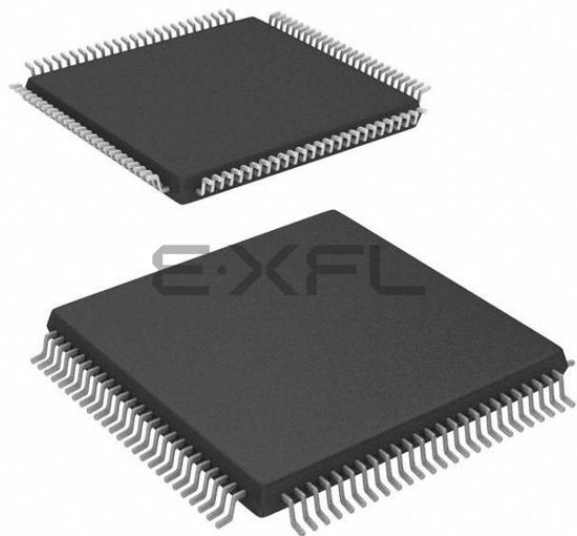




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Details

Product Status	Obsolete
Core Processor	8051
Core Size	8-Bit
Speed	67MHz
Connectivity	EBI/EMI, I ² C, LINbus, SPI, UART/USART
Peripherals	CapSense, DMA, LCD, POR, PWM, WDT
Number of I/O	62
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	4K x 8
Voltage - Supply (Vcc/Vdd)	1.71V ~ 5.5V
Data Converters	A/D 16x20b; D/A 4x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	100-LQFP
Supplier Device Package	100-TQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/infineon-technologies/cy8c3865axi-018

Table 4-3. Data Transfer Instructions *(continued)*

Mnemonic	Description	Bytes	Cycles
MOV Direct, Direct	Move direct byte to direct byte	3	3
MOV Direct, @Ri	Move indirect RAM to direct byte	2	3
MOV Direct, #data	Move immediate data to direct byte	3	3
MOV @Ri, A	Move accumulator to indirect RAM	1	2
MOV @Ri, Direct	Move direct byte to indirect RAM	2	3
MOV @Ri, #data	Move immediate data to indirect RAM	2	2
MOV DPTR, #data16	Load data pointer with 16 bit constant	3	3
MOVC A, @A+DPTR	Move code byte relative to DPTR to accumulator	1	5
MOVC A, @A + PC	Move code byte relative to PC to accumulator	1	4
MOVX A,@Ri	Move external RAM (8-bit) to accumulator	1	4
MOVX A, @DPTR	Move external RAM (16-bit) to accumulator	1	3
MOVX @Ri, A	Move accumulator to external RAM (8-bit)	1	5
MOVX @DPTR, A	Move accumulator to external RAM (16-bit)	1	4
PUSH Direct	Push direct byte onto stack	2	3
POP Direct	Pop direct byte from stack	2	2
XCH A, Rn	Exchange register with accumulator	1	2
XCH A, Direct	Exchange direct byte with accumulator	2	3
XCH A, @Ri	Exchange indirect RAM with accumulator	1	3
XCHD A, @Ri	Exchange low order indirect digit RAM with accumulator	1	3

Table 4-4. Boolean Instructions

Mnemonic	Description	Bytes	Cycles
CLR C	Clear carry	1	1
CLR bit	Clear direct bit	2	3
SETB C	Set carry	1	1
SETB bit	Set direct bit	2	3
CPL C	Complement carry	1	1
CPL bit	Complement direct bit	2	3
ANL C, bit	AND direct bit to carry	2	2
ANL C, /bit	AND complement of direct bit to carry	2	2
ORL C, bit	OR direct bit to carry	2	2
ORL C, /bit	OR complement of direct bit to carry	2	2
MOV C, bit	Move direct bit to carry	2	2
MOV bit, C	Move carry to direct bit	2	3
JC rel	Jump if carry is set	2	3
JNC rel	Jump if no carry is set	2	3
JB bit, rel	Jump if direct bit is set	3	5
JNB bit, rel	Jump if direct bit is not set	3	5
JBC bit, rel	Jump if direct bit is set and clear bit	3	5

In addition to the register or bit address modes used with the lower 48 bytes, the lower 128 bytes can be accessed with direct or indirect addressing. With direct addressing mode, the upper 128 bytes map to the SFRs. With indirect addressing mode, the upper 128 bytes map to RAM. Stack operations use indirect addressing; the 8051 stack space is 256 bytes. See the “Addressing Modes” section on page 11.

5.6.3 SFRs

The SFR space provides access to frequently accessed registers. The memory map for the SFR memory space is shown in [Table 5-2](#).

Table 5-2. SFR Map

Address	0/8	1/9	2/A	3/B	4/C	5/D	6/E	7/F
0xF8	SFRPRT15DR	SFRPRT15PS	SFRPRT15SEL	–	–	–	–	–
0xF0	B	–	SFRPRT12SEL	–	–	–	–	–
0xE8	SFRPRT12DR	SFRPRT12PS	MXAX	–	–	–	–	–
0xE0	ACC	–	–	–	–	–	–	–
0xD8	SFRPRT6DR	SFRPRT6PS	SFRPRT6SEL	–	–	–	–	–
0xD0	PSW	–	–	–	–	–	–	–
0xC8	SFRPRT5DR	SFRPRT5PS	SFRPRT5SEL	–	–	–	–	–
0xC0	SFRPRT4DR	SFRPRT4PS	SFRPRT4SEL	–	–	–	–	–
0xB8				–	–	–	–	–
0xB0	SFRPRT3DR	SFRPRT3PS	SFRPRT3SEL	–	–	–	–	–
0xA8	IE	–	–	–	–	–	–	–
0xA0	P2AX	–	SFRPRT1SEL	–	–	–	–	–
0x98	SFRPRT2DR	SFRPRT2PS	SFRPRT2SEL	–	–	–	–	–
0x90	SFRPRT1DR	SFRPRT1PS	–	DPX0	–	DPX1	–	–
0x88	–	SFRPRT0PS	SFRPRT0SEL	–	–	–	–	–
0x80	SFRPRT0DR	SP	DPL0	DPH0	DPL1	DPH1	DPS	–

The CY8C38 family provides the standard set of registers found on industry standard 8051 devices. In addition, the CY8C38 devices add SFRs to provide direct access to the I/O ports on the device. The following sections describe the SFRs added to the CY8C38 family.

XData Space Access SFRs

The 8051 core features dual DPTR registers for faster data transfer operations. The data pointer select SFR, DPS, selects which data pointer register, DPTR0 or DPTR1, is used for the following instructions:

- MOVX @DPTR, A
- MOVX A, @DPTR
- MOVC A, @A+DPTR
- JMP @A+DPTR
- INC DPTR
- MOV DPTR, #data16

The extended data pointer SFRs, DPX0, DPX1, MXAX, and P2AX, hold the most significant parts of memory addresses during access to the xdata space. These SFRs are used only with the MOVX instructions.

During a MOVX instruction using the DPTR0/DPTR1 register, the most significant byte of the address is always equal to the contents of DPX0/DPX1.

During a MOVX instruction using the R0 or R1 register, the most significant byte of the address is always equal to the contents of MXAX, and the next most significant byte is always equal to the contents of P2AX.

I/O Port SFRs

The I/O ports provide digital input sensing, output drive, pin interrupts, connectivity for analog inputs and outputs, LCD, and access to peripherals through the DSI. Full information on I/O ports is found in [I/O System and Routing](#) on page 28.

I/O ports are linked to the CPU through the PHUB and are also available in the SFRs. Using the SFRs allows faster access to a limited set of I/O port registers, while using the PHUB allows boot configuration and access to all I/O port registers.

Each SFR supported I/O port provides three SFRs:

- SFRPRTxDR sets the output data state of the port (where x is port number and includes ports 0–6, 12 and 15).
- The SFRPRTxSEL selects whether the PHUB PRTxDR register or the SFRPRTxDR controls each pin's output buffer within the port. If a SFRPRTxSEL[y] bit is high, the corresponding SFRPRTxDR[y] bit sets the output state for that pin. If a SFRPRTxSEL[y] bit is low, the corresponding PRTxDR[y] bit sets the output state of the pin (where y varies from 0 to 7).
- The SFRPRTxPS is a read only register that contains pin state values of the port pins.

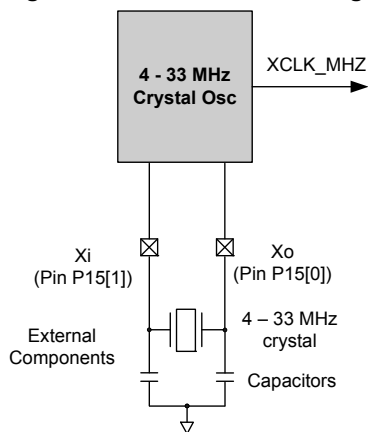
The 33-kHz clock (CLK33K) comes from a divide-by-3 operation on CLK100K. This output can be used as a reduced accuracy version of the 32.768-kHz ECO clock with no need for a crystal.

