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Understanding [Embedded - CPLDs \(Complex Programmable Logic Devices\)](#)

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

Applications of Embedded - CPLDs

Details

Product Status	Obsolete
Programmable Type	In System Programmable
Delay Time tpd(1) Max	10 ns
Voltage Supply - Internal	3V ~ 3.6V
Number of Logic Elements/Blocks	4
Number of Macrocells	64
Number of Gates	1250
Number of I/O	34
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	44-TQFP
Supplier Device Package	44-TQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epm3064ati44-10n

...and More Features

- PCI compatible
- Bus-friendly architecture including programmable slew-rate control
- Open-drain output option
- Programmable macrocell flipflops with individual clear, preset, clock, and clock enable controls
- Programmable power-saving mode for a power reduction of over 50% in each macrocell
- Configurable expander product-term distribution, allowing up to 32 product terms per macrocell
- Programmable security bit for protection of proprietary designs
- Enhanced architectural features, including:
 - 6 or 10 pin- or logic-driven output enable signals
 - Two global clock signals with optional inversion
 - Enhanced interconnect resources for improved routability
 - Programmable output slew-rate control
- Software design support and automatic place-and-route provided by Altera's development systems for Windows-based PCs and Sun SPARCstations, and HP 9000 Series 700/800 workstations
- Additional design entry and simulation support provided by EDIF 2.0.0 and 3.0.0 netlist files, library of parameterized modules (LPM), Verilog HDL, VHDL, and other interfaces to popular EDA tools from third-party manufacturers such as Cadence, Exemplar Logic, Mentor Graphics, OrCAD, Synopsys, Synplicity, and VeriBest
- Programming support with the Altera master programming unit (MPU), MasterBlaster™ communications cable, ByteBlasterMV™ parallel port download cable, BitBlaster™ serial download cable as well as programming hardware from third-party manufacturers and any in-circuit tester that supports Jam™ Standard Test and Programming Language (STAPL) Files (.jam), Jam STAPL Byte-Code Files (.jbc), or Serial Vector Format Files (.svf)

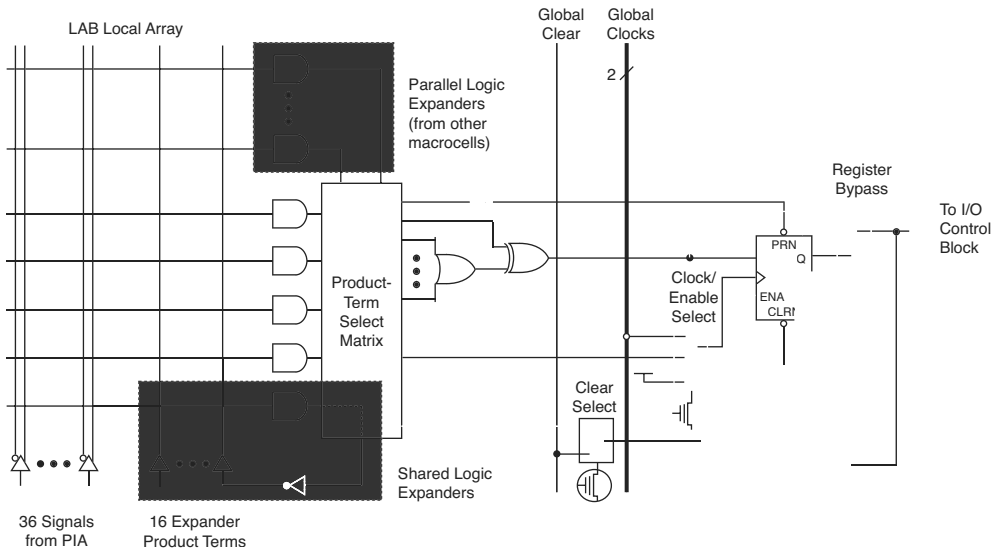
General Description

MAX 3000A devices are low-cost, high-performance devices based on the Altera MAX architecture. Fabricated with advanced CMOS technology, the EEPROM-based MAX 3000A devices operate with a 3.3-V supply voltage and provide 600 to 10,000 usable gates, ISP, pin-to-pin delays as fast as 4.5 ns, and counter speeds of up to 227.3 MHz. MAX 3000A devices in the -4, -5, -6, -7, and -10 speed grades are compatible with the timing requirements of the PCI Special Interest Group (PCI SIG) *PCI Local Bus Specification, Revision 2.2*. See Table 2.

Macrocells

MAX 3000A macrocells can be individually configured for either sequential or combinatorial logic operation. Macrocells consist of three functional blocks: logic array, product-term select matrix, and programmable register. Figure 2 shows a MAX 3000A macrocell.

Figure 2. MAX 3000A Macrocell



Combinatorial logic is implemented in the logic array, which provides five product terms per macrocell. The product-term select matrix allocates these product terms for use as either primary logic inputs (to the OR and XOR gates) to implement combinatorial functions, or as secondary inputs to the macrocell's register preset, clock, and clock enable control functions.

Two kinds of expander product terms ("expanders") are available to supplement macrocell logic resources:

- Shareable expanders, which are inverted product terms that are fed back into the logic array
- Parallel expanders, which are product terms borrowed from adjacent macrocells

The Altera development system automatically optimizes product-term allocation according to the logic requirements of the design.

