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Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	64MHz
Connectivity	I ² C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, LCD, POR, PWM, WDT
Number of I/O	69
Program Memory Size	32KB (16K x 16)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	2K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 5.5V
Data Converters	A/D 24x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	80-TQFP
Supplier Device Package	80-TQFP (12x12)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18f85k90-i-ptsl

PIC18F87K90 FAMILY

NOTES:

PIC18F87K90 FAMILY

FIGURE 5-4: TIME-OUT SEQUENCE ON POWER-UP ($\overline{\text{MCLR}}$ NOT TIED TO V_{DD}): CASE 1

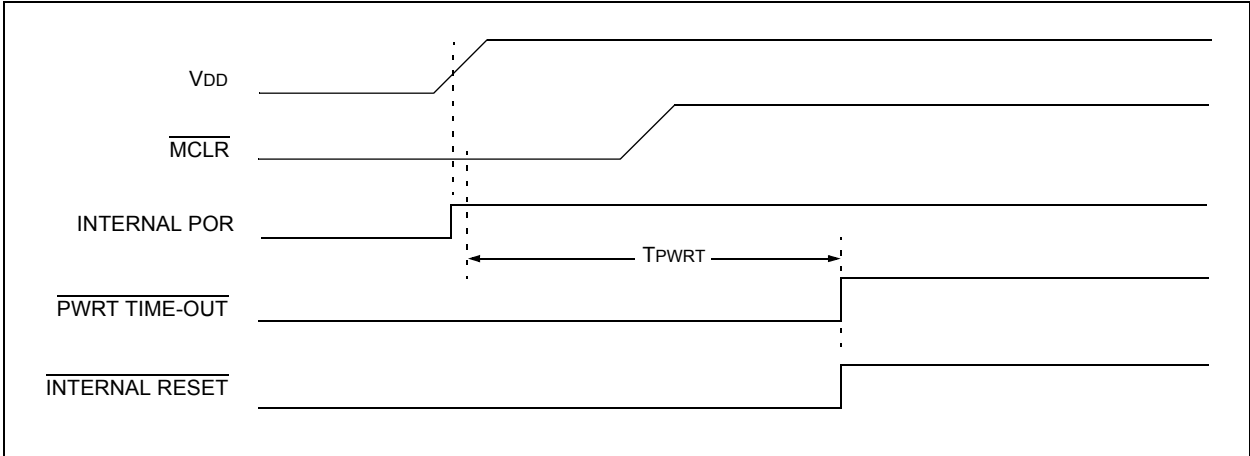


FIGURE 5-5: TIME-OUT SEQUENCE ON POWER-UP ($\overline{\text{MCLR}}$ NOT TIED TO V_{DD}): CASE 2

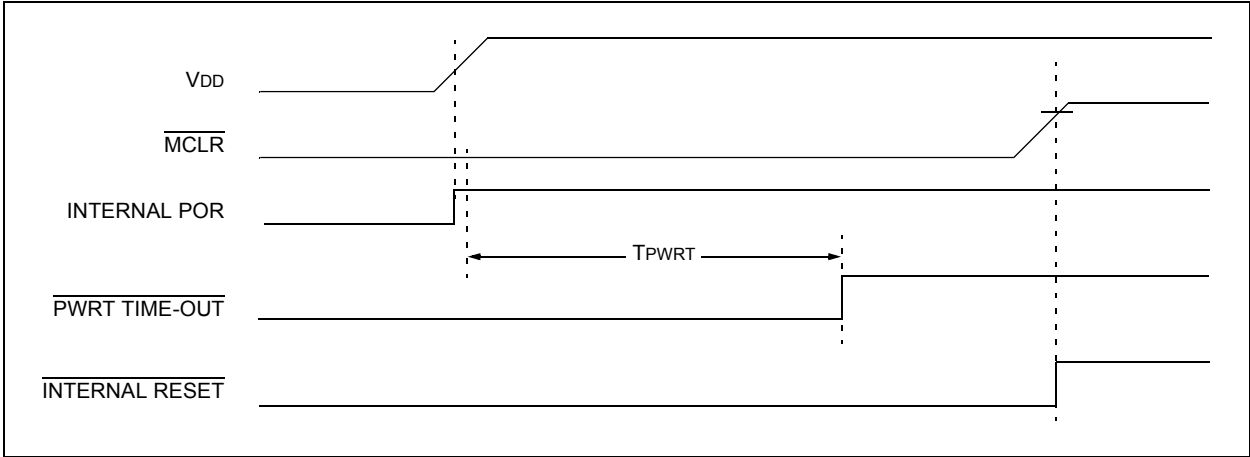
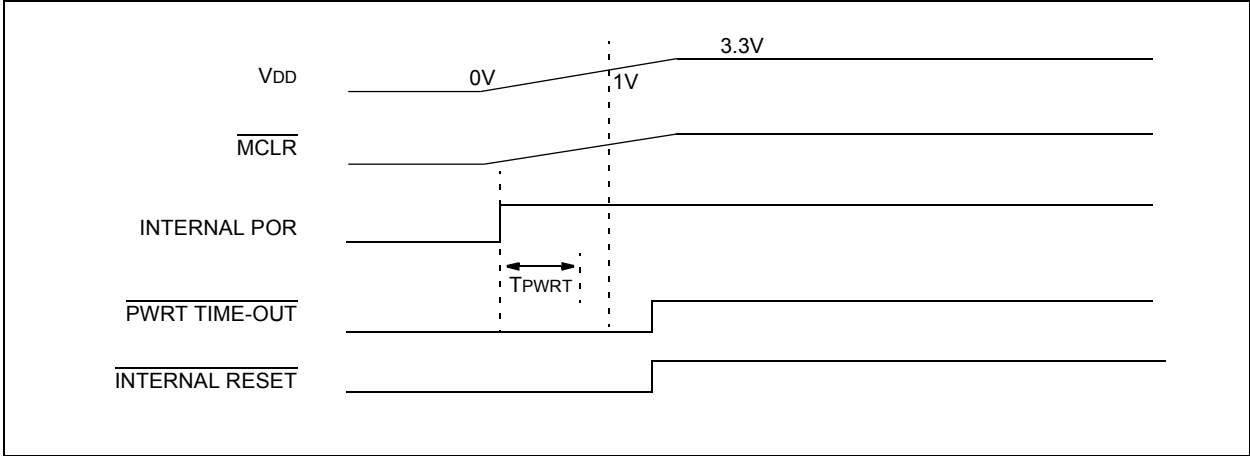


FIGURE 5-6: SLOW RISE TIME ($\overline{\text{MCLR}}$ TIED TO V_{DD} , V_{DD} RISE $> T_{PWRT}$)



11.6 PORTE, TRISE and LATE Registers

PORTE is an 8-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISE and LATE.

All pins on PORTE are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output. The RE7 pin is also configurable for open-drain output when ECCP2 is active on this pin. Open-drain configuration is selected by setting the CCP2OD control bit (ODCON1<6>)

Note: These pins are configured as digital inputs on any device Reset.

Each of the PORTE pins has a weak internal pull-up. A single control bit can turn off all the pull-ups. This is performed by clearing bit, REPU (PADCFG1<6>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on any device Reset.

Pins, RE<6:3>, are multiplexed with the LCD common drives. I/O port functions are available only on those PORTE pins according to which commons are active. The configuration is determined by the LMUX<1:0> control bits (LCDCON<1:0>). The availability is summarized in Table 11-9.

TABLE 11-9: PORTE PINS AVAILABLE IN DIFFERENT LCD DRIVE CONFIGURATIONS⁽¹⁾

LCDCON <1:0>	Active LCD Commons	PORTE Pins Available for I/O
00	COM0	RE6, RE5, RE4
01	COM0, COM1	RE6, RE5
10	COM0, COM1 and COM2	RE6
11	All (COM0 through COM3)	None

Note 1: If the LCD bias voltages are generated using the internal resistor ladder, the LCDBIASx pins are also available as I/O ports (RE0, RE1 and RE2).

