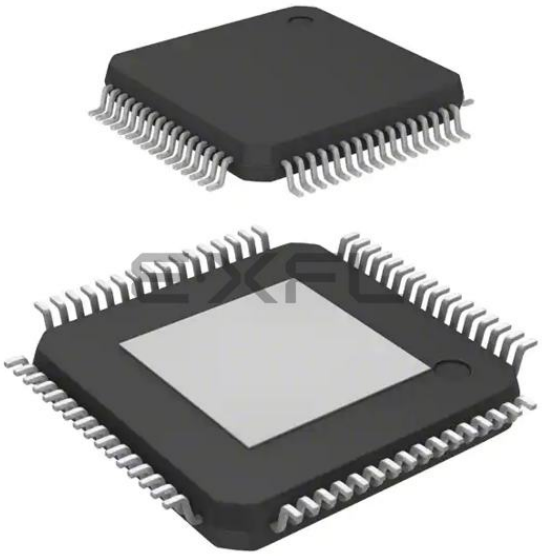
Infineon Technologies - XC2336B40F80LRABKXUMA1 Datasheet



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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Not For New Designs
Core Processor	C166SV2
Core Size	16/32-Bit
Speed	80MHz
Connectivity	CANbus, EBI/EMI, I²C, LINbus, SPI, SSC, UART/USART, USI
Peripherals	I²S, POR, PWM, WDT
Number of I/O	38
Program Memory Size	320KB (320K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	34K x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	A/D 9x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	64-LQFP Exposed Pad
Supplier Device Package	PG-LQFP-64-6
Purchase URL	https://www.e-xfl.com/product-detail/infineon-technologies/xc2336b40f80lrabkxuma1

1.3 Definition of Feature Variants

The XC2336B types are offered with several Flash memory sizes. **Table 3** and **Table 4** describe the location of the available Flash memory.

Table 3 Continuous Flash Memory Ranges

Total Flash Size	1st Range[1)]	2nd Range	3rd Range
320 Kbytes	$C0'0000_H$ … $C0'EFFF_H$	$C1'0000_H$ … $C4'FFFF_H$	n.a.
192 Kbytes	$C0'0000_H$ … $C0'EFFF_H$	$C1'0000_H$ … $C1'FFFF_H$	$C4'0000_H$ … $C4'FFFF_H$

1) The uppermost 4-Kbyte sector of the first Flash segment is reserved for internal use ($C0'F000_H$ to $C0'FFFF_H$).

Table 4 Flash Memory Module Allocation (in Kbytes)

Total Flash Size	Flash 0[1)]	Flash 1
320	256	64
192	128	64

1) The uppermost 4-Kbyte sector of the first Flash segment is reserved for internal use ($C0'F000_H$ to $C0'FFFF_H$).

The XC2336B types are offered with different interface options. **Table 5** lists the available channels for each option.

Table 5 Interface Channel Association

Total Number	Available Channels / Message Objects
7 ADC0 channels	CH0, CH2, Ch4, CH8, CH10, CH13, CH15
2 ADC1 channels	CH0, CH4
2 CAN nodes	CAN0, CAN1 64 message objects
4 serial channels	U0C0, U0C1, U1C0, U1C1

The XC2336B types are offered with several SRAM memory sizes. **Figure 1** shows the allocation rules for PSRAM and DSRAM. Note that the rules differ:

- PSRAM allocation starts from the **lower** address
- DSRAM allocation starts from the **higher** address

For example 8 Kbytes of PSRAM will be allocated at E0’0000h-E0’1FFFh and 8 Kbytes of DSRAM will be at 00’C000h-00’DFFFh.

