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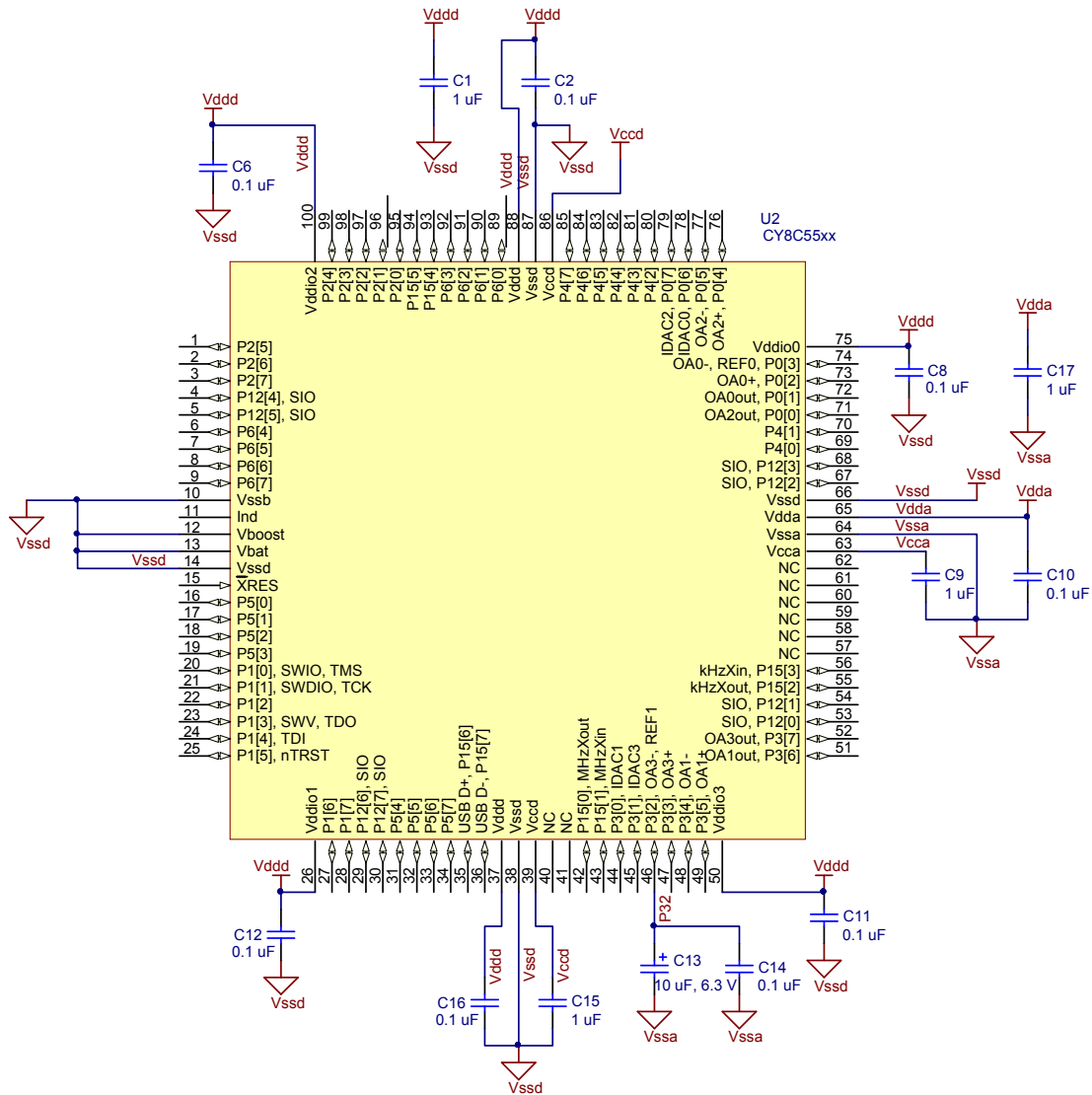
"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "[Embedded - Microcontrollers](#)"

Details

Product Status	Obsolete
Core Processor	8051
Core Size	8-Bit
Speed	67MHz
Connectivity	EBI/EMI, I ² C, LINbus, SPI, UART/USART
Peripherals	CapSense, DMA, LCD, POR, PWM, WDT
Number of I/O	25
Program Memory Size	64KB (64K x 8)
Program Memory Type	FLASH
EEPROM Size	2K x 8
RAM Size	8K x 8
Voltage - Supply (Vcc/Vdd)	1.71V ~ 5.5V
Data Converters	A/D 16x20b; D/A 4x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	48-BSSOP (0.295", 7.50mm Width)
Supplier Device Package	48-SSOP
Purchase URL	https://www.e-xfl.com/product-detail/infineon-technologies/cy8c3866pvi-066

Figure 2-5. Example Schematic for 100-pin TQFP Part with Power Connections



Note The two Vccd pins must be connected together with as short a trace as possible. A trace under the device is recommended, as shown in [Figure 2-6](#) on page 10.

4.4.4.7 Nested DMA

One TD may modify another TD, as the TD configuration space is memory mapped similar to any other peripheral. For example, a first TD loads a second TD's configuration and then calls the second TD. The second TD moves data as required by the application. When complete, the second TD calls the first TD, which again updates the second TD's configuration. This process repeats as often as necessary.

4.5 Interrupt Controller

The interrupt controller provides a mechanism for hardware resources to change program execution to a new address, independent of the current task being executed by the main code. The interrupt controller provides enhanced features not found on original 8051 interrupt controllers:

- Thirty-two interrupt vectors
- Jumps directly to ISR anywhere in code space with dynamic vector addresses
- Multiple sources for each vector
- Flexible interrupt to vector matching
- Each interrupt vector is independently enabled or disabled
- Each interrupt can be dynamically assigned one of eight priorities
- Eight level nestable interrupts
- Multiple I/O interrupt vectors
- Software can send interrupts
- Software can clear pending interrupts

When an interrupt is pending, the current instruction is completed and the program counter is pushed onto the stack. Code execution then jumps to the program address provided by the vector. After the ISR is completed, a RETI instruction is executed and returns execution to the instruction following the previously interrupted instruction. To do this the RETI instruction pops the program counter from the stack.

If the same priority level is assigned to two or more interrupts, the interrupt with the lower vector number is executed first. Each interrupt vector may choose from three interrupt sources: Fixed Function, DMA, and UDB. The fixed function interrupts are direct connections to the most common interrupt sources and provide the lowest resource cost connection. The DMA interrupt sources provide direct connections to the two DMA interrupt sources provided per DMA channel. The third interrupt source for vectors is from the UDB digital routing array. This allows any digital signal available to the UDB array to be used as an interrupt source.

Fixed function interrupts and all interrupt sources may be routed to any interrupt vector using the UDB interrupt source connections.

Table 4-8. Interrupt Vector Table

#	Fixed Function	DMA	UDB
0	LVD	phub_termout0[0]	udb_intr[0]
1	ECC	phub_termout0[1]	udb_intr[1]
2	Reserved	phub_termout0[2]	udb_intr[2]
3	Sleep (Pwr Mgr)	phub_termout0[3]	udb_intr[3]
4	PICU[0]	phub_termout0[4]	udb_intr[4]
5	PICU[1]	phub_termout0[5]	udb_intr[5]
6	PICU[2]	phub_termout0[6]	udb_intr[6]
7	PICU[3]	phub_termout0[7]	udb_intr[7]
8	PICU[4]	phub_termout0[8]	udb_intr[8]
9	PICU[5]	phub_termout0[9]	udb_intr[9]
10	PICU[6]	phub_termout0[10]	udb_intr[10]
11	PICU[12]	phub_termout0[11]	udb_intr[11]
12	PICU[15]	phub_termout0[12]	udb_intr[12]
13	Comparators Combined	phub_termout0[13]	udb_intr[13]
14	Switched Caps Combined	phub_termout0[14]	udb_intr[14]
15	I ² C	phub_termout0[15]	udb_intr[15]
16	CAN	phub_termout1[0]	udb_intr[16]
17	Timer/Counter0	phub_termout1[1]	udb_intr[17]
18	Timer/Counter1	phub_termout1[2]	udb_intr[18]
19	Timer/Counter2	phub_termout1[3]	udb_intr[19]
20	Timer/Counter3	phub_termout1[4]	udb_intr[20]
21	USB SOF Int	phub_termout1[5]	udb_intr[21]
22	USB Arb Int	phub_termout1[6]	udb_intr[22]
23	USB Bus Int	phub_termout1[7]	udb_intr[23]
24	USB Endpoint[0]	phub_termout1[8]	udb_intr[24]
25	USB Endpoint Data	phub_termout1[9]	udb_intr[25]
26	Reserved	phub_termout1[10]	udb_intr[26]
27	Reserved	phub_termout1[11]	udb_intr[27]
28	DFB Int	phub_termout1[12]	udb_intr[28]
29	Decimator Int	phub_termout1[13]	udb_intr[29]
30	PHUB Error Int	phub_termout1[14]	udb_intr[30]

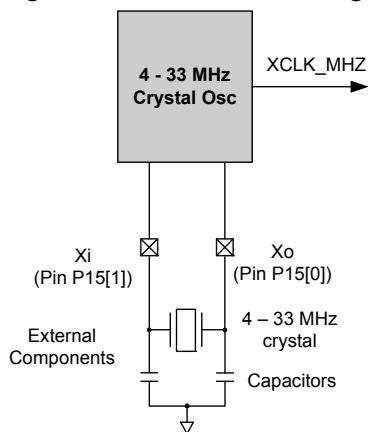
The 33-kHz clock (CLK33K) comes from a divide-by-3 operation on CLK100K. This output can be used as a reduced accuracy version of the 32.768-kHz ECO clock with no need for a crystal.

