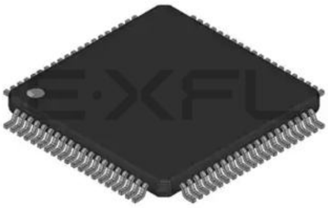
Infineon Technologies - XC2387A56F80LAAKFUMA1 Datasheet



Details	
Product Status	Obsolete
Core Processor	C166SV2
Core Size	16/32-Bit
Speed	80MHz
Connectivity	CANbus, EBI/EMI, I²C, LINbus, SPI, SSC, UART/USART, USI
Peripherals	I²S, POR, PWM, WDT
Number of I/O	118
Program Memory Size	448KB (448K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	64K x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	A/D 24x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	144-LQFP Exposed Pad
Supplier Device Package	PG-LQFP-144-4
Purchase URL	https://www.e-xfl.com/product-detail/infineon-technologies/xc2387a56f80laakfuma1

16/32-Bit Single-Chip Microcontroller with 32-Bit Performance

XC238xA (XC2000 Family)

1 Summary of Features

For a quick overview and easy reference, the features of the XC238xA are summarized here.

- High-performance CPU with five-stage pipeline and MPU
  - 12.5 ns instruction cycle at 80 MHz CPU clock (single-cycle execution)
  - One-cycle 32-bit addition and subtraction with 40-bit result
  - One-cycle multiplication (16 × 16 bit)
  - Background division (32 / 16 bit) in 21 cycles
  - One-cycle multiply-and-accumulate (MAC) instructions
  - Enhanced Boolean bit manipulation facilities
  - Zero-cycle jump execution
  - Additional instructions to support HLL and operating systems
  - Register-based design with multiple variable register banks
  - Fast context switching support with two additional local register banks
  - 16 Mbytes total linear address space for code and data
  - 1024 Bytes on-chip special function register area (C166 Family compatible)
  - Integrated Memory Protection Unit (MPU)
- Interrupt system with 16 priority levels for up to 96 sources
  - Selectable external inputs for interrupt generation and wake-up
  - Fastest sample-rate 12.5 ns
- Eight-channel interrupt-driven single-cycle data transfer with Peripheral Event Controller (PEC), 24-bit pointers cover total address space
- Clock generation from internal or external clock sources, using on-chip PLL or prescaler
- Hardware CRC-Checker with Programmable Polynomial to Supervise On-Chip Memory Areas
- On-chip memory modules
  - 8 Kbytes on-chip stand-by RAM (SBRAM)
  - 2 Kbytes on-chip dual-port RAM (DPRAM)
  - Up to 16 Kbytes on-chip data SRAM (DSRAM)
  - Up to 32 Kbytes on-chip program/data SRAM (PSRAM)
  - Up to 832 Kbytes on-chip program memory (Flash memory)
  - Memory content protection through Error Correction Code (ECC)
- On-Chip Peripheral Modules
  - Multi-functional general purpose timer unit with 5 timers
  - 16-channel general purpose capture/compare unit (CAPCOM2)
  - Up to 4 capture/compare units for flexible PWM signal generation (CCU6x)

The XC238xA types are offered with several SRAM memory sizes. Figure 1 shows the allocation rules for PSRAM and DSRAM. Note that the rules differ:

- PSRAM allocation starts from the **lower** address
- DSRAM allocation starts from the **higher** address

For example 8 Kbytes of PSRAM will be allocated at E0’0000h-E0’1FFFh and 8 Kbytes of DSRAM will be at 00’C000h-00’DFFFh.

