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Understanding Embedded - CPLDs (Complex Programmable Logic Devices)

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

Applications of Embedded - CPLDs

Details	
Product Status	Obsolete
Programmable Type	In System Programmable
Delay Time tpd(1) Max	10 ns
Voltage Supply - Internal	3V ~ 3.6V
Number of Logic Elements/Blocks	8
Number of Macrocells	128
Number of Gates	2500
Number of I/O	96
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epm3128ati144-10n

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

...and More Features

- PCI compatible
- Bus-friendly architecture including programmable slew-rate control
- Open–drain output option
- Programmable macrocell flipflops with individual clear, preset, clock, and clock enable controls
- Programmable power–saving mode for a power reduction of over 50% in each macrocell
- Configurable expander product–term distribution, allowing up to 32 product terms per macrocell
- Programmable security bit for protection of proprietary designs
- Enhanced architectural features, including:
 - 6 or 10 pin– or logic–driven output enable signals
 - Two global clock signals with optional inversion
 - Enhanced interconnect resources for improved routability
 - Programmable output slew–rate control
- Software design support and automatic place—and—route provided by Altera's development systems for Windows—based PCs and Sun SPARCstations, and HP 9000 Series 700/800 workstations
- Additional design entry and simulation support provided by EDIF 2 0 0 and 3 0 0 netlist files, library of parameterized modules (LPM), Verilog HDL, VHDL, and other interfaces to popular EDA tools from third–party manufacturers such as Cadence, Exemplar Logic, Mentor Graphics, OrCAD, Synopsys, Synplicity, and VeriBest
- Programming support with the Altera master programming unit (MPU), MasterBlasterTM communications cable, ByteBlasterMVTM parallel port download cable, BitBlasterTM serial download cable as well as programming hardware from third–party manufacturers and any in–circuit tester that supports JamTM Standard Test and Programming Language (STAPL) Files (.jam), Jam STAPL Byte-Code Files (.jbc), or Serial Vector Format Files (.svf)

General Description

MAX 3000A devices are low–cost, high–performance devices based on the Altera MAX architecture. Fabricated with advanced CMOS technology, the EEPROM–based MAX 3000A devices operate with a 3.3-V supply voltage and provide 600 to 10,000 usable gates, ISP, pin-to-pin delays as fast as 4.5 ns, and counter speeds of up to 227.3 MHz. MAX 3000A devices in the -4, -5, -6, -7, and -10 speed grades are compatible with the timing requirements of the PCI Special Interest Group (PCI SIG) *PCI Local Bus Specification, Revision 2.2.* See Table 2.

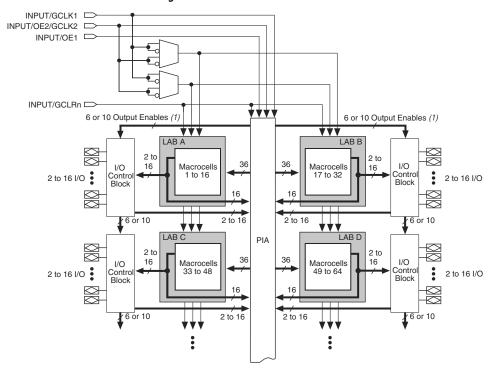


Figure 1. MAX 3000A Device Block Diagram

Note:

(1) EPM3032A, EPM3064A, EPM3128A, and EPM3256A devices have six output enables. EPM3512A devices have 10 output enables.

Logic Array Blocks

The MAX 3000A device architecture is based on the linking of high–performance LABs. LABs consist of 16–macrocell arrays, as shown in Figure 1. Multiple LABs are linked together via the PIA, a global bus that is fed by all dedicated input pins, I/O pins, and macrocells.

Each LAB is fed by the following signals:

- 36 signals from the PIA that are used for general logic inputs
- Global controls that are used for secondary register functions

Expander Product Terms

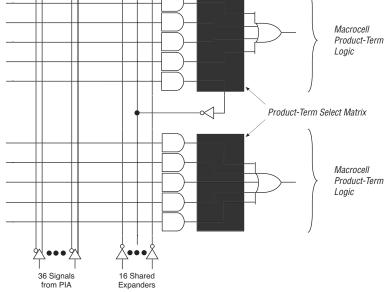
Although most logic functions can be implemented with the five product terms available in each macrocell, highly complex logic functions require additional product terms. Another macrocell can be used to supply the required logic resources. However, the MAX 3000A architecture also offers both shareable and parallel expander product terms ("expanders") that provide additional product terms directly to any macrocell in the same LAB. These expanders help ensure that logic is synthesized with the fewest possible logic resources to obtain the fastest possible speed.