6.1.2 External Oscillators

6.1.2.1 MHz External Crystal Oscillator

The MHzECO provides high frequency, high precision clocking using an external crystal (see Figure 6-2). It supports a wide variety of crystal types, in the range of 4 to 33 MHz. When used in conjunction with the PLL, it can generate CPU and system clocks up to the device's maximum frequency (see PLL). The GPIO pins connecting to the external crystal and capacitors are fixed. MHzECO accuracy depends on the crystal chosen.

Figure 6-2. MHzECO Block Diagram

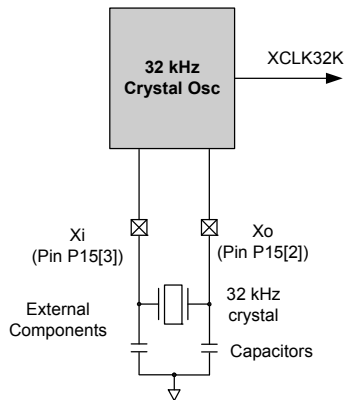


6.1.2.2 32.768-kHz ECO

The 32.768-kHz external crystal oscillator (32kHzECO) provides precision timing with minimal power consumption using an external 32.768-kHz watch crystal (see Figure 6-3). The 32kHzECO also connects directly to the sleep timer and provides the source for the RTC. The RTC uses a 1-second interrupt to implement the RTC functionality in firmware.

The oscillator works in two distinct power modes. This allows users to trade off power consumption with noise immunity from neighboring circuits. The GPIO pins connected to the external crystal and capacitors are fixed.

Figure 6-3. 32kHzECO Block Diagram



6.1.2.3 Digital System Interconnect

The DSI provides routing for clocks taken from external clock oscillators connected to I/O. The oscillators can also be generated within the device in the digital system and UDBs.

While the primary DSI clock input provides access to all clocking resources, up to eight other DSI clocks (internally or externally generated) may be routed directly to the eight digital clock dividers. This is only possible if there are multiple precision clock sources.

6.1.3 Clock Distribution

All seven clock sources are inputs to the central clock distribution system. The distribution system is designed to create multiple high precision clocks. These clocks are customized for the design's requirements and eliminate the common problems found with limited resolution prescalers attached to peripherals. The clock distribution system generates several types of clock trees.

- The system clock is used to select and supply the fastest clock in the system for general system clock requirements and clock synchronization of the PSoC device.
- Bus clock 16-bit divider uses the system clock to generate the system's bus clock used for data transfers. Bus clock is the source clock for the CPU clock divider.
- Eight fully programmable 16-bit clock dividers generate digital system clocks for general use in the digital system, as configured by the design's requirements. Digital system clocks can generate custom clocks derived from any of the seven clock sources for any purpose. Examples include baud rate generators, accurate PWM periods, and timer clocks, and many others. If more than eight digital clock dividers are required, the UDBs and fixed function timer/counter/PWMs can also generate clocks.
- Four 16-bit clock dividers generate clocks for the analog system components that require clocking, such as ADC and mixers. The analog clock dividers include skew control to ensure that critical analog events do not occur simultaneously with digital switching events. This is done to reduce analog system noise.

Each clock divider consists of an 8-input multiplexer, a 16-bit clock divider (divide by 2 and higher) that generates ~50 percent duty cycle clocks, system clock resynchronization logic, and deglitch logic. The outputs from each digital clock tree can be routed into the digital system interconnect and then brought back into the clock system as an input, allowing clock chaining of up to 32 bits.

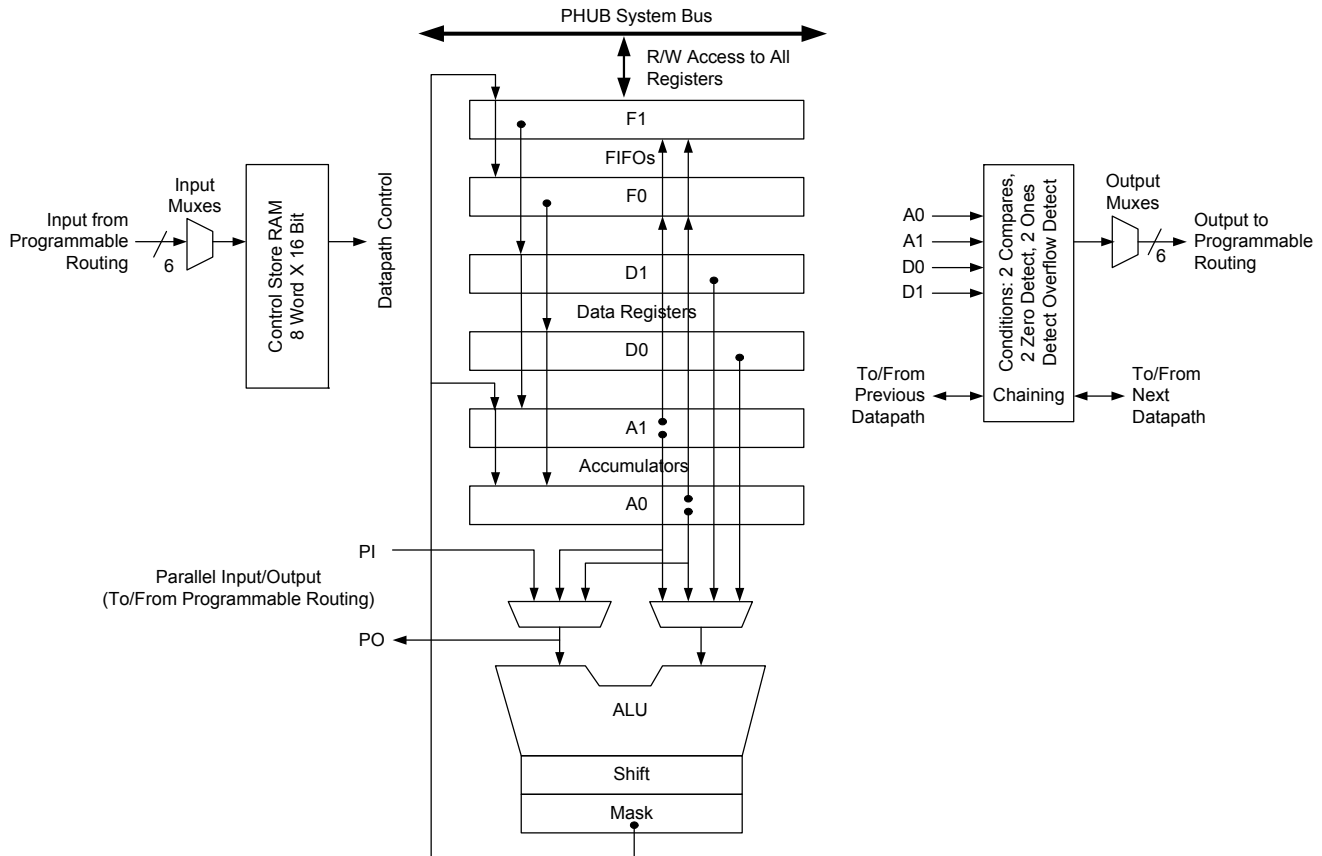
6.1.4 USB Clock Domain

The USB clock domain is unique in that it operates largely asynchronously from the main clock network. The USB logic contains a synchronous bus interface to the chip, while running on an asynchronous clock to process USB data. The USB logic requires a 48 MHz frequency. This frequency can be generated from different sources, including DSI clock at 48 MHz or doubled value of 24 MHz from internal oscillator, DSI signal, or crystal oscillator.

