



Welcome to E-XFL.COM

#### What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

#### Applications of "<u>Embedded -</u> <u>Microcontrollers</u>"

#### Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	20MHz
Connectivity	I <sup>2</sup> C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	35
Program Memory Size	7KB (4K x 14)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	256 x 8
Voltage - Supply (Vcc/Vdd)	2V ~ 5.5V
Data Converters	A/D 14x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	44-VQFN Exposed Pad
Supplier Device Package	44-QFN (8×8)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16f884t-i-ml

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

IADL	ADLE 3: 40-FIN FUP ALLOCATION TABLE (PICTOF004/08/)									
0/1	40-Pin PDIP	Analog	Comparators	Timers	ECCP	EUSART	MSSP	Interrupt	dn-llud	Basic
RA0	2	AN0/ULPWU	C12IN0-	—	_	_	—	—		—
RA1	3	AN1	C12IN1-	—	_	_	_			_
RA2	4	AN2	C2IN+	—	_	_				VREF-/CVREF
RA3	5	AN3	C1IN+	—	_	_			_	VREF+
RA4	6		C1OUT	T0CKI	—		—	_	_	_
RA5	7	AN4	C2OUT	—	_	_	SS			_
RA6	14	—	_	—	_		_	_	_	OSC2/CLKOUT
RA7	13	—		—	_	_				OSC1/CLKIN
RB0	33	AN12	_	—	—	_	—	IOC/INT	Y	—
RB1	34	AN10	C12IN3-	—	—	_	—	IOC	Y	—
RB2	35	AN8	_	—	_	_	_	IOC	Y	—
RB3	36	AN9	C12IN2-	—	—	_	—	IOC	Y	PGM
RB4	37	AN11	_	—	_	_	_	IOC	Y	—
RB5	38	AN13	_	T1G	_	-	_	IOC	Y	—
RB6	39	—	—	—	_	_	_	IOC	Y	ICSPCLK
RB7	40	—	_	—	_	-	_	IOC	Y	ICSPDAT
RC0	15	—	_	T1OSO/T1CKI	_	_	_	_	_	—
RC1	16	—		T1OSI	CCP2		_			—
RC2	17	—	_	—	CCP1/P1A	_	—		_	—
RC3	18	—	_	—	—		SCK/SCL	_		—
RC4	23	—	_	—	—	-	SDI/SDA		_	—
RC5	24	—		—	—		SDO			—
RC6	25	—	_	—	—	TX/CK	_	_		—
RC7	26	—		—	—	RX/DT	_			—
RD0	19	—		—	—		—	_		—
RD1	20	—		—	—		_	_		—
RD2	21	—	_	—	—	_	—	—	_	_
RD3	22	—	_	—	—	-		—	—	_
RD4	27	—	_	_	—	_	—	_	_	_
RD5	28	—	_	—	P1B	-		—	—	_
RD6	29	—	_	—	P1C	_	—	—	_	_
RD7	30	—		—	P1D		_	_		—
RE0	8	AN5		—	—	_	—	_	_	—
RE1	9	AN6		—	—		_			—
RE2	10	AN7	_	—	—		_	_		—
RE3	1	—	_	—	—	_	_	_	Y(1)	MCLR/VPP
	11	_	—		—	—		—		Vdd
_	32	_	_	_	_	_		_	_	Vdd
	12	_	_		_	_			—	Vss
	31	_	_		_	_		_	_	Vss