For registered functions, each macrocell flipflop can be individually programmed to implement D, T, JK, or SR operation with programmable clock control. The flipflop can be bypassed for combinatorial operation. During design entry, the designer specifies the desired flipflop type; the Altera development system software then selects the most efficient flipflop operation for each registered function to optimize resource utilization.

Each programmable register can be clocked in three different modes:

- Global clock signal mode, which achieves the fastest clock-to-output performance.
- Global clock signal enabled by an active-high clock enable. A clock enable is generated by a product term. This mode provides an enable on each flipflop while still achieving the fast clock-to-output performance of the global clock.
- Array clock implemented with a product term. In this mode, the flipflop can be clocked by signals from buried macrocells or I/O pins.

Two global clock signals are available in MAX 3000A devices. As shown in Figure 1, these global clock signals can be the true or the complement of either of the two global clock pins, GCLK1 or GCLK2.

Each register also supports asynchronous preset and clear functions. As shown in Figure 2, the product-term select matrix allocates product terms to control these operations. Although the product-term-driven preset and clear from the register are active high, active-low control can be obtained by inverting the signal within the logic array. In addition, each register clear function can be individually driven by the active-low dedicated global clear pin (GCLRn).

All registers are cleared upon power-up. By default, all registered outputs drive low when the device is powered up. You can set the registered outputs to drive high upon power-up through the Quartus® II software. Quartus II software uses the NOT Gate Push-Back method, which uses an additional macrocell to set the output high. To set this in the Quartus II software, go to the Assignment Editor and set the **Power-Up Level** assignment for the register to **High**.

Expander Product Terms

Although most logic functions can be implemented with the five product terms available in each macrocell, highly complex logic functions require additional product terms. Another macrocell can be used to supply the required logic resources. However, the MAX 3000A architecture also offers both shareable and parallel expander product terms ("expanders") that provide additional product terms directly to any macrocell in the same LAB. These expanders help ensure that logic is synthesized with the fewest possible logic resources to obtain the fastest possible speed.

Shareable Expanders

Each LAB has 16 shareable expanders that can be viewed as a pool of uncommitted single product terms (one from each macrocell) with inverted outputs that feed back into the logic array. Each shareable expander can be used and shared by any or all macrocells in the LAB to build complex logic functions. Shareable expanders incur a small delay (t_{SEXP}). Figure 3 shows how shareable expanders can feed multiple macrocells.

Figure 3. MAX 3000A Shareable Expanders

Shareable expanders can be shared by any or all macrocells in an LAB.

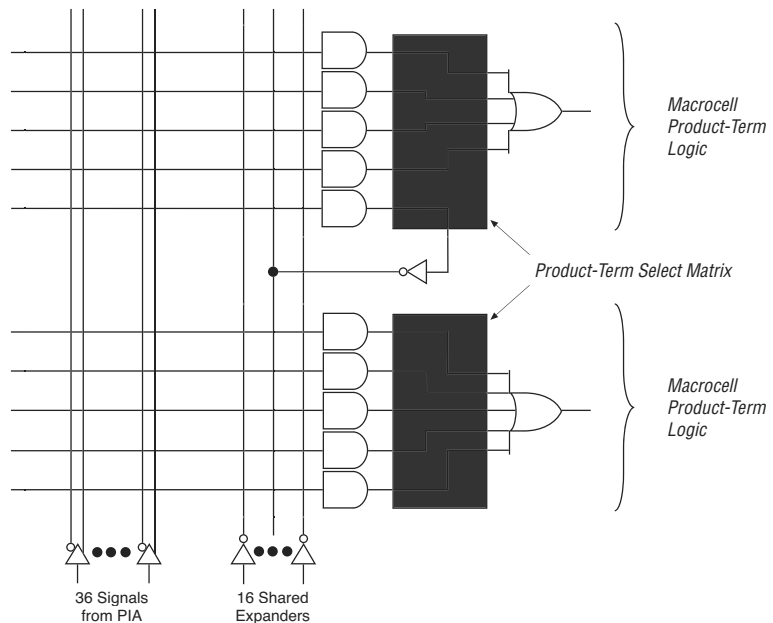
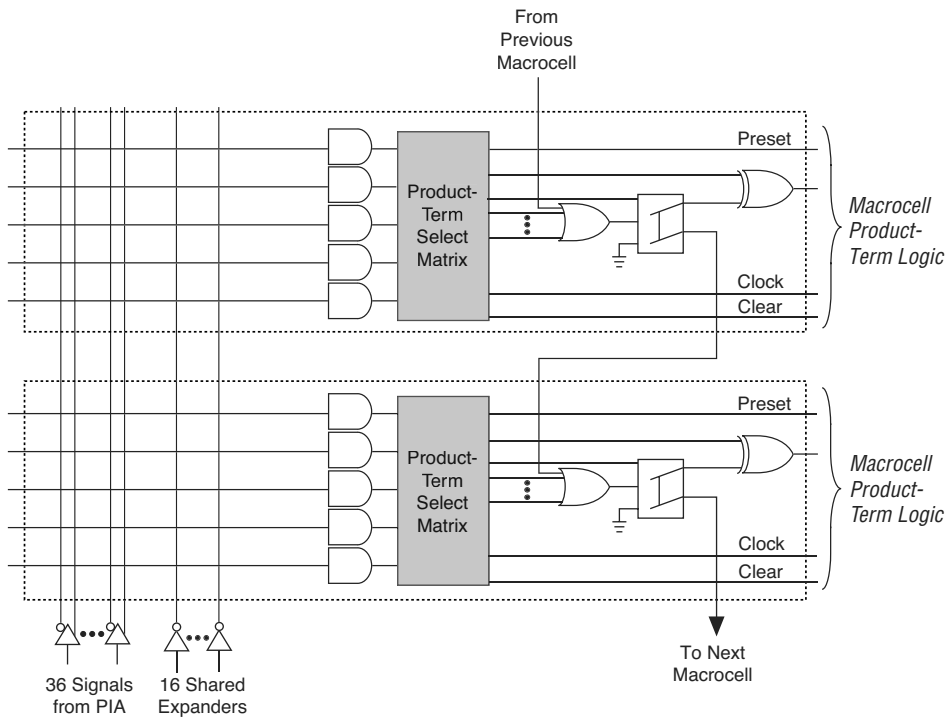


