available on PIC16(L)F1704 only.

REGISTER 11-21: WPUC: WEAK PULL-UP PORTC REGISTER

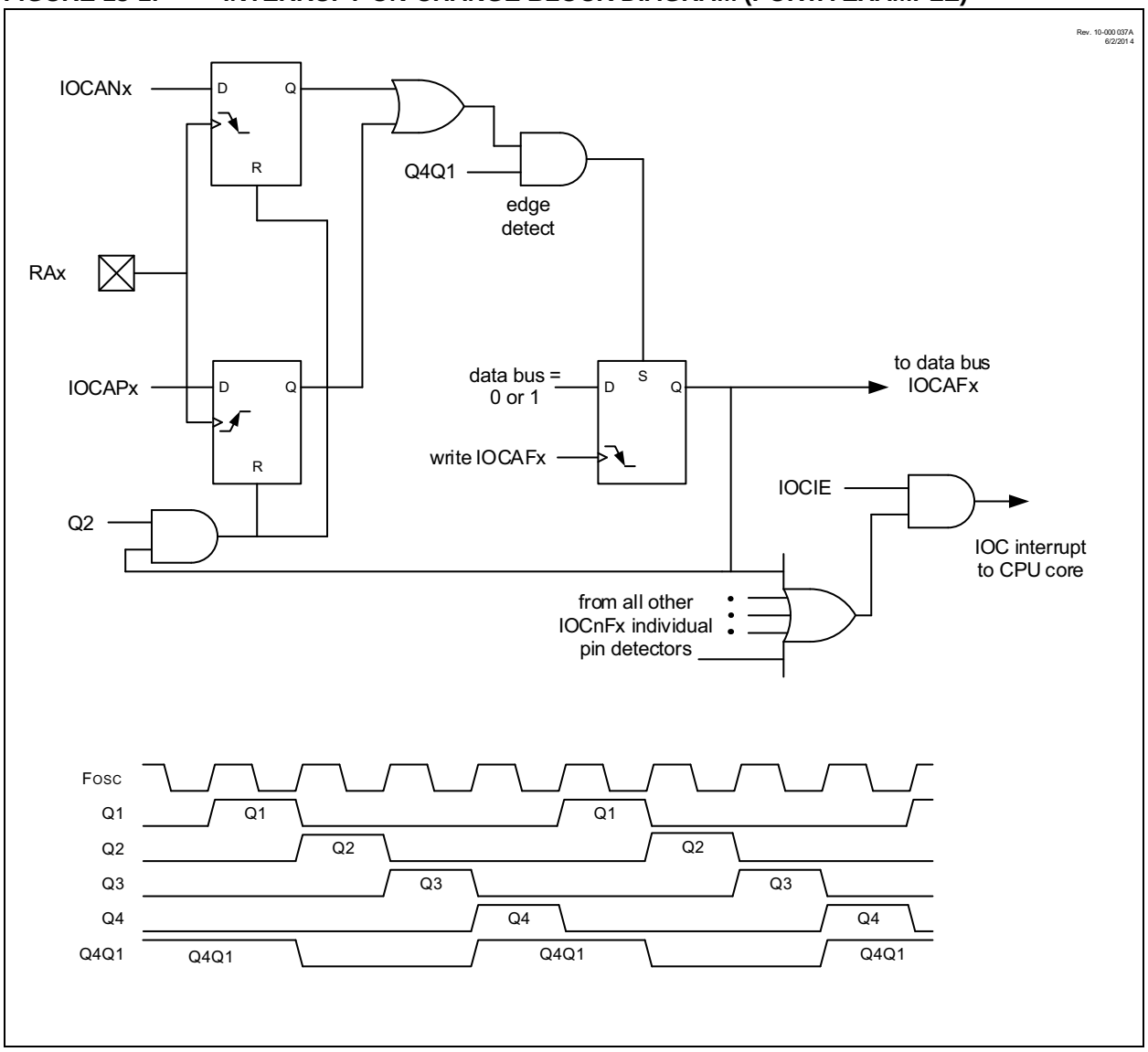
R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1
WPUC7(3)	WPUC6(3)	WPUC5	WPUC4	WPUC3	WPUC2	WPUC1	WPUC0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **WPUC<7:0>**: Weak Pull-up Register bits(3)
1 = Pull-up enabled
0 = Pull-up disabled

Note 1: Global $\overline{\text{WPUEN}}$ bit of the OPTION_REG register must be cleared for individual pull-ups to be enabled.
2: The weak pull-up device is automatically disabled if the pin is configured as an output.
3: WPUC<7:6> are available on PIC16(L)F1708 only.

FIGURE 13-1: INTERRUPT-ON-CHANGE BLOCK DIAGRAM (PORTA EXAMPLE)



14.0 FIXED VOLTAGE REFERENCE (FVR)

The Fixed Voltage Reference, or FVR, is a stable voltage reference, independent of VDD, with 1.024V, 2.048V or 4.096V selectable output levels. The output of the FVR can be configured to supply a reference voltage to the following:

- ADC input channel
- ADC positive reference
- Comparator positive input
- Digital-to-Analog Converter (DAC)

The FVR can be enabled by setting the FVREN bit of the FVRCON register.

14.1 Independent Gain Amplifiers

The output of the FVR supplied to the ADC, Comparators, and DAC is routed through two independent programmable gain amplifiers. Each amplifier can be programmed for a gain of 1x, 2x or 4x, to produce the three possible voltage levels.