## TABLE 3: 40-PIN PDIP ALLOCATION TABLE (PIC16F884/887)

**Note 1:** Pull-up activated only with external MCLR configuration.

IADL	$ABLE 4. \qquad 44-FIN QFN ALLOCATION TABLE (FIC 10F004/007)$									
0/1	44-Pin QFN	Analog	Comparators	Timers	ECCP	EUSART	MSSP	Interrupt	Pull-up	Basic
RA0	19	AN0/ULPWU	C12IN0-	—	_	_	_	_		_
RA1	20	AN1	C12IN1-	—	_	_	_	_		—
RA2	21	AN2	C2IN+	_	—	_	—	—	—	VREF-/CVREF
RA3	22	AN3	C1IN+	—	_	_	_	_		VREF+
RA4	23	_	C1OUT	TOCKI	_	_		_		_
RA5	24	AN4	C2OUT	_	_	_	SS	_	_	_
RA6	33	_	_	—	_	_		_		OSC2/CLKOUT
RA7	32	_	_		_	_	_	_	_	OSC1/CLKIN
RB0	9	AN12	_	_	_	_	_	IOC/INT	Y	_
RB1	10	AN10	C12IN3-	_	_	_	_	IOC	Y	_
RB2	11	AN8	_	_	_	_	_	IOC	Y	_
RB3	12	AN9	C12IN2-	_	_	_	_	IOC	Y	PGM
RB4	14	AN11	_		_	_	_	IOC	Y	
RB5	15	AN13	_	T1G	_	_	_	IOC	Y	_
RB6	16	_	_		_	_	_	IOC	Y	ICSPCLK
RB7	17		_	_	_	_	_	IOC	Y	ICSPDAT
RC0	34		_	T10S0/T1CKI	_	_	_	_		_
RC1	35	_	_	T10SI	CCP2	_	_	_		_
RC2	36	_	_		CCP1/P1A	_	_	_		
RC3	37	_	_		_	_	SCK/SCL	_		_
RC4	42				_	_	SDI/SDA	_		_
RC5	43	_	_		_	_	SDO	_	_	
RC6	44				_	TX/CK	_	_		_
RC7	1	_	_	_	_	RX/DT	_	_	_	_
RD0	38	_	_	—	_	_		_		_
RD1	39	_	_	_	_	_	_	_	_	_
RD2	40	—		—	_	_	_	—		_
RD3	41	_	_		_		_	_	_	
RD4	2	—		—	_	_	_	—		_
RD5	3	_	_		P1B		_	_	_	
RD6	4	—		—	P1C	_	_	—		_
RD7	5	_		—	P1D	_	_	_		_
RE0	25	AN5	_	_	—	_	—	—	—	
RE1	26	AN6		_	_	_	_	_		_
RE2	27	AN7	_	—	_	_		—		_
RE3	18	_		—	_	_	_	_	Y(1)	MCLR/VPP
_	7	_	_	_	_	_	_	_		Vdd
_	8	_	_	_	_	_	_	_	_	Vdd
_	28				_	_	_	_		Vdd
_	6	_	_	_	_	_	_	_	_	Vss
_	30	_	—		_	_	_	_	_	Vss
_	31	_	_	_	_	_	_	_	_	Vss
_	13	_	—		_	_	_	_	_	NC (no connect)
_	29		_	_	_	_	_	_	_	NC (no connect)

TABLE 4: 44-PIN QFN ALLOCATION TABLE (PIC16F884/887)

**Note 1:** Pull-up activated only with external MCLR configuration.

## TABLE 1-2: PIC16F884/887 PINOUT DESCRIPTION (CONTINUED)

Name	Function	Input Type	Output Type	Description
RB4/AN11	RB4	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	AN11	AN	-	A/D Channel 11.
RB5/AN13/T1G	RB5	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	AN13	AN	—	A/D Channel 13.
	T1G	ST	_	Timer1 Gate input.
RB6/ICSPCLK	RB6	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	ICSPCLK	ST	—	Serial Programming Clock.
RB7/ICSPDAT	RB7	TTL	CMOS	General purpose I/O. Individually controlled interrupt-on-change. Individually enabled pull-up.
	ICSPDAT	ST	TTL	ICSP™ Data I/O.
RC0/T1OSO/T1CKI	RC0	ST	CMOS	General purpose I/O.
	T1OSO	—	XTAL	Timer1 oscillator output.
	T1CKI	ST		Timer1 clock input.
RC1/T1OSI/CCP2	RC1	ST	CMOS	General purpose I/O.
	T1OSI	XTAL		Timer1 oscillator input.
	CCP2	ST	CMOS	Capture/Compare/PWM2.
RC2/P1A/CCP1	RC2	ST	CMOS	General purpose I/O.
	P1A	ST	CMOS	PWM output.
	CCP1	—	CMOS	Capture/Compare/PWM1.
RC3/SCK/SCL	RC3	ST	CMOS	General purpose I/O.
	SCK	ST	CMOS	SPI clock.
	SCL	ST	OD	I <sup>∠</sup> C <sup>™</sup> clock.
RC4/SDI/SDA	RC4	ST	CMOS	General purpose I/O.
	SDI	ST		SPI data input.
/	SDA	ST	OD	I <sup>2</sup> C data input/output.
RC5/SDO	RC5	ST	CMOS	General purpose I/O.
	SDO	_	CMOS	SPI data output.
RC6/TX/CK	RC6	SI	CMOS	General purpose I/O.
	IX		CMOS	EUSART asynchronous transmit.
	CK	SI	CMOS	EUSART synchronous clock.
RC7/RX/D1	RC7	SI	CMOS	General purpose I/O.
	RX	51	_	
PD0			CMOS	EUSART synchronous data.
RD0	RD0		CMOS	
RDI	RD1		CMOS	
RD2	RD2		CMOS	
PD/	RD3	11L 771	CMOS	
RD5/P1B		TTI	CMOS	General purpose I/O
	P1R		CMOS	
	RDe	 	CMOS	
	P1C		CMOS	
Locondi AN - Analog				P compatible input or output $P$ $OD = Open Drain$
TTL = TTL compat HV = High Voltage	ible input	ST XTAL	= Schm = Cryst	all

## 2.0 MEMORY ORGANIZATION

## 2.1 Program Memory Organization

The PIC16F882/883/884/886/887 devices have a 13-bit program counter capable of addressing a  $2K \times 14$  (0000h-07FFh) for the PIC16F882,  $4K \times 14$  (0000h-0FFFh) for the PIC16F883/PIC16F884, and  $8K \times 14$  (0000h-1FFFh) for the PIC16F886/PIC16F887 program memory space. Accessing a location above these boundaries will cause a wrap-around within the first  $8K \times 14$  space. The Reset vector is at 0000h and the interrupt vector is at 0004h (see Figures 2-2 and 2-3).

