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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e300
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	533MHz
Co-Processors/DSP	Security; SEC
RAM Controllers	DDR
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (2)
SATA	-
USB	USB 2.0 + PHY (2)
Voltage - I/O	2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	Cryptography, Random Number Generator
Package / Case	672-LBGA
Supplier Device Package	672-LBGA (35x35)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8347evvajdb

  - Data chaining and direct mode
  - Interrupt on completed segment and chain
- DUART
  - Two 4-wire interfaces (RxD, TxD, RTS, CTS)
  - Programming model compatible with the original 16450 UART and the PC16550D
- Serial peripheral interface (SPI) for master or slave
- General-purpose parallel I/O (GPIO)
  - 52 parallel I/O pins multiplexed on various chip interfaces
- System timers
  - Periodic interrupt timer
  - Real-time clock
  - Software watchdog timer
  - Eight general-purpose timers
- Designed to comply with IEEE Std. 1149.1™, JTAG boundary scan
- Integrated PCI bus and SDRAM clock generation

Figure 3 shows the undershoot and overshoot voltage of the PCI interface of the MPC8347E for the 3.3-V signals, respectively.

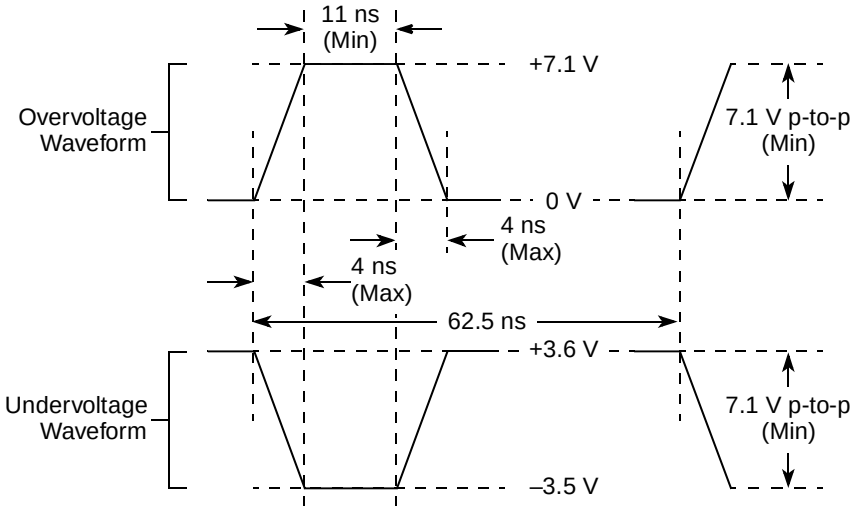

Figure 3. Maximum AC Waveforms on PCI Interface for 3.3-V Signaling

2.1.3 Output Driver Characteristics

Table 3 provides information on the characteristics of the output driver strengths. The values are preliminary estimates.

Table 3. Output Drive Capability

Driver Type	Output Impedance (Ω)	Supply Voltage
Local bus interface utilities signals	40	OV_{DD} = 3.3 V
PCI signals (not including PCI output clocks)	25	
PCI output clocks (including PCI_SYNC_OUT)	40	
DDR signal	18	GV_{DD} = 2.5 V
TSEC/10/100 signals	40	LV_{DD} = 2.5/3.3 V
DUART, system control, I^2C, JTAG, USB	40	OV_{DD} = 3.3 V
GPIO signals	40	OV_{DD} = 3.3 V, LV_{DD} = 2.5/3.3 V

2.2 Power Sequencing

MPC8347E does not require the core supply voltage and I/O supply voltages to be applied in any particular order. Note that during the power ramp up, before the power supplies are stable, there may be a period of time that I/O pins are actively driven. After the power is stable, as long as $\overline{\text{PORESET}}$ is asserted, most I/O pins are three-stated. To minimize the time that I/O pins are actively driven, it is recommended to apply core voltage before I/O voltage and assert $\overline{\text{PORESET}}$ before the power supplies fully ramp up.

Table 5 shows the estimated typical I/O power dissipation for MPC8347E.

Table 5. MPC8347E Typical I/O Power Dissipation

Interface	Parameter	DDR2 GV_{DD} (1.8 V)	DDR1 GV_{DD} (2.5 V)	OV_{DD} (3.3 V)	LV_{DD} (3.3 V)	LV_{DD} (2.5 V)	Unit	Comments
DDR I/O 65% utilization 2.5 V Rs = 20 Ω Rt = 50 Ω 2 pair of clocks	200 MHz, 32 bits	—	0.42	—	—	—	W	—
	200 MHz, 64 bits	—	0.55	—	—	—	W	—
	266 MHz, 32 bits	—	0.5	—	—	—	W	—
	266 MHz, 64 bits	—	0.66	—	—	—	W	—
	300 MHz,[1] 32 bits	—	0.54	—	—	—	W	—
	300 MHz,[1] 64 bits	—	0.7	—	—	—	W	—
	333 MHz,[1] 32 bits	—	0.58	—	—	—	W	—
	333 MHz,[1] 64 bits	—	0.76	—	—	—	W	—
	400 MHz,[1] 32 bits	—	—	—	—	—		—
	400 MHz,[1] 64 bits	—	—	—	—	—		—
PCI I/O load = 30 pF	33 MHz, 32 bits	—	—	0.04	—	—	W	—
	66 MHz, 32 bits	—	—	0.07	—	—	W	—
Local bus I/O load = 25 pF	167 MHz, 32 bits	—	—	0.34	—	—	W	—
	133 MHz, 32 bits	—	—	0.27	—	—	W	—
	83 MHz, 32 bits	—	—	0.17	—	—	W	—
	66 MHz, 32 bits	—	—	0.14	—	—	W	—
	50 MHz, 32 bits	—	—	0.11	—	—	W	—
TSEC I/O load = 25 pF	MII	—	—	—	0.01	—	W	Multiply by number of interfaces used.
	GMII or TBI	—	—	—	0.06	—	W	
	RGMII or RTBI	—	—	—	—	0.04	W	
USB	12 MHz	—	—	0.01	—	—	W	Multiply by 2 if using 2 ports.
	480 MHz	—	—	0.2	—	—	W	
Other I/O		—	—	0.01	—	—	W	—

[1] TBGA package only.

6.2 DDR SDRAM AC Electrical Characteristics

This section provides the AC electrical characteristics for the DDR SDRAM interface.