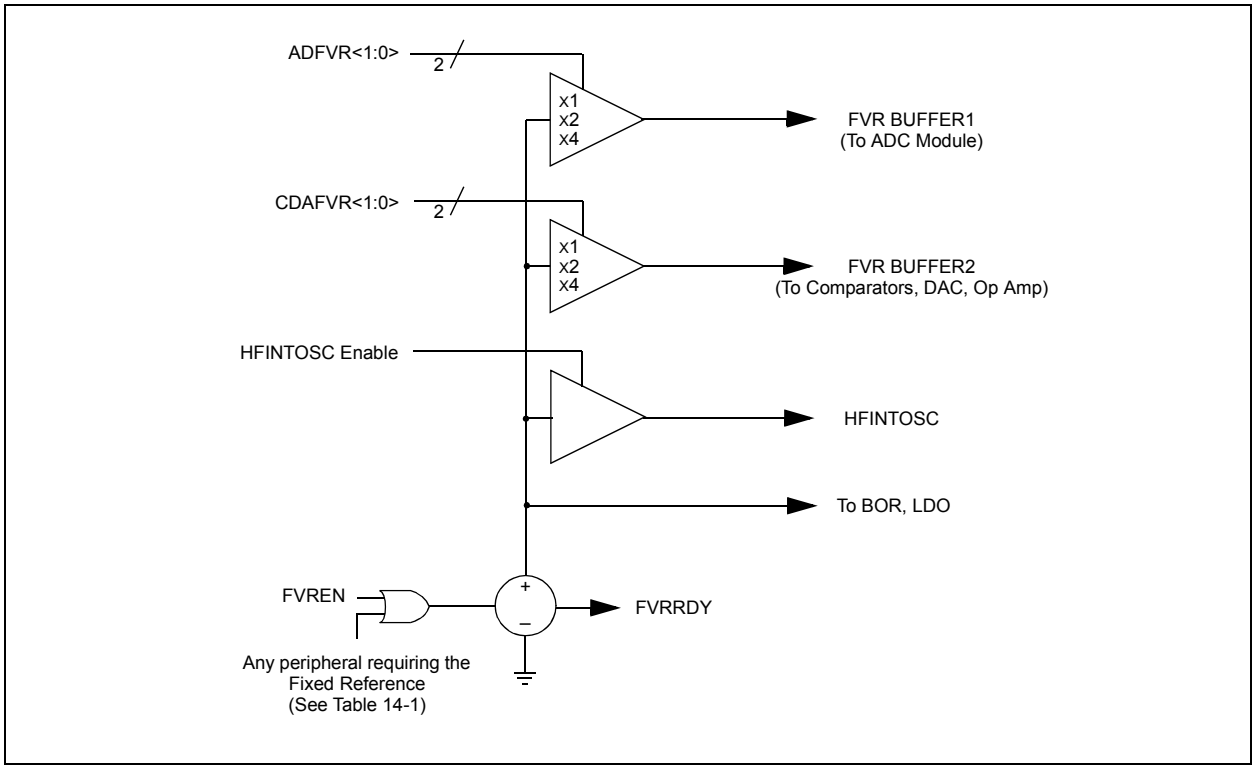
The ADFVR<1:0> bits of the FVRCON register are used to enable and configure the gain amplifier settings for the reference supplied to the ADC module. Reference **Section 20.0 "Analog-to-Digital Converter (ADC) Module"** for additional information.

The CDAFVR<1:0> bits of the FVRCON register are used to enable and configure the gain amplifier settings for the reference supplied to the DAC and comparator module. Reference **Section 22.0 "8-Bit Digital-to-Analog Converter (DAC1) Module"** and **Section 16.0 "Comparator Module"** for additional information.

14.2 FVR Stabilization Period

When the Fixed Voltage Reference module is enabled, it requires time for the reference and amplifier circuits to stabilize. Once the circuits stabilize and are ready for use, the FVRRDY bit of the FVRCON register will be set. See Figure 33-74: Wake from Sleep, VREGPM = 0.

FIGURE 14-1: VOLTAGE REFERENCE BLOCK DIAGRAM



18.5.4 RISING EVENT DEAD-BAND

Rising event dead band delays the turn-on of the primary outputs from when complementary outputs are turned off. The rising event dead-band time starts when the rising_ event output goes true.

See Section 18.5.1, Asynchronous delay chain dead-band delay and Section 18.5.2, Synchronous counter dead-band delay for more information on setting the rising edge dead-band time.

18.5.5 FALLING EVENT DEAD-BAND

Falling event dead band delays the turn-on of complementary outputs from when the primary outputs are turned off. The falling event dead-band time starts when the falling_ event output goes true.

See Section 18.5.1, Asynchronous delay chain dead-band delay and Section 18.5.2, Synchronous counter dead-band delay for more information on setting the rising edge dead-band time.

18.5.6 DEAD-BAND OVERLAP

There are two cases of dead-band overlap:

- Rising-to-falling
- Falling-to-rising

18.5.6.1 Rising-to-Falling Overlap

In this case, the falling event occurs while the rising event dead-band counter is still counting. When this happens, the primary drives are suppressed and the dead-band extends by the falling event dead-band time. At the termination of the extended dead-band time, the complementary drive goes true.

18.5.6.2 Falling-to-Rising Overlap

In this case, the rising event occurs while the falling event dead-band counter is still counting. When this happens, the complementary drive is suppressed and the dead-band extends by the rising event dead-band time. At the termination of the extended dead-band time, the primary drive goes true.

