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Details

Product Status	Active
Core Processor	ARM® Cortex®-M4
Core Size	32-Bit Single-Core
Speed	100MHz
Connectivity	I ² C, IrDA, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, DMA, POR, PWM, WDT
Number of I/O	47
Program Memory Size	1MB (1M x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	80K x 8
Voltage - Supply (Vcc/Vdd)	1.62V ~ 3.6V
Data Converters	A/D 11x10b; D/A 1x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	64-VFQFN Exposed Pad
Supplier Device Package	64-QFN (9x9)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/atsam4n16ba-mu

11.4.1.17 Exceptions and Interrupts

The Cortex-M4 processor supports interrupts and system exceptions. The processor and the *Nested Vectored Interrupt Controller* (NVIC) prioritize and handle all exceptions. An exception changes the normal flow of software control. The processor uses the Handler mode to handle all exceptions except for reset. See “Exception Entry” and “Exception Return” for more information.

The NVIC registers control interrupt handling. See “Nested Vectored Interrupt Controller (NVIC)” for more information.

11.4.1.18 Data Types

The processor supports the following data types:

- 32-bit words
- 16-bit halfwords
- 8-bit bytes
- The processor manages all data memory accesses as little-endian. Instruction memory and *Private Peripheral Bus* (PPB) accesses are always little-endian. See “Memory Regions, Types and Attributes” for more information.

11.4.1.19 Cortex Microcontroller Software Interface Standard (CMSIS)

For a Cortex-M4 microcontroller system, the *Cortex Microcontroller Software Interface Standard* (CMSIS) defines:

- A common way to:
 - Access peripheral registers
 - Define exception vectors
- The names of:
 - The registers of the core peripherals
 - The core exception vectors
- A device-independent interface for RTOS kernels, including a debug channel.

The CMSIS includes address definitions and data structures for the core peripherals in the Cortex-M4 processor.

The CMSIS simplifies the software development by enabling the reuse of template code and the combination of CMSIS-compliant software components from various middleware vendors. Software vendors can expand the CMSIS to include their peripheral definitions and access functions for those peripherals.

This document includes the register names defined by the CMSIS, and gives short descriptions of the CMSIS functions that address the processor core and the core peripherals.

Note: This document uses the register short names defined by the CMSIS. In a few cases, these differ from the architectural short names that might be used in other documents.

The following sections give more information about the CMSIS:

- Section 11.5.3 “Power Management Programming Hints”
- Section 11.6.2 “CMSIS Functions”
- Section 11.8.2.1 “NVIC Programming Hints”.

11.6.4.6 LDM and STM

Load and Store Multiple registers.

Syntax

op{*addr_mode*}{*cond*} *Rn*{!}, *reglist*

where:

op is one of:

LDM Load Multiple registers.

STM Store Multiple registers.

addr_mode is any one of the following:

IA Increment address After each access. This is the default.

DB Decrement address Before each access.

cond is an optional condition code, see “Conditional Execution” .

Rn is the register on which the memory addresses are based.

! is an optional writeback suffix.

If ! is present, the final address, that is loaded from or stored to, is written back into *Rn*.

reglist is a list of one or more registers to be loaded or stored, enclosed in braces. It can contain register ranges. It must be comma separated if it contains more than one register or register range, see “Examples” .

LDM and LDMFD are synonyms for LDMIA. LDMFD refers to its use for popping data from Full Descending stacks.

LDMEA is a synonym for LDMDB, and refers to its use for popping data from Empty Ascending stacks.

STM and STMEA are synonyms for STMIA. STMEA refers to its use for pushing data onto Empty Ascending stacks.

STMFD is a synonym for STMDB, and refers to its use for pushing data onto Full Descending stacks

Operation

LDM instructions load the registers in *reglist* with word values from memory addresses based on *Rn*.

STM instructions store the word values in the registers in *reglist* to memory addresses based on *Rn*.

For LDM, LDMIA, LDMFD, STM, STMIA, and STMEA the memory addresses used for the accesses are at 4-byte intervals ranging from *Rn* to *Rn* + 4 * (*n*-1), where *n* is the number of registers in *reglist*. The accesses happen in order of increasing register numbers, with the lowest numbered register using the lowest memory address and the highest number register using the highest memory address. If the writeback suffix is specified, the value of *Rn* + 4 * (*n*-1) is written back to *Rn*.

For LDMDB, LDMEA, STMDB, and STMFD the memory addresses used for the accesses are at 4-byte intervals ranging from *Rn* to *Rn* - 4 * (*n*-1), where *n* is the number of registers in *reglist*. The accesses happen in order of decreasing register numbers, with the highest numbered register using the highest memory address and the lowest number register using the lowest memory address. If the writeback suffix is specified, the value of *Rn* - 4 * (*n*-1) is written back to *Rn*.

The PUSH and POP instructions can be expressed in this form. See “PUSH and POP” for details.

Restrictions

In these instructions:

- *Rn* must not be PC
- *reglist* must not contain SP
- In any STM instruction, *reglist* must not contain PC

11.6.4.9 CLREX

Clear Exclusive.

Syntax

```
CLREX{cond}
```

where:

cond is an optional condition code, see “Conditional Execution” .

Operation

Use CLREX to make the next STREX, STREXB, or STREXH instruction write 1 to its destination register and fail to perform the store. It is useful in exception handler code to force the failure of the store exclusive if the exception occurs between a load exclusive instruction and the matching store exclusive instruction in a synchronization operation.

See “Synchronization Primitives” for more information.

Condition Flags

These instructions do not change the flags.

Examples

```
CLREX
```

11.6.11.1 BKPT

Breakpoint.

Syntax

```
BKPT #imm
```

where:

imm is an expression evaluating to an integer in the range 0-255 (8-bit value).