6.2.1 DDR SDRAM Input AC Timing Specifications

Table 13 provides the input AC timing specifications for the DDR SDRAM interface.

Table 13. DDR SDRAM Input AC Timing Specifications

At recommended operating conditions with GV_{DD} of 2.5 V ± 5%.

Parameter	Symbol	Min	Max	Unit	Notes
AC input low voltage	V_{IL}	—	MV_{REF} – 0.31	V	
AC input high voltage	V_{IH}	MV_{REF} + 0.31	GV_{DD} + 0.3	V	
MDQS—MDQ/MECC input skew per byte 333 MHz 266 MHz	t_{DISKEW}	—	 750 1125	ps	1

Note:

1. Maximum possible skew between a data strobe (MDQS[n]) and any corresponding bit of data (MDQ[8n + {0...7}] if 0 <= n <= 7) or ECC (MECC[{0...7}] if n = 8).

Figure 4 illustrates the DDR input timing diagram showing the t_{DISKEW} timing parameter.

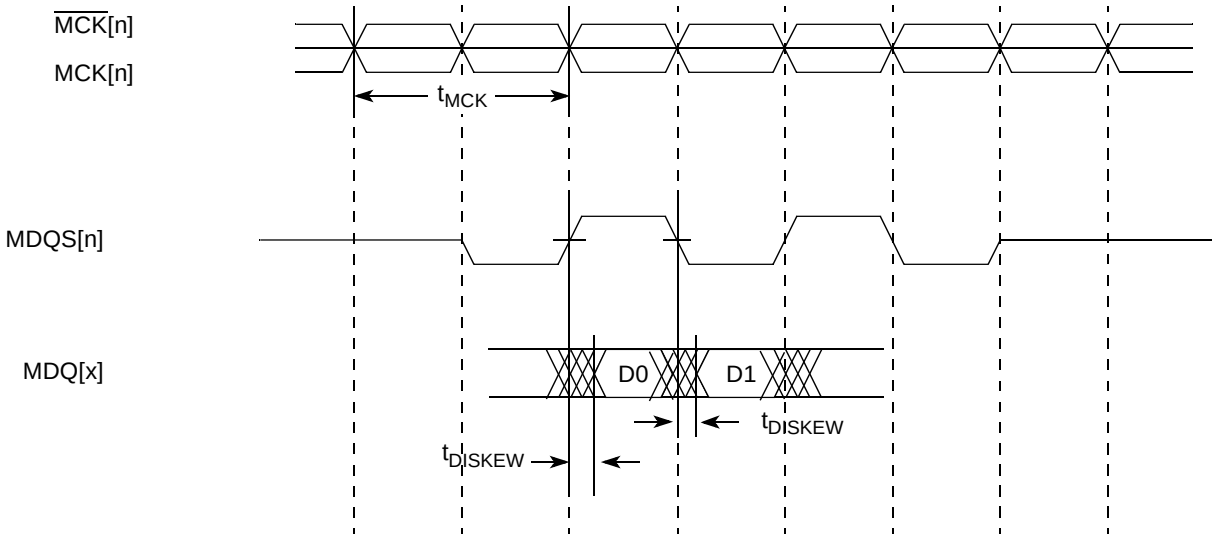

Figure 4. DDR Input Timing Diagram

6.2.2 DDR SDRAM Output AC Timing Specifications

Table 14 and Table 15 provide the output AC timing specifications and measurement conditions for the DDR SDRAM interface.

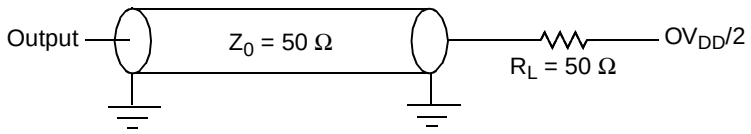

Figure 6. DDR AC Test Load

Table 15 shows the DDR SDRAM measurement conditions.

Table 15. DDR SDRAM Measurement Conditions

Symbol	DDR	Unit	Notes
V_{TH}	MV_{REF} ± 0.31 V	V	1
V_{OUT}	0.5 × GV_{DD}	V	2

Notes:
1. Data input threshold measurement point.
2. Data output measurement point.

Figure 7 shows the DDR SDRAM output timing diagram for source synchronous mode.