18.6 Blanking Control

Input blanking is a function, whereby, the event inputs can be masked or blanked for a short period of time. This is to prevent electrical transients caused by the turn-on/off of power components from generating a false input event.

The COG contains two blanking counters: one triggered by the rising event and the other triggered by the falling event. The counters are cross coupled with the events they are blanking. The falling event blanking counter is used to blank rising input events and the rising event blanking counter is used to blank falling input events. Once started, blanking extends for the time specified by the corresponding blanking counter.

Blanking is timed by counting COG_clock periods from zero up to the value in the blanking count register. Use Equation 18-1 to calculate blanking times.

18.6.1 FALLING EVENT BLANKING OF RISING EVENT INPUTS

The falling event blanking counter inhibits rising event inputs from triggering a rising event. The falling event blanking time starts when the rising event output drive goes false.

The falling event blanking time is set by the value contained in the COGxBLKF register (Register 18-13). Blanking times are calculated using the formula shown in Equation 18-1.

When the COGxBLKF value is zero, falling event blanking is disabled and the blanking counter output is true, thereby, allowing the event signal to pass straight through to the event trigger circuit.

18.6.2 RISING EVENT BLANKING OF FALLING EVENT INPUTS

The rising event blanking counter inhibits falling event inputs from triggering a falling event. The rising event blanking time starts when the falling event output drive goes false.

The rising event blanking time is set by the value contained in the COGxBLKR register (Register 18-12).

When the COGxBLKR value is zero, rising event blanking is disabled and the blanking counter output is true, thereby, allowing the event signal to pass straight through to the event trigger circuit.

18.6.3 BLANKING TIME UNCERTAINTY

When the rising and falling sources that trigger the blanking counters are asynchronous to the COG_clock, it creates uncertainty in the blanking time. The maximum uncertainty is equal to one COG_clock period. Refer to Equation 18-1 and Example 18-1 for more detail.

18.7 Phase Delay

It is possible to delay the assertion of either or both the rising event and falling events. This is accomplished by placing a non-zero value in COGxPHR or COGxPHF phase-delay count register, respectively (Register 18-14 and Register 18-15). Refer to Figure 18-10 for COG operation with CCP1 and phase delay. The delay from the input rising event signal switching to the actual assertion of the events is calculated the same as the dead-band and blanking delays. Refer to Equation 18-1.

REGISTER 18-6: COGxFSIM: COG FALLING EVENT SOURCE INPUT MODE REGISTER

U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	GxFSIM6	GxFSIM5	GxFSIM4	GxFSIM3	GxFSIM2	GxFSIM1	GxFSIM0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7 **Unimplemented:** Read as '0'

bit 6 **GxFSIM6:** COGx Falling Event Input Source 6 Mode bit
GxFIS6 = 1:
1 = PWM3 output high-to-low transition will cause a falling event after falling event phase delay
0 = PWM3 output low level will cause an immediate falling event
GxFIS6 = 0:
PWM3 output has no effect on falling event

bit 5 **GxFSIM5:** COGx Falling Event Input Source 5 Mode bit
GxFIS5 = 1:
1 = CCP2 output high-to-low transition will cause a falling event after falling event phase delay
0 = CCP2 output low level will cause an immediate falling event
GxFIS5 = 0:
CCP2 output has no effect on falling event

bit 4 **GxFSIM4:** COGx Falling Event Input Source 4 Mode bit
GxFIS4 = 1:
1 = CCP1 high-to-low transition will cause a falling event after falling event phase delay
0 = CCP1 low level will cause an immediate falling event
GxFIS4 = 0:
CCP1 has no effect on falling event

bit 3 **GxFSIM3:** COGx Falling Event Input Source 3 Mode bit
GxFIS3 = 1:
1 = CLC1 output high-to-low transition will cause a falling event after falling event phase delay
0 = CLC1 output low level will cause an immediate falling event
GxFIS3 = 0:
CLC1 output has no effect on falling event

bit 2 **GxFSIM2:** COGx Falling Event Input Source 2 Mode bit
GxFIS2 = 1:
1 = Comparator 2 high-to-low transition will cause a falling event after falling event phase delay
0 = Comparator 2 low level will cause an immediate falling event
GxFIS2 = 0:
Comparator 2 has no effect on falling event

bit 1 **GxFSIM1:** COGx Falling Event Input Source 1 Mode bit
GxFIS1 = 1:
1 = Comparator 1 high-to-low transition will cause a falling event after falling event phase delay
0 = Comparator 1 low level will cause an immediate falling event
GxFIS1 = 0:
Comparator 1 has no effect on falling event

bit 0 **GxFSIM0:** COGx Falling Event Input Source 0 Mode bit
GxFIS0 = 1:
1 = Pin selected with COGxPPS control high-to-low transition will cause a falling event after falling event phase delay
0 = Pin selected with COGxPPS control low level will cause an immediate falling event
GxFIS0 = 0:
Pin selected with COGxPPS control has no effect on falling event

REGISTER 18-9: COGxSTR: COG STEERING CONTROL REGISTER 1

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
GxSDATD	GxSDATC	GxSDATB	GxSDATA	GxSTRD	GxSTRC	GxSTRB	GxSTRA
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7 **GxSDATD:** COGxD Static Output Data bit
1 = COGxD static data is high
0 = COGxD static data is low

bit 6 **GxSDATC:** COGxC Static Output Data bit
1 = COGxC static data is high
0 = COGxC static data is low

bit 5 **GxSDATB:** COGxB Static Output Data bit
1 = COGxB static data is high
0 = COGxB static data is low

bit 4 **GxSDATA:** COGxA Static Output Data bit
1 = COGxA static data is high
0 = COGxA static data is low

