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Understanding Embedded - CPLDs (Complex Programmable Logic Devices)

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

Applications of Embedded - CPLDs

Details	
Product Status	Obsolete
Programmable Type	In System Programmable
Delay Time tpd(1) Max	7.5 ns
Voltage Supply - Internal	3V ~ 3.6V
Number of Logic Elements/Blocks	8
Number of Macrocells	128
Number of Gates	2500
Number of I/O	80
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	100-TQFP
Supplier Device Package	100-TQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epm3128atc100-7n

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

...and More Features

- PCI compatible
- Bus-friendly architecture including programmable slew-rate control
- Open–drain output option
- Programmable macrocell flipflops with individual clear, preset, clock, and clock enable controls
- Programmable power–saving mode for a power reduction of over 50% in each macrocell
- Configurable expander product–term distribution, allowing up to 32 product terms per macrocell
- Programmable security bit for protection of proprietary designs
- Enhanced architectural features, including:
 - 6 or 10 pin– or logic–driven output enable signals
 - Two global clock signals with optional inversion
 - Enhanced interconnect resources for improved routability
 - Programmable output slew–rate control
- Software design support and automatic place—and—route provided by Altera's development systems for Windows—based PCs and Sun SPARCstations, and HP 9000 Series 700/800 workstations
- Additional design entry and simulation support provided by EDIF 2 0 0 and 3 0 0 netlist files, library of parameterized modules (LPM), Verilog HDL, VHDL, and other interfaces to popular EDA tools from third–party manufacturers such as Cadence, Exemplar Logic, Mentor Graphics, OrCAD, Synopsys, Synplicity, and VeriBest
- Programming support with the Altera master programming unit (MPU), MasterBlasterTM communications cable, ByteBlasterMVTM parallel port download cable, BitBlasterTM serial download cable as well as programming hardware from third–party manufacturers and any in–circuit tester that supports JamTM Standard Test and Programming Language (STAPL) Files (.jam), Jam STAPL Byte-Code Files (.jbc), or Serial Vector Format Files (.svf)

General Description

MAX 3000A devices are low–cost, high–performance devices based on the Altera MAX architecture. Fabricated with advanced CMOS technology, the EEPROM–based MAX 3000A devices operate with a 3.3-V supply voltage and provide 600 to 10,000 usable gates, ISP, pin-to-pin delays as fast as 4.5 ns, and counter speeds of up to 227.3 MHz. MAX 3000A devices in the -4, -5, -6, -7, and -10 speed grades are compatible with the timing requirements of the PCI Special Interest Group (PCI SIG) *PCI Local Bus Specification, Revision 2.2.* See Table 2.

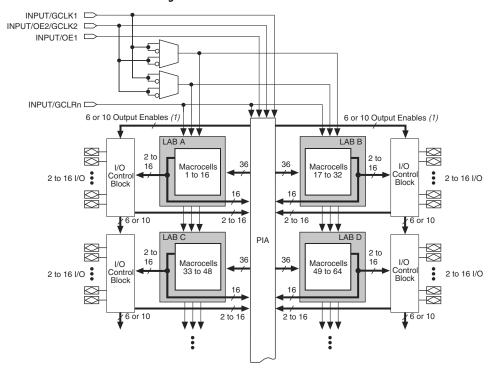


Figure 1. MAX 3000A Device Block Diagram

Note:

(1) EPM3032A, EPM3064A, EPM3128A, and EPM3256A devices have six output enables. EPM3512A devices have 10 output enables.

Logic Array Blocks

The MAX 3000A device architecture is based on the linking of high–performance LABs. LABs consist of 16–macrocell arrays, as shown in Figure 1. Multiple LABs are linked together via the PIA, a global bus that is fed by all dedicated input pins, I/O pins, and macrocells.

