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Understanding Embedded - CPLDs (Complex Programmable Logic Devices)

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

Applications of Embedded - CPLDs

Details	
Product Status	Obsolete
Programmable Type	In System Programmable
Delay Time tpd(1) Max	10 ns
Voltage Supply - Internal	3V ~ 3.6V
Number of Logic Elements/Blocks	16
Number of Macrocells	256
Number of Gates	5000
Number of I/O	158
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epm3256aqi208-10n

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

...and More Features

- PCI compatible
- Bus-friendly architecture including programmable slew-rate control
- Open–drain output option
- Programmable macrocell flipflops with individual clear, preset, clock, and clock enable controls
- Programmable power–saving mode for a power reduction of over 50% in each macrocell
- Configurable expander product–term distribution, allowing up to 32 product terms per macrocell
- Programmable security bit for protection of proprietary designs
- Enhanced architectural features, including:
 - 6 or 10 pin– or logic–driven output enable signals
 - Two global clock signals with optional inversion
 - Enhanced interconnect resources for improved routability
 - Programmable output slew–rate control
- Software design support and automatic place—and—route provided by Altera's development systems for Windows—based PCs and Sun SPARCstations, and HP 9000 Series 700/800 workstations
- Additional design entry and simulation support provided by EDIF 2 0 0 and 3 0 0 netlist files, library of parameterized modules (LPM), Verilog HDL, VHDL, and other interfaces to popular EDA tools from third–party manufacturers such as Cadence, Exemplar Logic, Mentor Graphics, OrCAD, Synopsys, Synplicity, and VeriBest
- Programming support with the Altera master programming unit (MPU), MasterBlasterTM communications cable, ByteBlasterMVTM parallel port download cable, BitBlasterTM serial download cable as well as programming hardware from third–party manufacturers and any in–circuit tester that supports JamTM Standard Test and Programming Language (STAPL) Files (.jam), Jam STAPL Byte-Code Files (.jbc), or Serial Vector Format Files (.svf)

General Description

MAX 3000A devices are low–cost, high–performance devices based on the Altera MAX architecture. Fabricated with advanced CMOS technology, the EEPROM–based MAX 3000A devices operate with a 3.3-V supply voltage and provide 600 to 10,000 usable gates, ISP, pin-to-pin delays as fast as 4.5 ns, and counter speeds of up to 227.3 MHz. MAX 3000A devices in the -4, -5, -6, -7, and -10 speed grades are compatible with the timing requirements of the PCI Special Interest Group (PCI SIG) *PCI Local Bus Specification, Revision 2.2.* See Table 2.

MAX 3000A devices contain 32 to 512 macrocells, combined into groups of 16 macrocells called logic array blocks (LABs). Each macrocell has a programmable—AND/fixed—OR array and a configurable register with independently programmable clock, clock enable, clear, and preset functions. To build complex logic functions, each macrocell can be supplemented with shareable expander and high–speed parallel expander product terms to provide up to 32 product terms per macrocell.

MAX 3000A devices provide programmable speed/power optimization. Speed-critical portions of a design can run at high speed/full power, while the remaining portions run at reduced speed/low power. This speed/power optimization feature enables the designer to configure one or more macrocells to operate at 50% or lower power while adding only a nominal timing delay. MAX 3000A devices also provide an option that reduces the slew rate of the output buffers, minimizing noise transients when non-speed-critical signals are switching. The output drivers of all MAX 3000A devices can be set for 2.5 V or 3.3 V, and all input pins are 2.5–V, 3.3–V, and 5.0-V tolerant, allowing MAX 3000A devices to be used in mixed-voltage systems.

MAX 3000A devices are supported by Altera development systems, which are integrated packages that offer schematic, text—including VHDL, Verilog HDL, and the Altera Hardware Description Language (AHDL)—and waveform design entry, compilation and logic synthesis, simulation and timing analysis, and device programming. The software provides EDIF 2 0 0 and 3 0 0, LPM, VHDL, Verilog HDL, and other interfaces for additional design entry and simulation support from other industry–standard PC– and UNIX–workstation–based EDA tools. The software runs on Windows–based PCs, as well as Sun SPARCstation, and HP 9000 Series 700/800 workstations.



For more information on development tools, see the MAX+PLUS II Programmable Logic Development System & Software Data Sheet and the Quartus Programmable Logic Development System & Software Data Sheet.

Functional Description

The MAX 3000A architecture includes the following elements:

- Logic array blocks (LABs)
- Macrocells
- Expander product terms (shareable and parallel)
- Programmable interconnect array (PIA)
- I/O control blocks

The MAX 3000A architecture includes four dedicated inputs that can be used as general–purpose inputs or as high–speed, global control signals (clock, clear, and two output enable signals) for each macrocell and I/O pin. Figure 1 shows the architecture of MAX 3000A devices.