6.1.2 External Oscillators

6.1.2.1 MHz External Crystal Oscillator

The MHzECO provides high frequency, high precision clocking using an external crystal (see Figure 6-2). It supports a wide variety of crystal types, in the range of 4 to 33 MHz. When used in conjunction with the PLL, it can generate CPU and system clocks up to the device's maximum frequency (see PLL). The GPIO pins connecting to the external crystal and capacitors are fixed. MHzECO accuracy depends on the crystal chosen.

Figure 6-2. MHzECO Block Diagram

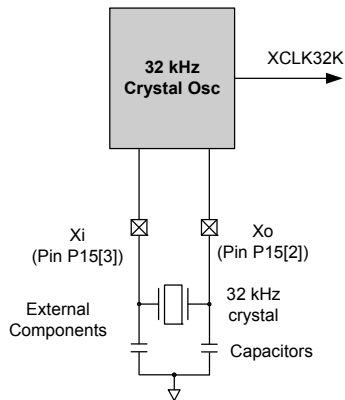


6.1.2.2 32.768-kHz ECO

The 32.768-kHz external crystal oscillator (32kHzECO) provides precision timing with minimal power consumption using an external 32.768-kHz watch crystal (see Figure 6-3). The 32kHzECO also connects directly to the sleep timer and provides the source for the RTC. The RTC uses a 1-second interrupt to implement the RTC functionality in firmware.

The oscillator works in two distinct power modes. This allows users to trade off power consumption with noise immunity from neighboring circuits. The GPIO pins connected to the external crystal and capacitors are fixed.

Figure 6-3. 32kHzECO Block Diagram



6.1.2.3 Digital System Interconnect

The DSI provides routing for clocks taken from external clock oscillators connected to I/O. The oscillators can also be generated within the device in the digital system and UDBs.

While the primary DSI clock input provides access to all clocking resources, up to eight other DSI clocks (internally or externally generated) may be routed directly to the eight digital clock dividers. This is only possible if there are multiple precision clock sources.

6.1.3 Clock Distribution

All seven clock sources are inputs to the central clock distribution system. The distribution system is designed to create multiple high precision clocks. These clocks are customized for the design's requirements and eliminate the common problems found with limited resolution prescalers attached to peripherals. The clock distribution system generates several types of clock trees.

- The system clock is used to select and supply the fastest clock in the system for general system clock requirements and clock synchronization of the PSoC device.
- Bus clock 16-bit divider uses the system clock to generate the system's bus clock used for data transfers. Bus clock is the source clock for the CPU clock divider.
- Eight fully programmable 16-bit clock dividers generate digital system clocks for general use in the digital system, as configured by the design's requirements. Digital system clocks can generate custom clocks derived from any of the seven clock sources for any purpose. Examples include baud rate generators, accurate PWM periods, and timer clocks, and many others. If more than eight digital clock dividers are required, the UDBs and fixed function timer/counter/PWMs can also generate clocks.
- Four 16-bit clock dividers generate clocks for the analog system components that require clocking, such as ADC and mixers. The analog clock dividers include skew control to ensure that critical analog events do not occur simultaneously with digital switching events. This is done to reduce analog system noise.

Each clock divider consists of an 8-input multiplexer, a 16-bit clock divider (divide by 2 and higher) that generates ~50 percent duty cycle clocks, system clock resynchronization logic, and deglitch logic. The outputs from each digital clock tree can be routed into the digital system interconnect and then brought back into the clock system as an input, allowing clock chaining of up to 32 bits.

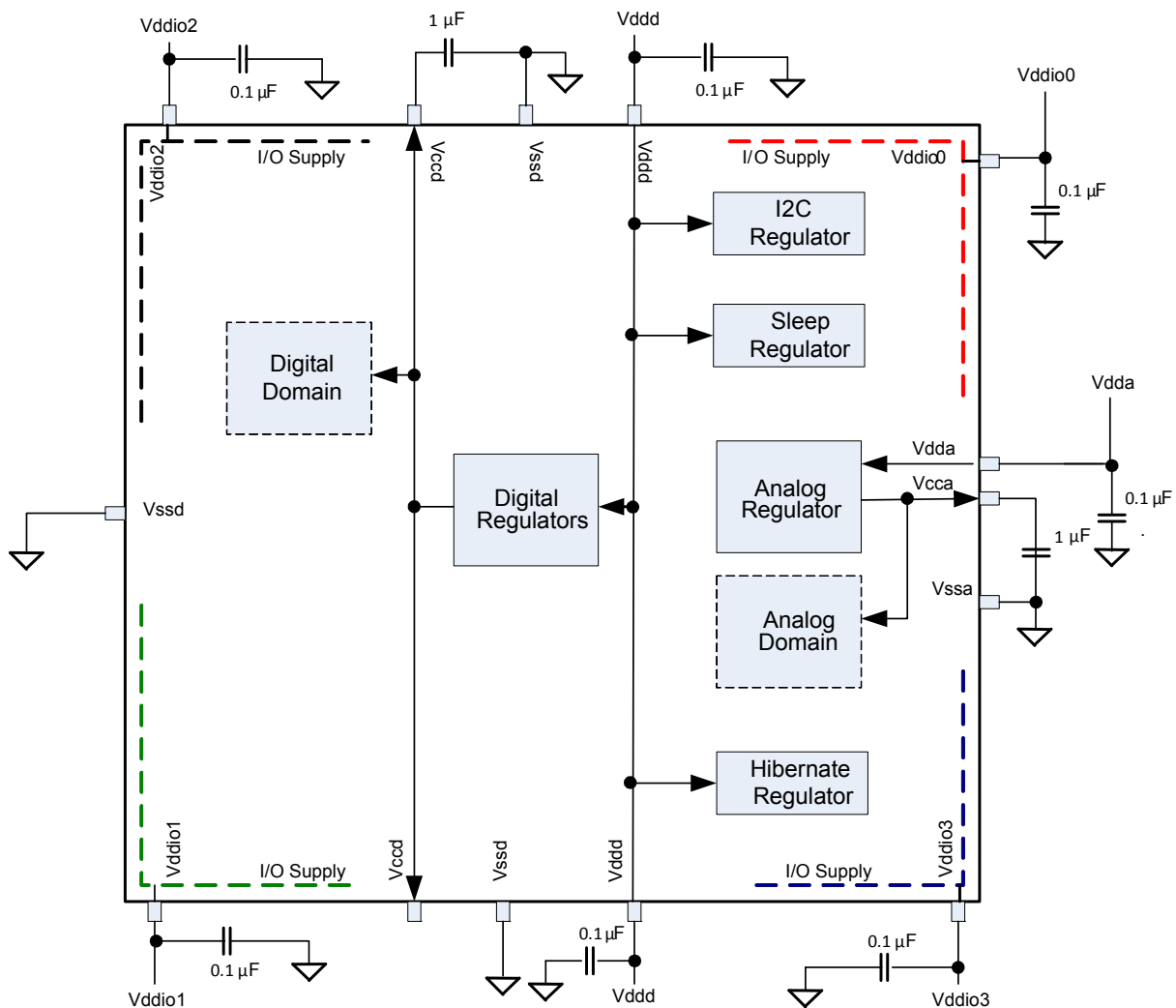
6.1.4 USB Clock Domain

The USB clock domain is unique in that it operates largely asynchronously from the main clock network. The USB logic contains a synchronous bus interface to the chip, while running on an asynchronous clock to process USB data. The USB logic requires a 48 MHz frequency. This frequency can be generated from different sources, including DSI clock at 48 MHz or doubled value of 24 MHz from internal oscillator, DSI signal, or crystal oscillator.

6.2 Power System

The power system consists of separate analog, digital, and I/O supply pins, labeled Vdda, Vddd, and Vddiox, respectively. It also includes two internal 1.8-V regulators that provide the digital (Vccd) and analog (Vcca) supplies for the internal core logic. The output pins of the regulators (Vccd and Vcca) and the Vddio pins must have capacitors connected as shown in [Figure 6-4](#). The two Vccd pins must be shorted together, with as short a trace as possible, and connected to a 1-μF ±10-percent X5R capacitor. The power system also contains a sleep regulator, an I²C regulator, and a hibernate regulator.

Figure 6-4. PSoC Power System



Note The two Vccd pins must be connected together with as short a trace as possible. A trace under the device is recommended, as shown in [Figure 2-6](#) on page 10.

■ **ALVI, DLVI, AHVI** – Analog/digital low voltage interrupt, analog high voltage interrupt

Interrupt circuits are available to detect when V_{DDA} and V_{DDD} go outside a voltage range. For AHVI, V_{DDA} is compared to a fixed trip level. For ALVI and DLVI, V_{DDA} and V_{DDD} are compared to trip levels that are programmable, as listed in Table 6-5. ALVI and DLVI can also be configured to generate a device reset instead of an interrupt.