Each LAB is fed by the following signals:

- 36 signals from the PIA that are used for general logic inputs
- Global controls that are used for secondary register functions

Expander Product Terms

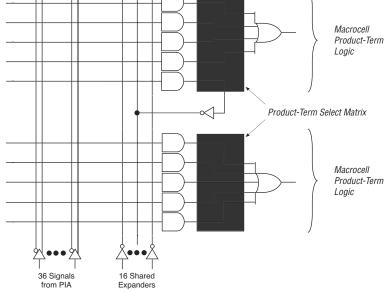
Although most logic functions can be implemented with the five product terms available in each macrocell, highly complex logic functions require additional product terms. Another macrocell can be used to supply the required logic resources. However, the MAX 3000A architecture also offers both shareable and parallel expander product terms ("expanders") that provide additional product terms directly to any macrocell in the same LAB. These expanders help ensure that logic is synthesized with the fewest possible logic resources to obtain the fastest possible speed.

Shareable Expanders

Each LAB has 16 shareable expanders that can be viewed as a pool of uncommitted single product terms (one from each macrocell) with inverted outputs that feed back into the logic array. Each shareable expander can be used and shared by any or all macrocells in the LAB to build complex logic functions. Shareable expanders incur a small delay (t_{SFXP}) . Figure 3 shows how shareable expanders can feed multiple macrocells.

Figure 3. MAX 3000A Shareable Expanders

Shareable expanders can be shared by any or all macrocells in an LAB.



Parallel Expanders

Parallel expanders are unused product terms that can be allocated to a neighboring macrocell to implement fast, complex logic functions. Parallel expanders allow up to 20 product terms to directly feed the macrocell OR logic, with five product terms provided by the macrocell and 15 parallel expanders provided by neighboring macrocells in the LAB.

The Altera development system compiler can automatically allocate up to three sets of up to five parallel expanders to the macrocells that require additional product terms. Each set of five parallel expanders incurs a small, incremental timing delay (t_{PEXP}). For example, if a macrocell requires 14 product terms, the compiler uses the five dedicated product terms within the macrocell and allocates two sets of parallel expanders; the first set includes five product terms, and the second set includes four product terms, increasing the total delay by $2 \times t_{PEXP}$.

Two groups of eight macrocells within each LAB (e.g., macrocells 1 through 8 and 9 through 16) form two chains to lend or borrow parallel expanders. A macrocell borrows parallel expanders from lower–numbered macrocells. For example, macrocell 8 can borrow parallel expanders from macrocell 7, from macrocells 7 and 6, or from macrocells 7, 6, and 5. Within each group of eight, the lowest–numbered macrocell can only lend parallel expanders and the highest–numbered macrocell can only borrow them. Figure 4 shows how parallel expanders can be borrowed from a neighboring macrocell.

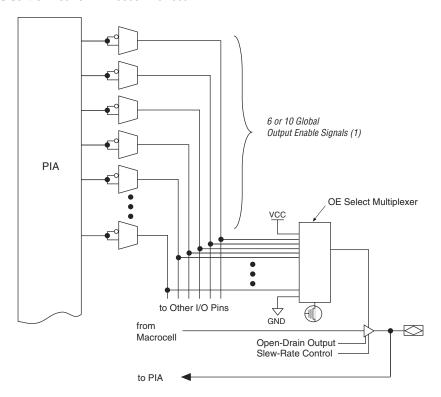


Figure 6. I/O Control Block of MAX 3000A Devices

Note:

(1) EPM3032A, EPM3064A, EPM3128A, and EPM3256A devices have six output enables. EPM3512A devices have 10 output enables.

When the tri–state buffer control is connected to ground, the output is tri-stated (high impedance), and the $\rm I/O$ pin can be used as a dedicated input. When the tri–state buffer control is connected to $\rm V_{CC}$, the output is enabled.

The MAX 3000A architecture provides dual I/O feedback, in which macrocell and pin feedbacks are independent. When an I/O pin is configured as an input, the associated macrocell can be used for buried logic.

In-System Programmability

MAX 3000A devices can be programmed in–system via an industry–standard four–pin IEEE Std. 1149.1-1990 (JTAG) interface. In-system programmability (ISP) offers quick, efficient iterations during design development and debugging cycles. The MAX 3000A architecture internally generates the high programming voltages required to program its EEPROM cells, allowing in–system programming with only a single 3.3–V power supply. During in–system programming, the I/O pins are tri–stated and weakly pulled–up to eliminate board conflicts. The pull–up value is nominally 50 k Ω

MAX 3000A devices have an enhanced ISP algorithm for faster programming. These devices also offer an ISP_Done bit that ensures safe operation when in–system programming is interrupted. This ISP_Done bit, which is the last bit programmed, prevents all I/O pins from driving until the bit is programmed.