For registered functions, each macrocell flipflop can be individually programmed to implement D, T, JK, or SR operation with programmable clock control. The flipflop can be bypassed for combinatorial operation. During design entry, the designer specifies the desired flipflop type; the Altera development system software then selects the most efficient flipflop operation for each registered function to optimize resource utilization.

Each programmable register can be clocked in three different modes:

- Global clock signal mode, which achieves the fastest clock–to–output performance.
- Global clock signal enabled by an active—high clock enable. A clock enable is generated by a product term. This mode provides an enable on each flipflop while still achieving the fast clock—to—output performance of the global clock.
- Array clock implemented with a product term. In this mode, the flipflop can be clocked by signals from buried macrocells or I/O pins.

Two global clock signals are available in MAX 3000A devices. As shown in Figure 1, these global clock signals can be the true or the complement of either of the two global clock pins, GCLK1 or GCLK2.

Each register also supports asynchronous preset and clear functions. As shown in Figure 2, the product–term select matrix allocates product terms to control these operations. Although the product–term–driven preset and clear from the register are active high, active–low control can be obtained by inverting the signal within the logic array. In addition, each register clear function can be individually driven by the active–low dedicated global clear pin (GCLRn).

All registers are cleared upon power-up. By default, all registered outputs drive low when the device is powered up. You can set the registered outputs to drive high upon power-up through the Quartus[®] II software. Quartus II software uses the NOT Gate Push-Back method, which uses an additional macrocell to set the output high. To set this in the Quartus II software, go to the Assignment Editor and set the **Power-Up Level** assignment for the register to **High**.

Parallel Expanders

Parallel expanders are unused product terms that can be allocated to a neighboring macrocell to implement fast, complex logic functions. Parallel expanders allow up to 20 product terms to directly feed the macrocell OR logic, with five product terms provided by the macrocell and 15 parallel expanders provided by neighboring macrocells in the LAB.

The Altera development system compiler can automatically allocate up to three sets of up to five parallel expanders to the macrocells that require additional product terms. Each set of five parallel expanders incurs a small, incremental timing delay (t_{PEXP}). For example, if a macrocell requires 14 product terms, the compiler uses the five dedicated product terms within the macrocell and allocates two sets of parallel expanders; the first set includes five product terms, and the second set includes four product terms, increasing the total delay by $2 \times t_{PEXP}$.

Two groups of eight macrocells within each LAB (e.g., macrocells 1 through 8 and 9 through 16) form two chains to lend or borrow parallel expanders. A macrocell borrows parallel expanders from lower–numbered macrocells. For example, macrocell 8 can borrow parallel expanders from macrocell 7, from macrocells 7 and 6, or from macrocells 7, 6, and 5. Within each group of eight, the lowest–numbered macrocell can only lend parallel expanders and the highest–numbered macrocell can only borrow them. Figure 4 shows how parallel expanders can be borrowed from a neighboring macrocell.

Programming Sequence

During in-system programming, instructions, addresses, and data are shifted into the MAX 3000A device through the TDI input pin. Data is shifted out through the TDO output pin and compared against the expected data.

Programming a pattern into the device requires the following six ISP stages. A stand-alone verification of a programmed pattern involves only stages 1, 2, 5, and 6.

- Enter ISP. The enter ISP stage ensures that the I/O pins transition smoothly from user mode to ISP mode. The enter ISP stage requires 1 ms.
- Check ID. Before any program or verify process, the silicon ID is checked. The time required to read this silicon ID is relatively small compared to the overall programming time.
- 3. *Bulk Erase*. Erasing the device in-system involves shifting in the instructions to erase the device and applying one erase pulse of 100 ms.
- Program. Programming the device in-system involves shifting in the address and data and then applying the programming pulse to program the EEPROM cells. This process is repeated for each EEPROM address.
- Verify. Verifying an Altera device in-system involves shifting in addresses, applying the read pulse to verify the EEPROM cells, and shifting out the data for comparison. This process is repeated for each EEPROM address.
- 6. Exit ISP. An exit ISP stage ensures that the I/O pins transition smoothly from ISP mode to user mode. The exit ISP stage requires 1 ms.

Programming Times

The time required to implement each of the six programming stages can be broken into the following two elements:

- A pulse time to erase, program, or read the EEPROM cells.
- A shifting time based on the test clock (TCK) frequency and the number of TCK cycles to shift instructions, address, and data into the device.