Pins, RE2, RE1 and RE0, are multiplexed with the functions of LCDBIAS3, LCDBIAS2 and LCDBIAS1. When LCD bias generation is required (in any application where the device is connected to an external LCD), these pins cannot be used as digital I/O. These pins can be used as digital I/O, however, when the internal resistor ladder is used for bias generation.

PORTE is also multiplexed with the Enhanced PWM Outputs B and C for ECCP1 and ECCP3, and Outputs B, C and D for ECCP2. For all devices, their default assignments are on PORTE<6:0>. On 80-pin devices, the multiplexing for the outputs of ECCP1 and ECCP3 is controlled by the ECCPMX Configuration bit. Clearing this bit reassigns the P1B/P1C and P3B/P3C outputs to PORTH.

RE7 is multiplexed with the LCD segment drive (SEG31) that is controlled by the LCDSE3<7> bit. I/O port function is only available when the segment is disabled. RE7 can also be configured as the alternate peripheral pin for the ECCP2 module. This is done by clearing the CCP2MX Configuration bit.

RE3 can also be configured as the Reference Clock Output (REFO) from the system clock. For further details, refer to **Section 3.7 “Reference Clock Output”**.

EXAMPLE 11-5: INITIALIZING PORTE

```
CLRF    PORTE    ; Initialize PORTE by
                ; clearing output
                ; data latches
CLRF    LATE     ; Alternate method
                ; to clear output
                ; data latches
MOVLW   03h      ; Value used to
                ; initialize data
                ; direction
MOVWF   TRISE    ; Set RE<1:0> as inputs
                ; RE<7:2> as outputs
```

15.0 TIMER3/5/7 MODULES

The Timer3/5/7 timer/counter modules incorporate these features:

- Software-selectable operation as a 16-bit timer or counter
- Readable and writable 8-bit registers (TMRxH and TMRxL)
- Selectable clock source (internal or external) with device clock or SOSC oscillator internal options
- Interrupt-on-overflow
- Module Reset on ECCP Special Event Trigger

Timer7 is unimplemented for devices with a program memory of 32 Kbytes (PIC18FX5K90).

Note: Throughout this section, generic references are used for register and bit names that are the same, except for an 'x' variable that indicates the item's association with the Timer3, Timer5 or Timer7 module. For example, the control register is named TxCON and refers to T3CON, T5CON and T7CON.

A simplified block diagram of the Timer3/5/7 module is shown in Figure 15-1.

The Timer3/5/7 module is controlled through the TxCON register (Register 15-1). It also selects the clock source options for the ECCP modules. (For more information, see **Section 19.1.1 “ECCP Module and Timer Resources”**.)

The FOSC clock source should not be used with the ECCP capture/compare features. If the timer will be used with the capture or compare features, always select one of the other timer clocking options.

PIC18F87K90 FAMILY

17.1.3 ALRMVALH AND ALRMVALL REGISTER MAPPINGS

REGISTER 17-14: ALRMMNTH: ALARM MONTH VALUE REGISTER⁽¹⁾

U-0		U-0		U-0		R/W-x		R/W-x		R/W-x		R/W-x		R/W-x	
—		—		—		MHTTEN0		MTHONE3		MTHONE2		MTHONE1		MTHONE0	
bit 7														bit 0	

Legend:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'
 -n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

- bit 7-5 **Unimplemented:** Read as '0'
 bit 4 **MHTTEN0:** Binary Coded Decimal Value of Month's Tens Digit bits
 Contains a value of '0' or '1'.
 bit 3-0 **MTHONE<3:0>:** Binary Coded Decimal Value of Month's Ones Digit bits
 Contains a value from 0 to 9.

Note 1: A write to this register is only allowed when RTCWREN = 1.

REGISTER 17-15: ALRMDAY: ALARM DAY VALUE REGISTER⁽¹⁾

U-0		U-0		R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	DAYTEN1	DAYTEN0	DAYONE3	DAYONE2	DAYONE1	DAYONE0		
bit 7								bit 0	

Legend:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'
 -n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

- bit 7-6 **Unimplemented:** Read as '0'
 bit 5-4 **DAYTEN<1:0>:** Binary Coded Decimal Value of Day's Tens Digit bits
 Contains a value from 0 to 3.
 bit 3-0 **DAYONE<3:0>:** Binary Coded Decimal Value of Day's Ones Digit bits
 Contains a value from 0 to 9.

Note 1: A write to this register is only allowed when RTCWREN = 1.

17.2.4 LEAP YEAR

Since the year range on the RTCC module is 2000 to 2099, the leap year calculation is determined by any year divisible by four in the above range. Only February is affected in a leap year.

February will have 29 days in a leap year and 28 days in any other year.

17.2.5 GENERAL FUNCTIONALITY

All Timer registers containing a time value of seconds or greater are writable. The user configures the time by writing the required year, month, day, hour, minutes and seconds to the Timer registers, via Register Pointers. (See **Section 17.2.8 “Register Mapping”**.)

The timer uses the newly written values and proceeds with the count from the required starting point.

The RTCC is enabled by setting the RTCEN bit (RTCCFG<7>). If enabled, while adjusting these registers, the timer still continues to increment. However, any time the MINSEC register is written to, both of the timer prescalers are reset to '0'. This allows fraction of a second synchronization.

The Timer registers are updated in the same cycle as the write instruction's execution by the CPU. The user must ensure that when RTCEN = 1, the updated registers will not be incremented at the same time. This can be accomplished in several ways:

- By checking the RTCSYNC bit (RTCCFG<4>)
- By checking the preceding digits from which a carry can occur
- By updating the registers immediately following the seconds pulse (or an alarm interrupt)

The user has visibility to the half-second field of the counter. This value is read-only and can be reset only by writing to the lower half of the SECONDS register.

17.2.6 SAFETY WINDOW FOR REGISTER READS AND WRITES

The RTCSYNC bit indicates a time window during which the RTCC Clock Domain registers can be safely read and written without concern about a rollover. When RTCSYNC = 0, the registers can be safely accessed by the CPU.

Whether RTCSYNC = 1 or 0, the user should employ a firmware solution to ensure that the data read did not fall on a rollover boundary, resulting in an invalid or partial read. This firmware solution would consist of reading each register twice and then comparing the two values. If the two values matched, then a rollover did not occur.

17.2.7 WRITE LOCK

In order to perform a write to any of the RTCC Timer registers, the RTCWREN bit (RTCCFG<5>) must be set.

To avoid accidental writes to the RTCC Timer register, it is recommended that the RTCWREN bit (RTCCFG<5>) be kept clear when not writing to the register. For the RTCWREN bit to be set, there is only one instruction cycle time window allowed between the 55h/AA sequence and the setting of RTCWREN. For that reason, it is recommended that users follow the code example in Example 17-1.

EXAMPLE 17-1: SETTING THE RTCWREN BIT

```
movlw    0x55
movwf    EEC0N2
movlw    0xAA
movwf    EEC0N2
bsf      RTCCFG, RTCWREN
```

17.2.8 REGISTER MAPPING

To limit the register interface, the RTCC Timer and Alarm Timer registers are accessed through corresponding Register Pointers. The RTCC Value register window (RTCVALH and RTCVALL) uses the RTCPTRx bits (RTCCFG<1:0>) to select the required Timer register pair.

By reading or writing to the RTCVALH register, the RTCC Pointer value (RTCPTR<1:0>) decrements by '1' until it reaches '00'. When '00' is reached, the MINUTES and SECONDS value is accessible through RTCVALH and RTCVALL until the pointer value is manually changed.

