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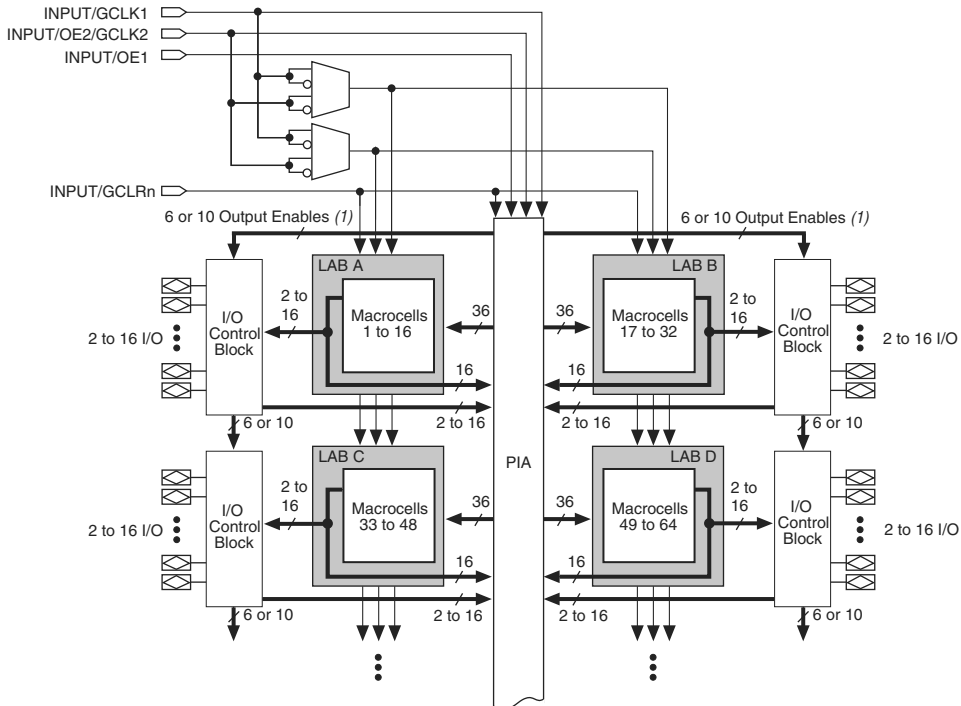
### Understanding [Embedded - CPLDs \(Complex Programmable Logic Devices\)](#)

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

### Applications of Embedded - CPLDs

#### Details

Product Status	Obsolete
Programmable Type	In System Programmable
Delay Time tpd(1) Max	10 ns
Voltage Supply - Internal	3V ~ 3.6V
Number of Logic Elements/Blocks	2
Number of Macrocells	32
Number of Gates	600
Number of I/O	34
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	44-LCC (J-Lead)
Supplier Device Package	44-PLCC (16.59x16.59)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/intel/epm3032alc44-10n">https://www.e-xfl.com/product-detail/intel/epm3032alc44-10n</a>

**Figure 1. MAX 3000A Device Block Diagram****Note:**

- (1) EPM3032A, EPM3064A, EPM3128A, and EPM3256A devices have six output enables. EPM3512A devices have 10 output enables.

**Logic Array Blocks**

The MAX 3000A device architecture is based on the linking of high-performance LABs. LABs consist of 16-macrocell arrays, as shown in Figure 1. Multiple LABs are linked together via the PIA, a global bus that is fed by all dedicated input pins, I/O pins, and macrocells.

Each LAB is fed by the following signals:

- 36 signals from the PIA that are used for general logic inputs
- Global controls that are used for secondary register functions

## Expander Product Terms

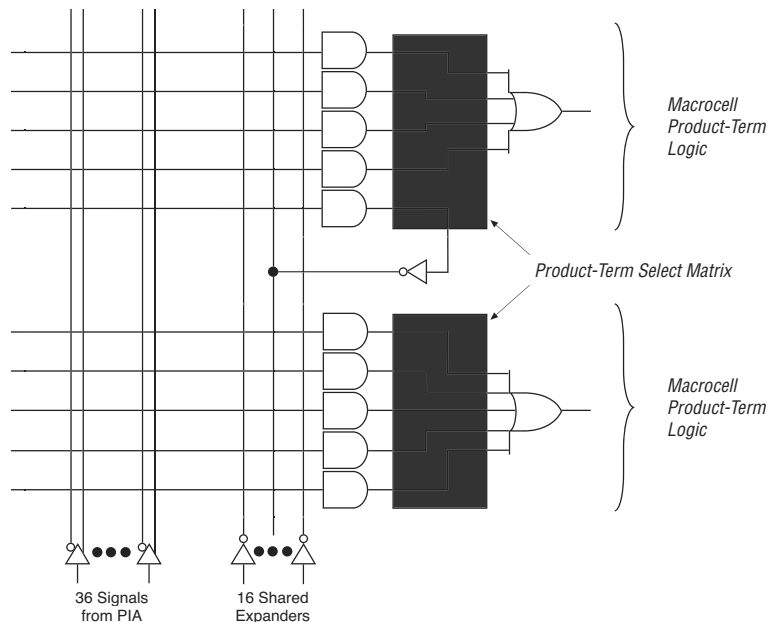
Although most logic functions can be implemented with the five product terms available in each macrocell, highly complex logic functions require additional product terms. Another macrocell can be used to supply the required logic resources. However, the MAX 3000A architecture also offers both shareable and parallel expander product terms ("expanders") that provide additional product terms directly to any macrocell in the same LAB. These expanders help ensure that logic is synthesized with the fewest possible logic resources to obtain the fastest possible speed.