Programming with External Hardware

MAX 3000A devices can be programmed on Windows–based PCs with an Altera Logic Programmer card, MPU, and the appropriate device adapter. The MPU performs continuity checking to ensure adequate electrical contact between the adapter and the device.



For more information, see the Altera Programming Hardware Data Sheet.

The Altera software can use text—or waveform—format test vectors created with the Altera Text Editor or Waveform Editor to test the programmed device. For added design verification, designers can perform functional testing to compare the functional device behavior with the results of simulation.

Data I/O, BP Microsystems, and other programming hardware manufacturers also provide programming support for Altera devices.



For more information, see *Programming Hardware Manufacturers*.

IEEE Std. 1149.1 (JTAG) Boundary–Scan Support

MAX 3000A devices include the JTAG BST circuitry defined by IEEE Std. 1149.1–1990. Table 7 describes the JTAG instructions supported by MAX 3000A devices. The pin-out tables found on the Altera web site (http://www.altera.com) or the *Altera Digital Library* show the location of the JTAG control pins for each device. If the JTAG interface is not required, the JTAG pins are available as user I/O pins.

Table 7. MAX 3000A	JTAG Instructions
JTAG Instruction	Description
SAMPLE/PRELOAD	Allows a snapshot of signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern output at the device pins
EXTEST	Allows the external circuitry and board–level interconnections to be tested by forcing a test pattern at the output pins and capturing test results at the input pins
BYPASS	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through a selected device to adjacent devices during normal device operation
IDCODE	Selects the IDCODE register and places it between the TDI and TDO pins, allowing the IDCODE to be serially shifted out of TDO
USERCODE	Selects the 32-bit USERCODE register and places it between the TDI and TDO pins, allowing the USERCODE value to be shifted out of TDO
ISP Instructions	These instructions are used when programming MAX 3000A devices via the JTAG ports with the MasterBlaster, ByteBlasterMV, or BitBlaster cable, or when using a Jam STAPL file, JBC file, or SVF file via an embedded processor or test equipment

The instruction register length of MAX 3000A devices is 10 bits. The IDCODE and USERCODE register length is 32 bits. Tables 8 and 9 show the boundary–scan register length and device IDCODE information for MAX 3000A devices.

Table 8. MAX 3000A Boundary–Scan Register Length								
Device Boundary–Scan Register Length								
EPM3032A	96							
EPM3064A	192							
EPM3128A	288							
EPM3256A	480							
EPM3512A	624							

Table 9. 32-Bit MAX 3000A Device IDCODE ValueNote (1)									
Device		IDCODE (32 bits)							
	Version (4 Bits)	Part Number (16 Bits)	Manufacturer's Identity (11 Bits)	1 (1 Bit) (2)					
EPM3032A	0001	0111 0000 0011 0010	00001101110	1					
EPM3064A	0001	0111 0000 0110 0100	00001101110	1					
EPM3128A	0001	0111 0001 0010 1000	00001101110	1					
EPM3256A	0001	0111 0010 0101 0110	00001101110	1					
EPM3512A	0001	0111 0101 0001 0010	00001101110	1					

Notes:

- (1) The most significant bit (MSB) is on the left.
- (2) The least significant bit (LSB) for all JTAG IDCODEs is 1.



See Application Note 39 (IEEE 1149.1 (JTAG) Boundary–Scan Testing in Altera Devices) for more information on JTAG BST.

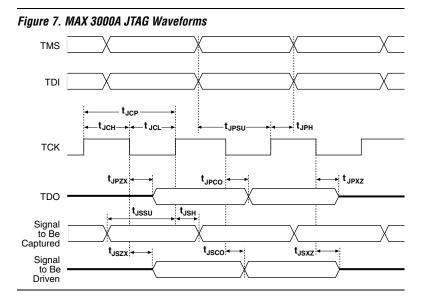


Figure 7 shows the timing information for the JTAG signals.

Table 10 shows the JTAG timing parameters and values for MAX 3000A devices.

Table 1	O. JTAG Timing Parameters & Values for MAX 30	00A De	vices	
Symbol	Parameter	Min	Max	Unit
t _{JCP}	TCK clock period	100		ns
t _{JCH}	TCK clock high time	50		ns
t _{JCL}	TCK clock low time	50		ns
t _{JPSU}	JTAG port setup time	20		ns
t _{JPH}	JTAG port hold time	45		ns
t _{JPCO}	JTAG port clock to output		25	ns
t _{JPZX}	JTAG port high impedance to valid output		25	ns
t _{JPXZ}	JTAG port valid output to high impedance		25	ns
t _{JSSU}	Capture register setup time	20		ns
t _{JSH}	Capture register hold time	45		ns
t _{JSCO}	Update register clock to output		25	ns
t _{JSZX}	Update register high impedance to valid output		25	ns
t _{JSXZ}	Update register valid output to high impedance		25	ns

Programmable Speed/Power Control

MAX 3000A devices offer a power–saving mode that supports low-power operation across user–defined signal paths or the entire device. This feature allows total power dissipation to be reduced by 50% or more because most logic applications require only a small fraction of all gates to operate at maximum frequency.