7.2.2 Datapath Module

The datapath contains an 8-bit single cycle ALU, with associated compare and condition generation logic. This datapath block is optimized to implement embedded functions, such as timers, counters, integrators, PWMs, PRS, CRC, shifters and dead band generators and many others.

Figure 7-8. Datapath Top Level



7.2.2.1 Working Registers

The datapath contains six primary working registers, which are accessed by CPU firmware or DMA during normal operation.

Table 7-1. Working Datapath Registers

Name	Function	Description
A0 and A1	Accumulators	These are sources and sinks for the ALU and also sources for the compares.
D0 and D1	Data Registers	These are sources for the ALU and sources for the compares.
F0 and F1	FIFOs	These are the primary interface to the system bus. They can be a data source for the data registers and accumulators or they can capture data from the accumulators or ALU. Each FIFO is four bytes deep.

7.2.2.2 Dynamic Datapath Configuration RAM

Dynamic configuration is the ability to change the datapath function and internal configuration on a cycle-by-cycle basis, under sequencer control. This is implemented using the 8-word × 16-bit configuration RAM, which stores eight unique 16-bit wide configurations. The address input to this RAM controls the sequence, and can be routed from any block connected to the UDB routing matrix, most typically PLD logic, I/O pins, or from the outputs of this or other datapath blocks.

ALU

The ALU performs eight general purpose functions. They are:

- Increment
- Decrement
- Add
- Subtract
- Logical AND
- Logical OR
- Logical XOR
- Pass, used to pass a value through the ALU to the shift register, mask, or another UDB register.

The PSoC Creator software program provides a user friendly interface to configure the analog connections between the GPIO and various analog resources and connections from one analog resource to another. PSoC Creator also provides component libraries that allow you to configure the various analog blocks to perform application specific functions (PGA, transimpedance amplifier, voltage DAC, current DAC, and so on). The tool also generates API interface libraries that allow you to write firmware that allows the communication between the analog peripheral and CPU/Memory.

8.1 Analog Routing

The CY8C38 family of devices has a flexible analog routing architecture that provides the capability to connect GPIOs and different analog blocks, and also route signals between different analog blocks. One of the strong points of this flexible routing architecture is that it allows dynamic routing of input and output connections to the different analog blocks.

8.1.1 Features

- Flexible, configurable analog routing architecture
- 16 analog globals (AG) and two analog mux buses (AMUXBUS) to connect GPIOs and the analog blocks

- Each GPIO is connected to one analog global and one analog mux bus
- Eight analog local buses (abus) to route signals between the different analog blocks
- Multiplexers and switches for input and output selection of the analog blocks

8.1.2 Functional Description

Analog globals (AGs) and analog mux buses (AMUXBUS) provide analog connectivity between GPIOs and the various analog blocks. There are 16 AGs in the CY8C38 family. The analog routing architecture is divided into four quadrants as shown in [Figure 8-2](#). Each quadrant has four analog globals (AGL[0..3], AGL[4..7], AGR[0..3], AGR[4..7]). Each GPIO is connected to the corresponding AG through an analog switch. The analog mux bus is a shared routing resource that connects to every GPIO through an analog switch. There are two AMUXBUS routes in CY8C38, one in the left half (AMUXBUSL) and one in the right half (AMUXBUSR), as shown in [Figure 8-2](#) on page 48.

Analog local buses (abus) are routing resources located within the analog subsystem and are used to route signals between different analog blocks. There are eight abus routes in CY8C38, four in the left half (abusl [0:3]) and four in the right half (abusr [0:3]) as shown in Figure 8-2. Using the abus saves the analog globals and analog mux buses from being used for interconnecting the analog blocks.

Multiplexers and switches exist on the various buses to direct signals into and out of the analog blocks. A multiplexer can have only one connection on at a time, whereas a switch can have multiple connections on simultaneously. In Figure 8-2, multiplexers are indicated by grayed ovals and switches are indicated by transparent ovals.

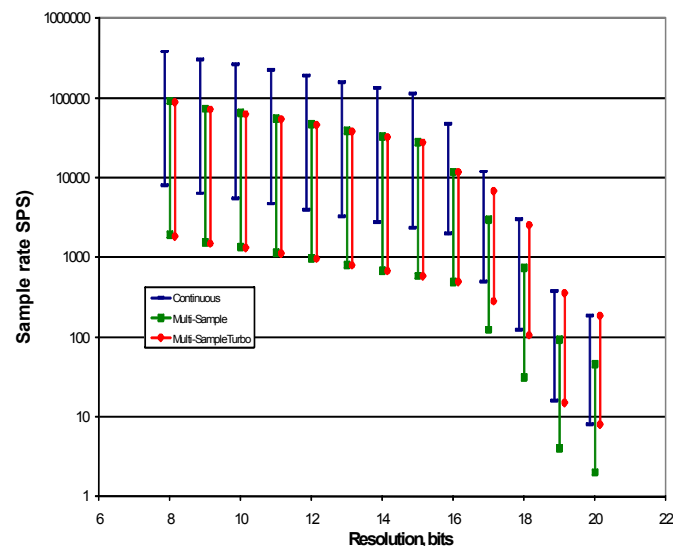
8.2 Delta-sigma ADC

The CY8C38 device contains one delta-sigma ADC. This ADC offers differential input, high resolution and excellent linearity, making it a good ADC choice for both audio signal processing and measurement applications. The converter's nominal operation is 16 bits at 48 ksp/s. The ADC can be configured to output 20-bit resolution at data rates of up to 187 sps. At a fixed clock rate, resolution can be traded for faster data rates as shown in Table 8-1 and Figure 8-3.

Table 8-1. Delta-sigma ADC Performance

Bits	Maximum Sample Rate (sps)	SINAD (dB)
20	187	—
16	48 k	84
12	192 k	66
8	384 k	43

Figure 8-3. Delta-sigma ADC Sample Rates, Range = ±1.024 V



8.2.1 Functional Description

The ADC connects and configures three basic components, input buffer, delta-sigma modulator, and decimator. The basic block diagram is shown in Figure 8-4. The signal from the input muxes is delivered to the delta-sigma modulator either directly or through the input buffer. The delta-sigma modulator performs the actual analog to digital conversion. The modulator over-samples the input and generates a serial data stream output. This high speed data stream is not useful for most applications without some type of post processing, and so is passed to the decimator through the Analog Interface block. The decimator converts the high speed serial data stream into parallel ADC results. The modulator/decimator frequency response is $[(\sin x)/x]^4$; a typical frequency response is shown in Figure 8-5.

Figure 8-4. Delta-sigma ADC Block Diagram

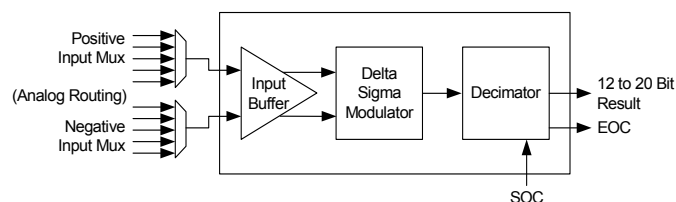
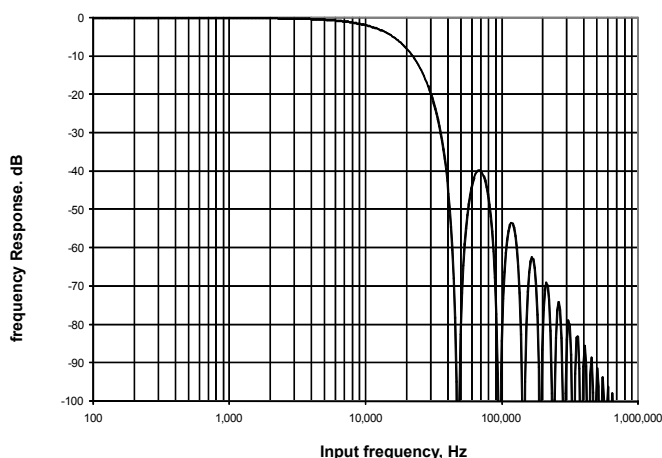


Figure 8-5. Delta-sigma ADC Frequency Response, Normalized to Output, Sample Rate = 48 kHz

