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What is "[Embedded - Microcontrollers](#)"?

"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "[Embedded - Microcontrollers](#)"

Details

Product Status	Obsolete
Core Processor	8051
Core Size	8-Bit
Speed	67MHz
Connectivity	EBI/EMI, I ² C, LINbus, SPI, UART/USART
Peripherals	CapSense, DMA, LCD, POR, PWM, WDT
Number of I/O	25
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	4K x 8
Voltage - Supply (Vcc/Vdd)	1.71V ~ 5.5V
Data Converters	A/D 16x20b; D/A 4x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	48-BSSOP (0.295", 7.50mm Width)
Supplier Device Package	48-SSOP
Purchase URL	https://www.e-xfl.com/product-detail/infineon-technologies/cy8c3865pvi-060

For more details on the peripherals see the “[Example Peripherals](#)” section on page 35 of this datasheet. For information on UDBs, DSI, and other digital blocks, see the “[Digital Subsystem](#)” section on page 34 of this datasheet.

PSoC’s analog subsystem is the second half of its unique configurability. All analog performance is based on a highly accurate absolute voltage reference with less than 0.1-percent error over temperature and voltage. The configurable analog subsystem includes:

- Analog muxes
- Comparators
- Voltage references
- Analog-to-digital converter (ADC)
- Digital-to-analog converters (DACs)
- Digital filter block (DFB)

All GPIO pins can route analog signals into and out of the device using the internal analog bus. This allows the device to interface up to 62 discrete analog signals. The heart of the analog subsystem is a fast, accurate, configurable delta-sigma ADC with these features:

- Less than 100 μ V offset
- A gain error of 0.2 percent
- INL less than ± 2 LSB
- DNL less than ± 1 LSB
- SINAD better than 84 dB in 16-bit mode

This converter addresses a wide variety of precision analog applications, including some of the most demanding sensors. The output of the ADC can optionally feed the programmable DFB through the DMA without CPU intervention. You can configure the DFB to perform IIR and FIR digital filters and several user-defined custom functions. The DFB can implement filters with up to 64 taps. It can perform a 48-bit multiply-accumulate (MAC) operation in one clock cycle.

Four high-speed voltage or current DACs support 8-bit output signals at an update rate of up to 8 Msps. They can be routed out of any GPIO pin. You can create higher resolution voltage PWM DAC outputs using the UDB array. This can be used to create a pulse width modulated (PWM) DAC of up to 10 bits, at up to 48 kHz. The digital DACs in each UDB support PWM, PRS, or delta-sigma algorithms with programmable widths. In addition to the ADC, DACs, and DFB, the analog subsystem provides multiple:

- Uncommitted opamps
- Configurable switched capacitor/continuous time (SC/CT) blocks. These support:
 - Transimpedance amplifiers
 - Programmable gain amplifiers
 - Mixers
 - Other similar analog components

See the “[Analog Subsystem](#)” section on page 46 of this datasheet for more details.

PSoC’s 8051 CPU subsystem is built around a single cycle pipelined 8051 8-bit processor running at up to 67 MHz. The CPU subsystem includes a programmable nested vector interrupt controller, DMA controller, and RAM. PSoC’s nested vector interrupt controller provides low latency by allowing the CPU to vector directly to the first address of the interrupt service routine, bypassing the jump instruction required by other architectures. The DMA controller enables peripherals to exchange data without CPU involvement. This allows the CPU to run slower (saving power) or use those CPU cycles to improve the performance of firmware algorithms. The single cycle 8051 CPU runs ten times faster than a standard 8051 processor. The processor speed itself is configurable, allowing you to tune active power consumption for specific applications.

PSoC’s nonvolatile subsystem consists of flash, byte-writeable EEPROM, and nonvolatile configuration options. It provides up to 64 KB of on-chip flash. The CPU can reprogram individual blocks of flash, enabling bootloaders. You can enable an error correcting code (ECC) for high reliability applications. A powerful and flexible protection model secures the user’s sensitive information, allowing selective memory block locking for read and write protection. Up to 2 KB of byte-writeable EEPROM is available on-chip to store application data. Additionally, selected configuration options such as boot speed and pin drive mode are stored in nonvolatile memory. This allows settings to activate immediately after POR.

The three types of PSoC I/O are extremely flexible. All I/Os have many drive modes that are set at POR. PSoC also provides up to four I/O voltage domains through the Vddio pins. Every GPIO has analog I/O, LCD drive^[4], CapSense^[5], flexible interrupt generation, slew rate control, and digital I/O capability. The SIOs on PSoC allow V_{OH} to be set independently of Vddio when used as outputs. When SIOs are in input mode they are high impedance. This is true even when the device is not powered or when the pin voltage goes above the supply voltage. This makes the SIO ideally suited for use on an I²C bus where the PSoC may not be powered when other devices on the bus are. The SIO pins also have high current sink capability for applications such as LED drives. The programmable input threshold feature of the SIO can be used to make the SIO function as a general purpose analog comparator. For devices with Full-Speed USB the USB physical interface is also provided (USBIO). When not using USB these pins may also be used for limited digital functionality and device programming. All of the features of the PSoC I/Os are covered in detail in the “[I/O System and Routing](#)” section on page 28 of this datasheet.

The PSoC device incorporates flexible internal clock generators, designed for high stability and factory trimmed for high accuracy. The internal main oscillator (IMO) is the master clock base for the system, and has 1-percent accuracy at 3 MHz. The IMO can be configured to run from 3 MHz up to 62 MHz. Multiple clock derivatives can be generated from the main clock frequency to meet application needs. The device provides a PLL to generate system clock frequencies up to 67 MHz from the IMO, external crystal, or external reference clock.

Notes

4. This feature on select devices only. See [Ordering Information](#) on page 105 for details.
5. GPIOs with opamp outputs are not recommended for use with CapSense.

Figure 2-4. 100-pin TQFP Part Pinout

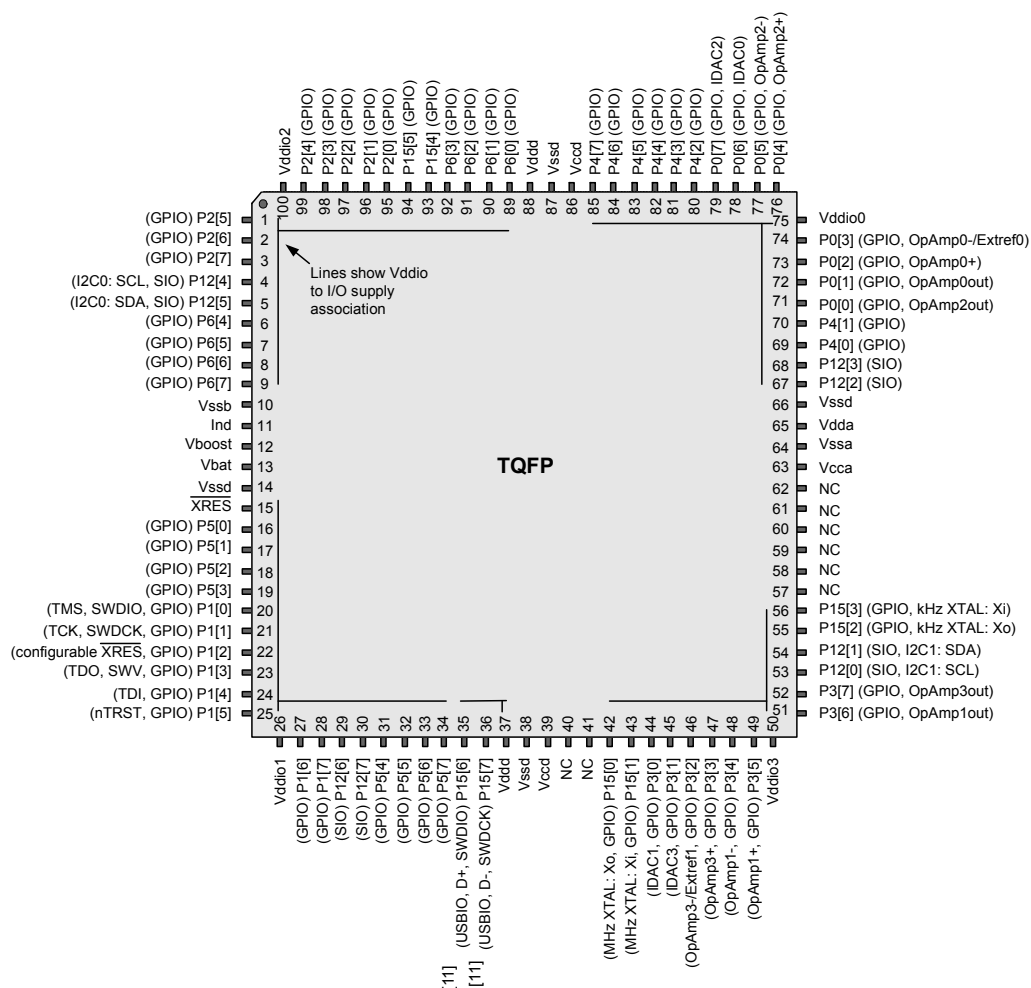


Figure 2-5 and Figure 2-6 show an example schematic and an example PCB layout, for the 100-pin TQFP part, for optimal analog performance on a two-layer board.