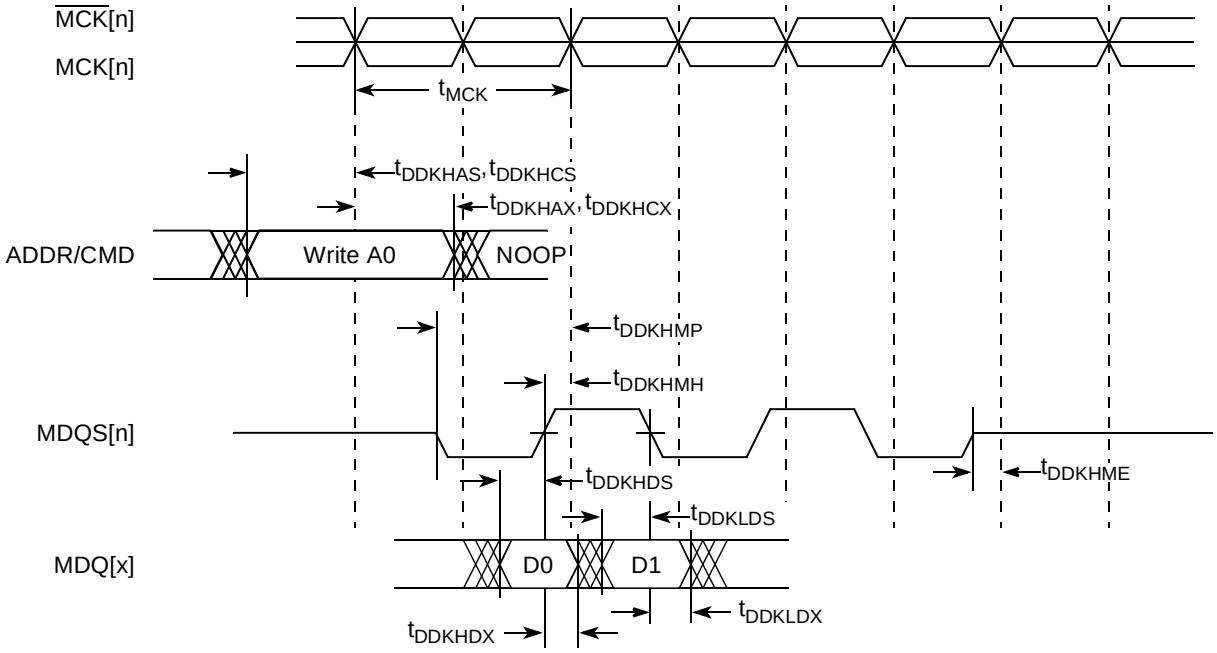

Figure 7. DDR SDRAM Output Timing Diagram for Source Synchronous Mode

Table 16 provides approximate delay information that can be expected for the address and command signals of the DDR controller for various loadings, which can be useful for a system utilizing the DLL. These numbers are the result of simulations for one topology. The delay numbers will strongly depend on the topology used. These delay numbers show the total delay for the address and command to arrive at the DRAM devices. The actual delay could be different than the delays seen in simulation, depending on the system topology. If a heavily loaded system is used, the DLL loop may need to be adjusted to meet setup requirements at the DRAM.

Figure 12 shows the MII receive AC timing diagram.

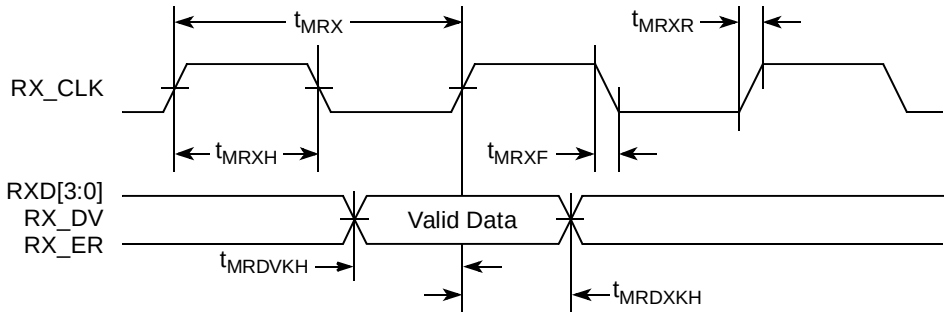

Figure 12. MII Receive AC Timing Diagram

8.2.3 TBI AC Timing Specifications

This section describes the TBI transmit and receive AC timing specifications.

8.2.3.1 TBI Transmit AC Timing Specifications

Table 25 provides the TBI transmit AC timing specifications.

Table 25. TBI Transmit AC Timing Specifications

At recommended operating conditions with LV_{DD}/OV_{DD} of 3.3 V ± 10%.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit
GTX_CLK clock period	t_{TTX}	—	8.0	—	ns
GTX_CLK duty cycle	t_{TTXH}/t_{TTX}	40	—	60	%
GTX_CLK to TBI data TXD[7:0], TX_ER, TX_EN delay	t_{TTKHDX}	1.0	—	5.0	ns
GTX_CLK clock rise, V_{IL}(min) to V_{IH}(max)	t_{TTXR}	—	—	1.0	ns
GTX_CLK clock fall time, V_{IH}(max) to V_{IL}(min)	t_{TTXF}	—	—	1.0	ns
GTX_CLK125 reference clock period	t_{G125}[2]	—	8.0	—	ns
GTX_CLK125 reference clock duty cycle	t_{G125H}/t_{G125}	45	—	55	ns

Notes:

1. The symbols for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{TTKHDV} symbolizes the TBI transmit timing (TT) with respect to the time from t_{TTX} (K) going high (H) until the referenced data signals (D) reach the valid state (V) or setup time. Also, t_{TTKHDX} symbolizes the TBI transmit timing (TT) with respect to the time from t_{TTX} (K) going high (H) until the referenced data signals (D) reach the invalid state (X) or hold time. In general, the clock reference symbol is based on three letters representing the clock of a particular function. For example, the subscript of t_{TTX} represents the TBI (T) transmit (TX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. This symbol represents the external GTX_CLK125 and does not follow the original symbol naming convention

10 Local Bus

This section describes the DC and AC electrical specifications for the local bus interface of the MPC8347E.

10.1 Local Bus DC Electrical Characteristics

Table 33 provides the DC electrical characteristics for the local bus interface.

Table 33. Local Bus DC Electrical Characteristics

Parameter	Symbol	Min	Max	Unit
High-level input voltage	V_{IH}	2	OV_{DD} + 0.3	V
Low-level input voltage	V_{IL}	–0.3	0.8	V
Input current	I_{IN}	—	±5	μA
High-level output voltage, I_{OH} = –100 μA	V_{OH}	OV_{DD} – 0.2	—	V
Low-level output voltage, I_{OL} = 100 μA	V_{OL}	—	0.2	V

10.2 Local Bus AC Electrical Specification

Table 34 and Table 35 describe the general timing parameters of the local bus interface of the MPC8347E.

Table 34. Local Bus General Timing Parameters—DLL On

Parameter	Symbol[1]	Min	Max	Unit	Notes
Local bus cycle time	t_{LBK}	7.5	—	ns	2
Input setup to local bus clock (except LUPWAIT)	$t_{LBIVKH1}$	1.5	—	ns	3, 4
LUPWAIT input setup to local bus clock	$t_{LBIVKH2}$	2.2	—	ns	3, 4
Input hold from local bus clock (except LUPWAIT)	$t_{LBIXKH1}$	1.0	—	ns	3, 4
LUPWAIT Input hold from local bus clock	$t_{LBIXKH2}$	1.0	—	ns	3, 4
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT1}$	1.5	—	ns	5
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT2}$	3	—	ns	6
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT3}$	2.5	—	ns	7
Local bus clock to LALE rise	t_{LBKHLR}	—	4.5	ns	
Local bus clock to output valid (except LAD/LDP and LALE)	$t_{LBKHOV1}$	—	4.5	ns	
Local bus clock to data valid for LAD/LDP	$t_{LBKHOV2}$	—	4.5	ns	3
Local bus clock to address valid for LAD	$t_{LBKHOV3}$	—	4.5	ns	3
Output hold from local bus clock (except LAD/LDP and LALE)	$t_{LBKHOX1}$	1	—	ns	3

Table 34. Local Bus General Timing Parameters—DLL On (continued)

