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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	ST7
Core Size	8-Bit
Speed	8MHz
Connectivity	I²C, SCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	32
Program Memory Size	16KB (16K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	512 x 8
Voltage - Supply (Vcc/Vdd)	3.8V ~ 5.5V
Data Converters	A/D 12x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	44-LQFP
Supplier Device Package	-
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/st72f325j4tc

Table of Contents

Address	Block	Register Label	Register Name	Reset Status	Remarks
002Ch 002Dh	MCC	MCCSR MCCBCR	Main Clock Control / Status Register Main Clock Controller: Beep Control Register	00h 00h	R/W R/W
002Eh to 0030h	Reserved Area (3 Bytes)				
0031h 0032h 0033h 0034h 0035h 0036h 0037h 0038h 0039h 003Ah 003Bh 003Ch 003Dh 003Eh 003Fh	TIMER A	TACR2 TACR1 TACSR TAIC1HR TAIC1LR TAOC1HR TAOC1LR TACHR TACLR TAACHR TAACLR TAIC2HR TAIC2LR TAOC2HR TAOC2LR	Timer A Control Register 2 Timer A Control Register 1 Timer A Control/Status Register Timer A Input Capture 1 High Register Timer A Input Capture 1 Low Register Timer A Output Compare 1 High Register Timer A Output Compare 1 Low Register Timer A Counter High Register Timer A Counter Low Register Timer A Alternate Counter High Register Timer A Alternate Counter Low Register Timer A Input Capture 2 High Register Timer A Input Capture 2 Low Register Timer A Output Compare 2 High Register Timer A Output Compare 2 Low Register	00h 00h xxxx x0xx b xxh xxh 80h 00h FFh FCh FFh FCh xxh xxh 80h 00h	R/W R/W R/W Read Only Read Only R/W R/W Read Only Read Only Read Only Read Only Read Only Read Only R/W R/W
0040h	Reserved Area (1 Byte)				
0041h 0042h 0043h 0044h 0045h 0046h 0047h 0048h 0049h 004Ah 004Bh 004Ch 004Dh 004Eh 004Fh	TIMER B	TBCR2 TBCR1 TBCSR TBIC1HR TBIC1LR TBOC1HR TBOC1LR TBCHR TBCLR TBACHR TBACLR TBIC2HR TBIC2LR TBOC2HR TBOC2LR	Timer B Control Register 2 Timer B Control Register 1 Timer B Control/Status Register Timer B Input Capture 1 High Register Timer B Input Capture 1 Low Register Timer B Output Compare 1 High Register Timer B Output Compare 1 Low Register Timer B Counter High Register Timer B Counter Low Register Timer B Alternate Counter High Register Timer B Alternate Counter Low Register Timer B Input Capture 2 High Register Timer B Input Capture 2 Low Register Timer B Output Compare 2 High Register Timer B Output Compare 2 Low Register	00h 00h xxxx x0xx b xxh xxh 80h 00h FFh FCh FFh FCh xxh xxh 80h 00h	R/W R/W R/W Read Only Read Only R/W R/W Read Only Read Only Read Only Read Only Read Only Read Only R/W R/W
0050h 0051h 0052h 0053h 0054h 0055h 0056h 0057h	SCI	SCISR SCIDR SCIBRR SCICR1 SCICR2 SCIERPR SCIETPR	SCI Status Register SCI Data Register SCI Baud Rate Register SCI Control Register 1 SCI Control Register 2 SCI Extended Receive Prescaler Register Reserved area SCI Extended Transmit Prescaler Register	C0h xxh 00h x000 0000b 00h 00h --- 00h	Read Only R/W R/W R/W R/W R/W R/W

4 FLASH PROGRAM MEMORY

4.1 Introduction

The ST7 dual voltage High Density Flash (HDFlash) is a non-volatile memory that can be electrically erased as a single block or by individual sectors and programmed on a Byte-by-Byte basis using an external V_{PP} supply.

The HDFlash devices can be programmed and erased off-board (plugged in a programming tool) or on-board using ICP (In-Circuit Programming) or IAP (In-Application Programming).

The array matrix organisation allows each sector to be erased and reprogrammed without affecting other sectors.

4.2 Main Features

- Three Flash programming modes:
  - Insertion in a programming tool. In this mode, all sectors including option bytes can be programmed or erased.
  - ICP (In-Circuit Programming). In this mode, all sectors including option bytes can be programmed or erased without removing the device from the application board.
  - IAP (In-Application Programming) In this mode, all sectors except Sector 0, can be programmed or erased without removing the device from the application board and while the application is running.
- ICT (In-Circuit Testing) for downloading and executing user application test patterns in RAM
- Read-out protection
- Register Access Security System (RASS) to prevent accidental programming or erasing

4.3 Structure

The Flash memory is organised in sectors and can be used for both code and data storage.

Depending on the overall Flash memory size in the microcontroller device, there are up to three user sectors (see Table 5). Each of these sectors can be erased independently to avoid unnecessary erasing of the whole Flash memory when only a partial erasing is required.

The first two sectors have a fixed size of 4 Kbytes (see Figure 7). They are mapped in the upper part of the ST7 addressing space so the reset and interrupt vectors are located in Sector 0 (F000h-FFFFh).

Table 5. Sectors available in Flash devices

Flash Size (bytes)	Available Sectors
4K	Sector 0
8K	Sectors 0,1
> 8K	Sectors 0,1, 2

4.3.1 Read-out Protection

Read-out protection, when selected, provides a protection against Program Memory content extraction and against write access to Flash memory. Even if no protection can be considered as totally unbreakable, the feature provides a very high level of protection for a general purpose microcontroller.

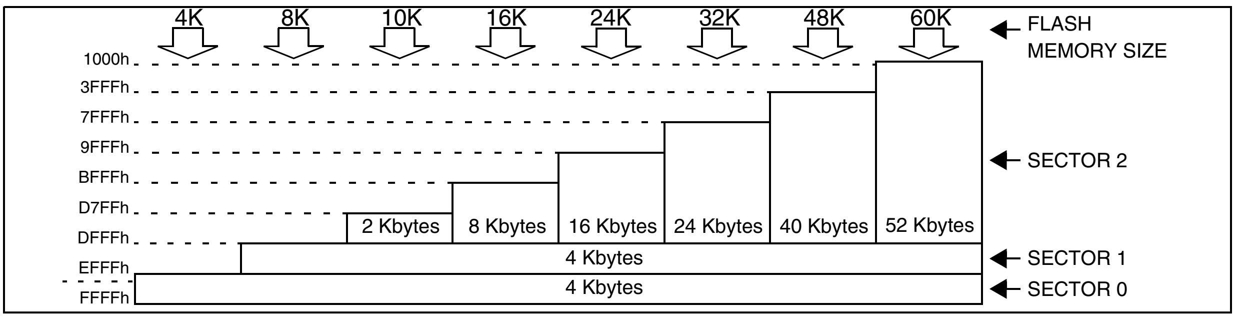
In flash devices, this protection is removed by reprogramming the option. In this case, the entire program memory is first automatically erased and the device can be reprogrammed.