Figure 4. MAX 3000A Parallel Expanders

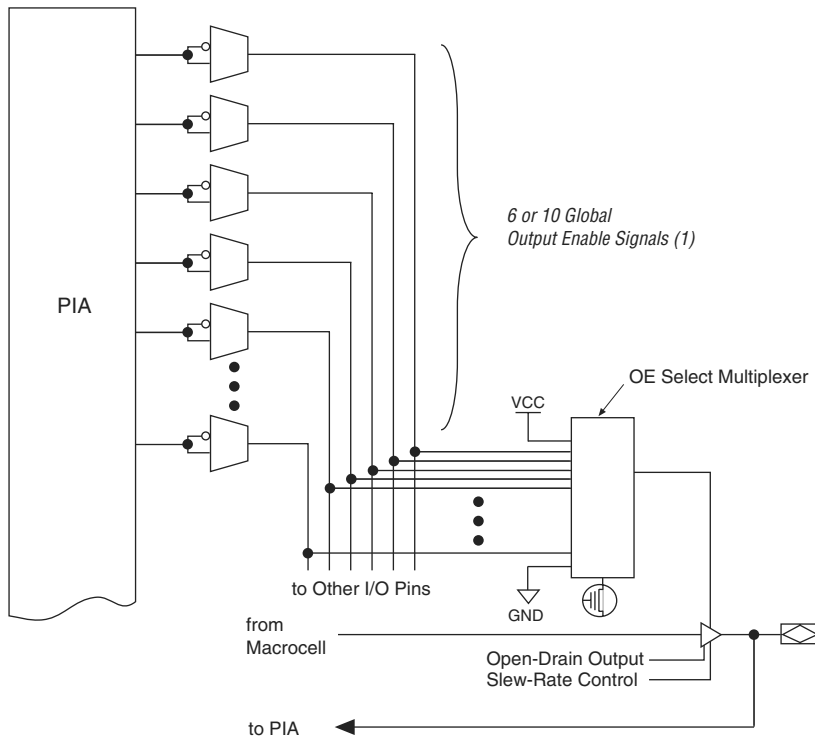
Unused product terms in a macrocell can be allocated to a neighboring macrocell.



Programmable Interconnect Array

Logic is routed between LABs on the PIA. This global bus is a programmable path that connects any signal source to any destination on the device. All MAX 3000A dedicated inputs, I/O pins, and macrocell outputs feed the PIA, which makes the signals available throughout the entire device. Only the signals required by each LAB are actually routed from the PIA into the LAB. Figure 5 shows how the PIA signals are routed into the LAB. An EEPROM cell controls one input to a two-input AND gate, which selects a PIA signal to drive into the LAB.

Figure 6. I/O Control Block of MAX 3000A Devices

**Note:**

- (1) EPM3032A, EPM3064A, EPM3128A, and EPM3256A devices have six output enables. EPM3512A devices have 10 output enables.

When the tri-state buffer control is connected to ground, the output is tri-stated (high impedance), and the I/O pin can be used as a dedicated input. When the tri-state buffer control is connected to V_{CC} , the output is enabled.

The MAX 3000A architecture provides dual I/O feedback, in which macrocell and pin feedbacks are independent. When an I/O pin is configured as an input, the associated macrocell can be used for buried logic.

By combining the pulse and shift times for each of the programming stages, the program or verify time can be derived as a function of the TCK frequency, the number of devices, and specific target device(s). Because different ISP-capable devices have a different number of EEPROM cells, both the total fixed and total variable times are unique for a single device.

Programming a Single MAX 3000A Device

The time required to program a single MAX 3000A device in-system can be calculated from the following formula:

$$t_{PROG} = t_{PPULSE} + \frac{Cycle_{PTCK}}{f_{TCK}}$$

where: t_{PROG} = Programming time
 t_{PPULSE} = Sum of the fixed times to erase, program, and verify the EEPROM cells
 $Cycle_{PTCK}$ = Number of TCK cycles to program a device
 f_{TCK} = TCK frequency

The ISP times for a stand-alone verification of a single MAX 3000A device can be calculated from the following formula:

$$t_{VER} = t_{VPULSE} + \frac{Cycle_{VTCK}}{f_{TCK}}$$

where: t_{VER} = Verify time
 t_{VPULSE} = Sum of the fixed times to verify the EEPROM cells
 $Cycle_{VTCK}$ = Number of TCK cycles to verify a device

The programming times described in Tables 4 through 6 are associated with the worst-case method using the enhanced ISP algorithm.