The designer can program each individual macrocell in a MAX 3000A device for either high–speed or low–power operation. As a result, speed-critical paths in the design can run at high speed, while the remaining paths can operate at reduced power. Macrocells that run at low power incur a nominal timing delay adder (t_{LPA}) for the t_{LAD} , t_{LAC} , t_{IC} , t_{ACI} , t_{EN} , t_{CPPW} and t_{SEXP} parameters.

Output Configuration

MAX 3000A device outputs can be programmed to meet a variety of system–level requirements.

MultiVolt I/O Interface

The MAX 3000A device architecture supports the MultiVolt I/O interface feature, which allows MAX 3000A devices to connect to systems with differing supply voltages. MAX 3000A devices in all packages can be set for 2.5–V, 3.3–V, or 5.0–V I/O pin operation. These devices have one set of V_{CC} pins for internal operation and input buffers (VCCINT), and another set for I/O output drivers (VCCIO).

The VCCIO pins can be connected to either a 3.3–V or 2.5–V power supply, depending on the output requirements. When the VCCIO pins are connected to a 2.5–V power supply, the output levels are compatible with 2.5–V systems. When the VCCIO pins are connected to a 3.3–V power supply, the output high is at 3.3 V and is therefore compatible with 3.3-V or 5.0–V systems. Devices operating with V_{CCIO} levels lower than 3.0 V incur a nominally greater timing delay of t_{OD2} instead of t_{OD1} . Inputs can always be driven by 2.5–V, 3.3–V, or 5.0–V signals.

1	able I	1 summarizes	the MA	X 3000A	Multi V	olt I/C) supp	ort.
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Table 11. MAX 3000A MultiVolt I/O Support								
V _{CCIO} Voltage Input Signal (V) Output Signal (V)								
	2.5	3.3	5.0	2.5	3.3	5.0		
2.5	✓	✓	✓	✓				
3.3	✓	✓	✓	✓	✓	✓		

Note:

(1) When $V_{\rm CCIO}$ is 3.3 V, a MAX 3000A device can drive a 2.5–V device that has 3.3–V tolerant inputs.

Open-Drain Output Option

MAX 3000A devices provide an optional open–drain (equivalent to open-collector) output for each I/O pin. This open–drain output enables the device to provide system–level control signals (e.g., interrupt and write enable signals) that can be asserted by any of several devices. It can also provide an additional wired–OR plane.

Open-drain output pins on MAX 3000A devices (with a pull-up resistor to the 5.0-V supply) can drive 5.0-V CMOS input pins that require a high $V_{\rm IH}$. When the open-drain pin is active, it will drive low. When the pin is inactive, the resistor will pull up the trace to 5.0 V, thereby meeting CMOS requirements. The open-drain pin will only drive low or tri-state; it will never drive high. The rise time is dependent on the value of the pull-up resistor and load impedance. The $I_{\rm OL}$ current specification should be considered when selecting a pull-up resistor

Slew-Rate Control

The output buffer for each MAX 3000A I/O pin has an adjustable output slew rate that can be configured for low–noise or high–speed performance. A faster slew rate provides high–speed transitions for high-performance systems. However, these fast transitions may introduce noise transients into the system. A slow slew rate reduces system noise, but adds a nominal delay of 4 to 5 ns. When the configuration cell is turned off, the slew rate is set for low–noise performance. Each I/O pin has an individual EEPROM bit that controls the slew rate, allowing designers to specify the slew rate on a pin–by–pin basis. The slew rate control affects both the rising and falling edges of the output signal.

Design Security

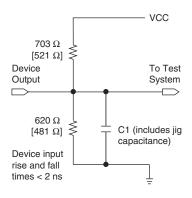
All MAX 3000A devices contain a programmable security bit that controls access to the data programmed into the device. When this bit is programmed, a design implemented in the device cannot be copied or retrieved. This feature provides a high level of design security because programmed data within EEPROM cells is invisible. The security bit that controls this function, as well as all other programmed data, is reset only when the device is reprogrammed.