Operation

The BKPT instruction causes the processor to enter Debug state. Debug tools can use this to investigate system state when the instruction at a particular address is reached.

imm is ignored by the processor. If required, a debugger can use it to store additional information about the breakpoint.

The BKPT instruction can be placed inside an IT block, but it executes unconditionally, unaffected by the condition specified by the IT instruction.

Condition Flags

This instruction does not change the flags.

Examples

```
BKPT 0xAB ; Breakpoint with immediate value set to 0xAB (debugger can  
          ; extract the immediate value by locating it using the PC)
```

Note: ARM does not recommend the use of the BKPT instruction with an immediate value set to 0xAB for any purpose other than Semi-hosting.

11.8 Nested Vectored Interrupt Controller (NVIC)

This section describes the NVIC and the registers it uses. The NVIC supports:

- 1 to 30 interrupts.
- A programmable priority level of 0-15 for each interrupt. A higher level corresponds to a lower priority, so level 0 is the highest interrupt priority.
- Level detection of interrupt signals.
- Dynamic reprioritization of interrupts.
- Grouping of priority values into group priority and subpriority fields.
- Interrupt tail-chaining.
- An external *Non-maskable interrupt* (NMI)

The processor automatically stacks its state on exception entry and unstacks this state on exception exit, with no instruction overhead. This provides low latency exception handling.

11.8.1 Level-sensitive Interrupts

The processor supports level-sensitive interrupts. A level-sensitive interrupt is held asserted until the peripheral deasserts the interrupt signal. Typically, this happens because the ISR accesses the peripheral, causing it to clear the interrupt request.

When the processor enters the ISR, it automatically removes the pending state from the interrupt (see “Hardware and Software Control of Interrupts”). For a level-sensitive interrupt, if the signal is not deasserted before the processor returns from the ISR, the interrupt becomes pending again, and the processor must execute its ISR again. This means that the peripheral can hold the interrupt signal asserted until it no longer requires servicing.

11.8.1.1 Hardware and Software Control of Interrupts

The Cortex-M4 latches all interrupts. A peripheral interrupt becomes pending for one of the following reasons:

- The NVIC detects that the interrupt signal is HIGH and the interrupt is not active
- The NVIC detects a rising edge on the interrupt signal
- A software writes to the corresponding interrupt set-pending register bit, see “Interrupt Set-pending Registers” , or to the NVIC_STIR register to make an interrupt pending, see “Software Trigger Interrupt Register” .

A pending interrupt remains pending until one of the following:

- The processor enters the ISR for the interrupt. This changes the state of the interrupt from pending to active. Then:
 - For a level-sensitive interrupt, when the processor returns from the ISR, the NVIC samples the interrupt signal. If the signal is asserted, the state of the interrupt changes to pending, which might cause the processor to immediately re-enter the ISR. Otherwise, the state of the interrupt changes to inactive.
- Software writes to the corresponding interrupt clear-pending register bit.
For a level-sensitive interrupt, if the interrupt signal is still asserted, the state of the interrupt does not change. Otherwise, the state of the interrupt changes to inactive.

11.8.2 NVIC Design Hints and Tips

Ensure that the software uses correctly aligned register accesses. The processor does not support unaligned accesses to NVIC registers. See the individual register descriptions for the supported access sizes.

An interrupt can enter a pending state even if it is disabled. Disabling an interrupt only prevents the processor from taking that interrupt.

- **SIZE: Size of the MPU Protection Region**

The minimum permitted value is 3 (b00010).

The SIZE field defines the size of the MPU memory region specified by the MPU_RNR. as follows:

$$(\text{Region size in bytes}) = 2^{(\text{SIZE}+1)}$$

The smallest permitted region size is 32B, corresponding to a SIZE value of 4. The table below gives an example of SIZE values, with the corresponding region size and value of N in the MPU_RBAR.

SIZE Value	Region Size	Value of N ⁽¹⁾	Note
b00100 (4)	32 B	5	Minimum permitted size
b01001 (9)	1 KB	10	-
b10011 (19)	1 MB	20	-
b11101 (29)	1 GB	30	-
b11111 (31)	4 GB	b01100	Maximum possible size

Note: 1. In the MPU_RBAR, see “MPU Region Base Address Register”

- **ENABLE: Region Enable**

Note: For information about access permission, see “MPU Access Permission Attributes” .

12.5.6.2 Asynchronous Mode

The TPIU is configured in asynchronous mode, trace data are output using the single TRACESWO pin. The TRACESWO signal is multiplexed with the TDO signal of the JTAG Debug Port. As a consequence, asynchronous trace mode is only available when the serial wire debug mode is selected since TDO signal is used in JTAG debug mode.

Two encoding formats are available for the single pin output:

- Manchester encoded stream. This is the reset value.
- NRZ-based UART byte structure

12.5.6.3 How to Configure the TPIU

This example only concerns the asynchronous trace mode.

- Set the TRCENA bit to 1 into the Debug Exception and Monitor Register (0xE00EDFC) to enable the use of trace and debug blocks.
- Write 0x2 into the Selected Pin Protocol Register.
 - Select the Serial Wire Output – NRZ.
- Write 0x100 into the Formatter and Flush Control Register.
- Set the suitable clock prescaler value into the Async Clock Prescaler Register to scale the baud rate of the asynchronous output (this can be done automatically by the debugging tool).

12.5.7 IEEE 1149.1 JTAG Boundary Scan

IEEE 1149.1 JTAG Boundary Scan allows pin-level access independent of the device packaging technology.