Table 4. MAX 3000A t_{PULSE} & $Cycle_{TCK}$ Values

Device	Programming		Stand-Alone Verification	
	t_{PULSE} (s)	$Cycle_{PTCK}$	t_{VPULSE} (s)	$Cycle_{VTCK}$
EPM3032A	2.00	55,000	0.002	18,000
EPM3064A	2.00	105,000	0.002	35,000
EPM3128A	2.00	205,000	0.002	68,000
EPM3256A	2.00	447,000	0.002	149,000
EPM3512A	2.00	890,000	0.002	297,000

Tables 5 and 6 show the in-system programming and stand alone verification times for several common test clock frequencies.

Table 5. MAX 3000A In-System Programming Times for Different Test Clock Frequencies

Device	f_{TCK}								Units
	10 MHz	5 MHz	2 MHz	1 MHz	500 kHz	200 kHz	100 kHz	50 kHz	
EPM3032A	2.01	2.01	2.03	2.06	2.11	2.28	2.55	3.10	s
EPM3064A	2.01	2.02	2.05	2.11	2.21	2.53	3.05	4.10	s
EPM3128A	2.02	2.04	2.10	2.21	2.41	3.03	4.05	6.10	s
EPM3256A	2.05	2.09	2.23	2.45	2.90	4.24	6.47	10.94	s
EPM3512A	2.09	2.18	2.45	2.89	3.78	6.45	10.90	19.80	s

Table 6. MAX 3000A Stand-Alone Verification Times for Different Test Clock Frequencies

Device	f_{TCK}								Units
	10 MHz	5 MHz	2 MHz	1 MHz	500 kHz	200 kHz	100 kHz	50 kHz	
EPM3032A	0.00	0.01	0.01	0.02	0.04	0.09	0.18	0.36	s
EPM3064A	0.01	0.01	0.02	0.04	0.07	0.18	0.35	0.70	s
EPM3128A	0.01	0.02	0.04	0.07	0.14	0.34	0.68	1.36	s
EPM3256A	0.02	0.03	0.08	0.15	0.30	0.75	1.49	2.98	s
EPM3512A	0.03	0.06	0.15	0.30	0.60	1.49	2.97	5.94	s

Figure 7 shows the timing information for the JTAG signals.

Figure 7. MAX 3000A JTAG Waveforms

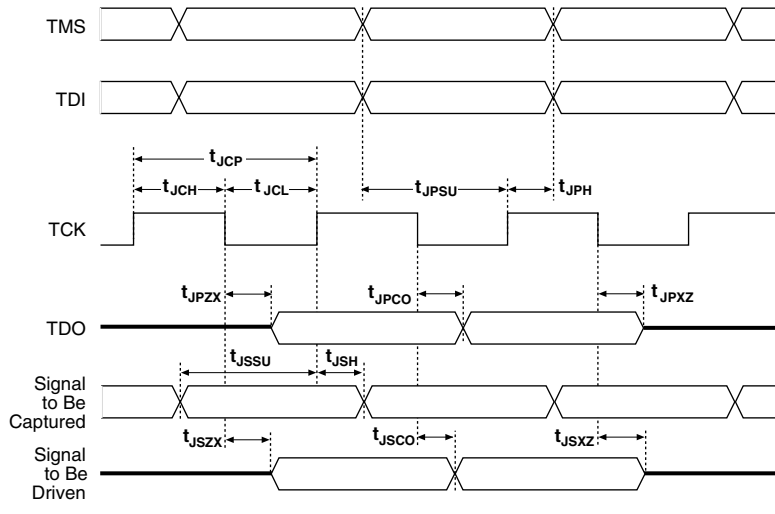


Table 10 shows the JTAG timing parameters and values for MAX 3000A devices.

Symbol	Parameter	Min	Max	Unit
t_{JCP}	TCK clock period	100		ns
t_{JCH}	TCK clock high time	50		ns
t_{JCL}	TCK clock low time	50		ns
t_{JPSU}	JTAG port setup time	20		ns
t_{JPH}	JTAG port hold time	45		ns
t_{JPCO}	JTAG port clock to output		25	ns
t_{JPZX}	JTAG port high impedance to valid output		25	ns
t_{JPXZ}	JTAG port valid output to high impedance		25	ns
t_{JSSU}	Capture register setup time	20		ns
t_{JSH}	Capture register hold time	45		ns
t_{JSCO}	Update register clock to output		25	ns
t_{JSZX}	Update register high impedance to valid output		25	ns
t_{JSXZ}	Update register valid output to high impedance		25	ns

Programmable Speed/Power Control

MAX 3000A devices offer a power-saving mode that supports low-power operation across user-defined signal paths or the entire device. This feature allows total power dissipation to be reduced by 50% or more because most logic applications require only a small fraction of all gates to operate at maximum frequency.