Generic Testing

MAX 3000A devices are fully tested. Complete testing of each programmable EEPROM bit and all internal logic elements ensures 100% programming yield. AC test measurements are taken under conditions equivalent to those shown in Figure 8. Test patterns can be used and then erased during early stages of the production flow.

Figure 8. MAX 3000A AC Test Conditions

Power supply transients can affect AC measurements. Simultaneous transitions of multiple outputs should be avoided for accurate measurement. Threshold tests must not be performed under AC conditions. Large-amplitude, fastground-current transients normally occur as the device outputs discharge the load capacitances. When these transients flow through the parasitic inductance between the device ground pin and the test system ground, significant reductions in observable noise immunity can result. Numbers in brackets are for 2.5-V outputs. Numbers without brackets are for 3.3-V devices or outputs.



Operating Conditions

Tables 12 through 15 provide information on absolute maximum ratings, recommended operating conditions, DC operating conditions, and capacitance for MAX 3000A devices.

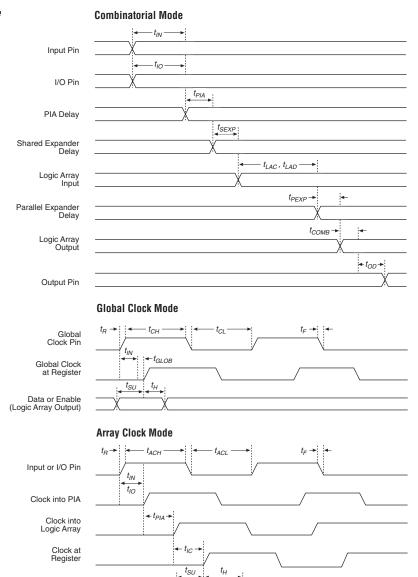
Table 1	Table 12. MAX 3000A Device Absolute Maximum Ratings Note (1)									
Symbol	Parameter	Min	Max	Unit						
V _{CC}	Supply voltage	With respect to ground (2)	-0.5	4.6	V					
VI	DC input voltage	1	-2.0	5.75	V					
I _{OUT}	DC output current, per pin		-25	25	mA					
T _{STG}	Storage temperature	No bias	-65	150	° C					
T_A	Ambient temperature	Under bias	-65	135	° C					
T_{J}	Junction temperature	PQFP and TQFP packages, under bias		135	° C					

Symbol	Parameter	Conditions	Min	Max	Unit
V _{CCINT}	Supply voltage for internal logic and input buffers	(10)	3.0	3.6	V
V _{CCIO}	Supply voltage for output drivers, 3.3–V operation		3.0	3.6	V
	Supply voltage for output drivers, 2.5–V operation		2.3	2.7	V
V _{CCISP}	Supply voltage during ISP		3.0	3.6	V
V _I	Input voltage	(3)	-0.5	5.75	V
V _O	Output voltage		0	V _{CCIO}	V
T _A	Ambient temperature	Commercial range	0	70	° C
		Industrial range	-40	85	° C
T _J	Junction temperature	Commercial range	0	90	° C
		Industrial range (11)	-40	105	° C
t _R	Input rise time			40	ns
t _F	Input fall time			40	ns

Table 1	Table 14. MAX 3000A Device DC Operating Conditions Note (4)								
Symbol	Parameter	Conditions	Min	Max	Unit				
V _{IH}	High-level input voltage		1.7	5.75	V				
V _{IL}	Low-level input voltage		-0.5	0.8	V				
V _{OH}	3.3–V high–level TTL output voltage	$I_{OH} = -8 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V } (5)$	2.4		V				
	3.3–V high–level CMOS output voltage	$I_{OH} = -0.1 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V } (5)$	V _{CCIO} - 0.2		V				
	2.5-V high-level output voltage	2.1		٧					
		$I_{OH} = -1 \text{ mA DC}, V_{CCIO} = 2.30 \text{ V } (5)$	2.0		V				
		$I_{OH} = -2 \text{ mA DC}, V_{CCIO} = 2.30 \text{ V } (5)$	1.7		V				
V_{OL}	3.3-V low-level TTL output voltage	I _{OL} = 8 mA DC, V _{CCIO} = 3.00 V <i>(6)</i>		0.4	V				
	3.3–V low–level CMOS output voltage	$I_{OL} = 0.1 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V } (6)$		0.2	V				
	2.5-V low-level output voltage	I _{OL} = 100 μA DC, V _{CCIO} = 2.30 V (6)		0.2	V				
		I _{OL} = 1 mA DC, V _{CCIO} = 2.30 V (6)		0.4	V				
		I _{OL} = 2 mA DC, V _{CCIO} = 2.30 V (6)		0.7	٧				
II	Input leakage current	V _I = -0.5 to 5.5 V (7)	-10	10	μА				
I _{OZ}	Tri-state output off-state current	V _I = -0.5 to 5.5 V (7)	-10	10	μА				
R _{ISP}	Value of I/O pin pull–up resistor when programming in–system or during power–up	V _{CCIO} = 2.3 to 3.6 V (8)	20	74	kΩ				

Figure 11. MAX 3000A Switching Waveforms

 t_R & t_F < 2 ns. Inputs are driven at 3 V for a logic high and 0 V for a logic low. All timing characteristics are measured at 1.5 V.