TABLE 17-3: RTCVALH AND RTCVALL REGISTER MAPPING

RTCPTR<1:0>	RTCC Value Register Window	
	RTCVALH	RTCVALL
00	MINUTES	SECONDS
01	WEEKDAY	HOURS
10	MONTH	DAY
11	—	YEAR

The Alarm Value register windows (ALRMVALH and ALRMVALL) use the ALRMPTR bits (ALRMCFG<1:0>) to select the desired alarm register pair.

By reading or writing to the ALRMVALH register, the Alarm Pointer value, ALRMPTR<1:0>, decrements by one until it reaches '00'. When it reaches '00', the ALRMMIN and ALRMSEC value is accessible through ALRMVALH and ALRMVALL until the pointer value is manually changed.

PIC18F87K90 FAMILY

20.2 LCD Clock Source Selection

The LCD driver module has three possible clock sources:

- $(F_{osc}/4)/8192$
- SOSC Clock/32
- INTRC/32

The first clock source is the system clock divided by 8,192 ($(F_{osc}/4)/8192$). This divider ratio is chosen to provide about 1 kHz output when the system clock is 8 MHz. The divider is not programmable. Instead, the LCD prescaler bits, LCDPS<3:0>, are used to set the LCD frame clock rate.

The second clock source is the SOSC oscillator/32. This also outputs about 1 kHz when a 32.768 kHz crystal is used with the SOSC oscillator. To use the SOSC oscillator as a clock source, set the SOSCEN (T1CON<3>) bit.

The third clock source is a 31.25 kHz internal RC oscillator/32 that provides approximately 1 kHz output.

The second and third clock sources may be used to continue running the LCD while the processor is in Sleep.

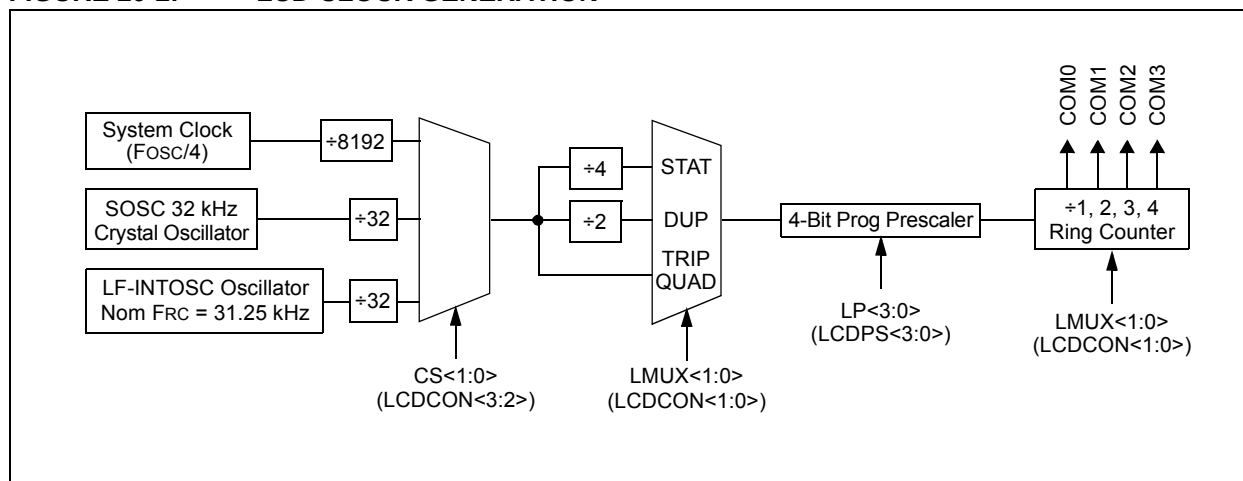
These clock sources are selected through the bits CS<1:0> (LCDCON<3:2>).

20.2.1 LCD PRESCALER

A 16-bit counter is available as a prescaler for the LCD clock. The prescaler is not directly readable or writable. Its value is set by the LP<3:0> bits (LCDPS<3:0>) that determines the prescaler assignment and prescale ratio.

Selectable prescale values are from 1:1 through 1:32,768, in power-of-2 increments.

FIGURE 20-2: LCD CLOCK GENERATION



PIC18F87K90 FAMILY

REGISTER 21-2: SSPxCON1: MSSPx CONTROL REGISTER 1 (SPI MODE)

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
WCOL	SSPOV ⁽¹⁾	SSPEN ⁽²⁾	CKP	SSPM3 ⁽³⁾	SSPM2 ⁽³⁾	SSPM1 ⁽³⁾	SSPM0 ⁽³⁾
bit 7							bit 0

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

-n = Value at POR

'1' = Bit is set

'0' = Bit is cleared

x = Bit is unknown

- bit 7 **WCOL:** Write Collision Detect bit
1 = The SSPxBUF register is written while it is still transmitting the previous word (must be cleared in software)
0 = No collision
- bit 6 **SSPOV:** Receive Overflow Indicator bit⁽¹⁾
SPI Slave mode:
1 = A new byte is received while the SSPxBUF register is still holding the previous data. In case of overflow, the data in SSPxSR is lost. Overflow can only occur in Slave mode. The user must read the SSPxBUF, even if only transmitting data, to avoid setting overflow (must be cleared in software).
0 = No overflow
- bit 5 **SSPEN:** Master Synchronous Serial Port Enable bit⁽²⁾
1 = Enables serial port and configures SCKx, SDOx, SDIx and $\overline{SSx}$ as serial port pins
0 = Disables serial port and configures these pins as I/O port pins
- bit 4 **CKP:** Clock Polarity Select bit
1 = Idle state for clock is a high level
0 = Idle state for clock is a low level
- bit 3-0 **SSPM<3:0>:** Master Synchronous Serial Port Mode Select bits⁽³⁾
1010 = SPI Master mode: Clock = Fosc/8
0101 = SPI Slave mode: Clock = SCKx pin; $\overline{SSx}$ pin control is disabled; $\overline{SSx}$ can be used as an I/O pin
0100 = SPI Slave mode: Clock = SCKx pin; $\overline{SSx}$ pin control is enabled
0011 = SPI Master mode: Clock = TMR2 Output/2
0010 = SPI Master mode: Clock = Fosc/64
0001 = SPI Master mode: Clock = Fosc/16
0000 = SPI Master mode: Clock = Fosc/4

Note 1: In Master mode, the overflow bit is not set since each new reception (and transmission) is initiated by writing to the SSPxBUF register.

2: When enabled, these pins must be properly configured as inputs or outputs.

3: Bit combinations not specifically listed here are either reserved or implemented in I²C™ mode only.

21.4.8 I²C™ MASTER MODE START CONDITION TIMING

To initiate a Start condition, the user sets the Start Enable bit, SEN (SSPxCON2<0>). If the SDAx and SCLx pins are sampled high, the Baud Rate Generator is reloaded with the contents of SSPxADD<6:0> and starts its count. If SCLx and SDAx are both sampled high when the Baud Rate Generator times out (TBRG), the SDAx pin is driven low. The action of the SDAx being driven low while SCLx is high is the Start condition and causes the S bit (SSPxSTAT<3>) to be set. Following this, the Baud Rate Generator is reloaded with the contents of SSPxADD<6:0> and resumes its count. When the Baud Rate Generator times out (TBRG), the SEN bit (SSPxCON2<0>) will be automatically cleared by hardware. The Baud Rate Generator is suspended, leaving the SDAx line held low and the Start condition is complete.