Table 6-5. Analog/Digital Low Voltage Interrupt, Analog High Voltage Interrupt

Interrupt	Supply	Normal Voltage Range	Available Trip Settings	Accuracy
DLVI	V _{DDD}	1.71 V–5.5 V	1.70 V–5.45 V in 250 mV increments	±2%
ALVI	V _{DDA}	1.71 V–5.5 V	1.70 V–5.45 V in 250 mV increments	±2%
AHVI	V _{DDA}	1.71 V–5.5 V	5.75 V	±2%

The monitors are disabled until after IPOR. During sleep mode these circuits are periodically activated (buzzed). If an interrupt occurs during buzzing then the system first enters its wakeup sequence. The interrupt is then recognized and may be serviced.

6.3.1.2 Other Reset Sources

■ **XRES** – External reset

PSoC 3 has either a single GPIO pin that is configured as an external reset or a dedicated XRES pin. Either the dedicated XRES pin or the GPIO pin, if configured, holds the part in reset while held active (low). The response to an XRES is the same as to an IPOR reset.

The external reset is active low. It includes an internal pull-up resistor. XRES is active during sleep and hibernate modes.

■ **SRES** – Software reset

A reset can be commanded under program control by setting a bit in the software reset register. This is done either directly by the program or indirectly by DMA access. The response to a SRES is the same as after an IPOR reset.

Another register bit exists to disable this function.

■ **DRES** – Digital logic reset

A logic signal can be routed from the UDBs or other digital peripheral source through the DSI to the Configurable XRES pin, P1[2], to generate a hardware-controlled reset. The pin must be placed in XRES mode. The response to a DRES is the same as after an IPOR reset.

■ **WRES** – Watchdog timer reset

The watchdog reset detects when the software program is no longer being executed correctly. To indicate to the watchdog timer that it is running correctly, the program must periodically reset the timer. If the timer is not reset before a user-specified amount of time, then a reset is generated.

Note IPOR disables the watchdog function. The program must enable the watchdog function at an appropriate point in the code by setting a register bit. When this bit is set, it cannot be cleared again except by an IPOR power on reset event.

Note

14. GPIOs with opamp outputs are not recommended for use with CapSense.

6.4 I/O System and Routing

PSoC I/Os are extremely flexible. Every GPIO has analog and digital I/O capability. All I/Os have a large number of drive modes, which are set at POR. PSoC also provides up to four individual I/O voltage domains through the V_{DDIO} pins.

There are two types of I/O pins on every device; those with USB provide a third type. Both GPIO and SIO provide similar digital functionality. The primary differences are their analog capability and drive strength. Devices that include USB also provide two USBIO pins that support specific USB functionality as well as limited GPIO capability.

All I/O pins are available for use as digital inputs and outputs for both the CPU and digital peripherals. In addition, all I/O pins can generate an interrupt. The flexible and advanced capabilities of the PSoC I/O, combined with any signal to any pin routability, greatly simplify circuit design and board layout. All GPIO pins can be used for analog input, CapSense^[14], and LCD segment drive, while SIO pins are used for voltages in excess of V_{DDA} and for programmable output voltages.

■ **Features supported by both GPIO and SIO:**

- User programmable port reset state
- Separate I/O supplies and voltages for up to four groups of I/O
- Digital peripherals use DSI to connect the pins
- Input or output or both for CPU and DMA
- Eight drive modes
- Every pin can be an interrupt source configured as rising edge, falling edge or both edges. If required, level sensitive interrupts are supported through the DSI
- Dedicated port interrupt vector for each port
- Slew rate controlled digital output drive mode
- Access port control and configuration registers on either port basis or pin basis
- Separate port read (PS) and write (DR) data registers to avoid read modify write errors
- Special functionality on a pin by pin basis

■ **Additional features only provided on the GPIO pins:**

- LCD segment drive on LCD equipped devices
- CapSense^[14]
- Analog input and output capability
- Continuous 100 µA clamp current capability
- Standard drive strength down to 1.7 V

■ **Additional features only provided on SIO pins:**

- Higher drive strength than GPIO
- Hot swap capability (5 V tolerance at any operating V_{DD})
- Programmable and regulated high input and output drive levels down to 1.2 V
- No analog input, CapSense, or LCD capability
- Over voltage tolerance up to 5.5 V
- SIO can act as a general purpose analog comparator

■ **USBIO features:**

- Full speed USB 2.0 compliant I/O
- Highest drive strength for general purpose use
- Input, output, or both for CPU and DMA
- Input, output, or both for digital peripherals
- Digital output (CMOS) drive mode
- Each pin can be an interrupt source configured as rising edge, falling edge, or both edges