#### FIGURE 2-1: PROGRAM MEMORY MAP AND STACK FOR THE PIC16F882



## FIGURE 2-2:

#### PROGRAM MEMORY MAP AND STACK FOR THE PIC16F883/PIC16F884



## FIGURE 2-3:

#### PROGRAM MEMORY MAP AND STACK FOR THE PIC16F886/PIC16F887



### 2.2.2.5 PIE2 Register

The PIE2 register contains the interrupt enable bits, as shown in Register 2-5.

**Note:** Bit PEIE of the INTCON register must be set to enable any peripheral interrupt.

## **REGISTER DEFINITIONS: PIE2**

## REGISTER 2-5: PIE2: PERIPHERAL INTERRUPT ENABLE REGISTER 2

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	U-0	R/W-0
OSFIE	C2IE	C1IE	EEIE	BCLIE	ULPWUIE	—	CCP2IE
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	OSFIE: Oscillator Fail Interrupt Enable bit
	<ul><li>1 = Enables oscillator fail interrupt</li><li>0 = Disables oscillator fail interrupt</li></ul>
bit 6	C2IE: Comparator C2 Interrupt Enable bit
	<ul><li>1 = Enables Comparator C2 interrupt</li><li>0 = Disables Comparator C2 interrupt</li></ul>
bit 5	C1IE: Comparator C1 Interrupt Enable bit
	<ul><li>1 = Enables Comparator C1 interrupt</li><li>0 = Disables Comparator C1 interrupt</li></ul>
bit 4	EEIE: EEPROM Write Operation Interrupt Enable bit
	<ul><li>1 = Enables EEPROM write operation interrupt</li><li>0 = Disables EEPROM write operation interrupt</li></ul>
bit 3	BCLIE: Bus Collision Interrupt Enable bit
	<ul><li>1 = Enables Bus Collision interrupt</li><li>0 = Disables Bus Collision interrupt</li></ul>
bit 2	ULPWUIE: Ultra Low-Power Wake-up Interrupt Enable bit
	<ul><li>1 = Enables Ultra Low-Power Wake-up interrupt</li><li>0 = Disables Ultra Low-Power Wake-up interrupt</li></ul>
bit 1	Unimplemented: Read as '0'
bit 0	CCP2IE: CCP2 Interrupt Enable bit
	<ul><li>1 = Enables CCP2 interrupt</li><li>0 = Disables CCP2 interrupt</li></ul>

## 3.2.3 PIN DESCRIPTIONS AND DIAGRAMS

Each PORTA pin is multiplexed with other functions. The pins and their combined functions are briefly described here. For specific information about individual functions such as the comparator or the A/D Converter (ADC), refer to the appropriate section in this data sheet.

## 3.2.3.1 RA0/AN0/ULPWU/C12IN0-

Figure 3-1 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC
- a negative analog input to Comparator C1 or C2
- an analog input for the Ultra Low-Power Wake-up



FIGURE 3-1: BLOCK DIAGRAM OF RA0

## 3.4.4 PIN DESCRIPTIONS AND DIAGRAMS

Each PORTB pin is multiplexed with other functions. The pins and their combined functions are briefly described here. For specific information about individual functions such as the SSP,  $I^2C$  or interrupts, refer to the appropriate section in this data sheet.

### 3.4.4.1 RB0/AN12/INT

Figure 3-9 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC
- an external edge triggered interrupt

## 3.4.4.2 RB1/AN10/P1C<sup>(1)</sup>/C12IN3-

Figure 3-9 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC
- a PWM output<sup>(1)</sup>
- an analog input to Comparator C1 or C2

Note 1: P1C is available on PIC16F882/883/886 only.

## 3.4.4.3 RB2/AN8/P1B<sup>(1)</sup>

Figure 3-9 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC
- a PWM output<sup>(1)</sup>

Note 1: P1B is available on PIC16F882/883/886 only.

### 3.4.4.4 RB3/AN9/PGM/C12IN2-

Figure 3-9 shows the diagram for this pin. This pin is configurable to function as one of the following:

- a general purpose I/O
- an analog input for the ADC
- Low-voltage In-Circuit Serial Programming enable
   pin
- an analog input to Comparator C1 or C2

### FIGURE 3-9: BL

#### BLOCK DIAGRAM OF RB<3:0>



## 4.6 Clock Switching

The system clock source can be switched between external and internal clock sources via software using the System Clock Select (SCS) bit of the OSCCON register.