Parameter	Symbol[1]	Min	Max	Unit	Notes
Output hold from local bus clock for LAD/LDP	$t_{LBKHOX2}$	1	—	ns	3
Local bus clock to output high impedance for LAD/LDP	t_{LBKHOZ}	—	3.8	ns	8

Notes:
1. The symbols for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{LBIXKH1}$ symbolizes local bus timing (LB) for the input (I) to go invalid (X) with respect to the time the t_{LBK} clock reference (K) goes high (H), in this case for clock one (1). Also, t_{LBKHOX} symbolizes local bus timing (LB) for the t_{LBK} clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.
2. All timings are in reference to the rising edge of LSYNC_IN.
3. All signals are measured from $OV_{DD}/2$ of the rising edge of LSYNC_IN to $0.4 \times OV_{DD}$ of the signal in question for 3.3 V signaling levels.
4. Input timings are measured at the pin.
5. $t_{LBOTOT1}$ should be used when RCWH[LALE] is not set and when the load on the LALE output pin is at least 10 pF less than the load on the LAD output pins.
6. $t_{LBOTOT2}$ should be used when RCWH[LALE] is set and when the load on the LALE output pin is at least 10 pF less than the load on the LAD output pins.
7. $t_{LBOTOT3}$ should be used when RCWH[LALE] is set and when the load on the LALE output pin equals the load on the LAD output pins.
8. For active/float timing measurements, the Hi-Z or off-state is defined to be when the total current delivered through the component pin is less than or equal to that of the leakage current specification.

Table 35. Local Bus General Timing Parameters—DLL Bypass[9]

Parameter	Symbol[1]	Min	Max	Unit	Notes
Local bus cycle time	t_{LBK}	15	—	ns	2
Input setup to local bus clock	t_{LBIVKH}	7	—	ns	3, 4
Input hold from local bus clock	t_{LBIXKH}	1.0	—	ns	3, 4
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT1}$	1.5	—	ns	5
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT2}$	3	—	ns	6
LALE output fall to LAD output transition (LATCH hold time)	$t_{LBOTOT3}$	2.5	—	ns	7

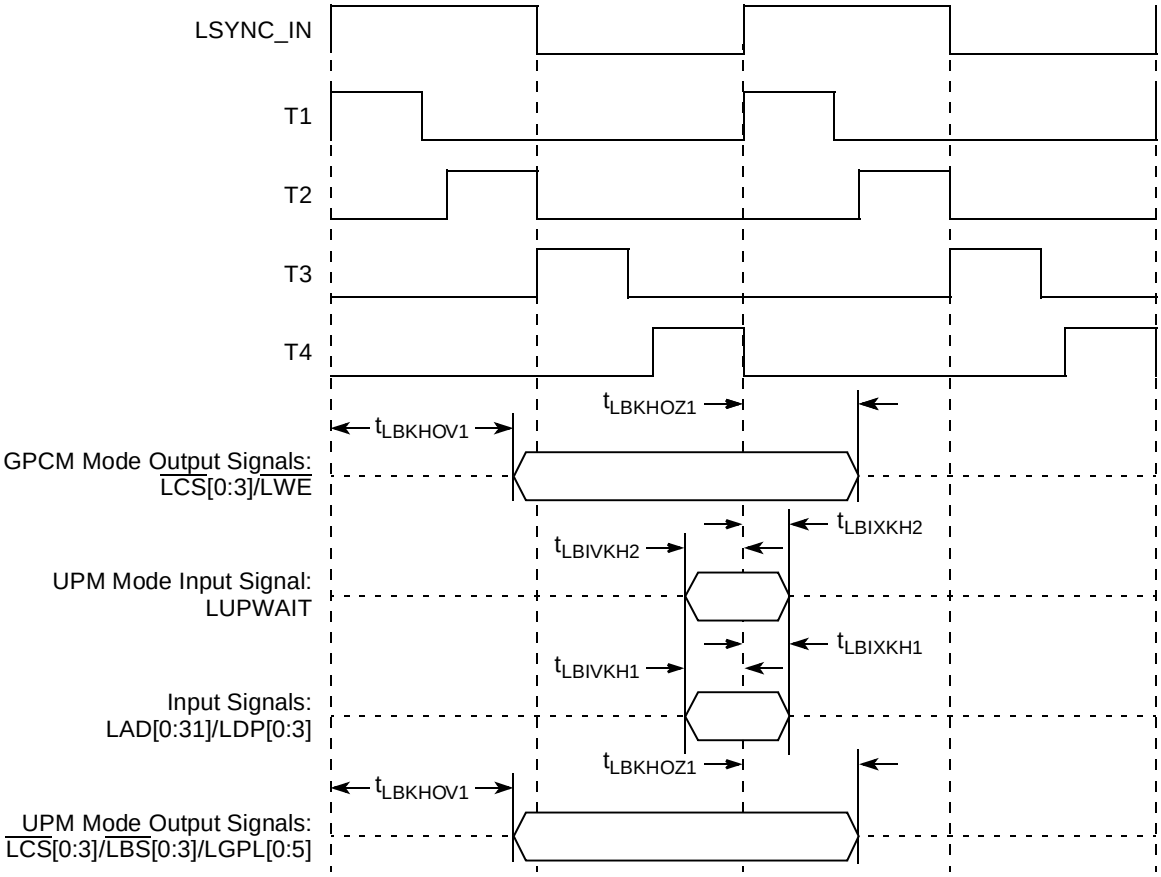

Figure 25. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 4 (DLL Enabled)

11 JTAG

This section describes the DC and AC electrical specifications for the IEEE Std. 1149.1 (JTAG) interface of the MPC8347E

11.1 JTAG DC Electrical Characteristics

Table 36 provides the DC electrical characteristics for the IEEE Std. 1149.1 (JTAG) interface of the MPC8347E.

Table 36. JTAG interface DC Electrical Characteristics

Characteristic	Symbol	Condition	Min	Max	Unit
Input high voltage	V_{IH}		OV_{DD} – 0.3	OV_{DD} + 0.3	V
Input low voltage	V_{IL}		–0.3	0.8	V
Input current	I_{IN}			±5	μA
Output high voltage	V_{OH}	I_{OH} = –8.0 mA	2.4	—	V
Output low voltage	V_{OL}	I_{OL} = 8.0 mA	—	0.5	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V

11.2 JTAG AC Timing Specifications

This section describes the AC electrical specifications for the IEEE Std. 1149.1 (JTAG) interface of the MPC8347E. Table 37 provides the JTAG AC timing specifications as defined in Figure 27 through Figure 30.