Figure 6-8. GPIO Block Diagram

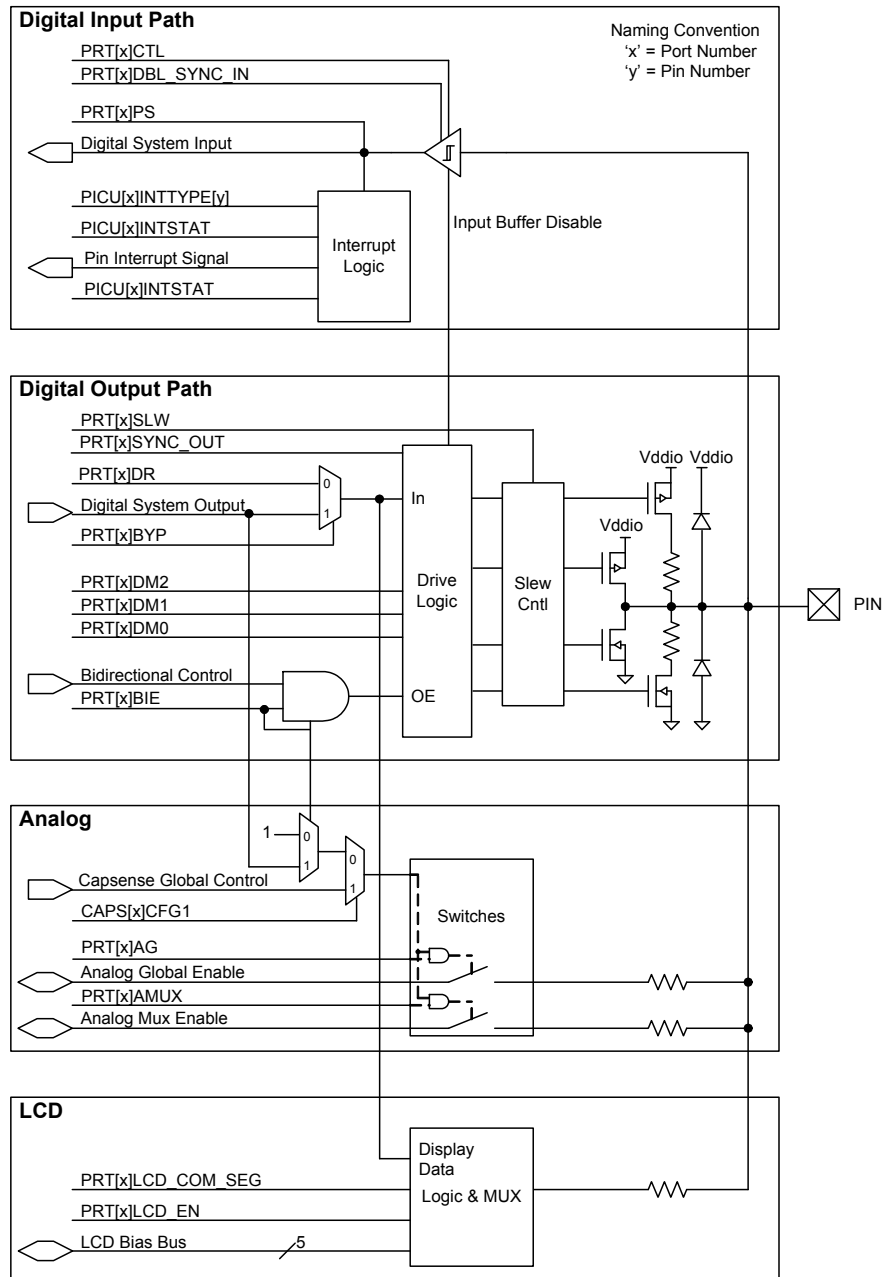


Figure 6-9. SIO Input/Output Block Diagram

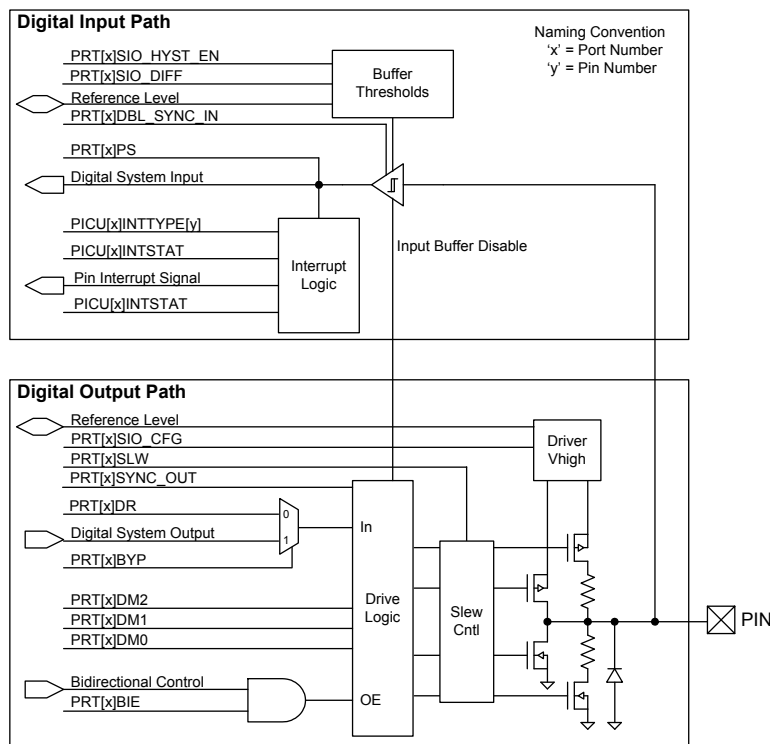
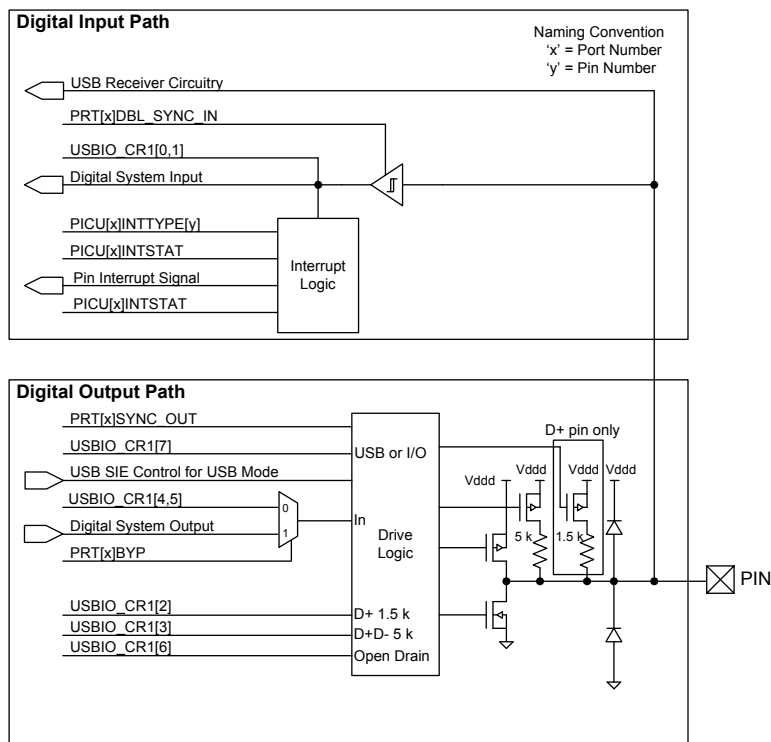


Figure 6-10. USBIO Block Diagram



6.4.8 Analog Connections

These connections apply only to GPIO pins. All GPIO pins may be used as analog inputs or outputs. The analog voltage present on the pin must not exceed the V_{ddio} supply voltage to which the GPIO belongs. Each GPIO may connect to one of the analog global busses or to one of the analog mux buses to connect any pin to any internal analog resource such as ADC or comparators. In addition, select pins provide direct connections to specific analog features such as the high current DACs or uncommitted opamps.

6.4.9 CapSense

This section applies only to GPIO pins. All GPIO pins may be used to create CapSense buttons and sliders^[16]. See the “CapSense” section on page 54 for more information.

6.4.10 LCD Segment Drive

This section applies only to GPIO pins. All GPIO pins may be used to generate Segment and Common drive signals for direct glass drive of LCD glass. See the “LCD Direct Drive” section on page 53 for details.

6.4.11 Adjustable Output Level

This section applies only to SIO pins. SIO port pins support the ability to provide a regulated high output level for interface to external signals that are lower in voltage than the SIO's respective V_{ddio}. SIO pins are individually configurable to output either the standard V_{ddio} level or the regulated output, which is based on an internally generated reference. Typically a voltage DAC (VDAC) is used to generate the reference (see Figure 6-12). The “DAC” section on page 54 has more details on VDAC use and reference routing to the SIO pins. Resistive pullup and pull-down drive modes are not available with SIO in regulated output mode.

6.4.12 Adjustable Input Level

This section applies only to SIO pins. SIO pins by default support the standard CMOS and LVTTTL input levels but also support a differential mode with programmable levels. SIO pins are grouped into pairs. Each pair shares a reference generator block which, is used to set the digital input buffer reference level for interface to external signals that differ in voltage from V_{ddio}. The reference sets the pins voltage threshold for a high logic level (see Figure 6-12). Available input thresholds are:

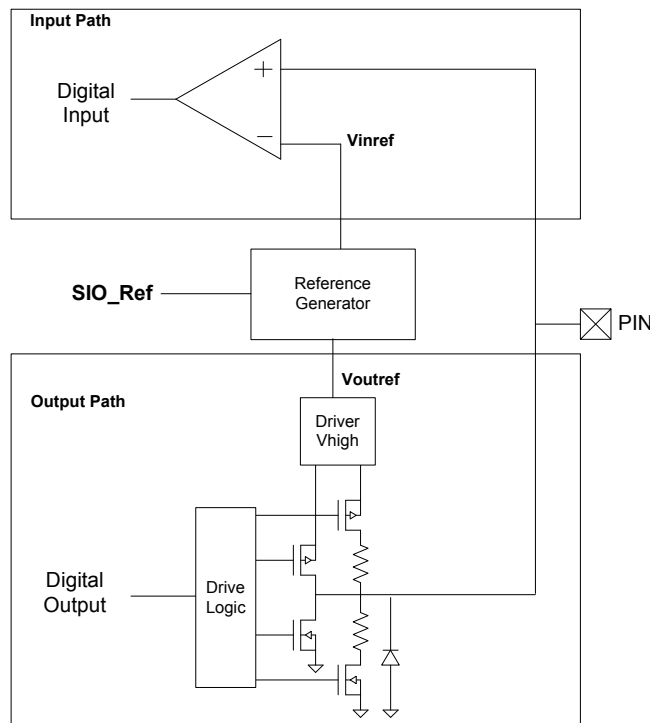
- $0.5 \times V_{ddio}$
- $0.4 \times V_{ddio}$
- $0.5 \times V_{REF}$
- V_{REF}

Typically a voltage DAC (VDAC) generates the V_{REF} reference. “DAC” section on page 54 has more details on VDAC use and reference routing to the SIO pins.

Note

16. GPIOs with opamp outputs are not recommended for use with CapSense.

Figure 6-12. SIO Reference for Input and Output



6.4.13 SIO as Comparator

This section applies only to SIO pins. The adjustable input level feature of the SIOs as explained in the [Adjustable Input Level](#) section can be used to construct a comparator. The threshold for the comparator is provided by the SIO's reference generator. The reference generator has the option to set the analog signal routed through the analog global line as threshold for the comparator. Note that a pair of SIO pins share the same threshold. The digital input path in Figure 6-9 on page 30 illustrates this functionality. In the figure, 'Reference level' is the analog signal routed through the analog global. The hysteresis feature can also be enabled for the input buffer of the SIO, which increases noise immunity for the comparator.

6.4.14 Hot Swap

This section applies only to SIO pins. SIO pins support 'hot swap' capability to plug into an application without loading the signals that are connected to the SIO pins even when no power is applied to the PSoC device. This allows the unpowered PSoC to maintain a high impedance load to the external device while also preventing the PSoC from being powered through a GPIO pin's protection diode.

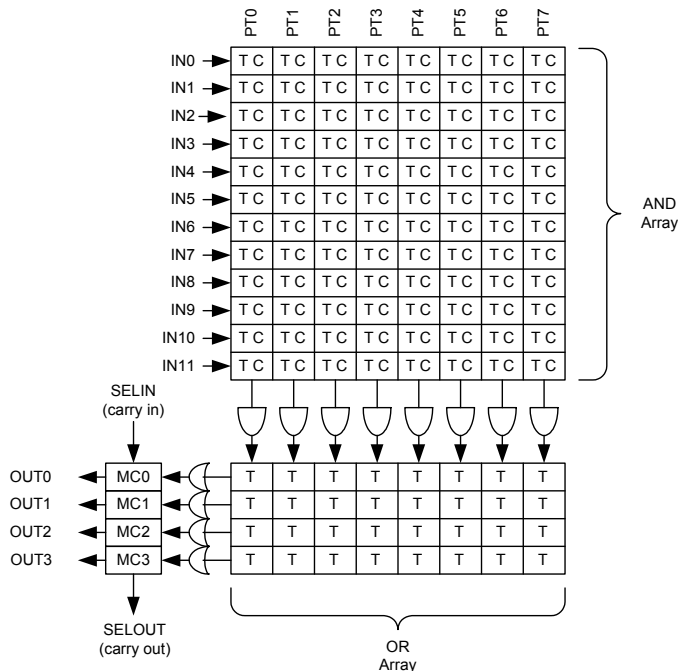
The main component blocks of the UDB are:

- **PLD blocks** – There are two small PLDs per UDB. These blocks take inputs from the routing array and form registered or combinational sum-of-products logic. PLDs are used to implement state machines, state bits, and combinational logic equations. PLD configuration is automatically generated from graphical primitives.
- **Datapath module** – This 8-bit wide datapath contains structured logic to implement a dynamically configurable ALU, a variety of compare configurations and condition generation. This block also contains input/output FIFOs, which are the primary parallel data interface between the CPU/DMA system and the UDB.
- **Status and control module** – The primary role of this block is to provide a way for CPU firmware to interact and synchronize with UDB operation.
- **Clock and reset module** – This block provides the UDB clocks and reset selection and control.

