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Understanding Embedded - DSP (Digital Signal Processors)

Embedded - DSP (Digital Signal Processors) are specialized microprocessors designed to perform complex mathematical computations on digital signals in real-time. Unlike general-purpose processors, DSPs are optimized for high-speed numeric processing tasks, making them ideal for applications that require efficient and precise manipulation of digital data. These processors are fundamental in converting and processing signals in various forms, including audio, video, and communication signals, ensuring that data is accurately interpreted and utilized in embedded systems.

Applications of Embedded - DSP (Digital Signal Processors)

Details	
Product Status	Obsolete
Type	SC140 Core
Interface	DSI, Ethernet, RS-232
Clock Rate	400MHz
Non-Volatile Memory	External
On-Chip RAM	1.436MB
Voltage - I/O	3.30V
Voltage - Core	1.10V
Operating Temperature	-40°C ~ 105°C (TJ)
Mounting Type	Surface Mount
Package / Case	431-BFBGA, FCBGA
Supplier Device Package	431-FCPBGA (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/msc8122tvt6400v

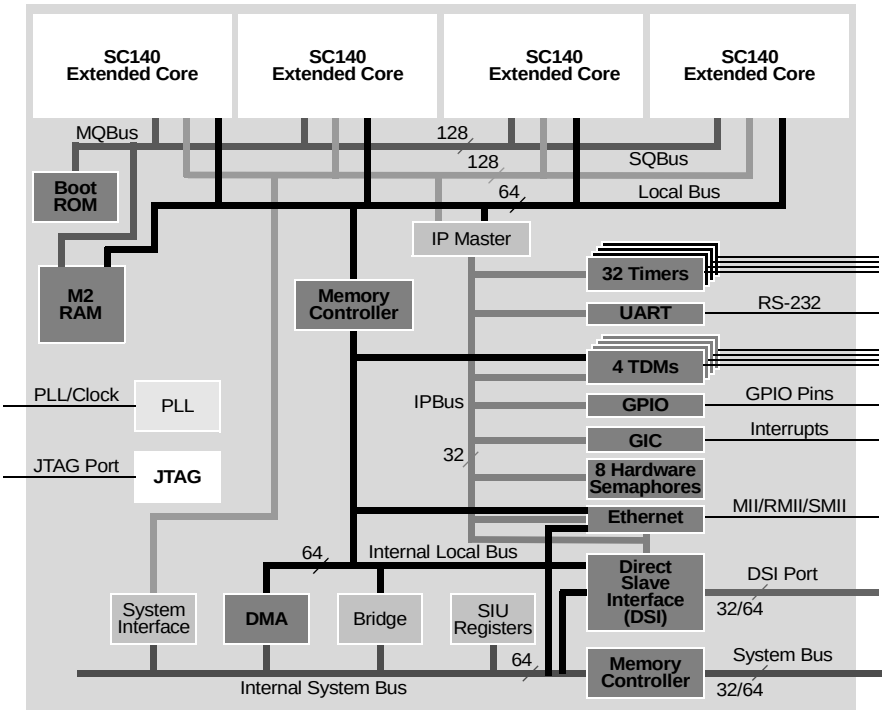
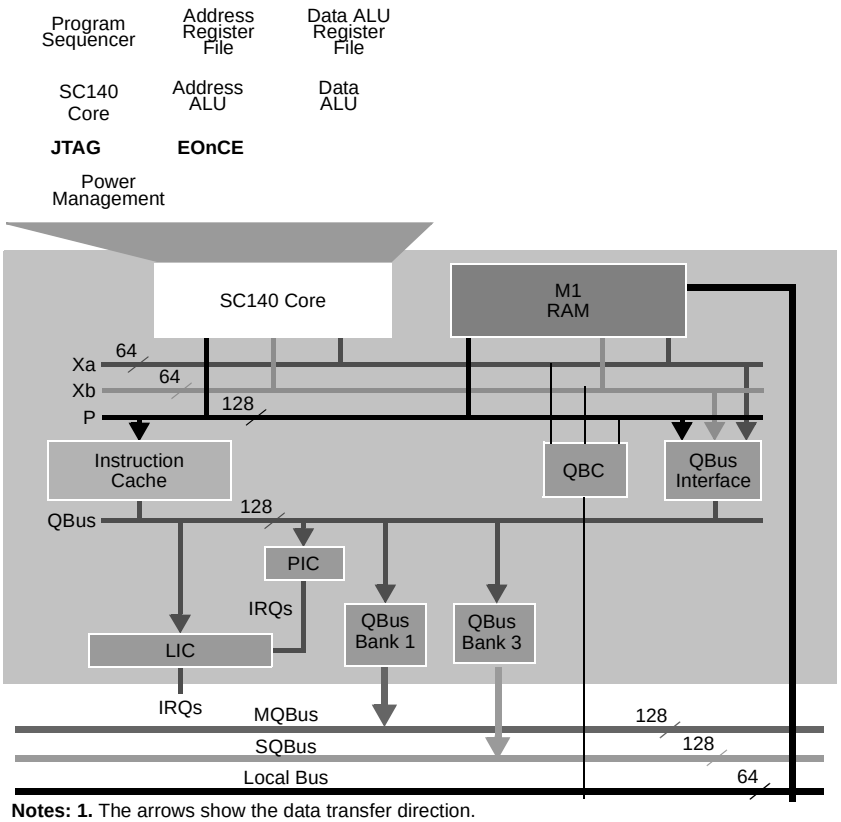


Figure 1. MSC8122 Block Diagram



Notes: 1. The arrows show the data transfer direction.
2. The QBus interface includes a bus switch, write buffer, fetch unit, and a control unit that defines four QBus banks. In addition, the QBC handles internal memory contentions.