- The two pins labeled Vddd must be connected together.
- The two pins labeled Vccd must be connected together, with capacitance added, as shown in Figure 2-5 and Power System on page 24. The trace between the two Vccd pins should be as short as possible.
- The two pins labeled Vssd must be connected together.

Note

11. Pins are Do Not Use (DNU) on devices without USB. The pin must be left floating.

4.4.4.7 Nested DMA

One TD may modify another TD, as the TD configuration space is memory mapped similar to any other peripheral. For example, a first TD loads a second TD's configuration and then calls the second TD. The second TD moves data as required by the application. When complete, the second TD calls the first TD, which again updates the second TD's configuration. This process repeats as often as necessary.

4.5 Interrupt Controller

The interrupt controller provides a mechanism for hardware resources to change program execution to a new address, independent of the current task being executed by the main code. The interrupt controller provides enhanced features not found on original 8051 interrupt controllers:

- Thirty-two interrupt vectors
- Jumps directly to ISR anywhere in code space with dynamic vector addresses
- Multiple sources for each vector
- Flexible interrupt to vector matching
- Each interrupt vector is independently enabled or disabled
- Each interrupt can be dynamically assigned one of eight priorities
- Eight level nestable interrupts
- Multiple I/O interrupt vectors
- Software can send interrupts
- Software can clear pending interrupts

When an interrupt is pending, the current instruction is completed and the program counter is pushed onto the stack. Code execution then jumps to the program address provided by the vector. After the ISR is completed, a RETI instruction is executed and returns execution to the instruction following the previously interrupted instruction. To do this the RETI instruction pops the program counter from the stack.

If the same priority level is assigned to two or more interrupts, the interrupt with the lower vector number is executed first. Each interrupt vector may choose from three interrupt sources: Fixed Function, DMA, and UDB. The fixed function interrupts are direct connections to the most common interrupt sources and provide the lowest resource cost connection. The DMA interrupt sources provide direct connections to the two DMA interrupt sources provided per DMA channel. The third interrupt source for vectors is from the UDB digital routing array. This allows any digital signal available to the UDB array to be used as an interrupt source.

Fixed function interrupts and all interrupt sources may be routed to any interrupt vector using the UDB interrupt source connections.

Table 4-8. Interrupt Vector Table

#	Fixed Function	DMA	UDB
0	LVD	phub_termout0[0]	udb_intr[0]
1	ECC	phub_termout0[1]	udb_intr[1]
2	Reserved	phub_termout0[2]	udb_intr[2]
3	Sleep (Pwr Mgr)	phub_termout0[3]	udb_intr[3]
4	PICU[0]	phub_termout0[4]	udb_intr[4]
5	PICU[1]	phub_termout0[5]	udb_intr[5]
6	PICU[2]	phub_termout0[6]	udb_intr[6]
7	PICU[3]	phub_termout0[7]	udb_intr[7]
8	PICU[4]	phub_termout0[8]	udb_intr[8]
9	PICU[5]	phub_termout0[9]	udb_intr[9]
10	PICU[6]	phub_termout0[10]	udb_intr[10]
11	PICU[12]	phub_termout0[11]	udb_intr[11]
12	PICU[15]	phub_termout0[12]	udb_intr[12]
13	Comparators Combined	phub_termout0[13]	udb_intr[13]
14	Switched Caps Combined	phub_termout0[14]	udb_intr[14]
15	I ² C	phub_termout0[15]	udb_intr[15]
16	CAN	phub_termout1[0]	udb_intr[16]
17	Timer/Counter0	phub_termout1[1]	udb_intr[17]
18	Timer/Counter1	phub_termout1[2]	udb_intr[18]
19	Timer/Counter2	phub_termout1[3]	udb_intr[19]
20	Timer/Counter3	phub_termout1[4]	udb_intr[20]
21	USB SOF Int	phub_termout1[5]	udb_intr[21]
22	USB Arb Int	phub_termout1[6]	udb_intr[22]
23	USB Bus Int	phub_termout1[7]	udb_intr[23]
24	USB Endpoint[0]	phub_termout1[8]	udb_intr[24]
25	USB Endpoint Data	phub_termout1[9]	udb_intr[25]
26	Reserved	phub_termout1[10]	udb_intr[26]
27	Reserved	phub_termout1[11]	udb_intr[27]
28	DFB Int	phub_termout1[12]	udb_intr[28]
29	Decimator Int	phub_termout1[13]	udb_intr[29]
30	PHUB Error Int	phub_termout1[14]	udb_intr[30]

6.2.1 Power Modes

PSoC 3 devices have four different power modes, as shown in [Table 6-2](#) and [Table 6-3](#). The power modes allow a design to easily provide required functionality and processing power while simultaneously minimizing power consumption and maximizing battery life in low-power and portable devices.

PSoC 3 power modes, in order of decreasing power consumption are:

- Active
- Alternate Active
- Sleep
- Hibernate

Active is the main processing mode. Its functionality is configurable. Each power controllable subsystem is enabled or disabled by using separate power configuration template registers. In alternate active mode, fewer subsystems are enabled, reducing power. In sleep mode most resources are disabled regardless of the template settings. Sleep mode is optimized to provide timed sleep intervals and Real Time Clock functionality. The lowest power mode is hibernate, which retains register and SRAM state, but no clocks, and allows wakeup only from I/O pins. [Figure 6-5](#) on page 26 illustrates the allowable transitions between power modes

Table 6-2. Power Modes

Power Modes	Description	Entry Condition	Wakeup Source	Active Clocks	Regulator
Active	Primary mode of operation, all peripherals available (programmable)	Wakeup, reset, manual register entry	Any interrupt	Any (programmable)	All regulators available. Digital and analog regulators can be disabled if external regulation used.
Alternate Active	Similar to Active mode, and is typically configured to have fewer peripherals active to reduce power. One possible configuration is to use the UDBs for processing, with the CPU turned off	Manual register entry	Any interrupt	Any (programmable)	All regulators available. Digital and analog regulators can be disabled if external regulation used.
Sleep	All subsystems automatically disabled	Manual register entry	Comparator, PICU, I ² C, RTC, CTW, LVD	ILO/kHzECO	Both digital and analog regulators buzzed. Digital and analog regulators can be disabled if external regulation used.
Hibernate	All subsystems automatically disabled Lowest power consuming mode with all peripherals and internal regulators disabled, except hibernate regulator is enabled Configuration and memory contents retained	Manual register entry	PICU		Only hibernate regulator active.

Table 6-3. Power Modes Wakeup Time and Power Consumption

Sleep Modes	Wakeup Time	Current (typ)	Code Execution	Digital Resources	Analog Resources	Clock Sources Available	Wakeup Sources	Reset Sources
Active	–	1.2 mA ^[13]	Yes	All	All	All	–	All
Alternate Active	–	–	User defined	All	All	All	–	All
Sleep	<15 µs	1 µA	No	I ² C	Comparator	ILO/kHzECO	Comparator, PICU, I ² C, RTC, CTW, LVD	XRES, LVD, WDR
Hibernate	<100 µs	200 nA	No	None	None	None	PICU	XRES

Note

13. Bus clock off. Execute from CPU instruction buffer at 6 MHz. See [Table 11-2](#) on page 60.

voltage. In standby mode, most boost functions are disabled, thus reducing power consumption of the boost circuit. The converter can be configured to provide low-power, low-current regulation in the standby mode. The external 32-kHz crystal can be used to generate inductor boost pulses on the rising and falling edge of the clock when the output voltage is less than the programmed value. This is called automatic thump mode (ATM).

The boost typically draws 200 μ A in active mode and 12 μ A in standby mode. The boost operating modes must be used in conjunction with chip power modes to minimize the total chip power consumption. Table 6-4 lists the boost power modes available in different chip power modes.

Table 6-4. Chip and Boost Power Modes Compatibility

Chip Power Modes	Boost Power Modes
Chip – Active mode	Boost can be operated in either active or standby mode.
Chip – Sleep mode	Boost can be operated in either active or standby mode. However, it is recommended to operate boost in standby mode for low-power consumption
Chip – Hibernate mode	Boost can only be operated in active mode. However, it is recommended not to use boost in chip hibernate mode due to high current consumption in boost active mode

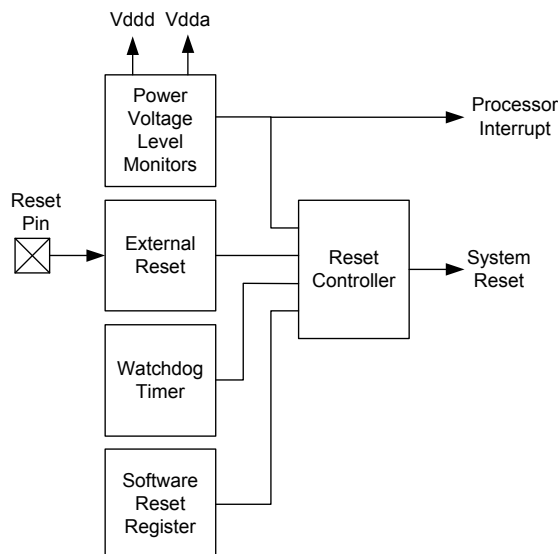
If the boost converter is not used in a given application, tie the Vbat, Vssb, and Vboost pins to ground and leave the Ind pin unconnected.