Read-out protection selection depends on the device type:

– In Flash devices it is enabled and removed through the FMP_R bit in the option byte.

– In ROM devices it is enabled by mask option specified in the Option List.

Figure 7. Memory Map and Sector Address



SYSTEM INTEGRITY MANAGEMENT (Cont'd)

6.4.3 Clock Security System (CSS)

The Clock Security System (CSS) protects the ST7 against breakdowns, spikes and overfrequencies occurring on the main clock source (f_{OSC}). It is based on a clock filter and a clock detection control with an internal safe oscillator (f_{SFOSC}).

6.4.3.1 Clock Filter Control

The PLL has an integrated glitch filtering capability making it possible to protect the internal clock from overfrequencies created by individual spikes. This feature is available only when the PLL is enabled. If glitches occur on f_{OSC} (for example, due to loose connection or noise), the CSS filters these automatically, so the internal CPU frequency (f_{CPU}) continues deliver a glitch-free signal (see Figure 20).

6.4.3.2 Clock detection Control

If the clock signal disappears (due to a broken or disconnected resonator...), the safe oscillator delivers a low frequency clock signal (f_{SFOSC}) which allows the ST7 to perform some rescue operations.

Automatically, the ST7 clock source switches back from the safe oscillator (f_{SFOSC}) if the main clock source (f_{OSC}) recovers.

When the internal clock (f_{CPU}) is driven by the safe oscillator (f_{SFOSC}), the application software is notified by hardware setting the CSSD bit in the SICSR register. An interrupt can be generated if the CSSIE bit has been previously set.

These two bits are described in the SICSR register description.

6.4.4 Low Power Modes

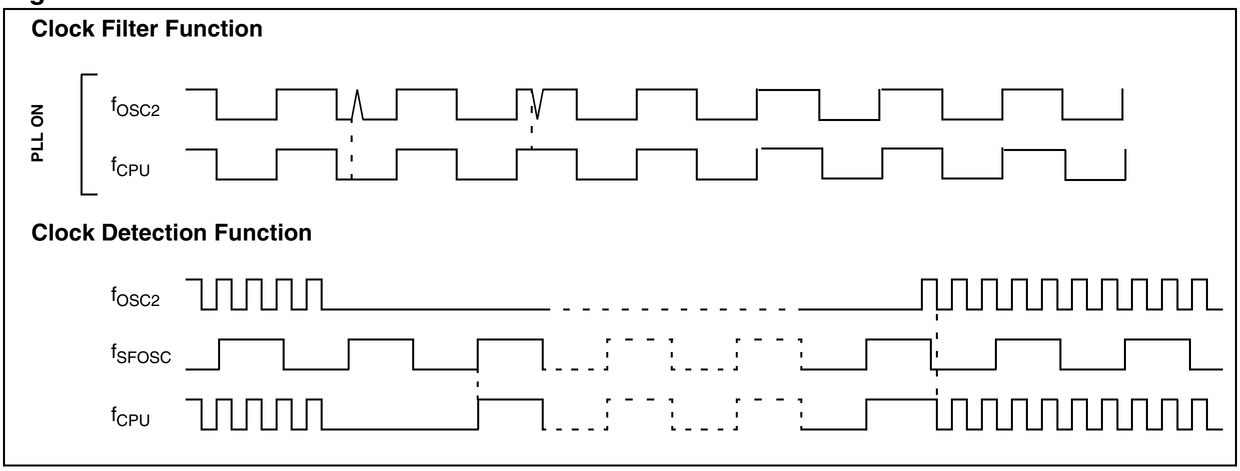
Mode	Description
WAIT	No effect on SI. CSS and AVD interrupts cause the device to exit from Wait mode.
HALT	The SICSR register is frozen.The CSS (including the safe oscillator) is disabled until HALT mode is exited. The previous CSS configuration resumes when the MCU is woken up by an interrupt with "exit from HALT mode" capability or from the counter reset value when the MCU is woken up by a RESET.

6.4.4.1 Interrupts

The CSS orAVD interrupt events generate an interrupt if the corresponding Enable Control Bit (CSSIE or AVDIE) is set and the interrupt mask in the CC register is reset (RIM instruction).

Interrupt Event	Event Flag	Enable Control Bit	Exit from Wait	Exit from Halt
CSS event detection (safe oscillator activated as main clock)	CSSD	CSSIE	Yes	No
AVD event	AVDF	AVDIE	Yes	No

Figure 20. Clock Filter Function



SYSTEM INTEGRITY MANAGEMENT (Cont'd)

6.4.5 Register Description

SYSTEM INTEGRITY (SI) CONTROL/STATUS REGISTER (SICSR)

Read/Write

Reset Value: 000x 000x (00h)

7							0
AVDS	AVDIE	AVDF	LVDRF	0	CSSIE	CSSD	WDGRF

Bit 7 = **AVDS** *Voltage Detection selection*
This bit is set and cleared by software. Voltage Detection is available only if the LVD is enabled by option byte.
0: Voltage detection on V_{DD} supply
1: Voltage detection on EVD pin

Bit 6 = **AVDIE** *Voltage Detector interrupt enable*
This bit is set and cleared by software. It enables an interrupt to be generated when the AVDF flag changes (toggles). The pending interrupt information is automatically cleared when software enters the AVD interrupt routine.
0: AVD interrupt disabled
1: AVD interrupt enabled

Bit 5 = **AVDF** *Voltage Detector flag*
This read-only bit is set and cleared by hardware. If the AVDIE bit is set, an interrupt request is generated when the AVDF bit changes value. Refer to Figure 18 and to Section 6.4.2.1 for additional details.
0: V_{DD} or V_{EVD} over $V_{IT+(AVD)}$ threshold
1: V_{DD} or V_{EVD} under $V_{IT-(AVD)}$ threshold

Bit 4 = **LVDRF** *LVD reset flag*
This bit indicates that the last Reset was generated by the LVD block. It is set by hardware (LVD reset) and cleared by software (writing zero). See WDGRF flag description for more details. When the LVD is disabled by OPTION BYTE, the LVDRF bit value is undefined.

Bit 3 = Reserved, must be kept cleared.

Bit 2 = **CSSIE** *Clock security syst interrupt enable*
This bit enables the interrupt when a disturbance is detected by the Clock Security System (CSSD bit set). It is set and cleared by software.
0: Clock security system interrupt disabled
1: Clock security system interrupt enabled
When the CSS is disabled by OPTION BYTE, the CSSIE bit has no effect.

Bit 1 = **CSSD** *Clock security system detection*
This bit indicates that the safe oscillator of the Clock Security System block has been selected by hardware due to a disturbance on the main clock signal (f_{OSC}). It is set by hardware and cleared by reading the SICSR register when the original oscillator recovers.
0: Safe oscillator is not active
1: Safe oscillator has been activated
When the CSS is disabled by OPTION BYTE, the CSSD bit value is forced to 0.