ISP simplifies the manufacturing flow by allowing devices to be mounted on a printed circuit board (PCB) with standard pick—and—place equipment before they are programmed. MAX 3000A devices can be programmed by downloading the information via in—circuit testers, embedded processors, the MasterBlaster communications cable, the ByteBlasterMV parallel port download cable, and the BitBlaster serial download cable. Programming the devices after they are placed on the board eliminates lead damage on high—pin—count packages (e.g., QFP packages) due to device handling. MAX 3000A devices can be reprogrammed after a system has already shipped to the field. For example, product upgrades can be performed in the field via software or modem.

The Jam STAPL programming and test language can be used to program MAX 3000A devices with in–circuit testers, PCs, or embedded processors.



For more information on using the Jam STAPL programming and test language, see *Application Note 88* (Using the Jam Language for ISP & ICR via an Embedded Processor), *Application Note 122* (Using Jam STAPL for ISP & ICR via an Embedded Processor) and AN 111 (Embedded Programming Using the 8051 and Jam Byte-Code).

The ISP circuitry in MAX 3000A devices is compliant with the IEEE Std. 1532 specification. The IEEE Std. 1532 is a standard developed to allow concurrent ISP between multiple PLD vendors.

Programming Sequence

During in-system programming, instructions, addresses, and data are shifted into the MAX 3000A device through the TDI input pin. Data is shifted out through the TDO output pin and compared against the expected data.

Programming a pattern into the device requires the following six ISP stages. A stand-alone verification of a programmed pattern involves only stages 1, 2, 5, and 6.

- Enter ISP. The enter ISP stage ensures that the I/O pins transition smoothly from user mode to ISP mode. The enter ISP stage requires 1 ms.
- Check ID. Before any program or verify process, the silicon ID is checked. The time required to read this silicon ID is relatively small compared to the overall programming time.
- 3. *Bulk Erase*. Erasing the device in-system involves shifting in the instructions to erase the device and applying one erase pulse of 100 ms.
- Program. Programming the device in-system involves shifting in the address and data and then applying the programming pulse to program the EEPROM cells. This process is repeated for each EEPROM address.
- Verify. Verifying an Altera device in-system involves shifting in addresses, applying the read pulse to verify the EEPROM cells, and shifting out the data for comparison. This process is repeated for each EEPROM address.
- 6. Exit ISP. An exit ISP stage ensures that the I/O pins transition smoothly from ISP mode to user mode. The exit ISP stage requires 1 ms.

Programming Times

The time required to implement each of the six programming stages can be broken into the following two elements:

- A pulse time to erase, program, or read the EEPROM cells.
- A shifting time based on the test clock (TCK) frequency and the number of TCK cycles to shift instructions, address, and data into the device.

The programming times described in Tables 4 through 6 are associated with the worst-case method using the enhanced ISP algorithm.

Table 4. MAX 3000A t _{PU}	_{LSE} & Cycle _{TCK} Values	3					
Device	Progra	ımming	Stand-Alone Verification				
	t _{PPULSE} (s)	Cycle _{PTCK}	t _{VPULSE} (s)	Cycle _{VTCK}			
EPM3032A	2.00	55,000	0.002	18,000			
EPM3064A	2.00	105,000	0.002	35,000			
EPM3128A	2.00	205,000	0.002	68,000			
EPM3256A	2.00	447,000	0.002	149,000			
EPM3512A	2.00	890,000	0.002	297,000			

Tables 5 and 6 show the in-system programming and stand alone verification times for several common test clock frequencies.