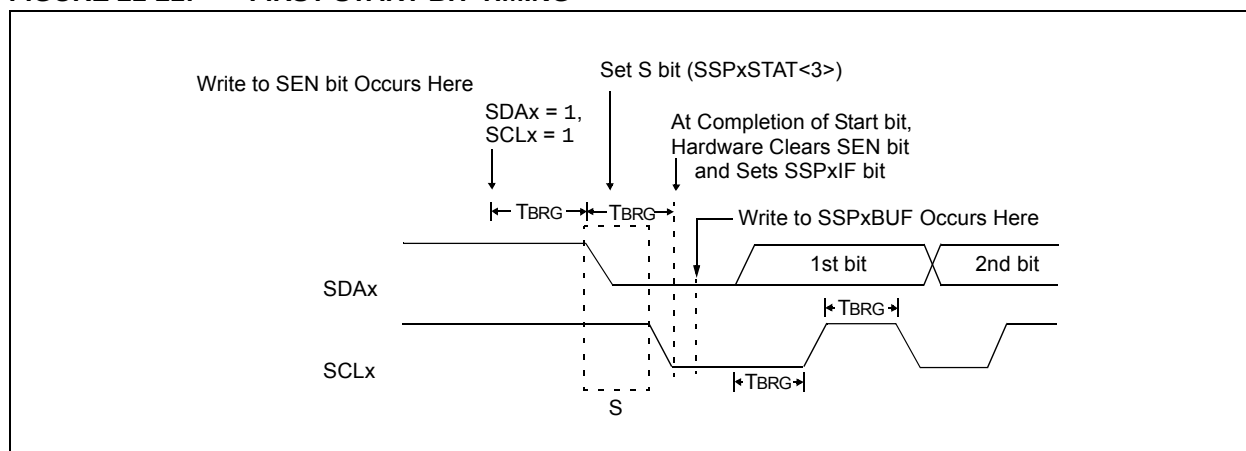
Note: If, at the beginning of the Start condition, the SDAx and SCLx pins are already sampled low, or if during the Start condition, the SCLx line is sampled low before the SDAx line is driven low, a bus collision occurs, the Bus Collision Interrupt Flag, BCLxIF, is set, the Start condition is aborted and the I²C module is reset into its Idle state.

21.4.8.1 WCOL Status Flag

If the user writes the SSPxBUF when a Start sequence is in progress, the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).

Note: Because queueing of events is not allowed, writing to the lower 5 bits of SSPxCON2 is disabled until the Start condition is complete.

FIGURE 21-21: FIRST START BIT TIMING



21.4.10 I²C™ MASTER MODE TRANSMISSION

Transmission of a data byte, a 7-bit address or the other half of a 10-bit address, is accomplished by simply writing a value to the SSPxBUF register. This action will set the Buffer Full flag bit, BF, and allow the Baud Rate Generator to begin counting and start the next transmission. Each bit of address/data will be shifted out onto the SDAx pin after the falling edge of SCLx is asserted (see data hold time specification Parameter 106). SCLx is held low for one Baud Rate Generator rollover count (TBRG). Data should be valid before SCLx is released high (see data setup time specification Parameter 107). When the SCLx pin is released high, it is held that way for TBRG. The data on the SDAx pin must remain stable for that duration and some hold time after the next falling edge of SCLx. After the eighth bit is shifted out (the falling edge of the eighth clock), the BF flag is cleared and the master releases SDAx. This allows the slave device being addressed to respond with an ACK bit during the ninth bit time if an address match occurred, or if data was received properly. The status of ACK is written into the ACKDT bit on the falling edge of the ninth clock. If the master receives an Acknowledge, the Acknowledge Status bit, ACKSTAT, is cleared; if not, the bit is set. After the ninth clock, the SSPxIF bit is set and the master clock (Baud Rate Generator) is suspended until the next data byte is loaded into the SSPxBUF, leaving SCLx low and SDAx unchanged (Figure 21-23).

After the write to the SSPxBUF, each bit of the address will be shifted out on the falling edge of SCLx until all seven address bits and the R/W bit are completed. On the falling edge of the eighth clock, the master will deassert the SDAx pin, allowing the slave to respond with an Acknowledge. On the falling edge of the ninth clock, the master will sample the SDAx pin to see if the address was recognized by a slave. The status of the ACK bit is loaded into the ACKSTAT status bit (SSPxCON2<6>). Following the falling edge of the ninth clock transmission of the address, the SSPxIF flag is set, the BF flag is cleared and the Baud Rate Generator is turned off until another write to the SSPxBUF takes place, holding SCLx low and allowing SDAx to float.

21.4.10.1 BF Status Flag

In Transmit mode, the BF bit (SSPxSTAT<0>) is set when the CPU writes to SSPxBUF and is cleared when all 8 bits are shifted out.

21.4.10.2 WCOL Status Flag

If the user writes the SSPxBUF when a transmit is already in progress (i.e., SSPxSR is still shifting out a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur) after

2 Tcy after the SSPxBUF write. If SSPxBUF is rewritten within 2 Tcy, the WCOL bit is set and SSPxBUF is updated. This may result in a corrupted transfer.

The user should verify that the WCOL bit is clear after each write to SSPxBUF to ensure the transfer is correct. In all cases, WCOL must be cleared in software.

21.4.10.3 ACKSTAT Status Flag

In Transmit mode, the ACKSTAT bit (SSPxCON2<6>) is cleared when the slave has sent an Acknowledge (ACK = 0) and is set when the slave does not Acknowledge (ACK = 1). A slave sends an Acknowledge when it has recognized its address (including a general call), or when the slave has properly received its data.

21.4.11 I²C MASTER MODE RECEPTION

Master mode reception is enabled by programming the Receive Enable bit, RCEN (SSPxCON2<3>).

Note: The MSSP module must be in an inactive state before the RCEN bit is set or the RCEN bit will be disregarded.

The Baud Rate Generator begins counting, and on each rollover, the state of the SCLx pin changes (high-to-low/low-to-high) and data is shifted into the SSPxSR. After the falling edge of the eighth clock, the receive enable flag is automatically cleared, the contents of the SSPxSR are loaded into the SSPxBUF, the BF flag bit is set, the SSPxIF flag bit is set and the Baud Rate Generator is suspended from counting, holding SCLx low. The MSSP is now in Idle state awaiting the next command. When the buffer is read by the CPU, the BF flag bit is automatically cleared. The user can then send an Acknowledge bit at the end of reception by setting the Acknowledge Sequence Enable bit, ACKEN (SSPxCON2<4>).

21.4.11.1 BF Status Flag

In receive operation, the BF bit is set when an address or data byte is loaded into SSPxBUF from SSPxSR. It is cleared when the SSPxBUF register is read.

21.4.11.2 SSPOV Status Flag

In receive operation, the SSPOV bit is set when 8 bits are received into the SSPxSR and the BF flag bit is already set from a previous reception.

21.4.11.3 WCOL Status Flag

If the user writes the SSPxBUF when a receive is already in progress (i.e., SSPxSR is still shifting in a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).

PIC18F87K90 FAMILY

21.4.12 ACKNOWLEDGE SEQUENCE TIMING

An Acknowledge sequence is enabled by setting the Acknowledge Sequence Enable bit, ACKEN (SSPxCON2<4>). When this bit is set, the SCLx pin is pulled low and the contents of the Acknowledge data bit are presented on the SDAx pin. If the user wishes to generate an Acknowledge, then the ACKDT bit should be cleared. If not, the user should set the ACKDT bit before starting an Acknowledge sequence. The Baud Rate Generator then counts for one rollover period (TBRG) and the SCLx pin is deasserted (pulled high). When the SCLx pin is sampled high (clock arbitration), the Baud Rate Generator counts for TBRG; the SCLx pin is then pulled low. Following this, the ACKEN bit is automatically cleared, the Baud Rate Generator is turned off and the MSSP module then goes into an inactive state (Figure 21-25).