Bit 0 = **WDGRF** *Watchdog reset flag*
This bit indicates that the last Reset was generated by the Watchdog peripheral. It is set by hardware (watchdog reset) and cleared by software (writing zero) or an LVD Reset (to ensure a stable cleared state of the WDGRF flag when CPU starts).
Combined with the LVDRF flag information, the flag description is given by the following table.

RESET Sources	LVDRF	WDGRF
External $\overline{\text{RESET}}$ pin	0	0
Watchdog	0	1
LVD	1	X

Application notes

The LVDRF flag is not cleared when another RESET type occurs (external or watchdog), the LVDRF flag remains set to keep trace of the original failure.
In this case, a watchdog reset can be detected by software while an external reset can not.

CAUTION: When the LVD is not activated with the associated option byte, the WDGRF flag can not be used in the application.

INTERRUPTS (Cont'd)

Servicing Pending Interrupts

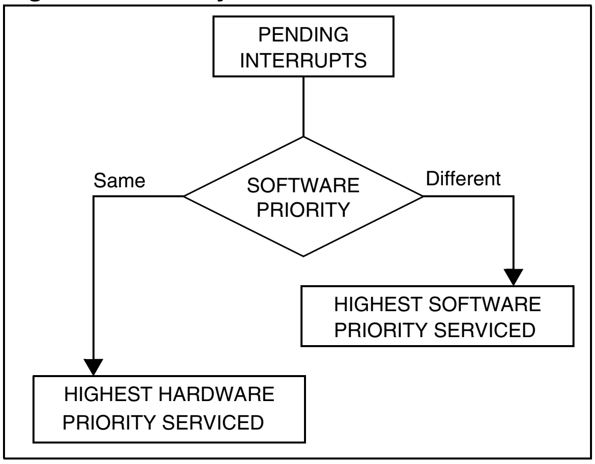
As several interrupts can be pending at the same time, the interrupt to be taken into account is determined by the following two-step process:

– the highest software priority interrupt is serviced,

– if several interrupts have the same software priority then the interrupt with the highest hardware priority is serviced first.

Figure 22 describes this decision process.

Figure 22. Priority Decision Process



When an interrupt request is not serviced immediately, it is latched and then processed when its software priority combined with the hardware priority becomes the highest one.

Note 1: The hardware priority is exclusive while the software one is not. This allows the previous process to succeed with only one interrupt.
Note 2: TLI,RESET and TRAP can be considered as having the highest software priority in the decision process.

Different Interrupt Vector Sources

Two interrupt source types are managed by the ST7 interrupt controller: the non-maskable type (RESET, TRAP) and the maskable type (external or from internal peripherals).

Non-Maskable Sources

These sources are processed regardless of the state of the I1 and I0 bits of the CC register (see Figure 21). After stacking the PC, X, A and CC registers (except for RESET), the corresponding vector is loaded in the PC register and the I1 and I0 bits of the CC are set to disable interrupts (level 3). These sources allow the processor to exit HALT mode.

■ TRAP (Non Maskable Software Interrupt)

This software interrupt is serviced when the TRAP instruction is executed. It will be serviced according to the flowchart in Figure 21.

Caution: TRAP can be interrupted by a TLI.

■ RESET

The RESET source has the highest priority in the ST7. This means that the first current routine has the highest software priority (level 3) and the highest hardware priority.
See the RESET chapter for more details.

Maskable Sources

Maskable interrupt vector sources can be serviced if the corresponding interrupt is enabled and if its own interrupt software priority (in ISPRx registers) is higher than the one currently being serviced (I1 and I0 in CC register). If any of these two conditions is false, the interrupt is latched and thus remains pending.

■ TLI (Top Level Hardware Interrupt)

This hardware interrupt occurs when a specific edge is detected on the dedicated TLI pin. It will be serviced according to the flowchart in Figure 21 as a trap.
Caution: A TRAP instruction must not be used in a TLI service routine.

■ External Interrupts

External interrupts allow the processor to exit from HALT low power mode. External interrupt sensitivity is software selectable through the External Interrupt Control register (EICR).
External interrupt triggered on edge will be latched and the interrupt request automatically cleared upon entering the interrupt service routine.
If several input pins of a group connected to the same interrupt line are selected simultaneously, these will be logically ORed.

■ Peripheral Interrupts

Usually the peripheral interrupts cause the MCU to exit from HALT mode except those mentioned in the “Interrupt Mapping” table. A peripheral interrupt occurs when a specific flag is set in the peripheral status registers and if the corresponding enable bit is set in the peripheral control register.
The general sequence for clearing an interrupt is based on an access to the status register followed by a read or write to an associated register.
Note: The clearing sequence resets the internal latch. A pending interrupt (i.e. waiting for being serviced) will therefore be lost if the clear sequence is executed.

POWER SAVING MODES (Cont'd)

8.4.2 HALT MODE

The HALT mode is the lowest power consumption mode of the MCU. It is entered by executing the 'HALT' instruction when the OIE bit of the Main Clock Controller Status register (MCCSR) is cleared (see section 10.2 on page 61 for more details on the MCCSR register).

The MCU can exit HALT mode on reception of either a specific interrupt (see Table 9, "Interrupt Mapping," on page 41) or a RESET. When exiting HALT mode by means of a RESET or an interrupt, the oscillator is immediately turned on and the 256 or 4096 CPU cycle delay is used to stabilize the oscillator. After the start up delay, the CPU resumes operation by servicing the interrupt or by fetching the reset vector which woke it up (see Figure 32).

When entering HALT mode, the I[1:0] bits in the CC register are forced to '10b'to enable interrupts. Therefore, if an interrupt is pending, the MCU wakes up immediately.

In HALT mode, the main oscillator is turned off causing all internal processing to be stopped, including the operation of the on-chip peripherals. All peripherals are not clocked except the ones which get their clock supply from another clock generator (such as an external or auxiliary oscillator).

The compatibility of Watchdog operation with HALT mode is configured by the "WDGHALT" option bit of the option byte. The HALT instruction when executed while the Watchdog system is enabled, can generate a Watchdog RESET (see section 14.1 on page 181 for more details).