Table 37. JTAG AC Timing Specifications (Independent of CLKIN)[1]

At recommended operating conditions (see Table 2).

Parameter	Symbol[2]	Min	Max	Unit	Notes
JTAG external clock frequency of operation	f_{JTG}	0	33.3	MHz	
JTAG external clock cycle time	t_{JTG}	30	—	ns	
JTAG external clock pulse width measured at 1.4 V	t_{JTKHKL}	15	—	ns	
JTAG external clock rise and fall times	t_{JTGR}, t_{JTGF}	0	2	ns	
$\overline{TRST}$ assert time	t_{TRST}	25	—	ns	3
Input setup times: Boundary-scan data TMS, TDI	 t_{JTDVKH} t_{JTIVKH}	 4 4	 — —	ns	4
Input hold times: Boundary-scan data TMS, TDI	 t_{JTDXKH} t_{JTIXKH}	 10 10	 — —	ns	4
Valid times: Boundary-scan data TDO	 t_{JTKLDV} t_{JTKLOV}	 2 2	 11 11	ns	5

Table 41. PCI AC Timing Specifications at 66 MHz[1] (continued)

Parameter	Symbol[2]	Min	Max	Unit	Notes
Input hold from clock	t_{PCIXKH}	0	—	ns	3, 5

Notes:
1. PCI timing depends on M66EN and the ratio between PCI1/PCI2. Refer to the PCI chapter of the reference manual for a description of M66EN.
2. The symbols for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{PCIVKH} symbolizes PCI timing (PC) with respect to the time the input signals (I) reach the valid state (V) relative to the PCI_SYNC_IN clock, t_{SYS}, reference (K) going to the high (H) state or setup time. Also, t_{PCRHFV} symbolizes PCI timing (PC) with respect to the time hard reset (R) went high (H) relative to the frame signal (F) going to the valid (V) state.
3. See the timing measurement conditions in the *PCI 2.2 Local Bus Specifications*.
4. For active/float timing measurements, the Hi-Z or off-state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
5. Input timings are measured at the pin.

Table 42 provides the PCI AC timing specifications at 33 MHz.

Table 42. PCI AC Timing Specifications at 33 MHz

Parameter	Symbol[1]	Min	Max	Unit	Notes
Clock to output valid	t_{PCKHOV}	—	11	ns	2
Output hold from clock	t_{PCKHOX}	2	—	ns	2
Clock to output high impedance	t_{PCKHOZ}	—	14	ns	2, 3
Input setup to clock	t_{PCIVKH}	3.0	—	ns	2, 4
Input hold from clock	t_{PCIXKH}	0	—	ns	2, 4

Notes:
1. The symbols for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{PCIVKH} symbolizes PCI timing (PC) with respect to the time the input signals (I) reach the valid state (V) relative to the PCI_SYNC_IN clock, t_{SYS}, reference (K) going to the high (H) state or setup time. Also, t_{PCRHFV} symbolizes PCI timing (PC) with respect to the time hard reset (R) went high (H) relative to the frame signal (F) going to the valid (V) state.
2. See the timing measurement conditions in the *PCI 2.2 Local Bus Specifications*.
3. For active/float timing measurements, the Hi-Z or off-state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
4. Input timings are measured at the pin.

Figure 33 provides the AC test load for PCI.

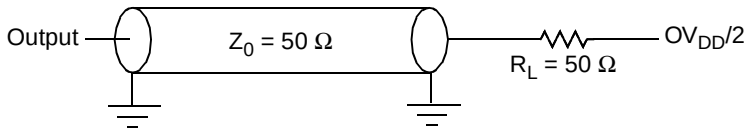

Figure 33. PCI AC Test Load

18 Package and Pin Listings

This section details package parameters, pin assignments, and dimensions. The MPC8347E is available in two packages—a tape ball grid array (TBGA) and a plastic ball grid array (PBGA). See Section 18.1, "Package Parameters for the MPC8347E TBGA," Section 18.2, "Mechanical Dimensions for the MPC8347E TBGA,Section 18.3, "Package Parameters for the MPC8347E PBGA," and Section 18.4, "Mechanical Dimensions for the MPC8347E PBGA."

18.1 Package Parameters for the MPC8347E TBGA

The package parameters are provided in the following list. The package type is 35 mm × 35 mm, 672 tape ball grid array (TBGA).

Package outline	35 mm × 35 mm
Interconnects	672
Pitch	1.00 mm
Module height (typical)	1.46 mm
Solder balls	62 Sn/36 Pb/2 Ag (ZU package) 95.5 Sn/0.5 Cu/4Ag (VV package)
Ball diameter (typical)	0.64 mm

18.5 Pinout Listings

Table 51 provides the pinout listing for the MPC8347E, 672 TBGA package.

Table 51. MPC8347E (TBGA) Pinout Listing

Signal	Package Pin Number	Pin Type	Power Supply	Notes
PCI				
PCI_INTA/IRQ_OUT	B34	O	OV_{DD}	2
PCI_RESET_OUT	C33	O	OV_{DD}	
PCI_AD[31:0]	G30, G32, G34, H31, H32, H33, H34, J29, J32, J33, L30, K31, K33, K34, L33, L34, P34, R29, R30, R33, R34, T31, T32, T33, U31, U34, V31, V32, V33, V34, W33, W34	I/O	OV_{DD}	
PCI_C/BE[3:0]	J30, M31, P33, T34	I/O	OV_{DD}	
PCI_PAR	P32	I/O	OV_{DD}	
PCI_FRAME	M32	I/O	OV_{DD}	5
PCI_TRDY	N29	I/O	OV_{DD}	5
PCI_IRDY	M34	I/O	OV_{DD}	5
PCI_STOP	N31	I/O	OV_{DD}	5
PCI_DEVSEL	N30	I/O	OV_{DD}	5
PCI_IDSEL	J31	I	OV_{DD}	
PCI_SERR	N34	I/O	OV_{DD}	5
PCI_PERR	N33	I/O	OV_{DD}	5
PCI_REQ[0]	D32	I/O	OV_{DD}	
PCI_REQ[1]/CPCI1_HS_ES	D34	I	OV_{DD}	
PCI_REQ[2:4]	E34, F32, G29	I	OV_{DD}	
PCI_GNT0	C34	I/O	OV_{DD}	
PCI_GNT1/CPCI1_HS_LED	D33	O	OV_{DD}	
PCI_GNT2/CPCI1_HS_ENUM	E33	O	OV_{DD}	
PCI_GNT[3:4]	F31, F33	O	OV_{DD}	
M66EN	A19	I	OV_{DD}	
DDR SDRAM Memory Interface				
MDQ[0:63]	D5, A3, C3, D3, C4, B3, C2, D4, D2, E5, G2, H6, E4, F3, G4, G3, H1, J2, L6, M6, H2, K6, L2, M4, N2, P4, R2, T4, P6, P3, R1, T2, AB5, AA3, AD6, AE4, AB4, AC2, AD3, AE6, AE3, AG4, AK5, AK4, AE2, AG6, AK3, AK2, AL2, AL1, AM5, AP5, AM2, AN1, AP4, AN5, AJ7, AN7, AM8, AJ9, AP6, AL7, AL9, AN8	I/O	GV_{DD}	