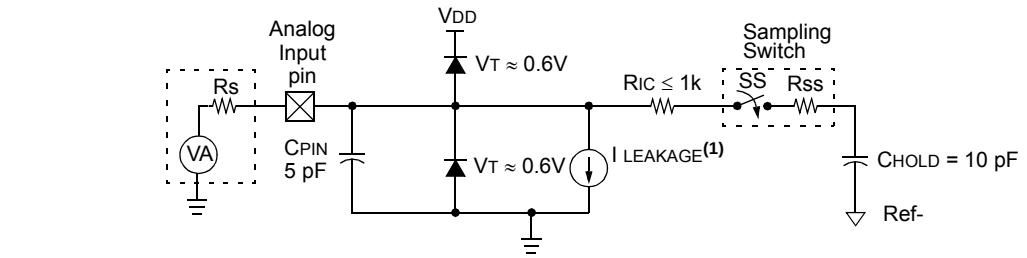
bit 3 **GxSTRD:** COGxD Steering Control bit
1 = COGxD output has the COGxD waveform with polarity control from GxPOLD bit
0 = COGxD output is the static data level determined by the GxSDATD bit

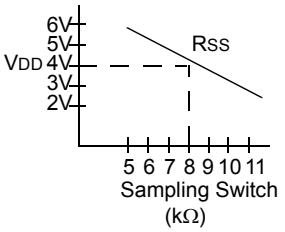
bit 2 **GxSTRC:** COGxC Steering Control bit
1 = COGxC output has the COGxC waveform with polarity control from GxPOLC bit
0 = COGxC output is the static data level determined by the GxSDATC bit

bit 1 **GxSTRB:** COGxB Steering Control bit
1 = COGxB output has the COGxB waveform with polarity control from GxPOLB bit
0 = COGxB output is the static data level determined by the GxSDATB bit

bit 0 **GxSTRA:** COGxA Steering Control bit
1 = COGxA output has the COGxA waveform with polarity control from GxPOLA bit
0 = COGxA output is the static data level determined by the GxSDATA bit

FIGURE 20-4: ANALOG INPUT MODEL



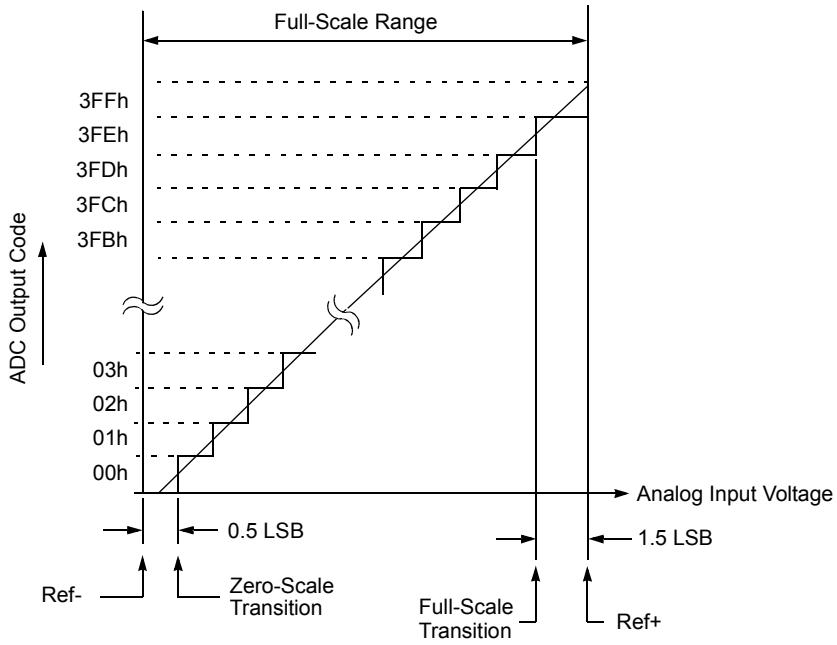


Legend:

CHOLD	=	Sample/Hold Capacitance
CPIN	=	Input Capacitance
I LEAKAGE	=	Leakage current at the pin due to various junctions
RIC	=	Interconnect Resistance
RSS	=	Resistance of Sampling Switch
SS	=	Sampling Switch
VT	=	Threshold Voltage

Note 1: Refer to Table 32-4: I/O Ports (parameter D060).

FIGURE 20-5: ADC TRANSFER FUNCTION



25.3 Timer1 Prescaler

Timer1 has four prescaler options allowing 1, 2, 4 or 8 divisions of the clock input. The T1CKPS bits of the T1CON register control the prescale counter. The prescale counter is not directly readable or writable; however, the prescaler counter is cleared upon a write to TMR1H or TMR1L.

25.4 Timer1 (Secondary) Oscillator

A dedicated low-power 32.768 kHz oscillator circuit is built-in between pins SOSCI (input) and SOSCO (amplifier output). This internal circuit is to be used in conjunction with an external 32.768 kHz crystal.

The oscillator circuit is enabled by setting the T1OSCEN bit of the T1CON register. The oscillator will continue to run during Sleep.