Shareable Expanders

Each LAB has 16 shareable expanders that can be viewed as a pool of uncommitted single product terms (one from each macrocell) with inverted outputs that feed back into the logic array. Each shareable expander can be used and shared by any or all macrocells in the LAB to build complex logic functions. Shareable expanders incur a small delay (t_{SFXP}) . Figure 3 shows how shareable expanders can feed multiple macrocells.

Figure 3. MAX 3000A Shareable Expanders

Shareable expanders can be shared by any or all macrocells in an LAB.



Parallel Expanders

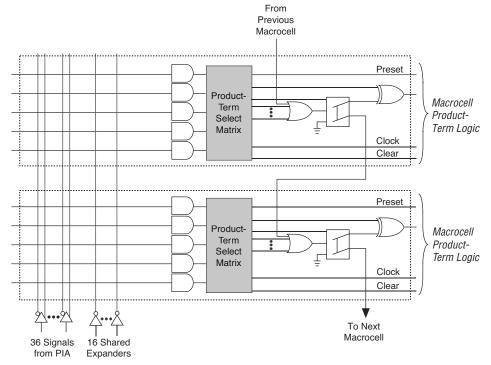
Parallel expanders are unused product terms that can be allocated to a neighboring macrocell to implement fast, complex logic functions. Parallel expanders allow up to 20 product terms to directly feed the macrocell OR logic, with five product terms provided by the macrocell and 15 parallel expanders provided by neighboring macrocells in the LAB.

The Altera development system compiler can automatically allocate up to three sets of up to five parallel expanders to the macrocells that require additional product terms. Each set of five parallel expanders incurs a small, incremental timing delay (t_{PEXP}). For example, if a macrocell requires 14 product terms, the compiler uses the five dedicated product terms within the macrocell and allocates two sets of parallel expanders; the first set includes five product terms, and the second set includes four product terms, increasing the total delay by $2 \times t_{PEXP}$.

Two groups of eight macrocells within each LAB (e.g., macrocells 1 through 8 and 9 through 16) form two chains to lend or borrow parallel expanders. A macrocell borrows parallel expanders from lower–numbered macrocells. For example, macrocell 8 can borrow parallel expanders from macrocell 7, from macrocells 7 and 6, or from macrocells 7, 6, and 5. Within each group of eight, the lowest–numbered macrocell can only lend parallel expanders and the highest–numbered macrocell can only borrow them. Figure 4 shows how parallel expanders can be borrowed from a neighboring macrocell.

Figure 4. MAX 3000A Parallel Expanders

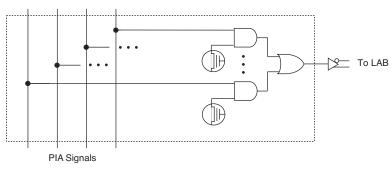
Unused product terms in a macrocell can be allocated to a neighboring macrocell.



Programmable Interconnect Array

Logic is routed between LABs on the PIA. This global bus is a programmable path that connects any signal source to any destination on the device. All MAX 3000A dedicated inputs, I/O pins, and macrocell outputs feed the PIA, which makes the signals available throughout the entire device. Only the signals required by each LAB are actually routed from the PIA into the LAB. Figure 5 shows how the PIA signals are routed into the LAB. An EEPROM cell controls one input to a two-input AND gate, which selects a PIA signal to drive into the LAB.

Figure 5. MAX 3000A PIA Routing



While the routing delays of channel-based routing schemes in masked or FPGAs are cumulative, variable, and path-dependent, the MAX 3000A PIA has a predictable delay. The PIA makes a design's timing performance easy to predict.