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
7	P6.0	O0 / I	DA/A	**Bit 0 of Port 6, General Purpose Input/Output**
	EMUX0	O1	DA/A	**External Analog MUX Control Output 0 (ADC0)**
	$\overline{\text{BRKOUT}}$	O3	DA/A	**OCDS Break Signal Output**
	ADCx_REQGTyG	I	DA/A	**External Request Gate Input for ADC0/1**
	U1C1_DX0E	I	DA/A	**USIC1 Channel 1 Shift Data Input**
8	P6.1	O0 / I	DA/A	**Bit 1 of Port 6, General Purpose Input/Output**
	EMUX1	O1	DA/A	**External Analog MUX Control Output 1 (ADC0)**
	T3OUT	O2	DA/A	**GPT12E Timer T3 Toggle Latch Output**
	U1C1_DOUT	O3	DA/A	**USIC1 Channel 1 Shift Data Output**
	ADCx_REQTRyE	I	DA/A	**External Request Trigger Input for ADC0/1**
	ESR1_6	I	DA/A	**ESR1 Trigger Input 6**
10	P15.0	I	In/A	**Bit 0 of Port 15, General Purpose Input**
	ADC1_CH0	I	In/A	**Analog Input Channel 0 for ADC1**
11	P15.4	I	In/A	**Bit 4 of Port 15, General Purpose Input**
	ADC1_CH4	I	In/A	**Analog Input Channel 4 for ADC1**
	T6INA	I	In/A	**GPT12E Timer T6 Count/Gate Input**
12	V_{AREF}	-	PS/A	**Reference Voltage for A/D Converters ADC0/1**
13	V_{AGND}	-	PS/A	**Reference Ground for A/D Converters ADC0/1**
14	P5.0	I	In/A	**Bit 0 of Port 5, General Purpose Input**
	ADC0_CH0	I	In/A	**Analog Input Channel 0 for ADC0**
15	P5.2	I	In/A	**Bit 2 of Port 5, General Purpose Input**
	ADC0_CH2	I	In/A	**Analog Input Channel 2 for ADC0**
	TDI_A	I	In/A	**JTAG Test Data Input**
19	P5.4	I	In/A	**Bit 4 of Port 5, General Purpose Input**
	ADC0_CH4	I	In/A	**Analog Input Channel 4 for ADC0**
	T3EUDA	I	In/A	**GPT12E Timer T3 External Up/Down Control Input**
	TMS_A	I	In/A	**JTAG Test Mode Selection Input**

3 Functional Description

The architecture of the XC2336B combines advantages of RISC, CISC, and DSP processors with an advanced peripheral subsystem in a well-balanced design. On-chip memory blocks allow the design of compact systems-on-silicon with maximum performance suited for computing, control, and communication.

The on-chip memory blocks (program code memory and SRAM, dual-port RAM, data SRAM) and the generic peripherals are connected to the CPU by separate high-speed buses. Another bus, the LXBus, connects additional on-chip resources and external resources. This bus structure enhances overall system performance by enabling the concurrent operation of several subsystems of the XC2336B.

The block diagram gives an overview of the on-chip components and the advanced internal bus structure of the XC2336B.

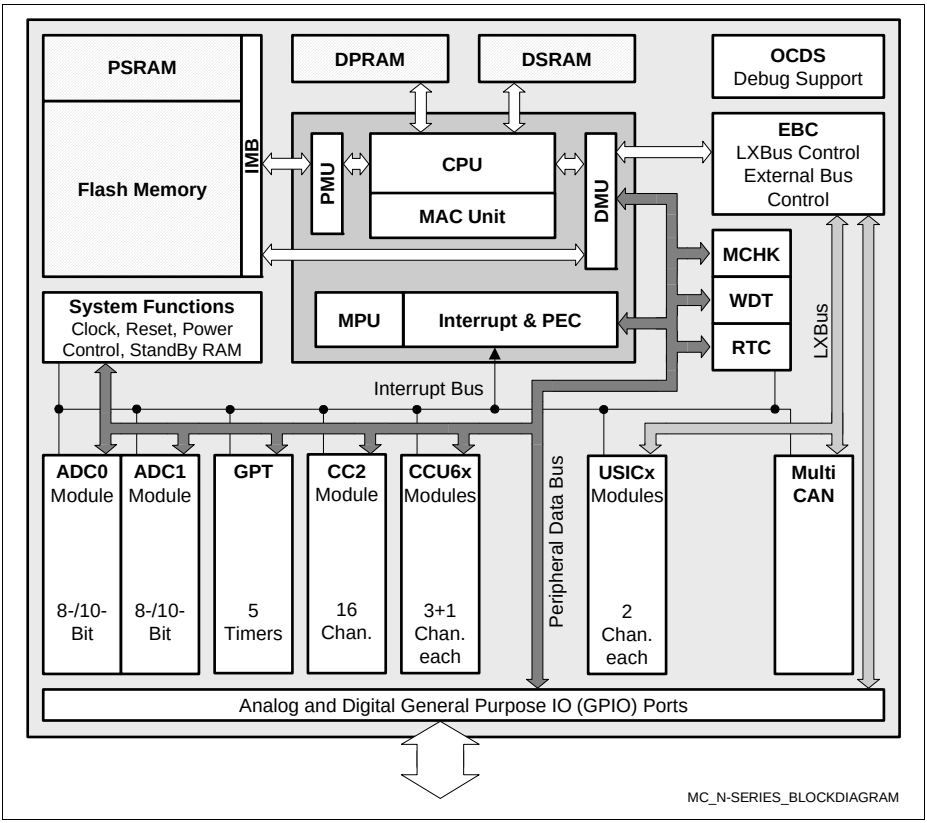

Figure 4 Block Diagram

Up to 16 Kbytes of on-chip Data SRAM (DSRAM) are used for storage of general user data. The DSRAM is accessed via a separate interface and is optimized for data access.