Table 5. MAX 3000A In-System Programming Times for Different Test Clock Frequencies									
Device				1	TCK				Units
	10 MHz	5 MHz	2 MHz	1 MHz	500 kHz	200 kHz	100 kHz	50 kHz	
EPM3032A	2.01	2.01	2.03	2.06	2.11	2.28	2.55	3.10	S
EPM3064A	2.01	2.02	2.05	2.11	2.21	2.53	3.05	4.10	S
EPM3128A	2.02	2.04	2.10	2.21	2.41	3.03	4.05	6.10	S
EPM3256A	2.05	2.09	2.23	2.45	2.90	4.24	6.47	10.94	S
EPM3512A	2.09	2.18	2.45	2.89	3.78	6.45	10.90	19.80	s

Table 6. MAX 3000A Stand-Alone Verification Times for Different Test Clock Frequencies									
Device				1	TCK				Units
	10 MHz	5 MHz	2 MHz	1 MHz	500 kHz	200 kHz	100 kHz	50 kHz	
EPM3032A	0.00	0.01	0.01	0.02	0.04	0.09	0.18	0.36	S
EPM3064A	0.01	0.01	0.02	0.04	0.07	0.18	0.35	0.70	S
EPM3128A	0.01	0.02	0.04	0.07	0.14	0.34	0.68	1.36	S
EPM3256A	0.02	0.03	0.08	0.15	0.30	0.75	1.49	2.98	S
EPM3512A	0.03	0.06	0.15	0.30	0.60	1.49	2.97	5.94	S

Programming with External Hardware

MAX 3000A devices can be programmed on Windows–based PCs with an Altera Logic Programmer card, MPU, and the appropriate device adapter. The MPU performs continuity checking to ensure adequate electrical contact between the adapter and the device.



For more information, see the Altera Programming Hardware Data Sheet.

The Altera software can use text—or waveform—format test vectors created with the Altera Text Editor or Waveform Editor to test the programmed device. For added design verification, designers can perform functional testing to compare the functional device behavior with the results of simulation.

Data I/O, BP Microsystems, and other programming hardware manufacturers also provide programming support for Altera devices.



For more information, see *Programming Hardware Manufacturers*.

IEEE Std. 1149.1 (JTAG) Boundary–Scan Support

MAX 3000A devices include the JTAG BST circuitry defined by IEEE Std. 1149.1–1990. Table 7 describes the JTAG instructions supported by MAX 3000A devices. The pin-out tables found on the Altera web site (http://www.altera.com) or the *Altera Digital Library* show the location of the JTAG control pins for each device. If the JTAG interface is not required, the JTAG pins are available as user I/O pins.

Table 7. MAX 3000A	JTAG Instructions
JTAG Instruction	Description
SAMPLE/PRELOAD	Allows a snapshot of signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern output at the device pins
EXTEST	Allows the external circuitry and board–level interconnections to be tested by forcing a test pattern at the output pins and capturing test results at the input pins
BYPASS	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through a selected device to adjacent devices during normal device operation
IDCODE	Selects the IDCODE register and places it between the TDI and TDO pins, allowing the IDCODE to be serially shifted out of TDO
USERCODE	Selects the 32-bit USERCODE register and places it between the TDI and TDO pins, allowing the USERCODE value to be shifted out of TDO
ISP Instructions	These instructions are used when programming MAX 3000A devices via the JTAG ports with the MasterBlaster, ByteBlasterMV, or BitBlaster cable, or when using a Jam STAPL file, JBC file, or SVF file via an embedded processor or test equipment

Open-Drain Output Option

MAX 3000A devices provide an optional open–drain (equivalent to open-collector) output for each I/O pin. This open–drain output enables the device to provide system–level control signals (e.g., interrupt and write enable signals) that can be asserted by any of several devices. It can also provide an additional wired–OR plane.

Open-drain output pins on MAX 3000A devices (with a pull-up resistor to the 5.0-V supply) can drive 5.0-V CMOS input pins that require a high $V_{\rm IH}$. When the open-drain pin is active, it will drive low. When the pin is inactive, the resistor will pull up the trace to 5.0 V, thereby meeting CMOS requirements. The open-drain pin will only drive low or tri-state; it will never drive high. The rise time is dependent on the value of the pull-up resistor and load impedance. The $I_{\rm OL}$ current specification should be considered when selecting a pull-up resistor

Slew-Rate Control

The output buffer for each MAX 3000A I/O pin has an adjustable output slew rate that can be configured for low–noise or high–speed performance. A faster slew rate provides high–speed transitions for high-performance systems. However, these fast transitions may introduce noise transients into the system. A slow slew rate reduces system noise, but adds a nominal delay of 4 to 5 ns. When the configuration cell is turned off, the slew rate is set for low–noise performance. Each I/O pin has an individual EEPROM bit that controls the slew rate, allowing designers to specify the slew rate on a pin–by–pin basis. The slew rate control affects both the rising and falling edges of the output signal.