I/O Control Blocks

The I/O control block allows each I/O pin to be individually configured for input, output, or bidirectional operation. All I/O pins have a tri–state buffer that is individually controlled by one of the global output enable signals or directly connected to ground or $V_{CC}.$ Figure 6 shows the I/O control block for MAX 3000A devices. The I/O control block has 6 or 10 global output enable signals that are driven by the true or complement of two output enable signals, a subset of the I/O pins, or a subset of the I/O macrocells.

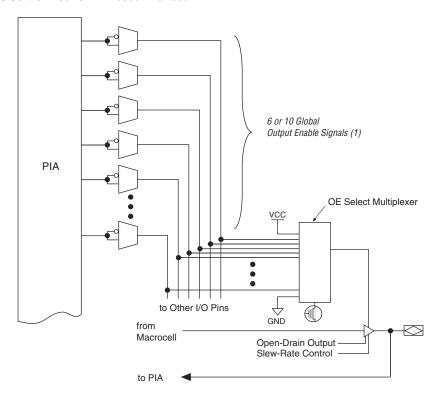


Figure 6. I/O Control Block of MAX 3000A Devices

Note:

(1) EPM3032A, EPM3064A, EPM3128A, and EPM3256A devices have six output enables. EPM3512A devices have 10 output enables.

When the tri–state buffer control is connected to ground, the output is tri-stated (high impedance), and the $\rm I/O$ pin can be used as a dedicated input. When the tri–state buffer control is connected to $\rm V_{CC}$, the output is enabled.

The MAX 3000A architecture provides dual I/O feedback, in which macrocell and pin feedbacks are independent. When an I/O pin is configured as an input, the associated macrocell can be used for buried logic.

In-System Programmability

MAX 3000A devices can be programmed in–system via an industry–standard four–pin IEEE Std. 1149.1-1990 (JTAG) interface. In-system programmability (ISP) offers quick, efficient iterations during design development and debugging cycles. The MAX 3000A architecture internally generates the high programming voltages required to program its EEPROM cells, allowing in–system programming with only a single 3.3–V power supply. During in–system programming, the I/O pins are tri–stated and weakly pulled–up to eliminate board conflicts. The pull–up value is nominally 50 k Ω

MAX 3000A devices have an enhanced ISP algorithm for faster programming. These devices also offer an ISP_Done bit that ensures safe operation when in–system programming is interrupted. This ISP_Done bit, which is the last bit programmed, prevents all I/O pins from driving until the bit is programmed.

ISP simplifies the manufacturing flow by allowing devices to be mounted on a printed circuit board (PCB) with standard pick—and—place equipment before they are programmed. MAX 3000A devices can be programmed by downloading the information via in—circuit testers, embedded processors, the MasterBlaster communications cable, the ByteBlasterMV parallel port download cable, and the BitBlaster serial download cable. Programming the devices after they are placed on the board eliminates lead damage on high—pin—count packages (e.g., QFP packages) due to device handling. MAX 3000A devices can be reprogrammed after a system has already shipped to the field. For example, product upgrades can be performed in the field via software or modem.

The Jam STAPL programming and test language can be used to program MAX 3000A devices with in–circuit testers, PCs, or embedded processors.



For more information on using the Jam STAPL programming and test language, see *Application Note 88* (Using the Jam Language for ISP & ICR via an Embedded Processor), *Application Note 122* (Using Jam STAPL for ISP & ICR via an Embedded Processor) and AN 111 (Embedded Programming Using the 8051 and Jam Byte-Code).

The ISP circuitry in MAX 3000A devices is compliant with the IEEE Std. 1532 specification. The IEEE Std. 1532 is a standard developed to allow concurrent ISP between multiple PLD vendors.

Programming Sequence

During in-system programming, instructions, addresses, and data are shifted into the MAX 3000A device through the TDI input pin. Data is shifted out through the TDO output pin and compared against the expected data.

Programming a pattern into the device requires the following six ISP stages. A stand-alone verification of a programmed pattern involves only stages 1, 2, 5, and 6.