7.2.1 PLD Module

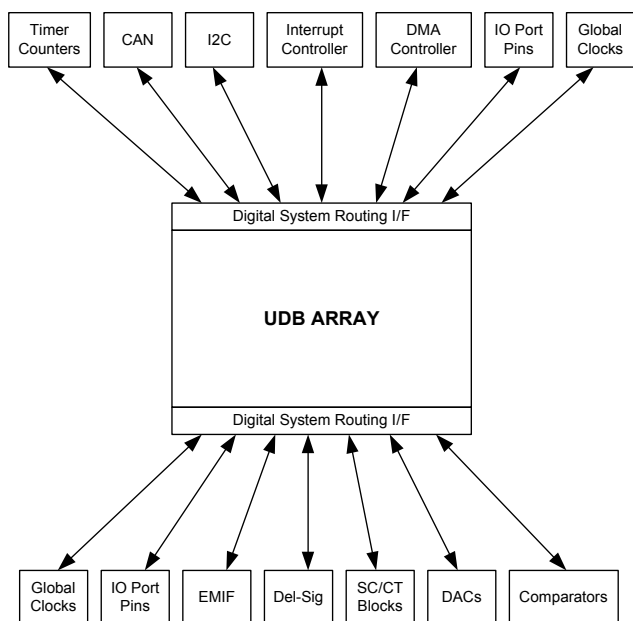
The primary purpose of the PLD blocks is to implement logic expressions, state machines, sequencers, lookup tables, and decoders. In the simplest use model, consider the PLD blocks as a standalone resource onto which general purpose RTL is synthesized and mapped. The more common and efficient use model is to create digital functions from a combination of PLD and datapath blocks, where the PLD implements only the random logic and state portion of the function while the datapath (ALU) implements the more structured elements.

Figure 7-7. PLD 12C4 Structure



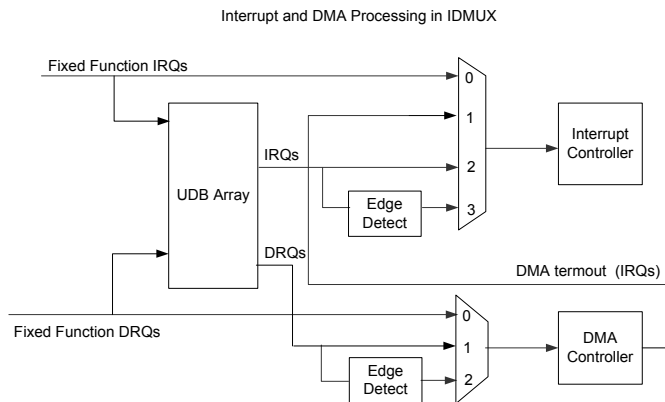
One 12C4 PLD block is shown in Figure 7-7. This PLD has 12 inputs, which feed across eight product terms. Each product term (AND function) can be from 1 to 12 inputs wide, and in a given product term, the true (T) or complement (C) of each input can be selected. The product terms are summed (OR function) to create the PLD outputs. A sum can be from 1 to 8 product terms wide. The 'C' in 12C4 indicates that the width of the OR gate (in this case 8) is constant across all outputs (rather than variable as in a 22V10 device). This PLA like structure gives maximum flexibility and insures that all inputs and outputs are permutable for ease of allocation by the software tools. There are two 12C4 PLDs in each UDB.

Figure 7-13. Digital System Interconnect



Interrupt and DMA routing is very flexible in the CY8C38 programmable architecture. In addition to the numerous fixed function peripherals that can generate interrupt requests, any data signal in the UDB array routing can also be used to generate a request. A single peripheral may generate multiple independent interrupt requests simplifying system and firmware design. Figure 7-14 shows the structure of the IDMUX (Interrupt/DMA Multiplexer).

Figure 7-14. Interrupt and DMA Processing in the IDMUX



7.4.1 I/O Port Routing

There are a total of 20 DSI routes to a typical 8-bit I/O port, 16 for data and four for drive strength control.

When an I/O pin is connected to the routing, there are two primary connections available, an input and an output. In conjunction with drive strength control, this can implement a bidirectional I/O pin. A data output signal has the option to be

single synchronized (pipelined) and a data input signal has the option to be double synchronized. The synchronization clock is the system clock (see Figure 6-1). Normally all inputs from pins are synchronized as this is required if the CPU interacts with the signal or any signal derived from it. Asynchronous inputs have rare uses. An example of this is a feed through of combinational PLD logic from input pins to output pins.

Figure 7-15. I/O Pin Synchronization Routing

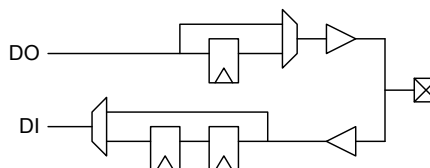
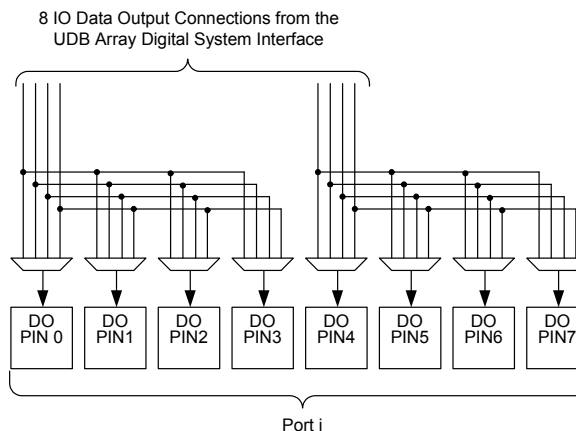
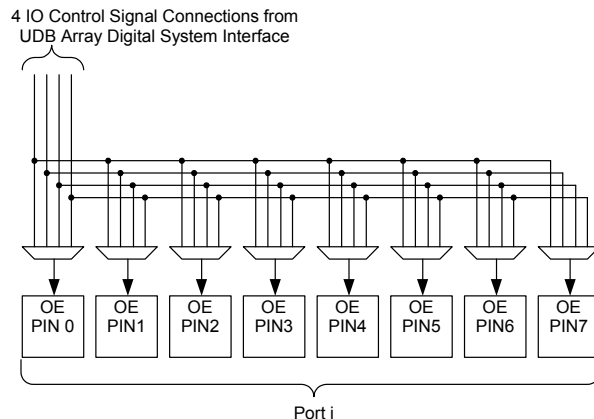


Figure 7-16. I/O Pin Output Connectivity



There are four more DSI connections to a given I/O port to implement dynamic output enable control of pins. This connectivity gives a range of options, from fully ganged 8-bits controlled by one signal, to up to four individually controlled pins. The output enable signal is useful for creating tri-state bidirectional pins and buses.

Figure 7-17. I/O Pin Output Enable Connectivity



Analog local buses (abus) are routing resources located within the analog subsystem and are used to route signals between different analog blocks. There are eight abus routes in CY8C38, four in the left half (abusl [0:3]) and four in the right half (abusr [0:3]) as shown in Figure 8-2. Using the abus saves the analog globals and analog mux buses from being used for interconnecting the analog blocks.

Multiplexers and switches exist on the various buses to direct signals into and out of the analog blocks. A multiplexer can have only one connection on at a time, whereas a switch can have multiple connections on simultaneously. In Figure 8-2, multiplexers are indicated by grayed ovals and switches are indicated by transparent ovals.

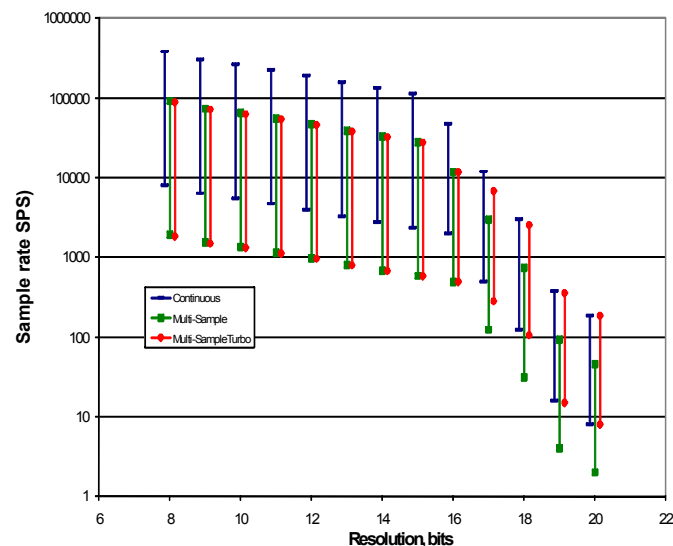
8.2 Delta-sigma ADC

The CY8C38 device contains one delta-sigma ADC. This ADC offers differential input, high resolution and excellent linearity, making it a good ADC choice for both audio signal processing and measurement applications. The converter's nominal operation is 16 bits at 48 ksp/s. The ADC can be configured to output 20-bit resolution at data rates of up to 187 sps. At a fixed clock rate, resolution can be traded for faster data rates as shown in Table 8-1 and Figure 8-3.