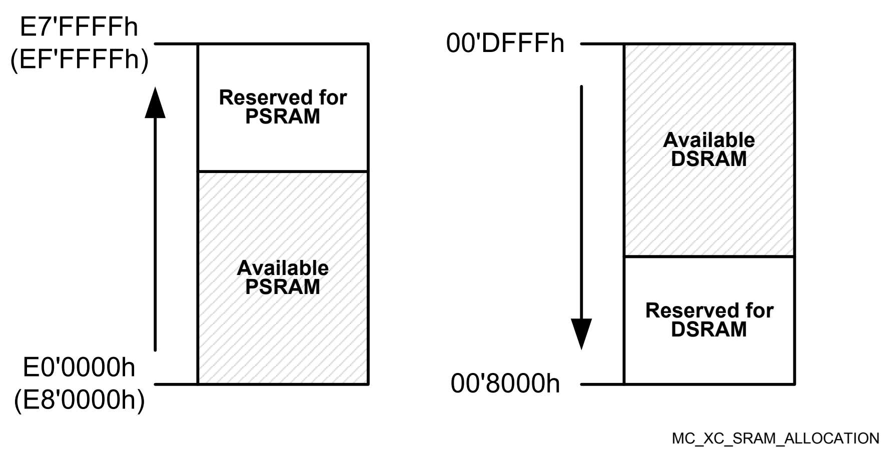


Figure 1 SRAM Allocation

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
105	P10.7	O0 / I	St/B	**Bit 7 of Port 10, General Purpose Input/Output**
	U0C1_DOUT	O1	St/B	**USIC0 Channel 1 Shift Data Output**
	CCU60_COUT63	O2	St/B	**CCU60 Channel 3 Output**
	AD7	OH / IH	St/B	**External Bus Interface Address/Data Line 7**
	U0C1_DX0B	I	St/B	**USIC0 Channel 1 Shift Data Input**
	CCU60_CCPOS0A	I	St/B	**CCU60 Position Input 0**
	T4INB	I	St/B	**GPT12E Timer T4 Count/Gate Input**
106	P0.7	O0 / I	St/B	**Bit 7 of Port 0, General Purpose Input/Output**
	U1C1_DOUT	O1	St/B	**USIC1 Channel 1 Shift Data Output**
	U1C0_SELO3	O2	St/B	**USIC1 Channel 0 Select/Control 3 Output**
	A7	OH	St/B	**External Bus Interface Address Line 7**
	U1C1_DX0B	I	St/B	**USIC1 Channel 1 Shift Data Input**
	CCU61_CTRAPB	I	St/B	**CCU61 Emergency Trap Input**
107	P3.7	O0 / I	St/B	**Bit 7 of Port 3, General Purpose Input/Output**
	U2C1_DOUT	O1	St/B	**USIC2 Channel 1 Shift Data Output**
	U2C0_SELO3	O2	St/B	**USIC2 Channel 0 Select/Control 3 Output**
	U0C0_SELO7	O3	St/B	**USIC0 Channel 0 Select/Control 7 Output**
	U2C1_DX0B	I	St/B	**USIC2 Channel 1 Shift Data Input**

3.6 Interrupt System

The architecture of the XC238xA supports several mechanisms for fast and flexible response to service requests; these can be generated from various sources internal or external to the microcontroller. Any of these interrupt requests can be programmed to be serviced by the Interrupt Controller or by the Peripheral Event Controller (PEC).

Where in a standard interrupt service the current program execution is suspended and a branch to the interrupt vector table is performed, just one cycle is ‘stolen’ from the current CPU activity to perform a PEC service. A PEC service implies a single byte or word data transfer between any two memory locations with an additional increment of either the PEC source pointer, the destination pointer, or both. An individual PEC transfer counter is implicitly decremented for each PEC service except when performing in the continuous transfer mode. When this counter reaches zero, a standard interrupt is performed to the corresponding source-related vector location. PEC services are particularly well suited to supporting the transmission or reception of blocks of data. The XC238xA has eight PEC channels, each whith fast interrupt-driven data transfer capabilities.

With a minimum interrupt response time of 7/11[1] CPU clocks, the XC238xA can react quickly to the occurrence of non-deterministic events.

Interrupt Nodes and Source Selection

The interrupt system provides 96 physical nodes with separate control register containing an interrupt request flag, an interrupt enable flag and an interrupt priority bit field. Most interrupt sources are assigned to a dedicated node. A particular subset of interrupt sources shares a set of nodes. The source selection can be programmed using the interrupt source selection (ISSR) registers.

External Request Unit (ERU)

A dedicated External Request Unit (ERU) is provided to route and preprocess selected on-chip peripheral and external interrupt requests. The ERU features 4 programmable input channels with event trigger logic (ETL) a routing matrix and 4 output gating units (OGU). The ETL features rising edge, falling edge, or both edges event detection. The OGU combines the detected interrupt events and provides filtering capabilities depending on a programmable pattern match or miss.

Trap Processing

The XC238xA provides efficient mechanisms to identify and process exceptions or error conditions that arise during run-time, the so-called ‘Hardware Traps’. A hardware trap causes an immediate system reaction similar to a standard interrupt service (branching

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[1] Depending if the jump cache is used or not.

to a dedicated vector table location). The occurrence of a hardware trap is also indicated by a single bit in the trap flag register (TFR). Unless another higher-priority trap service is in progress, a hardware trap will interrupt any ongoing program execution. In turn, hardware trap services can normally not be interrupted by standard or PEC interrupts.

Depending on the package option up to 3 External Service Request (ESR) pins are provided. The ESR unit processes their input values and allows to implement user controlled trap functions (System Requests SR0 and SR1). In this way reset, wakeup and power control can be efficiently realized.