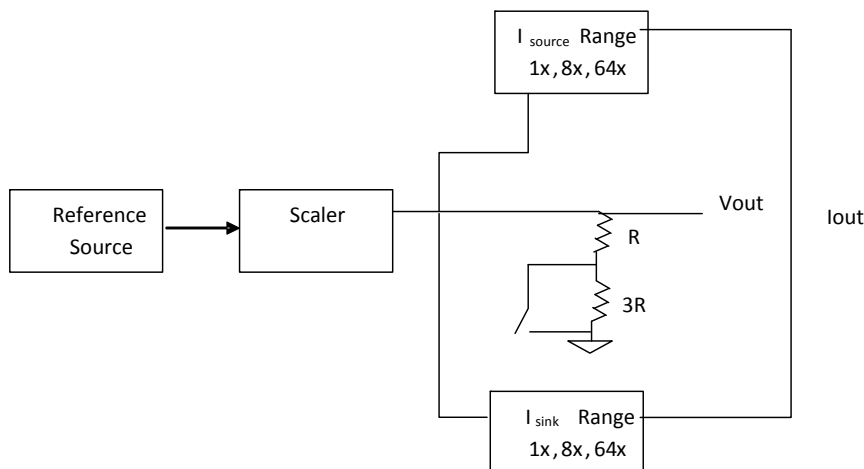


Resolution and sample rate are controlled by the Decimator. Data is pipelined in the decimator; the output is a function of the last four samples. When the input multiplexer is switched, the output data is not valid until after the fourth sample after the switch.

8.2.2 Operational Modes

The ADC can be configured by the user to operate in one of four modes: Single Sample, Multi Sample, Continuous, or Multi Sample (Turbo). All four modes are started by either a write to the start bit in a control register or an assertion of the Start of Conversion (SoC) signal. When the conversion is complete, a status bit is set and the output signal End of Conversion (EoC)

Figure 8-12. DAC Block Diagram



8.9.1 Current DAC

The current DAC (IDAC) can be configured for the ranges 0 to 32 μA , 0 to 256 μA , and 0 to 2.048 mA. The IDAC can be configured to source or sink current.

8.9.2 Voltage DAC

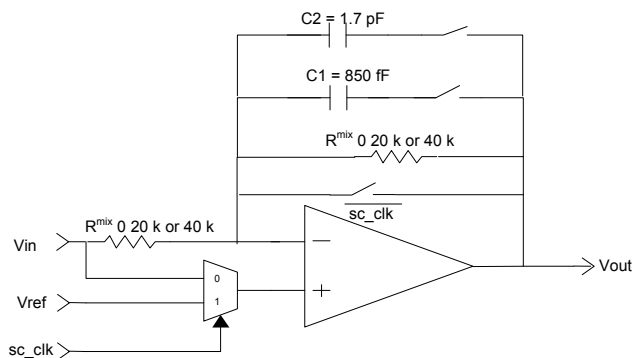
For the voltage DAC (VDAC), the current DAC output is routed through resistors. The two ranges available for the VDAC are 0 to 1.024 V and 0 to 4.096 V. In voltage mode any load connected to the output of a DAC should be purely capacitive (the output of the VDAC is not buffered).

8.10 Up/Down Mixer

In continuous time mode, the SC/CT block components are used to build an up or down mixer. Any mixing application contains an input signal frequency and a local oscillator frequency. The polarity of the clock, Fclk, switches the amplifier between inverting or noninverting gain. The output is the product of the input and the switching function from the local oscillator, with frequency components at the local oscillator plus and minus the signal frequency ($F_{\text{clk}} + F_{\text{in}}$ and $F_{\text{clk}} - F_{\text{in}}$) and reduced-level frequency components at odd integer multiples of the local oscillator frequency. The local oscillator frequency is provided by the selected clock source for the mixer.

Continuous time up and down mixing works for applications with input signals and local oscillator frequencies up to 1 MHz.

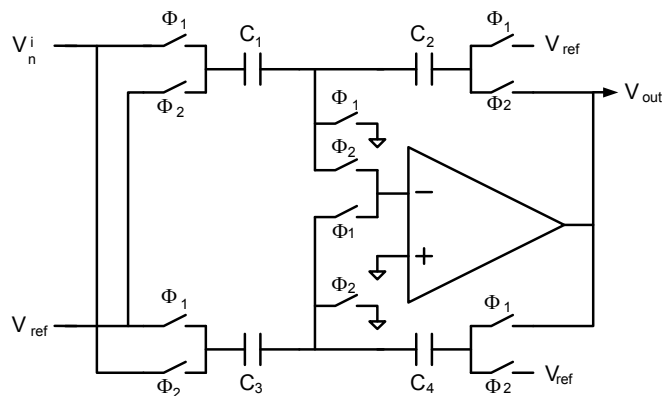
Figure 8-13. Mixer Configuration



8.11 Sample and Hold

The main application for a sample and hold, is to hold a value stable while an ADC is performing a conversion. Some applications require multiple signals to be sampled simultaneously, such as for power calculations (V and I).

Figure 8-14. Sample and Hold Topology
(Φ_1 and Φ_2 are opposite phases of a clock)



8.11.1 Down Mixer

The SC/CT block can be used as a mixer to down convert an input signal. This circuit is a high bandwidth passive sample network that can sample input signals up to 14 MHz. This sampled value is then held using the opamp with a maximum clock rate of 4 MHz. The output frequency is at the difference between the input frequency and the highest integer multiple of the Local Oscillator that is less than the input.

8.11.2 First Order Modulator – SC Mode

A first order modulator is constructed by placing the SC/CT block in an integrator mode and using a comparator to provide a 1-bit feedback to the input. Depending on this bit, a reference voltage is either subtracted or added to the input signal. The block output is the output of the comparator and not the integrator in the modulator case. The signal is downshifted and buffered and then processed by a decimator to make a delta-sigma converter or a counter to make an incremental converter. The accuracy of the sampled data from the first-order modulator is determined from several factors.

The main application for this modulator is for a low-frequency ADC with high accuracy. Applications include strain gauges, thermocouples, precision voltage, and current measurement.

9. Programming, Debug Interfaces, Resources

PSoC devices include extensive support for programming, testing, debugging, and tracing both hardware and firmware. Three interfaces are available: JTAG, SWD, and SWV. JTAG and SWD support all programming and debug features of the device. JTAG also supports standard JTAG scan chains for board level test and chaining multiple JTAG devices to a single JTAG connection.

Complete Debug on Chip (DoC) functionality enables full device debugging in the final system using the standard production device. It does not require special interfaces, debugging pods, simulators, or emulators. Only the standard programming connections are required to fully support debug.

The PSoC Creator IDE software provides fully integrated programming and debug support for PSoC devices. The low cost MiniProg3 programmer and debugger is designed to provide full programming and debug support of PSoC devices in conjunction with the PSoC Creator IDE. PSoC JTAG, SWD, and SWV interfaces are fully compatible with industry standard third party tools.

All DOC circuits are disabled by default and can only be enabled in firmware. If not enabled, the only way to reenale them is to erase the entire device, clear flash protection, and reprogram the device with new firmware that enables DOC. Disabling DOC features, robust flash protection, and hiding custom analog and digital functionality inside the PSoC device provide a level of security not possible with multichip application solutions. Additionally, all device interfaces can be permanently disabled (Device Security) for applications concerned about phishing

attacks due to a maliciously reprogrammed device. Permanently disabling interfaces is not recommended in most applications because you cannot access the device later. Because all programming, debug, and test interfaces are disabled when device security is enabled, PSoCs with Device Security enabled may not be returned for failure analysis.

Table 9-1. Debug Configurations

Debug and Trace Configuration	GPIO Pins Used
All debug and trace disabled	0
JTAG	4 or 5
SWD	2
SWV	1
SWD + SWV	3

9.1 JTAG Interface

The IEEE 1149.1 compliant JTAG interface exists on four or five pins (the nTRST pin is optional). The JTAG clock frequency can be up to 8 MHz, or 1/3 of the CPU clock frequency for 8 and 16-bit transfers, or 1/5 of the CPU clock frequency for 32-bit transfers, whichever is least. By default, the JTAG pins are enabled on new devices but the JTAG interface can be disabled, allowing these pins to be used as General Purpose I/O (GPIO) instead. The JTAG interface is used for programming the flash memory, debugging, I/O scan chains, and JTAG device chaining.

9.2 Serial Wire Debug Interface

The SWD interface is the preferred alternative to the JTAG interface. It requires only two pins instead of the four or five needed by JTAG. SWD provides all of the programming and debugging features of JTAG at the same speed. SWD does not provide access to scan chains or device chaining. The SWD clock frequency can be up to 1/3 of the CPU clock frequency.