Design Security

All MAX 3000A devices contain a programmable security bit that controls access to the data programmed into the device. When this bit is programmed, a design implemented in the device cannot be copied or retrieved. This feature provides a high level of design security because programmed data within EEPROM cells is invisible. The security bit that controls this function, as well as all other programmed data, is reset only when the device is reprogrammed.

Generic Testing

MAX 3000A devices are fully tested. Complete testing of each programmable EEPROM bit and all internal logic elements ensures 100% programming yield. AC test measurements are taken under conditions equivalent to those shown in Figure 8. Test patterns can be used and then erased during early stages of the production flow.

Symbol	Parameter	Conditions	Min	Max	Unit
V _{CCINT}	Supply voltage for internal logic and input buffers	(10)	3.0	3.6	V
V _{CCIO}	Supply voltage for output drivers, 3.3–V operation		3.0	3.6	V
	Supply voltage for output drivers, 2.5–V operation		2.3	2.7	V
V _{CCISP}	Supply voltage during ISP		3.0	3.6	V
V _I	Input voltage	(3)	-0.5	5.75	V
V _O	Output voltage		0	V _{CCIO}	V
T _A	Ambient temperature	Commercial range	0	70	° C
		Industrial range	-40	85	° C
T _J	Junction temperature	Commercial range	0	90	° C
		Industrial range (11)	-40	105	° C
t _R	Input rise time			40	ns
t _F	Input fall time			40	ns

Table 1	4. MAX 3000A Device DC Opera	ating Conditions Note (4)			
Symbol	Parameter	Conditions	Min	Max	Unit
V _{IH}	High-level input voltage		1.7	5.75	V
V _{IL}	Low-level input voltage		-0.5	0.8	V
V _{OH}	3.3–V high–level TTL output voltage	$I_{OH} = -8 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V } (5)$	2.4		V
	3.3–V high–level CMOS output voltage	$I_{OH} = -0.1 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V } (5)$	V _{CCIO} - 0.2		V
	2.5-V high-level output voltage	$I_{OH} = -100 \mu A DC, V_{CCIO} = 2.30 V (5)$	2.1		٧
		$I_{OH} = -1 \text{ mA DC}, V_{CCIO} = 2.30 \text{ V } (5)$	2.0		V
		$I_{OH} = -2 \text{ mA DC}, V_{CCIO} = 2.30 \text{ V } (5)$	1.7		V
V_{OL}	3.3-V low-level TTL output voltage	I _{OL} = 8 mA DC, V _{CCIO} = 3.00 V <i>(6)</i>		0.4	V
	3.3–V low–level CMOS output voltage	$I_{OL} = 0.1 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V } (6)$		0.2	V
	2.5-V low-level output voltage	I _{OL} = 100 μA DC, V _{CCIO} = 2.30 V (6)		0.2	V
		I _{OL} = 1 mA DC, V _{CCIO} = 2.30 V (6)		0.4	V
		I _{OL} = 2 mA DC, V _{CCIO} = 2.30 V (6)		0.7	٧
II	Input leakage current	V _I = -0.5 to 5.5 V (7)	-10	10	μА
I _{OZ}	Tri-state output off-state current	V _I = -0.5 to 5.5 V (7)	-10	10	μА
R _{ISP}	Value of I/O pin pull–up resistor when programming in–system or during power–up	V _{CCIO} = 2.3 to 3.6 V (8)	20	74	kΩ

Tables 16 through 23 show EPM3032A, EPM3064A, EPM3128A, EPM3256A, and EPM3512A timing information.