The designer can program each individual macrocell in a MAX 3000A device for either high-speed or low-power operation. As a result, speed-critical paths in the design can run at high speed, while the remaining paths can operate at reduced power. Macrocells that run at low power incur a nominal timing delay adder (t_{LPA}) for the t_{LAD} , t_{LAC} , t_{IC} , t_{ACL} , t_{EN} , t_{CPPW} and t_{SEXP} parameters.

Output Configuration

MAX 3000A device outputs can be programmed to meet a variety of system-level requirements.

MultiVolt I/O Interface

The MAX 3000A device architecture supports the MultiVolt I/O interface feature, which allows MAX 3000A devices to connect to systems with differing supply voltages. MAX 3000A devices in all packages can be set for 2.5-V, 3.3-V, or 5.0-V I/O pin operation. These devices have one set of V_{CC} pins for internal operation and input buffers (V_{CCINT}), and another set for I/O output drivers (V_{CCIO}).

The V_{CCIO} pins can be connected to either a 3.3-V or 2.5-V power supply, depending on the output requirements. When the V_{CCIO} pins are connected to a 2.5-V power supply, the output levels are compatible with 2.5-V systems. When the V_{CCIO} pins are connected to a 3.3-V power supply, the output high is at 3.3 V and is therefore compatible with 3.3-V or 5.0-V systems. Devices operating with V_{CCIO} levels lower than 3.0 V incur a nominally greater timing delay of t_{OD2} instead of t_{OD1} . Inputs can always be driven by 2.5-V, 3.3-V, or 5.0-V signals.

Table 11 summarizes the MAX 3000A MultiVolt I/O support.

Table 11. MAX 3000A MultiVolt I/O Support						
V_{CCIO} Voltage	Input Signal (V)			Output Signal (V)		
	2.5	3.3	5.0	2.5	3.3	5.0
2.5	✓	✓	✓	✓		
3.3	✓	✓	✓	✓	✓	✓

Note:

- (1) When V_{CCIO} is 3.3 V, a MAX 3000A device can drive a 2.5-V device that has 3.3-V tolerant inputs.

Table 19. EPM3064A Internal Timing Parameters (Part 1 of 2) *Note (1)*

Symbol	Parameter	Conditions	Speed Grade						Unit
			-4		-7		-10		
			Min	Max	Min	Max	Min	Max	
t_{IN}	Input pad and buffer delay			0.6		1.1		1.4	ns
t_{IO}	I/O input pad and buffer delay			0.6		1.1		1.4	ns
t_{SEXP}	Shared expander delay			1.8		3.0		3.9	ns
t_{PEXP}	Parallel expander delay			0.4		0.7		0.9	ns
t_{LAD}	Logic array delay			1.5		2.5		3.2	ns
t_{LAC}	Logic control array delay			0.6		1.0		1.2	ns
t_{IOE}	Internal output enable delay			0.0		0.0		0.0	ns
t_{OD1}	Output buffer and pad delay, slow slew rate = off $V_{CCIO} = 3.3\text{ V}$	$C1 = 35\text{ pF}$		0.8		1.3		1.8	ns
t_{OD2}	Output buffer and pad delay, slow slew rate = off $V_{CCIO} = 2.5\text{ V}$	$C1 = 35\text{ pF}$		1.3		1.8		2.3	ns
t_{OD3}	Output buffer and pad delay, slow slew rate = on $V_{CCIO} = 2.5\text{ V}$ or 3.3 V	$C1 = 35\text{ pF}$		5.8		6.3		6.8	ns
t_{ZX1}	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 3.3\text{ V}$	$C1 = 35\text{ pF}$		4.0		4.0		5.0	ns
t_{ZX2}	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 2.5\text{ V}$	$C1 = 35\text{ pF}$		4.5		4.5		5.5	ns
t_{ZX3}	Output buffer enable delay, slow slew rate = on $V_{CCIO} = 2.5\text{ V}$ or 3.3 V	$C1 = 35\text{ pF}$		9.0		9.0		10.0	ns
t_{XZ}	Output buffer disable delay	$C1 = 5\text{ pF}$		4.0		4.0		5.0	ns
t_{SU}	Register setup time		1.3		2.0		2.9		ns
t_H	Register hold time		0.6		1.0		1.3		ns
t_{RD}	Register delay			0.7		1.2		1.6	ns
t_{COMB}	Combinatorial delay			0.6		0.9		1.3	ns
t_{IC}	Array clock delay			1.2		1.9		2.5	ns
t_{EN}	Register enable time			0.6		1.0		1.2	ns
t_{GLOB}	Global control delay			1.0		1.5		2.2	ns
t_{PRE}	Register preset time			1.3		2.1		2.9	ns

Table 19. EPM3064A Internal Timing Parameters (Part 2 of 2) *Note (1)*

Symbol	Parameter	Conditions	Speed Grade						Unit
			−4		−7		−10		
			Min	Max	Min	Max	Min	Max	
t_{CLR}	Register clear time			1.3		2.1		2.9	ns
t_{PIA}	PIA delay	(2)		1.0		1.7		2.3	ns
t_{LPA}	Low-power adder	(5)		3.5		4.0		5.0	ns