Note: The actual size of the DSRAM depends on the quoted device type.

2 Kbytes of on-chip Dual-Port RAM (DPRAM) provide storage for user-defined variables, for the system stack, and for general purpose register banks. A register bank can consist of up to 16 word-wide (R0 to R15) and/or byte-wide (RL0, RH0, …, RL7, RH7) General Purpose Registers (GPRs).
The upper 256 bytes of the DPRAM are directly bit addressable. When used by a GPR, any location in the DPRAM is bit addressable.

8 Kbytes of on-chip Stand-By SRAM (SBRAM) provide storage for system-relevant user data that must be preserved while the major part of the device is powered down. The SBRAM is accessed via a specific interface and is powered in domain M.

1024 bytes (2 × 512 bytes) of the address space are reserved for the Special Function Register areas (SFR space and ESFR space). SFRs are word-wide registers which are used to control and monitor functions of the different on-chip units. Unused SFR addresses are reserved for future members of the XC2000 Family. In order to ensure upward compatibility they should either not be accessed or written with zeros.

In order to meet the requirements of designs where more memory is required than is available on chip, up to 12 Mbytes (approximately, see **Table 8**) of external RAM and/or ROM can be connected to the microcontroller. The External Bus Interface also provides access to external peripherals.

The on-chip Flash memory stores code, constant data, and control data. The 320 Kbytes of on-chip Flash memory consist of 1 module of 64 Kbytes (preferably for data storage) and 1 module of 256 Kbytes. Each module is organized in 4-Kbyte sectors. The uppermost 4-Kbyte sector of segment 0 (located in Flash module 0) is used internally to store operation control parameters and protection information.

Note: The actual size of the Flash memory depends on the chosen device type.

Each sector can be separately write protected[1)], erased and programmed (in blocks of 128 Bytes). The complete Flash area can be read-protected. A user-defined password sequence temporarily unlocks protected areas. The Flash modules combine 128-bit read access with protected and efficient writing algorithms for programming and erasing. Dynamic error correction provides extremely high read data security for all read access operations. Access to different Flash modules can be executed in parallel.
For Flash parameters, please see **Section 4.6**.

1) To save control bits, sectors are clustered for protection purposes, they remain separate for programming/erasing.

to a dedicated vector table location). The occurrence of a hardware trap is also indicated by a single bit in the trap flag register (TFR). Unless another higher-priority trap service is in progress, a hardware trap will interrupt any ongoing program execution. In turn, hardware trap services can normally not be interrupted by standard or PEC interrupts.

Depending on the package option up to 3 External Service Request (ESR) pins are provided. The ESR unit processes their input values and allows to implement user controlled trap functions (System Requests SR0 and SR1). In this way reset, wakeup and power control can be efficiently realized.

Software interrupts are supported by the 'TRAP' instruction in combination with an individual trap (interrupt) number. Alternatively to emulate an interrupt by software a program can trigger interrupt requests by writing the Interrupt Request (IR) bit of an interrupt control register.

3.6 On-Chip Debug Support (OCDS)

The On-Chip Debug Support system built into the XC2336B provides a broad range of debug and emulation features. User software running on the XC2336B can be debugged within the target system environment.

The OCDS is controlled by an external debugging device via the debug interface. This either consists of the 2-pin Device Access Port (DAP) or of the JTAG port conforming to IEEE-1149. The debug interface can be completed with an optional break interface.

The debugger controls the OCDS with a set of dedicated registers accessible via the debug interface (DAP or JTAG). In addition the OCDS system can be controlled by the CPU, e.g. by a monitor program. An injection interface allows the execution of OCDS-generated instructions by the CPU.

Multiple breakpoints can be triggered by on-chip hardware, by software, or by an external trigger input. Single stepping is supported, as is the injection of arbitrary instructions and read/write access to the complete internal address space. A breakpoint trigger can be answered with a CPU halt, a monitor call, a data transfer, or/and the activation of an external signal.

Tracing of data can be obtained via the debug interface, or via the external bus interface for increased performance.

Tracing of program execution is supported by the XC2000 Family emulation device. With this device the DAP can operate on clock rates of up to 20 MHz.

The DAP interface uses two interface signals, the JTAG interface uses four interface signals, to communicate with external circuitry. The debug interface can be amended with two optional break lines.

Table 9 Compare Modes (cont'd)

Compare Modes	Function
Mode 2	Interrupt-only compare mode; Only one compare interrupt per timer period is generated
Mode 3	Pin set '1' on match; pin reset '0' on compare timer overflow; Only one compare event per timer period is generated
Double Register Mode	Two registers operate on one pin; Pin toggles on each compare match; Several compare events per timer period are possible
Single Event Mode	Generates single edges or pulses; Can be used with any compare mode

When a capture/compare register has been selected for capture mode, the current contents of the allocated timer will be latched ('captured') into the capture/compare register in response to an external event at the port pin associated with this register. In addition, a specific interrupt request for this capture/compare register is generated. Either a positive, a negative, or both a positive and a negative external signal transition at the pin can be selected as the triggering event.