Figure 2. StarCore SC140 DSP Extended Core Block Diagram

Top View

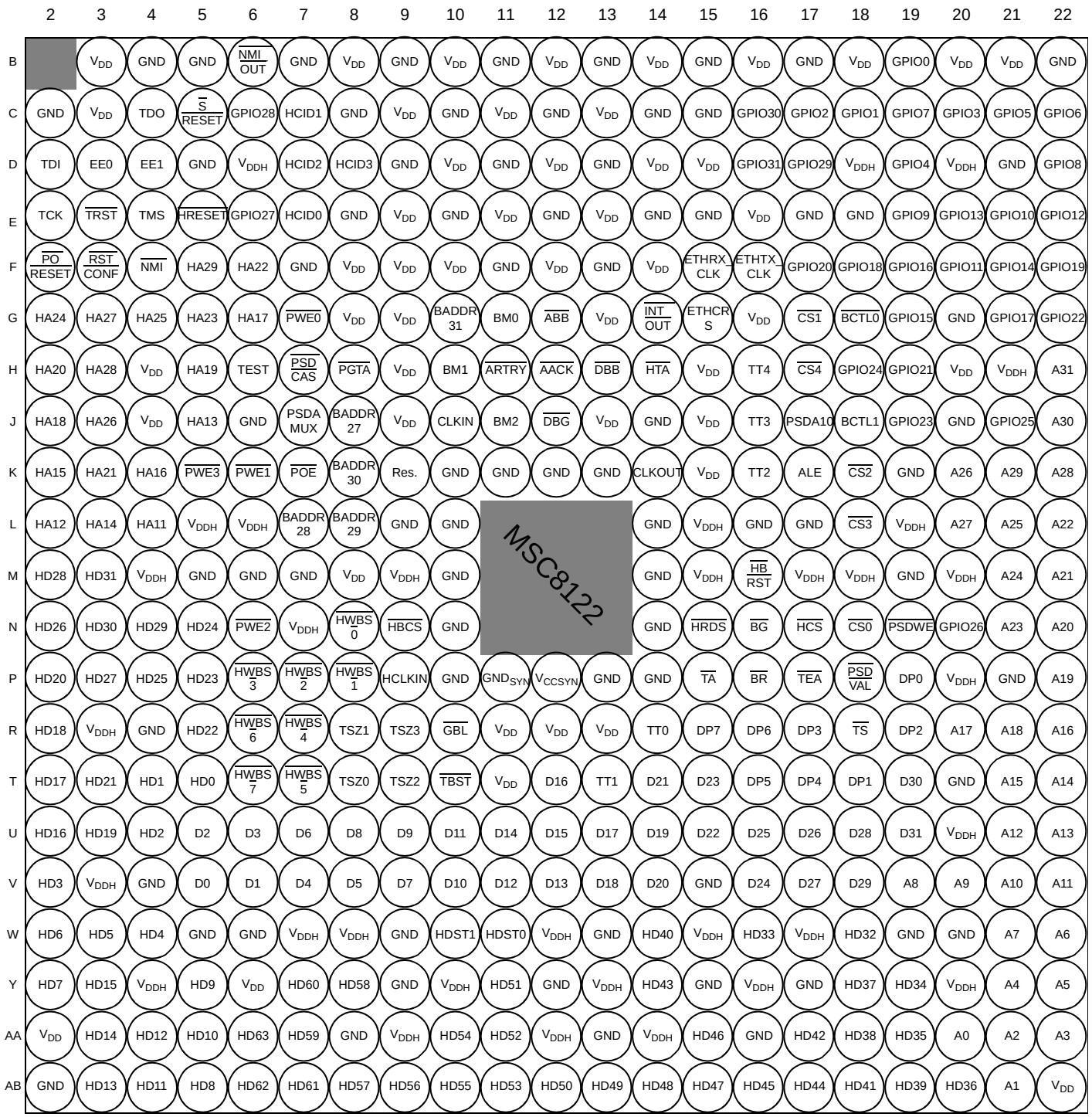


Figure 3. MSC8122 Package, Top View

1.2 Signal List By Ball Location

Table 1 presents signal list sorted by ball number. -

Table 1. MSC8122 Signal Listing by Ball Designator

Des.	Signal Name	Des.	Signal Name
B3	V_{DD}	C18	GPIO1/TIMER0/CHIP_ID1/$\overline{IRQ5}$/ETHTXD1
B4	GND	C19	GPIO7/TDM3RCLK/$\overline{IRQ5}$/ETHTXD3
B5	GND	C20	GPIO3/TDM3TSYN/$\overline{IRQ1}$/ETHTXD2
B6	$\overline{NMI\_OUT}$	C21	GPIO5/TDM3TDAT/$\overline{IRQ3}$/ETHRXD3
B7	GND	C22	GPIO6/TDM3RSYN/$\overline{IRQ4}$/ETHRXD2
B8	V_{DD}	D2	TDI
B9	GND	D3	EE0
B10	V_{DD}	D4	EE1
B11	GND	D5	GND
B12	V_{DD}	D6	V_{DDH}
B13	GND	D7	HCID2
B14	V_{DD}	D8	HCID3/HA8
B15	GND	D9	GND
B16	V_{DD}	D10	V_{DD}
B17	GND	D11	GND
B18	V_{DD}	D12	V_{DD}
B19	GPIO0/CHIP_ID0/$\overline{IRQ4}$/ETHTXD0	D13	GND
B20	V_{DD}	D14	V_{DD}
B21	V_{DD}	D15	V_{DD}
B22	GND	D16	GPIO31/TIMER3/SCL
C2	GND	D17	GPIO29/CHIP_ID3/ETHTX_EN
C3	V_{DD}	D18	V_{DDH}
C4	TDO	D19	GPIO4/TDM3TCLK/$\overline{IRQ2}$/ETHTX_ER
C5	$\overline{SRESET}$	D20	V_{DDH}
C6	GPIO28/UTXD/DREQ2	D21	GND
C7	HCID1	D22	GPIO8/TDM3RDAT/$\overline{IRQ6}$/ETHCOL
C8	GND	E2	TCK
C9	V_{DD}	E3	$\overline{TRST}$
C10	GND	E4	TMS
C11	V_{DD}	E5	$\overline{HRESET}$
C12	GND	E6	GPIO27/URXD/DREQ1
C13	V_{DD}	E7	HCID0
C14	GND	E8	GND
C15	GND	E9	V_{DD}
C16	GPIO30/TIMER2/TMCLK/SDA	E10	GND
C17	GPIO2/TIMER1/CHIP_ID2/$\overline{IRQ6}$	E11	V_{DD}

Table 1. MSC8122 Signal Listing by Ball Designator (continued)

Des.	Signal Name	Des.	Signal Name
H21	V_{DDH}	K15	V_{DD}
H22	A31	K16	TT2/$\overline{CS5}$
J2	HA18	K17	ALE
J3	HA26	K18	$\overline{CS2}$
J4	V_{DD}	K19	GND
J5	HA13	K20	A26
J6	GND	K21	A29
J7	PSDAMUX/PGPL5	K22	A28
J8	BADDR27	L2	HA12
J9	V_{DD}	L3	HA14
J10	CLKIN	L4	HA11
J11	BM2/TC2/BNKSEL2	L5	V_{DDH}
J12	$\overline{DBG}$	L6	V_{DDH}
J13	V_{DD}	L7	BADDR28
J14	GND	L8	$\overline{IRQ5}$/BADDR29
J15	V_{DD}	L9	GND
J16	TT3/$\overline{CS6}$	L10	GND
J17	PSDA10/PGPL0	L14	GND
J18	$\overline{BCTL1}$/$\overline{CS5}$	L15	V_{DDH}
J19	GPIO23/TDM0TDAT/$\overline{IRQ13}$	L16	GND
J20	GND	L17	GND
J21	GPIO25/TDM0RCLK/$\overline{IRQ15}$	L18	$\overline{CS3}$
J22	A30	L19	V_{DDH}
K2	HA15	L20	A27
K3	HA21	L21	A25
K4	HA16	L22	A22
K5	$\overline{PWE3}$/$\overline{PSDDQM3}$/$\overline{PBS3}$	M2	HD28
K6	$\overline{PWE1}$/$\overline{PSDDQM1}$/$\overline{PBS1}$	M3	HD31
K7	$\overline{POE}$/$\overline{PSDRAS}$/PGPL2	M4	V_{DDH}
K8	$\overline{IRQ2}$/BADDR30	M5	GND
K9	Reserved	M6	GND
K10	GND	M7	GND
K11	GND	M8	V_{DD}
K12	GND	M9	V_{DDH}
K13	GND	M10	GND
K14	CLKOUT	M14	GND

Table 1. MSC8122 Signal Listing by Ball Designator (continued)

Des.	Signal Name	Des.	Signal Name
M15	V_{DDH}	P12	V_{CCSYN}
M16	$\overline{\text{HBRST}}$	P13	GND
M17	V_{DDH}	P14	GND
M18	V_{DDH}	P15	$\overline{\text{TA}}$
M19	GND	P16	$\overline{\text{BR}}$
M20	V_{DDH}	P17	$\overline{\text{TEA}}$
M21	A24	P18	$\overline{\text{PSDVAL}}$
M22	A21	P19	DP0/DREQ1/$\overline{\text{EXT\_BR2}}$
N2	HD26	P20	V_{DDH}
N3	HD30	P21	GND
N4	HD29	P22	A19
N5	HD24	R2	HD18
N6	$\overline{\text{PWE2}}$/$\overline{\text{PSDDQM2}}$/$\overline{\text{PBS2}}$	R3	V_{DDH}
N7	V_{DDH}	R4	GND
N8	$\overline{\text{HWBS0}}$/$\overline{\text{HDBS0}}$/$\overline{\text{HWBE0}}$/$\overline{\text{HDBE0}}$	R5	HD22
N9	$\overline{\text{HBCS}}$	R6	$\overline{\text{HWBS6}}$/$\overline{\text{HDBS6}}$/$\overline{\text{HWBE6}}$/$\overline{\text{HDBE6}}$/$\overline{\text{PWE6}}$/$\overline{\text{PSDDQM6}}$/$\overline{\text{PBS6}}$
N10	GND	R7	$\overline{\text{HWBS4}}$/$\overline{\text{HDBS4}}$/$\overline{\text{HWBE4}}$/$\overline{\text{HDBE4}}$/$\overline{\text{PWE4}}$/$\overline{\text{PSDDQM4}}$/$\overline{\text{PBS4}}$
N14	GND	R8	TSZ1
N15	$\overline{\text{HRDS}}$/HRW/$\overline{\text{HRDE}}$	R9	TSZ3
N16	$\overline{\text{BG}}$	R10	$\overline{\text{IRQ1}}$/$\overline{\text{GBL}}$
N17	$\overline{\text{HCS}}$	R11	V_{DD}
N18	$\overline{\text{CS0}}$	R12	V_{DD}
N19	$\overline{\text{PSDWE}}$/PGPL1	R13	V_{DD}
N20	GPIO26/TDM0RDAT	R14	TT0/HA7
N21	A23	R15	$\overline{\text{IRQ7}}$/DP7/DREQ4
N22	A20	R16	$\overline{\text{IRQ6}}$/DP6/DREQ3
P2	HD20	R17	$\overline{\text{IRQ3}}$/DP3/DREQ2/$\overline{\text{EXT\_BR3}}$
P3	HD27	R18	$\overline{\text{TS}}$
P4	HD25	R19	$\overline{\text{IRQ2}}$/DP2/$\overline{\text{DACK2}}$/$\overline{\text{EXT\_DBG2}}$
P5	HD23	R20	A17
P6	$\overline{\text{HWBS3}}$/$\overline{\text{HDBS3}}$/$\overline{\text{HWBE3}}$/$\overline{\text{HDBE3}}$	R21	A18
P7	$\overline{\text{HWBS2}}$/$\overline{\text{HDBS2}}$/$\overline{\text{HWBE2}}$/$\overline{\text{HDBE2}}$	R22	A16
P8	$\overline{\text{HWBS1}}$/$\overline{\text{HDBS1}}$/$\overline{\text{HWBE1}}$/$\overline{\text{HDBE1}}$	T2	HD17
P9	HCLKIN	T3	HD21
P10	GND	T4	HD1/DSISYNC
P11	GND_{SYN}	T5	HD0/SWTE