### Shareable Expanders

Each LAB has 16 shareable expanders that can be viewed as a pool of uncommitted single product terms (one from each macrocell) with inverted outputs that feed back into the logic array. Each shareable expander can be used and shared by any or all macrocells in the LAB to build complex logic functions. Shareable expanders incur a small delay ( $t_{SEXP}$ ). Figure 3 shows how shareable expanders can feed multiple macrocells.

**Figure 3. MAX 3000A Shareable Expanders**

*Shareable expanders can be shared by any or all macrocells in an LAB.*

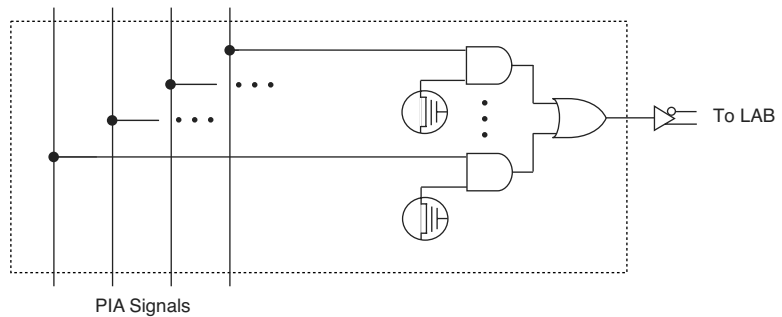


### *Parallel Expanders*

Parallel expanders are unused product terms that can be allocated to a neighboring macrocell to implement fast, complex logic functions. Parallel expanders allow up to 20 product terms to directly feed the macrocell OR logic, with five product terms provided by the macrocell and 15 parallel expanders provided by neighboring macrocells in the LAB.

The Altera development system compiler can automatically allocate up to three sets of up to five parallel expanders to the macrocells that require additional product terms. Each set of five parallel expanders incurs a small, incremental timing delay ( $t_{PEXP}$ ). For example, if a macrocell requires 14 product terms, the compiler uses the five dedicated product terms within the macrocell and allocates two sets of parallel expanders; the first set includes five product terms, and the second set includes four product terms, increasing the total delay by  $2 \times t_{PEXP}$ .

Two groups of eight macrocells within each LAB (e.g., macrocells 1 through 8 and 9 through 16) form two chains to lend or borrow parallel expanders. A macrocell borrows parallel expanders from lower-numbered macrocells. For example, macrocell 8 can borrow parallel expanders from macrocell 7, from macrocells 7 and 6, or from macrocells 7, 6, and 5. Within each group of eight, the lowest-numbered macrocell can only lend parallel expanders and the highest-numbered macrocell can only borrow them. Figure 4 shows how parallel expanders can be borrowed from a neighboring macrocell.

**Figure 5. MAX 3000A PIA Routing**

While the routing delays of channel-based routing schemes in masked or FPGAs are cumulative, variable, and path-dependent, the MAX 3000A PIA has a predictable delay. The PIA makes a design's timing performance easy to predict.

### I/O Control Blocks

The I/O control block allows each I/O pin to be individually configured for input, output, or bidirectional operation. All I/O pins have a tri-state buffer that is individually controlled by one of the global output enable signals or directly connected to ground or  $V_{CC}$ . Figure 6 shows the I/O control block for MAX 3000A devices. The I/O control block has 6 or 10 global output enable signals that are driven by the true or complement of two output enable signals, a subset of the I/O pins, or a subset of the I/O macrocells.

## In-System Programmability

MAX 3000A devices can be programmed in-system via an industry-standard four-pin IEEE Std. 1149.1-1990 (JTAG) interface. In-system programmability (ISP) offers quick, efficient iterations during design development and debugging cycles. The MAX 3000A architecture internally generates the high programming voltages required to program its EEPROM cells, allowing in-system programming with only a single 3.3-V power supply. During in-system programming, the I/O pins are tri-stated and weakly pulled-up to eliminate board conflicts. The pull-up value is nominally 50 k $\Omega$ .

MAX 3000A devices have an enhanced ISP algorithm for faster programming. These devices also offer an `ISP_Done` bit that ensures safe operation when in-system programming is interrupted. This `ISP_Done` bit, which is the last bit programmed, prevents all I/O pins from driving until the bit is programmed.

ISP simplifies the manufacturing flow by allowing devices to be mounted on a printed circuit board (PCB) with standard pick-and-place equipment before they are programmed. MAX 3000A devices can be programmed by downloading the information via in-circuit testers, embedded processors, the MasterBlaster communications cable, the ByteBlasterMV parallel port download cable, and the BitBlaster serial download cable. Programming the devices after they are placed on the board eliminates lead damage on high-pin-count packages (e.g., QFP packages) due to device handling. MAX 3000A devices can be reprogrammed after a system has already shipped to the field. For example, product upgrades can be performed in the field via software or modem.