Figure 31. HALT Timing Overview

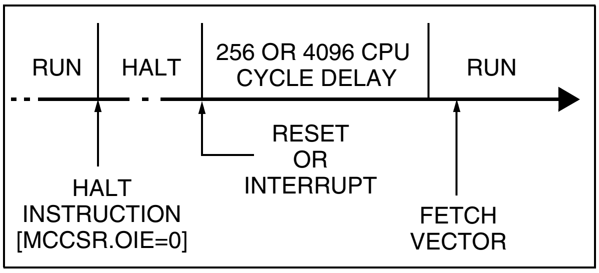
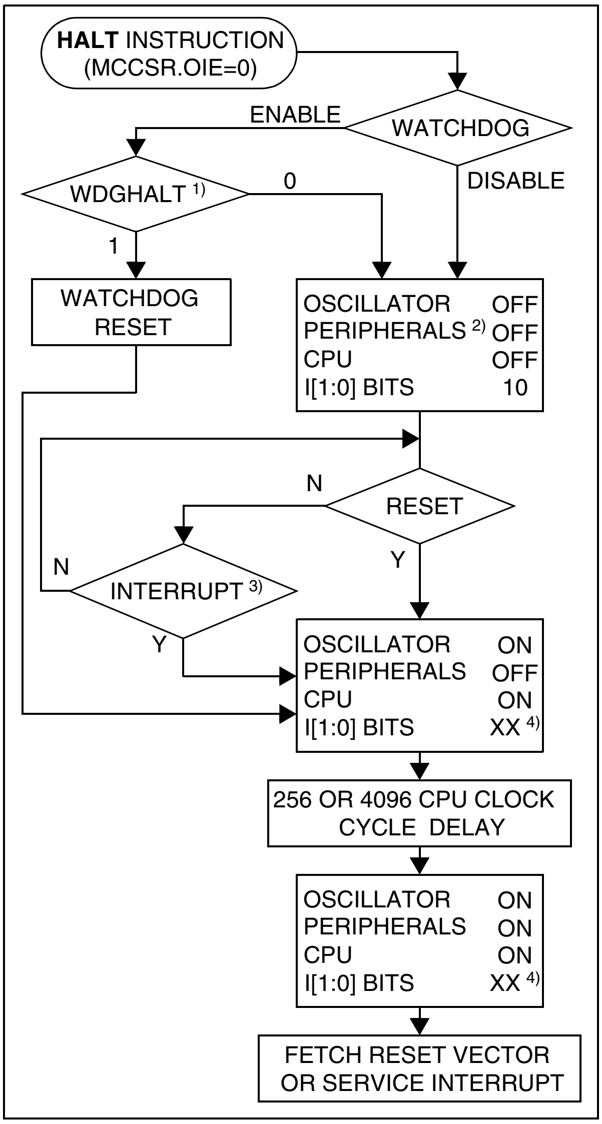


Figure 32. HALT Mode Flow-chart



Notes:

1. WDGHALT is an option bit. See option byte section for more details.
2. Peripheral clocked with an external clock source can still be active.
3. Only some specific interrupts can exit the MCU from HALT mode (such as external interrupt). Refer to Table 9, "Interrupt Mapping," on page 41 for more details.
4. Before servicing an interrupt, the CC register is pushed on the stack. The I[1:0] bits of the CC register are set to the current software priority level of the interrupt routine and recovered when the CC register is popped.

Table 15. Watchdog Timer Register Map and Reset Values

Address (Hex.)	Register Label	7	6	5	4	3	2	1	0
002Ah	**WDGCR** Reset Value	WDGA 0	T6 1	T5 1	T4 1	T3 1	T2 1	T1 1	T0 1

ON-CHIP PERIPHERALS (Cont'd)

10.3.2 Functional Description

Counter

The free running 8-bit counter is fed by the output of the prescaler, and is incremented on every rising edge of the clock signal.

It is possible to read or write the contents of the counter on the fly by reading or writing the Counter Access register (ARTCAR).

When a counter overflow occurs, the counter is automatically reloaded with the contents of the ARTARR register (the prescaler is not affected).

Counter clock and prescaler

The counter clock frequency is given by:

$$f_{COUNTER} = f_{INPUT} / 2^{CC[2:0]}$$

The timer counter's input clock (f_{INPUT}) feeds the 7-bit programmable prescaler, which selects one of the 8 available taps of the prescaler, as defined by CC[2:0] bits in the Control/Status Register (ARTCSR). Thus the division factor of the prescaler can be set to 2^n (where n = 0, 1,..7).

This f_{INPUT} frequency source is selected through the EXCL bit of the ARTCSR register and can be either the f_{CPU} or an external input frequency f_{EXT}.

The clock input to the counter is enabled by the TCE (Timer Counter Enable) bit in the ARTCSR register. When TCE is reset, the counter is stopped and the prescaler and counter contents are frozen. When TCE is set, the counter runs at the rate of the selected clock source.

Counter and Prescaler Initialization

After RESET, the counter and the prescaler are cleared and $f_{INPUT} = f_{CPU}$.

The counter can be initialized by:

– Writing to the ARTARR register and then setting the FCRL (Force Counter Re-Load) and the TCE (Timer Counter Enable) bits in the ARTCSR register.

– Writing to the ARTCAR counter access register,

In both cases the 7-bit prescaler is also cleared, whereupon counting will start from a known value.

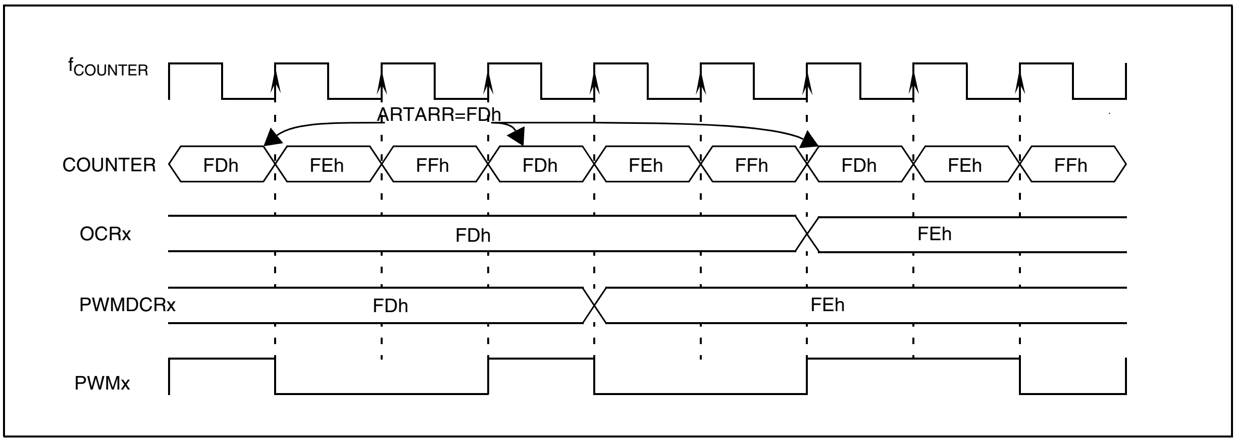
Direct access to the prescaler is not possible.

Output compare control

The timer compare function is based on four different comparisons with the counter (one for each PWMx output). Each comparison is made between the counter value and an output compare register (OCRx) value. This OCRx register can not be accessed directly, it is loaded from the duty cycle register (PWMDCRx) at each overflow of the counter.

This double buffering method avoids glitch generation when changing the duty cycle on the fly.