IEEE 1149.1 JTAG Boundary Scan is enabled when TST is tied to low, while JTAGSEL is high during power-up and must be kept in this state during the whole boundary scan operation. The SAMPLE, EXTEST and BYPASS functions are implemented. In SWD/JTAG debug mode, the ARM processor responds with a non-JTAG chip ID that identifies the processor. This is not IEEE 1149.1 JTAG-compliant.

It is not possible to switch directly between JTAG Boundary Scan and SWJ Debug Port operations. A chip reset must be performed after JTAGSEL is changed. A Boundary-scan Descriptor Language (BSDL) file to set up the test is provided on www.atmel.com.

12.5.7.1 JTAG Boundary-scan Register

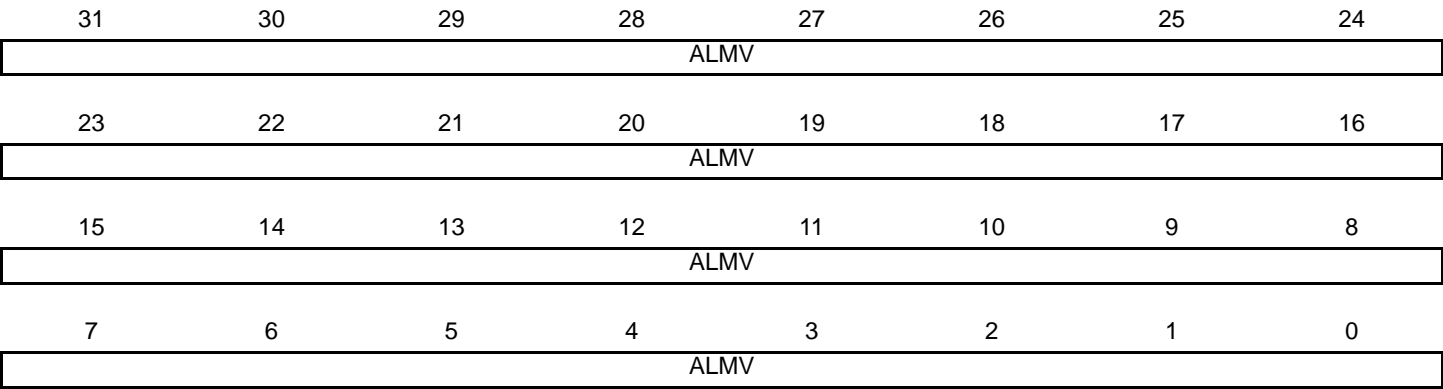
The Boundary-scan Register (BSR) contains a number of bits which corresponds to active pins and associated control signals.

Each SAM4 input/output pin corresponds to a 3-bit register in the BSR. The OUTPUT bit contains data that can be forced on the pad. The INPUT bit facilitates the observability of data applied to the pad. The CONTROL bit selects the direction of the pad.

For more information, please refer to BSDL files available for the SAM4 Series.

14.5.2 Real-time Timer Alarm Register

Name: RTT_AR
Address: 0x400E1434
Access: Read-write



- **ALMV: Alarm Value**
Defines the alarm value (ALMV+1) compared with the Real-time Timer.
Note: The alarm interrupt must be disabled (ALMIEN must be cleared in RTT_MR register) when writing a new ALMV value.

The following checks are performed:

1. Century (check if it is in range 19 - 20 or 13-14 in Persian mode)
2. Year (BCD entry check)
3. Date (check range 01 - 31)
4. Month (check if it is in BCD range 01 - 12, check validity regarding "date")
5. Day (check range 1 - 7)
6. Hour (BCD checks: in 24-hour mode, check range 00 - 23 and check that AM/PM flag is not set if RTC is set in 24-hour mode; in 12-hour mode check range 01 - 12)
7. Minute (check BCD and range 00 - 59)
8. Second (check BCD and range 00 - 59)

Note: If the 12-hour mode is selected by means of the RTC_MR register, a 12-hour value can be programmed and the returned value on RTC_TIMR will be the corresponding 24-hour value. The entry control checks the value of the AM/PM indicator (bit 22 of RTC_TIMR register) to determine the range to be checked.

15.5.5 RTC Internal Free Running Counter Error Checking

To improve the reliability and security of the RTC, a permanent check is performed on the internal free running counters to report non-BCD or invalid date/time values.

An error is reported by TDERR bit in the status register (RTC_SR) if an incorrect value has been detected. The flag can be cleared by programming the TDERRCLR in the RTC status clear control register (RTC_SCCR).

Anyway the TDERR error flag will be set again if the source of the error has not been cleared before clearing the TDERR flag. The clearing of the source of such error can be done either by reprogramming a correct value on RTC_CALR and/or RTC_TIMR registers.

The RTC internal free running counters may automatically clear the source of TDERR due to their roll-over (i.e. every 10 seconds for SECONDS[3:0] bitfield in RTC_TIMR register). In this case the TDERR is held high until a clear command is asserted by TDERRCLR bit in RTC_SCCR register.

15.5.6 Updating Time/Calendar

To update any of the time/calendar fields, the user must first stop the RTC by setting the corresponding field in the Control Register. Bit UPDTIM must be set to update time fields (hour, minute, second) and bit UPDCAL must be set to update calendar fields (century, year, month, date, day).

Then the user must poll or wait for the interrupt (if enabled) of bit ACKUPD in the Status Register. Once the bit reads 1, it is mandatory to clear this flag by writing the corresponding bit in RTC_SCCR. The user can now write to the appropriate Time and Calendar register.