Note:	The oscillator requires a start-up and stabilization time before use. Thus, T1OSCEN should be set and a suitable delay observed prior to using Timer1. A suitable delay similar to the OST delay can be implemented in software by clearing the TMR1IF bit then presetting the TMR1H:TMR1L register pair to FC00h. The TMR1IF flag will be set when 1024 clock cycles have elapsed, thereby indicating that the oscillator is running and reasonably stable.

25.5 Timer1 Operation in Asynchronous Counter Mode

If the control bit $\overline{\text{T1SYNC}}$ of the T1CON register is set, the external clock input is not synchronized. The timer increments asynchronously to the internal phase clocks. If the external clock source is selected then the timer will continue to run during Sleep and can generate an interrupt on overflow, which will wake-up the processor. However, special precautions in software are needed to read/write the timer (see **Section 25.5.1 "Reading and Writing Timer1 in Asynchronous Counter Mode"**).

Note:	When switching from synchronous to asynchronous operation, it is possible to skip an increment. When switching from asynchronous to synchronous operation, it is possible to produce an additional increment.

25.5.1 READING AND WRITING TIMER1 IN ASYNCHRONOUS COUNTER MODE

Reading TMR1H or TMR1L while the timer is running from an external asynchronous clock will ensure a valid read (taken care of in hardware). However, the user should keep in mind that reading the 16-bit timer in two 8-bit values itself, poses certain problems, since the timer may overflow between the reads.

For writes, it is recommended that the user simply stop the timer and write the desired values. A write contention may occur by writing to the timer registers, while the register is incrementing. This may produce an unpredictable value in the TMR1H:TMR1L register pair.

25.6 Timer1 Gate

Timer1 can be configured to count freely or the count can be enabled and disabled using Timer1 gate circuitry. This is also referred to as Timer1 Gate Enable.

Timer1 gate can also be driven by multiple selectable sources.

25.6.1 TIMER1 GATE ENABLE

The Timer1 Gate Enable mode is enabled by setting the TMR1GE bit of the T1GCON register. The polarity of the Timer1 Gate Enable mode is configured using the T1GPOL bit of the T1GCON register.

When Timer1 Gate Enable mode is enabled, Timer1 will increment on the rising edge of the Timer1 clock source. When Timer1 Gate Enable mode is disabled, no incrementing will occur and Timer1 will hold the current count. See Figure 25-3 for timing details.

TABLE 25-3: TIMER1 GATE ENABLE SELECTIONS

T1CLK	T1GPOL	T1G	Timer1 Operation
↑	0	0	Counts
↑	0	1	Holds Count
↑	1	0	Holds Count
↑	1	1	Counts

27.0 CAPTURE/COMPARE/PWM MODULES

The Capture/Compare/PWM module is a peripheral which allows the user to time and control different events, and to generate Pulse-Width Modulation (PWM) signals. In Capture mode, the peripheral allows the timing of the duration of an event. The Compare mode allows the user to trigger an external event when a predetermined amount of time has expired. The PWM mode can generate Pulse-Width Modulated signals of varying frequency and duty cycle.

This family of devices contains two standard Capture/Compare/PWM modules (CCP1 and CCP2).

The Capture and Compare functions are identical for all CCP modules.

Note 1:	In devices with more than one CCP module, it is very important to pay close attention to the register names used. A number placed after the module acronym is used to distinguish between separate modules. For example, the CCP1CON and CCP2CON control the same operational aspects of two completely different CCP modules.
2:	Throughout this section, generic references to a CCP module in any of its operating modes may be interpreted as being equally applicable to CCPx module. Register names, module signals, I/O pins, and bit names may use the generic designator 'x' to indicate the use of a numeral to distinguish a particular module, when required.

27.1 Capture Mode

The Capture mode function described in this section is available and identical for all CCP modules.

Capture mode makes use of the 16-bit Timer1 resource. When an event occurs on the CCPx pin, the 16-bit CCPRxH:CCPRxL register pair captures and stores the 16-bit value of the TMR1H:TMR1L register pair, respectively. An event is defined as one of the following and is configured by the CCPxM<3:0> bits of the CCPxCON register:

- Every falling edge
- Every rising edge
- Every 4th rising edge
- Every 16th rising edge

When a capture is made, the Interrupt Request Flag bit CCPxIF of the PIRx register is set. The interrupt flag must be cleared in software. If another capture occurs before the value in the CCPRxH, CCPRxL register pair is read, the old captured value is overwritten by the new captured value.

Figure 27-1 shows a simplified diagram of the capture operation.

27.1.1 CCP PIN CONFIGURATION

In Capture mode, the CCPx pin should be configured as an input by setting the associated TRIS control bit.

Note:	If the CCPx pin is configured as an output, a write to the port can cause a capture condition.