SWD uses two pins, either two of the JTAG pins (TMS and TCK) or the USBIO D+ and D– pins. The USBIO pins are useful for in system programming of USB solutions that would otherwise require a separate programming connector. One pin is used for the data clock and the other is used for data input and output. SWD can be enabled on only one of the pin pairs at a time. This only happens if, within 8 μ s (key window) after reset, that pin pair (JTAG or USB) receives a predetermined sequence of 1s and 0s. SWD is used for debugging or for programming the flash memory.

The SWD interface can be enabled from the JTAG interface or disabled, allowing its pins to be used as GPIO. Unlike JTAG, the SWD interface can always be reacquired on any device during the key window. It can then be used to reenale the JTAG interface, if desired. When using SWD or JTAG pins as standard GPIO, make sure that the GPIO functionality and PCB circuits do not interfere with SWD or JTAG use.

Table 11-2. DC Specifications (continued)

Parameter	Description	Conditions		Min	Typ	Max	Units
	Sleep Mode ^[22]						
	CPU = OFF RTC = ON (= ECO32K ON, in low-power mode) Sleep timer = ON (= ILO ON at 1 kHz) ^[23] WDT = OFF I ² C Wake = OFF Comparator = OFF POR = ON Boost = OFF SIO pins in single ended input, unregulated output mode	V _{DD} = V _{DDIO} = 4.5–5.5 V	T = –40 °C	–	–	–	μA
			T = 25 °C	–	–	–	μA
			T = 85 °C	–	–	–	μA
		V _{DD} = V _{DDIO} = 2.7–3.6 V	T = –40 °C	–	–	–	μA
			T = 25 °C	–	1	–	μA
			T = 85 °C	–	–	–	μA
		V _{DD} = V _{DDIO} = 1.71–1.95 V	T = –40 °C	–	–	–	μA
			T = 25 °C	–	–	–	μA
			T = 85 °C	–	–	–	μA
	Comparator = ON CPU = OFF RTC = OFF Sleep timer = OFF WDT = OFF I2C Wake = OFF POR = ON Boost = OFF SIO pins in single ended input, unregulated output mode	V _{DD} = V _{DDIO} = 2.7–3.6V	T = 25 °C	–	–	–	μA
	I2C Wake = ON CPU = OFF RTC = OFF Sleep timer = OFF WDT = OFF Comparator = OFF POR = ON Boost = OFF SIO pins in single ended input, unregulated output mode	V _{DD} = V _{DDIO} = 2.7–3.6V	T = 25 °C	–	–	–	μA
	Hibernate Mode ^[22]						
	Hibernate mode current All regulators and oscillators off. SRAM retention GPIO interrupts are active Boost = OFF SIO pins in single ended input, unregulated output mode	V _{DD} = V _{DDIO} = 4.5–5.5 V	T = –40 °C	–	–	–	nA
			T = 25 °C	–	–	–	nA
			T = 85 °C	–	–	–	nA
V _{DD} = V _{DDIO} = 2.7–3.6 V		T = –40 °C	–	–	–	nA	
		T = 25 °C	–	200	–	nA	
		T = 85 °C	–	–	–	nA	
V _{DD} = V _{DDIO} = 1.71–1.95 V		T = –40 °C	–	–	–	nA	
		T = 25 °C	–	–	–	nA	
		T = 85 °C	–	–	–	nA	

Notes

 22. If V_{CCD} and V_{CCA} are externally regulated, the voltage difference between V_{CCD} and V_{CCA} must be less than 50 mV.

23. Sleep timer generates periodic interrupts to wake up the CPU. This specification applies only to those times that the CPU is off.

11.5.3 Voltage Reference

Table 11-25. Voltage Reference Specifications

See also ADC external reference specifications in *Section 11.5.2*.

Parameter	Description	Conditions	Min	Typ	Max	Units
V _{REF}	Precision reference voltage	Initial trimming	1.023 (−0.1%)	1.024	1.025 (+0.1%)	V
	Temperature drift ^[36]		–	–	20	ppm/°C
	Long term drift		–	100	–	ppm/Khr
	Thermal cycling drift (stability) ^[36]		–	100	–	ppm

11.5.4 Analog Globals

Table 11-26. Analog Globals Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
R _{ppag}	Resistance pin-to-pin through analog global ^[37]	V _{DDA} = 3.0 V	–	939	1461	Ω
R _{ppmuxbus}	Resistance pin-to-pin through analog mux bus ^[37]	V _{DDA} = 3.0 V	–	721	1135	Ω

11.5.5 Comparator

Table 11-27. Comparator DC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
V _{OS}	Input offset voltage in fast mode	Factory trim, V _{dda} > 2.7 V, V _{in} ≥ 0.5 V	–		10	mV
	Input offset voltage in slow mode	Factory trim, V _{in} ≥ 0.5 V	–		9	mV
V _{OS}	Input offset voltage in fast mode ^[38]	Custom trim	–	–	4	mV
	Input offset voltage in slow mode ^[38]	Custom trim	–	–	4	mV
V _{OS}	Input offset voltage in ultra low-power mode		–	±12	–	mV
V _{HYST}	Hysteresis	Hysteresis enable mode	–	10	32	mV
V _{ICM}	Input common mode voltage	High current / fast mode	V _{SSA}	–	V _{DDA} – 0.1	V
		Low current / slow mode	V _{SSA}	–	V _{DDA}	V
		Ultra low power mode	V _{SSA}	–	V _{DDA} – 0.9	V
CMRR	Common mode rejection ratio		–	50	–	dB
I _{CMP}	High current mode/fast mode ^[36]		–	–	400	μA
	Low current mode/slow mode ^[36]		–	–	100	μA
	Ultra low-power mode ^[36]		–	6	–	μA

Notes

36. Based on device characterization (Not production tested).

37. The resistance of the analog global and analog mux bus is high if V_{DDA} ≤ 2.7 V, and the chip is in either sleep or hibernate mode. Use of analog global and analog mux bus under these conditions is not recommended.

38. The recommended procedure for using a custom trim value for the on-chip comparators can be found in the TRM.

Figure 11-33. IDAC Operating Current vs Temperature, Range = 255 μ A, Code = 0, Source Mode

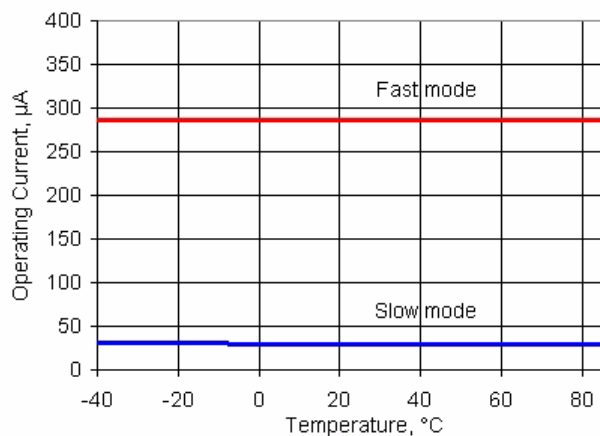


Figure 11-34. IDAC Operating Current vs Temperature, Range = 255 μ A, Code = 0, Sink Mode

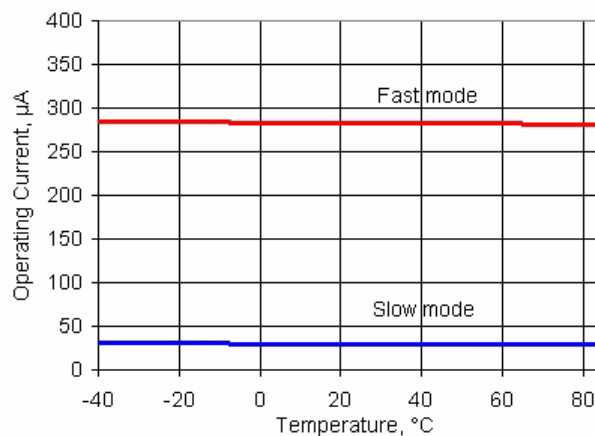


Table 11-30. IDAC AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
F_{DAC}	Update rate		–	–	8	Msp/s
T_{SETTLE}	Settling time to 0.5 LSB	Range = 31.875 μ A or 255 μ A, full scale transition, fast mode, 600 Ω 15-pF load	–	–	125	ns