Table 1. MSC8122 Signal Listing by Ball Designator (continued)

Des.	Signal Name	Des.	Signal Name
T6	HWBS7/HDBS7/HWBE7/HDBE7/PWE7/PSDDQM7/PBS7	U21	A12
T7	HWBS5/HDBS5/HWBE5/HDBE5/PWE5/PSDDQM5/PBS5	U22	A13
T8	TSZ0	V2	HD3/MODCK1
T9	TSZ2	V3	V_{DDH}
T10	TBST	V4	GND
T11	V_{DD}	V5	D0
T12	D16	V6	D1
T13	TT1	V7	D4
T14	D21	V8	D5
T15	D23	V9	D7
T16	IRQ5/DP5/DACK4/EXT_BG3	V10	D10
T17	IRQ4/DP4/DACK3/EXT_DBG3	V11	D12
T18	IRQ1/DP1/DACK1/EXT_BG2	V12	D13
T19	D30	V13	D18
T20	GND	V14	D20
T21	A15	V15	GND
T22	A14	V16	D24
U2	HD16	V17	D27
U3	HD19	V18	D29
U4	HD2/DSI64	V19	A8
U5	D2	V20	A9
U6	D3	V21	A10
U7	D6	V22	A11
U8	D8	W2	HD6
U9	D9	W3	HD5/CNFGS
U10	D11	W4	HD4/MODCK2
U11	D14	W5	GND
U12	D15	W6	GND
U13	D17	W7	V_{DDH}
U14	D19	W8	V_{DDH}
U15	D22	W9	GND
U16	D25	W10	HDST1/HA10
U17	D26	W11	HDST0/HA9
U18	D28	W12	V_{DDH}
U19	D31	W13	GND
U20	V_{DDH}	W14	HD40/D40/ETHRXD0

2 Electrical Characteristics

This document contains detailed information on power considerations, DC/AC electrical characteristics, and AC timing specifications. For additional information, see the *MSC8122 Reference Manual*.

2.1 Maximum Ratings

CAUTION

This device contains circuitry protecting against damage due to high static voltage or electrical fields; however, normal precautions should be taken to avoid exceeding maximum voltage ratings. Reliability is enhanced if unused inputs are tied to an appropriate logic voltage level (for example, either GND or V_{DD}).

In calculating timing requirements, adding a maximum value of one specification to a minimum value of another specification does not yield a reasonable sum. A maximum specification is calculated using a worst case variation of process parameter values in one direction. The minimum specification is calculated using the worst case for the same parameters in the opposite direction. Therefore, a "maximum" value for a specification never occurs in the same device with a "minimum" value for another specification; adding a maximum to a minimum represents a condition that can never exist.

Table 2 describes the maximum electrical ratings for the MSC8122.

Table 2. Absolute Maximum Ratings

Rating	Symbol	Value	Unit
Core and PLL supply voltage	V_{DD}	–0.2 to 1.6	V
I/O supply voltage	V_{DDH}	–0.2 to 4.0	V
Input voltage	V_{IN}	–0.2 to 4.0	V
Maximum operating temperature: • Standard range • Extended range	T_J	 90 105	 °C °C
Minimum operating temperature • Standard range • Extended range	T_J	 0 –40	 °C °C
Storage temperature range	T_{STG}	–55 to +150	°C

Notes:
1. Functional operating conditions are given in **Table 3.**
2. Absolute maximum ratings are stress ratings only, and functional operation at the maximum is not guaranteed. Stress beyond the listed limits may affect device reliability or cause permanent damage.
3. **Section 3.5**, *Thermal Considerations* includes a formula for computing the chip junction temperature (T_J).

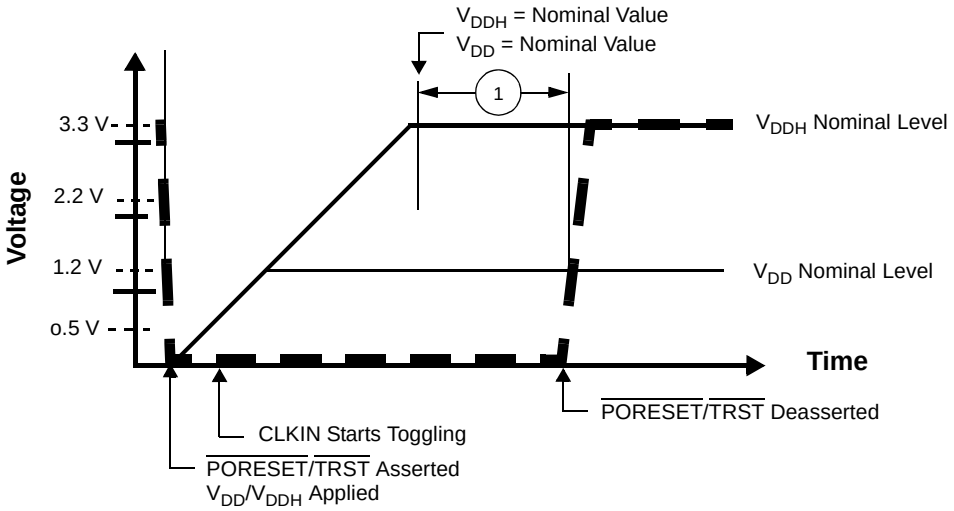


Figure 6. Start-Up Sequence: V_{DD} and V_{DDH} Raised Together

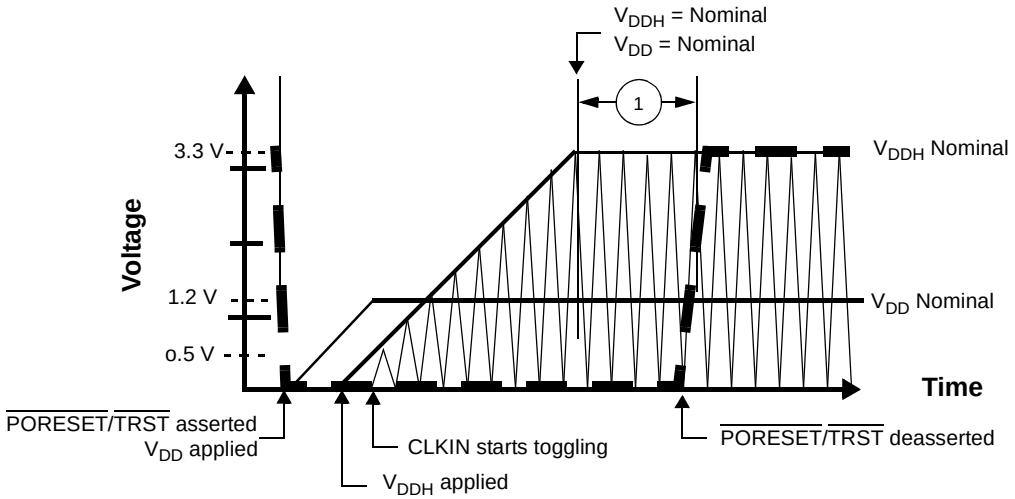


Figure 7. Start-Up Sequence: V_{DD} Raised Before V_{DDH} with CLKIN Started with V_{DDH}

2.5.4.3 Reset Timing Tables

Table 12 and **Figure 9** describe the reset timing for a reset configuration write through the direct slave interface (DSI) or through the system bus.