4.6.1 SYSTEM CLOCK SELECT (SCS) BIT

The System Clock Select (SCS) bit of the OSCCON register selects the system clock source that is used for the CPU and peripherals.

- When the SCS bit of the OSCCON register = 0, the system clock source is determined by configuration of the FOSC<2:0> bits in the Configuration Word Register 1 (CONFIG1).
- When the SCS bit of the OSCCON register = 1, the system clock source is chosen by the internal oscillator frequency selected by the IRCF<2:0> bits of the OSCCON register. After a Reset, the SCS bit of the OSCCON register is always cleared.
- Note: Any automatic clock switch, which may occur from Two-Speed Start-up or Fail-Safe Clock Monitor, does not update the SCS bit of the OSCCON register. The user can monitor the OSTS bit of the OSCCON register to determine the current system clock source.

## 4.6.2 OSCILLATOR START-UP TIME-OUT STATUS (OSTS) BIT

The Oscillator Start-up Time-out Status (OSTS) bit of the OSCCON register indicates whether the system clock is running from the external clock source, as defined by the FOSC<2:0> bits in the Configuration Word Register 1 (CONFIG1), or from the internal clock source. In particular, OSTS indicates that the Oscillator Start-up Timer (OST) has timed out for LP, XT or HS modes.

## 4.7 Two-Speed Clock Start-up Mode

Two-Speed Start-up mode provides additional power savings by minimizing the latency between external oscillator start-up and code execution. In applications that make heavy use of the Sleep mode, Two-Speed Start-up will remove the external oscillator start-up time from the time spent awake and can reduce the overall power consumption of the device.

This mode allows the application to wake-up from Sleep, perform a few instructions using the INTOSC as the clock source and go back to Sleep without waiting for the primary oscillator to become stable.

**Note:** Executing a SLEEP instruction will abort the oscillator start-up time and will cause the OSTS bit of the OSCCON register to remain clear.

When the oscillator module is configured for LP, XT or HS modes, the Oscillator Start-up Timer (OST) is enabled (see **Section 4.4.1 "Oscillator Start-up Timer (OST)**"). The OST will suspend program execution until 1024 oscillations are counted. Two-Speed Start-up mode minimizes the delay in code execution by operating from the internal oscillator as the OST is counting. When the OST count reaches 1024 and the OSTS bit of the OSCCON register is set, program execution switches to the external oscillator.

### 4.7.1 TWO-SPEED START-UP MODE CONFIGURATION

Two-Speed Start-up mode is configured by the following settings:

- IESO (of the Configuration Word Register 1) = 1; Internal/External Switchover bit (Two-Speed Start-up mode enabled).
- SCS (of the OSCCON register) = 0.
- FOSC<2:0> bits in the Configuration Word Register 1 (CONFIG1) configured for LP, XT or HS mode.

Two-Speed Start-up mode is entered after:

- Power-on Reset (POR) and, if enabled, after Power-up Timer (PWRT) has expired, or
- Wake-up from Sleep.

If the external clock oscillator is configured to be anything other than LP, XT or HS mode, then Twospeed Start-up is disabled. This is because the external clock oscillator does not require any stabilization time after POR or an exit from Sleep.

## 9.3 A/D Acquisition Requirements

For the ADC to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The Analog Input model is shown in Figure 9-4. The source impedance (Rs) and the internal sampling switch (Rss) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (Rss) impedance varies over the device voltage (VDD), see Figure 9-4. The maximum recommended impedance for analog sources is 10 k $\Omega$ . As the source impedance is decreased, the acquisition time may be decreased. After the analog input channel is selected (or changed), an A/D acquisition must be done before the conversion can be started. To calculate the minimum acquisition time, Equation 9-1 may be used. This equation assumes that 1/2 LSb error is used (1024 steps for the ADC). The 1/2 LSb error is the maximum error allowed for the ADC to meet its specified resolution.

## EQUATION 9-1: ACQUISITION TIME EXAMPLE

Assumptions: Temperature = 50°C and external impedance of 10k 
$$\Omega$$
 5.0V VDD  

$$T_{ACQ} = Amplifier Settling Time + Hold Capacitor Charging Time + Temperature Coefficient
= TAMP + TC + TCOFF
= 2µs + TC + [(Temperature - 25°C)(0.05µs/°C)]
The value for Tc can be approximated with the following equations:
$$V_{APPLIED} \left( 1 - \frac{1}{(2^{n+1}) - 1} \right) = V_{CHOLD} \qquad ;[1] V_{CHOLD} charged to within 1/2 lsb
V_{APPLIED} \left( 1 - \frac{1}{(2^{n+1}) - 1} \right) = V_{CHOLD} \qquad ;[2] V_{CHOLD} charge response to V_{APPLIED}
$$V_{APPLIED} \left( 1 - e^{\frac{-TC}{RC}} \right) = V_{CHOLD} \qquad ;[2] V_{CHOLD} charge response to V_{APPLIED}
V_{APPLIED} \left( 1 - e^{\frac{-TC}{RC}} \right) = V_{APPLIED} \left( 1 - \frac{1}{(2^{n+1}) - 1} \right) \qquad ;combining [1] and [2]$$
Solving for TC:  