6.3 Reset

CY8C38 has multiple internal and external reset sources available. The reset sources are:

- **Power source monitoring** – The analog and digital power voltages, Vdda, Vddd, Vcca, and Vccd are monitored in several different modes during power up, active mode, and sleep mode (buzzing). If any of the voltages goes outside predetermined ranges then a reset is generated. The monitors are programmable to generate an interrupt to the processor under certain conditions before reaching the reset thresholds.
- **External** – The device can be reset from an external source by pulling the reset pin (XRES) low. The XRES pin includes an internal pull-up to Vddio1. Vddd, Vdda, and Vddio1 must all have voltage applied before the part comes out of reset.
- **Watchdog timer** – A watchdog timer monitors the execution of instructions by the processor. If the watchdog timer is not reset by firmware within a certain period of time, the watchdog timer generates a reset.
- **Software** – The device can be reset under program control.

Figure 6-7. Resets



The term **device reset** indicates that the processor as well as analog and digital peripherals and registers are reset.

A reset status register holds the source of the most recent reset or power voltage monitoring interrupt. The program may examine this register to detect and report exception conditions. This register is cleared after a power-on reset.

6.3.1 Reset Sources

6.3.1.1 Power Voltage Level Monitors

■ IPOR – Initial power-on reset

At initial power on, IPOR monitors the power voltages V_{DDP} and V_{DDA}, both directly at the pins and at the outputs of the corresponding internal regulators. The trip level is not precise. It is set to approximately 1 volt, which is below the lowest specified operating voltage but high enough for the internal circuits to be reset and to hold their reset state. The monitor generates a reset pulse that is at least 100 ns wide. It may be much wider if one or more of the voltages ramps up slowly.

To save power the IPOR circuit is disabled when the internal digital supply is stable. Voltage supervision is then handed off to the precise low voltage reset (PRES) circuit. When the voltage is high enough for PRES to release, the IMO starts.

■ PRES – Precise low voltage reset

This circuit monitors the outputs of the analog and digital internal regulators after power up. The regulator outputs are compared to a precise reference voltage. The response to a PRES trip is identical to an IPOR reset.

In normal operating mode, the program cannot disable the digital PRES circuit. The analog regulator can be disabled, which also disables the analog portion of the PRES. The PRES circuit is disabled automatically during sleep and hibernate modes, with one exception: During sleep mode the regulators are periodically activated (buzzed) to provide supervisory services and to reduce wakeup time. At these times the PRES circuit is also buzzed to allow periodic voltage monitoring.

Figure 6-9. SIO Input/Output Block Diagram

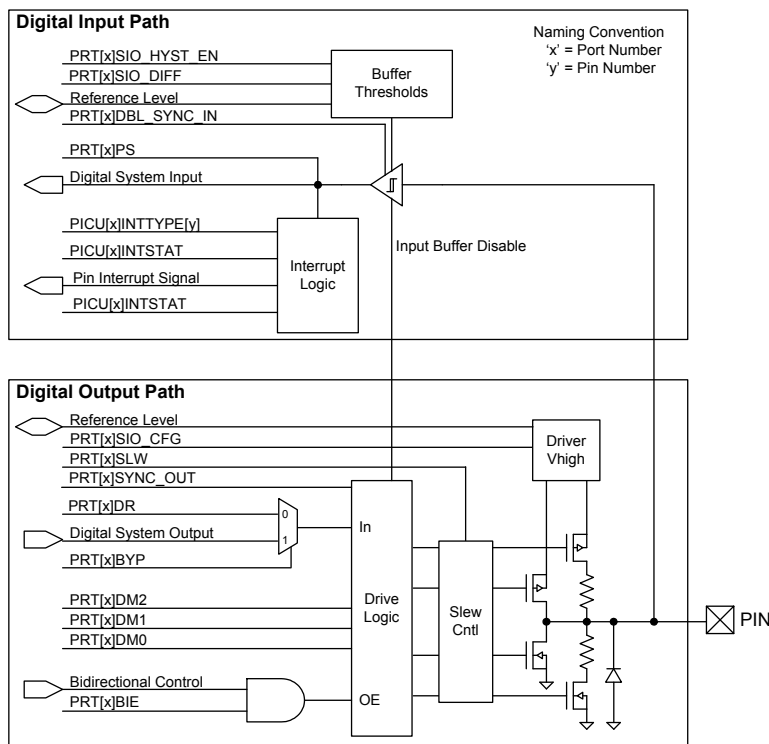
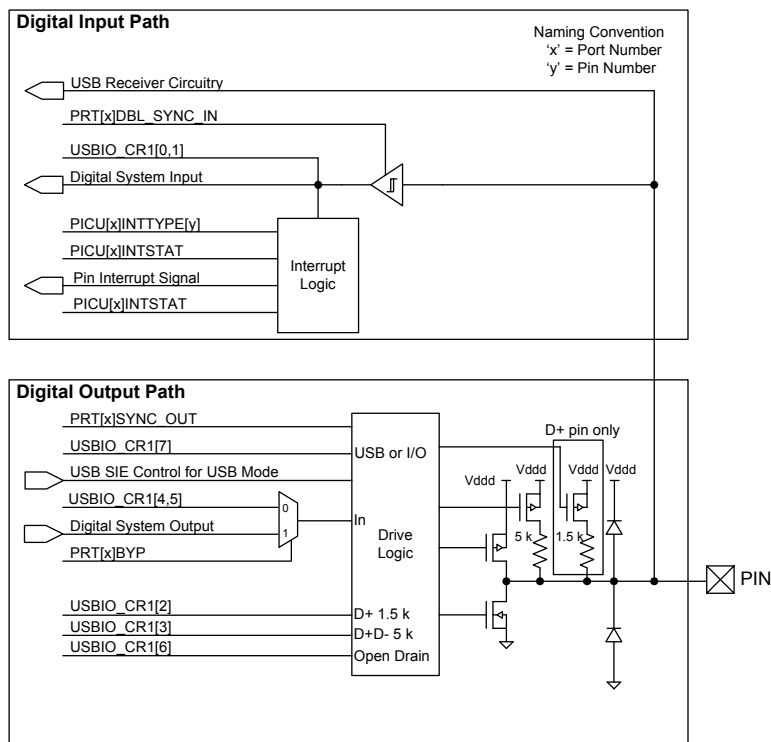


Figure 6-10. USBIO Block Diagram



6.4.15 Over Voltage Tolerance

All I/O pins provide an over voltage tolerance feature at any operating V_{DD} .

- There are no current limitations for the SIO pins as they present a high impedance load to the external circuit where $V_{DDIO} \leq V_{IN} \leq 5.5\text{ V}$.
- The GPIO pins must be limited to 100 μA using a current limiting resistor. GPIO pins clamp the pin voltage to approximately one diode above the V_{DDIO} supply where $V_{DDIO} \leq V_{IN} \leq V_{DDA}$.
- In case of a GPIO pin configured for analog input/output, the analog voltage on the pin must not exceed the V_{DDIO} supply voltage to which the GPIO belongs.

A common application for this feature is connection to a bus such as I^2C where different devices are running from different supply voltages. In the I^2C case, the PSoC chip is configured into the Open Drain, Drives Low mode for the SIO pin. This allows an external pull-up to pull the I^2C bus voltage above the PSoC pin supply. For example, the PSoC chip could operate at 1.8 V, and an external device could run from 5 V. Note that the SIO pin's V_{IH} and V_{IL} levels are determined by the associated V_{DDIO} supply pin. The I/O pin must be configured into a high impedance drive mode, open drain low drive mode, or pull-down drive mode, for over voltage tolerance to work properly. Absolute maximum ratings for the device must be observed for all I/O pins.

6.4.16 Reset Configuration

While reset is active all I/Os are reset to and held in the High Impedance Analog state. After reset is released, the state can be reprogrammed on a port-by-port basis to pull-down or pull-up. To ensure correct reset operation, the port reset configuration data is stored in special nonvolatile registers. The stored reset data is automatically transferred to the port reset configuration registers at reset release.

6.4.17 Low-Power Functionality

In all low-power modes the I/O pins retain their state until the part is awakened and changed or reset. To awaken the part, use a pin interrupt, because the port interrupt logic continues to function in all low-power modes.

6.4.18 Special Pin Functionality

Some pins on the device include additional special functionality in addition to their GPIO or SIO functionality. The specific special function pins are listed in [Pinouts](#) on page 5. The special features are:

- Digital
 - 4- to 33-MHz crystal oscillator
 - 32.768-kHz crystal oscillator
 - Wake from sleep on I^2C address match. Any pin can be used for I^2C if wake from sleep is not required.
 - JTAG interface pins
 - SWD interface pins
 - SWV interface pins
 - External reset
- Analog
 - Opamp inputs and outputs
 - High current IDAC outputs
 - External reference inputs

6.4.19 JTAG Boundary Scan

The device supports standard JTAG boundary scan chains on all I/O pins for board level test.

7. Digital Subsystem

The digital programmable system creates application specific combinations of both standard and advanced digital peripherals and custom logic functions. These peripherals and logic are then interconnected to each other and to any pin on the device, providing a high level of design flexibility and IP security.

The features of the digital programmable system are outlined here to provide an overview of capabilities and architecture. You do not need to interact directly with the programmable digital system at the hardware and register level. PSoC Creator provides a high level schematic capture graphical interface to automatically place and route resources similar to PLDs.