Software interrupts are supported by the 'TRAP' instruction in combination with an individual trap (interrupt) number. Alternatively to emulate an interrupt by software a program can trigger interrupt requests by writing the Interrupt Request (IR) bit of an interrupt control register.

3.7 On-Chip Debug Support (OCDS)

The On-Chip Debug Support system built into the XC238xA provides a broad range of debug and emulation features. User software running on the XC238xA can be debugged within the target system environment.

The OCDS is controlled by an external debugging device via the debug interface. This either consists of the 2-pin Device Access Port (DAP) or of the JTAG port conforming to IEEE-1149. The debug interface can be completed with an optional break interface.

The debugger controls the OCDS with a set of dedicated registers accessible via the debug interface (DAP or JTAG). In addition the OCDS system can be controlled by the CPU, e.g. by a monitor program. An injection interface allows the execution of OCDS-generated instructions by the CPU.

Multiple breakpoints can be triggered by on-chip hardware, by software, or by an external trigger input. Single stepping is supported, as is the injection of arbitrary instructions and read/write access to the complete internal address space. A breakpoint trigger can be answered with a CPU halt, a monitor call, a data transfer, or/and the activation of an external signal.

Tracing data can be obtained via the debug interface, or via the external bus interface for increased performance.

Tracing of program execution is supported by the XC2000 Family emulation device.

The DAP interface uses two interface signals, the JTAG interface uses four interface signals, to communicate with external circuitry. The debug interface can be amended with two optional break lines.

3.16 Watchdog Timer

The Watchdog Timer is one of the fail-safe mechanisms which have been implemented to prevent the controller from malfunctioning for longer periods of time.

The Watchdog Timer is always enabled after an application reset of the chip. It can be disabled and enabled at any time by executing the instructions DISWDT and ENWDT respectively. The software has to service the Watchdog Timer before it overflows. If this is not the case because of a hardware or software failure, the Watchdog Timer overflows, generating a prewarning interrupt and then a reset request.

The Watchdog Timer is a 16-bit timer clocked with the system clock divided by 16,384 or 256. The Watchdog Timer register is set to a prespecified reload value (stored in WDTREL) in order to allow further variation of the monitored time interval. Each time it is serviced by the application software, the Watchdog Timer is reloaded and the prescaler is cleared.

Time intervals between 3.2 μs and 13.42 s can be monitored (@ 80 MHz).
The default Watchdog Timer interval after power-up is 6.5 ms (@ 10 MHz).

3.17 Clock Generation

The Clock Generation Unit can generate the system clock signal f_{SYS} for the XC238xA from a number of external or internal clock sources:

- External clock signals with pad voltage or core voltage levels
- External crystal or resonator using the on-chip oscillator
- On-chip clock source for operation without crystal/resonator
- Wake-up clock (ultra-low-power) to further reduce power consumption

The programmable on-chip PLL with multiple prescalers generates a clock signal for maximum system performance from standard crystals, a clock input signal, or from the on-chip clock source. See also **Section 4.6.2**.

The Oscillator Watchdog (OWD) generates an interrupt if the crystal oscillator frequency falls below a certain limit or stops completely. In this case, the system can be supplied with an emergency clock to enable operation even after an external clock failure.

All available clock signals can be output on one of two selectable pins.

Table 11 Instruction Set Summary (cont'd)

Mnemonic	Description	Bytes
NOP	Null operation	2
CoMUL/CoMAC	Multiply (and accumulate)	4
CoADD/CoSUB	Add/Subtract	4
Co(A)SHR	(Arithmetic) Shift right	4
CoSHL	Shift left	4
CoLOAD/STORE	Load accumulator/Store MAC register	4
CoCMP	Compare	4
CoMAX/MIN	Maximum/Minimum	4
CoABS/CoRND	Absolute value/Round accumulator	4
CoMOV	Data move	4
CoNEG/NOP	Negate accumulator/Null operation	4

1) The Enter Power Down Mode instruction is not used in the XC238xA, due to the enhanced power control scheme. PWRDN will be correctly decoded, but will trigger no action.

4 Electrical Parameters

The operating range for the XC238xA is defined by its electrical parameters. For proper operation the specified limits must be respected when integrating the device in its target environment.

4.1 General Parameters

These parameters are valid for all subsequent descriptions, unless otherwise noted.

4.1.1 Absolut Maximum Rating Conditions

Stresses above the values listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only. Functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for an extended time may affect device reliability.

During absolute maximum rating overload conditions ($V_{IN} > V_{DDP}$ or $V_{IN} < V_{SS}$) the voltage on V_{DDP} pins with respect to ground (V_{SS}) must not exceed the values defined by the absolute maximum ratings.