$$T_{C} = -C_{HOLD} (RIC + RSS + RS) ln(1/2047) = -10pF(1k\Omega + 7k\Omega + 10k\Omega) ln(0.0004885) = 1.37µs$$
Therefore:  

$$T_{ACQ} = 2MS + 1.37MS + [(50°C - 25°C)(0.05MS/°C)] = 4.67MS$$$$$$

**Note 1:** The reference voltage (VREF) has no effect on the equation, since it cancels itself out.

- 2: The charge holding capacitor (CHOLD) is not discharged after each conversion.
- **3:** The maximum recommended impedance for analog sources is  $10 \text{ k}\Omega$ . This is required to meet the pin leakage specification.

© 2006-2015 Microchip Technology Inc.

## 11.5.4 OPERATION IN SLEEP MODE

In Sleep mode, the TMR2 register will not increment and the state of the module will not change. If the CCPx pin is driving a value, it will continue to drive that value. When the device wakes up, TMR2 will continue from its previous state.

#### 11.5.5 CHANGES IN SYSTEM CLOCK FREQUENCY

The PWM frequency is derived from the system clock frequency. Any changes in the system clock frequency will result in changes to the PWM frequency. See Section 4.0 "Oscillator Module (With Fail-Safe Clock Monitor)" for additional details.

### 11.5.6 EFFECTS OF RESET

Any Reset will force all ports to Input mode and the CCP registers to their Reset states.

### 11.5.7 SETUP FOR PWM OPERATION

The following steps should be taken when configuring the CCP module for PWM operation:

- 1. Disable the PWM pin (CCPx) output drivers as an input by setting the associated TRIS bit.
- 2. Set the PWM period by loading the PR2 register.
- Configure the CCP module for the PWM mode by loading the CCPxCON register with the appropriate values.
- Set the PWM duty cycle by loading the CCPRxL register and DCxB<1:0> bits of the CCPxCON register.
- 5. Configure and start Timer2:
  - Clear the TMR2IF interrupt flag bit of the PIR1 register.
  - Set the Timer2 prescale value by loading the T2CKPS bits of the T2CON register.
  - Enable Timer2 by setting the TMR2ON bit of the T2CON register.
- 6. Enable PWM output after a new PWM cycle has started:
  - Wait until Timer2 overflows (TMR2IF bit of the PIR1 register is set).
  - Enable the CCPx pin output driver by clearing the associated TRIS bit.

## **REGISTER DEFINITIONS: PWM CONTROL**

### REGISTER 11-4: PWM1CON: ENHANCED PWM CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
PRSEN	PDC6	PDC5	PDC4	PDC3	PDC2	PDC1	PDC0
bit 7				·	•		bit 0
Legend:							
R = Readable bit W = Writable bit			U = Unimplemented bit, read as '0'				
-n = Value at POR		'1' = Bit is set		'0' = Bit is cleared		x = Bit is unknown	
<u></u>							

bit 7 **PRSEN:** PWM Restart Enable bit

1 = Upon auto-shutdown, the ECCPASE bit clears automatically once the shutdown event goes away; the PWM restarts automatically

0 = Upon auto-shutdown, ECCPASE must be cleared in software to restart the PWM

#### bit 6-0 PDC<6:0>: PWM Delay Count bits

PDCn = Number of Fosc/4 (4 \* Tosc) cycles between the scheduled time when a PWM signal **should** transition active and the **actual** time it transitions active.

## 12.1 EUSART Asynchronous Mode

The EUSART transmits and receives data using the standard non-return-to-zero (NRZ) format. NRZ is implemented with two levels: a VOH mark state which represents a '1' data bit, and a VOL space state which represents a '0' data bit. NRZ refers to the fact that consecutively transmitted data bits of the same value stay at the output level of that bit without returning to a neutral level between each bit transmission. An NRZ transmission port idles in the mark state. Each character transmission consists of one Start bit followed by eight or nine data bits and is always terminated by one or more Stop bits. The Start bit is always a space and the Stop bits are always marks. The most common data format is eight bits. Each transmitted bit persists for a period of 1/(Baud Rate). An on-chip dedicated 8-bit/16bit Baud Rate Generator is used to derive standard baud rate frequencies from the system oscillator. See Table 12-5 for examples of baud rate configurations.

The EUSART transmits and receives the LSb first. The EUSART's transmitter and receiver are functionally independent, but share the same data format and baud rate. Parity is not supported by the hardware, but can be implemented in software and stored as the ninth data bit.

### 12.1.1 EUSART ASYNCHRONOUS TRANSMITTER

The EUSART transmitter block diagram is shown in Figure 12-1. The heart of the transmitter is the serial Transmit Shift Register (TSR), which is not directly accessible by software. The TSR obtains its data from the transmit buffer, which is the TXREG register.