Figure 11-39. VDAC Full Scale Error vs Temperature, 1 V Mode

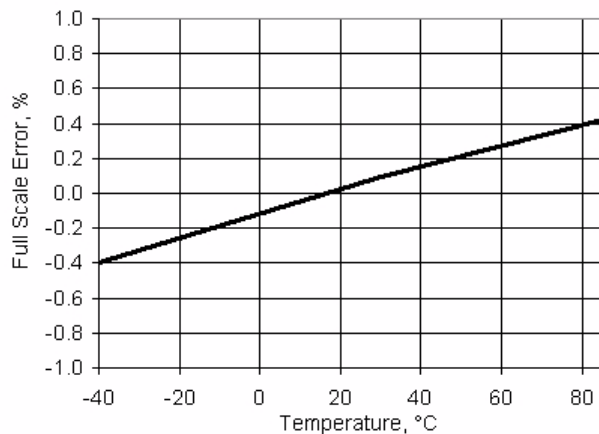


Figure 11-40. VDAC Full Scale Error vs Temperature, 4 V Mode

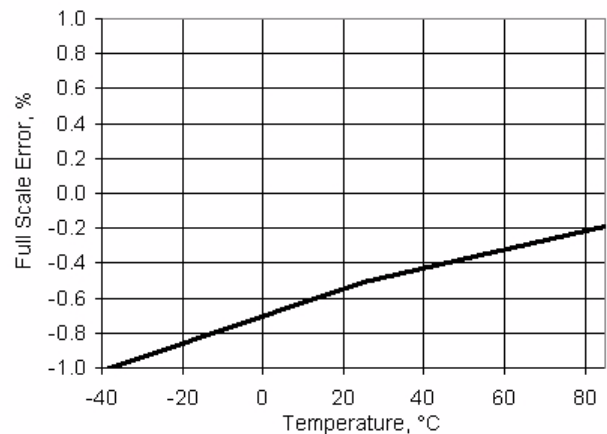


Figure 11-41. VDAC Operating Current vs Temperature, 1V Mode, Slow Mode

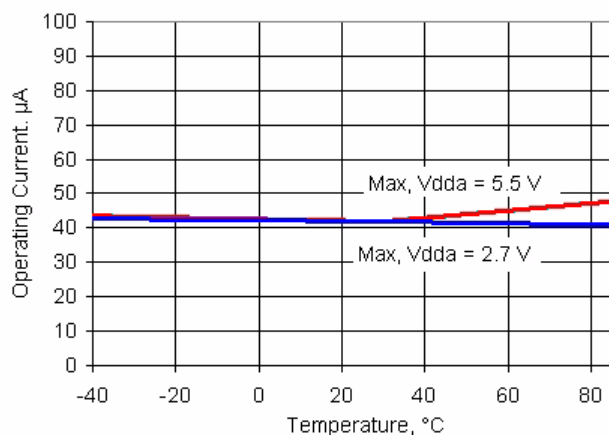


Figure 11-42. VDAC Operating Current vs Temperature, 1 V Mode, Fast Mode

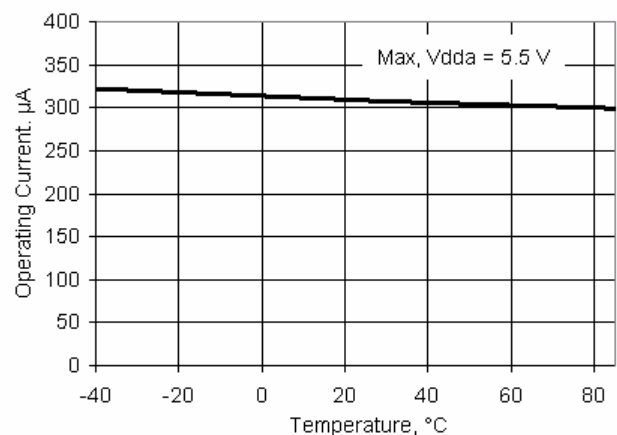
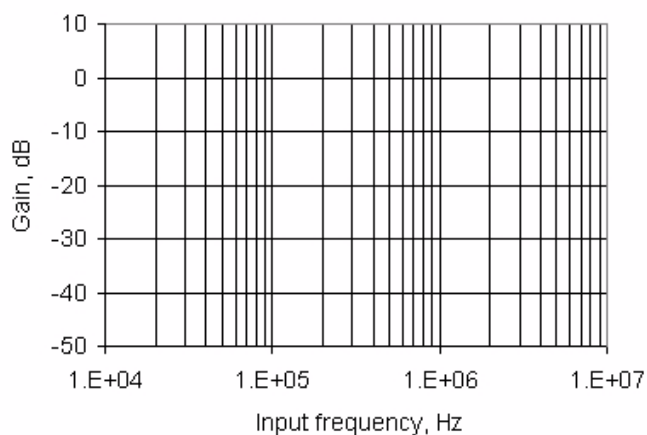
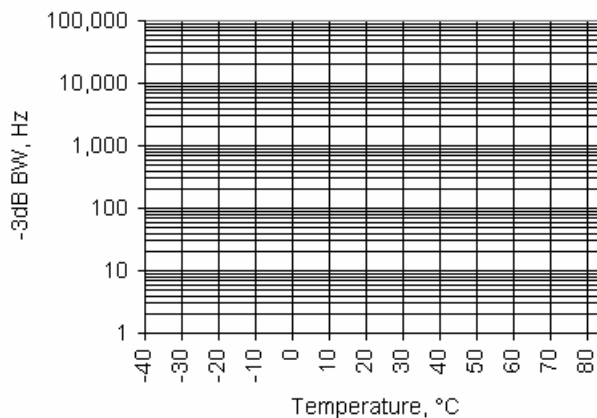
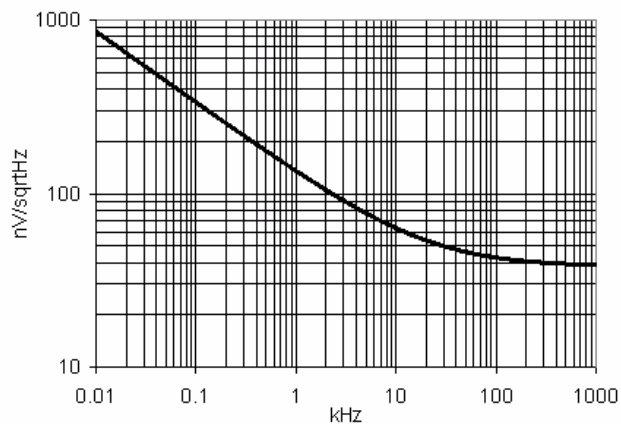


Table 11-32. VDAC AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
F _{DAC}	Update rate	1 V scale	–	–	1000	ksps
		4 V scale	–	–	250	ksps
T _{settleP}	Settling time to 0.1%, step 25% to 75%	1 V scale, Cload = 15 pF	–	0.45	1	µs
		4 V scale, Cload = 15 pF	–	0.8	3.2	µs
T _{settleN}	Settling time to 0.1%, step 75% to 25%	1 V scale, Cload = 15 pF	–	0.45	1	µs
		4 V scale, Cload = 15 pF	–	0.7	3	µs

Table 11-38. PGA AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
BW1	–3 dB bandwidth	Power mode = high, gain = 1, input = 100 mV peak-to-peak	6.7	8	–	MHz
SR1	Slew rate	Power mode = high, gain = 1, 20% to 80%	3	–	–	V/μs
e _n	Input noise density	Power mode = high, V _{dda} = 5 V, at 100 kHz	–	43	–	nV/sqrtHz

Figure 11-44. Gain vs. Frequency, at Different Gain Settings, V_{dda} = 3.3 V, Power Mode = High

Figure 11-45. Bandwidth vs. Temperature, at Different Gain Settings, Power Mode = High

Figure 11-46. Noise vs. Frequency, V_{dda} = 5 V, Power Mode = High


11.5.11 Temperature Sensor

Table 11-39. Temperature Sensor Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
	Temp sensor accuracy	Range: –40 °C to +85 °C	–	±5	–	°C