Table 51. MPC8347E (TBGA) Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
DUART				
UART_SOUT[1:2]/MSRCID[0:1]/LSRCID[0:1]	AK27, AN29	O	OV_{DD}	
UART_SIN[1:2]/MSRCID[2:3]/LSRCID[2:3]	AL28, AM29	I/O	OV_{DD}	
$\overline{\text{UART\_CTS}}$[1]/MSRCID4/LSRCID4	AP30	I/O	OV_{DD}	
$\overline{\text{UART\_CTS}}$[2]/MDVAL/ LDVAL	AN30	I/O	OV_{DD}	
$\overline{\text{UART\_RTS}}$[1:2]	AP31, AM30	O	OV_{DD}	
I^2C interface				
IIC1_SDA	AK29	I/O	OV_{DD}	2
IIC1_SCL	AP32	I/O	OV_{DD}	2
IIC2_SDA	AN31	I/O	OV_{DD}	2
IIC2_SCL	AM31	I/O	OV_{DD}	2
SPI				
SPIMOSI	AN32	I/O	OV_{DD}	
SPIMISO	AP33	I/O	OV_{DD}	
SPICLK	AK30	I/O	OV_{DD}	
SPISEL	AL31	I	OV_{DD}	
Clocks				
PCI_CLK_OUT[0:4]	AN9, AP9, AM10, AN10, AJ11	O	OV_{DD}	
PCI_SYNC_IN/PCI_CLOCK	AK12	I	OV_{DD}	
PCI_SYNC_OUT	AP11	O	OV_{DD}	3
RTC/PIT_CLOCK	AM32	I	OV_{DD}	
CLKIN	AM9	I	OV_{DD}	
JTAG				
TCK	E20	I	OV_{DD}	
TDI	F20	I	OV_{DD}	4
TDO	B20	O	OV_{DD}	3
TMS	A20	I	OV_{DD}	4
$\overline{\text{TRST}}$	B19	I	OV_{DD}	4
Test				
TEST	D22	I	OV_{DD}	6
TEST_SEL	AL13	I	OV_{DD}	7
PMC				
$\overline{\text{QUIESCE}}$	A18	O	OV_{DD}	

Table 51. MPC8347E (TBGA) Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
No Connection				
NC	W32, AA31, AA32, AA33, AA34, AB31, AB32, AB33, AB34, AC29, AC31, AC33, AC34, AD30, AD32, AD33, AD34, AE29, AE30, AH32, AH33, AH34, AM33, AJ31, AJ32, AJ33, AJ34, AK32, AK33, AK34, AM34, AL33, AL34, AK31, AH30, AC32, AE32, AH31, AL32, AG34, AE33, AF32, AE34, AF34, AF33, AG33, AG32, AL11, AM11, AP10, Y32, Y34, Y31, Y33	—	—	

Notes:

1. This pin is an open-drain signal. A weak pull-up resistor (1 kΩ) should be placed on this pin to OV_{DD}.
2. This pin is an open-drain signal. A weak pull-up resistor (2–10 kΩ) should be placed on this pin to OV_{DD}.
3. During reset, this output is actively driven rather than three-stated.
4. These JTAG pins have weak internal pull-up P-FETs that are always enabled.
5. This pin should have a weak pull-up if the chip is in PCI host mode. Follow the PCI specifications.
6. This pin must always be tied to GND.
7. This pin must always be pulled up to OV_{DD}.
8. This pin must always be left not connected.
9. Thermal sensitive resistor.
10. It is recommended that MDIC0 be tied to GRD using an 18 Ω resistor and MDIC1 be tied to DDR power using an 18 Ω resistor.
11. TSEC1_TXD[3] is required an external pull-up resistor. For proper functionality of the device, this pin must be pulled up or actively driven high during a hard reset. No external pull-down resistors are allowed to be attached to this net.

Table 52 provides the pinout listing for the MPC8347E, 620 PBGA package.

Table 52. MPC8347E (PBGA) Pinout Listing

Signal	Package Pin Number	Pin Type	Power Supply	Notes
PCI				
$\overline{\text{PCI1\_INTA}}$/$\overline{\text{IRQ\_OUT}}$	D20	O	OV_{DD}	2
$\overline{\text{PCI1\_RESET\_OUT}}$	B21	O	OV_{DD}	
PCI1_AD[31:0]	E19, D17, A16, A18, B17, B16, D16, B18, E17, E16, A15, C16, D15, D14, C14, A12, D12, B11, C11, E12, A10, C10, A9, E11, E10, B9, B8, D9, A8, C9, D8, C8	I/O	OV_{DD}	
PCI1_C/$\overline{\text{BE}}$[3:0]	A17, A14, A11, B10	I/O	OV_{DD}	
PCI1_PAR	D13	I/O	OV_{DD}	
$\overline{\text{PCI1\_FRAME}}$	B14	I/O	OV_{DD}	5
$\overline{\text{PCI1\_TRDY}}$	A13	I/O	OV_{DD}	5

19 Clocking

Figure 41 shows the internal distribution of the clocks.