The Jam STAPL programming and test language can be used to program MAX 3000A devices with in-circuit testers, PCs, or embedded processors.



For more information on using the Jam STAPL programming and test language, see *Application Note 88 (Using the Jam Language for ISP & ICR via an Embedded Processor)*, *Application Note 122 (Using Jam STAPL for ISP & ICR via an Embedded Processor)* and *AN 111 (Embedded Programming Using the 8051 and Jam Byte-Code)*.

The ISP circuitry in MAX 3000A devices is compliant with the IEEE Std. 1532 specification. The IEEE Std. 1532 is a standard developed to allow concurrent ISP between multiple PLD vendors.

## Programming Sequence

During in-system programming, instructions, addresses, and data are shifted into the MAX 3000A device through the TDI input pin. Data is shifted out through the TDO output pin and compared against the expected data.

Programming a pattern into the device requires the following six ISP stages. A stand-alone verification of a programmed pattern involves only stages 1, 2, 5, and 6.

1. *Enter ISP.* The enter ISP stage ensures that the I/O pins transition smoothly from user mode to ISP mode. The enter ISP stage requires 1 ms.
2. *Check ID.* Before any program or verify process, the silicon ID is checked. The time required to read this silicon ID is relatively small compared to the overall programming time.
3. *Bulk Erase.* Erasing the device in-system involves shifting in the instructions to erase the device and applying one erase pulse of 100 ms.
4. *Program.* Programming the device in-system involves shifting in the address and data and then applying the programming pulse to program the EEPROM cells. This process is repeated for each EEPROM address.
5. *Verify.* Verifying an Altera device in-system involves shifting in addresses, applying the read pulse to verify the EEPROM cells, and shifting out the data for comparison. This process is repeated for each EEPROM address.
6. *Exit ISP.* An exit ISP stage ensures that the I/O pins transition smoothly from ISP mode to user mode. The exit ISP stage requires 1 ms.

## Programming Times

The time required to implement each of the six programming stages can be broken into the following two elements:

- A pulse time to erase, program, or read the EEPROM cells.
- A shifting time based on the test clock (TCK) frequency and the number of TCK cycles to shift instructions, address, and data into the device.

## Programming with External Hardware

MAX 3000A devices can be programmed on Windows-based PCs with an Altera Logic Programmer card, MPU, and the appropriate device adapter. The MPU performs continuity checking to ensure adequate electrical contact between the adapter and the device.



For more information, see the *Altera Programming Hardware Data Sheet*.

The Altera software can use text- or waveform-format test vectors created with the Altera Text Editor or Waveform Editor to test the programmed device. For added design verification, designers can perform functional testing to compare the functional device behavior with the results of simulation.

Data I/O, BP Microsystems, and other programming hardware manufacturers also provide programming support for Altera devices.



For more information, see *Programming Hardware Manufacturers*.

## IEEE Std. 1149.1 (JTAG) Boundary-Scan Support

MAX 3000A devices include the JTAG BST circuitry defined by IEEE Std. 1149.1–1990. Table 7 describes the JTAG instructions supported by MAX 3000A devices. The pin-out tables found on the Altera web site (<http://www.altera.com>) or the *Altera Digital Library* show the location of the JTAG control pins for each device. If the JTAG interface is not required, the JTAG pins are available as user I/O pins.

**Table 7. MAX 3000A JTAG Instructions**

JTAG Instruction	Description
SAMPLE/PRELOAD	Allows a snapshot of signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern output at the device pins
EXTEST	Allows the external circuitry and board-level interconnections to be tested by forcing a test pattern at the output pins and capturing test results at the input pins
BYPASS	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through a selected device to adjacent devices during normal device operation
IDCODE	Selects the IDCODE register and places it between the TDI and TDO pins, allowing the IDCODE to be serially shifted out of TDO
USERCODE	Selects the 32-bit USERCODE register and places it between the TDI and TDO pins, allowing the USERCODE value to be shifted out of TDO
ISP Instructions	These instructions are used when programming MAX 3000A devices via the JTAG ports with the MasterBlaster, ByteBlasterMV, or BitBlaster cable, or when using a Jam STAPL file, JBC file, or SVF file via an embedded processor or test equipment



Figure 7 shows the timing information for the JTAG signals.