The main components of the digital programmable system are:

- UDB – These form the core functionality of the digital programmable system. UDBs are a collection of uncommitted logic (PLD) and structural logic (Datapath) optimized to create all common embedded peripherals and customized functionality that are application or design specific.
- Universal digital block array – UDB blocks are arrayed within a matrix of programmable interconnect. The UDB array structure is homogeneous and allows for flexible mapping of digital functions onto the array. The array supports extensive and flexible routing interconnects between UDBs and the Digital System Interconnect.
- Digital system interconnect (DSI) – Digital signals from UDBs, fixed function peripherals, I/O pins, interrupts, DMA, and other system core signals are attached to the digital system interconnect to implement full featured device connectivity. The DSI allows any digital function to any pin or other feature routability when used with the universal digital block array.

Figure 7-1. CY8C38 Digital Programmable Architecture

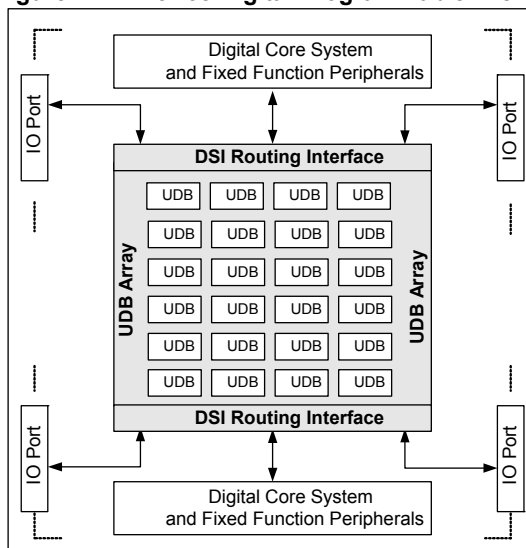
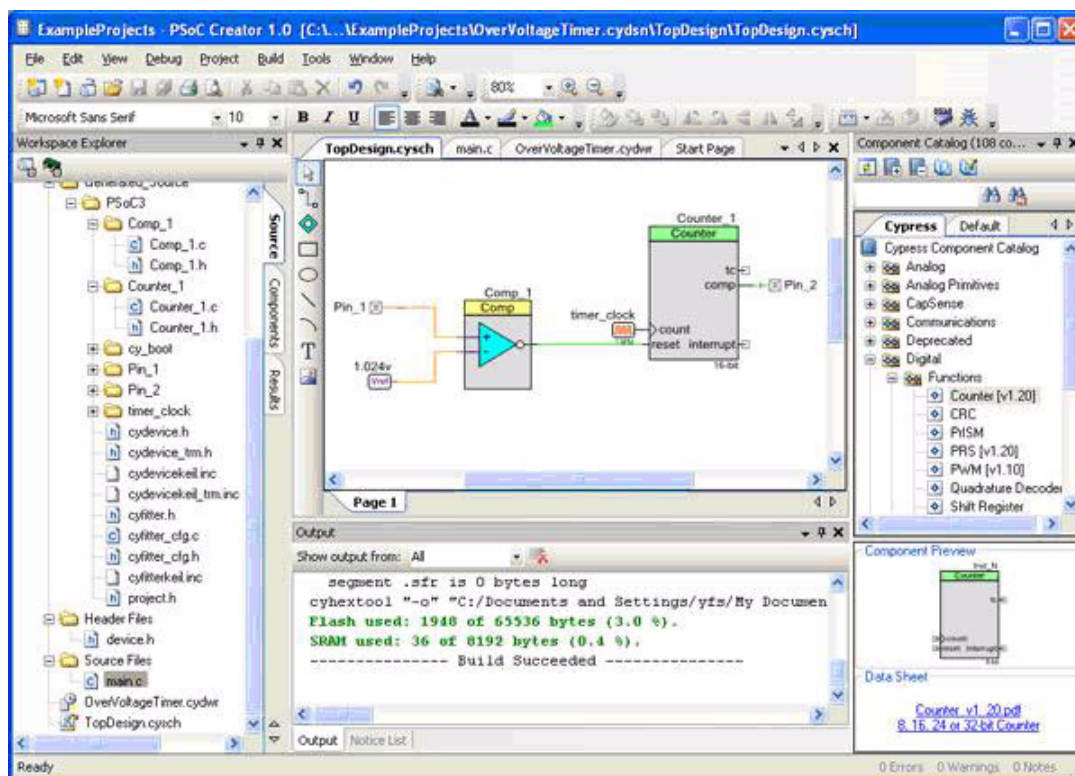
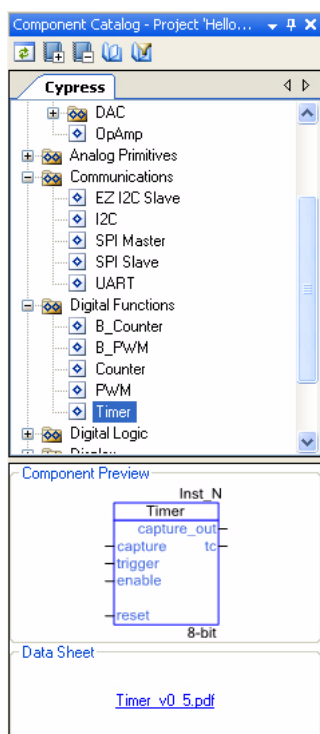


Figure 7-2. PSoC Creator Framework



7.1.4.2 Component Catalog

Figure 7-3. Component Catalog



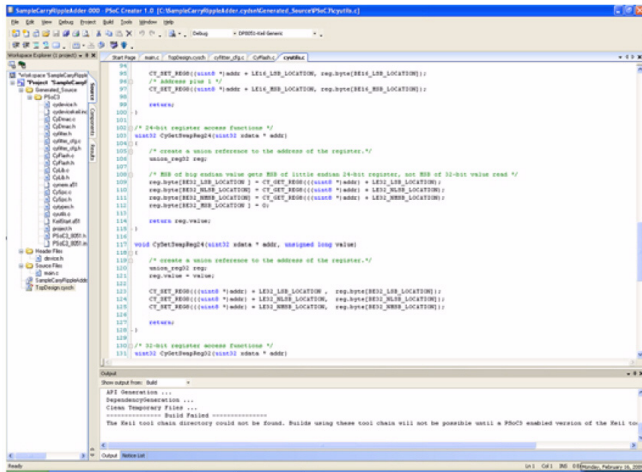
The component catalog is a repository of reusable design elements that select device functionality and customize your PSoC device. It is populated with an impressive selection of content; from simple primitives such as logic gates and device registers, through the digital timers, counters and PWMs, plus analog components such as ADC, DACs, and filters, and communication protocols, such as I²C, USB, and CAN. See [Example Peripherals](#) on page 35 for more details about available peripherals. All content is fully characterized and carefully documented in datasheets with code examples, AC/DC specifications, and user code ready APIs.

7.1.4.3 Design Reuse

The symbol editor gives you the ability to develop reusable components that can significantly reduce future design time. Just draw a symbol and associate that symbol with your proven design. PSoC Creator allows for the placement of the new symbol anywhere in the component catalog along with the content provided by Cypress. You can then reuse your content as many times as you want, and in any number of projects, without ever having to revisit the details of the implementation.

7.1.4.4 Software Development

Figure 7-4. Code Editor

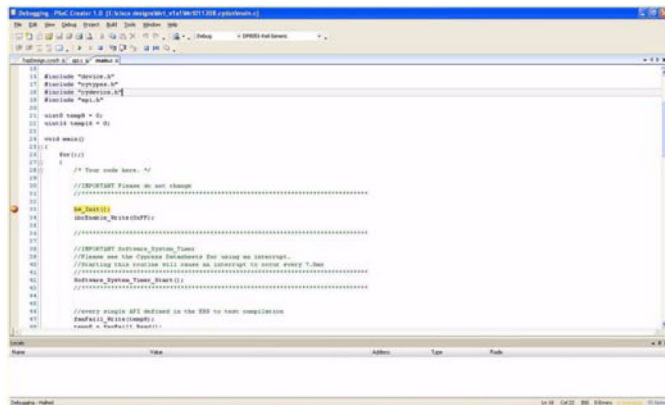


Anchoring the tool is a modern, highly customizable user interface. It includes project management and integrated editors for C and assembler source code, as well the design entry tools.

Project build control leverages compiler technology from top commercial vendors such as ARM® Limited, Keil™, and CodeSourcery (GNU). Free versions of Keil C51 and GNU C Compiler (GCC) for ARM, with no restrictions on code size or end product distribution, are included with the tool distribution. Upgrading to more optimizing compilers is a snap with support for the professional Keil C51 product and ARM RealView™ compiler.

7.1.4.5 Nonintrusive Debugging

Figure 7-5. PSoC Creator Debugger



With JTAG (4-wire) and SWD (2-wire) debug connectivity available on all devices, the PSoC Creator debugger offers full control over the target device with minimum intrusion. Breakpoints and code execution commands are all readily available from toolbar buttons and an impressive lineup of windows—register, locals, watch, call stack, memory and peripherals—make for an unparalleled level of visibility into the system.

PSoC Creator contains all the tools necessary to complete a design, and then to maintain and extend that design for years to come. All steps of the design flow are carefully integrated and optimized for ease-of-use and to maximize productivity.