Table 12 Absolute Maximum Rating Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Output current on a pin when high value is driven	I_{OH} SR	-30	–	–	mA	
Output current on a pin when low value is driven	I_{OL} SR	–	–	30	mA	
Overload current	I_{OV} SR	-10	–	10	mA	[1]
Absolute sum of overload currents	$\Sigma\|I_{OV}\|$ SR	–	–	100	mA	[1]
Junction Temperature	T_J SR	-40	–	150	°C	
Storage Temperature	T_{ST} SR	-65	–	150	°C	
Digital supply voltage for IO pads and voltage regulators	V_{DDPA}, V_{DDPB} SR	-0.5	–	6.0	V	
Voltage on any pin with respect to ground (Vss)	V_{IN} SR	-0.5	–	V_{DDP} + 0.5	V	$V_{IN} \leq V_{DDP(max)}$

[1] Overload condition occurs if the input voltage V_{IN} is out of the absolute maximum rating range. In this case the current must be limited to the listed values by design measures.

4.2 DC Parameters

These parameters are static or average values that may be exceeded during switching transitions (e.g. output current).

Leakage current is strongly dependent on the operating temperature and the voltage level at the respective pin. The maximum values in the following tables apply under worst case conditions, i.e. maximum temperature and an input level equal to the supply voltage.
The value for the leakage current in an application can be determined by using the respective leakage derating formula (see tables) with values from that application.

The pads of the XC238xA are designed to operate in various driver modes. The DC parameter specifications refer to the pad current limits specified in **Section 4.6.4**.

Supply Voltage Restrictions

The XC238xA can operate within a wide supply voltage range from 3.0 V to 5.5 V. However, during operation this supply voltage must remain within 10 percent of the selected nominal supply voltage. It cannot vary across the full operating voltage range.

Because of the supply voltage restriction and because electrical behavior depends on the supply voltage, the parameters are specified separately for the upper and the lower voltage range.

During operation, the supply voltages may only change with a maximum speed of dV/dt < 1 V/ms.

During power-on sequences, the supply voltages may only change with a maximum speed of dV/dt < 5 V/µs, i.e. the target supply voltage may be reached earliest after approx. 1 µs.

Note: To limit the speed of supply voltage changes, the employment of external buffer capacitors at pins V_{DDPA}/V_{DDPB} *is recommended.*

Sample time and conversion time of the XC238xA's A/D converters are programmable. The timing above can be calculated using **Table 19**.
The limit values for f_{ADCI} must not be exceeded when selecting the prescaler value.

Table 19 A/D Converter Computation Table

GLOBCTR.5-0 (DIVA)	A/D Converter Analog Clock f_{ADCI}	INPCRx.7-0 (STC)	Sample Time[1] t_S
000000_B	f_{SYS}	00_H	$t_{ADCI} \times 2$
000001_B	$f_{SYS} / 2$	01_H	$t_{ADCI} \times 3$
000010_B	$f_{SYS} / 3$	02_H	$t_{ADCI} \times 4$
:	f_{SYS} / (DIVA+1)	:	$t_{ADCI} \times$ (STC+2)
111110_B	$f_{SYS} / 63$	FE_H	$t_{ADCI} \times 256$
111111_B	$f_{SYS} / 64$	FF_H	$t_{ADCI} \times 257$

1) The selected sample time is doubled if broken wire detection is active (due to the presampling phase).

Converter Timing Example A:

Assumptions: f_{SYS} = 80 MHz (i.e. t_{SYS} = 12.5 ns), DIVA = 03_H, STC = 00_H
Analog clock f_{ADCI} = f_{SYS} / 4 = 20 MHz, i.e. t_{ADCI} = 50 ns
Sample time t_S = $t_{ADCI} \times 2$ = 100 ns

Conversion 10-bit:

$t_{C10} = 13 \times t_{ADCI} + 2 \times t_{SYS} = 13 \times 50\text{ ns} + 2 \times 12.5\text{ ns} = 0.675\ \mu\text{s}$

Conversion 8-bit:

$t_{C8} = 11 \times t_{ADCI} + 2 \times t_{SYS} = 11 \times 50\text{ ns} + 2 \times 12.5\text{ ns} = 0.575\ \mu\text{s}$

Converter Timing Example B:

Assumptions: f_{SYS} = 40 MHz (i.e. t_{SYS} = 25 ns), DIVA = 02_H, STC = 03_H
Analog clock f_{ADCI} = f_{SYS} / 3 = 13.3 MHz, i.e. t_{ADCI} = 75 ns
Sample time t_S = $t_{ADCI} \times 5$ = 375 ns

Conversion 10-bit:

$t_{C10} = 16 \times t_{ADCI} + 2 \times t_{SYS} = 16 \times 75\text{ ns} + 2 \times 25\text{ ns} = 1.25\ \mu\text{s}$

Conversion 8-bit:

$t_{C8} = 14 \times t_{ADCI} + 2 \times t_{SYS} = 14 \times 75\text{ ns} + 2 \times 25\text{ ns} = 1.10\ \mu\text{s}$

4.6.2 Definition of Internal Timing

The internal operation of the XC238xA is controlled by the internal system clock f_{SYS}.