21.4.12.1 WCOL Status Flag

If the user writes the SSPxBUF when an Acknowledge sequence is in progress, then WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

21.4.13 STOP CONDITION TIMING

A Stop bit is asserted on the SDAx pin at the end of a receive/transmit by setting the Stop Sequence Enable bit, PEN (SSPxCON2<2>). At the end of a receive/transmit, the SCLx line is held low after the falling edge of the ninth clock. When the PEN bit is set, the master will assert the SDAx line low. When the SDAx line is sampled low, the Baud Rate Generator is reloaded and counts down to 0. When the Baud Rate Generator times out, the SCLx pin will be brought high and one TBRG (Baud Rate Generator rollover count) later, the SDAx pin will be deasserted. When the SDAx pin is sampled high while SCLx is high, the P bit (SSPxSTAT<4>) is set. A TBRG later, the PEN bit is cleared and the SSPxIF bit is set (see Figure 21-26).

21.4.13.1 WCOL Status Flag

If the user writes the SSPxBUF when a Stop sequence is in progress, then the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).

FIGURE 21-25: ACKNOWLEDGE SEQUENCE WAVEFORM

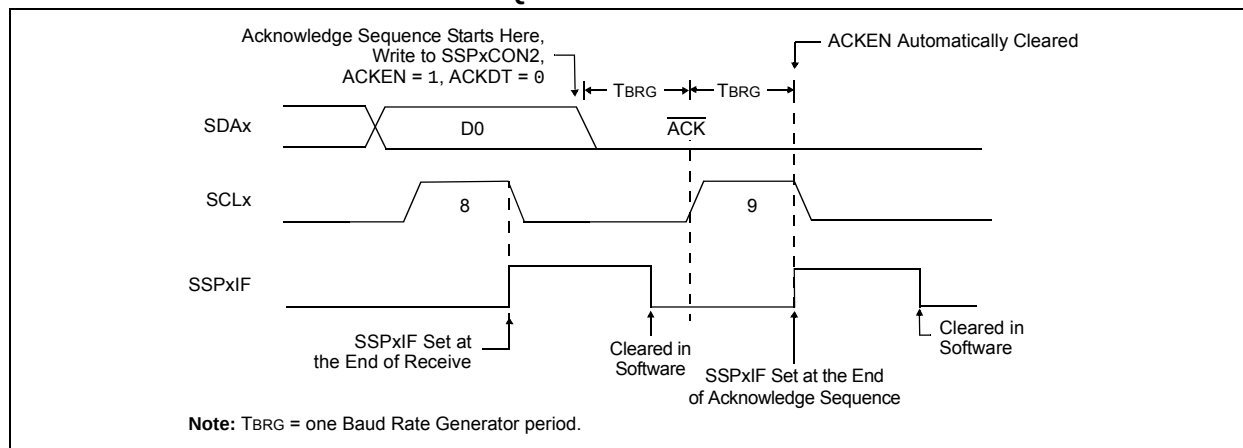
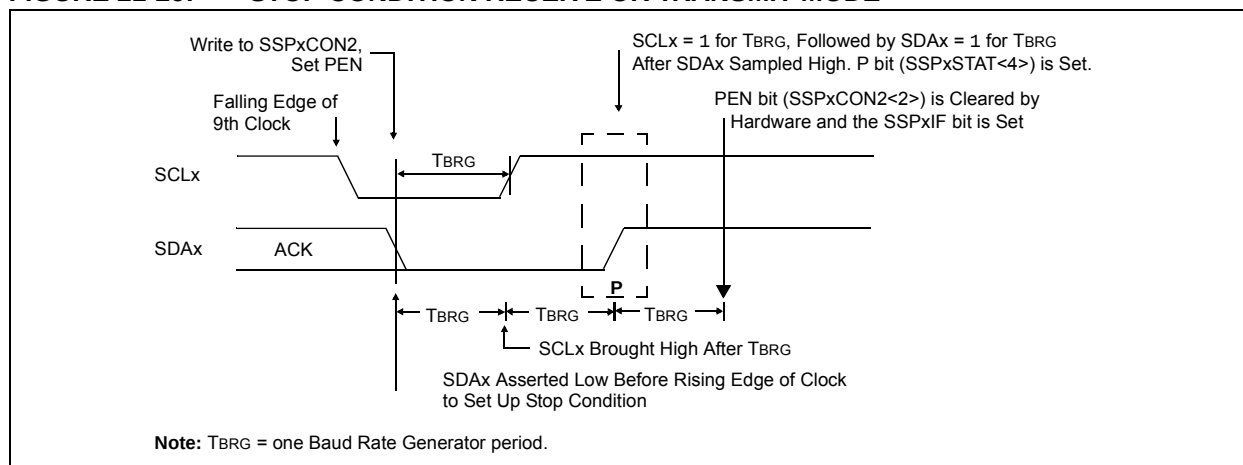
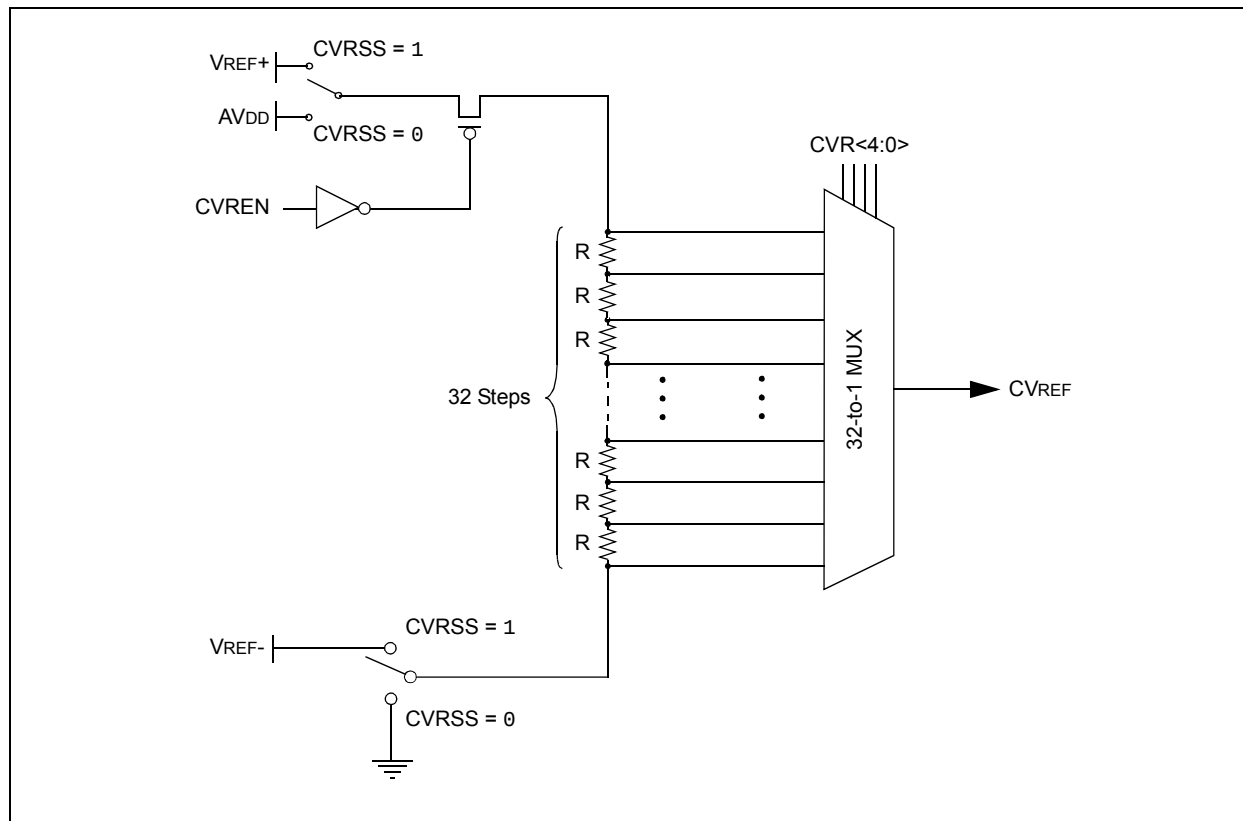


FIGURE 21-26: STOP CONDITION RECEIVE OR TRANSMIT MODE



PIC18F87K90 FAMILY

FIGURE 25-1: COMPARATOR VOLTAGE REFERENCE BLOCK DIAGRAM



25.2 Voltage Reference Accuracy/Error

The full range of voltage reference cannot be realized due to the construction of the module. The transistors on the top and bottom of the resistor ladder network (Figure 25-1) keep CVREF from approaching the reference source rails. The voltage reference is derived from the reference source; therefore, the CVREF output changes with fluctuations in that source. The tested absolute accuracy of the voltage reference can be found in **Section 31.0 “Electrical Characteristics”**.