FIGURE 27-1: CAPTURE MODE OPERATION BLOCK DIAGRAM

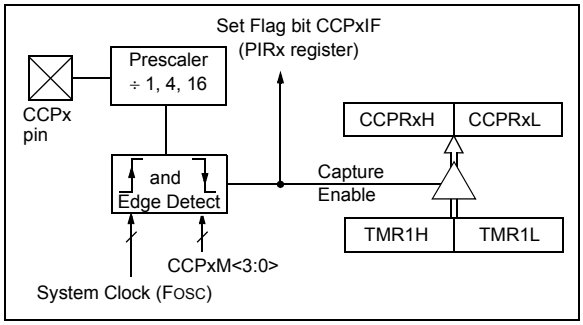
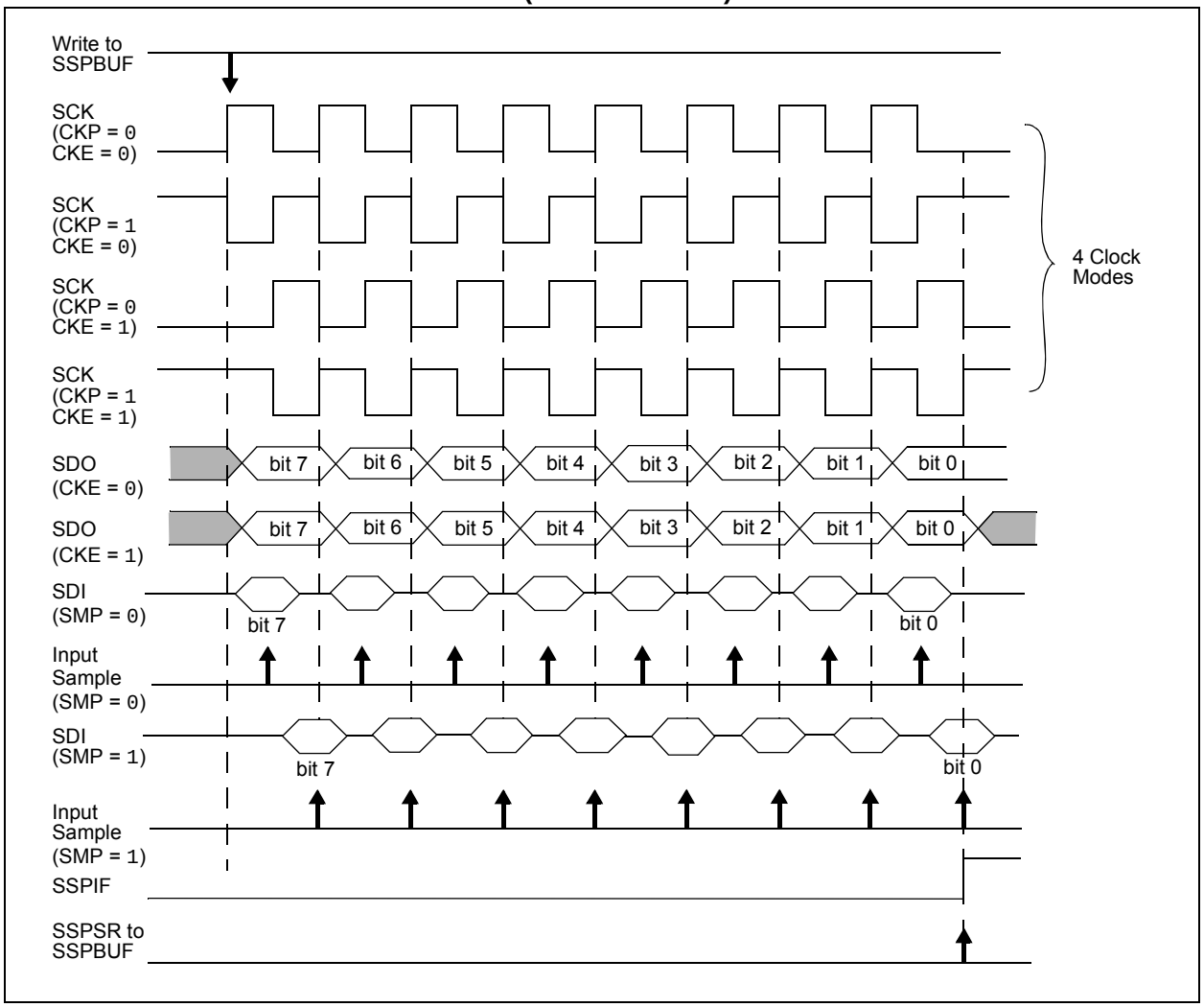


FIGURE 28-6: SPI MODE WAVEFORM (MASTER MODE)



29.5 EUSART Synchronous Mode

Synchronous serial communications are typically used in systems with a single master and one or more slaves. The master device contains the necessary circuitry for baud rate generation and supplies the clock for all devices in the system. Slave devices can take advantage of the master clock by eliminating the internal clock generation circuitry.

There are two signal lines in Synchronous mode: a bidirectional data line and a clock line. Slaves use the external clock supplied by the master to shift the serial data into and out of their respective receive and transmit shift registers. Since the data line is bidirectional, synchronous operation is half-duplex only. Half-duplex refers to the fact that master and slave devices can receive and transmit data but not both simultaneously. The EUSART can operate as either a master or slave device.

Start and Stop bits are not used in synchronous transmissions.

29.5.1 SYNCHRONOUS MASTER MODE

The following bits are used to configure the EUSART for synchronous master operation:

- SYNC = 1
- CSRC = 1
- SREN = 0 (for transmit); SREN = 1 (for receive)
- CREN = 0 (for transmit); CREN = 1 (for receive)
- SPEN = 1

Setting the SYNC bit of the TXSTA register configures the device for synchronous operation. Setting the CSRC bit of the TXSTA register configures the device as a master. Clearing the SREN and CREN bits of the RCSTA register ensures that the device is in the Transmit mode, otherwise the device will be configured to receive. Setting the SPEN bit of the RCSTA register enables the EUSART.