Table 8-1. Delta-sigma ADC Performance

Bits	Maximum Sample Rate (sps)	SINAD (dB)
20	187	—
16	48 k	84
12	192 k	66
8	384 k	43

Figure 8-3. Delta-sigma ADC Sample Rates, Range = ± 1.024 V



8.2.1 Functional Description

The ADC connects and configures three basic components, input buffer, delta-sigma modulator, and decimator. The basic block diagram is shown in Figure 8-4. The signal from the input muxes is delivered to the delta-sigma modulator either directly or through the input buffer. The delta-sigma modulator performs the actual analog to digital conversion. The modulator over-samples the input and generates a serial data stream output. This high speed data stream is not useful for most applications without some type of post processing, and so is passed to the decimator through the Analog Interface block. The decimator converts the high speed serial data stream into parallel ADC results. The modulator/decimator frequency response is $[(\sin x)/x]^4$; a typical frequency response is shown in Figure 8-5.

Figure 8-4. Delta-sigma ADC Block Diagram

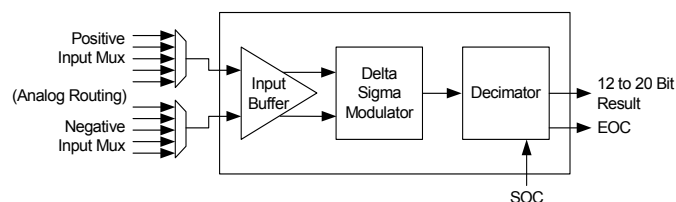
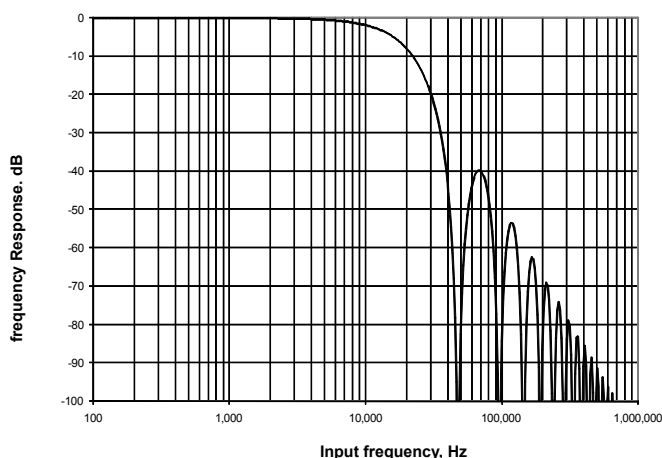


Figure 8-5. Delta-sigma ADC Frequency Response, Normalized to Output, Sample Rate = 48 kHz



Resolution and sample rate are controlled by the Decimator. Data is pipelined in the decimator; the output is a function of the last four samples. When the input multiplexer is switched, the output data is not valid until after the fourth sample after the switch.

8.2.2 Operational Modes

The ADC can be configured by the user to operate in one of four modes: Single Sample, Multi Sample, Continuous, or Multi Sample (Turbo). All four modes are started by either a write to the start bit in a control register or an assertion of the Start of Conversion (SoC) signal. When the conversion is complete, a status bit is set and the output signal End of Conversion (EoC)

10. Development Support

The CY8C38 family has a rich set of documentation, development tools, and online resources to assist you during your development process. Visit psoc.cypress.com/getting-started to find out more.

10.1 Documentation

A suite of documentation, supports the CY8C38 family to ensure that you can find answers to your questions quickly. This section contains a list of some of the key documents.

Software User Guide: A step-by-step guide for using PSoC Creator. The software user guide shows you how the PSoC Creator build process works in detail, how to use source control with PSoC Creator, and much more.

Component Datasheets: The flexibility of PSoC allows the creation of new peripherals (components) long after the device has gone into production. Component datasheets provide all of the information needed to select and use a particular component, including a functional description, API documentation, example code, and AC/DC specifications.

Application Notes: PSoC application notes discuss a particular application of PSoC in depth; examples include brushless DC motor control and on-chip filtering. Application notes often include example projects in addition to the application note document.

Technical Reference Manual: The Technical Reference Manual (TRM) contains all the technical detail you need to use a PSoC device, including a complete description of all PSoC registers.

10.2 Online

In addition to print documentation, the Cypress PSoC forums connect you with fellow PSoC users and experts in PSoC from around the world, 24 hours a day, 7 days a week.

10.3 Tools

With industry standard cores, programming, and debugging interfaces, the CY8C38 family is part of a development tool ecosystem. Visit us at www.cypress.com/go/psoccreator for the latest information on the revolutionary, easy to use PSoC Creator IDE, supported third party compilers, programmers, debuggers, and development kits.

Table 11-2. DC Specifications (continued)

Parameter	Description	Conditions		Min	Typ	Max	Units
	Sleep Mode ^[22]						
	CPU = OFF RTC = ON (= ECO32K ON, in low-power mode) Sleep timer = ON (= ILO ON at 1 kHz) ^[23] WDT = OFF I ² C Wake = OFF Comparator = OFF POR = ON Boost = OFF SIO pins in single ended input, unregulated output mode	V _{DD} = V _{DDIO} = 4.5–5.5 V	T = –40 °C	–	–	–	μA
			T = 25 °C	–	–	–	μA
			T = 85 °C	–	–	–	μA
		V _{DD} = V _{DDIO} = 2.7–3.6 V	T = –40 °C	–	–	–	μA
			T = 25 °C	–	1	–	μA
			T = 85 °C	–	–	–	μA
		V _{DD} = V _{DDIO} = 1.71–1.95 V	T = –40 °C	–	–	–	μA
			T = 25 °C	–	–	–	μA
			T = 85 °C	–	–	–	μA
	Comparator = ON CPU = OFF RTC = OFF Sleep timer = OFF WDT = OFF I2C Wake = OFF POR = ON Boost = OFF SIO pins in single ended input, unregulated output mode	V _{DD} = V _{DDIO} = 2.7–3.6V	T = 25 °C	–	–	–	μA
	I2C Wake = ON CPU = OFF RTC = OFF Sleep timer = OFF WDT = OFF Comparator = OFF POR = ON Boost = OFF SIO pins in single ended input, unregulated output mode	V _{DD} = V _{DDIO} = 2.7–3.6V	T = 25 °C	–	–	–	μA
	Hibernate Mode ^[22]						
	Hibernate mode current All regulators and oscillators off. SRAM retention GPIO interrupts are active Boost = OFF SIO pins in single ended input, unregulated output mode	V _{DD} = V _{DDIO} = 4.5–5.5 V	T = –40 °C	–	–	–	nA
			T = 25 °C	–	–	–	nA
			T = 85 °C	–	–	–	nA
		V _{DD} = V _{DDIO} = 2.7–3.6 V	T = –40 °C	–	–	–	nA
			T = 25 °C	–	200	–	nA
			T = 85 °C	–	–	–	nA
		V _{DD} = V _{DDIO} = 1.71–1.95 V	T = –40 °C	–	–	–	nA
T = 25 °C			–	–	–	nA	
T = 85 °C			–	–	–	nA	

Notes

22. If V_{CCD} and V_{CCA} are externally regulated, the voltage difference between V_{CCD} and V_{CCA} must be less than 50 mV.

23. Sleep timer generates periodic interrupts to wake up the CPU. This specification applies only to those times that the CPU is off.

Table 11-28. Comparator AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
T _{RESP}	Response time, high current mode ^[39]	50 mV overdrive, measured pin-to-pin	–	75	110	ns
	Response time, low current mode ^[39]	50 mV overdrive, measured pin-to-pin	–	155	200	ns
	Response time, ultra low-power mode ^[39]	50 mV overdrive, measured pin-to-pin	–	55	–	μs

11.5.6 Current Digital-to-analog Converter(IDAC)

See the IDAC component datasheet in PSoC Creator for full electrical specifications and APIs.

Unless otherwise specified, all charts and graphs show typical values.