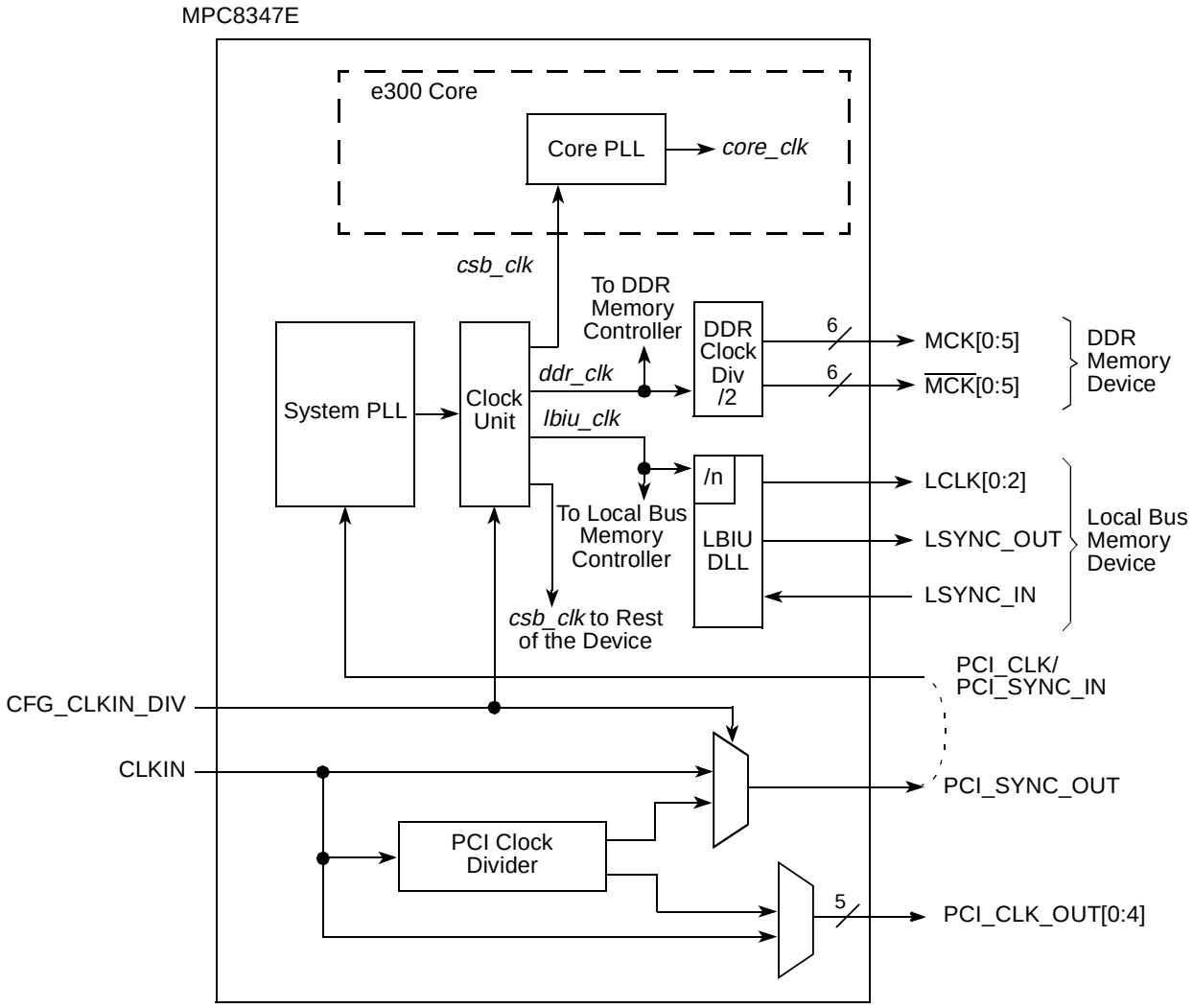

Figure 41. MPC8347E Clock Subsystem

The primary clock source can be one of two inputs, CLKIN or PCI_CLK, depending on whether the device is configured in PCI host or PCI agent mode. When the MPC8347E is configured as a PCI host device, CLKIN is its primary input clock. CLKIN feeds the PCI clock divider (÷2) and the multiplexors for PCI_SYNC_OUT and PCI_CLK_OUT. The CFG_CLKIN_DIV configuration input selects whether CLKIN or CLKIN/2 is driven out on the PCI_SYNC_OUT signal. The OCCR[PCICD*n*] parameters select whether CLKIN or CLKIN/2 is driven out on the PCI_CLK_OUT*n* signals.

PCI_SYNC_OUT is connected externally to PCI_SYNC_IN to allow the internal clock subsystem to synchronize to the system PCI clocks. PCI_SYNC_OUT must be connected properly to PCI_SYNC_IN, with equal delay to all PCI agent devices in the system, to allow the MPC8347E to function. When the MPC8347E is configured as a PCI agent device, PCI_CLK is the primary input clock and the CLKIN signal should be tied to GND.

19.1 System PLL Configuration

The system PLL is controlled by the RCWL[SPMF] parameter. Table 56 shows the multiplication factor encodings for the system PLL.

Table 56. System PLL Multiplication Factors

RCWL[SPMF]	System PLL Multiplication Factor
0000	× 16
0001	Reserved
0010	× 2
0011	× 3
0100	× 4
0101	× 5
0110	× 6
0111	× 7
1000	× 8
1001	× 9
1010	× 10
1011	× 11
1100	× 12
1101	× 13
1110	× 14
1111	× 15

As described in Section 19, “Clocking,” the LBIUCM, DDRCM, and SPMF parameters in the reset configuration word low and the CFG_CLKIN_DIV configuration input signal select the ratio between the primary clock input (CLKIN or PCI_CLK) and the internal coherent system bus clock (*csb_clk*). Table 57 and Table 58 show the expected frequency values for the CSB frequency for select *csb_clk* to CLKIN/PCI_SYNC_IN ratios.

many natural convection and especially closed box applications, the board temperature at the perimeter (edge) of the package is approximately the same as the local air temperature near the device. Specifying the local ambient conditions explicitly as the board temperature provides a more precise description of the local ambient conditions that determine the temperature of the device.

At a known board temperature, the junction temperature is estimated using the following equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

T_J = junction temperature (°C)
T_A = ambient temperature for the package (°C)
$R_{\theta JA}$ = junction-to-ambient thermal resistance (°C/W)
P_D = power dissipation in the package (W)

When the heat loss from the package case to the air can be ignored, acceptable predictions of junction temperature can be made. The application board should be similar to the thermal test condition: the component is soldered to a board with internal planes.