Once the update is finished, the user must reset (0) UPDTIM and/or UPDCAL in the Control

When entering programming mode of the calendar fields, the time fields remain enabled. When entering the programming mode of the time fields, both time and calendar fields are stopped. This is due to the location of the calendar logic circuitry (downstream for low-power considerations). It is highly recommended to prepare all the fields to be updated before entering programming mode. In successive update operations, the user must wait at least one second after resetting the UPDTIM/UPDCAL bit in the RTC_CR (Control Register) before setting these bits again. This is done by waiting for the SEC flag in the Status Register before setting UPDTIM/UPDCAL bit. After resetting UPDTIM/UPDCAL, the SEC flag must also be cleared.

18.3 General Purpose Backup Registers (GPBR) User Interface

Table 18-1. Register Mapping

Offset	Register	Name	Access	Reset
0x0	General Purpose Backup Register 0	SYS_GPBR0	Read-write	—
...
0x1C	General Purpose Backup Register 7	SYS_GPBR7	Read-write	—

forces MAINCK to be the source clock for the master clock (MCK). Then, regardless of the PMC configuration, a clock failure detection automatically forces the 12/8/4 MHz Fast RC Oscillator to be the source clock for MAINCK. If the Fast RC Oscillator is disabled when a clock failure detection occurs, it is automatically re-enabled by the clock failure detection mechanism.

It takes 2 slow clock RC oscillator cycles to detect and switch from the 3 to 20 MHz Crystal, or Ceramic Resonator-based Oscillator, to the 12/8/4 MHz Fast RC Oscillator if the Master Clock source is Main Clock, or 3 slow clock RC oscillator cycles if the Master Clock source is PLLACK .

A clock failure detection activates a fault output that is connected to the Pulse Width Modulator (PWM) Controller. With this connection, the PWM controller is able to force its outputs and to protect the driven device, if a clock failure is detected. This fault output remains active until the defect is detected and until it is cleared by the bit FOCLR in the PMC Fault Output Clear Register (PMC_FOCR).

The user can know the status of the fault output at any time by reading the FOS bit in the PMC_SR register.

25.12 Slow Crystal Clock Frequency Monitor

The frequency of the slow clock crystal oscillator can be monitored by means of logic driven by the main RC oscillator known as a reliable clock source. This function is enabled by configuring the XT32KFME bit of the Main Oscillator Register (CKGR_MOR).

An error flag (XT32KERR in PMC_SR) is asserted when the slow clock crystal oscillator frequency is out of the +/-10% nominal frequency value (i.e. 32768 kHz). The error flag can be cleared only if the slow clock frequency monitoring is disabled.

When the main RC oscillator frequency is 4 MHz, the accuracy of the measurement is +/-40% as this frequency is not trimmed during production. Therefore, +/-10% accuracy is obtained only if the RC oscillator frequency is configured for 8 or 12 MHz.

The monitored clock frequency is declared invalid if at least 4 consecutive clock period measurement results are over the nominal period +/-10%.

Due to the possible frequency variation of the embedded main RC oscillator acting as reference clock for the monitor logic, any slow clock crystal frequency deviation over +/-10% of the nominal frequency is systematically reported as an error by means of XT32KERR in PMC_SR. Between -1% and -10% and +1% and +10%, the error is not systematically reported.

Thus only a crystal running at 32768 kHz frequency ensures that the error flag will not be asserted. The permitted drift of the crystal is 10000ppm (1%), which allows any standard crystal to be used.

If the main RC frequency needs to be changed while the slow clock frequency monitor is operating, the monitoring must be stopped prior to change the main RC frequency. Then it can be re-enabled as soon as MOSCRCS is set in PMC_SR register.

The error flag can be defined as an interrupt source of the PMC by setting the XT32KERR bit of PMC_IER.

25.13 Programming Sequence

1. Enabling the Main Oscillator:

The main oscillator is enabled by setting the MOSCXTEN field in the Main Oscillator Register (CKGR_MOR). The user can define a start-up time. This can be achieved by writing a value in the MOSCXTST field in CKGR_MOR. Once this register has been correctly configured, the user must wait for MOSCXTS field in the PMC_SR register to be set. This can be done either by polling the status register, or by waiting the interrupt line to be raised if the associated interrupt to MOSCXTS has been enabled in the PMC_IER register.

Start Up Time = $8 * \text{MOSCXTST} / \text{SLCK} = 56$ Slow Clock Cycles.

The main oscillator will be enabled (MOSCXTS bit set) after 56 Slow Clock Cycles.

27.7 Parallel Input/Output Controller (PIO) User Interface

Each I/O line controlled by the PIO Controller is associated with a bit in each of the PIO Controller User Interface registers. Each register is 32 bits wide. If a parallel I/O line is not defined, writing to the corresponding bits has no effect. Undefined bits read zero. If the I/O line is not multiplexed with any peripheral, the I/O line is controlled by the PIO Controller and PIO_PSR returns 1 systematically.