The contents of all registers selected for one of the five compare modes are continuously compared with the contents of the allocated timers.

When a match occurs between the timer value and the value in a capture/compare register, specific actions will be taken based on the compare mode selected.

With its maximum resolution of 2 system clock cycles, the **GPT2 module** provides precise event control and time measurement. It includes two timers (T5, T6) and a capture/reload register (CAPREL). Both timers can be clocked with an input clock which is derived from the CPU clock via a programmable prescaler or with external signals. The counting direction (up/down) for each timer can be programmed by software or altered dynamically with an external signal on a port pin (TxEUD). Concatenation of the timers is supported with the output toggle latch (T6OTL) of timer T6, which changes its state on each timer overflow/underflow.

The state of this latch may be used to clock timer T5, and/or it may be output on pin T6OUT. The overflows/underflows of timer T6 can also be used to clock the CAPCOM2 timers and to initiate a reload from the CAPREL register.

The CAPREL register can capture the contents of timer T5 based on an external signal transition on the corresponding port pin (CAPIN); timer T5 may optionally be cleared after the capture procedure. This allows the XC2336B to measure absolute time differences or to perform pulse multiplication without software overhead.

The capture trigger (timer T5 to CAPREL) can also be generated upon transitions of GPT1 timer T3 inputs T3IN and/or T3EUD. This is especially advantageous when T3 operates in Incremental Interface Mode.

MultiCAN Features

- CAN functionality conforming to CAN specification V2.0 B active for each CAN node (compliant to ISO 11898)
- Independent CAN nodes
- Set of independent message objects (shared by the CAN nodes)
- Dedicated control registers for each CAN node
- Data transfer rate up to 1 Mbit/s, individually programmable for each node
- Flexible and powerful message transfer control and error handling capabilities
- Full-CAN functionality for message objects:
  - Can be assigned to one of the CAN nodes
  - Configurable as transmit or receive objects, or as message buffer FIFO
  - Handle 11-bit or 29-bit identifiers with programmable acceptance mask for filtering
  - Remote Monitoring Mode, and frame counter for monitoring
- Automatic Gateway Mode support
- 16 individually programmable interrupt nodes
- Analyzer mode for CAN bus monitoring

3.14 System Timer

The System Timer consists of a programmable prescaler and two concatenated timers (10 bits and 6 bits). Both timers can generate interrupt requests. The clock source can be selected and the timers can also run during power reduction modes.

Therefore, the System Timer enables the software to maintain the current time for scheduling functions or for the implementation of a clock.

3.15 Watchdog Timer

The Watchdog Timer is one of the fail-safe mechanisms which have been implemented to prevent the controller from malfunctioning for longer periods of time.

The Watchdog Timer is always enabled after an application reset of the chip. It can be disabled and enabled at any time by executing the instructions DISWDT and ENWDT respectively. The software has to service the Watchdog Timer before it overflows. If this is not the case because of a hardware or software failure, the Watchdog Timer overflows, generating a prewarning interrupt and then a reset request.

The Watchdog Timer is a 16-bit timer clocked with the system clock divided by 16,384 or 256. The Watchdog Timer register is set to a prespecified reload value (stored in WDTREL) in order to allow further variation of the monitored time interval. Each time it is serviced by the application software, the Watchdog Timer is reloaded and the prescaler is cleared.

Time intervals between 3.2 µs and 13.4 s can be monitored (@ 80 MHz).

The default Watchdog Timer interval after power-up is 6.5 ms (@ 10 MHz).

Table 11 Instruction Set Summary (cont'd)

Mnemonic	Description	Bytes
NOP	Null operation	2
CoMUL/CoMAC	Multiply (and accumulate)	4
CoADD/CoSUB	Add/Subtract	4
Co(A)SHR	(Arithmetic) Shift right	4
CoSHL	Shift left	4
CoLOAD/STORE	Load accumulator/Store MAC register	4
CoCMP	Compare	4
CoMAX/MIN	Maximum/Minimum	4
CoABS/CoRND	Absolute value/Round accumulator	4
CoMOV	Data move	4
CoNEG/NOP	Negate accumulator/Null operation	4

1) The Enter Power Down Mode instruction is not used in the XC2336B, due to the enhanced power control scheme. PWRDN will be correctly decoded, but will trigger no action.