- Enter ISP. The enter ISP stage ensures that the I/O pins transition smoothly from user mode to ISP mode. The enter ISP stage requires 1 ms.
- Check ID. Before any program or verify process, the silicon ID is checked. The time required to read this silicon ID is relatively small compared to the overall programming time.
- 3. *Bulk Erase*. Erasing the device in-system involves shifting in the instructions to erase the device and applying one erase pulse of 100 ms.
- Program. Programming the device in-system involves shifting in the address and data and then applying the programming pulse to program the EEPROM cells. This process is repeated for each EEPROM address.
- Verify. Verifying an Altera device in-system involves shifting in addresses, applying the read pulse to verify the EEPROM cells, and shifting out the data for comparison. This process is repeated for each EEPROM address.
- 6. Exit ISP. An exit ISP stage ensures that the I/O pins transition smoothly from ISP mode to user mode. The exit ISP stage requires 1 ms.

Programming Times

The time required to implement each of the six programming stages can be broken into the following two elements:

- A pulse time to erase, program, or read the EEPROM cells.
- A shifting time based on the test clock (TCK) frequency and the number of TCK cycles to shift instructions, address, and data into the device.

By combining the pulse and shift times for each of the programming stages, the program or verify time can be derived as a function of the TCK frequency, the number of devices, and specific target device(s). Because different ISP-capable devices have a different number of EEPROM cells, both the total fixed and total variable times are unique for a single device.

Programming a Single MAX 3000A Device

The time required to program a single MAX 3000A device in-system can be calculated from the following formula:

$$t_{PROG} = t_{PPULSE} + \frac{Cycle_{PTCK}}{f_{TCK}}$$

where: $t_{PROG} = Programming time$ $t_{PPULSE} = Sum of the fixed times to erase, program, and$

verify the EEPROM cells

 $Cycle_{PTCK}$ = Number of TCK cycles to program a device

= TCK frequency

The ISP times for a stand-alone verification of a single MAX 3000A device can be calculated from the following formula:

$$t_{VER} = t_{VPULSE} + \frac{Cycle_{VTCK}}{f_{TCK}}$$

where: t_{VER} = Verify time t_{VPULSE} = Sum of the fixed times to verify the EEPROM cells $Cycle_{VTCK}$ = Number of TCK cycles to verify a device

Symbol	Parameter	Conditions	Min	Max	Unit
V _{CCINT}	Supply voltage for internal logic and input buffers	(10)	3.0	3.6	V
V _{CCIO}	Supply voltage for output drivers, 3.3–V operation		3.0	3.6	V
	Supply voltage for output drivers, 2.5–V operation		2.3	2.7	V
V _{CCISP}	Supply voltage during ISP		3.0	3.6	V
V _I	Input voltage	(3)	-0.5	5.75	V
V _O	Output voltage		0	V _{CCIO}	V
T _A	Ambient temperature	Commercial range	0	70	° C
		Industrial range	-40	85	° C
T _J	Junction temperature	Commercial range	0	90	° C
		Industrial range (11)	-40	105	° C
t _R	Input rise time			40	ns
t _F	Input fall time			40	ns

Table 1	Table 14. MAX 3000A Device DC Operating Conditions Note (4)									
Symbol	Parameter	Conditions	Min	Max	Unit					
V _{IH}	High-level input voltage		1.7	5.75	V					
V _{IL}	Low-level input voltage		-0.5	0.8	V					
V _{OH}	3.3–V high–level TTL output voltage	$I_{OH} = -8 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V } (5)$	2.4		V					
	3.3–V high–level CMOS output voltage	$I_{OH} = -0.1 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V } (5)$	V _{CCIO} - 0.2		V					
	2.5-V high-level output voltage	$I_{OH} = -100 \mu A DC, V_{CCIO} = 2.30 V (5)$	2.1		٧					
		$I_{OH} = -1 \text{ mA DC}, V_{CCIO} = 2.30 \text{ V } (5)$	2.0		V					
		$I_{OH} = -2 \text{ mA DC}, V_{CCIO} = 2.30 \text{ V } (5)$	1.7		٧					
V_{OL}	3.3-V low-level TTL output voltage	I _{OL} = 8 mA DC, V _{CCIO} = 3.00 V <i>(6)</i>		0.4	V					
	3.3–V low–level CMOS output voltage	$I_{OL} = 0.1 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V } (6)$		0.2	V					
	2.5-V low-level output voltage	I _{OL} = 100 μA DC, V _{CCIO} = 2.30 V (6)		0.2	V					
		I _{OL} = 1 mA DC, V _{CCIO} = 2.30 V (6)		0.4	V					
		I _{OL} = 2 mA DC, V _{CCIO} = 2.30 V (6)		0.7	٧					
II	Input leakage current	V _I = -0.5 to 5.5 V (7)	-10	10	μА					
I _{OZ}	Tri-state output off-state current	V _I = -0.5 to 5.5 V (7)	-10	10	μА					
R _{ISP}	Value of I/O pin pull–up resistor when programming in–system or during power–up	V _{CCIO} = 2.3 to 3.6 V (8)	20	74	kΩ					