Table 11-29. IDAC DC Specifications

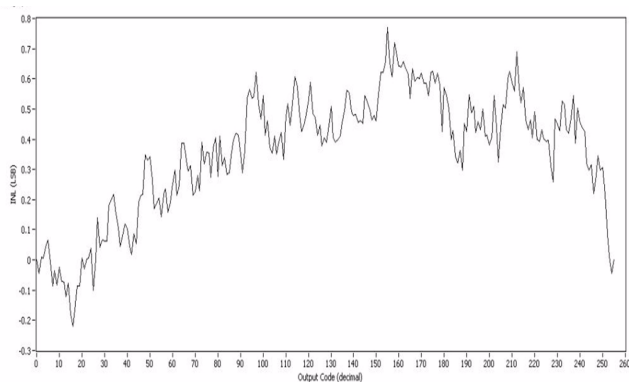
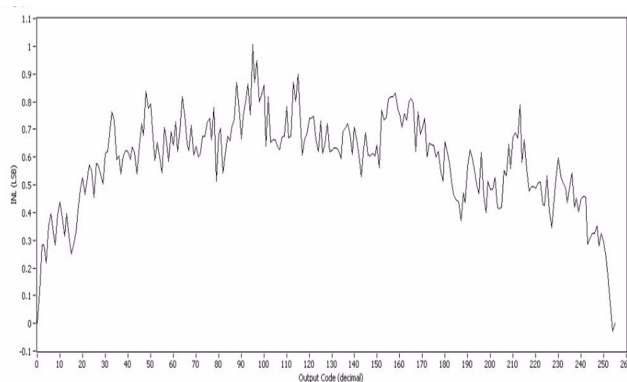
Parameter	Description	Conditions	Min	Typ	Max	Units
	Resolution		–	–	8	bits
I _{OUT}	Output current at code = 255	Range = 2.048 mA, code = 255, V _{DDA} ≥ 2.7 V, Rload = 600 Ω	–	2.048	–	mA
		Range = 2.048 mA, High mode, code = 255, V _{DDA} ≤ 2.7 V, Rload = 300 Ω	–	2.048	–	mA
		Range = 255 μA, code = 255, Rload = 600 Ω	–	255	–	μA
		Range = 31.875 μA, code = 255, Rload = 600 Ω	–	31.875	–	μA
	Monotonicity		–	–	Yes	
E _{zs}	Zero scale error		–	0	±1	LSB
E _g	Gain error	Range = 2.048 mA, 25 °C	–	–	±2.5	%
		Range = 255 μA, 25 °C	–	–	±2.5	%
		Range = 31.875 μA, 25 °C	–	–	±3.5	%
TC _{Eg}	Temperature coefficient of gain error	Range = 2.048 mA	–	–	0.04	% / °C
		Range = 255 μA	–	–	0.04	% / °C
		Range = 31.875 μA	–	–	0.05	% / °C
INL	Integral nonlinearity	Sink mode, range = 255 μA, Codes 8 – 255, Rload = 2.4 kΩ, Cload = 15 pF	–	±0.9	±1	LSB
		Source mode, range = 255 μA, Codes 8 – 255, Rload = 2.4 kΩ, Cload = 15 pF	–	±1.2	±1.5	LSB
DNL	Differential nonlinearity	Sink mode, range = 255 μA, Rload = 2.4 kΩ, Cload = 15 pF	–	±0.3	±1	LSB
		Source mode, range = 255 μA, Rload = 2.4 kΩ, Cload = 15 pF	–	±0.3	±1	LSB
V _{compliance}	Dropout voltage, source or sink mode	Voltage headroom at max current, Rload to Vdda or Rload to Vssa, Vdiff from Vdda	1	–	–	V

Note

39. Based on device characterization (Not production tested).

Table 11-29. IDAC DC Specifications (continued)

Parameter	Description	Conditions	Min	Typ	Max	Units
I_{DD}	Operating current, code = 0	Slow mode, source mode, range = 31.875 μ A	—	—	44	μ A
		Slow mode, source mode, range = 255 μ A,	—	—	33	μ A
		Slow mode, source mode, range = 2.04 mA	—	—	33	μ A
		Slow mode, sink mode, range = 31.875 μ A	—	—	36	μ A
		Slow mode, sink mode, range = 255 μ A	—	—	33	μ A
		Slow mode, sink mode, range = 2.04 mA	—	—	33	μ A
		Fast mode, source mode, range = 31.875 μ A	—	—	310	μ A
		Fast mode, source mode, range = 255 μ A	—	—	305	μ A
		Fast mode, source mode, range = 2.04 mA	—	—	305	μ A
		Fast mode, sink mode, range = 31.875 μ A	—	—	310	μ A
		Fast mode, sink mode, range = 255 μ A	—	—	300	μ A
		Fast mode, sink mode, range = 2.04 mA	—	—	300	μ A

Figure 11-25. IDAC INL vs Input Code, Range = 255 μ A, Source Mode

Figure 11-26. IDAC INL vs Input Code, Range = 255 μ A, Sink Mode


11.5.10 Programmable Gain Amplifier

The PGA is created using a SC/CT analog block; see the PGA component datasheet in PSoC Creator for full electrical specifications and APIs.

Unless otherwise specified, operating conditions are:

- Operating temperature = 25 °C for typical values
- Unless otherwise specified, all charts and graphs show typical values

Table 11-37. PGA DC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
V _{in}	Input voltage range	Power mode = minimum	V _{ssa}	–	V _{dda}	V
V _{os}	Input offset voltage	Power mode = high, gain = 1	–	–	10	mV
TCV _{os}	Input offset voltage drift with temperature	Power mode = high, gain = 1	–	–	±30	µV/°C
Ge ₁	Gain error, gain = 1		–	–	±0.15	%
Ge ₁₆	Gain error, gain = 16		–	–	±2.5	%
Ge ₅₀	Gain error, gain = 50		–	–	±5	%
V _{onl}	DC output nonlinearity	Gain = 1	–	–	±0.01	% of FSR
C _{in}	Input capacitance		–	–	7	pF
V _{oh}	Output voltage swing	Power mode = high, gain = 1, R _{load} = 100 kΩ to V _{DDA} / 2	V _{DDA} – 0.15	–	–	V
V _{ol}	Output voltage swing	Power mode = high, gain = 1, R _{load} = 100 kΩ to V _{DDA} / 2	–	–	V _{SSA} + 0.15	V
V _{src}	Output voltage under load	I _{load} = 250 µA, V _{dda} ≥ 2.7V, power mode = high	–	–	300	mV
I _{dd}	Operating current	Power mode = high	–	1.5	1.65	mA
PSRR	Power supply rejection ratio		48	–	–	dB