4.3.3 Power Consumption

The power consumed by the XC2336B depends on several factors such as supply voltage, operating frequency, active circuits, and operating temperature. The power consumption specified here consists of two components:

- The switching current I_S depends on the device activity
- The leakage current I_{LK} depends on the device temperature

To determine the actual power consumption, always both components, switching current I_S and leakage current I_{LK} must be added:

$I_{DDP} = I_S + I_{LK}$.

Note: The power consumption values are not subject to production test. They are verified by design/characterization.
To determine the total power consumption for dimensioning the external power supply, also the pad driver currents must be considered.

The given power consumption parameters and their values refer to specific operating conditions:

- **Active mode**:
  Regular operation, i.e. peripherals are active, code execution out of Flash.
- **Stopover mode**:
  Crystal oscillator and PLL stopped, Flash switched off, clock in domain DMP_1 stopped.

Note: The maximum values cover the complete specified operating range of all manufactured devices.
The typical values refer to average devices under typical conditions, such as nominal supply voltage, room temperature, application-oriented activity.
After a power reset, the decoupling capacitors for V_{DDIM} and V_{DDI1} are charged with the maximum possible current.

For additional information, please refer to **Section 5.2**, **Thermal Considerations**.

Note: Operating Conditions apply.

Table 18 Switching Power Consumption

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Power supply current (active) with all peripherals active and EVVRs on	I_{SACT} CC	–	6 + 0.6 x f_{SYS} [1]	8 + 1.0 x f_{SYS} [1]	mA	power_mode= active ; voltage_range= both [2][3][4]
Power supply current in stopover mode, EVVRs on	I_{SSO} CC	–	0.7	2.0	mA	power_mode= stopover ; voltage_range= both [4]

1) f_{SYS} in MHz

2) The pad supply voltage pins (V_{DDPB}) provide the input current for the on-chip EVVRs and the current consumed by the pin output drivers. A small current is consumed because the drivers input stages are switched. In Fast Startup Mode (with the Flash modules deactivated), the typical current is reduced to 3 + 0.6 x f_{SYS}.

3) Please consider the additional conditions described in section "Active Mode Power Supply Current".

4) The pad supply voltage has only a minor influence on this parameter.

Active Mode Power Supply Current

The actual power supply current in active mode not only depends on the system frequency but also on the configuration of the XC2336B's subsystem.

Besides the power consumed by the device logic the power supply pins also provide the current that flows through the pin output drivers.
A small current is consumed because the drivers' input stages are switched.

The IO power domains can be supplied separately. Power domain A (V_{DDPA}) supplies the A/D converters and Port 6. Power domain B (V_{DDPB}) supplies the on-chip EVVRs and all other ports.
During operation domain A draws a maximum current of 1.5 mA for each active A/D converter module from V_{DDPA}.

In Fast Startup Mode (with the Flash modules deactivated), the typical current is reduced to 3 + 0.6×f_{SYS} mA.

Note: A fraction of the leakage current flows through domain DMP_A (pin V_{DDPA}). This current can be calculated as 7,000 $\times e^{-\alpha}$, with α = 5000 / (273 + 1.3 $\times T_J$). For T_J = 150°C, this results in a current of 160 μA.

Leakage Power Consumption Calculation

The leakage power consumption can be calculated according to the following formula:

$I_{LK1} = 530{,}000 \times e^{-\alpha}$ with $\alpha = 5000 / (273 + B \times T_J)$