Figure 40. Output compare control



ON-CHIP PERIPHERALS (Cont'd)

Independent PWM signal generation

This mode allows up to four Pulse Width Modulated signals to be generated on the PWMx output pins with minimum core processing overhead. This function is stopped during HALT mode.

Each PWMx output signal can be selected independently using the corresponding OEx bit in the PWM Control register (PWMCR). When this bit is set, the corresponding I/O pin is configured as output push-pull alternate function.

The PWM signals all have the same frequency which is controlled by the counter period and the ARTARR register value.

$$f_{PWM} = f_{COUNTER} / (256 - ARTARR)$$

When a counter overflow occurs, the PWMx pin level is changed depending on the corresponding OPx (output polarity) bit in the PWMCR register.

When the counter reaches the value contained in one of the output compare register (OCRx) the corresponding PWMx pin level is restored.

It should be noted that the reload values will also affect the value and the resolution of the duty cycle of the PWM output signal. To obtain a signal on a PWMx pin, the contents of the OCRx register must be greater than the contents of the ARTARR register.

The maximum available resolution for the PWMx duty cycle is:

$$\text{Resolution} = 1 / (256 - ARTARR)$$

Note: To get the maximum resolution (1/256), the ARTARR register must be 0. With this maximum resolution, 0% and 100% can be obtained by changing the polarity.

Figure 41. PWM Auto-reload Timer Function

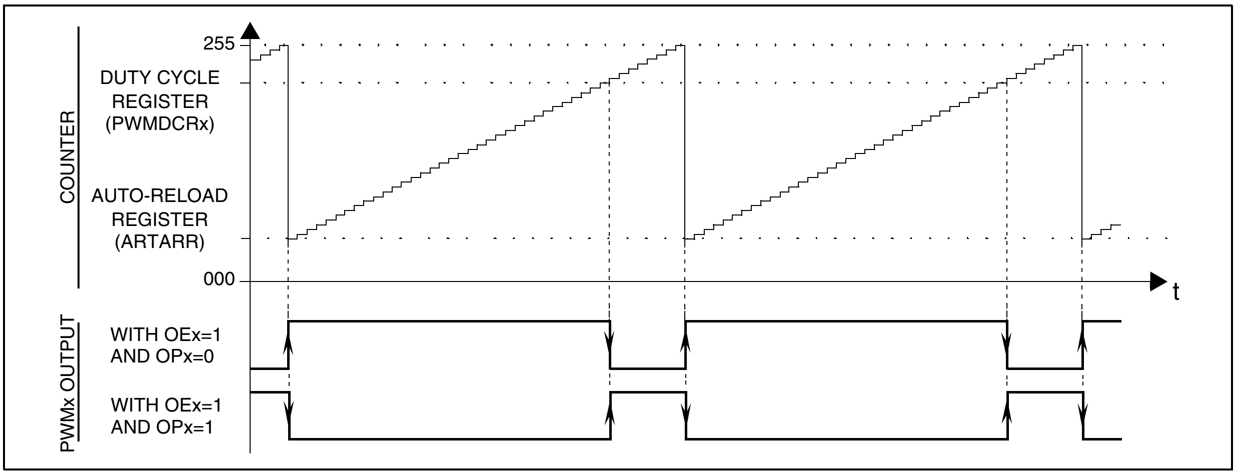
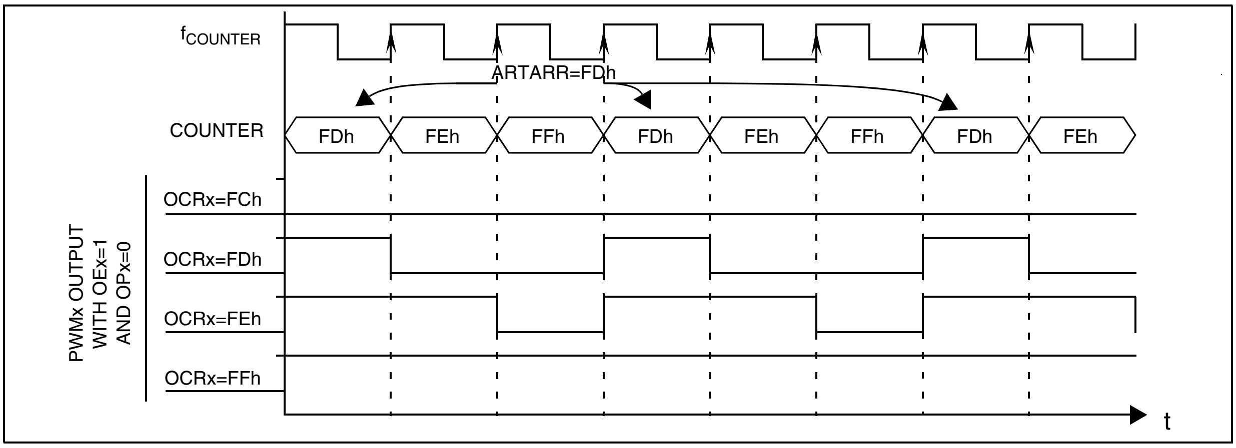


Figure 42. PWM Signal from 0% to 100% Duty Cycle



ON-CHIP PERIPHERALS (Cont'd)

10.3.3 Register Description

CONTROL / STATUS REGISTER (ARTCSR)

Read/Write

Reset Value: 0000 0000 (00h)

7							0
EXCL	CC2	CC1	CC0	TCE	FCRL	OIE	OVF

Bit 7 = **EXCL** *External Clock*
This bit is set and cleared by software. It selects the input clock for the 7-bit prescaler.
0: CPU clock.
1: External clock.

Bit 6:4 = **CC[2:0]** *Counter Clock Control*
These bits are set and cleared by software. They determine the prescaler division ratio from f_{INPUT}.

$f_{COUNTER}$	With f_{INPUT}=8 MHz	CC2	CC1	CC0
f_{INPUT}	8 MHz	0	0	0
f_{INPUT} / 2	4 MHz	0	0	1
f_{INPUT} / 4	2 MHz	0	1	0
f_{INPUT} / 8	1 MHz	0	1	1
f_{INPUT} / 16	500 kHz	1	0	0
f_{INPUT} / 32	250 kHz	1	0	1
f_{INPUT} / 64	125 kHz	1	1	0
f_{INPUT} / 128	62.5 kHz	1	1	1

Bit 3 = **TCE** *Timer Counter Enable*
This bit is set and cleared by software. It puts the timer in the lowest power consumption mode.
0: Counter stopped (prescaler and counter frozen).
1: Counter running.

Bit 2 = **FCRL** *Force Counter Re-Load*
This bit is write-only and any attempt to read it will yield a logical zero. When set, it causes the contents of ARTARR register to be loaded into the counter, and the content of the prescaler register to be cleared in order to initialize the timer before starting to count.

Bit 1 = **OIE** *Overflow Interrupt Enable*
This bit is set and cleared by software. It allows to enable/disable the interrupt which is generated when the OVF bit is set.
0: Overflow Interrupt disable.
1: Overflow Interrupt enable.