Table 12. Timing for a Reset Configuration Write through the DSI or System Bus

No.	Characteristics	Expression	Min	Max	Unit
1	Required external PORESET duration minimum • CLKIN = 20 MHz • CLKIN = 100 MHz (300 MHz core) • CLKIN = 133 MHz (400 MHz core) • CLKIN = 166 MHz (500 MHz core)	16/CLKIN	 800 160 120 96	 — — — —	 ns ns ns ns
2	Delay from deassertion of external PORESET to deassertion of internal PORESET • CLKIN = 20 MHz to 166 MHz	1024/CLKIN	 6.17	 51.2	 µs
3	Delay from de-assertion of internal PORESET to SPLL lock • CLKIN = 20 MHz (RDF = 1) • CLKIN = 100 MHz (RDF = 1) (300 MHz core) • CLKIN = 133 MHz (RDF = 2) (400 MHz core) • CLKIN = 166 MHz (RDF = 2) (500 MHz core)	6400/(CLKIN/RDF) (PLL reference clock-division factor)	 320 64 96 77	 320 64 96 77	 µs µs µs µs
5	Delay from SPLL to HRESET deassertion • REFCLK = 40 MHz to 166 MHz	512/REFCLK	3.08	12.8	µs
6	Delay from SPLL lock to SRESET deassertion • REFCLK = 40 MHz to 166 MHz	515/REFCLK	3.10	12.88	µs
7	Setup time from assertion of RSTCONF, CNFGS, DSISYNC, DSI64, CHIP_ID[0–3], BM[0–2], SWTE, and MODCK[1–2] before deassertion of PORESET		3	—	ns
8	Hold time from deassertion of PORESET to deassertion of RSTCONF, CNFGS, DSISYNC, DSI64, CHIP_ID[0–3], BM[0–2], SWTE, and MODCK[1–2]		5	—	ns
Note:	Timings are not tested, but are guaranteed by design.				

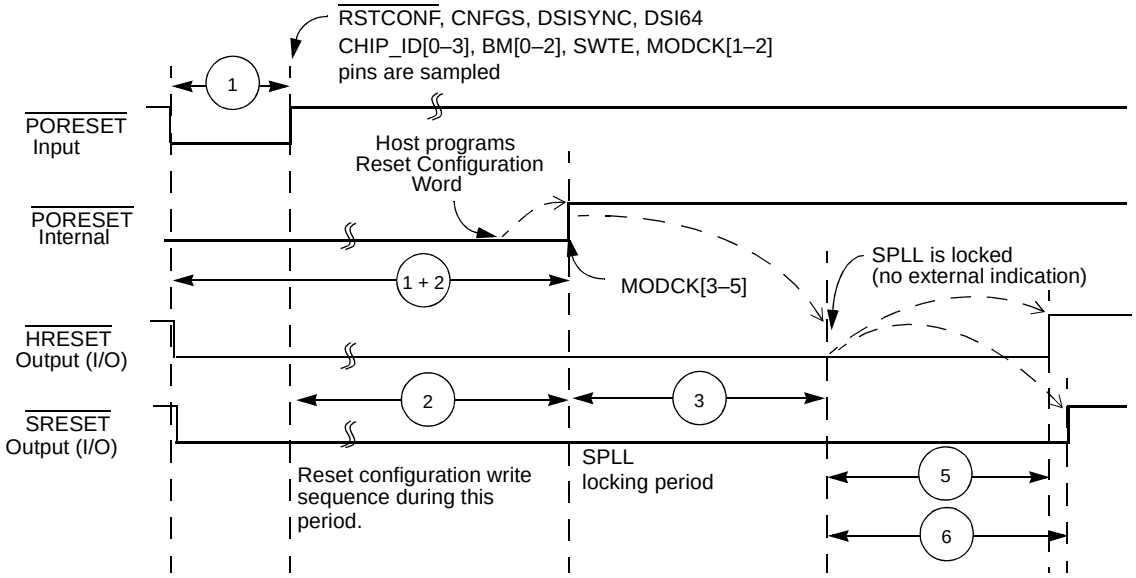


Figure 9. Timing Diagram for a Reset Configuration Write

Table 15. AC Timing for SIU Outputs

No.	Characteristic	Value for Bus Speed in MHz[3]				Units
		Ref = CLKIN			Ref = CLKOUT	
		1.1 V	1.2 V	1.2 V	1.2 V	
		100/133	133	166	100/133	
30[2]	Minimum delay from the 50% level of the REFCLK for all signals	0.9	0.8	0.8	1.0	ns
31	$\overline{\text{PSDVAL}}/\overline{\text{TEA}}/\overline{\text{TA}}$ max delay from the 50% level of the REFCLK rising edge	6.0	4.9	4.9	5.8	ns
32a	Address bus max delay from the 50% level of the REFCLK rising edge • Multi-master mode (SIUBCR[EBM] = 1) • Single-master mode (SIUBCR[EBM] = 0)	 6.4 5.3	 5.5 4.2	 5.5 3.9	 6.4 5.1	 ns ns
32b	Address attributes: TT[0–1]/$\overline{\text{TBST}}$/TSZ/$\overline{\text{GBL}}$ max delay from the 50% level of the REFCLK rising edge	6.4	5.1	5.1	6.0	ns
32c	Address attributes: TT[2–4]/TC max delay from the 50% level of the REFCLK rising edge	6.9	5.7	5.7	6.6	ns
32d	BADDR max delay from the 50% level of the REFCLK rising edge	5.2	4.2	4.2	5.1	ns
33a	Data bus max delay from the 50% level of the REFCLK rising edge • Data-pipeline mode • Non-pipeline mode	 4.8 7.1	 3.9 6.1	 3.7 6.1	 4.8 7.0	 ns ns
33b	DP max delay from the 50% level of the REFCLK rising edge • Data-pipeline mode • Non-pipeline mode	 6.0 7.5	 5.3 6.5	 5.3 6.5	 6.2 7.4	 ns ns
34	Memory controller signals/ALE/$\overline{\text{CS}}$[0–4] max delay from the 50% level of the REFCLK rising edge	5.1	4.2	3.9	5.1	ns
35a	$\overline{\text{DBG}}/\overline{\text{BG}}/\overline{\text{BR}}/\overline{\text{DBB}}$ max delay from the 50% level of the REFCLK rising edge	6.0	4.7	4.7	5.6	ns
35b	$\overline{\text{AACK}}/\overline{\text{ABB}}/\overline{\text{TS}}/\overline{\text{CS}}$[5–7] max delay from the 50% level of the REFCLK rising edge	5.5	4.5	4.5	5.4	ns

Notes:
1. Values are measured from the 50% level of the REFCLK rising edge to the 50% signal level and assume a 20 pF load except where otherwise specified.
2. Except for specification 30, which is specified for a 10 pF load, all timings in this table are specified for a 20 pF load. Decreasing the load results in a timing decrease at the rate of 0.3 ns per 5 pF decrease in load. Increasing the load results in a timing increase at the rate of 0.15 ns per 5 pF increase in load.
3. The maximum bus frequency depends on the mode:
   • In 60x-compatible mode connected to another MSC8122 device, the frequency is determined by adding the input and output longest timing values, which results in the total delay for 20 pF output capacitance. You must also account for other influences that can affect timing, such as on-board clock skews, on-board noise delays, and so on.
   • In single-master mode, the frequency depends on the timing of the devices connected to the MSC8122.
   • To achieve maximum performance on the bus in single-master mode, disable the $\overline{\text{DBB}}$ signal by writing a 1 to the SIUMCR[BDD] bit. See the SIU chapter in the *MSC8122 Reference Manual* for details.