 $V_{CCINT} = 3.3 V$

V_{CCIO} = 2.5 V

Temperature = 25 °C

150 I_{OL} 100 Typical I_O $V_{CCINT} = 3.3 V$ Output $V_{CCIO} = 3.3 V$ Current (mA) Temperature = 25 °C 50 I_{OH} 2 V_O Output Voltage (V) 2.5 V 150 I_{OL}

Figure 9. Output Drive Characteristics of MAX 3000A Devices

3.3 V

Power Sequencing & Hot-Socketing

Because MAX 3000A devices can be used in a mixed–voltage environment, they have been designed specifically to tolerate any possible power–up sequence. The $\rm V_{CCIO}$ and $\rm V_{CCINT}$ power planes can be powered in any order.

V_O Output Voltage (V)

Signals can be driven into MAX 3000A devices before and during power-up without damaging the device. In addition, MAX 3000A devices do not drive out during power-up. Once operating conditions are reached, MAX 3000A devices operate as specified by the user.

Altera Corporation 25

100

50

Typical I_O

Current (mA)

Output

Symbol	Parameter	Conditions			Speed	Grade			Unit
			_	4	-7		_	10	
			Min	Max	Min	Max	Min	Max	
t _{IN}	Input pad and buffer delay			0.7		1.2		1.5	ns
t _{IO}	I/O input pad and buffer delay			0.7		1.2		1.5	ns
t _{SEXP}	Shared expander delay			1.9		3.1		4.0	ns
t _{PEXP}	Parallel expander delay			0.5		0.8		1.0	ns
t_{LAD}	Logic array delay			1.5		2.5		3.3	ns
t _{LAC}	Logic control array delay			0.6		1.0		1.2	ns
t _{IOE}	Internal output enable delay			0.0		0.0		0.0	ns
t _{OD1}	Output buffer and pad delay, slow slew rate = off V _{CCIO} = 3.3 V	C1 = 35 pF		0.8		1.3		1.8	ns
t _{OD2}	Output buffer and pad delay, slow slew rate = off $V_{CCIO} = 2.5 \text{ V}$	C1 = 35 pF		1.3		1.8		2.3	ns
t _{OD3}	Output buffer and pad delay, slow slew rate = on $V_{CCIO} = 2.5 \text{ V or } 3.3 \text{ V}$	C1 = 35 pF		5.8		6.3		6.8	ns
t _{ZX1}	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 3.3 \text{ V}$	C1 = 35 pF		4.0		4.0		5.0	ns
t _{ZX2}	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 2.5 \text{ V}$	C1 = 35 pF		4.5		4.5		5.5	ns
t _{ZX3}	Output buffer enable delay, slow slew rate = on V _{CCIO} = 2.5 V or 3.3 V	C1 = 35 pF		9.0		9.0		10.0	ns
t_{XZ}	Output buffer disable delay	C1 = 5 pF		4.0		4.0		5.0	ns
t _{SU}	Register setup time		1.3		2.0		2.8		ns
t _H	Register hold time		0.6		1.0		1.3		ns
t_{RD}	Register delay			0.7		1.2		1.5	ns
t _{COMB}	Combinatorial delay			0.6		1.0		1.3	ns
t _{IC}	Array clock delay			1.2		2.0		2.5	ns
t _{EN}	Register enable time			0.6		1.0		1.2	ns
t _{GLOB}	Global control delay			0.8		1.3		1.9	ns
t _{PRE}	Register preset time			1.2		1.9		2.6	ns
t _{CLR}	Register clear time			1.2		1.9		2.6	ns