Parameter B must be replaced by

- 1.0 for typical values
- 1.3 for maximum values

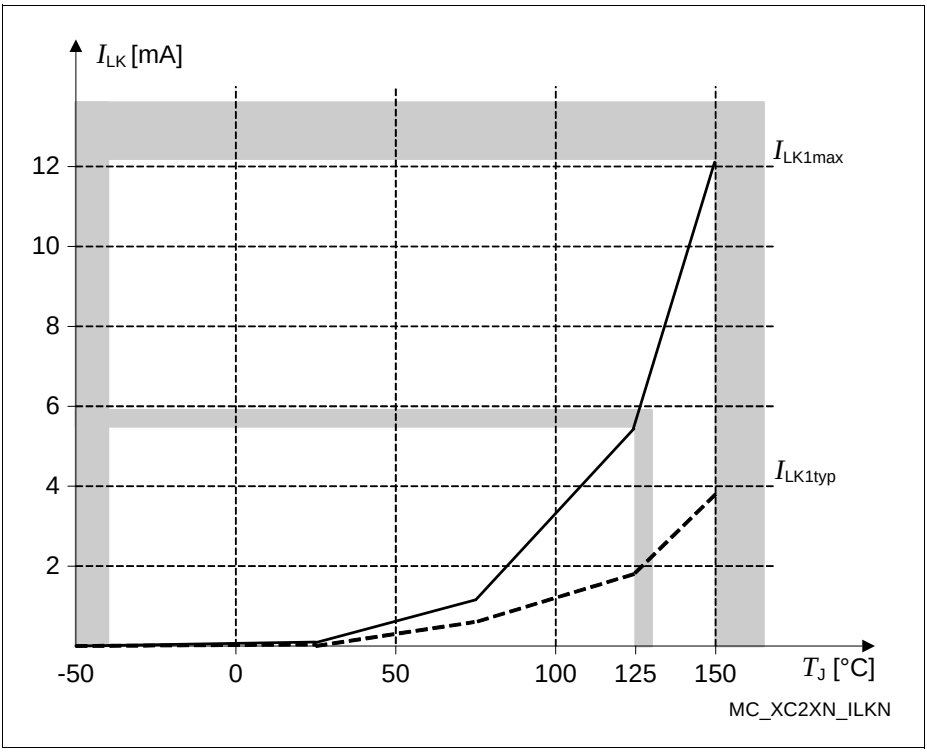

Figure 15 Leakage Supply Current as a Function of Temperature

Table 20 ADC Parameters (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Broken wire detection delay against VAGND[2]	t_{BWG} CC	–	–	50[3]		
Broken wire detection delay against VAREF[2]	t_{BWR} CC	–	–	50[4]		
Conversion time for 8-bit result[2]	t_{c8} CC	$(11+STC)$ x t_{ADCI} + 2 x t_{SYS}	–	–		
Conversion time for 10-bit result[2]	t_{c10} CC	$(13+STC)$ x t_{ADCI} + 2 x t_{SYS}	–	–		
Total Unadjusted Error	|TUE| CC	–	1	2	LSB	[5]
Wakeup time from analog powerdown, fast mode	t_{WAF} CC	–	–	4	µs	
Wakeup time from analog powerdown, slow mode	t_{WAS} CC	–	–	15	µs	
Analog reference ground	V_{AGND} SR	V_{SS} - 0.05	–	1.5	V	
Analog input voltage range	V_{AIN} SR	V_{AGND}	–	V_{AREF}	V	[6]
Analog reference voltage	V_{AREF} SR	V_{AGND} + 1.0	–	V_{DDPA} + 0.05	V	

1) These parameter values cover the complete operating range. Under relaxed operating conditions (temperature, supply voltage) typical values can be used for calculation. At room temperature and nominal supply voltage the following typical values can be used: $C_{AINTtyp}$ = 12 pF, $C_{AINStyp}$ = 5 pF, R_{AINtyp} = 1.0 kOhm, $C_{AREFTtyp}$ = 15 pF, $C_{AREFStyp}$ = 10 pF, $R_{AREFtyp}$ = 1.0 kOhm.

2) This parameter includes the sample time (also the additional sample time specified by STC), the time to determine the digital result and the time to load the result register with the conversion result. Values for the basic clock t_{ADCI} depend on programming.

3) The broken wire detection delay against V_{AGND} is measured in numbers of consecutive precharge cycles at a conversion rate of not more than 500 µs. Result below 10% (66_H)

4) The broken wire detection delay against V_{AREF} is measured in numbers of consecutive precharge cycles at a conversion rate of not more than 10 µs. This function is influenced by leakage current, in particular at high temperature. Result above 80% (332_H)

5) TUE is tested at V_{AREF} = V_{DDPA} = 5.0 V, V_{AGND} = 0 V. It is verified by design for all other voltages within the defined voltage range. The specified TUE is valid only if the absolute sum of input overload currents on analog port pins (see I_{OV} specification) does not exceed 10 mA, and if V_{AREF} and V_{AGND} remain stable during the measurement time.

6) V_{AIN} may exceed V_{AGND} or V_{AREF} up to the absolute maximum ratings. However, the conversion result in these cases will be $X000_H$ or $X3FF_H$, respectively.

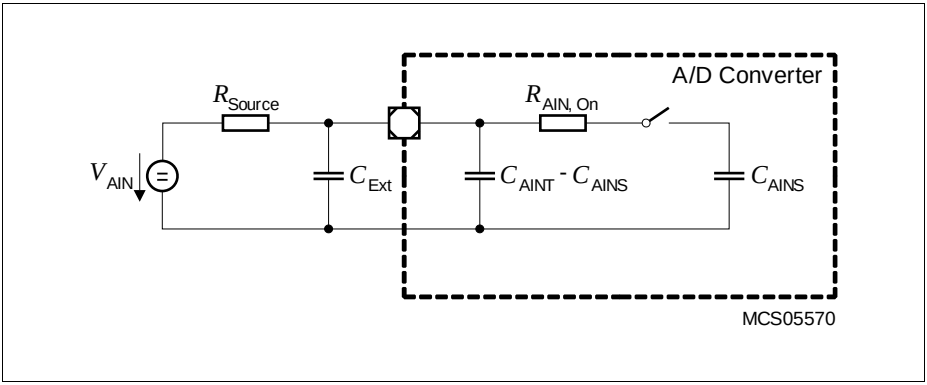

Figure 16 Equivalent Circuitry for Analog Inputs

4.7.4 Pad Properties

The output pad drivers of the XC2336B can operate in several user-selectable modes. Strong driver mode allows controlling external components requiring higher currents such as power bridges or LEDs. Reducing the driving power of an output pad reduces electromagnetic emissions (EME). In strong driver mode, selecting a slower edge reduces EME.