11.5.12 LCD Direct Drive

Table 11-40. LCD Direct Drive DC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
I_{CC}	LCD system operating current	Device sleep mode with wakeup at 400-Hz rate to refresh LCDs, bus clock = 3 Mhz, Vddio = Vdda = 3 V, 4 commons, 16 segments, 1/4 duty cycle, 50 Hz frame rate, no glass connected	–	38	–	μA
I_{CC_SEG}	Current per segment driver	Strong drive mode	–	260	–	μA
V_{BIAS}	LCD bias range (V_{BIAS} refers to the main output voltage(V0) of LCD DAC)	$V_{DDA} \geq 3 V$ and $V_{DDA} \geq V_{BIAS}$	2	–	5	V
	LCD bias step size	$V_{DDA} \geq 3 V$ and $V_{DDA} \geq V_{BIAS}$	–	$9.1 \times V_{DDA}$	–	mV
	LCD capacitance per segment/common driver	Drivers may be combined	–	500	5000	pF
	Long term segment offset		–	–	20	mV
I_{OUT}	Output drive current per segment driver)	Vddio = 5.5V, strong drive mode	355	–	710	μA

Table 11-41. LCD Direct Drive AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
f_{LCD}	LCD frame rate		10	50	150	Hz

11.7.4 SRAM

Table 11-61. SRAM DC Specifications

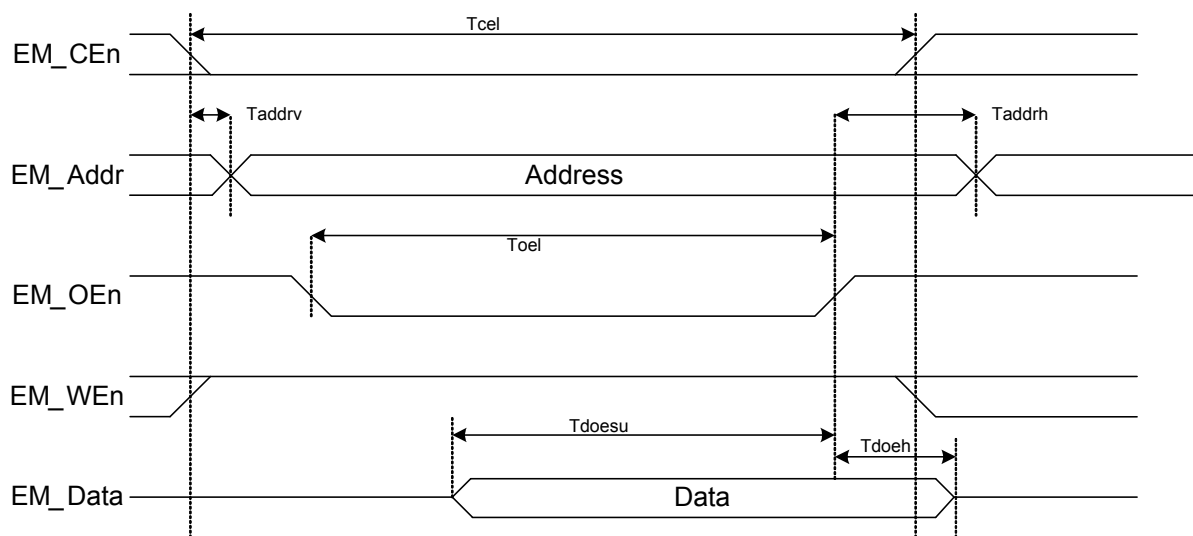
Parameter	Description	Conditions	Min	Typ	Max	Units
V _{SRAM}	SRAM retention voltage		1.2	–	–	V

Table 11-62. SRAM AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
F _{SRAM}	SRAM operating frequency		DC	–	67	MHz

11.7.5 External Memory Interface

Figure 11-48. Asynchronous Read Cycle Timing



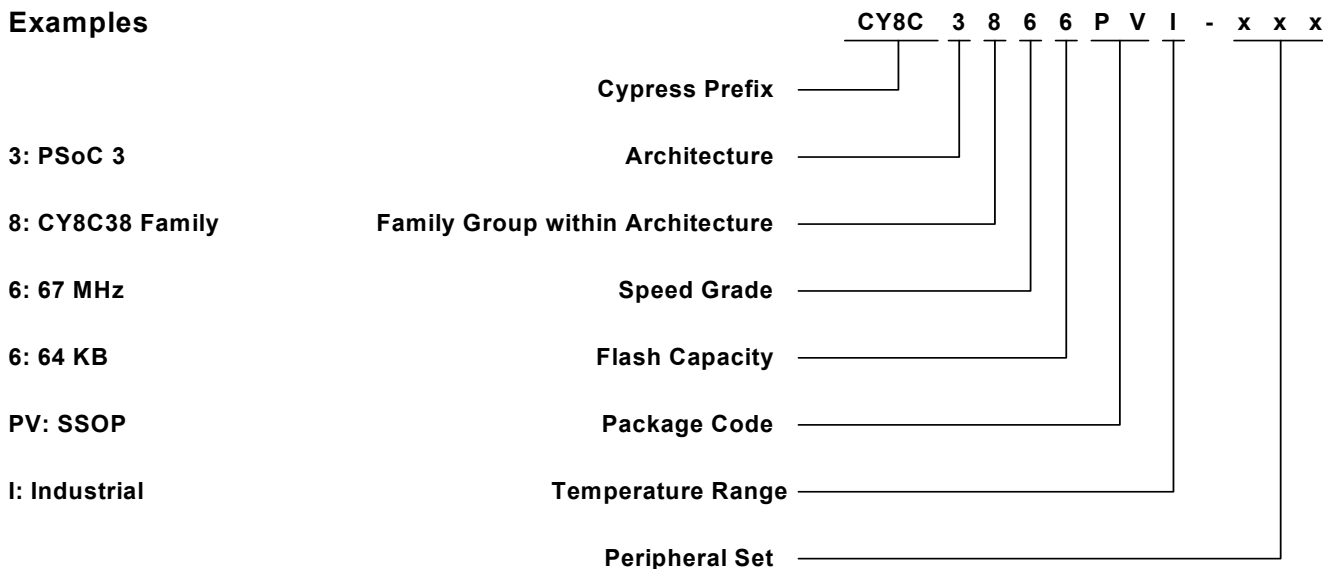
12.1 Part Numbering Conventions

PSoc 3 devices follow the part numbering convention described here. All fields are single character alphanumeric (0, 1, 2, ..., 9, A, B, ..., Z) unless stated otherwise.

CY8Cabcdefg-xxx

- a: Architecture
 - 3: PSoC 3
 - 5: PSoC 5
- b: Family group within architecture
 - 4: CY8C34 family
 - 6: CY8C36 family
 - 8: CY8C38 family
- c: Speed grade
 - 4: 48 MHz
 - 6: 67 MHz
- d: Flash capacity
 - 4: 16 KB
 - 5: 32 KB
 - 6: 64 KB
- ef: Package code
 - Two character alphanumeric
 - AX: TQFP
 - LT: QFN
 - PV: SSOP
- g: Temperature range
 - C: commercial
 - I: industrial
 - A: automotive
- xxx: Peripheral set
 - Three character numeric
 - No meaning is associated with these three characters.

Examples



All devices in the PSoC 3 CY8C38 family comply to RoHS-6 specifications, demonstrating the commitment by Cypress to lead-free products. Lead (Pb) is an alloying element in solders that has resulted in environmental concerns due to potential toxicity. Cypress uses nickel-palladium-gold (NiPdAu) technology for the majority of leadframe-based packages.

A high-level review of the Cypress Pb-free position is available on our website. Specific package information is also available. Package Material Declaration Datasheets (PMDDs) identify all substances contained within Cypress packages. PMDDs also confirm the absence of many banned substances. The information in the PMDDs will help Cypress customers plan for recycling or other “end of life” requirements.