#### 12.1.1.1 Enabling the Transmitter

The EUSART transmitter is enabled for asynchronous operations by configuring the following three control bits:

- TXEN = 1
- SYNC = 0
- SPEN = 1

All other EUSART control bits are assumed to be in their default state.

Setting the TXEN bit of the TXSTA register enables the transmitter circuitry of the EUSART. Clearing the SYNC bit of the TXSTA register configures the EUSART for asynchronous operation. Setting the SPEN bit of the RCSTA register enables the EUSART and automatically configures the TX/CK I/O pin as an output. If the TX/CK pin is shared with an analog peripheral the analog I/O function must be disabled by clearing the corresponding ANSEL bit.

- Note 1: When the SPEN bit is set the RX/DT I/O pin is automatically configured as an input, regardless of the state of the corresponding TRIS bit and whether or not the EUSART receiver is enabled. The RX/DT pin data can be read via a normal PORT read but PORT latch data output is precluded.
  - 2: The TXIF transmitter interrupt flag is set when the TXEN enable bit is set.

### 12.1.1.2 Transmitting Data

A transmission is initiated by writing a character to the TXREG register. If this is the first character, or the previous character has been completely flushed from the TSR, the data in the TXREG is immediately transferred to the TSR register. If the TSR still contains all or part of a previous character, the new character data is held in the TXREG until the Stop bit of the previous character has been transmitted. The pending character in the TXREG is then transferred to the TSR in one TCY immediately following the Stop bit sequence commences immediately following the transfer of the data to the TSR from the TXREG.

### 12.1.1.3 Transmit Interrupt Flag

The TXIF interrupt flag bit of the PIR1 register is set whenever the EUSART transmitter is enabled and no character is being held for transmission in the TXREG. In other words, the TXIF bit is only clear when the TSR is busy with a character and a new character has been queued for transmission in the TXREG. The TXIF flag bit is not cleared immediately upon writing TXREG. TXIF becomes valid in the second instruction cycle following the write execution. Polling TXIF immediately following the TXREG write will return invalid results. The TXIF bit is read-only, it cannot be set or cleared by software.

The TXIF interrupt can be enabled by setting the TXIE interrupt enable bit of the PIE1 register. However, the TXIF flag bit will be set whenever the TXREG is empty, regardless of the state of TXIE enable bit.

To use interrupts when transmitting data, set the TXIE bit only when there is more data to send. Clear the TXIE interrupt enable bit upon writing the last character of the transmission to the TXREG.





2: The EESART temples is the while the WOE bit is set.

### 12.3.3 BREAK CHARACTER SEQUENCE

The EUSART module has the capability of sending the special Break character sequences that are required by the LIN bus standard. A Break character consists of a Start bit, followed by 12 '0' bits and a Stop bit.

To send a Break character, set the SENDB and TXEN bits of the TXSTA register. The Break character transmission is then initiated by a write to the TXREG. The value of data written to TXREG will be ignored and all '0's will be transmitted.

The SENDB bit is automatically reset by hardware after the corresponding Stop bit is sent. This allows the user to preload the transmit FIFO with the next transmit byte following the Break character (typically, the Sync character in the LIN specification).

The TRMT bit of the TXSTA register indicates when the transmit operation is active or Idle, just as it does during normal transmission. See Figure 12-9 for the timing of the Break character sequence.

#### 12.3.3.1 Break and Sync Transmit Sequence

The following sequence will start a message frame header made up of a Break, followed by an auto-baud Sync byte. This sequence is typical of a LIN bus master.

- 1. Configure the EUSART for the desired mode.
- 2. Set the TXEN and SENDB bits to enable the Break sequence.
- 3. Load the TXREG with a dummy character to initiate transmission (the value is ignored).
- 4. Write '55h' to TXREG to load the Sync character into the transmit FIFO buffer.
- 5. After the Break has been sent, the SENDB bit is reset by hardware and the Sync character is then transmitted.

When the TXREG becomes empty, as indicated by the TXIF, the next data byte can be written to TXREG.

### 12.3.4 RECEIVING A BREAK CHARACTER

The Enhanced EUSART module can receive a Break character in two ways.

The first method to detect a Break character uses the FERR bit of the RCSTA register and the Received data as indicated by RCREG. The Baud Rate Generator is assumed to have been initialized to the expected baud rate.

A Break character has been received when;

- RCIF bit is set
- FERR bit is set
- RCREG = 00h

The second method uses the Auto-Wake-up feature described in **Section 12.3.2** "Auto-Wake-up on **Break**". By enabling this feature, the EUSART will sample the next two transitions on RX/DT, cause an RCIF interrupt, and receive the next data byte followed by another interrupt.

Note that following a Break character, the user will typically want to enable the Auto-Baud Detect feature. For both methods, the user can set the ABDEN bit of the BAUDCTL register before placing the EUSART in Sleep mode.