The dynamic behavior, i.e. the rise time and fall time, depends on the applied external capacitance that must be charged and discharged. Timing values are given for a capacitance of 20 pF, unless otherwise noted.

In general, the performance of a pad driver depends on the available supply voltage V_{DDP}. Therefore the following tables list the pad parameters for the upper voltage range and the lower voltage range, respectively.

Note: These parameters are not subject to production test but verified by design and/or characterization.

Note: Operating Conditions apply.

Table 28 is valid under the following conditions: $V_{DDP} \leq$ 5.5 V; V_{DDP}typ. 5 V; $V_{DDP} \geq$ 4.5 V

Table 28 Standard Pad Parameters for Upper Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Maximum output driver current (absolute value)[1)]	I_{Omax} CC	–	–	4.0	mA	Driver_Strength = Medium
		–	–	10	mA	Driver_Strength = Strong
		–	–	0.5	mA	Driver_Strength = Weak
Nominal output driver current (absolute value)	I_{Onom} CC	–	–	1.0	mA	Driver_Strength = Medium
		–	–	2.5	mA	Driver_Strength = Strong
		–	–	0.1	mA	Driver_Strength = Weak

Table 32 USIC SSC Slave Mode Timing for Upper Voltage Range (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Data input DX0 hold time from clock input DX1 receive edge[1)]	t_{13} SR	5	–	–	ns	
Data output DOUT valid time	t_{14} CC	7	–	33	ns	

1) These input timings are valid for asynchronous input signal handling of slave select input, shift clock input, and receive data input (bits DXnCR.DSEN = 0).

Table 33 is valid under the following conditions: C_L= 20 pF; *SSC*= slave ; voltage_range= lower

Table 33 USIC SSC Slave Mode Timing for Lower Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Select input DX2 setup to first clock input DX1 transmit edge[1)]	t_{10} SR	7	–	–	ns	
Select input DX2 hold after last clock input DX1 receive edge[1)]	t_{11} SR	7	–	–	ns	
Receive data input setup time to shift clock receive edge[1)]	t_{12} SR	7	–	–	ns	
Data input DX0 hold time from clock input DX1 receive edge[1)]	t_{13} SR	5	–	–	ns	
Data output DOUT valid time	t_{14} CC	8	–	41	ns	

1) These input timings are valid for asynchronous input signal handling of slave select input, shift clock input, and receive data input (bits DXnCR.DSEN = 0).

4.7.6 Debug Interface Timing

The debugger can communicate with the XC2336B either via the 2-pin DAP interface or via the standard JTAG interface.

Debug via DAP

The following parameters are applicable for communication through the DAP debug interface.

Note: These parameters are not subject to production test but verified by design and/or characterization.

Note: Operating Conditions apply.

Table 34 is valid under the following conditions: C_L= 20 pF; voltage_range= upper

Table 34 DAP Interface Timing for Upper Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
DAP0 clock period[1)]	t_{11} SR	25	–	–	ns	
DAP0 high time	t_{12} SR	8	–	–	ns	
DAP0 low time[1)]	t_{13} SR	8	–	–	ns	
DAP0 clock rise time	t_{14} SR	–	–	4	ns	
DAP0 clock fall time	t_{15} SR	–	–	4	ns	
DAP1 setup to DAP0 rising edge	t_{16} SR	6	–	–	ns	
DAP1 hold after DAP0 rising edge	t_{17} SR	6	–	–	ns	
DAP1 valid per DAP0 clock period[2)]	t_{19} CC	17	20	–	ns	

1) See the DAP chapter for clock rate restrictions in the Active::IDLE protocol state.

2) The Host has to find a suitable sampling point by analyzing the sync telegram response.

Table 35 is valid under the following conditions: C_L= 20 pF; voltage_range= lower

Table 35 DAP Interface Timing for Lower Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
DAP0 clock period[1)]	t_{11} SR	25	–	–	ns	
DAP0 high time	t_{12} SR	8	–	–	ns	
DAP0 low time[1)]	t_{13} SR	8	–	–	ns	
DAP0 clock rise time	t_{14} SR	–	–	4	ns	
DAP0 clock fall time	t_{15} SR	–	–	4	ns	
DAP1 setup to DAP0 rising edge	t_{16} SR	6	–	–	ns	
DAP1 hold after DAP0 rising edge	t_{17} SR	6	–	–	ns	
DAP1 valid per DAP0 clock period[2)]	t_{19} CC	12	17	–	ns	

1) See the DAP chapter for clock rate restrictions in the Active::IDLE protocol state.