	6. EPM3032A External 1	, 		Note (1)		•			T
Symbol	Parameter	Conditions			Speed	Grade	1		Unit
			_	4	_	-7		-10	
			Min	Max	Min	Max	Min	Max	
t _{PD1}	Input to non- registered output	C1 = 35 pF (2)		4.5		7.5		10	ns
t _{PD2}	I/O input to non– registered output	C1 = 35 pF (2)		4.5		7.5		10	ns
t _{SU}	Global clock setup time	(2)	2.9		4.7		6.3		ns
t _H	Global clock hold time	(2)	0.0		0.0		0.0		ns
t _{CO1}	Global clock to output delay	C1 = 35 pF	1.0	3.0	1.0	5.0	1.0	6.7	ns
t _{CH}	Global clock high time		2.0		3.0		4.0		ns
t _{CL}	Global clock low time		2.0		3.0		4.0		ns
t _{ASU}	Array clock setup time	(2)	1.6		2.5		3.6		ns
t _{AH}	Array clock hold time	(2)	0.3		0.5		0.5		ns
t _{ACO1}	Array clock to output delay	C1 = 35 pF (2)	1.0	4.3	1.0	7.2	1.0	9.4	ns
t _{ACH}	Array clock high time		2.0		3.0		4.0		ns
t _{ACL}	Array clock low time		2.0		3.0		4.0		ns
t _{CPPW}	Minimum pulse width for clear and preset	(3)	2.0		3.0		4.0		ns
t _{CNT}	Minimum global clock period	(2)		4.4		7.2		9.7	ns
f _{CNT}	Maximum internal global clock frequency	(2), (4)	227.3		138.9		103.1		MHz
t _{ACNT}	Minimum array clock period	(2)		4.4		7.2		9.7	ns
f _{ACNT}	Maximum internal array clock frequency	(2), (4)	227.3		138.9		103.1		MHz

Symbol	Parameter	Conditions			Speed	Grade			Unit
			_	4	_	-7	_	10	
			Min	Max	Min	Max	Min	Max	
t _{IN}	Input pad and buffer delay			0.7		1.2		1.5	ns
t _{IO}	I/O input pad and buffer delay			0.7		1.2		1.5	ns
t _{SEXP}	Shared expander delay			1.9		3.1		4.0	ns
t _{PEXP}	Parallel expander delay			0.5		0.8		1.0	ns
t_{LAD}	Logic array delay			1.5		2.5		3.3	ns
t _{LAC}	Logic control array delay			0.6		1.0		1.2	ns
t _{IOE}	Internal output enable delay			0.0		0.0		0.0	ns
t _{OD1}	Output buffer and pad delay, slow slew rate = off V _{CCIO} = 3.3 V	C1 = 35 pF		0.8		1.3		1.8	ns
t _{OD2}	Output buffer and pad delay, slow slew rate = off $V_{CCIO} = 2.5 \text{ V}$	C1 = 35 pF		1.3		1.8		2.3	ns
t _{OD3}	Output buffer and pad delay, slow slew rate = on $V_{CCIO} = 2.5 \text{ V or } 3.3 \text{ V}$	C1 = 35 pF		5.8		6.3		6.8	ns
t _{ZX1}	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 3.3 \text{ V}$	C1 = 35 pF		4.0		4.0		5.0	ns
t _{ZX2}	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 2.5 \text{ V}$	C1 = 35 pF		4.5		4.5		5.5	ns
t _{ZX3}	Output buffer enable delay, slow slew rate = on V _{CCIO} = 2.5 V or 3.3 V	C1 = 35 pF		9.0		9.0		10.0	ns
t_{XZ}	Output buffer disable delay	C1 = 5 pF		4.0		4.0		5.0	ns
t _{SU}	Register setup time		1.3		2.0		2.8		ns
t _H	Register hold time		0.6		1.0		1.3		ns
t_{RD}	Register delay			0.7		1.2		1.5	ns
t _{COMB}	Combinatorial delay			0.6		1.0		1.3	ns
t _{IC}	Array clock delay			1.2		2.0		2.5	ns
t _{EN}	Register enable time			0.6		1.0		1.2	ns
t _{GLOB}	Global control delay			0.8		1.3		1.9	ns
t _{PRE}	Register preset time			1.2		1.9		2.6	ns
t _{CLR}	Register clear time			1.2		1.9		2.6	ns

Symbol	Parameter	Conditions	Speed Grade				
			-7		-10		
			Min	Max	Min	Max	
t _{CNT}	Minimum global clock period	(2)		7.9		10.5	ns
f _{CNT}	Maximum internal global clock frequency	(2), (4)	126.6		95.2		MHz
t _{ACNT}	Minimum array clock period	(2)		7.9		10.5	ns
f _{ACNT}	Maximum internal array clock frequency	(2), (4)	126.6		95.2		MHz