25.3 Operation During Sleep

When the device wakes up from Sleep through an interrupt or a Watchdog Timer time-out, the contents of the CVRCON register are not affected. To minimize current consumption in Sleep mode, the voltage reference should be disabled.

25.4 Effects of a Reset

A device Reset disables the voltage reference by clearing bit, CVREN (CVRCON<7>). This Reset also disconnects the reference from the RF5 pin by clearing bit, CVROE (CVRCON<6>).

25.5 Connection Considerations

The voltage reference module operates independently of the comparator module. The output of the reference generator may be connected to the RF5 pin if the CVROE bit is set. Enabling the voltage reference output onto RF5, when it is configured as a digital input, will increase current consumption. Connecting RF5 as a digital output, with CVRSS enabled, will also increase current consumption.

The RF5 pin can be used as a simple D/A output with limited drive capability. Due to the limited current drive capability, a buffer must be used on the voltage reference output for external connections to VREF. Figure 25-2 shows an example buffering technique.

27.2 CTMU Operation

The CTMU works by using a fixed current source to charge a circuit. The type of circuit depends on the type of measurement being made.

In the case of charge measurement, the current is fixed and the amount of time the current is applied to the circuit is fixed. The amount of voltage read by the A/D becomes a measurement of the circuit's capacitance.

In the case of time measurement, the current, as well as the capacitance of the circuit, is fixed. In this case, the voltage read by the A/D is representative of the amount of time elapsed from the time the current source starts and stops charging the circuit.

If the CTMU is being used as a time delay, both capacitance and current source are fixed, as well as the voltage supplied to the comparator circuit. The delay of a signal is determined by the amount of time it takes the voltage to charge to the comparator threshold voltage.

27.2.1 THEORY OF OPERATION

The operation of the CTMU is based on the equation for charge:

$$C = I \cdot \frac{dV}{dT}$$

More simply, the amount of charge measured in coulombs in a circuit is defined as current in amperes (I) multiplied by the amount of time in seconds that the current flows (t). Charge is also defined as the capacitance in farads (C) multiplied by the voltage of the circuit (V). It follows that:

$$I \cdot t = C \cdot V$$

The CTMU module provides a constant, known current source. The A/D Converter is used to measure (V) in the equation, leaving two unknowns: capacitance (C) and time (t). The above equation can be used to calculate capacitance or time, by either the relationship using the known fixed capacitance of the circuit:

$$t = (C \cdot V)/I$$

or by:

$$C = (I \cdot t)/V$$

using a fixed time that the current source is applied to the circuit.

27.2.2 CURRENT SOURCE

At the heart of the CTMU is a precision current source, designed to provide a constant reference for measurements. The level of current is user-selectable across three ranges or a total of two orders of magnitude, with the ability to trim the output in $\pm 2\%$ increments (nominal). The current range is selected by the IRNG<1:0> bits (CTMUICON<1:0>), with a value of '00' representing the lowest range.

Current trim is provided by the ITRIM<5:0> bits (CTMUICON<7:2>). These six bits allow trimming of the current source in steps of approximately 2% per step. Half of the range adjusts the current source positively and the other half reduces the current source. A value of '000000' is the neutral position (no change). A value of '100000' is the maximum negative adjustment (approximately -62%) and '011111' is the maximum positive adjustment (approximately +62%).

27.2.3 EDGE SELECTION AND CONTROL

CTMU measurements are controlled by edge events occurring on the module's two input channels. Each channel, referred to as Edge 1 and Edge 2, can be configured to receive input pulses from one of the edge input pins (CTED1 and CTED2) or CCPx Special Event Triggers. The input channels are level-sensitive, responding to the instantaneous level on the channel rather than a transition between levels. The inputs are selected using the EDG1SEL and EDG2SEL bit pairs (CTMUCONL<3:2, 6:5>).

In addition to source, each channel can be configured for event polarity using the EDGE2POL and EDGE1POL bits (CTMUCONL<7,4>). The input channels can also be filtered for an edge event sequence (Edge 1 occurring before Edge 2) by setting the EDGSEQEN bit (CTMUCONH<2>).

27.2.4 EDGE STATUS

The CTMUCON register also contains two status bits, EDG2STAT and EDG1STAT (CTMUCONL<1:0>). Their primary function is to show if an edge response has occurred on the corresponding channel. The CTMU automatically sets a particular bit when an edge response is detected on its channel. The level-sensitive nature of the input channels also means that the status bits become set immediately if the channel's configuration is changed and matches the channel's current state.

The module uses the edge status bits to control the current source output to external analog modules (such as the A/D Converter). Current is only supplied to external modules when only one (not both) of the status bits is set. Current is shut off when both bits are either set or cleared. This allows the CTMU to measure current only during the interval between edges. After both status bits are set, it is necessary to clear them before another measurement is taken. Both bits should be cleared simultaneously, if possible, to avoid re-enabling the CTMU current source.

In addition to being set by the CTMU hardware, the edge status bits can also be set by software. This permits a user application to manually enable or disable the current source. Setting either (but not both) of the bits enables the current source. Setting or clearing both bits at once disables the source.

PIC18F87K90 FAMILY

REGISTER 28-10: CONFIG6L: CONFIGURATION REGISTER 6 LOW (BYTE ADDRESS 30000Ah)⁽²⁾

R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1
WRT7 ⁽¹⁾	WRT6 ⁽¹⁾	WRT5 ⁽¹⁾	WRT4 ⁽¹⁾	WRT3	WRT2	WRT1	WRT0
bit 7							bit 0

Legend:	C = Clearable bit	U = Unimplemented bit, read as '0'
R = Readable bit	W = Writable bit	'0' = Bit is cleared
-n = Value at POR	'1' = Bit is set	x = Bit is unknown

bit 7	WRT7: Write Protection bit ⁽¹⁾ 1 = Block 7 is not write-protected 0 = Block 7 is write-protected
bit 6	WRT6: Write Protection bit ⁽¹⁾ 1 = Block 6 is not write-protected 0 = Block 6 is write-protected
bit 5	WRT5: Write Protection bit ⁽¹⁾ 1 = Block 5 is not write-protected 0 = Block 5 is write-protected
bit 4	WRT4: Write Protection bit ⁽¹⁾ 1 = Block 4 is not write-protected 0 = Block 4 is write-protected
bit 3	WRT3: Write Protection bit 1 = Block 3 is not write-protected 0 = Block 3 is write-protected
bit 2	WRT2: Write Protection bit 1 = Block 2 is not write-protected 0 = Block 2 is write-protected
bit 1	WRT1: Write Protection bit 1 = Block 1 is not write-protected 0 = Block 1 is write-protected
bit 0	WRT0: Write Protection bit 1 = Block 0 is not write-protected 0 = Block 0 is write-protected

Note 1: This bit is only available on PIC18F67K90 and PIC18F87K90.

2: For the memory size of the blocks, refer to Figure 28-6.