#### 13.3.3 MASTER MODE

The master can initiate the data transfer at any time because it controls the SCK. The master determines when the slave is to broadcast data by the software protocol.

In Master mode, the data is transmitted/received as soon as the SSPBUF register is written to. If the SPI is only going to receive, the SDO output could be disabled (programmed as an input). The SSPSR register will continue to shift in the signal present on the SDI pin at the programmed clock rate. As each byte is received, it will be loaded into the SSPBUF register as a normal received byte (interrupts and Status bits appropriately set). This could be useful in receiver applications as a "Line Activity Monitor" mode. The clock polarity is selected by appropriately programming the CKP bit of the SSPCON register. This, then, would give waveforms for SPI communication as shown in Figure 13-2, Figure 13-4 and Figure 13-5, where the MSb is transmitted first. In Master mode, the SPI clock rate (bit rate) is user programmable to be one of the following:

- Fosc/4 (or Tcy)
- Fosc/16 (or 4 Tcy)
- Fosc/64 (or 16 Tcy)
- Timer2 output/2

This allows a maximum data rate (at 40 MHz) of 10.00 Mbps.

Figure 13-2 shows the waveforms for Master mode. When the CKE bit of the SSPSTAT register is set, the SDO data is valid before there is a clock edge on SCK. The change of the input sample is shown based on the state of the SMP bit of the SSPSTAT register. The time when the SSPBUF is loaded with the received data is shown.





## 13.4.16.2 Bus Collision During a Repeated Start Condition

During a Repeated Start condition, a bus collision occurs if:

- a) A low level is sampled on SDA when SCL goes from low level to high level.
- b) SCL goes low before SDA is asserted low, indicating that another master is attempting to transmit a data '1'.

When the user de-asserts SDA and the pin is allowed to float high, the BRG is loaded with SSPADD<6:0> and counts down to 0. The SCL pin is then de-asserted, and when sampled high, the SDA pin is sampled.

If SDA is low, a bus collision has occurred (i.e, another master is attempting to transmit a data '0', see Figure 13-24). If SDA is sampled high, the BRG is reloaded and begins counting. If SDA goes from highto-low before the BRG times out, no bus collision occurs because no two masters can assert SDA at exactly the same time.

If SCL goes from high-to-low before the BRG times out and SDA has not already been asserted, a bus collision occurs. In this case, another master is attempting to transmit a data '1' during the Repeated Start condition (Figure 13-25).

If at the end of the BRG time-out, both SCL and SDA are still high, the SDA pin is driven low and the BRG is reloaded and begins counting. At the end of the count, regardless of the status of the SCL pin, the SCL pin is driven low and the Repeated Start condition is complete.

FIGURE 13-24: BUS COLLISION DURING A REPEATED START CONDITION (CASE 1)



### FIGURE 13-25: BUS COLLISION DURING REPEATED START CONDITION (CASE 2)



## 15.2 Instruction Descriptions

ADDLW	Add literal and W
Syntax:	[label] ADDLW k
Operands:	$0 \leq k \leq 255$
Operation:	$(W) + k \to (W)$
Status Affected:	C, DC, Z
Description:	The contents of the W register are added to the 8-bit literal 'k' and the result is placed in the W register.

BCF	Bit Clear f
Syntax:	[ label ] BCF f,b
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b \leq 7 \end{array}$
Operation:	$0 \rightarrow (f < b >)$
Status Affected:	None
Description:	Bit 'b' in register 'f' is cleared.

ADDWF	Add W and f
Syntax:	[label] ADDWF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(W) + (f) $\rightarrow$ (destination)
Status Affected:	C, DC, Z
Description:	Add the contents of the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

BSF	Bit Set f
Syntax:	[label]BSF f,b
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b \leq 7 \end{array}$
Operation:	$1 \rightarrow (f < b >)$
Status Affected:	None
Description:	Bit 'b' in register 'f' is set.

ANDLW	AND literal with W	
Syntax:	[ <i>label</i> ] ANDLW k	
Operands:	$0 \leq k \leq 255$	
Operation:	(W) .AND. (k) $\rightarrow$ (W)	
Status Affected:	Z	
Description:	The contents of W register are AND'ed with the 8-bit literal 'k'. The result is placed in the W reg- ister.	

BTFSC	Bit Test f, Skip if Clear	
Syntax:	[ label ] BTFSC f,b	
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b \leq 7 \end{array}$	
Operation:	skip if (f <b>) = <math>0</math></b>	
Status Affected:	None	
Description:	None If bit 'b' in register 'f' is '1', the next instruction is executed. If bit 'b' in register 'f' is '0', the next instruction is discarded, and a NOP is executed instead, making this a 2-cycle instruction.	

ANDWF	AND W with f	
Syntax:	[label] ANDWF f,d	
Operands:	$0 \le f \le 127$ $d \in [0,1]$	
Operation:	(W) .AND. (f) $\rightarrow$ (destination)	
Status Affected:	Z	
Description:	AND the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.	