Bit 0 = **OVF** *Overflow Flag*
This bit is set by hardware and cleared by software reading the ARTCSR register. It indicates the transition of the counter from FFh to the ARTARR value.
0: New transition not yet reached
1: Transition reached

COUNTER ACCESS REGISTER (ARTCAR)

Read/Write

Reset Value: 0000 0000 (00h)

7							0
CA7	CA6	CA5	CA4	CA3	CA2	CA1	CA0

Bit 7:0 = **CA[7:0]** *Counter Access Data*

These bits can be set and cleared either by hardware or by software. The ARTCAR register is used to read or write the auto-reload counter "on the fly" (while it is counting).

AUTO-RELOAD REGISTER (ARTARR)

Read/Write

Reset Value: 0000 0000 (00h)

7							0
AR7	AR6	AR5	AR4	AR3	AR2	AR1	AR0

Bit 7:0 = **AR[7:0]** *Counter Auto-Reload Data*

These bits are set and cleared by software. They are used to hold the auto-reload value which is automatically loaded in the counter when an overflow occurs. At the same time, the PWM output levels are changed according to the corresponding OPx bit in the PWMCR register.

This register has two PWM management functions:

– Adjusting the PWM frequency
– Setting the PWM duty cycle resolution

PWM Frequency vs Resolution:

ARTARR value	Resolution	f_{PWM}	
		Min	Max
0	8-bit	~0.244 kHz	31.25 kHz
[0..127]	> 7-bit	~0.244 kHz	62.5 kHz
[128..191]	> 6-bit	~0.488 kHz	125 kHz
[192..223]	> 5-bit	~0.977 kHz	250 kHz
[224..239]	> 4-bit	~1.953 kHz	500 kHz

ON-CHIP PERIPHERALS (Cont'd)

PWM CONTROL REGISTER (PWMCR)

Read/Write

Reset Value: 0000 0000 (00h)

7							0
OE3	OE2	OE1	OE0	OP3	OP2	OP1	OP0

Bit 7:4 = **OE[3:0]** *PWM Output Enable*
These bits are set and cleared by software. They enable or disable the PWM output channels independently acting on the corresponding I/O pin.
0: PWM output disabled.
1: PWM output enabled.

Bit 3:0 = **OP[3:0]** *PWM Output Polarity*
These bits are set and cleared by software. They independently select the polarity of the four PWM output signals.

PWMx output level		OPx
Counter <= OCRx	**Counter > OCRx**	
1	0	0
0	1	1

Note: When an OPx bit is modified, the PWMx output signal polarity is immediately reversed.

DUTY CYCLE REGISTERS (PWMDCRx)

Read/Write

Reset Value: 0000 0000 (00h)

7							0
DC7	DC6	DC5	DC4	DC3	DC2	DC1	DC0

Bit 7:0 = **DC[7:0]** *Duty Cycle Data*

These bits are set and cleared by software.

A PWMDCRx register is associated with the OCRx register of each PWM channel to determine the second edge location of the PWM signal (the first edge location is common to all channels and given by the ARTARR register). These PWMDCR registers allow the duty cycle to be set independently for each PWM channel.

10.5 SERIAL PERIPHERAL INTERFACE (SPI)

10.5.1 Introduction

The Serial Peripheral Interface (SPI) allows full-duplex, synchronous, serial communication with external devices. An SPI system may consist of a master and one or more slaves however the SPI interface can not be a master in a multi-master system.

10.5.2 Main Features

- Full duplex synchronous transfers (on 3 lines)
- Simplex synchronous transfers (on 2 lines)
- Master or slave operation
- Six master mode frequencies ($f_{CPU}/4$ max.)
- $f_{CPU}/2$ max. slave mode frequency (see note)
- $\overline{SS}$ Management by software or hardware
- Programmable clock polarity and phase
- End of transfer interrupt flag
- Write collision, Master Mode Fault and Overrun flags

Note: In slave mode, continuous transmission is not possible at maximum frequency due to the software overhead for clearing status flags and to initiate the next transmission sequence.

10.5.3 General Description

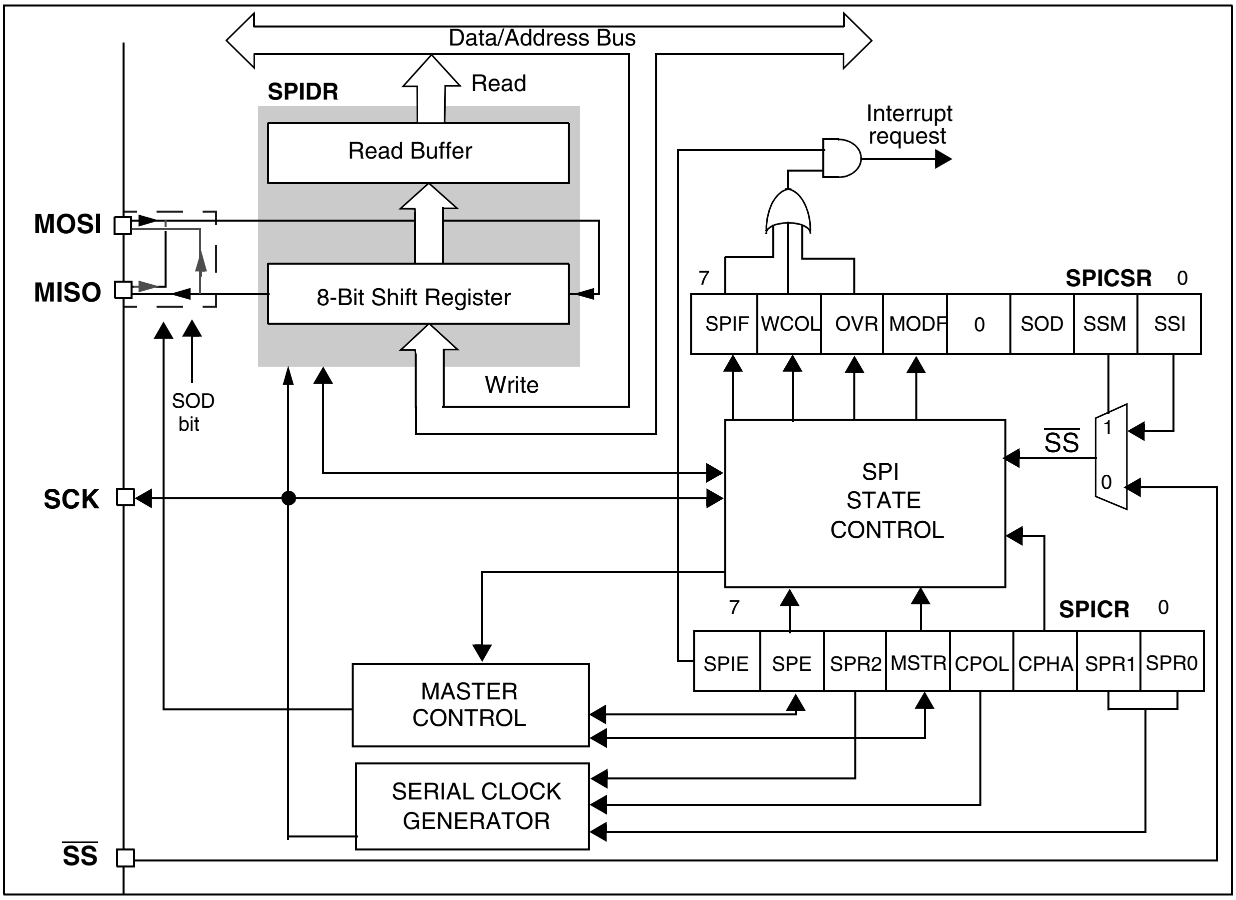
Figure 56 shows the serial peripheral interface (SPI) block diagram. There are 3 registers:

– SPI Control Register (SPICR)
– SPI Control/Status Register (SPICSR)
– SPI Data Register (SPIDR)

The SPI is connected to external devices through 4 pins:

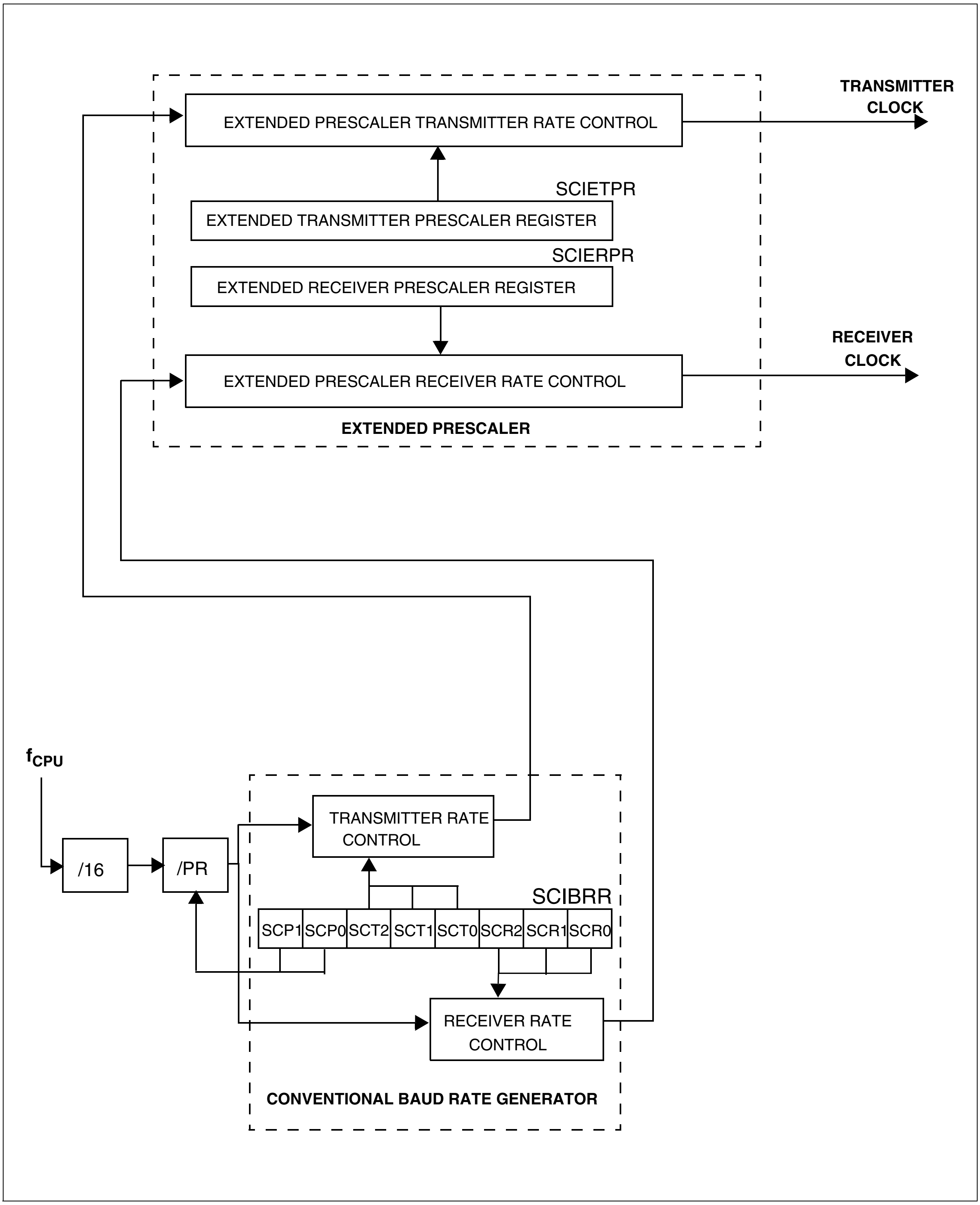
– MISO: Master In / Slave Out data
– MOSI: Master Out / Slave In data
– SCK: Serial Clock out by SPI masters and input by SPI slaves

Figure 56. Serial Peripheral Interface Block Diagram



SERIAL COMMUNICATIONS INTERFACE (Cont'd)

Figure 65. SCI Baud Rate and Extended Prescaler Block Diagram



I^2C BUS INTERFACE (Cont'd)

I^2C STATUS REGISTER 1 (SR1)

Read Only

Reset Value: 0000 0000 (00h)

7							0
EVF	ADD10	TRA	BUSY	BTF	ADSL	M/SL	SB

Bit 7 = **EVF** *Event flag.*
This bit is set by hardware as soon as an event occurs. It is cleared by software reading SR2 register in case of error event or as described in Figure 69. It is also cleared by hardware when the interface is disabled (PE=0).
0: No event
1: One of the following events has occurred:

– BTF=1 (Byte received or transmitted)
– ADSL=1 (Address matched in Slave mode while ACK=1)
– SB=1 (Start condition generated in Master mode)
– AF=1 (No acknowledge received after byte transmission)
– STOPF=1 (Stop condition detected in Slave mode)
– ARLO=1 (Arbitration lost in Master mode)
– BERR=1 (Bus error, misplaced Start or Stop condition detected)
– ADD10=1 (Master has sent header byte)
– Address byte successfully transmitted in Master mode.

Bit 6 = **ADD10** *10-bit addressing in Master mode.*
This bit is set by hardware when the master has sent the first byte in 10-bit address mode. It is cleared by software reading SR2 register followed by a write in the DR register of the second address byte. It is also cleared by hardware when the peripheral is disabled (PE=0).