Because the system clock signal f_{SYS} can be generated from a number of internal and external sources using different mechanisms, the duration of the system clock periods (TCSs) and their variation (as well as the derived external timing) depend on the mechanism used to generate f_{SYS}. This must be considered when calculating the timing for the XC238xA.

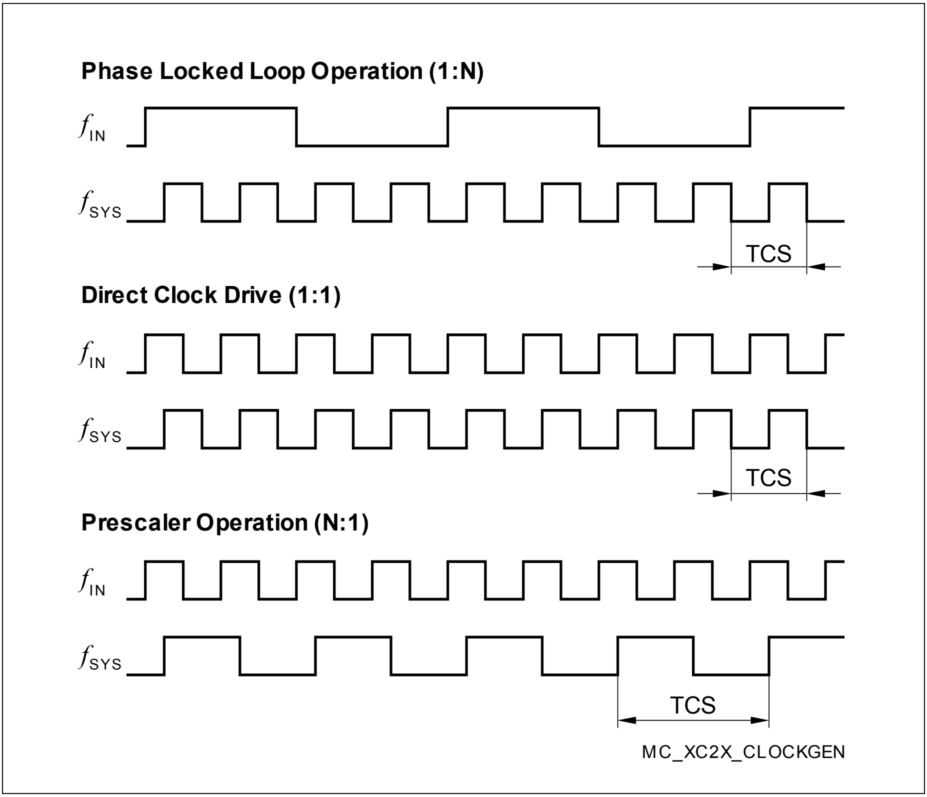


Figure 19 Generation Mechanisms for the System Clock

Note: The example of PLL operation shown in ***Figure 19*** *uses a PLL factor of 1:4; the example of prescaler operation uses a divider factor of 2:1.*

The specification of the external timing (AC Characteristics) depends on the period of the system clock (TCS).

The timing in the AC Characteristics refers to TCSs. Timing must be calculated using the minimum TCS possible under the given circumstances.
The actual minimum value for TCS depends on the jitter of the PLL. Because the PLL is constantly adjusting its output frequency to correspond to the input frequency (from crystal or oscillator), the accumulated jitter is limited. This means that the relative deviation for periods of more than one TCS is lower than for a single TCS (see formulas and **Figure 20**).

This is especially important for bus cycles using waitstates and for the operation of timers, serial interfaces, etc. For all slower operations and longer periods (e.g. pulse train generation or measurement, lower baudrates, etc.) the deviation caused by the PLL jitter is negligible.

The value of the accumulated PLL jitter depends on the number of consecutive VCO output cycles within the respective timeframe. The VCO output clock is divided by the output prescaler K2 to generate the system clock signal f_{SYS}. The number of VCO cycles is K2 × **T**, where **T** is the number of consecutive f_{SYS} cycles (TCS).

The maximum accumulated jitter (long-term jitter) D_{Tmax} is defined by:

$\mathbf{D_{Tmax}}$ **[ns] =** $\pm(220 / (K2 \times f_{SYS}) + 4.3)$

This maximum value is applicable, if either the number of clock cycles $T > (f_{SYS} / 1.2)$ or the prescaler value $K2 > 17$.