 $-t_{PIA}$

 $\leftarrow t_{OD} \rightarrow$

 $\leftarrow t_{CLR}, t_{PRE} \rightarrow$

Altera Corporation 27

 t_{RD}

 $\leftarrow t_{PIA} \rightarrow$

← t_{OD}

Data from Logic Array

Register to PIA to Logic Array

Register Output to Pin

Tables 16 through 23 show EPM3032A, EPM3064A, EPM3128A, EPM3256A, and EPM3512A timing information.

	6. EPM3032A External 1	, 		Note (1)		•			T
Symbol	Parameter	Conditions			Speed	Grade	1		Unit
			_	4	_	7	-1	10	
			Min	Max	Min	Max	Min	Max	
t _{PD1}	Input to non- registered output	C1 = 35 pF (2)		4.5		7.5		10	ns
t _{PD2}	I/O input to non– registered output	C1 = 35 pF (2)		4.5		7.5		10	ns
t _{SU}	Global clock setup time	(2)	2.9		4.7		6.3		ns
t _H	Global clock hold time	(2)	0.0		0.0		0.0		ns
t _{CO1}	Global clock to output delay	C1 = 35 pF	1.0	3.0	1.0	5.0	1.0	6.7	ns
t _{CH}	Global clock high time		2.0		3.0		4.0		ns
t _{CL}	Global clock low time		2.0		3.0		4.0		ns
t _{ASU}	Array clock setup time	(2)	1.6		2.5		3.6		ns
t _{AH}	Array clock hold time	(2)	0.3		0.5		0.5		ns
t _{ACO1}	Array clock to output delay	C1 = 35 pF (2)	1.0	4.3	1.0	7.2	1.0	9.4	ns
t _{ACH}	Array clock high time		2.0		3.0		4.0		ns
t _{ACL}	Array clock low time		2.0		3.0		4.0		ns
t _{CPPW}	Minimum pulse width for clear and preset	(3)	2.0		3.0		4.0		ns
t _{CNT}	Minimum global clock period	(2)		4.4		7.2		9.7	ns
f _{CNT}	Maximum internal global clock frequency	(2), (4)	227.3		138.9		103.1		MHz
t _{ACNT}	Minimum array clock period	(2)		4.4		7.2		9.7	ns
f _{ACNT}	Maximum internal array clock frequency	(2), (4)	227.3		138.9		103.1		MHz

Symbol	Parameter	Conditions	Speed Grade					Unit	
			_	4	-7		-10		
			Min	Max	Min	Max	Min	Max	
t _{IN}	Input pad and buffer delay			0.7		1.2		1.5	ns
t _{IO}	I/O input pad and buffer delay			0.7		1.2		1.5	ns
t _{SEXP}	Shared expander delay			1.9		3.1		4.0	ns
t _{PEXP}	Parallel expander delay			0.5		0.8		1.0	ns
t_{LAD}	Logic array delay			1.5		2.5		3.3	ns
t _{LAC}	Logic control array delay			0.6		1.0		1.2	ns
t _{IOE}	Internal output enable delay			0.0		0.0		0.0	ns
t _{OD1}	Output buffer and pad delay, slow slew rate = off V _{CCIO} = 3.3 V	C1 = 35 pF		0.8		1.3		1.8	ns
t _{OD2}	Output buffer and pad delay, slow slew rate = off $V_{CCIO} = 2.5 \text{ V}$	C1 = 35 pF		1.3		1.8		2.3	ns
t _{OD3}	Output buffer and pad delay, slow slew rate = on $V_{CCIO} = 2.5 \text{ V or } 3.3 \text{ V}$	C1 = 35 pF		5.8		6.3		6.8	ns
t _{ZX1}	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 3.3 \text{ V}$	C1 = 35 pF		4.0		4.0		5.0	ns
t _{ZX2}	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 2.5 \text{ V}$	C1 = 35 pF		4.5		4.5		5.5	ns
t _{ZX3}	Output buffer enable delay, slow slew rate = on V _{CCIO} = 2.5 V or 3.3 V	C1 = 35 pF		9.0		9.0		10.0	ns
t_{XZ}	Output buffer disable delay	C1 = 5 pF		4.0		4.0		5.0	ns
t _{SU}	Register setup time		1.3		2.0		2.8		ns
t _H	Register hold time		0.6		1.0		1.3		ns
t_{RD}	Register delay			0.7		1.2		1.5	ns
t _{COMB}	Combinatorial delay			0.6		1.0		1.3	ns
t _{IC}	Array clock delay			1.2		2.0		2.5	ns
t _{EN}	Register enable time			0.6		1.0		1.2	ns
t _{GLOB}	Global control delay			0.8		1.3		1.9	ns
t _{PRE}	Register preset time			1.2		1.9		2.6	ns
t _{CLR}	Register clear time			1.2		1.9		2.6	ns