Table 20	Table 20. EPM3128A External Timing Parameters Note (1)								
Symbol	Parameter	Conditions	Speed Grade U				Unit		
			-5 -7		' –10				
			Min	Max	Min	Max	Min	Max	
f _{ACNT}	Maximum internal array clock frequency	(2), (4)	192.3		129.9		98.0		MHz

Table 2	1. EPM3128A Internal Timing	g Parameters (I	Part 1 of	2) N	ote (1)				
Symbol	Parameter	Conditions			Speed	Grade			Unit
			_	·5	-	-7		10	
			Min	Max	Min	Max	Min	Max	
t _{IN}	Input pad and buffer delay			0.7		1.0		1.4	ns
t _{IO}	I/O input pad and buffer delay			0.7		1.0		1.4	ns
t _{SEXP}	Shared expander delay			2.0		2.9		3.8	ns
t _{PEXP}	Parallel expander delay			0.4		0.7		0.9	ns
t_{LAD}	Logic array delay			1.6		2.4		3.1	ns
t_{LAC}	Logic control array delay			0.7		1.0		1.3	ns
t _{IOE}	Internal output enable delay			0.0		0.0		0.0	ns
t _{OD1}	Output buffer and pad delay, slow slew rate = off $V_{CCIO} = 3.3 \text{ V}$	C1 = 35 pF		0.8		1.2		1.6	ns
t _{OD2}	Output buffer and pad delay, slow slew rate = off V _{CCIO} = 2.5 V	C1 = 35 pF		1.3		1.7		2.1	ns
t _{OD3}	Output buffer and pad delay, slow slew rate = on V _{CCIO} = 2.5 V or 3.3 V	C1 = 35 pF		5.8		6.2		6.6	ns
t _{ZX1}	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 3.3 \text{ V}$	C1 = 35 pF		4.0		4.0		5.0	ns
t _{ZX2}	Output buffer enable delay, slow slew rate = off V _{CCIO} = 2.5 V	C1 = 35 pF		4.5		4.5		5.5	ns
t _{ZX3}	Output buffer enable delay, slow slew rate = on V _{CCIO} = 2.5 V or 3.3 V	C1 = 35 pF		9.0		9.0		10.0	ns
t_{XZ}	Output buffer disable delay	C1 = 5 pF		4.0		4.0		5.0	ns

Table 21	1. EPM3128A Internal Tim	ning Parameters (F	Part 2 of	2) N	ote (1)				
Symbol	Parameter	Conditions	Speed Grade Ui						Unit
			_	- 5		-7		-10	
			Min	Max	Min	Max	Min	Max	
t _{SU}	Register setup time		1.4		2.1		2.9		ns
t _H	Register hold time		0.6		1.0		1.3		ns
t _{RD}	Register delay			0.8		1.2		1.6	ns
t _{COMB}	Combinatorial delay			0.5		0.9		1.3	ns
t _{IC}	Array clock delay			1.2		1.7		2.2	ns
t _{EN}	Register enable time			0.7		1.0		1.3	ns
t _{GLOB}	Global control delay			1.1		1.6		2.0	ns
t _{PRE}	Register preset time			1.4		2.0		2.7	ns
t _{CLR}	Register clear time			1.4		2.0		2.7	ns
t _{PIA}	PIA delay	(2)		1.4		2.0		2.6	ns
t_{LPA}	Low-power adder	(5)		4.0		4.0		5.0	ns

Table 22.	EPM3256A External Timing	Parameters	Note (1)				
Symbol	Parameter	Conditions		Speed	Grade		Unit
			=	-7		10	
			Min	Max	Min	Max	
t _{PD1}	Input to non–registered output	C1 = 35 pF (2)		7.5		10	ns
t _{PD2}	I/O input to non–registered output	C1 = 35 pF (2)		7.5		10	ns
t _{SU}	Global clock setup time	(2)	5.2		6.9		ns
t _H	Global clock hold time	(2)	0.0		0.0		ns
t _{CO1}	Global clock to output delay	C1 = 35 pF	1.0	4.8	1.0	6.4	ns
t _{CH}	Global clock high time		3.0		4.0		ns
t _{CL}	Global clock low time		3.0		4.0		ns
t _{ASU}	Array clock setup time	(2)	2.7		3.6		ns
t _{AH}	Array clock hold time	(2)	0.3		0.5		ns
t _{ACO1}	Array clock to output delay	C1 = 35 pF (2)	1.0	7.3	1.0	9.7	ns
t _{ACH}	Array clock high time		3.0		4.0		ns
t _{ACL}	Array clock low time		3.0		4.0		ns
t _{CPPW}	Minimum pulse width for clear and preset	(3)	3.0		4.0		ns