**Figure 7. MAX 3000A JTAG Waveforms**

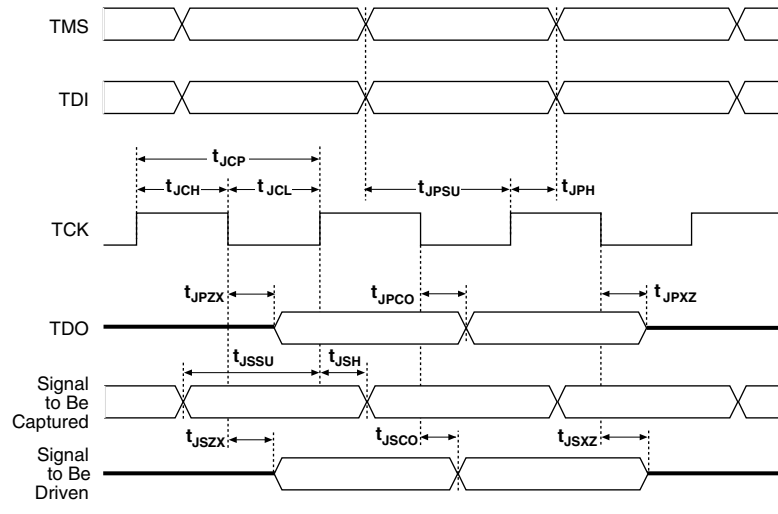


Table 10 shows the JTAG timing parameters and values for MAX 3000A devices.

<b>Table 10. JTAG Timing Parameters &amp; Values for MAX 3000A Devices</b>				
<b>Symbol</b>	<b>Parameter</b>	<b>Min</b>	<b>Max</b>	<b>Unit</b>
$t_{JCP}$	TCK clock period	100		ns
$t_{JCH}$	TCK clock high time	50		ns
$t_{JCL}$	TCK clock low time	50		ns
$t_{JPSU}$	JTAG port setup time	20		ns
$t_{JPH}$	JTAG port hold time	45		ns
$t_{JPCO}$	JTAG port clock to output		25	ns
$t_{JPZX}$	JTAG port high impedance to valid output		25	ns
$t_{JPXZ}$	JTAG port valid output to high impedance		25	ns
$t_{JSSU}$	Capture register setup time	20		ns
$t_{JSH}$	Capture register hold time	45		ns
$t_{JSCO}$	Update register clock to output		25	ns
$t_{JSZX}$	Update register high impedance to valid output		25	ns
$t_{JSXZ}$	Update register valid output to high impedance		25	ns

**Table 15. MAX 3000A Device Capacitance** Note (9)

Symbol	Parameter	Conditions	Min	Max	Unit
$C_{IN}$	Input pin capacitance	$V_{IN} = 0\text{ V}$ , $f = 1.0\text{ MHz}$		8	pF
$C_{I/O}$	I/O pin capacitance	$V_{OUT} = 0\text{ V}$ , $f = 1.0\text{ MHz}$		8	pF

**Notes to tables:**

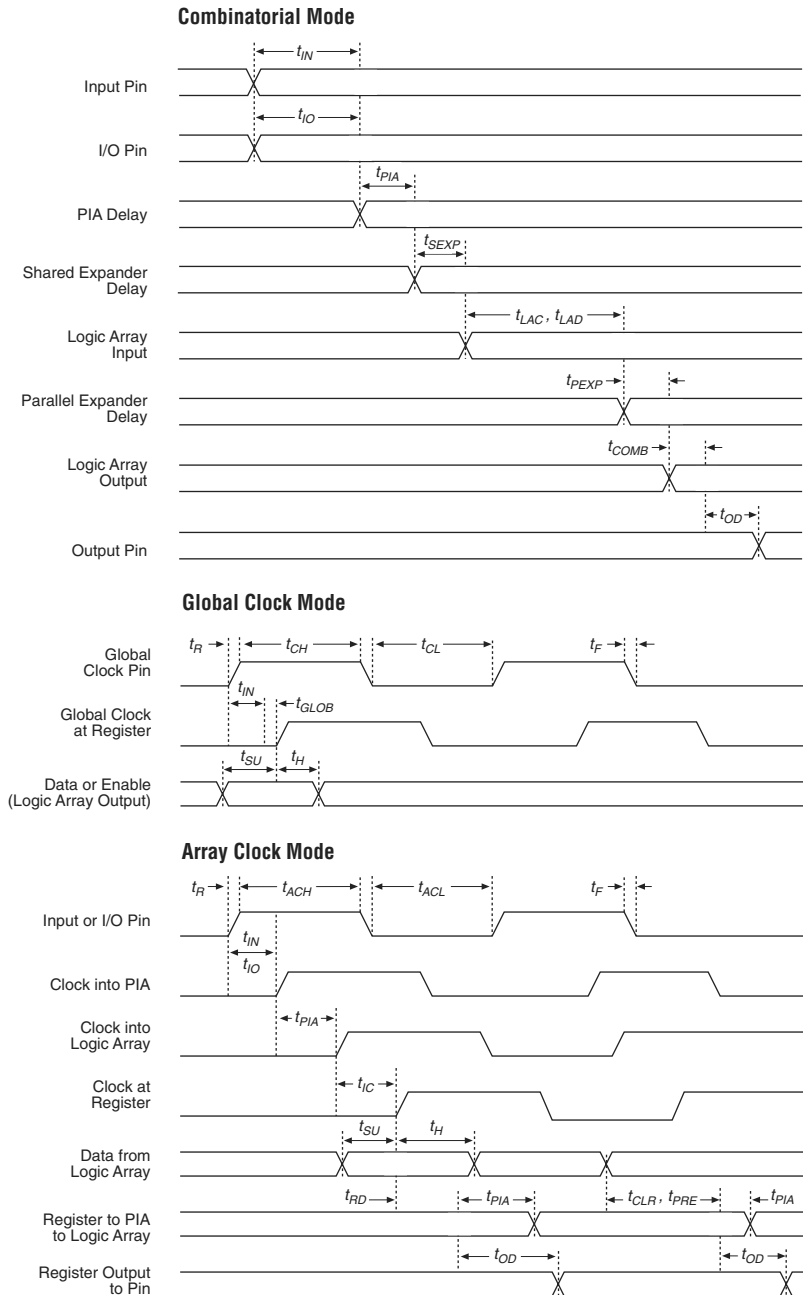
- (1) See the *Operating Requirements for Altera Devices Data Sheet*.
- (2) Minimum DC input voltage is  $-0.5\text{ V}$ . During transitions, the inputs may undershoot to  $-2.0\text{ V}$  or overshoot to  $5.75\text{ V}$  for input currents less than  $100\text{ mA}$  and periods shorter than  $20\text{ ns}$ .
- (3) All pins, including dedicated inputs, I/O pins, and JTAG pins, may be driven before  $V_{CCINT}$  and  $V_{CCIO}$  are powered.
- (4) These values are specified under the recommended operating conditions, as shown in Table 13 on page 23.
- (5) The parameter is measured with 50% of the outputs each sourcing the specified current. The  $I_{OH}$  parameter refers to high-level TTL or CMOS output current.
- (6) The parameter is measured with 50% of the outputs each sinking the specified current. The  $I_{OL}$  parameter refers to low-level TTL, PCI, or CMOS output current.
- (7) This value is specified during normal device operation. During power-up, the maximum leakage current is  $\pm 300\text{ }\mu\text{A}$ .
- (8) This pull-up exists while devices are programmed in-system and in unprogrammed devices during power-up.
- (9) Capacitance is measured at  $25^{\circ}\text{ C}$  and is sample-tested only. The  $\text{OE1}$  pin (high-voltage pin during programming) has a maximum capacitance of  $20\text{ pF}$ .
- (10) The POR time for all MAX 3000A devices does not exceed  $100\text{ }\mu\text{s}$ . The sufficient  $V_{CCINT}$  voltage level for POR is  $3.0\text{ V}$ . The device is fully initialized within the POR time after  $V_{CCINT}$  reaches the sufficient POR voltage level.
- (11) These devices support in-system programming for  $-40^{\circ}$  to  $100^{\circ}\text{ C}$ . For in-system programming support between  $-40^{\circ}$  and  $0^{\circ}\text{ C}$ , contact Altera Applications.