7.2 Universal Digital Block

The UDB represents an evolutionary step to the next generation of PSoC embedded digital peripheral functionality. The architecture in first generation PSoC digital blocks provides coarse programmability in which a few fixed functions with a small number of options are available. The new UDB architecture is the optimal balance between configuration granularity and efficient implementation. A cornerstone of this approach is to provide the ability to customize the devices digital operation to match application requirements.

To achieve this, UDBs consist of a combination of uncommitted logic (PLD), structured logic (Datapath), and a flexible routing scheme to provide interconnect between these elements, I/O connections, and other peripherals. UDB functionality ranges from simple self contained functions that are implemented in one UDB, or even a portion of a UDB (unused resources are available for other functions), to more complex functions that require multiple UDBs. Examples of basic functions are timers, counters, CRC generators, PWMs, dead band generators, and communications functions, such as UARTs, SPI, and I²C. Also, the PLD blocks and connectivity provide full featured general purpose programmable logic within the limits of the available resources.

Figure 7-6. UDB Block Diagram

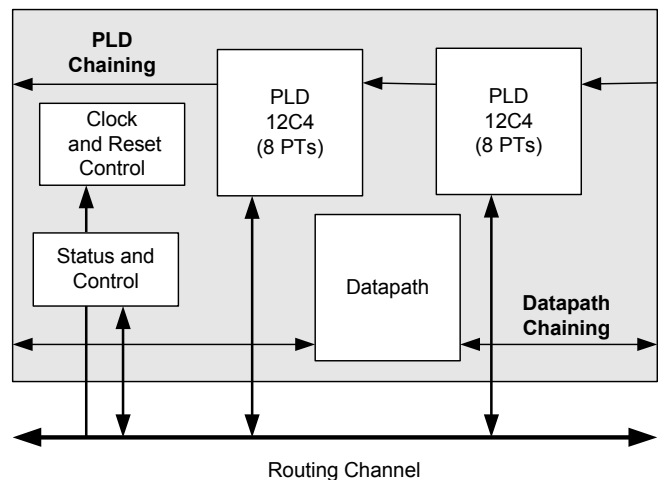
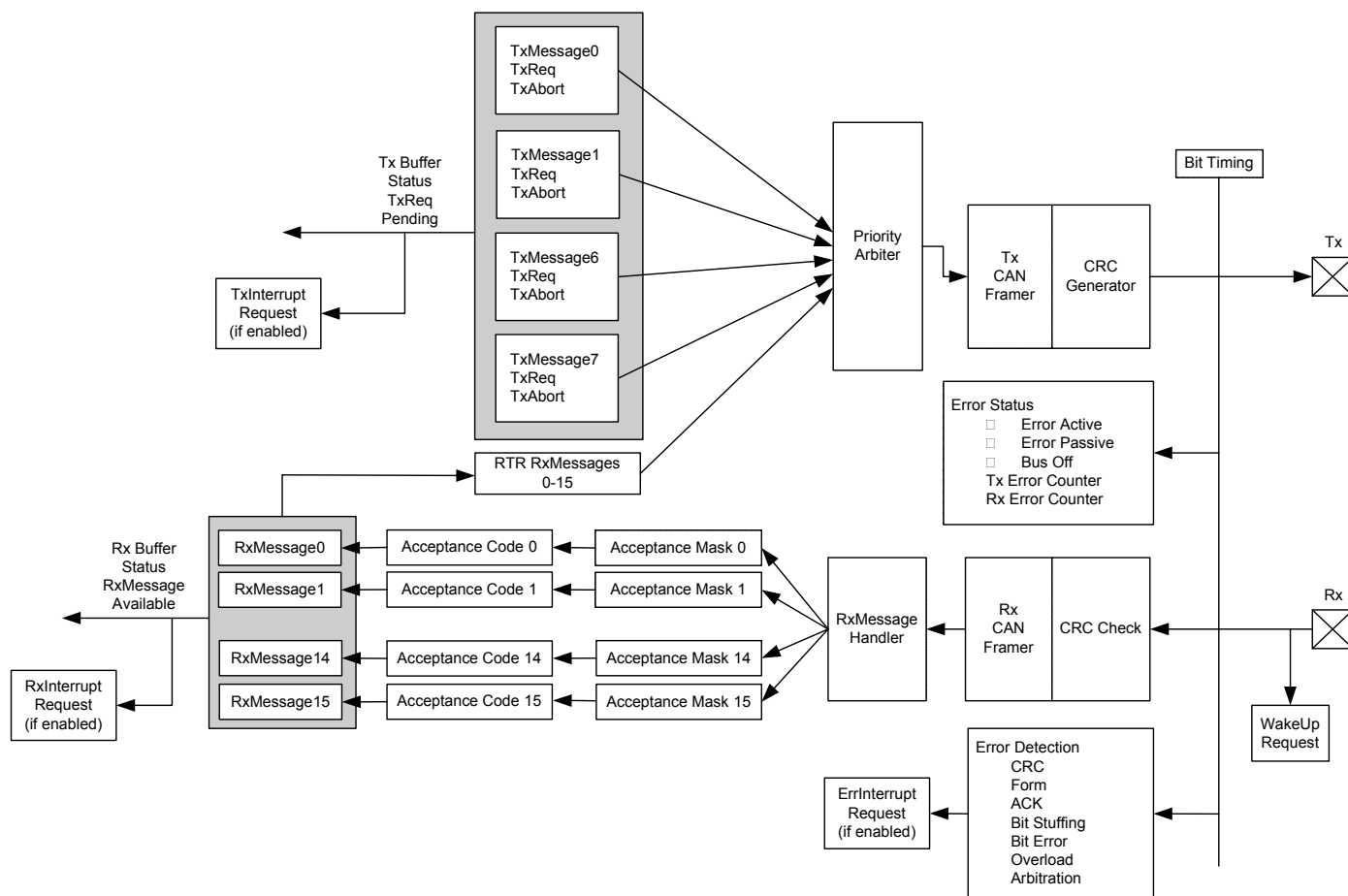


Figure 7-19. CAN Controller Block Diagram



7.6 USB

PSoC includes a dedicated Full-Speed (12 Mbps) USB 2.0 transceiver supporting all four USB transfer types: control, interrupt, bulk, and isochronous. PSoC Creator provides full configuration support. USB interfaces to hosts through two dedicated USBIO pins, which are detailed in the “[I/O System and Routing](#)” section on page 28.

USB includes the following features:

- Eight unidirectional data endpoints
- One bidirectional control endpoint 0 (EP0)
- Shared 512-byte buffer for the eight data endpoints
- Dedicated 8-byte buffer for EP0
- Three memory modes
 - Manual memory management with no DMA access
 - Manual memory management with manual DMA access
 - Automatic memory management with automatic DMA access
- Internal 3.3-V regulator for transceiver

- Internal 48-MHz main oscillator mode that auto locks to USB bus clock, requiring no external crystal for USB (USB equipped parts only)
- Interrupts on bus and each endpoint event, with device wakeup
- USB reset, suspend, and resume operations
- Bus-powered and self-powered modes

Figure 7-20. USB

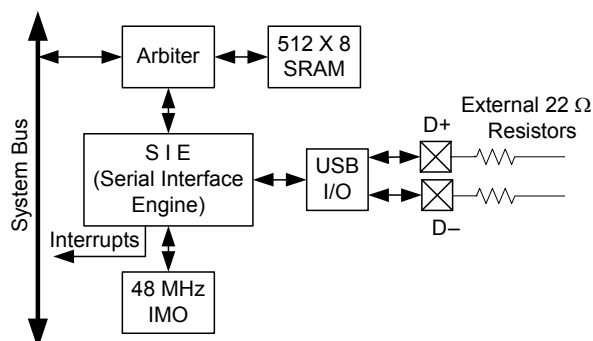
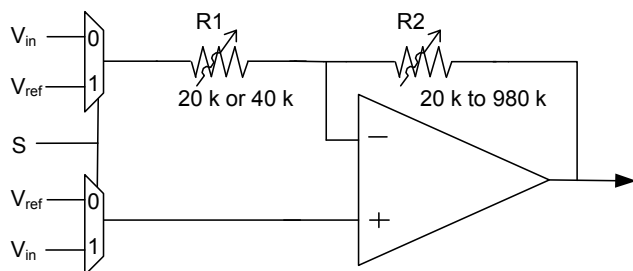


Table 8-3. Bandwidth

Gain	Bandwidth
1	6.0 MHz
24	340 kHz
48	220 kHz
50	215 kHz

Figure 8-9. PGA Resistor Settings



The PGA is used in applications where the input signal may not be large enough to achieve the desired resolution in the ADC, or dynamic range of another SC/CT block such as a mixer. The gain is adjustable at runtime, including changing the gain of the PGA prior to each ADC sample.