PIC18F87K90 FAMILY

29.1.1 STANDARD INSTRUCTION SET

ADDLW ADD Literal to W

Syntax: ADDLW k

Operands: $0 \leq k \leq 255$

Operation: $(W) + k \rightarrow W$

Status Affected: N, OV, C, DC, Z

Encoding:

0000	1111	kkkk	kkkk
------	------	------	------

Description: The contents of W are added to the 8-bit literal 'k' and the result is placed in W.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'k'	Process Data	Write to W

Example: ADDLW 15h

Before Instruction
W = 10h
After Instruction
W = 25h

ADDWF ADD W to f

Syntax: ADDWF f {,d {,a}}

Operands: $0 \leq f \leq 255$
 $d \in [0, 1]$
 $a \in [0, 1]$

Operation: $(W) + (f) \rightarrow \text{dest}$

Status Affected: N, OV, C, DC, Z

Encoding:

0010	01da	ffff	ffff
------	------	------	------

Description: Add W to register 'f'. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.

If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.

If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \leq 95$ (5Fh). See **Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode"** for details.

Words: 1
Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

Example: ADDWF REG, 0, 0

Before Instruction
W = 17h
REG = 0C2h
After Instruction
W = 0D9h
REG = 0C2h

Note: All PIC18 instructions may take an optional label argument preceding the instruction mnemonic for use in symbolic addressing. If a label is used, the instruction format then becomes: {label} instruction argument(s).

PIC18F87K90 FAMILY

LFSR Load FSR

Syntax: LFSR f, k

Operands: $0 \leq f \leq 2$
 $0 \leq k \leq 4095$

Operation: $k \rightarrow \text{FSRf}$

Status Affected: None

Encoding:

1110	1110	00ff	k ₁₁ kkk
1111	0000	k ₇ kkk	kkkk

Description: The 12-bit literal 'k' is loaded into the file select register pointed to by 'f'.

Words: 2

Cycles: 2

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'k' MSB	Process Data	Write literal 'k' MSB to FSRfH
Decode	Read literal 'k' LSB	Process Data	Write literal 'k' to FSRfL

Example: LFSR 2, 3ABh

After Instruction

FSR2H = 03h
 FSR2L = ABh

MOVF Move f

Syntax: MOVF f {,d {,a}}

Operands: $0 \leq f \leq 255$
 $d \in [0, 1]$
 $a \in [0, 1]$

Operation: $f \rightarrow \text{dest}$

Status Affected: N, Z

Encoding:

0101	00da	ffff	ffff
------	------	------	------

Description: The contents of register 'f' are moved to a destination dependent upon the status of 'd'. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f'. Location 'f' can be anywhere in the 256-byte bank.

If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.

If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \leq 95$ (5Fh). See **Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode"** for details.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write W

Example: MOVF REG, 0, 0

Before Instruction

REG = 22h
 W = FFh

After Instruction

REG = 22h
 W = 22h

PIC18F87K90 FAMILY

31.2 DC Characteristics: Power-Down and Supply Current PIC18F87K90 Family (Industrial/Extended)

PIC18F87K90 Family		Standard Operating Conditions (unless otherwise stated)			
		Operating temperature			
		-40°C ≤ TA ≤ +85°C for industrial			
		-40°C ≤ TA ≤ +125°C for extended			
Param No.	Device	Typ	Max	Units	Conditions
	Power-Down Current (IDD)⁽¹⁾				
	All devices	10	500	nA	-40°C
		20	500	nA	+25°C
		120	600	nA	+60°C
		630	1800	nA	+85°C
		4	9	μA	+125°C
	All devices	50	700	μA	-40°C
		60	700	nA	+25°C
		170	800	nA	+60°C
		700	2700	nA	+85°C
		5	11	μA	+125°C
	All devices	350	1300	nA	-40°C
		400	1400	nA	+25°C
		550	1500	nA	+60°C
		1350	4000	nA	+85°C
		6	12	μA	+125°C

- Note 1:** The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or VSS, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).
- 2:** The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.
The test conditions for all IDD measurements in active operation mode are:
OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;
MCLR = VDD; WDT enabled/disabled as specified.
- 3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4:** Voltage regulator disabled (ENVREG = 0, tied to VSS, $\overline{\text{RETEN}}$ (CONFIG1L<0>) = 1).
- 5:** Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and $\overline{\text{RETEN}}$ (CONFIG1L<0>) = 0).
- 6:** LCD glass is not connected; resistor current is not included.
- 7:** 48 MHz maximum frequency at 125°C.

PIC18F87K90 FAMILY

31.2 DC Characteristics: Power-Down and Supply Current PIC18F87K90 Family (Industrial/Extended) (Continued)

PIC18F87K90 Family		Standard Operating Conditions (unless otherwise stated)			
		Operating temperature			
		-40°C ≤ TA ≤ +85°C for industrial			
		-40°C ≤ TA ≤ +125°C for extended			
Param No.	Device	Typ	Max	Units	Conditions
Supply Current (I_{DD}) Cont.^(2,3)					
	All devices	130	390	μA	-40°C
		130	390	μA	+25°C
		130	390	μA	+85°C
		250	500	μA	+125°C
	All devices	270	790	μA	-40°C
		270	790	μA	+25°C
		270	790	μA	+85°C
		400	900	μA	+125°C
	All devices	430	990	μA	-40°C
		450	980	μA	+25°C
		460	980	μA	+85°C
		600	1300	μA	+125°C
	All devices	430	860	μA	-40°C
		530	900	μA	+25°C
		490	880	μA	+85°C
		750	1600	μA	+125°C
	All devices	850	1750	μA	-40°C
		850	1700	μA	+25°C
		850	1800	μA	+85°C
		1150	2400	μA	+125°C
	All devices	1.1	2.7	mA	-40°C
		1.1	2.6	mA	+25°C
		1.1	2.6	mA	+85°C
		2.0	4.0	mA	+125°C
	All devices	12	19	mA	-40°C
		12	19	mA	+25°C
		12	19	mA	+85°C
		13	22	mA	+125°C ⁽⁷⁾
	All devices	13	20	mA	-40°C
		13	20	mA	+25°C
		13	20	mA	+85°C
		14	23	mA	+125°C ⁽⁷⁾

- Note 1:** The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to V_{DD} or V_{SS}, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).
- 2:** The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.
The test conditions for all I_{DD} measurements in active operation mode are:
OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to V_{DD};
MCLR = V_{DD}; WDT enabled/disabled as specified.
- 3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4:** Voltage regulator disabled (ENVREG = 0, tied to V_{SS}, $\overline{\text{RETEN}}$ (CONFIG1L<0>) = 1).
- 5:** Voltage regulator enabled (ENVREG = 1, tied to V_{DD}, SRETEN (WDTCON<4>) = 1 and $\overline{\text{RETEN}}$ (CONFIG1L<0>) = 0).
- 6:** LCD glass is not connected; resistor current is not included.
- 7:** 48 MHz maximum frequency at 125°C.