Figure 9 shows the typical output drive characteristics of MAX 3000A devices.



**Figure 11. MAX 3000A Switching Waveforms**

$t_R$  &  $t_F < 2$  ns. Inputs are driven at 3 V for a logic high and 0 V for a logic low. All timing characteristics are measured at 1.5 V.



**Table 17. EPM3032A Internal Timing Parameters (Part 1 of 2)** *Note (1)*

Symbol	Parameter	Conditions	Speed Grade						Unit
			-4		-7		-10		
			Min	Max	Min	Max	Min	Max	
$t_{IN}$	Input pad and buffer delay			0.7		1.2		1.5	ns
$t_{IO}$	I/O input pad and buffer delay			0.7		1.2		1.5	ns
$t_{SEXP}$	Shared expander delay			1.9		3.1		4.0	ns
$t_{PEXP}$	Parallel expander delay			0.5		0.8		1.0	ns
$t_{LAD}$	Logic array delay			1.5		2.5		3.3	ns
$t_{LAC}$	Logic control array delay			0.6		1.0		1.2	ns
$t_{IOE}$	Internal output enable delay			0.0		0.0		0.0	ns
$t_{OD1}$	Output buffer and pad delay, slow slew rate = off $V_{CCIO} = 3.3\text{ V}$	$C1 = 35\text{ pF}$		0.8		1.3		1.8	ns
$t_{OD2}$	Output buffer and pad delay, slow slew rate = off $V_{CCIO} = 2.5\text{ V}$	$C1 = 35\text{ pF}$		1.3		1.8		2.3	ns
$t_{OD3}$	Output buffer and pad delay, slow slew rate = on $V_{CCIO} = 2.5\text{ V}$ or $3.3\text{ V}$	$C1 = 35\text{ pF}$		5.8		6.3		6.8	ns
$t_{ZX1}$	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 3.3\text{ V}$	$C1 = 35\text{ pF}$		4.0		4.0		5.0	ns
$t_{ZX2}$	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 2.5\text{ V}$	$C1 = 35\text{ pF}$		4.5		4.5		5.5	ns
$t_{ZX3}$	Output buffer enable delay, slow slew rate = on $V_{CCIO} = 2.5\text{ V}$ or $3.3\text{ V}$	$C1 = 35\text{ pF}$		9.0		9.0		10.0	ns
$t_{XZ}$	Output buffer disable delay	$C1 = 5\text{ pF}$		4.0		4.0		5.0	ns
$t_{SU}$	Register setup time		1.3		2.0		2.8		ns
$t_H$	Register hold time		0.6		1.0		1.3		ns
$t_{RD}$	Register delay			0.7		1.2		1.5	ns
$t_{COMB}$	Combinatorial delay			0.6		1.0		1.3	ns
$t_{IC}$	Array clock delay			1.2		2.0		2.5	ns
$t_{EN}$	Register enable time			0.6		1.0		1.2	ns
$t_{GLOB}$	Global control delay			0.8		1.3		1.9	ns
$t_{PRE}$	Register preset time			1.2		1.9		2.6	ns
$t_{CLR}$	Register clear time			1.2		1.9		2.6	ns