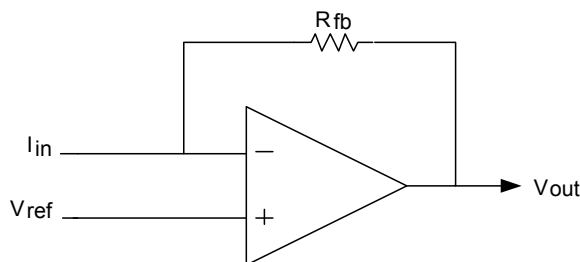
8.5.4 TIA

The Transimpedance Amplifier (TIA) converts an internal or external current to an output voltage. The TIA uses an internal feedback resistor in a continuous time configuration to convert input current to output voltage. For an input current I_{in} , the output voltage is $I_{in} \times R_{fb} + V_{REF}$, where V_{REF} is the value placed on the non inverting input. The feedback resistor R_{fb} is programmable between 20 K Ω and 1 M Ω through a configuration register. [Table 8-4](#) shows the possible values of R_{fb} and associated configuration settings.

Table 8-4. Feedback Resistor Settings

Configuration Word	Nominal R_{fb} (K Ω)
000b	20
001b	30
010b	40
011b	60
100b	120
101b	250
110b	500
111b	1000

Figure 8-10. Continuous Time TIA Schematic



The TIA configuration is used for applications where an external sensor's output is current as a function of some type of stimulus such as temperature, light, magnetic flux etc. In a common application, the voltage DAC output can be connected to the V_{REF} TIA input to allow calibration of the external sensor bias current by adjusting the voltage DAC output voltage.

8.6 LCD Direct Drive

The PSoC LCD driver system is a highly configurable peripheral designed to allow PSoC to directly drive a broad range of LCD glass. All voltages are generated on chip, eliminating the need for external components. With a high multiplex ratio of up to 1/16, the CY8C38 family LCD driver system can drive a maximum of 736 segments. The PSoC LCD driver module was also designed with the conservative power budget of portable devices in mind, enabling different LCD drive modes and power down modes to conserve power.

PSoC Creator provides an LCD segment drive component. The component wizard provides easy and flexible configuration of LCD resources. You can specify pins for segments and commons along with other options. The software configures the device to meet the required specifications. This is possible because of the programmability inherent to PSoC devices.

Key features of the PSoC LCD segment system are:

- LCD panel direct driving
- Type A (standard) and Type B (low-power) waveform support
- Wide operating voltage range support (2 V to 5 V) for LCD panels
- Static, 1/2, 1/3, 1/4, 1/5 bias voltage levels
- Internal bias voltage generation through internal resistor ladder
- Up to 62 total common and segment outputs
- Up to 1/16 multiplex for a maximum of 16 backplane/common outputs
- Up to 62 front plane/segment outputs for direct drive
- Drives up to 736 total segments (16 backplane \times 46 front plane)
- Up to 64 levels of software controlled contrast
- Ability to move display data from memory buffer to LCD driver through DMA (without CPU intervention)
- Adjustable LCD refresh rate from 10 Hz to 150 Hz
- Ability to invert LCD display for negative image
- Three LCD driver drive modes, allowing power optimization

8.11.1 Down Mixer

The SC/CT block can be used as a mixer to down convert an input signal. This circuit is a high bandwidth passive sample network that can sample input signals up to 14 MHz. This sampled value is then held using the opamp with a maximum clock rate of 4 MHz. The output frequency is at the difference between the input frequency and the highest integer multiple of the Local Oscillator that is less than the input.

8.11.2 First Order Modulator – SC Mode

A first order modulator is constructed by placing the SC/CT block in an integrator mode and using a comparator to provide a 1-bit feedback to the input. Depending on this bit, a reference voltage is either subtracted or added to the input signal. The block output is the output of the comparator and not the integrator in the modulator case. The signal is downshifted and buffered and then processed by a decimator to make a delta-sigma converter or a counter to make an incremental converter. The accuracy of the sampled data from the first-order modulator is determined from several factors.

The main application for this modulator is for a low-frequency ADC with high accuracy. Applications include strain gauges, thermocouples, precision voltage, and current measurement.

9. Programming, Debug Interfaces, Resources

PSoC devices include extensive support for programming, testing, debugging, and tracing both hardware and firmware. Three interfaces are available: JTAG, SWD, and SWV. JTAG and SWD support all programming and debug features of the device. JTAG also supports standard JTAG scan chains for board level test and chaining multiple JTAG devices to a single JTAG connection.

Complete Debug on Chip (DoC) functionality enables full device debugging in the final system using the standard production device. It does not require special interfaces, debugging pods, simulators, or emulators. Only the standard programming connections are required to fully support debug.

The PSoC Creator IDE software provides fully integrated programming and debug support for PSoC devices. The low cost MiniProg3 programmer and debugger is designed to provide full programming and debug support of PSoC devices in conjunction with the PSoC Creator IDE. PSoC JTAG, SWD, and SWV interfaces are fully compatible with industry standard third party tools.

All DOC circuits are disabled by default and can only be enabled in firmware. If not enabled, the only way to reenale them is to erase the entire device, clear flash protection, and reprogram the device with new firmware that enables DOC. Disabling DOC features, robust flash protection, and hiding custom analog and digital functionality inside the PSoC device provide a level of security not possible with multichip application solutions. Additionally, all device interfaces can be permanently disabled (Device Security) for applications concerned about phishing

attacks due to a maliciously reprogrammed device. Permanently disabling interfaces is not recommended in most applications because you cannot access the device later. Because all programming, debug, and test interfaces are disabled when device security is enabled, PSoCs with Device Security enabled may not be returned for failure analysis.

Table 9-1. Debug Configurations

Debug and Trace Configuration	GPIO Pins Used
All debug and trace disabled	0
JTAG	4 or 5
SWD	2
SWV	1
SWD + SWV	3

9.1 JTAG Interface

The IEEE 1149.1 compliant JTAG interface exists on four or five pins (the nTRST pin is optional). The JTAG clock frequency can be up to 8 MHz, or 1/3 of the CPU clock frequency for 8 and 16-bit transfers, or 1/5 of the CPU clock frequency for 32-bit transfers, whichever is least. By default, the JTAG pins are enabled on new devices but the JTAG interface can be disabled, allowing these pins to be used as General Purpose I/O (GPIO) instead. The JTAG interface is used for programming the flash memory, debugging, I/O scan chains, and JTAG device chaining.

9.2 Serial Wire Debug Interface

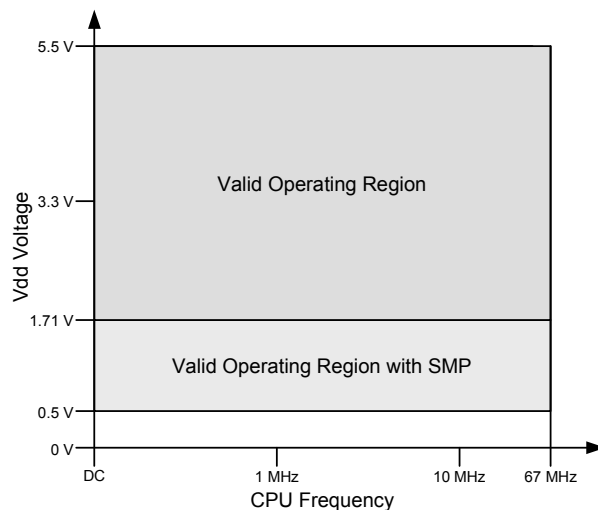
The SWD interface is the preferred alternative to the JTAG interface. It requires only two pins instead of the four or five needed by JTAG. SWD provides all of the programming and debugging features of JTAG at the same speed. SWD does not provide access to scan chains or device chaining. The SWD clock frequency can be up to 1/3 of the CPU clock frequency.

SWD uses two pins, either two of the JTAG pins (TMS and TCK) or the USBIO D+ and D– pins. The USBIO pins are useful for in system programming of USB solutions that would otherwise require a separate programming connector. One pin is used for the data clock and the other is used for data input and output. SWD can be enabled on only one of the pin pairs at a time. This only happens if, within 8 μ s (key window) after reset, that pin pair (JTAG or USB) receives a predetermined sequence of 1s and 0s. SWD is used for debugging or for programming the flash memory.

The SWD interface can be enabled from the JTAG interface or disabled, allowing its pins to be used as GPIO. Unlike JTAG, the SWD interface can always be reacquired on any device during the key window. It can then be used to reenale the JTAG interface, if desired. When using SWD or JTAG pins as standard GPIO, make sure that the GPIO functionality and PCB circuits do not interfere with SWD or JTAG use.