29.5.1.1 Master Clock

Synchronous data transfers use a separate clock line, which is synchronous with the data. A device configured as a master transmits the clock on the TX/CK line. The TX/CK pin output driver is automatically enabled when the EUSART is configured for synchronous transmit or receive operation. Serial data bits change on the leading edge to ensure they are valid at the trailing edge of each clock. One clock cycle is generated for each data bit. Only as many clock cycles are generated as there are data bits.

29.5.1.2 Clock Polarity

A clock polarity option is provided for Microwire compatibility. Clock polarity is selected with the SCKP bit of the BAUDCON register. Setting the SCKP bit sets the clock Idle state as high. When the SCKP bit is set, the data changes on the falling edge of each clock. Clearing the SCKP bit sets the Idle state as low. When the SCKP bit is cleared, the data changes on the rising edge of each clock.

29.5.1.3 Synchronous Master Transmission

Data is transferred out of the device on the RX/DT pin. The RX/DT and TX/CK pin output drivers are automatically enabled when the EUSART is configured for synchronous master transmit operation.

A transmission is initiated by writing a character to the TXREG register. If the TSR still contains all or part of a previous character the new character data is held in the TXREG until the last bit of the previous character has been transmitted. If this is the first character, or the previous character has been completely flushed from the TSR, the data in the TXREG is immediately transferred to the TSR. The transmission of the character commences immediately following the transfer of the data to the TSR from the TXREG.

Each data bit changes on the leading edge of the master clock and remains valid until the subsequent leading clock edge.

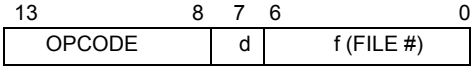
Note:	The TSR register is not mapped in data memory, so it is not available to the user.

29.5.1.4 Synchronous Master Transmission Set-up:

1. Initialize the SPBRGH, SPBRGL register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see **Section 29.4 “EUSART Baud Rate Generator (BRG)”**).
2. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
3. Disable Receive mode by clearing bits SREN and CREN.
4. Enable Transmit mode by setting the TXEN bit.
5. If 9-bit transmission is desired, set the TX9 bit.
6. If interrupts are desired, set the TXIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
7. If 9-bit transmission is selected, the ninth bit should be loaded in the TX9D bit.
8. Start transmission by loading data to the TXREG register.

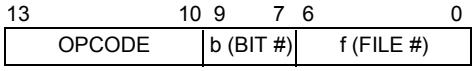
FIGURE 31-1: GENERAL FORMAT FOR INSTRUCTIONS

Byte-oriented file register operations



d = 0 for destination W
d = 1 for destination f
f = 7-bit file register address

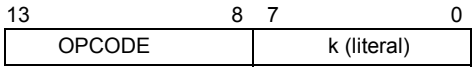
Bit-oriented file register operations



b = 3-bit bit address
f = 7-bit file register address

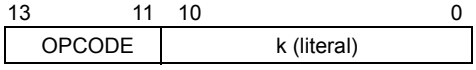
Literal and control operations

General



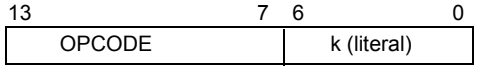
k = 8-bit immediate value

CALL and GOTO instructions only



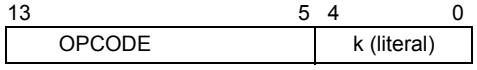
k = 11-bit immediate value

MOVLP instruction only



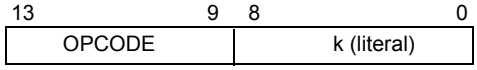
k = 7-bit immediate value

MOVLB instruction only



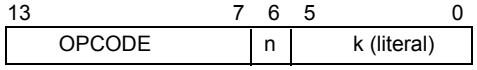
k = 5-bit immediate value

BRA instruction only



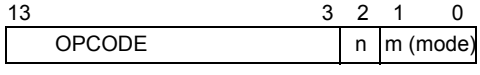
k = 9-bit immediate value

FSR Offset instructions



n = appropriate FSR
k = 6-bit immediate value

FSR Increment instructions



n = appropriate FSR
m = 2-bit mode value

OPCODE only

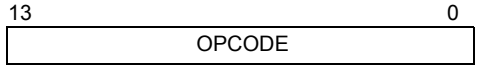


TABLE 32-2: SUPPLY CURRENT (I_{DD})(1,2)