Figure 11-43. V_{offset} Histogram, 1000 Samples, V_{dda} = 5 V

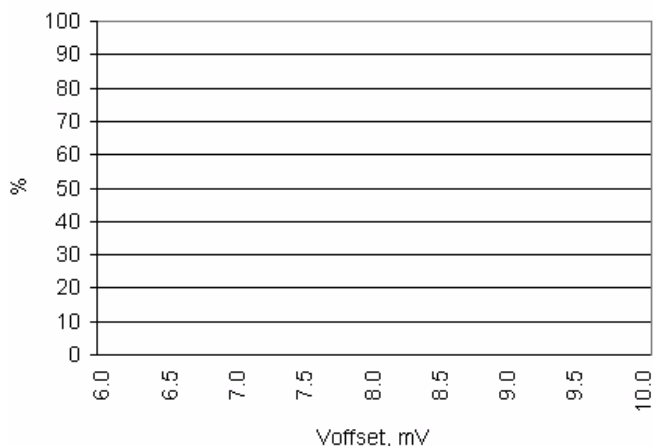


Figure 11-51. Synchronous Write Cycle Timing

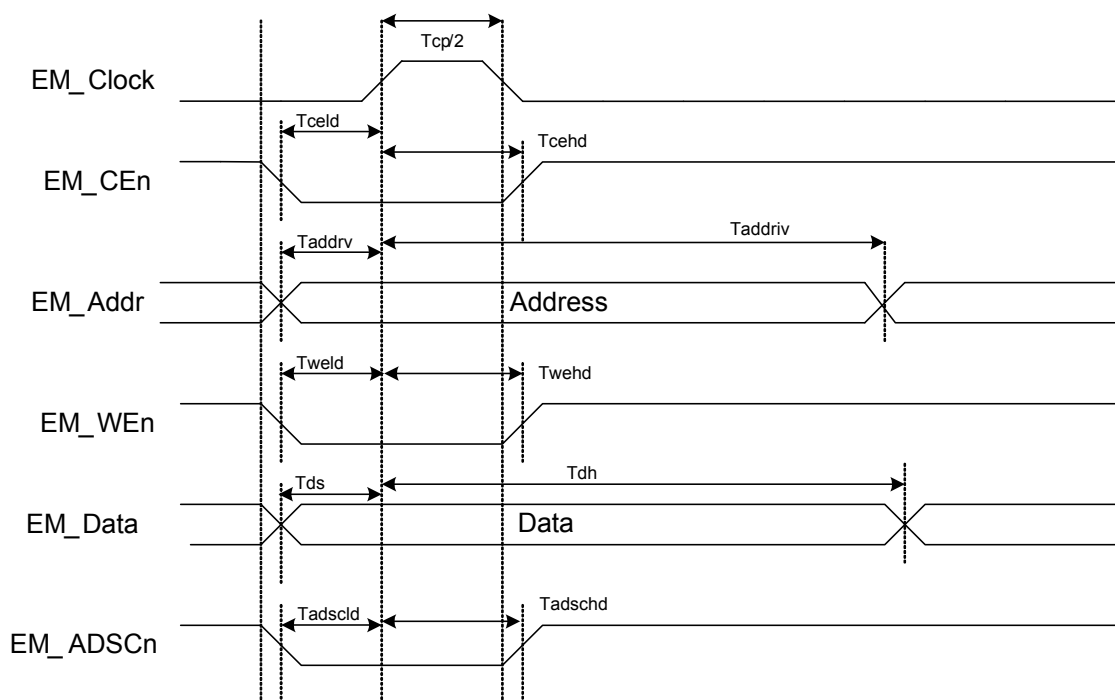


Table 11-66. Synchronous Write Cycle Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
T	EMIF clock Period ^[45]	V _{dda} ≥ 3.3 V	30.3	–	–	nS
Tcp/2	EM_Clock pulse high		T/2	–	–	nS
Tceld	EM_CEn low to EM_Clock high		5	–	–	nS
Tcehd	EM_Clock high to EM_CEn high		T/2 – 5	–	–	nS
Taddrv	EM_Addr valid to EM_Clock high		5	–	–	nS
Taddriv	EM_Clock high to EM_Addr invalid		T/2 – 5	–	–	nS
Tweld	EM_WEn low to EM_Clock high		5	–	–	nS
Twehd	EM_Clock high to EM_WEn high		T/2 – 5	–	–	nS
Tds	Data valid before EM_Clock high		5	–	–	nS
Tdh	Data invalid after EM_Clock high		T	–	–	nS
Tadscl	EM_ADSCn low to EM_Clock high		5	–	–	nS
Tadschd	EM_Clock high to EM_ADSCn high		T/2 – 5	–	–	nS

Note

45. Limited by GPIO output frequency, see [Table 11-10](#) on page 65.

11.8.3 Interrupt Controller

Table 11-71. Interrupt Controller AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
	Delay from interrupt signal input to ISR code execution from ISR code	Includes worse case completion of longest instruction DIV with 6 cycles	–	–	25	Tcy CPU

11.8.4 JTAG Interface

Table 11-72. JTAG Interface AC Specifications^[46]

Parameter	Description	Conditions	Min	Typ	Max	Units
f_TCK	TCK frequency	$3.3\text{ V} \leq V_{DD} \leq 5\text{ V}$	–	–	14 ^[47]	MHz
		$1.71\text{ V} \leq V_{DD} < 3.3\text{ V}$	–	–	7 ^[47]	MHz
T_TDI_setup	TDI setup before TCK high		$(T/10) - 5$	–	–	ns
T_TMS_setup	TMS setup before TCK high		T/4	–	–	
T_TDI_hold	TDI, TMS hold after TCK high	$T = 1/f_TCK$	T/4	–	–	
T_TDO_valid	TCK low to TDO valid	$T = 1/f_TCK$	2T/5	–	–	
T_TDO_hold	TDO hold after TCK high	$T = 1/f_TCK$	T/4	–	–	
	TCK to device outputs valid		–	–	2T/5	

11.8.5 SWD Interface

Table 11-73. SWD Interface AC Specifications^[46]

Parameter	Description	Conditions	Min	Typ	Max	Units
f_SWDCCK	SWDCLK frequency	$3.3\text{ V} \leq V_{DD} \leq 5\text{ V}$	–	–	14 ^[48]	MHz
		$1.71\text{ V} \leq V_{DD} < 3.3\text{ V}$	–	–	7 ^[48]	MHz
		$1.71\text{ V} \leq V_{DD} < 3.3\text{ V}$, SWD over USBIO pins	–	–	5.5 ^[48]	MHz
T_SWDI_setup	SWDIO input setup before SWDCCK high	$T = 1/f_SWDCCK$	T/4	–	–	
T_SWDI_hold	SWDIO input hold after SWDCCK high	$T = 1/f_SWDCCK$	T/4	–	–	
T_SWDO_valid	SWDCCK low to SWDIO output valid	$T = 1/f_SWDCCK$	2T/5	–	–	
T_SWDO_hold	SWDIO output hold after SWDCCK high	$T = 1/f_SWDCCK$	T/4	–	–	

11.8.6 SWV Interface

Table 11-74. SWV Interface AC Specifications^[46]

Parameter	Description	Conditions	Min	Typ	Max	Units
	SWV mode SWV bit rate		–	–	33	Mbit

Notes

46. Based on device characterization (Not production tested).

47. f_TCK must also be no more than 1/3 CPU clock frequency.

48. f_SWDCCK must also be no more than 1/3 CPU clock frequency.

Table 12-1. CY8C38 Family with Single Cycle 8051 (continued)

Part Number	MCU Core				Analog								Digital				I/O ^[60]				Package	JTAG ID ^[61]
	CPU Speed (MHz)	Flash (KB)	SRAM (KB)	EEPROM (KB)	LCD Segment Drive	ADC	DAC	Comparator	SC/CT Analog Blocks ^[58]	Opamps	DFB	CapSense	UDBs ^[59]	16-Bit Timer/PWM	FS USB	CAN 2.0b	Total I/O	GPIO	SIO	USBIO		
CY8C3866AXI-039	67	64	8	2	✓	20-bit Del-Sig	4	4	4	4	✓	✓	24	4	✓	–	72	62	8	2	100-pin TQFP	0×0E027069
CY8C3866LTI-030	67	64	8	2	✓	20-bit Del-Sig	4	4	4	4	✓	✓	24	4	✓	–	48	38	8	2	68-pin QFN	0×0E01E069
CY8C3866LTI-068	67	64	8	2	✓	20-bit Del-Sig	4	4	4	2	✓	✓	24	4	✓	✓	31	25	4	2	48-pin QFN	0×0E044069
CY8C3866PVI-069	67	64	8	2	✓	20-bit Del-Sig	4	4	4	2	✓	✓	24	4	✓	–	31	25	4	2	48-pin SSOP	0×0E045069
CY8C3866AXI-040	67	64	8	2	✓	20-bit Del-Sig	4	4	4	4	✓	✓	24	4	✓	✓	72	62	8	2	100-pin TQFP	0×0E028069
CY8C3866PVI-047	67	64	8	2	–	20-bit Del-Sig	4	4	4	2	✓	✓	24	4	–	✓	29	25	4	0	48-pin SSOP	0×0E02F069
CY8C3866PVI-070	67	64	8	2	✓	20-bit Del-Sig	4	4	4	2	✓	✓	24	4	–	✓	29	25	4	0	48-pin SSOP	0×0E046069
CY8C3866AXI-055	67	64	8	2	–	20-bit Del-Sig	4	4	4	4	✓	✓	24	4	–	✓	70	62	8	0	100-pin TQFP	0×0E037069
CY8C3866AXI-035	67	64	8	2	✓	20-bit Del-Sig	4	4	4	4	✓	✓	24	4	–	✓	70	62	8	0	100-pin TQFP	0×0E023069

Notes

58. Analog blocks support a variety of functionality including TIA, PGA, and mixers. See the [Example Peripherals](#) on page 35 for more information on how analog blocks can be used.

59. UDBs support a variety of functionality including SPI, LIN, UART, timer, counter, PWM, PRS, and others. Individual functions may use a fraction of a UDB or multiple UDBs. Multiple functions can share a single UDB. See the [Example Peripherals](#) on page 35 for more information on how UDBs can be used.

60. The I/O Count includes all types of digital I/O: GPIO, SIO, and the two USB I/O. See the [I/O System and Routing](#) on page 28 for details on the functionality of each of these types of I/O.

61. The JTAG ID has three major fields. The most significant nibble (left digit) is the version, followed by a 2 byte part number and a 3 nibble manufacturer ID.

13. Packaging

Table 13-1. Package Characteristics

Parameter	Description	Conditions	Min	Typ	Max	Units
T _A	Operating ambient temperature		-40	25.00	85	°C
T _J	Operating junction temperature		-40	–	100	°C
T _{ja}	Package θJA (48-pin SSOP)		–	45.16	–	°C/Watt
T _{ja}	Package θJA (48-pin QFN)		–	15.94	–	°C/Watt
T _{ja}	Package θJA (68-pin QFN)		–	11.72	–	°C/Watt
T _{ja}	Package θJA (100-pin TQFP)		–	30.52	–	°C/Watt
T _{jc}	Package θJC (48-pin SSOP)		–	27.84	–	°C/Watt
T _{jc}	Package θJC (48-pin QFN)		–	7.05	–	°C/Watt
T _{jc}	Package θJC (68-pin QFN)		–	6.32	–	°C/Watt
T _{jc}	Package θJC (100-pin TQFP)		–	9.04	–	°C/Watt
	Pb-free assemblies (20s to 40s) – Sn-Ag-Cu solder paste reflow temperature		235	–	245	°C
	Pb-free assemblies (20s to 40s) – Sn-Pb solder paste reflow temperature		205	–	220	°C

Table 13-2. Package Moisture Sensitivity Level (MSL), IPC/JEDEC J-STD-2

Package	MSL
48-pin SSOP	MSL 1
48-pin QFN	MSL 3
68-pin QFN	MSL 3
100-pin TQFP	MSL 3

Figure 13-1. 48-pin (300 mil) SSOP Package Outline