DECFSZ	Decrement f, Skip if 0	
Syntax:	[label] DECFSZ f,d	
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$	
Operation:	(f) - 1 $\rightarrow$ (destination); skip if result = 0	
Status Affected:	None	
Description:	The contents of register 'f' are decremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'. If the result is '1', the next instruction is executed. If the result is '0', then a NOP is executed instead, making it a 2-cycle instruction.	

GOTO	Unconditional Branch		
Syntax:	[ <i>label</i> ] GOTO k		
Operands:	$0 \le k \le 2047$		
Operation:	$k \rightarrow PC < 10:0>$ PCLATH<4:3> $\rightarrow$ PC<12:11>		
Status Affected:	None		
Description:	GOTO is an unconditional branch. The 11-bit immediate value is loaded into PC bits <10:0>. The upper bits of PC are loaded from PCLATH<4:3>. GOTO is a 2-cycle instruction		

INCF	Increment f		
Syntax:	[ <i>label</i> ] INCF f,d		
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$		
Operation:	(f) + 1 $\rightarrow$ (destination)		
Status Affected:	Z		
Description:	The contents of register 'f' are incremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.		

INCFSZ	Increment f, Skip if 0	
Syntax:	[label] INCFSZ f,d	
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$	
Operation:	(f) + 1 $\rightarrow$ (destination), skip if result = 0	
Status Affected:	None	
Description:	None The contents of register 'f' are incremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'. If the result is '1', the next instruction is executed. If the result is '0', a NOP is executed instead, making it a 2-cycle	

IORLW	Inclusive OR literal with W		
Syntax:	[ <i>label</i> ] IORLW k		
Operands:	$0 \le k \le 255$		
Operation:	(W) .OR. $k \rightarrow$ (W)		
Status Affected:	Z		
Description:	The contents of the W register are OR'ed with the 8-bit literal 'k'. The result is placed in the W register.		

IORWF	Inclusive OR W with f		
Syntax:	[ <i>label</i> ] IORWF f,d		
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$		
Operation:	(W) .OR. (f) $\rightarrow$ (destination)		
Status Affected:	Z		
Description:	Inclusive OR the W register with register 'f'. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.		

## 17.7 Timing Parameter Symbology

The timing parameter symbols have been created with one of the following formats:

#### 1. TppS2ppS

2. TppS

Т			
F	Frequency	Т	Time
Lowerc	ase letters (pp) and their meanings:		
рр			
сс	CCP1	osc	OSC1
ck	CLKOUT	rd	RD
CS	CS	rw	RD or WR
di	SDI	SC	SCK
do	SDO	SS	SS
dt	Data in	tO	TOCKI
io	I/O PORT	t1	T1CKI
mc	MCLR	wr	WR
Upperc	ase letters and their meanings:		
S			
F	Fall	Р	Period
Н	High	R	Rise
I	Invalid (High-impedance)	V	Valid
L	Low	Z	High-impedance

### FIGURE 17-3: LOAD CONDITIONS











## **PRODUCT IDENTIFICATION SYSTEM**

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

PART NO.	[X] <sup>(1)</sup> - X <u>/XX XXX</u> T Tape and Reel Temperature Package Pattern Option Range	<ul> <li>Examples:</li> <li>a) PIC16F883-E/P 301 = Extended Temp., PDIP package, 20 MHz, QTP pattern #301</li> <li>b) PIC16F883-I/SO = Industrial Temp., SOIC package 20 MHz</li> </ul>
Device:	PIC16F883, PIC16F883T <sup>(1)</sup> , PIC16F884, PIC16F884T <sup>(1)</sup> , PIC16F886, PIC16F886T <sup>(1)</sup> , PIC16F887, PIC16F887T <sup>(1)</sup> , VDD range 2.0V to 5.5V	pasiago, 20 m.2
Tape and Reel Option:	Blank = Standard packaging (tube or tray) T = Tape and Reel <sup>(1)</sup>	
Temperature Range:	I = $-40^{\circ}$ C to $+85^{\circ}$ C (Industrial) E = $-40^{\circ}$ C to $+125^{\circ}$ C (Extended)	
Package: <sup>(2)</sup>	ML=Quad Flat No Leads (QFN)P=Plastic DIPPT=Plastic Thin-Quad Flatpack (TQFP)SO=Plastic Small Outline (SOIC) (7.50 mm)SP=Skinny Plastic DIPSS=Plastic Shrink Small Outline	<ul> <li>Note 1: Tape and Reel identifier only appears in the catalog part number description. This identifier is used for ordering purposes and is not printed on the device package. Check with your Microchip Sales Office for package availability with the Tape and Reel option.</li> <li>2: For other small form-factor package</li> </ul>
Pattern:	QTP, SQTP, Code or Special Requirements (blank otherwise)	availability and marking information, please visit www.microchip.com/packaging or contact your local sales office.