**Table 17. EPM3032A Internal Timing Parameters (Part 2 of 2)** *Note (1)*

Symbol	Parameter	Conditions	Speed Grade						Unit
			−4		−7		−10		
			Min	Max	Min	Max	Min	Max	
$t_{PIA}$	PIA delay	(2)		0.9		1.5		2.1	ns
$t_{LPA}$	Low-power adder	(5)		2.5		4.0		5.0	ns

**Table 18. EPM3064A External Timing Parameters** *Note (1)*

Symbol	Parameter	Conditions	Speed Grade						Unit
			−4		−7		−10		
			Min	Max	Min	Max	Min	Max	
t <sub>PD1</sub>	Input to non-registered output	C1 = 35 pF (2)		4.5		7.5		10.0	ns
t <sub>PD2</sub>	I/O input to non-registered output	C1 = 35 pF (2)		4.5		7.5		10.0	ns
t <sub>SU</sub>	Global clock setup time	(2)	2.8		4.7		6.2		ns
t <sub>H</sub>	Global clock hold time	(2)	0.0		0.0		0.0		ns
t <sub>CO1</sub>	Global clock to output delay	C1 = 35 pF	1.0	3.1	1.0	5.1	1.0	7.0	ns
t <sub>CH</sub>	Global clock high time		2.0		3.0		4.0		ns
t <sub>CL</sub>	Global clock low time		2.0		3.0		4.0		ns
t <sub>ASU</sub>	Array clock setup time	(2)	1.6		2.6		3.6		ns
t <sub>AH</sub>	Array clock hold time	(2)	0.3		0.4		0.6		ns
t <sub>ACO1</sub>	Array clock to output delay	C1 = 35 pF (2)	1.0	4.3	1.0	7.2	1.0	9.6	ns
t <sub>ACH</sub>	Array clock high time		2.0		3.0		4.0		ns
t <sub>ACL</sub>	Array clock low time		2.0		3.0		4.0		ns
t <sub>CPPW</sub>	Minimum pulse width for clear and preset	(3)	2.0		3.0		4.0		ns
t <sub>CNT</sub>	Minimum global clock period	(2)		4.5		7.4		10.0	ns
f <sub>CNT</sub>	Maximum internal global clock frequency	(2), (4)	222.2		135.1		100.0		MHz
t <sub>ACNT</sub>	Minimum array clock period	(2)		4.5		7.4		10.0	ns
f <sub>ACNT</sub>	Maximum internal array clock frequency	(2), (4)	222.2		135.1		100.0		MHz

**Table 19. EPM3064A Internal Timing Parameters (Part 1 of 2)** *Note (1)*

Symbol	Parameter	Conditions	Speed Grade						Unit
			−4		−7		−10		
			Min	Max	Min	Max	Min	Max	
$t_{IN}$	Input pad and buffer delay			0.6		1.1		1.4	ns
$t_{IO}$	I/O input pad and buffer delay			0.6		1.1		1.4	ns
$t_{SEXP}$	Shared expander delay			1.8		3.0		3.9	ns
$t_{PEXP}$	Parallel expander delay			0.4		0.7		0.9	ns
$t_{LAD}$	Logic array delay			1.5		2.5		3.2	ns
$t_{LAC}$	Logic control array delay			0.6		1.0		1.2	ns
$t_{IOE}$	Internal output enable delay			0.0		0.0		0.0	ns
$t_{OD1}$	Output buffer and pad delay, slow slew rate = off $V_{CCIO} = 3.3\text{ V}$	$C1 = 35\text{ pF}$		0.8		1.3		1.8	ns
$t_{OD2}$	Output buffer and pad delay, slow slew rate = off $V_{CCIO} = 2.5\text{ V}$	$C1 = 35\text{ pF}$		1.3		1.8		2.3	ns
$t_{OD3}$	Output buffer and pad delay, slow slew rate = on $V_{CCIO} = 2.5\text{ V}$ or $3.3\text{ V}$	$C1 = 35\text{ pF}$		5.8		6.3		6.8	ns
$t_{ZX1}$	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 3.3\text{ V}$	$C1 = 35\text{ pF}$		4.0		4.0		5.0	ns
$t_{ZX2}$	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 2.5\text{ V}$	$C1 = 35\text{ pF}$		4.5		4.5		5.5	ns
$t_{ZX3}$	Output buffer enable delay, slow slew rate = on $V_{CCIO} = 2.5\text{ V}$ or $3.3\text{ V}$	$C1 = 35\text{ pF}$		9.0		9.0		10.0	ns
$t_{XZ}$	Output buffer disable delay	$C1 = 5\text{ pF}$		4.0		4.0		5.0	ns
$t_{SU}$	Register setup time		1.3		2.0		2.9		ns
$t_H$	Register hold time		0.6		1.0		1.3		ns
$t_{RD}$	Register delay			0.7		1.2		1.6	ns
$t_{COMB}$	Combinatorial delay			0.6		0.9		1.3	ns
$t_{IC}$	Array clock delay			1.2		1.9		2.5	ns
$t_{EN}$	Register enable time			0.6		1.0		1.2	ns
$t_{GLOB}$	Global control delay			1.0		1.5		2.2	ns
$t_{PRE}$	Register preset time			1.3		2.1		2.9	ns