Table 27-2. Register Mapping

Offset	Register	Name	Access	Reset
0x0000	PIO Enable Register	PIO_PER	Write-only	–
0x0004	PIO Disable Register	PIO_PDR	Write-only	–
0x0008	PIO Status Register	PIO_PSR	Read-only	(1)
0x000C	Reserved	–	–	–
0x0010	Output Enable Register	PIO_OER	Write-only	–
0x0014	Output Disable Register	PIO_ODR	Write-only	–
0x0018	Output Status Register	PIO_OSR	Read-only	0x0000 0000
0x001C	Reserved	–	–	–
0x0020	Glitch Input Filter Enable Register	PIO_IFER	Write-only	–
0x0024	Glitch Input Filter Disable Register	PIO_IFDR	Write-only	–
0x0028	Glitch Input Filter Status Register	PIO_IFSR	Read-only	0x0000 0000
0x002C	Reserved	–	–	–
0x0030	Set Output Data Register	PIO_SODR	Write-only	–
0x0034	Clear Output Data Register	PIO_CODR	Write-only	–
0x0038	Output Data Status Register	PIO_ODSR	Read-only or ⁽²⁾ Read-write	–
0x003C	Pin Data Status Register	PIO_PDSR	Read-only	(3)
0x0040	Interrupt Enable Register	PIO_IER	Write-only	–
0x0044	Interrupt Disable Register	PIO_IDR	Write-only	–
0x0048	Interrupt Mask Register	PIO_IMR	Read-only	0x00000000
0x004C	Interrupt Status Register ⁽⁴⁾	PIO_ISR	Read-only	0x00000000
0x0050	Multi-driver Enable Register	PIO_MDER	Write-only	–
0x0054	Multi-driver Disable Register	PIO_MDDR	Write-only	–
0x0058	Multi-driver Status Register	PIO_MDSR	Read-only	0x00000000
0x005C	Reserved	–	–	–
0x0060	Pull-up Disable Register	PIO_PUDR	Write-only	–
0x0064	Pull-up Enable Register	PIO_PUER	Write-only	–
0x0068	Pad Pull-up Status Register	PIO_PUSR	Read-only	(1)
0x006C	Reserved	–	–	–

27.7.28 PIO Input Filter Slow Clock Status Register

Name: PIO_IFSCSR

Address: 0x400E0E88 (PIOA), 0x400E1088 (PIOB), 0x400E1288 (PIOC)

Access: Read-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Glitch or Debouncing Filter Selection Status**

0: The Glitch Filter is able to filter glitches with a duration < Tmck2.

1: The Debouncing Filter is able to filter pulses with a duration < Tdiv_slclk/2.

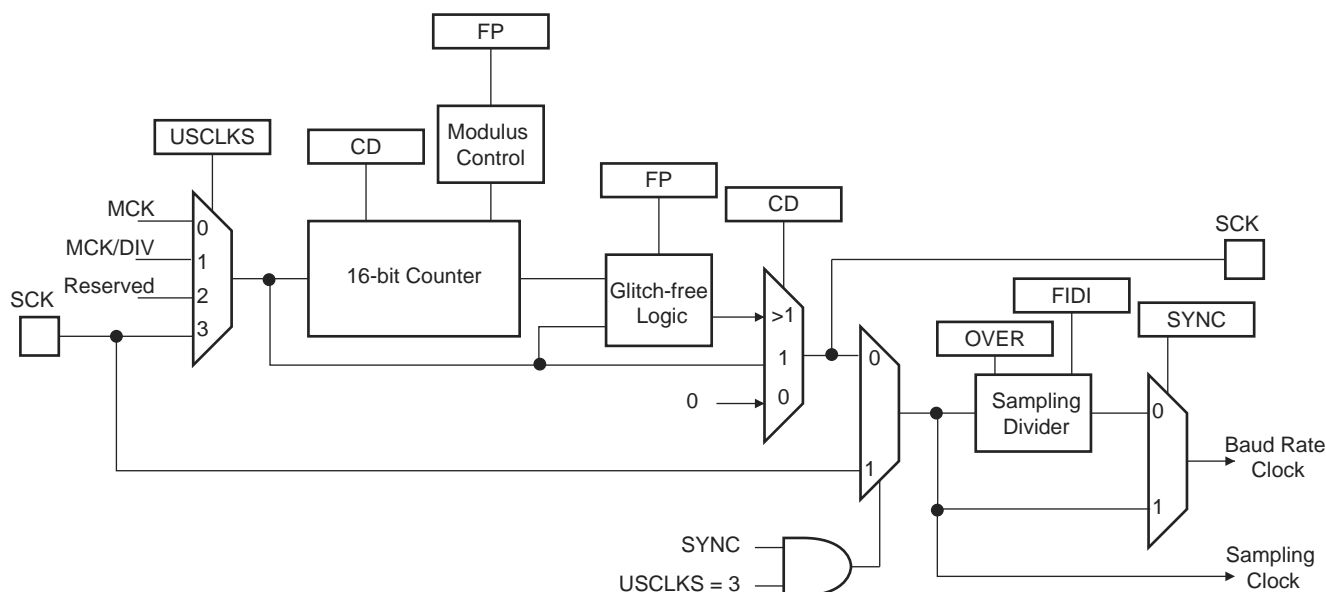
31.7.1.2 Fractional Baud Rate in Asynchronous Mode

The Baud Rate generator previously defined is subject to the following limitation: the output frequency changes by only integer multiples of the reference frequency. An approach to this problem is to integrate a fractional N clock generator that has a high resolution. The generator architecture is modified to obtain Baud Rate changes by a fraction of the reference source clock. This fractional part is programmed with the FP field in the Baud Rate Generator Register (US_BRGR). If FP is not 0, the fractional part is activated. The resolution is one eighth of the clock divider. This feature is only available when using USART normal mode. The fractional Baud Rate is calculated using the following formula:

$$Baudrate = \frac{SelectedClock}{\left(8(2 - Over)\left(CD + \frac{FP}{8}\right)\right)}$$

The modified architecture is presented below:

Figure 31-4. Fractional Baud Rate Generator



31.7.1.3 Baud Rate in Synchronous Mode or SPI Mode

If the USART is programmed to operate in synchronous mode, the selected clock is simply divided by the field CD in US_BRGR.

$$BaudRate = \frac{SelectedClock}{CD}$$

In synchronous mode, if the external clock is selected (USCLKS = 3), the clock is provided directly by the signal on the USART SCK pin. No division is active. The value written in US_BRGR has no effect. The external clock frequency must be at least 3 times lower than the system clock. In synchronous mode master (USCLKS = 0 or 1, CLK0 set to 1), the receive part limits the SCK maximum frequency to MCK/3 in USART mode, or MCK/6 in SPI mode.