Table 20. EPM3128A External Timing Parameters *Note (1)*

Symbol	Parameter	Conditions	Speed Grade						Unit
			–5		–7		–10		
			Min	Max	Min	Max	Min	Max	
t _{PD1}	Input to non–registered output	C1 = 35 pF (2)		5.0		7.5		10	ns
t _{PD2}	I/O input to non–registered output	C1 = 35 pF (2)		5.0		7.5		10	ns
t _{SU}	Global clock setup time	(2)	3.3		4.9		6.6		ns
t _H	Global clock hold time	(2)	0.0		0.0		0.0		ns
t _{CO1}	Global clock to output delay	C1 = 35 pF	1.0	3.4	1.0	5.0	1.0	6.6	ns
t _{CH}	Global clock high time		2.0		3.0		4.0		ns
t _{CL}	Global clock low time		2.0		3.0		4.0		ns
t _{ASU}	Array clock setup time	(2)	1.8		2.8		3.8		ns
t _{AH}	Array clock hold time	(2)	0.2		0.3		0.4		ns
t _{ACO1}	Array clock to output delay	C1 = 35 pF (2)	1.0	4.9	1.0	7.1	1.0	9.4	ns
t _{ACH}	Array clock high time		2.0		3.0		4.0		ns
t _{ACL}	Array clock low time		2.0		3.0		4.0		ns
t _{CPPW}	Minimum pulse width for clear and preset	(3)	2.0		3.0		4.0		ns
t _{CNT}	Minimum global clock period	(2)		5.2		7.7		10.2	ns
f _{CNT}	Maximum internal global clock frequency	(2), (4)	192.3		129.9		98.0		MHz
t _{ACNT}	Minimum array clock period	(2)		5.2		7.7		10.2	ns

Table 20. EPM3128A External Timing Parameters Note (1)

Symbol	Parameter	Conditions	Speed Grade						Unit
			−5		−7		−10		
			Min	Max	Min	Max	Min	Max	
f _{ACNT}	Maximum internal array clock frequency	(2), (4)	192.3		129.9		98.0		MHz

Table 21. EPM3128A Internal Timing Parameters (Part 1 of 2) Note (1)

Symbol	Parameter	Conditions	Speed Grade						Unit
			-5		-7		-10		
			Min	Max	Min	Max	Min	Max	
t_{IN}	Input pad and buffer delay			0.7		1.0		1.4	ns
t_{IO}	I/O input pad and buffer delay			0.7		1.0		1.4	ns
t_{SEXP}	Shared expander delay			2.0		2.9		3.8	ns
t_{PEXP}	Parallel expander delay			0.4		0.7		0.9	ns
t_{LAD}	Logic array delay			1.6		2.4		3.1	ns
t_{LAC}	Logic control array delay			0.7		1.0		1.3	ns
t_{IOE}	Internal output enable delay			0.0		0.0		0.0	ns
t_{OD1}	Output buffer and pad delay, slow slew rate = off $V_{CCIO} = 3.3\text{ V}$	$C1 = 35\text{ pF}$		0.8		1.2		1.6	ns
t_{OD2}	Output buffer and pad delay, slow slew rate = off $V_{CCIO} = 2.5\text{ V}$	$C1 = 35\text{ pF}$		1.3		1.7		2.1	ns
t_{OD3}	Output buffer and pad delay, slow slew rate = on $V_{CCIO} = 2.5\text{ V}$ or 3.3 V	$C1 = 35\text{ pF}$		5.8		6.2		6.6	ns
t_{ZX1}	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 3.3\text{ V}$	$C1 = 35\text{ pF}$		4.0		4.0		5.0	ns
t_{ZX2}	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 2.5\text{ V}$	$C1 = 35\text{ pF}$		4.5		4.5		5.5	ns
t_{ZX3}	Output buffer enable delay, slow slew rate = on $V_{CCIO} = 2.5\text{ V}$ or 3.3 V	$C1 = 35\text{ pF}$		9.0		9.0		10.0	ns
t_{XZ}	Output buffer disable delay	$C1 = 5\text{ pF}$		4.0		4.0		5.0	ns

Table 22. EPM3256A External Timing Parameters Note (1)

Symbol	Parameter	Conditions	Speed Grade				Unit
			−7		−10		
			Min	Max	Min	Max	
t _{CNT}	Minimum global clock period	(2)		7.9		10.5	ns
f _{CNT}	Maximum internal global clock frequency	(2), (4)	126.6		95.2		MHz
t _{ACNT}	Minimum array clock period	(2)		7.9		10.5	ns
f _{ACNT}	Maximum internal array clock frequency	(2), (4)	126.6		95.2		MHz

Table 23. EPM3256A Internal Timing Parameters (Part 1 of 2) Note (1)