**Table 25. EPM3512A Internal Timing Parameters (Part 2 of 2)** *Note (1)*

Symbol	Parameter	Conditions	Speed Grade				Unit
			-7		-10		
			Min	Max	Min	Max	
$t_{OD3}$	Output buffer and pad delay, slow slew rate = on $V_{CCIO} = 2.5\text{ V}$ or $3.3\text{ V}$	$C1 = 35\text{ pF}$		6.0		6.5	ns
$t_{ZX1}$	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 3.3\text{ V}$	$C1 = 35\text{ pF}$		4.0		5.0	ns
$t_{ZX2}$	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 2.5\text{ V}$	$C1 = 35\text{ pF}$		4.5		5.5	ns
$t_{ZX3}$	Output buffer enable delay, slow slew rate = on $V_{CCIO} = 3.3\text{ V}$	$C1 = 35\text{ pF}$		9.0		10.0	ns
$t_{XZ}$	Output buffer disable delay	$C1 = 5\text{ pF}$		4.0		5.0	ns
$t_{SU}$	Register setup time		2.1		3.0		ns
$t_H$	Register hold time		0.6		0.8		ns
$t_{FSU}$	Register setup time of fast input		1.6		1.6		ns
$t_{FH}$	Register hold time of fast input		1.4		1.4		ns
$t_{RD}$	Register delay			1.3		1.7	ns
$t_{COMB}$	Combinatorial delay			0.6		0.8	ns
$t_{IC}$	Array clock delay			1.8		2.3	ns
$t_{EN}$	Register enable time			1.0		1.3	ns
$t_{GLOB}$	Global control delay			1.7		2.2	ns
$t_{PRE}$	Register preset time			1.0		1.4	ns
$t_{CLR}$	Register clear time			1.0		1.4	ns
$t_{PIA}$	PIA delay	(2)		3.0		4.0	ns
$t_{LPA}$	Low-power adder	(5)		4.5		5.0	ns

**Notes to tables:**

- These values are specified under the recommended operating conditions, as shown in Table 13 on page 23. See Figure 11 on page 27 for more information on switching waveforms.
- These values are specified for a PIA fan-out of one LAB (16 macrocells). For each additional LAB fan-out in these devices, add an additional 0.1 ns to the PIA timing value.
- This minimum pulse width for preset and clear applies for both global clear and array controls. The  $t_{LPA}$  parameter must be added to this minimum width if the clear or reset signal incorporates the  $t_{LAD}$  parameter into the signal path.
- These parameters are measured with a 16-bit loadable, enabled, up/down counter programmed into each LAB.
- The  $t_{LPA}$  parameter must be added to the  $t_{LAD}$ ,  $t_{LAC}$ ,  $t_{IC}$ ,  $t_{EN}$ ,  $t_{SEXP}$ ,  $t_{ACL}$ , and  $t_{CPPW}$  parameters for macrocells running in low-power mode.



## Power Consumption

Supply power (P) versus frequency ( $f_{\text{MAX}}$ , in MHz) for MAX 3000A devices is calculated with the following equation:

$$P = P_{\text{INT}} + P_{\text{IO}} = I_{\text{CCINT}} \times V_{\text{CC}} + P_{\text{IO}}$$

The  $P_{\text{IO}}$  value, which depends on the device output load characteristics and switching frequency, can be calculated using the guidelines given in *Application Note 74 (Evaluating Power for Altera Devices)*.

The  $I_{\text{CCINT}}$  value depends on the switching frequency and the application logic. The  $I_{\text{CCINT}}$  value is calculated with the following equation:

$$I_{\text{CCINT}} =$$

$$(A \times \text{MC}_{\text{TON}}) + [B \times (\text{MC}_{\text{DEV}} - \text{MC}_{\text{TON}})] + (C \times \text{MC}_{\text{USED}} \times f_{\text{MAX}} \times \text{tog}_{\text{LC}})$$

The parameters in the  $I_{\text{CCINT}}$  equation are:

- $\text{MC}_{\text{TON}}$  = Number of macrocells with the Turbo Bit™ option turned on, as reported in the Quartus II or MAX+PLUS II Report File (.rpt)
- $\text{MC}_{\text{DEV}}$  = Number of macrocells in the device
- $\text{MC}_{\text{USED}}$  = Total number of macrocells in the design, as reported in the RPT File
- $f_{\text{MAX}}$  = Highest clock frequency to the device
- $\text{tog}_{\text{LC}}$  = Average percentage of logic cells toggling at each clock (typically 12.5%)
- A, B, C = Constants (shown in Table 26)