When either the external clock SCK or the internal clock divided (MCK/DIV) is selected, the value programmed in CD must be even if the user has to ensure a 50:50 mark/space ratio on the SCK pin. If the internal clock MCK is selected, the Baud Rate Generator ensures a 50:50 duty cycle on the SCK pin, even if the value programmed in CD is odd.

31.8.1 USART Control Register

Name: US_CR

Address: 0x40024000 (0), 0x40028000 (1), 0x4002C000 (2)

Access: Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	RTSDIS	RTSEN	–	–
15	14	13	12	11	10	9	8
RETTO	RSTNACK	RSTIT	SENDA	STTTO	STPBRK	STTBRK	RSTSTA
7	6	5	4	3	2	1	0
TXDIS	TXEN	RXDIS	RXEN	RSTTX	RSTRX	–	–

For SPI control, see “USART Control Register (SPI_MODE)” on page 631.

- **RSTRX: Reset Receiver**

0: No effect.

1: Resets the receiver.

- **RSTTX: Reset Transmitter**

0: No effect.

1: Resets the transmitter.

- **RXEN: Receiver Enable**

0: No effect.

1: Enables the receiver, if RXDIS is 0.

- **RXDIS: Receiver Disable**

0: No effect.

1: Disables the receiver.

- **TXEN: Transmitter Enable**

0: No effect.

1: Enables the transmitter if TXDIS is 0.

- **TXDIS: Transmitter Disable**

0: No effect.

1: Disables the transmitter.

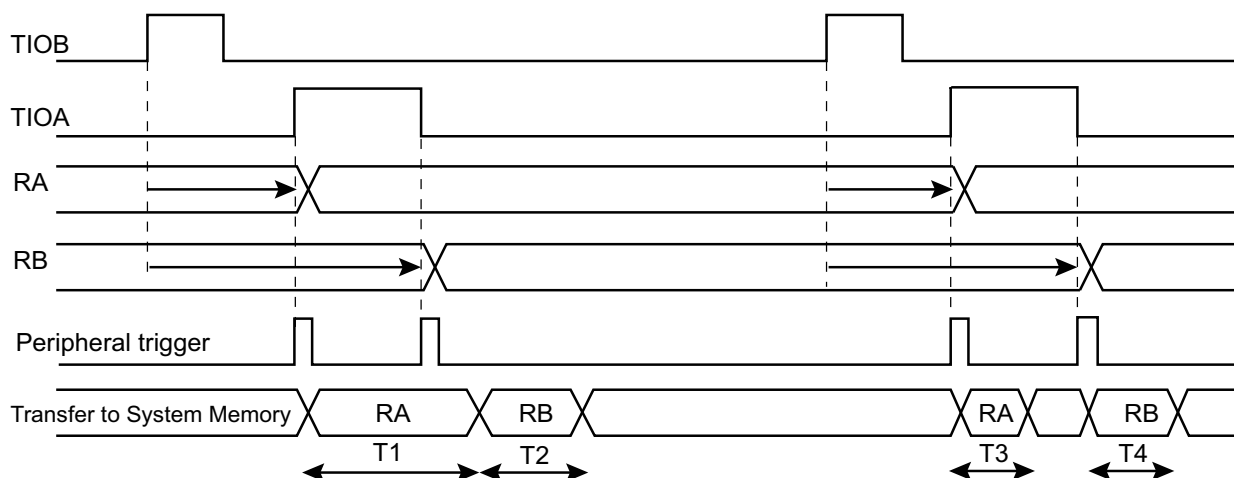
- **RSTSTA: Reset Status Bits**

0: No effect.

1: Resets the status bits PARE, FRAME, OVRE and RXBRK in US_CSR.

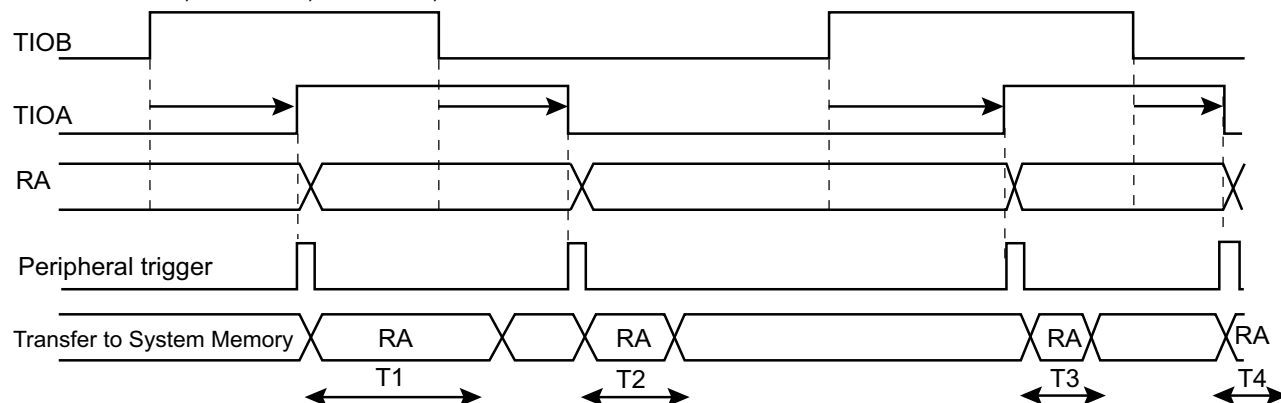
Figure 32-5. Example of Transfer with PDC

ETRGEDG = 1, LDRA = 1, LDRB = 2, ABETRG = 0



T1, T2, T3, T4 = System Bus load dependent ($t_{\min} = 8$ peripheral clocks)

ETRGEDG = 3, LDRA = 3, LDRB = 0, ABETRG = 0



T1, T2, T3, T4 = System Bus load dependent ($t_{\min} = 8$ peripheral clocks)

32.6.10 Trigger Conditions

In addition to the SYNC signal, the software trigger and the RC compare trigger, an external trigger can be defined. The ABETRG bit in the TC_CMR selects TIOA or TIOB input signal as an external trigger. The External Trigger Edge Selection parameter (ETRGEDG field in TC_CMR) defines the edge (rising, falling, or both) detected to generate an external trigger. If ETRGEDG = 0 (none), the external trigger is disabled.