PIC16LF1704/8		Standard Operating Conditions (unless otherwise stated)					
PIC16F1704/8		Standard Operating Conditions (unless otherwise stated)					
Param. No.	Device Characteristics	Min.	Typ.†	Max.	Units	Conditions	
						V_{DD}	Note
D009	LDO Regulator	—	75	—	µA	—	High Power mode, normal operation
		—	15	—	µA	—	Sleep, VREGCON<1> = 0
		—	0.3	—	µA	—	Sleep, VREGCON<1> = 1
D010		—	5.0	12	µA	1.8	F_{OSC} = 32 kHz, LP Oscillator mode **(Note 4)**, -40°C ≤ T_A ≤ +85°C
		—	8.0	18	µA	3.0	
D010		—	16	26	µA	2.3	F_{OSC} = 32 kHz, LP Oscillator mode **(Note 4, Note 5)**, -40°C ≤ T_A ≤ +85°C
		—	18	32	µA	3.0	
		—	22	35	µA	5.0	
D012		—	160	240	µA	1.8	F_{OSC} = 4 MHz, XT Oscillator mode
		—	280	380	µA	3.0	
D012		—	250	320	µA	2.3	F_{OSC} = 4 MHz, XT Oscillator mode **(Note 5)**
		—	320	420	µA	3.0	
		—	400	500	µA	5.0	
D014		—	140	180	µA	1.8	F_{OSC} = 4 MHz, External Clock (ECM), Medium-Power mode
		—	240	300	µA	3.0	
D014		—	210	280	µA	2.3	F_{OSC} = 4 MHz, External Clock (ECM), Medium-Power mode
		—	280	350	µA	3.0	
		—	360	420	µA	5.0	
D015		—	1.9	2.6	mA	3.0	F_{OSC} = 32 MHz, External Clock (ECH), High-Power mode (**Note 5**)
		—	2.4	3.0	mA	3.6	
D015		—	2	2.6	mA	3.0	F_{OSC} = 32 MHz, External Clock (ECH), High-Power mode (**Note 5**)
		—	2.2	2.8	mA	5.0	
D017		—	115	170	µA	1.8	F_{OSC} = 500 kHz, MFINTOSC mode
		—	135	200	µA	3.0	
D017		—	150	200	µA	2.3	F_{OSC} = 500 kHz, MFINTOSC mode
		—	170	220	µA	3.0	
		—	215	280	µA	5.0	
D019		—	0.7	1.1	mA	1.8	F_{OSC} = 16 MHz, HFINTOSC mode
		—	1.2	1.8	mA	3.0	
D019		—	0.9	1.5	mA	2.3	F_{OSC} = 16 MHz, HFINTOSC mode
		—	1.3	1.8	mA	3.0	
		—	1.4	2.0	mA	5.0	

† Data in "Typ." column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The test conditions for all I_{DD} measurements in active operation mode are: OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to V_{DD}; $\overline{MCLR}$ = V_{DD}; WDT disabled.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading and switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption.

3: For RC oscillator configurations, current through R_{EXT} is not included. The current through the resistor can be extended by the formula $I_R = V_{DD}/2R_{EXT}$ (mA) with R_{EXT} in kΩ.

4: FVR and BOR are disabled.

5: 8 MHz clock with 4x PLL enabled.

TABLE 32-2: SUPPLY CURRENT (I_{DD})(1,2) (CONTINUED)

PIC16LF1704/8		Standard Operating Conditions (unless otherwise stated)					
PIC16F1704/8		Standard Operating Conditions (unless otherwise stated)					
Param. No.	Device Characteristics	Min.	Typ.†	Max.	Units	Conditions	
						V_{DD}	Note
D020		—	2.3	3.0	mA	3.0	F_{OSC} = 32 MHz, HFINTOSC mode (**Note 5**)
		—	2.8	3.5	mA	3.6	
D020		—	2.4	3.1	mA	3.0	F_{OSC} = 32 MHz, HFINTOSC mode (**Note 5**)
		—	2.6	3.4	mA	5.0	
D022		—	2	3.0	mA	3.0	F_{OSC} = 32 MHz, HS Oscillator mode (**Note 5**)
		—	2.6	3.5	mA	3.6	
D022		—	2.1	3.0	mA	3.0	F_{OSC} = 32 MHz, HS Oscillator mode (**Note 5**)
		—	3	3.5	mA	5.0	

† Data in "Typ." column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The test conditions for all I_{DD} measurements in active operation mode are: OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to V_{DD}; $\overline{MCLR}$ = V_{DD}; WDT disabled.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading and switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption.

3: For RC oscillator configurations, current through R_{EXT} is not included. The current through the resistor can be extended by the formula $I_R = V_{DD}/2R_{EXT}$ (mA) with R_{EXT} in kΩ.

4: FVR and BOR are disabled.

5: 8 MHz clock with 4x PLL enabled.

TABLE 32-5: MEMORY PROGRAMMING SPECIFICATIONS

Standard Operating Conditions (unless otherwise stated)							
Param. No.	**Sym.**	**Characteristic**	**Min.**	**Typ.†**	**Max.**	**Units**	**Conditions**
		Program Memory Programming Specifications					
D110	VIHH	Voltage on $\overline{MCLR}$/VPP pin	8.0	—	9.0	V	**(Note 1, Note 2)**
D111	IDDP	Supply Current during Programming	—	—	10	mA	
D112	VBE	VDD for Bulk Erase	2.7	—	VDDMAX	V	
D113	VPEW	VDD for Write or Row Erase	VDDMIN	—	VDDMAX	V	
D114	IPPGM	Current on $\overline{MCLR}$/VPP during Erase/Write	—	—	1.0	mA	
D115	IDDPGM	Current on VDD during Erase/ Write	—	—	5.0	mA	
		Program Flash Memory					
D121	EP	Cell Endurance	10K	—	—	E/W	-40°C ≤ TA ≤ +85°C (**Note 3**)
D122	VPR	VDD for Read	VDDMIN	—	VDDMAX	V	
D123	TIW	Self-timed Write Cycle Time	—	2	2.5	ms	
D124	TRETD	Characteristic Retention	—	40	—	Year	Provided no other specifications are violated
D125	EHEFC	High-Endurance Flash Cell	100K	—	—	E/W	0°C ≤ TA ≤ +60°C, Lower byte last 128 addresses

† Data in "Typ." column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Required only if single-supply programming is disabled.

2: The MPLAB ICD2 does not support variable VPP output. Circuitry to limit the ICD2 VPP voltage must be placed between the ICD2 and target system when programming or debugging with the ICD2.

3: Self-write and Block Erase.