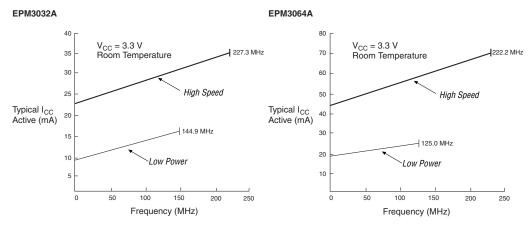
Symbol	Parameter	Conditions		Speed	Grade		Unit
			-7		-10		
			Min	Max	Min	Max	
t _{IN}	Input pad and buffer delay			0.9		1.2	ns
t _{IO}	I/O input pad and buffer delay			0.9		1.2	ns
t _{SEXP}	Shared expander delay			2.8		3.7	ns
t _{PEXP}	Parallel expander delay			0.5		0.6	ns
t_{LAD}	Logic array delay			2.2		2.8	ns
t_{LAC}	Logic control array delay			1.0		1.3	ns
t _{IOE}	Internal output enable delay			0.0		0.0	ns
t _{OD1}	Output buffer and pad delay, slow slew rate = off $V_{CCIO} = 3.3 \text{ V}$	C1 = 35 pF		1.2		1.6	ns
t _{OD2}	Output buffer and pad delay, slow slew rate = off V _{CCIO} = 2.5 V	C1 = 35 pF		1.7		2.1	ns
t _{OD3}	Output buffer and pad delay, slow slew rate = on V _{CCIO} = 2.5 V or 3.3 V	C1 = 35 pF		6.2		6.6	ns
t _{ZX1}	Output buffer enable delay, slow slew rate = off V _{CCIO} = 3.3 V	C1 = 35 pF		4.0		5.0	ns
t _{ZX2}	Output buffer enable delay, slow slew rate = off V_{CCIO} = 2.5 V	C1 = 35 pF		4.5		5.5	ns

Symbol	Parameter	Conditions		Speed	Grade		Unit
			-	·7	-1	10	
			Min	Max	Min	Max	
t _{OD3}	Output buffer and pad delay, slow slew rate = on $V_{CCIO} = 2.5 \text{ V or } 3.3 \text{ V}$	C1 = 35 pF		6.0		6.5	ns
t _{ZX1}	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 3.3 \text{ V}$	C1 = 35 pF		4.0		5.0	ns
t _{ZX2}	Output buffer enable delay, slow slew rate = off V _{CCIO} = 2.5 V	C1 = 35 pF		4.5		5.5	ns
t _{ZX3}	Output buffer enable delay, slow slew rate = on $V_{\rm CCIO} = 3.3 \ { m V}$	C1 = 35 pF		9.0		10.0	ns
t_{XZ}	Output buffer disable delay	C1 = 5 pF		4.0		5.0	ns
t _{SU}	Register setup time		2.1		3.0		ns
t _H	Register hold time		0.6		0.8		ns
t _{FSU}	Register setup time of fast input		1.6		1.6		ns
t _{FH}	Register hold time of fast input		1.4		1.4		ns
t _{RD}	Register delay			1.3		1.7	ns
t _{COMB}	Combinatorial delay			0.6		0.8	ns
t _{IC}	Array clock delay			1.8		2.3	ns
t _{EN}	Register enable time			1.0		1.3	ns
t _{GLOB}	Global control delay			1.7		2.2	ns
t _{PRE}	Register preset time			1.0		1.4	ns
t _{CLR}	Register clear time			1.0		1.4	ns
t _{PIA}	PIA delay	(2)		3.0		4.0	ns
t _{LPA}	Low-power adder	(5)		4.5		5.0	ns

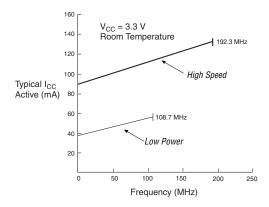
Notes to tables:

- (1) These values are specified under the recommended operating conditions, as shown in Table 13 on page 23. See Figure 11 on page 27 for more information on switching waveforms.
- (2) These values are specified for a PIA fan-out of one LAB (16 macrocells). For each additional LAB fan-out in these devices, add an additional 0.1 ns to the PIA timing value.
- (3) This minimum pulse width for preset and clear applies for both global clear and array controls. The t_{LPA} parameter must be added to this minimum width if the clear or reset signal incorporates the t_{LAD} parameter into the signal path.
- (4) These parameters are measured with a 16-bit loadable, enabled, up/down counter programmed into each LAB.
- (5) The t_{LPA} parameter must be added to the t_{LAD} , t_{LAC} , t_{IC} , t_{EN} , t_{SEXP} , $\mathbf{t_{ACL}}$, and $\mathbf{t_{CPPW}}$ parameters for macrocells running in low–power mode.

Figure 12. I_{CC} vs. Frequency for MAX 3000A Devices



EPM3128A



Device Pin-Outs

See the Altera web site (http://www.altera.com) or the *Altera Digital Library* for pin–out information.

Figures 14 through 18 show the package pin-out diagrams for MAX 3000A devices.

Figure 14. 44-Pin PLCC/TQFP Package Pin-Out Diagram

Package outlines not drawn to scale.

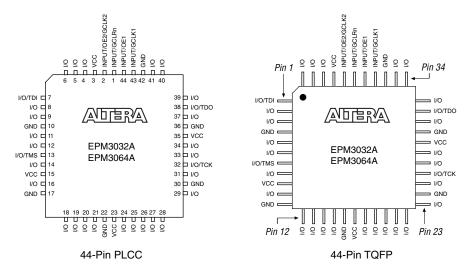


Figure 15. 100-Pin TQFP Package Pin-Out Diagram

Package outline not drawn to scale.

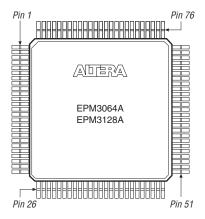


Figure 16. 144-Pin TQFP Package Pin-Out Diagram

Package outline not drawn to scale.

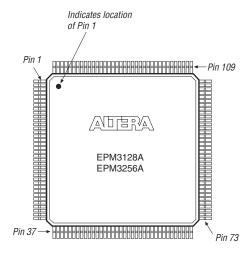
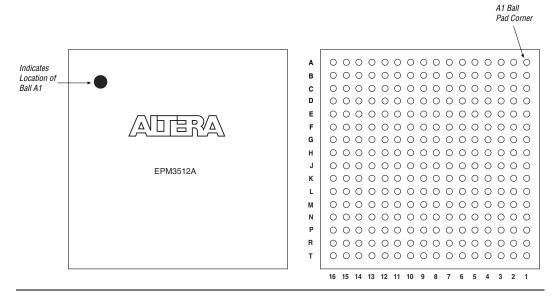


Figure 18. 256-Pin FineLine BGA Package Pin-Out Diagram

Package outline not drawn to scale.



Revision History

The information contained in the *MAX 3000A Programmable Logic Device Data Sheet* version 3.5 supersedes information published in previous versions. The following changes were made in the *MAX 3000A Programmable Logic Device Data Sheet* version 3.5:

Version 3.5

The following changes were made in the MAX 3000A Programmable Logic Device Data Sheet version 3.5:

■ New paragraph added before "Expander Product Terms".

Version 3.4

The following changes were made in the MAX 3000A Programmable Logic Device Data Sheet version 3.4:

■ Updated Table 1.

Version 3.3

The following changes were made in the MAX 3000A Programmable Logic Device Data Sheet version 3.3:

- Updated Tables 3, 13, and 26.
- Added Tables 4 through 6.
- Updated Figures 12 and 13.
- Added "Programming Sequence" on page 14 and "Programming Times" on page 14

Version 3.2

The following change were made in the MAX 3000A Programmable Logic Device Data Sheet version 3.2:

■ Updated the EPM3512 I_{CC} versus frequency graph in Figure 13.

Version 3.1

The following changes were made in the MAX 3000A Programmable Logic Device Data Sheet version 3.1:

- Updated timing information in Table 1 for the EPM3256A device.
- Updated *Note (10)* of Table 15.

Version 3.0

The following changes were made in the MAX 3000A Programmable Logic Device Data Sheet version 3.0:

- Added EPM3512A device.
- Updated Tables 2 and 3.

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