32.7.12 TC Interrupt Mask Register

Name: TC_IMRx [x=0..2]

Address: 0x4001002C (0)[0], 0x4001006C (0)[1], 0x400100AC (0)[2], 0x4001402C (1)[0], 0x4001406C (1)[1], 0x400140AC (1)[2]

Access: Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	RXBUFF	ENDRX
7	6	5	4	3	2	1	0
ETRGS	LDRBS	LDRAS	CPCS	CPBS	CPAS	LOVRS	COVFS

- **COVFS: Counter Overflow**

0: The Counter Overflow Interrupt is disabled.

1: The Counter Overflow Interrupt is enabled.

- **LOVRS: Load Overrun**

0: The Load Overrun Interrupt is disabled.

1: The Load Overrun Interrupt is enabled.

- **CPAS: RA Compare**

0: The RA Compare Interrupt is disabled.

1: The RA Compare Interrupt is enabled.

- **CPBS: RB Compare**

0: The RB Compare Interrupt is disabled.

1: The RB Compare Interrupt is enabled.

- **CPCS: RC Compare**

0: The RC Compare Interrupt is disabled.

1: The RC Compare Interrupt is enabled.

- **LDRAS: RA Loading**

0: The Load RA Interrupt is disabled.

1: The Load RA Interrupt is enabled.

- **LDRBS: RB Loading**

0: The Load RB Interrupt is disabled.

1: The Load RB Interrupt is enabled.

32.7.18 TC QDEC Interrupt Status Register

Name: TC_QISR

Address: 0x400100D4 (0), 0x400140D4 (1)

Access: Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	DIR
7	6	5	4	3	2	1	0
–	–	–	–	–	QERR	DIRCHG	IDX

- **IDX: Index**

0: No Index input change since the last read of TC_QISR.

1: The IDX input has changed since the last read of TC_QISR.

- **DIRCHG: Direction Change**

0: No change on rotation direction since the last read of TC_QISR.

1: The rotation direction changed since the last read of TC_QISR.

- **QERR: Quadrature Error**

0: No quadrature error since the last read of TC_QISR.

1: A quadrature error occurred since the last read of TC_QISR.

- **DIR: Direction**

Returns an image of the actual rotation direction.

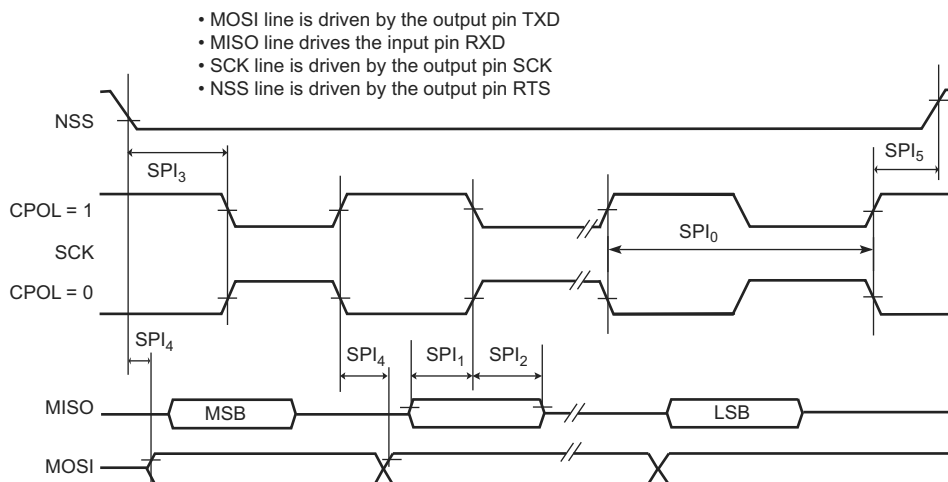
Note that in SPI master mode the SAM4N does not sample the data (MISO) on the opposite edge where data clocks out (MOSI) but the same edge is used as shown in Figure 36-14 and Figure 36-15.

36.10.4 USART in SPI Mode Timings

Timings are given in the following domains:

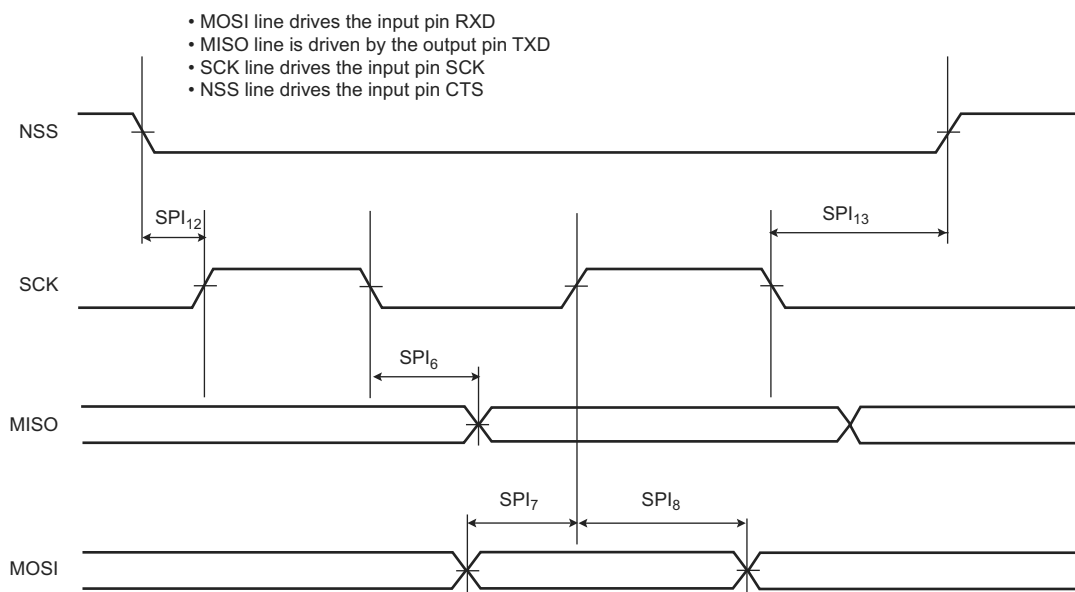
- 1.8V domain: VDDIO from 1.65V to 1.95V, maximum external capacitor = 20 pF
- 3.3V domain: VDDIO from 2.85V to 3.6V, maximum external capacitor = 40 pF

Figure 36-18. USART SPI Master Mode



The full speed is obtained for an input source impedance $< 50 \Omega$ or $t_{\text{TRACK}} = 500 \text{ ns}$.

Figure 36-19. USART SPI Slave Mode (Mode 1 or 2)



36.10.5 Two-wire Serial Interface Characteristics

Table 36-44 describes the requirements for devices connected to the Two-wire Serial Bus. For timing symbols refer to Figure 36-20.

Table 36-44. Two-wire Serial Bus Requirements

Symbol	Parameter	Condition	Min	Max	Unit
V_{IL}	Input Low-voltage		-0.3	$0.3 \times V_{DDIO}$	V
V_{IH}	Input High-voltage		$0.7 \times V_{DDIO}$	$V_{CC} + 0.3$	V
V_{hys}	Hysteresis of Schmitt Trigger Inputs		0.150	—	V
V_{OL}	Output Low-voltage	3 mA sink current	-	0.4	V
t_r	Rise Time for both TWD and TWCK		$20 + 0.1C_b^{(1)(2)}$	300	ns
t_{of}	Output Fall Time from V_{IHmin} to V_{ILmax}	$10 \text{ pF} < C_b < 400 \text{ pF}$ Figure 36-20	$20 + 0.1C_b^{(1)(2)}$	250	ns
$C_i^{(1)}$	Capacitance for each I/O Pin		—	10	pF
f_{TWCK}	TWCK Clock Frequency		0	400	kHz
R_p	Value of Pull-up Resistor	$f_{TWCK} \leq 100 \text{ kHz}$	$(V_{DDIO} - 0.4V) \div 3mA$	$1000ns \div C_b$	Ω
		$f_{TWCK} > 100 \text{ kHz}$	$(V_{DDIO} - 0.4V) \div 3mA$	$300ns \div C_b$	Ω
t_{LOW}	Low Period of the TWCK Clock	$f_{TWCK} \leq 100 \text{ kHz}$	(3)	—	μs
		$f_{TWCK} > 100 \text{ kHz}$	(3)	—	μs
t_{HIGH}	High period of the TWCK Clock	$f_{TWCK} \leq 100 \text{ kHz}$	(4)	—	μs
		$f_{TWCK} > 100 \text{ kHz}$	(4)	—	μs
$t_{h(start)}$	Hold Time (Repeated) START Condition	$f_{TWCK} \leq 100 \text{ kHz}$	t_{HIGH}	—	μs
		$f_{TWCK} > 100 \text{ kHz}$	t_{HIGH}	—	μs
$t_{su(start)}$	Set-up Time for a Repeated START Condition	$f_{TWCK} \leq 100 \text{ kHz}$	t_{HIGH}	—	μs
		$f_{TWCK} > 100 \text{ kHz}$	t_{HIGH}	—	μs
$t_{h(data)}$	Data Hold Time	$f_{TWCK} \leq 100 \text{ kHz}$	0	$3 \times t_{CPMCK}^{(5)}$	μs
		$f_{TWCK} > 100 \text{ kHz}$	0	$3 \times t_{CPMCK}^{(5)}$	μs
$t_{su(data)}$	Data Setup Time	$f_{TWCK} \leq 100 \text{ kHz}$	$t_{LOW} - 3 \times t_{CPMCK}^{(5)}$	—	ns
		$f_{TWCK} > 100 \text{ kHz}$	$t_{LOW} - 3 \times t_{CP_MCK}^{(5)}$	—	ns
$t_{su(stop)}$	Setup Time for STOP Condition	$f_{TWCK} \leq 100 \text{ kHz}$	t_{HIGH}	—	μs
		$f_{TWCK} > 100 \text{ kHz}$	t_{HIGH}	—	μs
t_{BUF}	Bus Free Time between a STOP and START Condition	$f_{TWCK} \leq 100 \text{ kHz}$	t_{LOW}	—	μs
		$f_{TWCK} > 100 \text{ kHz}$	t_{LOW}	—	μs

- Notes:
1. Required only for $f_{TWCK} > 100 \text{ kHz}$.
 2. C_b = capacitance of one bus line in pF. Per I2C Standard, C_b Max = 400 pF
 3. The TWCK low period is defined as follows: $t_{LOW} = ((CLDIV \times 2^{CKDIV}) + 4) \times t_{MCK}$
 4. The TWCK high period is defined as follows: $t_{HIGH} = ((CHDIV \times 2^{CKDIV}) + 4) \times t_{MCK}$
 5. t_{CPMCK} = MCK bus period.