Symbol	Parameter	Conditions	Speed Grade				Unit
			−7		−10		
			Min	Max	Min	Max	
t_{IN}	Input pad and buffer delay			0.9		1.2	ns
t_{IO}	I/O input pad and buffer delay			0.9		1.2	ns
t_{SEXP}	Shared expander delay			2.8		3.7	ns
t_{PEXP}	Parallel expander delay			0.5		0.6	ns
t_{LAD}	Logic array delay			2.2		2.8	ns
t_{LAC}	Logic control array delay			1.0		1.3	ns
t_{IOE}	Internal output enable delay			0.0		0.0	ns
t_{OD1}	Output buffer and pad delay, slow slew rate = off $V_{CCIO} = 3.3\text{ V}$	$C1 = 35\text{ pF}$		1.2		1.6	ns
t_{OD2}	Output buffer and pad delay, slow slew rate = off $V_{CCIO} = 2.5\text{ V}$	$C1 = 35\text{ pF}$		1.7		2.1	ns
t_{OD3}	Output buffer and pad delay, slow slew rate = on $V_{CCIO} = 2.5\text{ V}$ or 3.3 V	$C1 = 35\text{ pF}$		6.2		6.6	ns
t_{ZX1}	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 3.3\text{ V}$	$C1 = 35\text{ pF}$		4.0		5.0	ns
t_{ZX2}	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 2.5\text{ V}$	$C1 = 35\text{ pF}$		4.5		5.5	ns

Table 23. EPM3256A Internal Timing Parameters (Part 2 of 2) *Note (1)*

Symbol	Parameter	Conditions	Speed Grade				Unit
			−7		−10		
			Min	Max	Min	Max	
t_{ZX3}	Output buffer enable delay, slow slew rate = on $V_{CCIO} = 2.5\text{ V}$ or 3.3 V	$C1 = 35\text{ pF}$		9.0		10.0	ns
t_{XZ}	Output buffer disable delay	$C1 = 5\text{ pF}$		4.0		5.0	ns
t_{SU}	Register setup time		2.1		2.9		ns
t_H	Register hold time		0.9		1.2		ns
t_{RD}	Register delay			1.2		1.6	ns
t_{COMB}	Combinatorial delay			0.8		1.2	ns
t_{IC}	Array clock delay			1.6		2.1	ns
t_{EN}	Register enable time			1.0		1.3	ns
t_{GLOB}	Global control delay			1.5		2.0	ns
t_{PRE}	Register preset time			2.3		3.0	ns
t_{CLR}	Register clear time			2.3		3.0	ns
t_{PIA}	PIA delay	(2)		2.4		3.2	ns
t_{LPA}	Low-power adder	(5)		4.0		5.0	ns

Table 24. EPM3512A External Timing Parameters *Note (1)*

Symbol	Parameter	Conditions	Speed Grade				Unit
			-7		-10		
			Min	Max	Min	Max	
t _{PD1}	Input to non-registered output	C1 = 35 pF (2)		7.5		10.0	ns
t _{PD2}	I/O input to non-registered output	C1 = 35 pF (2)		7.5		10.0	ns
t _{SU}	Global clock setup time	(2)	5.6		7.6		ns
t _H	Global clock hold time	(2)	0.0		0.0		ns
t _{FSU}	Global clock setup time of fast input		3.0		3.0		ns
t _{FH}	Global clock hold time of fast input		0.0		0.0		ns
t _{CO1}	Global clock to output delay	C1 = 35 pF	1.0	4.7	1.0	6.3	ns
t _{CH}	Global clock high time		3.0		4.0		ns
t _{CL}	Global clock low time		3.0		4.0		ns
t _{ASU}	Array clock setup time	(2)	2.5		3.5		ns

Table 24. EPM3512A External Timing Parameters *Note (1)*

Symbol	Parameter	Conditions	Speed Grade				Unit
			-7		-10		
			Min	Max	Min	Max	
t _{AH}	Array clock hold time	(2)	0.2		0.3		ns
t _{ACO1}	Array clock to output delay	C1 = 35 pF (2)	1.0	7.8	1.0	10.4	ns
t _{ACH}	Array clock high time		3.0		4.0		ns
t _{ACL}	Array clock low time		3.0		4.0		ns
t _{CPPW}	Minimum pulse width for clear and preset	(3)	3.0		4.0		ns
t _{CNT}	Minimum global clock period	(2)		8.6		11.5	ns
f _{CNT}	Maximum internal global clock frequency	(2), (4)	116.3		87.0		MHz
t _{ACNT}	Minimum array clock period	(2)		8.6		11.5	ns
f _{ACNT}	Maximum internal array clock frequency	(2), (4)	116.3		87.0		MHz

Table 25. EPM3512A Internal Timing Parameters (Part 1 of 2) *Note (1)*

Symbol	Parameter	Conditions	Speed Grade				Unit
			-7		-10		
			Min	Max	Min	Max	
t_{IN}	Input pad and buffer delay			0.7		0.9	ns
t_{IO}	I/O input pad and buffer delay			0.7		0.9	ns
t_{FIN}	Fast input delay			3.1		3.6	ns
t_{SEXP}	Shared expander delay			2.7		3.5	ns
t_{PEXP}	Parallel expander delay			0.4		0.5	ns
t_{LAD}	Logic array delay			2.2		2.8	ns
t_{LAC}	Logic control array delay			1.0		1.3	ns
t_{IOE}	Internal output enable delay			0.0		0.0	ns
t_{OD1}	Output buffer and pad delay, slow slew rate = off $V_{CCIO} = 3.3\text{ V}$	C1 = 35 pF		1.0		1.5	ns
t_{OD2}	Output buffer and pad delay, slow slew rate = off $V_{CCIO} = 2.5\text{ V}$	C1 = 35 pF		1.5		2.0	ns

Power Consumption

Supply power (P) versus frequency (f_{MAX} , in MHz) for MAX 3000A devices is calculated with the following equation:

$$P = P_{\text{INT}} + P_{\text{IO}} = I_{\text{CCINT}} \times V_{\text{CC}} + P_{\text{IO}}$$

The P_{IO} value, which depends on the device output load characteristics and switching frequency, can be calculated using the guidelines given in *Application Note 74 (Evaluating Power for Altera Devices)*.