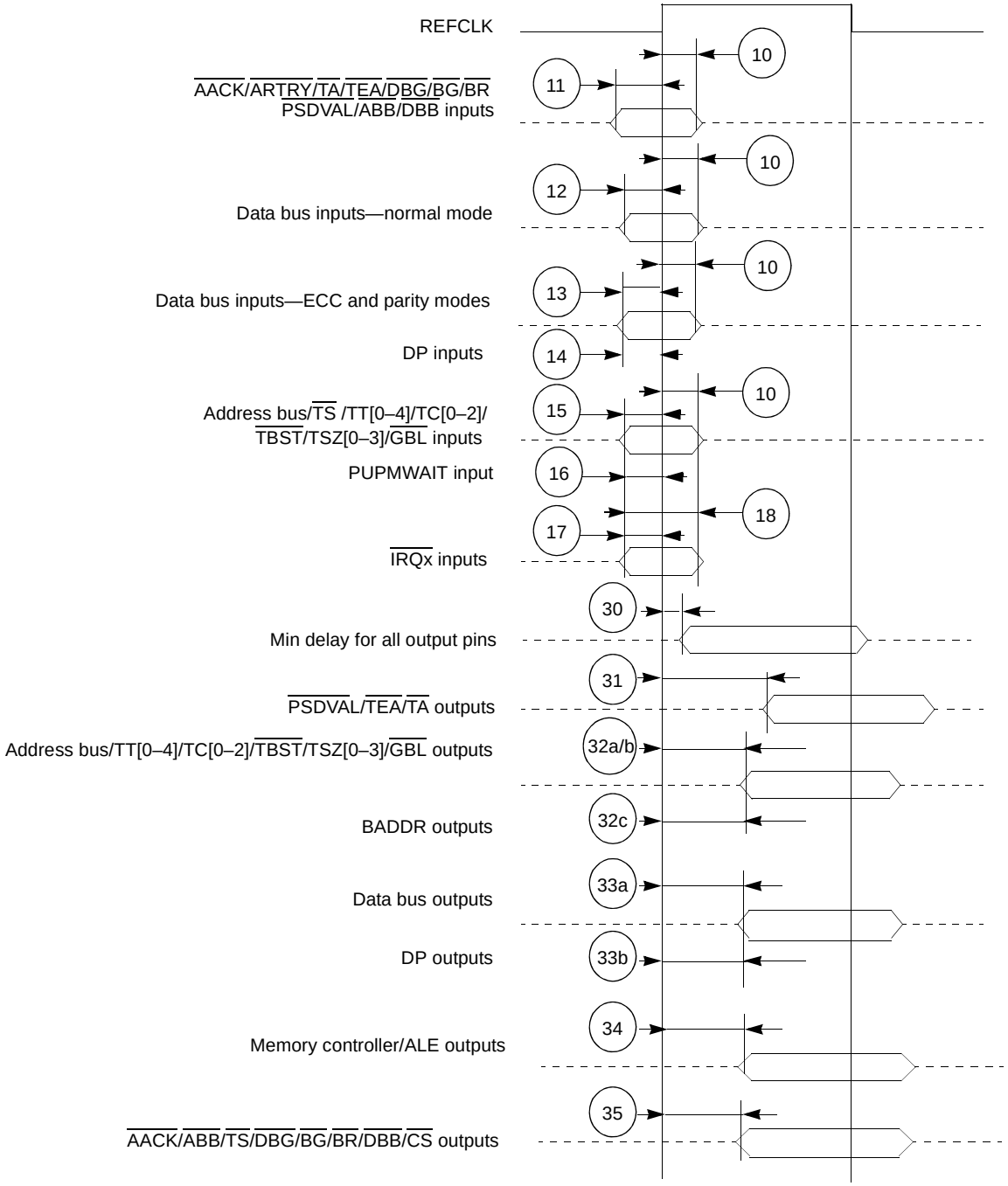


Figure 11. SIU Timing Diagram

2.5.5.3 DMA Data Transfers

Table 17 describes the DMA signal timing.

Table 17. DMA Signals

No.	Characteristic	Ref = CLKIN		Ref = CLKOUT (1.2 V only)		Units
		Min	Max	Min	Max	
37	DREQ set-up time before the 50% level of the falling edge of REFCLK	5.0	—	5.0	—	ns
38	DREQ hold time after the 50% level of the falling edge of REFCLK	0.5	—	0.5	—	ns
39	$\overline{\text{DONE}}$ set-up time before the 50% level of the rising edge of REFCLK	5.0	—	5.0	—	ns
40	$\overline{\text{DONE}}$ hold time after the 50% level of the rising edge of REFCLK	0.5	—	0.5	—	ns
41	$\overline{\text{DACK}}$/$\overline{\text{DRACK}}$/$\overline{\text{DONE}}$ delay after the 50% level of the REFCLK rising edge	0.5	7.5	0.5	8.4	ns

The DREQ signal is synchronized with REFCLK. To achieve fast response, a synchronized peripheral should assert DREQ according to the timings in **Table 17**. **Figure 13** shows synchronous peripheral interaction.

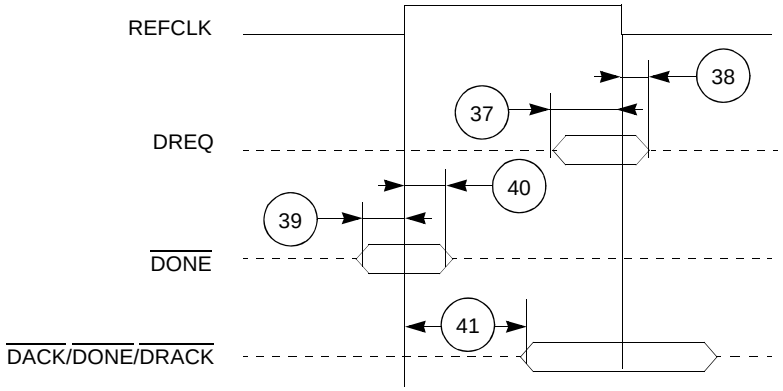
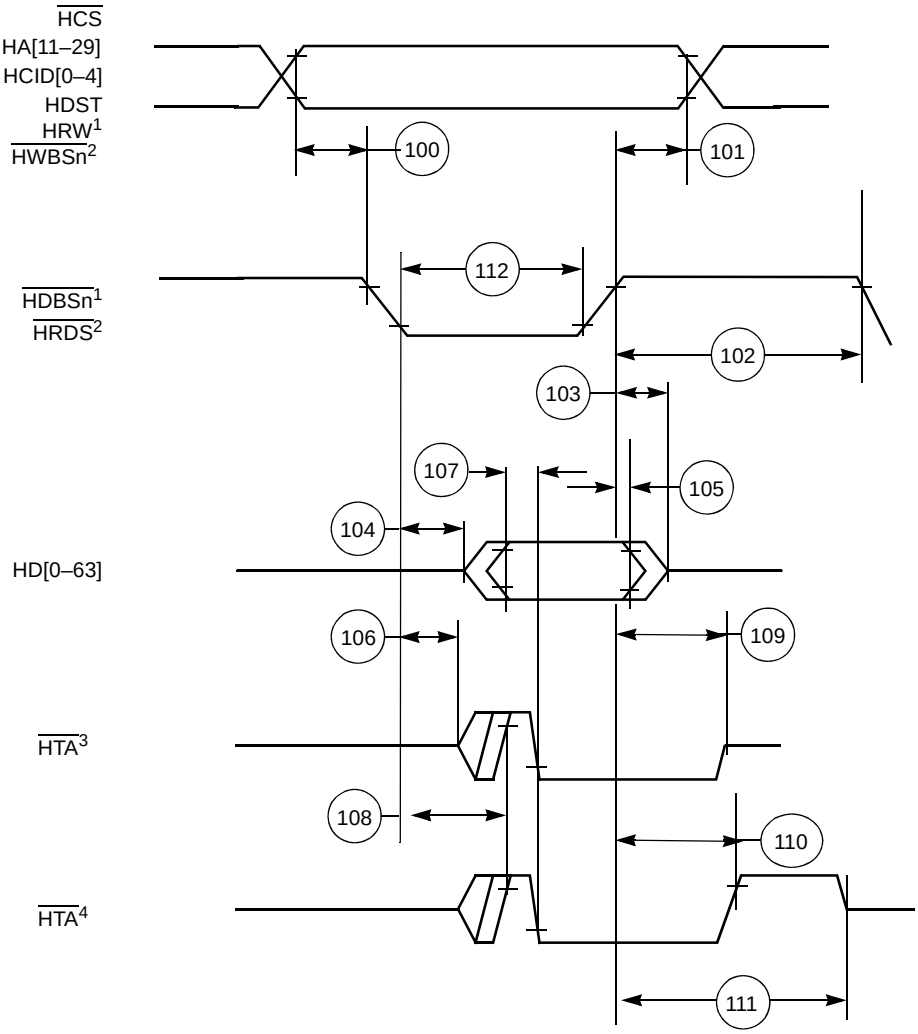