Table 11-3. AC Specifications^[24]

Parameter	Description	Conditions	Min	Typ	Max	Units
F _{CPU}	CPU frequency	$1.71\text{ V} \leq V_{DD} \leq 5.5\text{ V}$	DC	–	67	MHz
F _{BUSCLK}	Bus frequency	$1.71\text{ V} \leq V_{DD} \leq 5.5\text{ V}$	DC	–	67	MHz
Svdd	V _{DD} ramp rate		–	–	1	V/ns
T _{IO_INIT}	Time from V _{DDD} /V _{DDA} /V _{CCD} /V _{CCA} ≥ IPOR to I/O ports set to their reset states		–	–	10	μs
T _{STARTUP}	Time from V _{DDD} /V _{DDA} /V _{CCD} /V _{CCA} ≥ PRES to CPU executing code at reset vector	V _{CCA} /V _{DDA} = regulated from V _{DDA} /V _{DDD} , no PLL used, fast IMO boot mode (48 MHz typ.)	–	–	33	μs
		V _{CCA} /V _{CCD} = regulated from V _{DDA} /V _{DDD} , no PLL used, slow IMO boot mode (12 MHz typ.)	–	–	66	μs
T _{SLEEP}	Wakeup from sleep mode – Application of non-LVD interrupt to beginning of execution of next CPU instruction		–	–	15	μs
T _{HIBERNATE}	Wakeup from hibernate mode – Application of external interrupt to beginning of execution of next CPU instruction		–	–	100	μs

Figure 11-1. F_{CPU} vs. V_{DD}

Note

24. Based on device characterization (Not production tested).

11.4.2 SIO

Table 11-11. SIO DC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
Vinmax	Maximum input voltage	All allowed values of Vddio and Vddd, see Section 11.2.1	–	–	5.5	V
Vinref	Input voltage reference (Differential input mode)		0.5	–	$0.52 \times V_{DDIO}$	V
Voutref	Output voltage reference (Regulated output mode)					
		$V_{DDIO} > 3.7$	1	–	$V_{DDIO} - 1$	V
		$V_{DDIO} < 3.7$	1	–	$V_{DDIO} - 0.5$	V
V _{IH}	Input voltage high threshold					
	GPIO mode	CMOS input	$0.7 \times V_{DDIO}$	–	–	V
	Differential input mode ^[30]	Hysteresis disabled	SIO_ref + 0.2	–	–	V
V _{IL}	Input voltage low threshold					
	GPIO mode	CMOS input	–	–	$0.3 \times V_{DDIO}$	V
	Differential input mode ^[30]	Hysteresis disabled	–	–	SIO_ref – 0.2	V
V _{OH}	Output voltage high					
	Unregulated mode	$I_{OH} = 4 \text{ mA}$, $V_{DDIO} = 3.3 \text{ V}$	$V_{DDIO} - 0.4$	–	–	V
	Regulated mode ^[30]	$I_{OH} = 1 \text{ mA}$	SIO_ref – 0.65	–	SIO_ref + 0.2	V
	Regulated mode ^[30]	$I_{OH} = 0.1 \text{ mA}$	SIO_ref – 0.3	–	SIO_ref + 0.2	V
V _{OL}	Output voltage low					
		$V_{DDIO} = 3.30 \text{ V}$, $I_{OL} = 25 \text{ mA}$	–	–	0.8	V
		$V_{DDIO} = 1.80 \text{ V}$, $I_{OL} = 4 \text{ mA}$	–	–	0.4	V
Rpullup	Pull-up resistor		3.5	5.6	8.5	kΩ
Rpulldown	Pull-down resistor		3.5	5.6	8.5	kΩ
I _{IL}	Input leakage current (Absolute value) ^[31]					
	$V_{IH} \leq V_{ddio}$	25 °C, Vddio = 3.0 V, V _{IH} = 3.0 V	–	–	14	nA
	$V_{IH} > V_{ddio}$	25 °C, Vddio = 0 V, V _{IH} = 3.0 V	–	–	10	μA
C _{IN}	Input Capacitance ^[31]		–	–	7	pF
V _H	Input voltage hysteresis (Schmitt-Trigger) ^[31]	Single ended mode (GPIO mode)	–	40	–	mV
		Differential mode	–	35	–	mV
Idiode	Current through protection diode to V _{SSIO}		–	–	100	μA

Table 11-12. SIO AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
TriseF	Rise time in fast strong mode (90/10%) ^[31]	Clod = 25 pF, V _{DDIO} = 3.3 V	–	–	12	ns
TfallF	Fall time in fast strong mode (90/10%) ^[31]	Clod = 25 pF, V _{DDIO} = 3.3 V	–	–	12	ns
TriseS	Rise time in slow strong mode (90/10%) ^[31]	Clod = 25 pF, V _{DDIO} = 3.0 V	–	–	75	ns
TfallS	Fall time in slow strong mode (90/10%) ^[31]	Clod = 25 pF, V _{DDIO} = 3.0 V	–	–	60	ns

Notes

30. See Figure 6-9 on page 30 and Figure 6-12 on page 33 for more information on SIO reference.

31. Based on device characterization (Not production tested).

Table 11-14. USBIO AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
Tdrate	Full-speed data rate average bit rate		12 – 0.25%	12	12 + 0.25%	MHz
Tjr1	Receiver data jitter tolerance to next transition		–8	–	8	ns
Tjr2	Receiver data jitter tolerance to pair transition		–5	–	5	ns
Tdj1	Driver differential jitter to next transition		–3.5	–	3.5	ns
Tdj2	Driver differential jitter to pair transition		–4	–	4	ns
Tfdeop	Source jitter for differential transition to SE0 transition		–2	–	5	ns
Tfeopt	Source SE0 interval of EOP		160	–	175	ns
Tfeopr	Receiver SE0 interval of EOP		82	–	–	ns
Tfst	Width of SE0 interval during differential transition		–	–	14	ns
Fgpio_out	GPIO mode output operating frequency	$3\text{ V} \leq V_{\text{DDD}} \leq 5.5\text{ V}$	–	–	20	MHz
		$V_{\text{DDD}} = 1.71\text{ V}$	–	–	6	MHz
Tr_gpio	Rise time, GPIO mode, 10%/90% V_{DDD}	$V_{\text{DDD}} > 3\text{ V}$, 25 pF load	–	–	12	ns
		$V_{\text{DDD}} = 1.71\text{ V}$, 25 pF load	–	–	40	ns
Tf_gpio	Fall time, GPIO mode, 90%/10% V_{DDD}	$V_{\text{DDD}} > 3\text{ V}$, 25 pF load	–	–	12	ns
		$V_{\text{DDD}} = 1.71\text{ V}$, 25 pF load	–	–	40	ns

Table 11-15. USB Driver AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
Tr	Transition rise time		–	–	20	ns
Tf	Transition fall time		–	–	20	ns
TR	Rise/fall time matching	V_{USB_5} , $V_{\text{USB}_3.3}$, see USB DC Specifications on page 93	90%	–	111%	
Vcrs	Output signal crossover voltage		1.3	–	2	V

11.4.4 XRES

Table 11-16. XRES DC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
V_{IH}	Input voltage high threshold		$0.7 \times V_{\text{DDIO}}$	–	–	V
V_{IL}	Input voltage low threshold		–	–	$0.3 \times V_{\text{DDIO}}$	V
Rpullup	Pull-up resistor		3.5	5.6	8.5	k Ω
C_{IN}	Input capacitance ^[32]		–	3	–	pF
V_{H}	Input voltage hysteresis (Schmitt-Trigger) ^[32]		–	100	–	mV
Idiode	Current through protection diode to V_{DDIO} and V_{SSIO}		–	–	100	μA

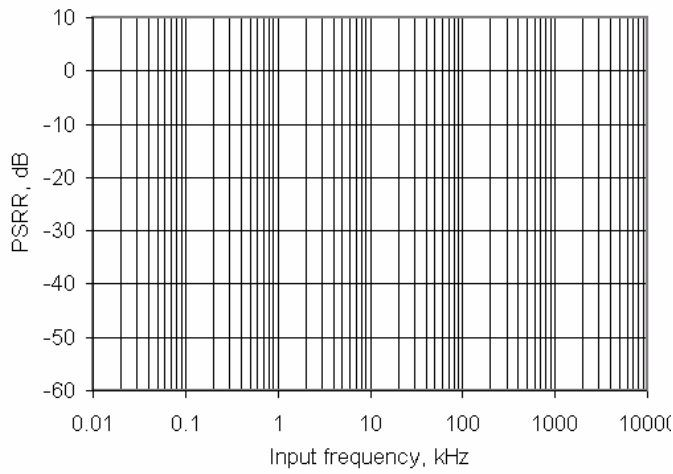
Table 11-17. XRES AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
T _{RESET}	Reset pulse width		1	–	–	μs

Note

32. Based on device characterization (Not production tested).

Figure 11-16. Opamp PSRR vs Frequency



11.5.7 Voltage Digital to Analog Converter (VDAC)

See the VDAC component datasheet in PSoc Creator for full electrical specifications and APIs.

Unless otherwise specified, all charts and graphs show typical values.