PIC18F87K90 FAMILY

RC3/SCK1/SCL1/SEG17	17, 26	POP	480
RC4/SDI1/SDA1/SEG16	17, 26	POR. See Power-on Reset.	
RC5/SDO1/SEG12	17, 26	PORTA	
RC6/TX1/CK1/SEG27	17, 26	Associated Registers	159
RC7/RX1/DT1/SEG28	17, 26	LATA Register	157
RD0/SEG0/CTPLS	18, 27	PORTA Register	157
RD1/SEG1/T5CKI/T7G	18, 27	TRISA Register	157
RD2/SEG2	18, 27	PORTB	
RD3/SEG3	18, 27	Associated Registers	162
RD4/SEG4/SDO2	18, 27	LATB Register	160
RD5/SEG5/SDI2/SDA2	18, 27	PORTB Register	160
RD6/SEG6/SCK2/SCL2	18, 27	RB7:RB4 Interrupt-on-Change Flag (RBIF Bit)	160
RD7/SEG7/SS2	18, 27	TRISB Register	160
RE0/LCDBIAS1/P2D	19, 28	PORTC	
RE1/LCDBIAS2/P2C	19, 28	Associated Registers	165
RE2/LCDBIAS3/P2B/CCP10	19, 28	LATC Register	163
RE3/COM0/P3C/CCP9/REFO	19, 28	PORTC Register	163
RE4/COM1/P3B/CCP8	19, 28	TRISC Register	163
RE5/COM2/P1C/CCP7	19, 28	PORTD	
RE6/COM3/P1B/CCP6	19, 28	Associated Registers	168
RE7/ECCP2/P2A/SEG31	28	LATD Register	166
RE7/ECCP2/SEG31/P2A	19	PORTD Register	166
RF1/AN6/C2OUT/SEG19/CTDIN	20, 29	TRISD Register	166
RF2/AN7/C1OUT/SEG20	20	PORTE	
RF2/AN7/C1OUT/SEG20/CTMUI	29	Associated Registers	171
RF3/AN8/SEG21/C2INB	29	LATE Register	169
RF3/AN8/SEG21/C2INB/CTMUI	20	Pins Available in Different LCD Drives	169
RF4/AN9/SEG22/C2INA	20, 29	PORTE Register	169
RF5/AN10	20	TRISE Register	169
RF5/AN10/CVREF/SEG23/C1INB	29	PORTF	
RF6/AN11/SEG24/C1INA	20, 29	Associated Registers	174
RF7/AN5/SS1/SEG25	20, 29	LATF Register	172
RG0/ECCP3/P3A	21, 30	PORTF Register	172
RG1/TX2/CK2/AN19/C3OUT	21, 30	TRISF Register	172
RG2/RX2/DT2/AN18/C3INA	21, 30	PORTG	
RG3/CCP4/AN17/P3D/C3INB	21, 30	Associated Registers	176
RG4/SEG26/RTCC/T7CKI/T5G/CCP5/AN16/P1D/ C3INC	21, 30	LATG Register	175
RH0/SEG47/AN23	31	PORTG Register	175
RH1/SEG46/AN22	31	TRISG Register	175
RH2/SEG45/AN21	31	PORTH	
RH3/SEG44/AN20	31	Associated Registers	179
RH4/SEG40/CCP9/P3C/AN12/C2INC	31	LATH Register	177
RH5/SEG41/CCP8/P3B/AN13/C2IND	31	PORTH Register	177
RH6/SEG42/CCP7/P1C/AN14/C1INC	31	TRISH Register	177
RH7/SEG43/CCP6/P1B/AN15	32	PORTJ	
RJ0	33	Associated Registers	181
RJ1/SEG33	33	LATJ Register	180
RJ2/SEG34	33	PORTJ Register	180
RJ3/SEG35	33	TRISJ Register	180
RJ4/SEG39	33	Power-Managed Modes	53
RJ5/SEG38	33	and PWM Operation	271
RJ6/SEG37	33	and SPI Operation	311
RJ7/SEG36	33	Clock Transitions and Status Indicators	54
VDD	22, 33	Entering	53
VDDCORE/VCAP	22, 33	Exiting Idle and Sleep Modes	65
Vss	22, 33	by Interrupt	65
Pinout I/O Descriptions		by Reset	65
PIC18F6XK90	14	by WDT Time-out	65
PIC18F8XK90	22	Without an Oscillator Start-up Delay	65
PIR5 Register		Idle Modes	58
TMR12IF Bit	138	PRI_IDLE	59
PLL	48	RC_IDLE	60
HSPLL and ECPLL Oscillator Modes	48	SEC_IDLE	59
Use with HF-INTOSC	49		

PIC18F87K90 FAMILY

Multiple Sleep Commands	54	Reader Response	567
Run Modes	54	Real-Time Clock and Calendar (RTCC)	217
PRI_RUN	54	Alarm	232
RC_RUN	55	Configuring	232
SEC_RUN	54	Interrupt	233
Selecting	53	Mask Settings	233
Sleep Mode	58	Alarm Value Registers (ALRMVAL)	226
OSC1, OSC2 Pin States	52	Clock Source	230
Summary (table)	53	Control Registers	219
Power-on Reset (POR)	71	Digit Carry Rules	230
Power-up Delays	52	Operation	229
Power-up Timer (PWRT)	52, 72	Calibration	232
Time-out Sequence	72	General Functionality	231
Prescaler, Timer0	185	Leap Year	231
Prescaler, Timer2	248	Register Mapping	231
PRI_IDLE Mode	59	Register Reads/Writes, Safety Window	231
PRI_RUN Mode	54	Write Lock	231
Program Counter (PC)	87	Register Interface	229
PCL, PCH and PCU Registers	87	Register Maps	235
PCLATH and PCLATU Registers	87	Alarm Value	235
Program Memory		RTCC Control	235
Extended Instruction Set	106	RTCC Value	235
Hard Memory Vectors	86	Registers	218
Instructions	91	Reset	234
Two-Word	91	Device	234
Interrupt Vector	86	Power-on Reset (POR)	234
Look-up Tables	89	RTCCEN Bit, Write	229
Memory Maps		Sleep Mode	234
Hard Vectors and Configuration Words	86	Value Registers (RTCVAL)	222
Reset Vector	86	Reference Clock Output	51
Program Verification	447	Register File	94
Programming, Device Instructions	451	Register File Summary	97–102
Pulse Steering	268	Registers	
Pulse-Width Modulation. <i>See</i> PWM (CCP Module).		ADCON0 (A/D Control 0)	374
PUSH	480	ADCON1 (A/D Control 1)	375
PUSH and POP Instructions	88	ADCON2 (A/D Control 2)	376
PUSHL	496	ADRESH (A/D Result High Byte Left Justified, ADFM = 0)	378
PWM (CCP Module)		ADRESH (A/D Result High Byte Right Justified, ADFM = 1)	379
Associated Registers with PWM and Timers	249	ADRESL (A/D Result Low Byte Left Justified, ADFM = 0)	378
Duty Cycle	248	ADRESL (A/D Result Low Byte Right Justified, ADFM = 1)	379
Example Frequencies/Resolutions, 40 MHz	248	ALRMCFG (Alarm Configuration)	221
Period	248	ALRMDAY (Alarm Day Value)	226
Setup for Operation	249	ALRMMHR (Alarm Hours Value)	227
TMR12 to PR12 Match Flag (TMR12IF Bit)	138	ALRMMIN (Alarm Minutes Value)	228
TMR2 to PR2 Match	248	ALRMMNTH (Alarm Month Value)	226
TMRx to PRx Match	213	ALRMRPT (Alarm Repeat)	222
PWM (ECCP Module)		ALRMSEC (Alarm Seconds Value)	228
Effects of a Reset	271	ALRMWD (Alarm Weekday Value)	227
Operation in Power-Managed Modes	271	ANCON0 (A/D Port Configuration 0)	380
Operation with Fail-Safe Clock Monitor	271	ANCON1 (A/D Port Configuration 1)	380
Pulse Steering	268	ANCON2 (A/D Port Configuration 2)	381
Steering Synchronization	270	BAUDCONx (Baud Rate Control)	352
PWM Mode. <i>See</i> Enhanced Capture/Compare/PWM.		CCPRxH (CCPx Period High Byte)	240
Q		CCPRxL (CCPx Period Low Byte)	240
Q Clock	248	CCPTMRS0 (CCP Timer Select 0)	253
R		CCPTMRS1 (CCPx Timer Select 1)	238
RAM. <i>See</i> Data Memory.		CCPTMRS2 (CCPx Timer Select 2)	239
RC Oscillator	47	CCPxCON (CCPx Control, CCP4-CCP10)	237
RC_IDLE Mode	60	CCPxCON (Enhanced Capture/Compare/ PWM x Control)	252
RC_RUN Mode	55		
RCALL	481		
RCON Register			
Bit Status During Initialization	74		