The I_{CCINT} value depends on the switching frequency and the application logic. The I_{CCINT} value is calculated with the following equation:

$$I_{\text{CCINT}} =$$

$$(A \times \text{MC}_{\text{TON}}) + [B \times (\text{MC}_{\text{DEV}} - \text{MC}_{\text{TON}})] + (C \times \text{MC}_{\text{USED}} \times f_{\text{MAX}} \times \text{tog}_{\text{LC}})$$

The parameters in the I_{CCINT} equation are:

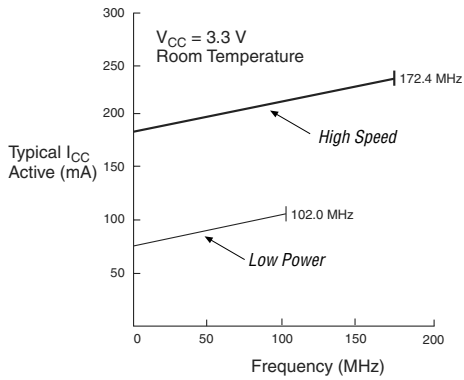
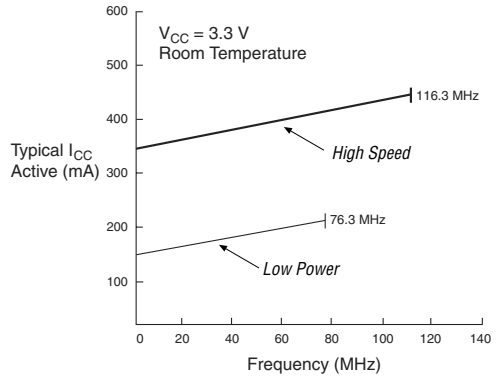
- MC_{TON} = Number of macrocells with the Turbo Bit™ option turned on, as reported in the Quartus II or MAX+PLUS II Report File (.rpt)
- MC_{DEV} = Number of macrocells in the device
- MC_{USED} = Total number of macrocells in the design, as reported in the RPT File
- f_{MAX} = Highest clock frequency to the device
- tog_{LC} = Average percentage of logic cells toggling at each clock (typically 12.5%)
- A, B, C = Constants (shown in Table 26)

Table 26. MAX 3000A I_{CC} Equation Constants

Device	A	B	C
EPM3032A	0.71	0.30	0.014
EPM3064A	0.71	0.30	0.014
EPM3128A	0.71	0.30	0.014
EPM3256A	0.71	0.30	0.014
EPM3512A	0.71	0.30	0.014

The I_{CCINT} calculation provides an I_{CC} estimate based on typical conditions using a pattern of a 16-bit, loadable, enabled, up/down counter in each LAB with no output load. Actual I_{CC} should be verified during operation because this measurement is sensitive to the actual pattern in the device and the environmental operating conditions.

Figures 12 and 13 show the typical supply current versus frequency for MAX 3000A devices.

Figure 13. I_{CC} vs. Frequency for MAX 3000A Devices**EPM3256A****EPM3512A**

Device Pin-Outs

See the Altera web site (<http://www.altera.com>) or the *Altera Digital Library* for pin-out information.

Figures 14 through 18 show the package pin-out diagrams for MAX 3000A devices.

Figure 14. 44-Pin PLCC/TQFP Package Pin-Out Diagram

Package outlines not drawn to scale.

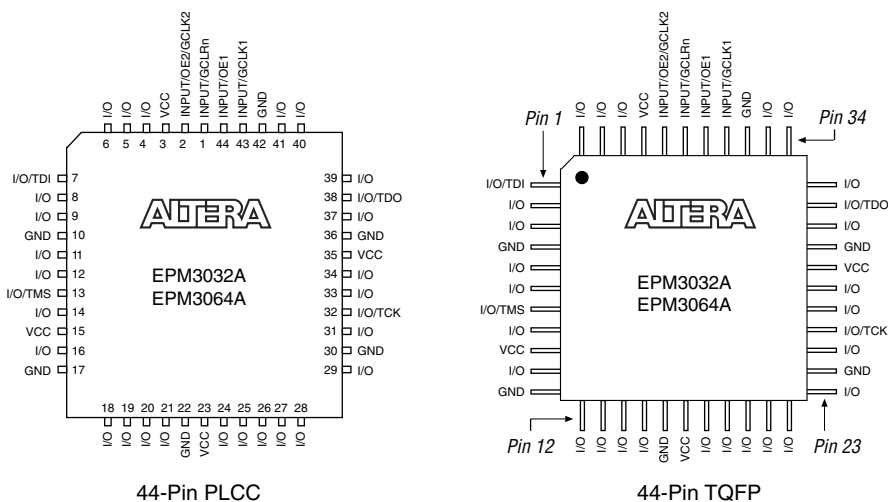


Figure 17. 208-Pin PQFP Package Pin-Out Diagram

Package outline not drawn to scale.