**Table 26. MAX 3000A  $I_{\text{CC}}$  Equation Constants**

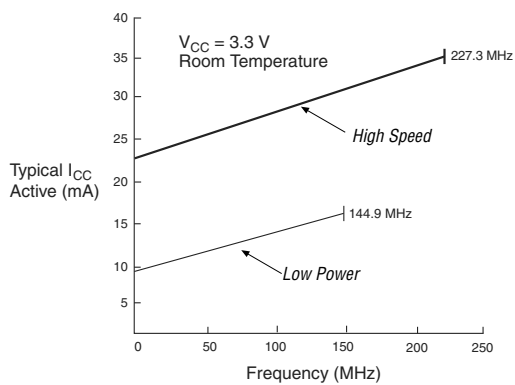
Device	A	B	C
EPM3032A	0.71	0.30	0.014
EPM3064A	0.71	0.30	0.014
EPM3128A	0.71	0.30	0.014
EPM3256A	0.71	0.30	0.014
EPM3512A	0.71	0.30	0.014

The  $I_{\text{CCINT}}$  calculation provides an  $I_{\text{CC}}$  estimate based on typical conditions using a pattern of a 16-bit, loadable, enabled, up/down counter in each LAB with no output load. Actual  $I_{\text{CC}}$  should be verified during operation because this measurement is sensitive to the actual pattern in the device and the environmental operating conditions.

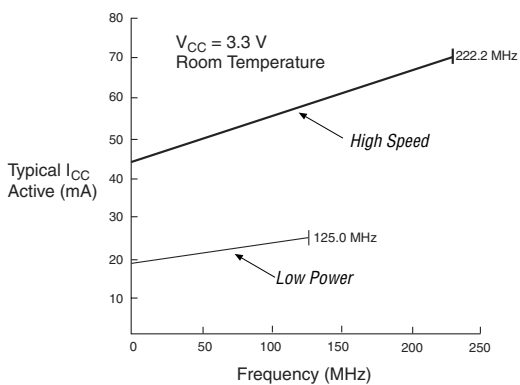
Figures 12 and 13 show the typical supply current versus frequency for MAX 3000A devices.

Figure 12.  $I_{CC}$  vs. Frequency for MAX 3000A Devices

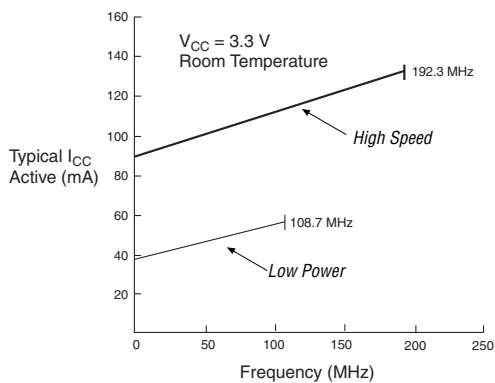
EPM3032A

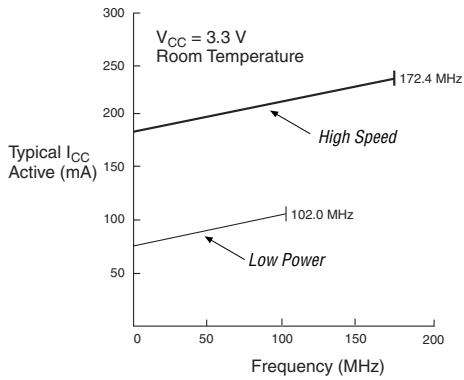
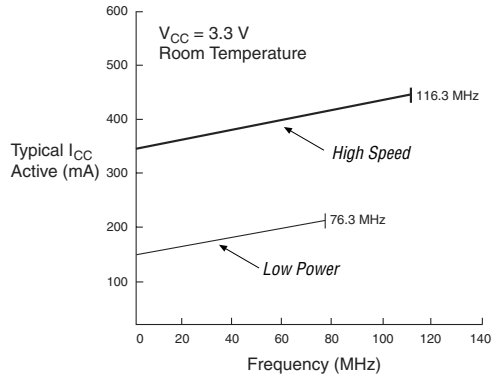


EPM3064A



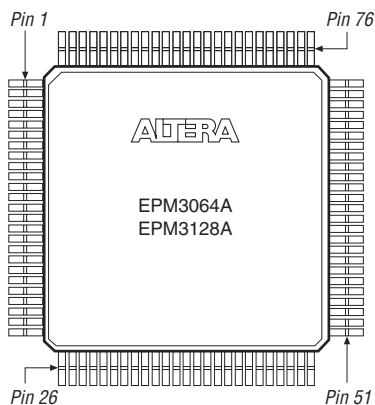
EPM3128A



**Figure 13.  $I_{CC}$  vs. Frequency for MAX 3000A Devices****EPM3256A****EPM3512A**

**Figure 15. 100-Pin TQFP Package Pin-Out Diagram**

Package outline not drawn to scale.

**Figure 16. 144-Pin TQFP Package Pin-Out Diagram**

Package outline not drawn to scale.