Symbol	Parameter	Conditions		Speed	Grade		Unit
			-7		-10		-
			Min	Max	Min	Max	
t _{CNT}	Minimum global clock period	(2)		7.9		10.5	ns
f _{CNT}	Maximum internal global clock frequency	(2), (4)	126.6		95.2		MHz
t _{ACNT}	Minimum array clock period	(2)		7.9		10.5	ns
f _{ACNT}	Maximum internal array clock frequency	(2), (4)	126.6		95.2		MHz

Symbol	Parameter	Conditions		Unit			
			-7		-10		
			Min	Max	Min	Max	
t _{IN}	Input pad and buffer delay			0.9		1.2	ns
t _{IO}	I/O input pad and buffer delay			0.9		1.2	ns
t _{SEXP}	Shared expander delay			2.8		3.7	ns
t _{PEXP}	Parallel expander delay			0.5		0.6	ns
t_{LAD}	Logic array delay			2.2		2.8	ns
t_{LAC}	Logic control array delay			1.0		1.3	ns
t _{IOE}	Internal output enable delay			0.0		0.0	ns
t _{OD1}	Output buffer and pad delay, slow slew rate = off $V_{CCIO} = 3.3 \text{ V}$	C1 = 35 pF		1.2		1.6	ns
t _{OD2}	Output buffer and pad delay, slow slew rate = off V _{CCIO} = 2.5 V	C1 = 35 pF		1.7		2.1	ns
t _{OD3}	Output buffer and pad delay, slow slew rate = on V _{CCIO} = 2.5 V or 3.3 V	C1 = 35 pF		6.2		6.6	ns
t _{ZX1}	Output buffer enable delay, slow slew rate = off V _{CCIO} = 3.3 V	C1 = 35 pF		4.0		5.0	ns
t _{ZX2}	Output buffer enable delay, slow slew rate = off V_{CCIO} = 2.5 V	C1 = 35 pF		4.5		5.5	ns

Symbol	Parameter	Conditions		Speed	Grade		Unit
			-7		-10		
			Min	Max	Min	Max	
t _{OD3}	Output buffer and pad delay, slow slew rate = on $V_{CCIO} = 2.5 \text{ V or } 3.3 \text{ V}$	C1 = 35 pF		6.0		6.5	ns
t _{ZX1}	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 3.3 \text{ V}$	C1 = 35 pF		4.0		5.0	ns
t _{ZX2}	Output buffer enable delay, slow slew rate = off V _{CCIO} = 2.5 V	C1 = 35 pF		4.5		5.5	ns
t _{ZX3}	Output buffer enable delay, slow slew rate = on $V_{\rm CCIO} = 3.3 \ { m V}$	C1 = 35 pF		9.0		10.0	ns
t_{XZ}	Output buffer disable delay	C1 = 5 pF		4.0		5.0	ns
t _{SU}	Register setup time		2.1		3.0		ns
t _H	Register hold time		0.6		0.8		ns
t _{FSU}	Register setup time of fast input		1.6		1.6		ns
t _{FH}	Register hold time of fast input		1.4		1.4		ns
t _{RD}	Register delay			1.3		1.7	ns
t _{COMB}	Combinatorial delay			0.6		0.8	ns
t _{IC}	Array clock delay			1.8		2.3	ns
t _{EN}	Register enable time			1.0		1.3	ns
t _{GLOB}	Global control delay			1.7		2.2	ns
t _{PRE}	Register preset time			1.0		1.4	ns
t _{CLR}	Register clear time			1.0		1.4	ns
t _{PIA}	PIA delay	(2)		3.0		4.0	ns
t _{LPA}	Low-power adder	(5)		4.5		5.0	ns

Notes to tables:

- (1) These values are specified under the recommended operating conditions, as shown in Table 13 on page 23. See Figure 11 on page 27 for more information on switching waveforms.
- (2) These values are specified for a PIA fan-out of one LAB (16 macrocells). For each additional LAB fan-out in these devices, add an additional 0.1 ns to the PIA timing value.
- (3) This minimum pulse width for preset and clear applies for both global clear and array controls. The t_{LPA} parameter must be added to this minimum width if the clear or reset signal incorporates the t_{LAD} parameter into the signal path.
- (4) These parameters are measured with a 16-bit loadable, enabled, up/down counter programmed into each LAB.
- (5) The t_{LPA} parameter must be added to the t_{LAD} , t_{LAC} , t_{IC} , t_{EN} , t_{SEXP} , $\mathbf{t_{ACL}}$, and $\mathbf{t_{CPPW}}$ parameters for macrocells running in low–power mode.

Power Consumption

Supply power (P) versus frequency (f_{MAX}, in MHz) for MAX 3000A devices is calculated with the following equation:

$$P = P_{INT} + P_{IO} = I_{CCINT} \times V_{CC} + P_{IO}$$

The $P_{\rm IO}$ value, which depends on the device output load characteristics and switching frequency, can be calculated using the guidelines given in *Application Note 74 (Evaluating Power for Altera Devices)*.

The I_{CCINT} value depends on the switching frequency and the application logic. The I_{CCINT} value is calculated with the following equation:

 $I_{CCINT} =$

$$(A \times MC_{TON}) + [B \times (MC_{DEV} - MC_{TON})] + (C \times MC_{USED} \times f_{MAX} \times tog_{LC})$$

The parameters in the I_{CCINT} equation are:

 MC_{TON} = Number of macrocells with the Turbo BitTM option turned

on, as reported in the Quartus II or MAX+PLUS II Report

File (.rpt)

 MC_{DEV} = Number of macrocells in the device

MC_{USED} = Total number of macrocells in the design, as reported in

the RPT File

 f_{MAX} = Highest clock frequency to the device

tog_{LC} = Average percentage of logic cells toggling at each clock

(typically 12.5%)

A, B, C = Constants (shown in Table 26)

Table 26. MAX 3000A I _{CC} Equation Constants								
Device	A	В	C					
EPM3032A	0.71	0.30	0.014					
EPM3064A	0.71	0.30	0.014					
EPM3128A	0.71	0.30	0.014					
EPM3256A	0.71	0.30	0.014					
EPM3512A	0.71	0.30	0.014					

The I_{CCINT} calculation provides an I_{CC} estimate based on typical conditions using a pattern of a 16–bit, loadable, enabled, up/down counter in each LAB with no output load. Actual I_{CC} should be verified during operation because this measurement is sensitive to the actual pattern in the device and the environmental operating conditions.

Figures 12 and 13 show the typical supply current versus frequency for MAX 3000A devices.

Figure 15. 100-Pin TQFP Package Pin-Out Diagram

Package outline not drawn to scale.

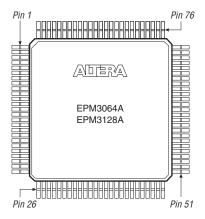
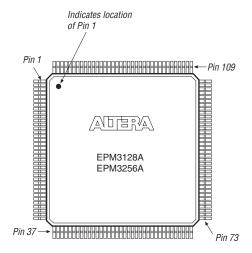


Figure 16. 144-Pin TQFP Package Pin-Out Diagram

Package outline not drawn to scale.



Version 3.3

The following changes were made in the MAX 3000A Programmable Logic Device Data Sheet version 3.3:

- Updated Tables 3, 13, and 26.
- Added Tables 4 through 6.
- Updated Figures 12 and 13.
- Added "Programming Sequence" on page 14 and "Programming Times" on page 14

Version 3.2

The following change were made in the MAX 3000A Programmable Logic Device Data Sheet version 3.2:

■ Updated the EPM3512 I_{CC} versus frequency graph in Figure 13.

Version 3.1

The following changes were made in the MAX 3000A Programmable Logic Device Data Sheet version 3.1:

- Updated timing information in Table 1 for the EPM3256A device.
- Updated *Note (10)* of Table 15.

Version 3.0

The following changes were made in the MAX 3000A Programmable Logic Device Data Sheet version 3.0:

- Added EPM3512A device.
- Updated Tables 2 and 3.

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