20.2.3 Experimental Determination of Junction Temperature

To determine the junction temperature of the device in the application after prototypes are available, use the thermal characterization parameter (Ψ_{JT}) to determine the junction temperature and a measure of the temperature at the top center of the package case using the following equation:

$$T_J = T_T + (\Psi_{JT} \times P_D)$$

where:

T_J = junction temperature (°C)
T_T = thermocouple temperature on top of package (°C)
Ψ_{JT} = junction-to-ambient thermal resistance (°C/W)
P_D = power dissipation in the package (W)

The thermal characterization parameter is measured per the JESD51-2 specification using a 40 gauge type T thermocouple epoxied to the top center of the package case. The thermocouple should be positioned so that the thermocouple junction rests on the package. A small amount of epoxy is placed over the thermocouple junction and over about 1 mm of wire extending from the junction. The thermocouple wire is placed flat against the package case to avoid measurement errors caused by cooling effects of the thermocouple wire.

20.2.4 Heat Sinks and Junction-to-Case Thermal Resistance

Some application environments require a heat sink to provide the necessary thermal management of the device. When a heat sink is used, the thermal resistance is expressed as the sum of a junction-to-case thermal resistance and a case-to-ambient thermal resistance:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$$

where:

$R_{\theta JA}$ = junction-to-ambient thermal resistance (°C/W)

$R_{\theta JC}$ = junction-to-case thermal resistance (°C/W)

$R_{\theta CA}$ = case-to-ambient thermal resistance (°C/W)

$R_{\theta JC}$ is device-related and cannot be influenced by the user. The user controls the thermal environment to change the case-to-ambient thermal resistance, $R_{\theta CA}$. For instance, the user can change the size of the heat sink, the air flow around the device, the interface material, the mounting arrangement on printed-circuit board, or change the thermal dissipation on the printed-circuit board surrounding the device.

The thermal performance of devices with heat sinks has been simulated with a few commercially available heat sinks. The heat sink choice is determined by the application environment (temperature, air flow, adjacent component power dissipation) and the physical space available. Because there is not a standard application environment, a standard heat sink is not required.

Table 63 and Table 64 show heat sink thermal resistance for TBGA and PBGA of the MPC8347E.

Table 63. Heat Sink and Thermal Resistance of MPC8347E (TBGA)

Heat Sink Assuming Thermal Grease	Air Flow	35 × 35 mm TBGA
		Thermal Resistance
AAVID 30 × 30 × 9.4 mm pin fin	Natural convection	10
AAVID 30 × 30 × 9.4 mm pin fin	1 m/s	6.5
AAVID 30 × 30 × 9.4 mm pin fin	2 m/s	5.6
AAVID 31 × 35 × 23 mm pin fin	Natural convection	8.4
AAVID 31 × 35 × 23 mm pin fin	1 m/s	4.7
AAVID 31 × 35 × 23 mm pin fin	2 m/s	4
Wakefield, 53 × 53 × 25 mm pin fin	Natural convection	5.7
Wakefield, 53 × 53 × 25 mm pin fin	1 m/s	3.5
Wakefield, 53 × 53 × 25 mm pin fin	2 m/s	2.7
MEI, 75 × 85 × 12 no adjacent board, extrusion	Natural convection	6.7
MEI, 75 × 85 × 12 no adjacent board, extrusion	1 m/s	4.1
MEI, 75 × 85 × 12 no adjacent board, extrusion	2 m/s	2.8
MEI, 75 × 85 × 12 mm, adjacent board, 40 mm side bypass	1 m/s	3.1

Table 64. Heat Sink and Thermal Resistance of MPC8347E (PBGA)

Heat Sink Assuming Thermal Grease	Air Flow	29 × 29 mm PBGA
		Thermal Resistance
AAVID 30 × 30 × 9.4 mm pin fin	Natural convection	13.5
AAVID 30 × 30 × 9.4 mm pin fin	1 m/s	9.6

21 System Design Information

This section provides electrical and thermal design recommendations for successful application of the MPC8347E.

21.1 System Clocking

The MPC8347E includes two PLLs:

1. The platform PLL generates the platform clock from the externally supplied CLKIN input. The frequency ratio between the platform and CLKIN is selected using the platform PLL ratio configuration bits as described in Section 19.1, “System PLL Configuration.”
2. The e300 core PLL generates the core clock as a slave to the platform clock. The frequency ratio between the e300 core clock and the platform clock is selected using the e300 PLL ratio configuration bits as described in Section 19.2, “Core PLL Configuration.”

21.2 PLL Power Supply Filtering

Each PLL gets power through independent power supply pins ($AV_{DD}1$, $AV_{DD}2$, respectively). The AV_{DD} level should always equal to V_{DD}, and preferably these voltages are derived directly from V_{DD} through a low frequency filter scheme.

There are a number of ways to provide power reliably to the PLLs, but the recommended solution is to provide four independent filter circuits as illustrated in Figure 42, one to each of the four AV_{DD} pins. Independent filters to each PLL reduce the opportunity to cause noise injection from one PLL to the other.

The circuit filters noise in the PLL resonant frequency range from 500 kHz to 10 MHz. It should be built with surface mount capacitors with minimum effective series inductance (ESL). Consistent with the recommendations of Dr. Howard Johnson in *High Speed Digital Design: A Handbook of Black Magic* (Prentice Hall, 1993), multiple small capacitors of equal value are recommended over a single large value capacitor.

To minimize noise coupled from nearby circuits, each circuit should be placed as closely as possible to the specific AV_{DD} pin being supplied. It should be possible to route directly from the capacitors to the AV_{DD} pin, which is on the periphery of package, without the inductance of vias.

Figure 42 shows the PLL power supply filter circuit.

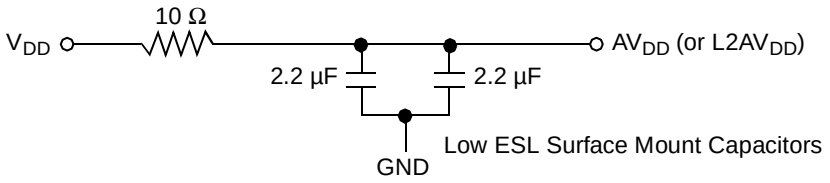

Figure 42. PLL Power Supply Filter Circuit