Figure 13. DMA Signals

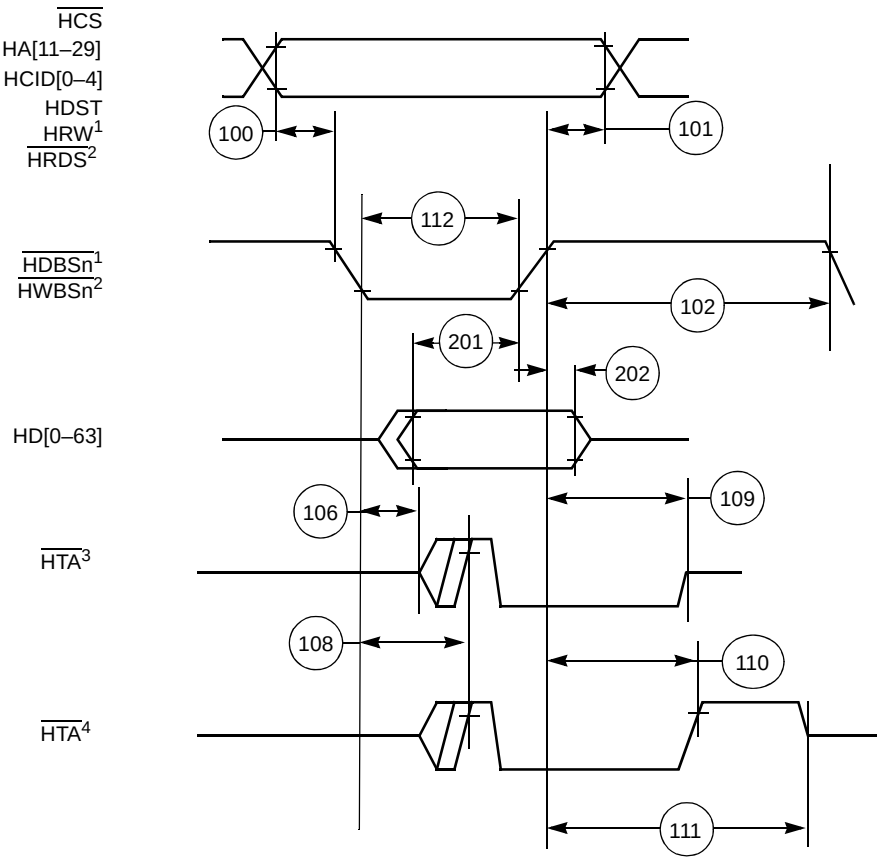
Figure 14 shows DSI asynchronous read signals timing.



Notes:
1. Used for single-strobe mode access.
2. Used for dual-strobe mode access.
3. HTA released at logic 0 (DCR[HTAAD] = 0) at end of access; used with pull-down implementation.
4. HTA released at logic 1 (DCR[HTAAD] = 1) at end of access; used with pull-up implementation.

Figure 14. Asynchronous Single- and Dual-Strobe Modes Read Timing Diagram

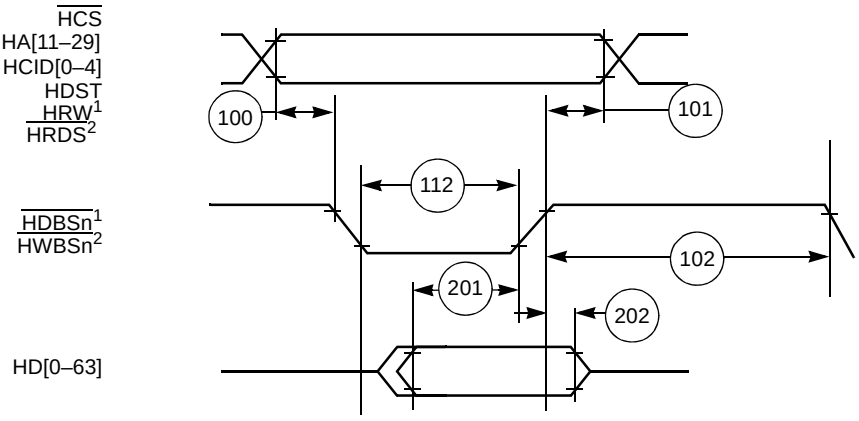
Figure 15 shows DSI asynchronous write signals timing.



Notes:
1. Used for single-strobe mode access.
2. Used for dual-strobe mode access.
3. HTA released at logic 0 (DCR[HTAAD] = 0) at end of access; used with pull-down implementation.
4. HTA released at logic 1 (DCR[HTAAD] = 1) at end of access; used with pull-up implementation.

Figure 15. Asynchronous Single- and Dual-Strobe Modes Write Timing Diagram

Figure 16 shows DSI asynchronous broadcast write signals timing.



Notes:
1. Used for single-strobe mode access.
2. Used for dual-strobe mode access.

Figure 16. Asynchronous Broadcast Write Timing Diagram

2.5.7 TDM Timing

Table 21. TDM Timing

No.	Characteristic	Expression	1.1 V Core		1.2 V Core		Units
			Min	Max	Min	Max	
300	TDMxRCLK/TDMxTCLK	TC[1]	16	—	16	—	ns
301	TDMxRCLK/TDMxTCLK high pulse width	(0.5 ± 0.1) × TC	7	—	7	—	ns
302	TDMxRCLK/TDMxTCLK low pulse width	(0.5 ± 0.1) × TC	7	—	7	—	ns
303	TDM receive all input set-up time		1.3	—	1.3	—	ns
304	TDM receive all input hold time		1.0	—	1.0	—	ns
305	TDMxTCLK high to TDMxTDAT/TDMxRCLK output active[2,3]		2.8	—	2.8	—	ns
306	TDMxTCLK high to TDMxTDAT/TDMxRCLK output		—	10.0	—	8.8	ns
307	All output hold time[4]		2.5	—	2.5	—	ns
308	TDMxTCLK high to TDmXTDAT/TDMxRCLK output high impedance[2,3]		—	10.7	—	10.5	ns
309	TDMxTCLK high to TDMXTSYN output valid[2]		—	9.7	—	8.5	ns
310	TDMxTSYN output hold time[4]		2.5	—	2.5	—	ns

Notes:
1. Values are based on a a maximum frequency of 62.5 MHz. The TDM interface supports any frequency below 62.5 MHz. Devices operating at 300 MHz are limited to a maximum TDMxRCLK/TDMxTCLK frequency of 50 MHz.
2. Values are based on 20 pF capacitive load.
3. When configured as an output, TDMxRCLK acts as a second data link. See the *MSC8122 Reference Manual* for details.
4. Values are based on 10 pF capacitive load.

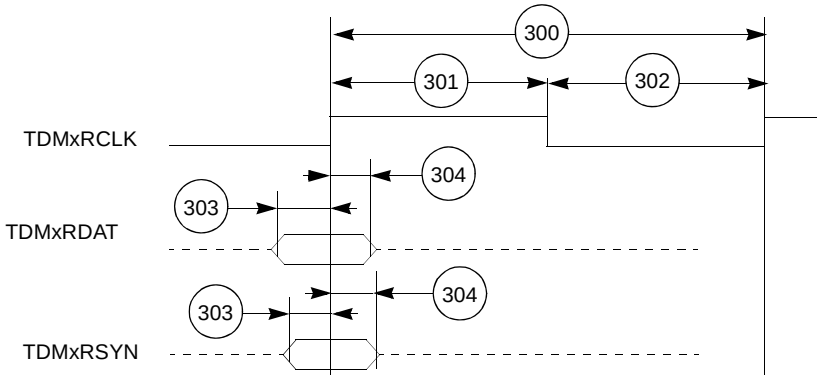


Figure 18. TDM Inputs Signals

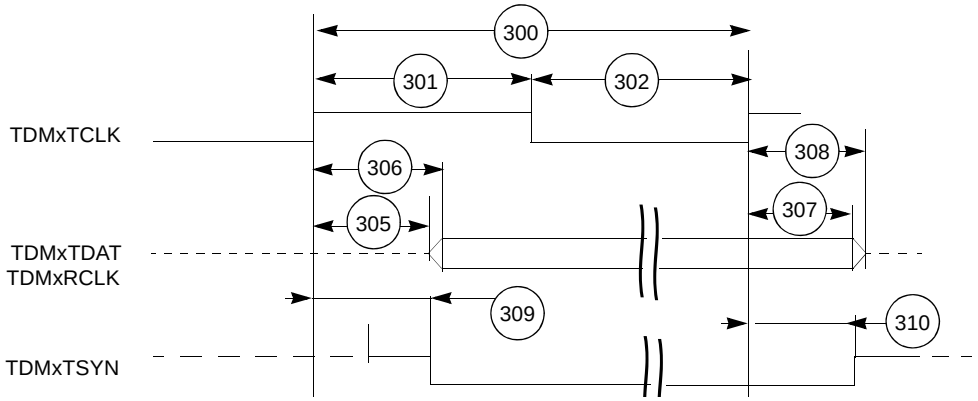


Figure 19. TDM Output Signals

- Never allow V_{DD} to exceed V_{DDH} + 0.8V.
- Design the V_{DDH} supply to prevent reverse current flow by adding a minimum 10 Ω resistor to GND to limit the current. Such a design yields an initial V_{DDH} level of V_{DD} – 0.8 V before it is enabled.