Symbol	Parameter	Conditions		Unit			
			-7		-10		
			Min	Max	Min	Max	1
t _{AH}	Array clock hold time	(2)	0.2		0.3		ns
t _{ACO1}	Array clock to output delay	C1 = 35 pF (2)	1.0	7.8	1.0	10.4	ns
t _{ACH}	Array clock high time		3.0		4.0		ns
t _{ACL}	Array clock low time		3.0		4.0		ns
t _{CPPW}	Minimum pulse width for clear and preset	(3)	3.0		4.0		ns
t _{CNT}	Minimum global clock period	(2)		8.6		11.5	ns
f _{CNT}	Maximum internal global clock frequency	(2), (4)	116.3		87.0		MHz
t _{ACNT}	Minimum array clock period	(2)		8.6		11.5	ns
f _{ACNT}	Maximum internal array clock frequency	(2), (4)	116.3		87.0		MHz

Table 25. EPM3512A Internal Timing Parameters (Part 1 of 2) Note (1)							
Symbol	Parameter	Conditions	Speed Grade				Unit
			-7		-10		
			Min	Max	Min	Max	
t _{IN}	Input pad and buffer delay			0.7		0.9	ns
t _{IO}	I/O input pad and buffer delay			0.7		0.9	ns
t _{FIN}	Fast input delay			3.1		3.6	ns
t _{SEXP}	Shared expander delay			2.7		3.5	ns
t _{PEXP}	Parallel expander delay			0.4		0.5	ns
t_{LAD}	Logic array delay			2.2		2.8	ns
t _{LAC}	Logic control array delay			1.0		1.3	ns
t _{IOE}	Internal output enable delay			0.0		0.0	ns
t _{OD1}	Output buffer and pad delay, slow slew rate = off V _{CCIO} = 3.3 V	C1 = 35 pF		1.0		1.5	ns
t _{OD2}	Output buffer and pad delay, slow slew rate = off V _{CCIO} = 2.5 V	C1 = 35 pF		1.5		2.0	ns

Symbol	Parameter	Conditions	Speed Grade				Unit
			-7		-10		=
			Min	Max	Min	Max	
t _{OD3}	Output buffer and pad delay, slow slew rate = on $V_{CCIO} = 2.5 \text{ V or } 3.3 \text{ V}$	C1 = 35 pF		6.0		6.5	ns
t _{ZX1}	Output buffer enable delay, slow slew rate = off $V_{CCIO} = 3.3 \text{ V}$	C1 = 35 pF		4.0		5.0	ns
t _{ZX2}	Output buffer enable delay, slow slew rate = off V _{CCIO} = 2.5 V	C1 = 35 pF		4.5		5.5	ns
t _{ZX3}	Output buffer enable delay, slow slew rate = on $V_{\rm CCIO} = 3.3 \ { m V}$	C1 = 35 pF		9.0		10.0	ns
t_{XZ}	Output buffer disable delay	C1 = 5 pF		4.0		5.0	ns
t _{SU}	Register setup time		2.1		3.0		ns
t _H	Register hold time		0.6		0.8		ns
t _{FSU}	Register setup time of fast input		1.6		1.6		ns
t _{FH}	Register hold time of fast input		1.4		1.4		ns
t _{RD}	Register delay			1.3		1.7	ns
t _{COMB}	Combinatorial delay			0.6		0.8	ns
t _{IC}	Array clock delay			1.8		2.3	ns
t _{EN}	Register enable time			1.0		1.3	ns
t _{GLOB}	Global control delay			1.7		2.2	ns
t _{PRE}	Register preset time			1.0		1.4	ns
t _{CLR}	Register clear time			1.0		1.4	ns
t _{PIA}	PIA delay	(2)		3.0		4.0	ns
t _{LPA}	Low-power adder	(5)		4.5		5.0	ns