2) The Host has to find a suitable sampling point by analyzing the sync telegram response.

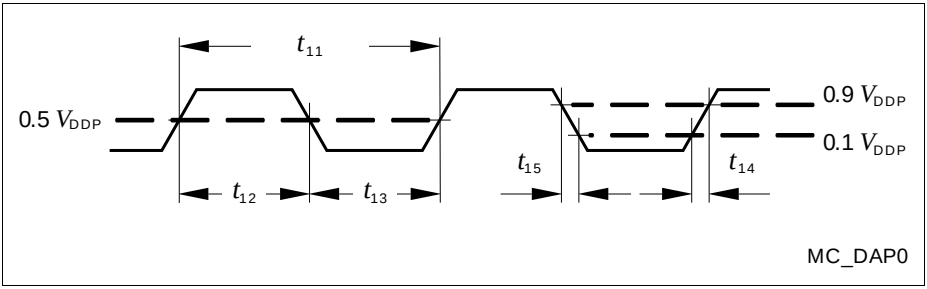

Figure 23 Test Clock Timing (DAP0)

Table 37 JTAG Interface Timing for Lower Voltage Range (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
TDI/TMS hold after TCK rising edge	t_7 SR	6	–	–	ns	
TDO valid from TCK falling edge (propagation delay)[1]	t_8 CC	–	32	36	ns	
TDO high impedance to valid output from TCK falling edge[2][1]	t_9 CC	–	32	36	ns	
TDO valid output to high impedance from TCK falling edge[1]	t_{10} CC	–	32	36	ns	
TDO hold after TCK falling edge[1]	t_{18} CC	5	–	–	ns	

1) The falling edge on TCK is used to generate the TDO timing.

2) The setup time for TDO is given implicitly by the TCK cycle time.

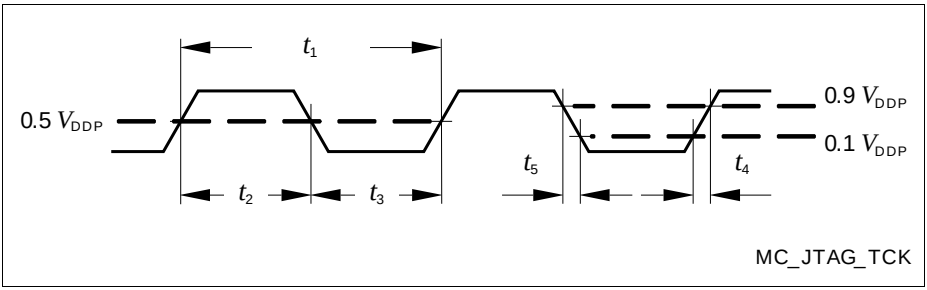

Figure 26 Test Clock Timing (TCK)

5 Package and Reliability

The XC2000 Family devices use the package type PG-LQFP (Plastic Green - Low Profile Quad Flat Package). The following specifications must be regarded to ensure proper integration of the XC2336B in its target environment.

5.1 Packaging

These parameters specify the packaging rather than the silicon.

Table 38 Package Parameters (PG-LQFP-64-6/-24)

Parameter	Symbol	Limit Values		Unit	Notes
		Min.	Max.		
Exposed Pad Dimension	Ex × Ey	–	5.6 × 5.6	mm	–
Power Dissipation	P_{DISS}	–	0.8	W	–
Thermal resistance Junction-Ambient	$R_{\Theta JA}$	–	40	K/W	No thermal via[1)]
			37	K/W	4-layer, no pad[2)]
			25	K/W	4-layer, pad[3)]

1) Device mounted on a 4-layer board without thermal vias; exposed pad not soldered.

2) Device mounted on a 4-layer JEDEC board (according to JESD 51-7) with thermal vias; exposed pad not soldered.

3) Device mounted on a 4-layer JEDEC board (according to JESD 51-7) with thermal vias; exposed pad soldered to the board.

Note: To improve the EMC behavior, it is recommended to connect the exposed pad to the board ground, independent of the thermal requirements.
Board layout examples are given in an application note.

Package Compatibility Considerations

The XC2336B is a member of the XC2000 Family of microcontrollers. It is also compatible to a certain extent with members of similar families or subfamilies.

Each package is optimized for the device it houses. Therefore, there may be slight differences between packages of the same pin-count but for different device types. In particular, the size of the Exposed Pad (if present) may vary.

If different device types are considered or planned for an application, it must be ensured that the board layout fits all packages under consideration.