In all other cases for a timeframe of **T** × TCS the accumulated jitter D_T is determined by:

$\mathbf{D_T}$ **[ns] =** $D_{Tmax} \times [(1 - 0.058 \times K2) \times (T - 1) / (0.83 \times f_{SYS} - 1) + 0.058 \times K2]$

f_{SYS} in [MHz] in all formulas.

Example, for a period of 3 TCSs @ 33 MHz and K2 = 4:

$D_{max} = \pm(220 / (4 \times 33) + 4.3) = 5.97$ ns (Not applicable directly in this case!)

$D_3 = 5.97 \times [(1 - 0.058 \times 4) \times (3 - 1) / (0.83 \times 33 - 1) + 0.058 \times 4]$
$= 5.97 \times [0.768 \times 2 / 26.39 + 0.232]$
$= 1.7$ ns

Example, for a period of 3 TCSs @ 33 MHz and K2 = 2:

$D_{max} = \pm(220 / (2 \times 33) + 4.3) = 7.63$ ns (Not applicable directly in this case!)

$D_3 = 7.63 \times [(1 - 0.058 \times 2) \times (3 - 1) / (0.83 \times 33 - 1) + 0.058 \times 2]$
$= 7.63 \times [0.884 \times 2 / 26.39 + 0.116]$
$= 1.4$ ns

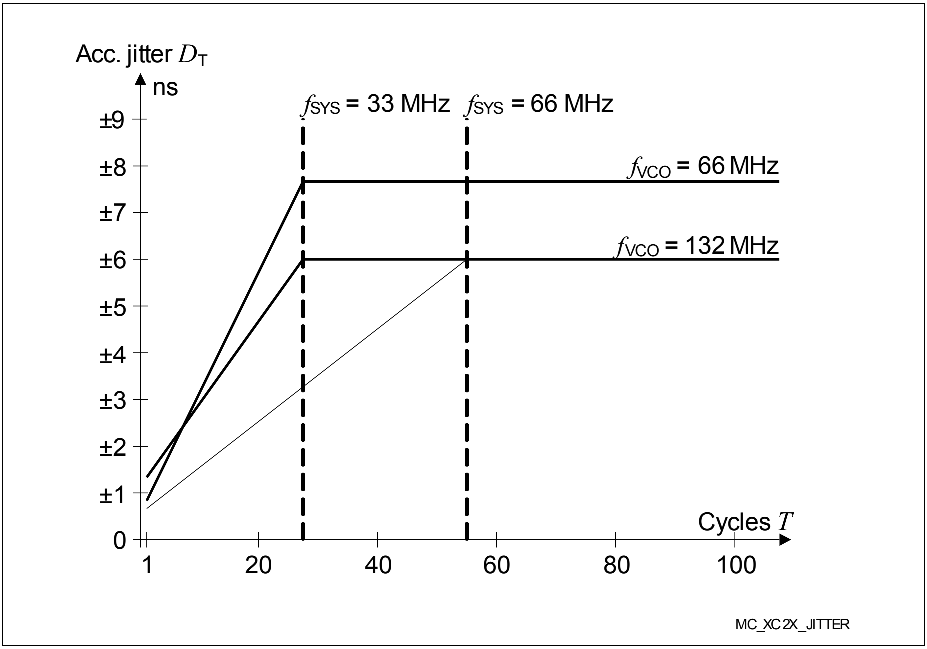


Figure 20 **Approximated Accumulated PLL Jitter**

Note: The specified PLL jitter values are valid if the capacitive load per pin does not exceed C_L = 20 pF.
The maximum peak-to-peak noise on the pad supply voltage (measured between V_{DDPB} pin 100 and V_{SS} pin 1) is limited to a peak-to-peak voltage of V_{PP} = 50 mV. This can be achieved by appropriate blocking of the supply voltage as close as possible to the supply pins and using PCB supply and ground planes.

Table 26 is valid under the following conditions:
$V_{DDP} \geq 4.5$ V; $V_{DDPtyp} = 5$ V; $V_{DDP} \leq 5.5$ V; $C_L \geq 20$ pF; $C_L \leq 100$ pF;

Table 26 Standard Pad Parameters for Upper Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Maximum output driver current (absolute value)[1]	I_{Omax} CC	–	–	10	mA	Strong driver
		–	–	4.0	mA	Medium driver
		–	–	0.5	mA	Weak driver
Nominal output driver current (absolute value)	I_{Onom} CC	–	–	2.5	mA	Strong driver
		–	–	1.0	mA	Medium driver
		–	–	0.1	mA	Weak driver
Rise and Fall times (10% - 90%)	t_{RF} CC	–	–	4.2 + 0.14 x C_L	ns	Strong driver; Sharp edge
		–	–	11.6 + 0.22 x C_L	ns	Strong driver; Medium edge
		–	–	20.6 + 0.22 x C_L	ns	Strong driver; Slow edge
		–	–	23 + 0.6 x C_L	ns	Medium driver
		–	–	212 + 1.9 x C_L	ns	Weak driver

[1] The total output current that may be drawn at a given time must be limited to protect the supply rails from damage. For any group of 16 neighboring output pins, the total output current in each direction (ΣI_{OL} and Σ-I_{OH}) must remain below 50 mA.