Notes to tables:

- (1) These values are specified under the recommended operating conditions, as shown in Table 13 on page 23. See Figure 11 on page 27 for more information on switching waveforms.
- (2) These values are specified for a PIA fan-out of one LAB (16 macrocells). For each additional LAB fan-out in these devices, add an additional 0.1 ns to the PIA timing value.
- (3) This minimum pulse width for preset and clear applies for both global clear and array controls. The t_{LPA} parameter must be added to this minimum width if the clear or reset signal incorporates the t_{LAD} parameter into the signal path.
- (4) These parameters are measured with a 16-bit loadable, enabled, up/down counter programmed into each LAB.
- (5) The t_{LPA} parameter must be added to the t_{LAD} , t_{LAC} , t_{IC} , t_{EN} , t_{SEXP} , $\mathbf{t_{ACL}}$, and $\mathbf{t_{CPPW}}$ parameters for macrocells running in low–power mode.

Power Consumption

Supply power (P) versus frequency (f_{MAX}, in MHz) for MAX 3000A devices is calculated with the following equation:

$$P = P_{INT} + P_{IO} = I_{CCINT} \times V_{CC} + P_{IO}$$

The $P_{\rm IO}$ value, which depends on the device output load characteristics and switching frequency, can be calculated using the guidelines given in *Application Note 74 (Evaluating Power for Altera Devices)*.

The I_{CCINT} value depends on the switching frequency and the application logic. The I_{CCINT} value is calculated with the following equation:

 $I_{CCINT} =$

$$(A \times MC_{TON}) + [B \times (MC_{DEV} - MC_{TON})] + (C \times MC_{USED} \times f_{MAX} \times tog_{LC})$$

The parameters in the I_{CCINT} equation are:

 MC_{TON} = Number of macrocells with the Turbo BitTM option turned

on, as reported in the Quartus II or MAX+PLUS II Report

File (.rpt)

 MC_{DEV} = Number of macrocells in the device

MC_{USED} = Total number of macrocells in the design, as reported in

the RPT File

 f_{MAX} = Highest clock frequency to the device

tog_{LC} = Average percentage of logic cells toggling at each clock

(typically 12.5%)

A, B, C = Constants (shown in Table 26)

Table 26. MAX 3000A I _{CC} Equation Constants						
Device	A	В	C			
EPM3032A	0.71	0.30	0.014			
EPM3064A	0.71	0.30	0.014			
EPM3128A	0.71	0.30	0.014			
EPM3256A	0.71	0.30	0.014			
EPM3512A	0.71	0.30	0.014			

The I_{CCINT} calculation provides an I_{CC} estimate based on typical conditions using a pattern of a 16–bit, loadable, enabled, up/down counter in each LAB with no output load. Actual I_{CC} should be verified during operation because this measurement is sensitive to the actual pattern in the device and the environmental operating conditions.

Figures 12 and 13 show the typical supply current versus frequency for MAX 3000A devices.

Figure 17. 208-Pin PQFP Package Pin-Out Diagram

Package outline not drawn to scale.

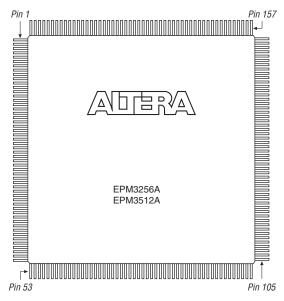
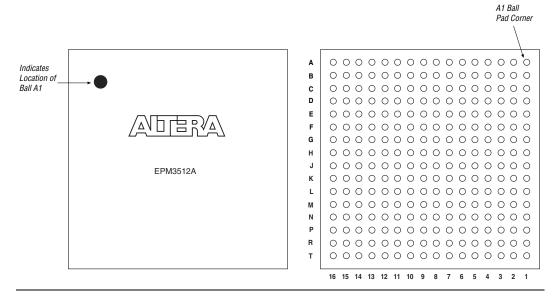


Figure 18. 256-Pin FineLine BGA Package Pin-Out Diagram

Package outline not drawn to scale.



Revision History

The information contained in the *MAX 3000A Programmable Logic Device Data Sheet* version 3.5 supersedes information published in previous versions. The following changes were made in the *MAX 3000A Programmable Logic Device Data Sheet* version 3.5:

Version 3.5

The following changes were made in the MAX 3000A Programmable Logic Device Data Sheet version 3.5:

■ New paragraph added before "Expander Product Terms".

Version 3.4

The following changes were made in the MAX 3000A Programmable Logic Device Data Sheet version 3.4:

■ Updated Table 1.