13. Packaging

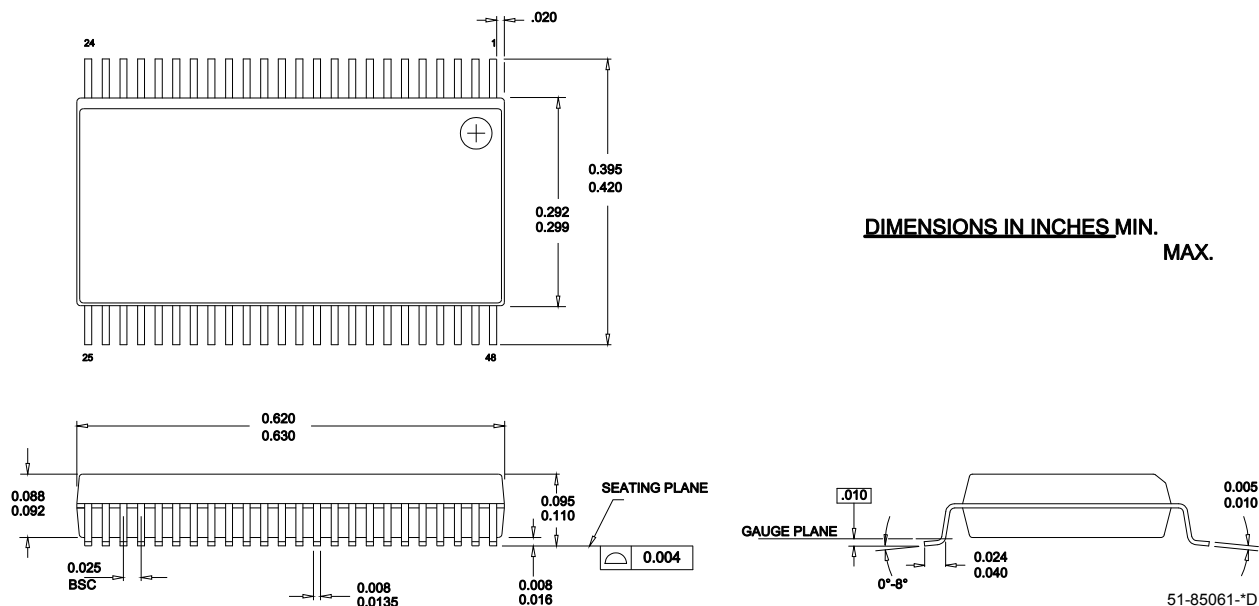
Table 13-1. Package Characteristics

Parameter	Description	Conditions	Min	Typ	Max	Units
T _A	Operating ambient temperature		-40	25.00	85	°C
T _J	Operating junction temperature		-40	–	100	°C
T _{ja}	Package θJA (48-pin SSOP)		–	45.16	–	°C/Watt
T _{ja}	Package θJA (48-pin QFN)		–	15.94	–	°C/Watt
T _{ja}	Package θJA (68-pin QFN)		–	11.72	–	°C/Watt
T _{ja}	Package θJA (100-pin TQFP)		–	30.52	–	°C/Watt
T _{jc}	Package θJC (48-pin SSOP)		–	27.84	–	°C/Watt
T _{jc}	Package θJC (48-pin QFN)		–	7.05	–	°C/Watt
T _{jc}	Package θJC (68-pin QFN)		–	6.32	–	°C/Watt
T _{jc}	Package θJC (100-pin TQFP)		–	9.04	–	°C/Watt
	Pb-free assemblies (20s to 40s) – Sn-Ag-Cu solder paste reflow temperature		235	–	245	°C
	Pb-free assemblies (20s to 40s) – Sn-Pb solder paste reflow temperature		205	–	220	°C

Table 13-2. Package Moisture Sensitivity Level (MSL), IPC/JEDEC J-STD-2

Package	MSL
48-pin SSOP	MSL 1
48-pin QFN	MSL 3
68-pin QFN	MSL 3
100-pin TQFP	MSL 3

Figure 13-1. 48-pin (300 mil) SSOP Package Outline



14. Acronyms

Table 14-1. Acronyms Used in this Document

Acronym	Description
abus	analog local bus
ADC	analog-to-digital converter
AG	analog global
AHB	AMBA (advanced microcontroller bus architecture) high-performance bus, an ARM data transfer bus
ALU	arithmetic logic unit
AMUXBUS	analog multiplexer bus
API	application programming interface
APSR	application program status register
ARM®	advanced RISC machine, a CPU architecture
ATM	automatic thump mode
BW	bandwidth
CAN	Controller Area Network, a communications protocol
CMRR	common-mode rejection ratio
CPU	central processing unit
CRC	cyclic redundancy check, an error-checking protocol
DAC	digital-to-analog converter, see also IDAC, VDAC
DFB	digital filter block
DIO	digital input/output, GPIO with only digital capabilities, no analog. See GPIO.
DMA	direct memory access, see also TD
DNL	differential nonlinearity, see also INL
DNU	do not use
DR	port write data registers
DRES	digital logic reset
DSI	digital system interconnect
DWT	data watchpoint and trace
ECC	error correcting code
ECO	external crystal oscillator
EEPROM	electrically erasable programmable read-only memory
EMI	electromagnetic interference
EMIF	external memory interface
EOC	end of conversion
EOF	end of frame
EPSR	execution program status register
ESD	electrostatic discharge

Table 14-1. Acronyms Used in this Document *(continued)*

Acronym	Description
ETM	embedded trace macrocell
FIR	finite impulse response, see also IIR
FPB	flash patch and breakpoint
FS	full-speed
GPIO	general-purpose input/output, applies to a PSoC pin
HVI	high-voltage interrupt, see also LVI, LVD
IC	integrated circuit
IDAC	current DAC, see also DAC, VDAC
IDE	integrated development environment
I ² C, or IIC	Inter-Integrated Circuit, a communications protocol
IIR	infinite impulse response, see also FIR
ILO	internal low-speed oscillator, see also IMO
IMO	internal main oscillator, see also ILO
INL	integral nonlinearity, see also DNL
I/O	input/output, see also GPIO, DIO, SIO, USBIO
IPOR	initial power-on reset
IPSR	interrupt program status register
IRQ	interrupt request
ITM	instrumentation trace macrocell
LCD	liquid crystal display
LIN	Local Interconnect Network, a communications protocol.
LR	link register
LUT	lookup table
LVD	low-voltage detect, see also LVI
LVI	low-voltage interrupt, see also HVI
LVTTL	low-voltage transistor-transistor logic
MAC	multiply-accumulate
MCU	microcontroller unit
MISO	master-in slave-out
NC	no connect
NMI	nonmaskable interrupt
NRZ	non-return-to-zero
NVIC	nested vectored interrupt controller
NVL	nonvolatile latch, see also WOL
opamp	operational amplifier
PAL	programmable array logic, see also PLD
PC	program counter
PCB	printed circuit board

Table 14-1. Acronyms Used in this Document *(continued)*

Acronym	Description
PGA	programmable gain amplifier
PHUB	peripheral hub
PHY	physical layer
PICU	port interrupt control unit
PLA	programmable logic array
PLD	programmable logic device, see also PAL
PLL	phase-locked loop
PMDD	package material declaration datasheet
POR	power-on reset
PRES	precise power-on reset
PRS	pseudo random sequence
PS	port read data register
PSoC®	Programmable System-on-Chip™
PSRR	power supply rejection ratio
PWM	pulse-width modulator
RAM	random-access memory
RISC	reduced-instruction-set computing
RMS	root-mean-square
RTC	real-time clock
RTL	register transfer language
RTR	remote transmission request
RX	receive
SAR	successive approximation register
SC/CT	switched capacitor/continuous time
SCL	I ² C serial clock
SDA	I ² C serial data
S/H	sample and hold
SINAD	signal to noise and distortion ratio
SIO	special input/output, GPIO with advanced features. See GPIO.
SOC	start of conversion
SOF	start of frame
SPI	Serial Peripheral Interface, a communications protocol
SR	slew rate
SRAM	static random access memory
SRES	software reset
SWD	serial wire debug, a test protocol
SWV	single-wire viewer
TD	transaction descriptor, see also DMA

Table 14-1. Acronyms Used in this Document *(continued)*

Acronym	Description
THD	total harmonic distortion
TIA	transimpedance amplifier
TRM	technical reference manual
TTL	transistor-transistor logic
TX	transmit
UART	Universal Asynchronous Transmitter Receiver, a communications protocol
UDB	universal digital block
USB	Universal Serial Bus
USBIO	USB input/output, PSoC pins used to connect to a USB port
VDAC	voltage DAC, see also DAC, IDAC
WDT	watchdog timer
WOL	write once latch, see also NVL
WRES	watchdog timer reset
XRES	external reset I/O pin
XTAL	crystal

15. Reference Documents

[PSoC® 3, PSoC® 5 Architecture TRM](#)

[PSoC® 3 Registers TRM](#)

16. Document Conventions

16.1 Units of Measure

Table 16-1. Units of Measure

Symbol	Unit of Measure
°C	degrees Celsius
dB	decibels
fF	femtofarads
Hz	hertz
KB	1024 bytes
kbps	kilobits per second
Khr	kilohours
kHz	kilohertz
kΩ	kilohms
ksps	kilosamples per second
LSB	least significant bit
Mbps	megabits per second
MHz	megahertz
MΩ	megaohms
Msp	megasamples per second

Description Title: PSoC® 3: CY8C38 Family Datasheet Programmable System-on-Chip (PSoC®) Document Number: 001-11729				
*Q	3179219	02/22/2011	MKEA	Updated conditions for flash data retention time Updated 100-pin TQFP package spec. Updated EEPROM AC specifications.

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