4.6.5 External Bus Timing

The following parameters specify the behavior of the XC238xA bus interface.

Note: These parameters are not subject to production test but verified by design and/or characterization.

Note: Operating Conditions apply.

Bus Interface Performance Limits

The output frequency at the bus interface pins is limited by the performance of the output drivers. The fast clock driver (used for CLKOUT) can drive 80-MHz signals, the standard drivers can drive 40-MHz signals

Therefore, the speed of the EBC must be limited, either by limiting the system frequency to f_{SYS} ≤ 80 MHz or by adding waitstates so that signal transitions have a minimum distance of 12.5 ns.

For a description of the bus protocol and the programming of its variable timing parameters, please refer to the User's Manual.

Table 28 EBC Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
CLKOUT Cycle Time[1]	t_5 CC	–	1 / f_{SYS}	–	ns	
CLKOUT high time	t_6 CC	2	–	–		
CLKOUT low time	t_7 CC	2	–	–		
CLKOUT rise time	t_8 CC	–	–	3	ns	
CLKOUT fall time	t_9 CC	–	–	3		

[1] The CLKOUT cycle time is influenced by PLL jitter. For longer periods the relative deviation decreases (see PLL deviation formula).

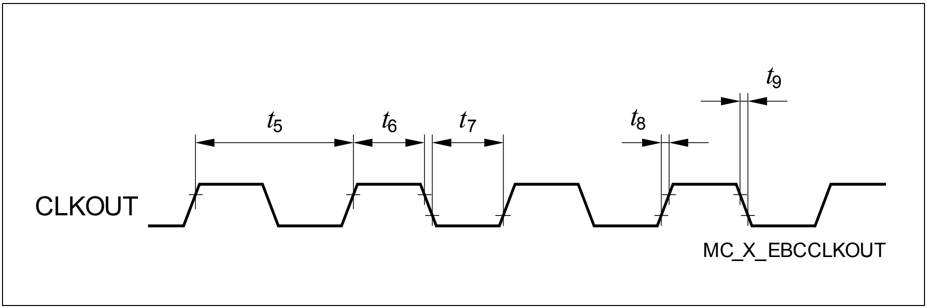


Figure 22 CLKOUT Signal Timing

4.6.7 Debug Interface Timing

The debugger can communicate with the XC238xA either via the 2-pin DAP interface or via the standard JTAG interface.

Debug via DAP

The following parameters are applicable for communication through the DAP debug interface.

Note: These parameters are not subject to production test but verified by design and/or characterization.

Note: Operating Conditions apply; C_L= 20 pF.

Table 38 DAP Interface Timing for Upper Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
DAP0 clock period	t_{11} SR	25[1]	–	–	ns	
DAP0 high time	t_{12} SR	8	–	–	ns	
DAP0 low time	t_{13} SR	8	–	–	ns	
DAP0 clock rise time	t_{14} SR	–	–	4	ns	
DAP0 clock fall time	t_{15} SR	–	–	4	ns	
DAP1 setup to DAP0 rising edge	t_{16} SR	6	–	–	ns	pad_type= standard
DAP1 hold after DAP0 rising edge	t_{17} SR	6	–	–	ns	pad_type= standard
DAP1 valid per DAP0 clock period[2]	t_{19} CC	17	20	–	ns	pad_type= standard

[1] The debug interface cannot operate faster than the overall system, therefore $t_{11} \geq t_{SYS}$.

[2] The Host has to find a suitable sampling point by analyzing the sync telegram response.