After power-up, V_{DDH} must not exceed V_{DD}/V_{CCSYN} by more than 2.6 V.

3.2 Power Supply Design Considerations

When used as a drop-in replacement in MSC8102 applications or when implementing a new design, use the guidelines described in *Migrating Designs from the MSC8102 to the MSC8122* (AN2716) and the *MSC8122 Design Checklist* (AN3374 for optimal system performance. *MSC8122 and MSC8126 Power Circuit Design Recommendations and Examples* (AN2937) provides detailed design information. See **Section 2.5.2** for start-up timing specifications.

Figure 33 shows the recommended power decoupling circuit for the core power supply. The voltage regulator and the decoupling capacitors should supply the required device current without any drop in voltage on the device pins. The voltage on the package pins should not drop below the minimum specified voltage level even for a very short spikes. This can be achieved by using the following guidelines:

- For the core supply, use a voltage regulator rated at 1.2 V with nominal rating of at least 3 A. This rating does not reflect actual average current draw, but is recommended because it resists changes imposed by transient spikes and has better voltage recovery time than supplies with lower current ratings.
- Decouple the supply using low-ESR capacitors mounted as close as possible to the socket. **Figure 33** shows three capacitors in parallel to reduce the resistance. Three capacitors is a recommended minimum number. If possible, mount at least one of the capacitors directly below the MSC8122 device.

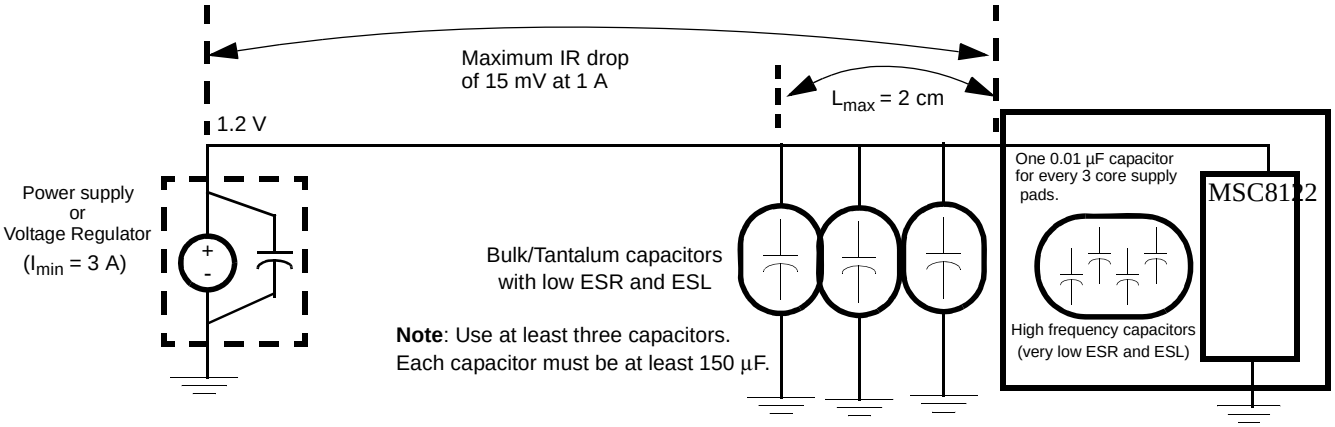


Figure 33. Core Power Supply Decoupling

Each V_{CC} and V_{DD} pin on the MSC8122 device should have a low-impedance path to the board power supply. Similarly, each GND pin should have a low-impedance path to the ground plane. The power supply pins drive distinct groups of logic on the chip. The V_{CC} power supply should have at least four 0.1 µF by-pass capacitors to ground located as closely as possible to the four sides of the package. The capacitor leads and associated printed circuit traces connecting to chip V_{CC}, V_{DD}, and GND should be kept to less than half an inch per capacitor lead. A four-layer board is recommended, employing two inner layers as V_{CC} and GND planes.

All output pins on the MSC8122 have fast rise and fall times. PCB trace interconnection length should be minimized to minimize undershoot and reflections caused by these fast output switching times. This recommendation particularly applies to the address and data buses. Maximum PCB trace lengths of six inches are recommended. For the DSI control signals in synchronous mode, ensure that the layout supports the DSI AC timing requirements and minimizes any signal crosstalk. Capacitance calculations should consider all device loads as well as parasitic capacitances due to the PCB traces. Attention to proper PCB layout and bypassing becomes especially critical in systems with higher capacitive loads because these loads create higher transient currents in the V_{CC}, V_{DD}, and GND circuits. Pull up all unused inputs or signals that will be inputs during reset.

Special care should be taken to minimize the noise levels on the PLL supply pins. There is one pair of PLL supply pins: V_{CCSYN}-GND_{SYN}. To ensure internal clock stability, filter the power to the V_{CCSYN} input with a circuit similar to the one in

Figure 34. For optimal noise filtering, place the circuit as close as possible to V_{CCSYN}. The 0.01-µF capacitor should be closest to V_{CCSYN}, followed by the 10-µF capacitor, the 10-nH inductor, and finally the 10-Ω resistor to V_{DD}. These traces should be kept short and direct. Provide an extremely low impedance path to the ground plane for GND_{SYN}. Bypass GND_{SYN} to V_{CCSYN} by a 0.01-µF capacitor located as close as possible to the chip package. For best results, place this capacitor on the backside of the PCB aligned with the depopulated void on the MSC8122 located in the square defined by positions, L11, L12, L13, M11, M12, M13, N11, N12, and N13.

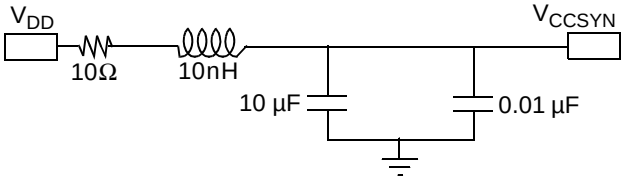


Figure 34. V_{CCSYN} Bypass

3.3 Connectivity Guidelines

Unused output pins can be disconnected, and unused input pins should be connected to the non-active value, via resistors to V_{DDH} or GND, except for the following:

- If the DSI is unused (DDR[DSIDIS] is set), $\overline{\text{HCS}}$ and $\overline{\text{HBCS}}$ must pulled up and all the rest of the DSI signals can be disconnected.
- When the DSI uses synchronous mode, $\overline{\text{HTA}}$ must be pulled up. In asynchronous mode, $\overline{\text{HTA}}$ should be pulled either up or down, depending on design requirements.
- $\overline{\text{HDST}}$ can be disconnected if the DSI is in big-endian mode, or if the DSI is in little-endian mode and the DCR[DSRFA] bit is set.
- When the DSI is in 64-bit data bus mode and DCR[BEM] is cleared, pull up $\overline{\text{HWBS[1–3]}}$/$\overline{\text{HDBS[1–3]}}$/$\overline{\text{HWBE[1–3]}}$/$\overline{\text{HDBE[1–3]}}$ and $\overline{\text{HWBS[4–7]}}$/$\overline{\text{HDBS[4–7]}}$/$\overline{\text{HWBE[4–7]}}$/$\overline{\text{HDBE[4–7]}}$/$\overline{\text{PWE[4–7]}}$/$\overline{\text{PSDDQM[4–7]}}$/$\overline{\text{PBS[4–7]}}$.
- When the DSI is in 32-bit data bus mode and DCR[BEM] is cleared, $\overline{\text{HWBS[1–3]}}$/$\overline{\text{HDBS[1–3]}}$/$\overline{\text{HWBE[1–3]}}$/$\overline{\text{HDBE[1–3]}}$ must be pulled up.
- When the DSI is in asynchronous mode, $\overline{\text{HBRST}}$ and HCLKIN should either be disconnected or pulled up.
- When the DSI uses sliding window address mode (DCR[SLDWA] = 1), the external HA[11–13] signals must be connected (tied) to the correct voltage levels so that the host can perform the first access to the DCR. After reset, the DSI expects full address mode (DCR[SLDWA] = 0). The DCR address in the DSI memory map is 0x1BE000, which requires the following connections:
  - HA11 must be pulled high (1)
  - HA12 must be pulled high (1)
  - HA13 must be pulled low (0)
- The following signals must be pulled up: $\overline{\text{HRESET}}$, $\overline{\text{SRESET}}$, $\overline{\text{ARTRY}}$, $\overline{\text{TA}}$, $\overline{\text{TEA}}$, $\overline{\text{PSDVAL}}$, and $\overline{\text{AACK}}$.
- In single-master mode (BCR[EBM] = 0) with internal arbitration (PPC_ACR[EARB] = 0):
  - $\overline{\text{BG}}$, $\overline{\text{DBG}}$, and $\overline{\text{TS}}$ can be left unconnected.
  - $\overline{\text{EXT\_BG[2–3]}}$, $\overline{\text{EXT\_DBG[2–3]}}$, and $\overline{\text{GBL}}$ can be left unconnected if they are multiplexed to the system bus functionality. For any other functionality, connect the signal lines based on the multiplexed functionality.
  - $\overline{\text{BR}}$ must be pulled up.
  - $\overline{\text{EXT\_BR[2–3]}}$ must be pulled up if multiplexed to the system bus functionality.
- If there is an external bus master (BCR[EBM] = 1):
  - $\overline{\text{BR}}$, $\overline{\text{BG}}$, $\overline{\text{DBG}}$, and $\overline{\text{TS}}$ must be pulled up.
  - $\overline{\text{EXT\_BR[2–3]}}$, $\overline{\text{EXT\_BG[2–3]}}$, and $\overline{\text{EXT\_DBG[2–3]}}$ must be pulled up if multiplexed to the system bus functionality.
- In single-master mode, $\overline{\text{ABB}}$ and $\overline{\text{DBB}}$ can be selected as $\overline{\text{IRQ}}$ inputs and be connected to the non-active value. In other modes, they must be pulled up.

Note: The MSC8122 does not support DLL-enabled mode. For the following two clock schemes, ensure that the DLL is disabled (that is, the DLLDIS bit in the Hard Reset Configuration Word is set).

- If no system synchronization is required (for example, the design does not use SDRAM), you can use any of the available clock modes.
- In the CLKIN synchronization mode, use the following connections:
  - Connect the oscillator output through a buffer to CLKIN.
  - Connect the CLKIN buffer output to the slave device (for example, SDRAM) making sure that the delay path between the clock buffer to the MSC8122 and the SDRAM is equal (that is, has a skew less than 100 ps).
  - Valid clock modes in this scheme are: 0, 7, 15, 19, 21, 23, 28, 29, 30, and 31.
- In CLKOUT synchronization mode (for 1.2 V devices), CLKOUT is the main clock to SDRAM. Use the following connections:
  - Connect the oscillator output through a buffer to CLKIN.
  - Connect CLKOUT through a zero-delay buffer to the slave device (for example, SDRAM) using the following guidelines:
    - The maximum delay between the slave and CLKOUT must not exceed 0.7 ns.
    - The maximum load on CLKOUT must not exceed 10 pF.
    - Use a zero-delay buffer with a jitter less than 0.3 ns.
  - All clock modes are valid in this clock scheme.

Note: See the Clock chapter in the *MSC8122 Reference Manual* for details.

- If the 60x-compatible system bus is not used and SIUMCR[PBSE] is set, $\overline{\text{PPBS}}$ can be disconnected. Otherwise, it should be pulled up.
- The following signals: SWTE, DSISYNC, DSI64, MODCK[1–2], CNFGS, CHIPID[0–3], RSTCONF and BM[0–2] are used to configure the MSC8122 and are sampled on the deassertion of the $\overline{\text{PORESET}}$ signal. Therefore, they should be tied to GND or V_{DDH} or through a pull-down or a pull-up resistor until the deassertion of the $\overline{\text{PORESET}}$ signal.
- When they are used, $\overline{\text{INT\_OUT}}$ (if SIUMCR[INTODC] is cleared), $\overline{\text{NMI\_OUT}}$, and $\overline{\text{IRQxx}}$ (if not full drive) signals must be pulled up.
- When the Ethernet controller is enabled and the SMII mode is selected, GPIO10 and GPIO14 must not be connected externally to any signal line.

Note: For details on configuration, see the *MSC8122 User's Guide* and *MSC8122 Reference Manual*. For additional information, refer to the *MSC8122 Design Checklist* (AN2787).

3.4 External SDRAM Selection

The external bus speed implemented in a system determines the speed of the SDRAM used on that bus. However, because of differences in timing characteristics among various SDRAM manufacturers, you may have use a faster speed rated SDRAM to assure efficient data transfer across the bus. For example, for 166 MHz operation, you may have to use 183 or 200 MHz SDRAM. Always perform a detailed timing analysis using the MSC8122 bus timing values and the manufacturer specifications for the SDRAM to ensure correct operation within your system design. The output delay listed in SDRAM specifications is usually given for a load of 30 pF. Scale the number to your specific board load using the typical scaling number provided by the SDRAM manufacturer.

7 Revision History

Table 31 provides a revision history for this data sheet.

Table 31. Document Revision History

Revision	Date	Description
0	May 2004	• Initial release.
1	Jun. 2004	• Updated timing number 32b. • Updated DSI timing specifications.
2	Sep 2004	• New orderable parts added with other core voltage and temperature options. • Updated thermal characteristics. • In Table 2-14, removed references to 30 pF. • Design guidelines and layout recommendations updated.
3	Nov. 2004	• Added 500 MHz core and 166 MHz bus speed options. • Definitions of GPIO[27–28] updated. • Bus, TDM, and GPIO timing updated. I^2C timing changed to GPIO timing. • GPIO[27–28] connections updated. MWBEn replaced with correct name HWBEn. • Design guidelines update.
4	Jan. 2005	• Package type changed to FC-PBGA for all frequencies. • Low-voltage 300 MHz power changed to 1.1 V. • $\overline{\text{HRESET}}$ and $\overline{\text{SRESET}}$ definitions updated. • Undershoot and overshoot values added for V_{DDH}. • RMII timing updated. • Design guidelines updated and reorganized.
5	Apr. 2005	• Added 400 MHz, 1.1 V core part. • Temperature range descriptions changed to standard and extended. • CLKOUT timing specifications added. • Device start-up guidelines added to design considerations and updated power supply guidelines. • Ordering information updated.
6	May 2005	• Multiple AC timing specifications updated.
7	May 2005	• Multiple AC timing specifications updated.
8	Jul. 2005	• Multiple AC timing specifications updated.
9	Jul. 2005	• AC specification table layout modified.
10	Sep. 2005	• ETHTX_EN type and $\overline{\text{TRST}}$ description updated. • Package drawing updated. • Clock specifications updated. • Start-up sequence updated.
11	Oct 2005	• V_{DDH} + 10% changed to V_{DDH} + 8% in **Figure 2-1**. • V_{DDH} +20% changed to V_{DDH} + 17% in **Figure 2-1**.
12	Apr 2006	• Reset timing updated to reflect actual values in **Table 2-11**.
13	Oct. 2006	• Added new timings 17 and 18 for IRQ set time and pulse width in **Table 2-13**
14	Dec. 2007	• Converted to new data sheet format. • Added PLL supply current to **Table 5** in **Section 2.4**. • Modified **Figure 5** in **Section 2.4** to make it clear that the time limits for undershoot referred to values below –0.3 V and not GND. • Added cross-references between **Sections 2.5.2** and **Section 3.1** and **3.2**. • Added power-sequence guidelines to **Sections 2.5.2**. • Added CLKIN jitter characteristic specifications to Table 9. • Added additional guidelines to prevent reverse current to **Section 3.1**. • Added connectivity guidelines for DSI in sliding windows mode to **Section 3.3**.
15	May 2008	• Changed V_{IL} maximum and reference value to 0.8 V in **Table 5**.
16	Dec. 2008	• Clarified the wording of note 2 in **Table 15** on p. 24.

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