Table 11-31. VDAC DC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
	Resolution		–	8	–	bits
INL1	Integral nonlinearity	1 V scale	–	±2.1	±2.5	LSB
DNL1	Differential nonlinearity	1 V scale	–	±0.3	±1	LSB
Rout	Output resistance	1 V scale	–	4	–	kΩ
		4 V scale	–	16	–	kΩ
V _{OUT}	Output voltage range, code = 255	1 V scale	–	1	–	V
		4 V scale, V _{dda} = 5 V	–	4	–	V
	Monotonicity		–	–	Yes	–
V _{OS}	Zero scale error		–	0	±0.9	LSB
E _g	Gain error	1 V scale, 25 °C	–	–	±2.5	%
		4 V scale, 25 °C	–	–	±2.5	%
TC_Eg	Temperature coefficient, gain error	1 V scale, 25 °C	–	–	0.03	%FSR / °C
		4 V scale, 25 °C	–	–	0.03	%FSR / °C
I _{DD}	Operating current	Slow mode	–	–	100	μA
		Fast mode	–	–	500	μA

Figure 11-35. VDAC INL vs Input Code, 1 V Mode

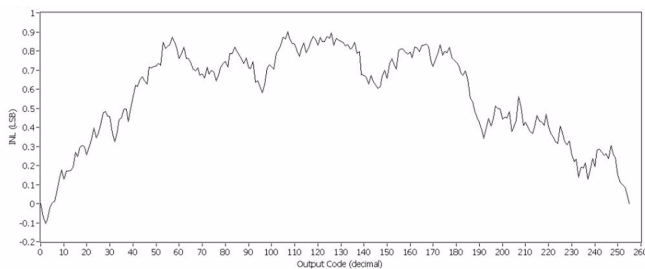


Figure 11-36. VDAC DNL vs Input Code, 1 V Mode

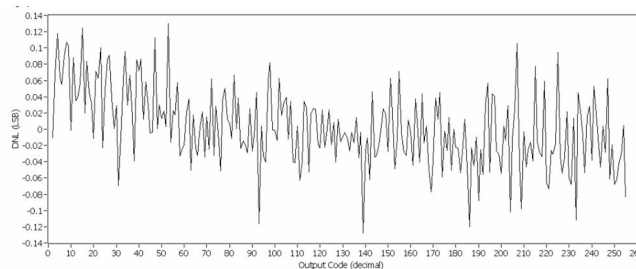


Figure 11-37. VDAC INL vs Temperature, 1 V Mode

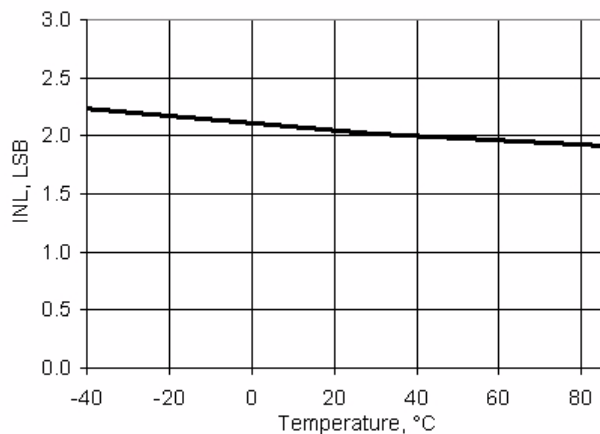
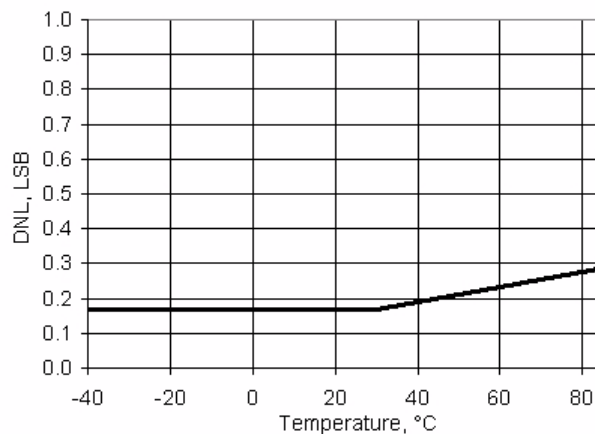


Figure 11-38. VDAC DNL vs Temperature, 1 V Mode



11.6.6 Digital Filter Block

Table 11-51. DFB DC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
	DFB operating current	64-tap FIR at F_{DFB}				
		100 kHz (1.3 ksps)	–	0.03	0.05	mA
		500 kHz (6.7 ksps)	–	0.16	0.27	mA
		1 MHz (13.4 ksps)	–	0.33	0.53	mA
		10 MHz (134 ksps)	–	3.3	5.3	mA
		48 MHz (644 ksps)	–	15.7	25.5	mA
		67 MHz (900 ksps)	–	21.8	35.6	mA

Table 11-52. DFB AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
F_{DFB}	DFB operating frequency		DC	–	67	MHz

11.6.7 USB

Table 11-53. USB DC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
V_{USB_5}	Device supply for USB operation	USB configured, USB regulator enabled	4.35	–	5.25	V
$V_{USB_3.3}$		USB configured, USB regulator bypassed	3.15	–	3.6	V
V_{USB_3}		USB configured, USB regulator bypassed ^[42]	2.85	–	3.6	V

Note

42. Rise/fall time matching (TR) not guaranteed, see [USB Driver AC Specifications](#) on page 68.

Table 11-63. Asynchronous Read Cycle Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
T	EMIF clock period ^[43]	$V_{DDA} \geq 3.3 \text{ V}$	30.3	–	–	nS
Tcel	EM_CEn low time		$2T - 5$	–	$2T + 5$	nS
Taddrv	EM_CEn low to EM_Addr valid		–	–	5	nS
Taddrh	Address hold time after EM_Wen high		T	–	–	nS
Toel	EM_OEn low time		$2T - 5$	–	$2T + 5$	nS
Tdoesu	Data to EM_OEn high setup time		$T + 15$	–	–	nS
Tdoeh	Data hold time after EM_OEn high		3	–	–	nS

Figure 11-49. Asynchronous Write Cycle Timing

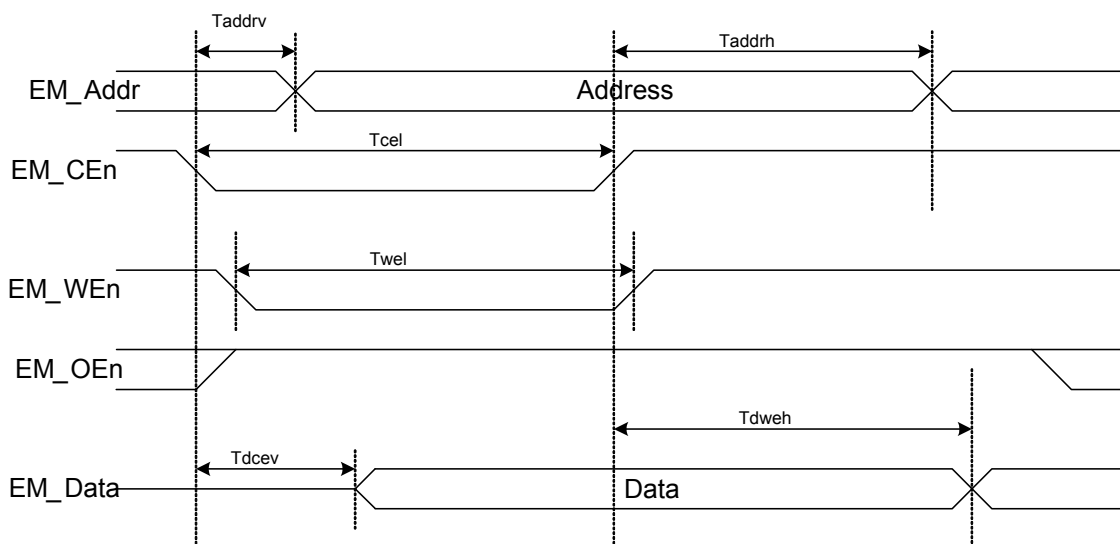


Table 11-64. Asynchronous Write Cycle Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
T	EMIF clock period ^[43]	$V_{DDA} \geq 3.3 \text{ V}$	30.3	–	–	nS
Tcel	EM_CEn low time		$T - 5$	–	$T + 5$	nS
Taddrv	EM_CEn low to EM_Addr valid		–	–	5	nS
Taddrh	Address hold time after EM_WEn high		T	–	–	nS
Twel	EM_WEn low time		$T - 5$	–	$T + 5$	nS
Tdcev	EM_CEn low to data valid		–	–	7	nS
Tdweh	Data hold time after EM_WEn high		T	–	–	nS

Note

43. Limited by GPIO output frequency, see [Table 11-10](#) on page 65.

Table 11-78. IMO AC Specifications (continued)

Parameter	Description	Conditions	Min	Typ	Max	Units
Jp-p	Jitter (peak to peak) ^[50]					
	F = 24 MHz		–	0.9	–	ns
	F = 3 MHz		–	1.6	–	ns
Jperiod	Jitter (long term) ^[50]					
	F = 24 MHz		–	0.9	–	ns
	F = 3 MHz		–	12	–	ns

11.9.3 Internal Low-Speed Oscillator

Table 11-79. ILO DC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
I _{CC}	Operating current	F _{OUT} = 1 kHz	–	0.3	1.7	μA
		F _{OUT} = 33 kHz	–	1.0	2.6	μA
		F _{OUT} = 100 kHz	–	1.0	2.6	μA
	Leakage current	Power down mode	–	2.0	15	nA

Table 11-80. ILO AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
	Startup time, all frequencies	Turbo mode	–	–	2	ms
F _{ILO}	ILO frequencies (trimmed)					
	100 kHz		45	100	200	kHz
	1 kHz		0.5	1	2	kHz
	ILO frequencies (untrimmed)					
	100 kHz		30	100	300	kHz
	1 kHz		0.3	1	3.5	kHz

11.9.4 External Crystal Oscillator

Table 11-81. ECO AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
F	Crystal frequency range		4	–	33	MHz

11.9.5 External Clock Reference

Table 11-82. External Clock Reference AC Specifications^[50]

Parameter	Description	Conditions	Min	Typ	Max	Units
	External frequency range		0	–	33	MHz
	Input duty cycle range	Measured at V _{DDIO} /2	30	50	70	%
	Input edge rate	V _{IL} to V _{IH}	0.1	–	–	V/ns

Note
Note

50. Based on device characterization (Not production tested).