Table 41 JTAG Interface Timing for Lower Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
TCK clock period	t_1 SR	50[1]	–	–	ns	[2]
TCK high time	t_2 SR	16	–	–	ns	
TCK low time	t_3 SR	16	–	–	ns	
TCK clock rise time	t_4 SR	–	–	8	ns	
TCK clock fall time	t_5 SR	–	–	8	ns	
TDI/TMS setup to TCK rising edge	t_6 SR	6	–	–	ns	
TDI/TMS hold after TCK rising edge	t_7 SR	6	–	–	ns	
TDO valid from TCK falling edge (propagation delay)[3]	t_8 CC	–	32	36	ns	
TDO high impedance to valid output from TCK falling edge[4][3]	t_9 CC	–	32	36	ns	
TDO valid output to high impedance from TCK falling edge[3]	t_{10} CC	–	32	36	ns	
TDO hold after TCK falling edge[3]	t_{18} CC	5	–	–	ns	

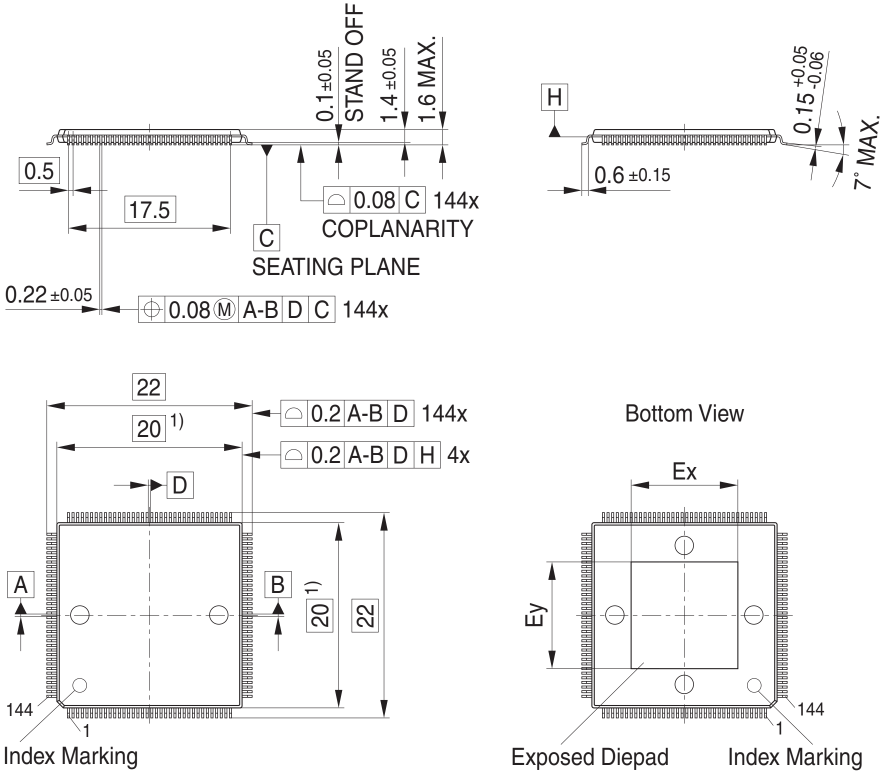
1) The debug interface cannot operate faster than the overall system, therefore $t_1 \geq t_{SYS}$.

2) Under typical conditions, the interface can operate at transfer rates up to 20 MHz.

3) The falling edge on TCK is used to generate the TDO timing.

4) The setup time for TDO is given implicitly by the TCK cycle time.

Package Outlines



1) Does not include plastic or metal protrusion of 0.25 max. per side

PG-LQFP-144-4, -13-PO V08

Figure 34 PG-LQFP-144-13/-23 (Plastic Green Thin Quad Flat Package)

All dimensions in mm.

You can find complete information about Infineon packages, packing and marking in our Infineon Internet Page “Packages”: **http://www.infineon.com/packages**

5.3 Quality Declarations

The operation lifetime of the XC238xA depends on the applied temperature profile in the application. For a typical example, please refer to **Table 44**; for other profiles, please contact your Infineon counterpart to calculate the specific lifetime within your application.

Table 43 Quality Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Operation lifetime	t_{OP} CC	–	–	20	a	See **Table 44** and **Table 45**
ESD susceptibility according to Human Body Model (HBM)	V_{HBM} SR	–	–	2 000	V	EIA/JESD22-A114-B
Moisture sensitivity level	MSL CC	–	–	3	–	JEDEC J-STD-020C

Table 44 Typical Usage Temperature Profile

Operating Time (Sum = 20 years)	Operating Temperat.	Notes
1 200 h	T_J = 150°C	Normal operation
3 600 h	T_J = 125°C	Normal operation
7 200 h	T_J = 110°C	Normal operation
12 000 h	T_J = 100°C	Normal operation
7 × 21 600 h	T_J = 0…10°C, …, 60…70°C	Power reduction

Table 45 Long Time Storage Temperature Profile

Operating Time (Sum = 20 years)	Operating Temperat.	Notes
2 000 h	T_J = 150°C	Normal operation
16 000 h	T_J = 125°C	Normal operation
6 000 h	T_J = 110°C	Normal operation
151 200 h	$T_J \leq$ 150°C	No operation

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