0: No ADD10 event occurred.
1: Master has sent first address byte (header)

Bit 5 = **TRA** *Transmitter/Receiver.*
When BTF is set, TRA=1 if a data byte has been transmitted. It is cleared automatically when BTF is cleared. It is also cleared by hardware after detection of Stop condition (STOPF=1), loss of bus arbitration (ARLO=1) or when the interface is disabled (PE=0).
0: Data byte received (if BTF=1)
1: Data byte transmitted

Bit 4 = **BUSY** *Bus busy.*
This bit is set by hardware on detection of a Start condition and cleared by hardware on detection of a Stop condition. It indicates a communication in progress on the bus. The BUSY flag of the I2CSR1 register is cleared if a Bus Error occurs.
0: No communication on the bus
1: Communication ongoing on the bus
Note:
– The BUSY flag is NOT updated when the interface is disabled (PE=0). This can have consequences when operating in Multimaster mode; i.e. a second active I^2C master commencing a transfer with an unset BUSY bit can cause a conflict resulting in lost data. A software workaround consists of checking that the I^2C is not busy before enabling the I^2C Multimaster cell.

Bit 3 = **BTF** *Byte transfer finished.*
This bit is set by hardware as soon as a byte is correctly received or transmitted with interrupt generation if ITE=1. It is cleared by software reading SR1 register followed by a read or write of DR register. It is also cleared by hardware when the interface is disabled (PE=0).

– Following a byte transmission, this bit is set after reception of the acknowledge clock pulse. In case an address byte is sent, this bit is set only after the EV6 event (See Figure 69). BTF is cleared by reading SR1 register followed by writing the next byte in DR register.
– Following a byte reception, this bit is set after transmission of the acknowledge clock pulse if ACK=1. BTF is cleared by reading SR1 register followed by reading the byte from DR register.

The SCL line is held low while BTF=1.

0: Byte transfer not done
1: Byte transfer succeeded

Bit 2 = **ADSL** *Address matched (Slave mode).*
This bit is set by hardware as soon as the received slave address matched with the OAR register content or a general call is recognized. An interrupt is generated if ITE=1. It is cleared by software reading SR1 register or by hardware when the interface is disabled (PE=0).

The SCL line is held low while ADSL=1.

0: Address mismatched or not received
1: Received address matched

INSTRUCTION SET OVERVIEW (Cont'd)

11.1.1 Inherent

All Inherent instructions consist of a single byte. The opcode fully specifies all the required information for the CPU to process the operation.

Inherent Instruction	Function
NOP	No operation
TRAP	S/W Interrupt
WFI	Wait For Interrupt (Low Power Mode)
HALT	Halt Oscillator (Lowest Power Mode)
RET	Sub-routine Return
IRET	Interrupt Sub-routine Return
SIM	Set Interrupt Mask (level 3)
RIM	Reset Interrupt Mask (level 0)
SCF	Set Carry Flag
RCF	Reset Carry Flag
RSP	Reset Stack Pointer
LD	Load
CLR	Clear
PUSH/POP	Push/Pop to/from the stack
INC/DEC	Increment/Decrement
TNZ	Test Negative or Zero
CPL, NEG	1 or 2 Complement
MUL	Byte Multiplication
SLL, SRL, SRA, RLC, RRC	Shift and Rotate Operations
SWAP	Swap Nibbles

11.1.2 Immediate

Immediate instructions have 2 bytes, the first byte contains the opcode, the second byte contains the operand value.

Immediate Instruction	Function
LD	Load
CP	Compare
BCP	Bit Compare
AND, OR, XOR	Logical Operations
ADC, ADD, SUB, SBC	Arithmetic Operations

11.1.3 Direct

In Direct instructions, the operands are referenced by their memory address.

The direct addressing mode consists of two submodes:

Direct (short)

The address is a byte, thus requires only one byte after the opcode, but only allows 00 - FF addressing space.

Direct (long)

The address is a word, thus allowing 64 Kbyte addressing space, but requires 2 bytes after the opcode.

11.1.4 Indexed (No Offset, Short, Long)

In this mode, the operand is referenced by its memory address, which is defined by the unsigned addition of an index register (X or Y) with an offset.

The indirect addressing mode consists of three submodes:

Indexed (No Offset)

There is no offset, (no extra byte after the opcode), and allows 00 - FF addressing space.

Indexed (Short)

The offset is a byte, thus requires only one byte after the opcode and allows 00 - 1FE addressing space.

Indexed (long)

The offset is a word, thus allowing 64 Kbyte addressing space and requires 2 bytes after the opcode.

11.1.5 Indirect (Short, Long)

The required data byte to do the operation is found by its memory address, located in memory (pointer).

The pointer address follows the opcode. The indirect addressing mode consists of two submodes:

Indirect (short)

The pointer address is a byte, the pointer size is a byte, thus allowing 00 - FF addressing space, and requires 1 byte after the opcode.

Indirect (long)

The pointer address is a byte, the pointer size is a word, thus allowing 64 Kbyte addressing space, and requires 1 byte after the opcode.

I/O PORT PIN CHARACTERISTICS (Cont'd)

Figure 84. Typical V_{OL} vs. V_{DD} (standard)

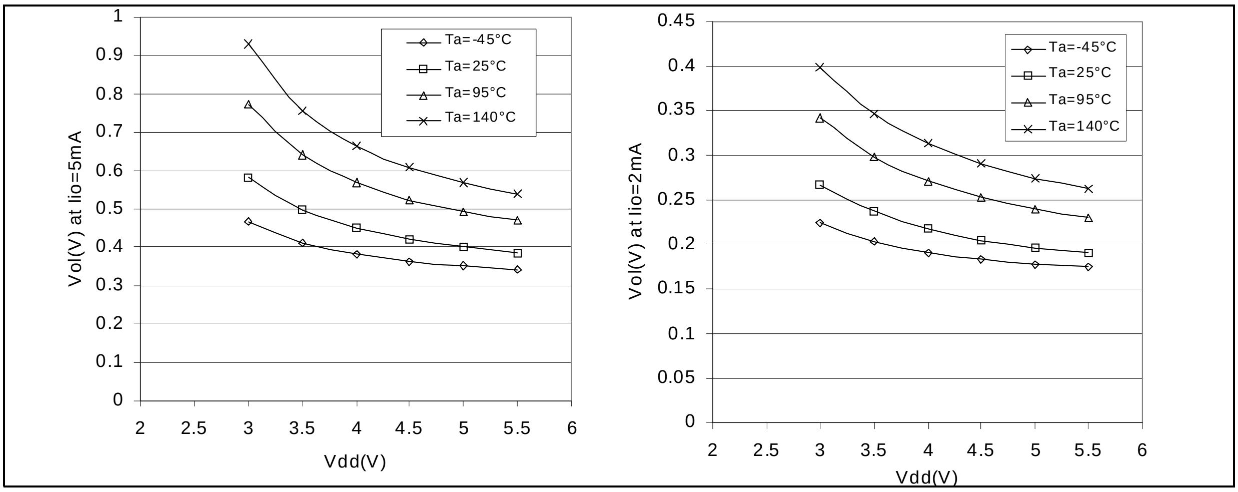


Figure 85. Typical V_{OL} vs. V_{DD} (high-sink)

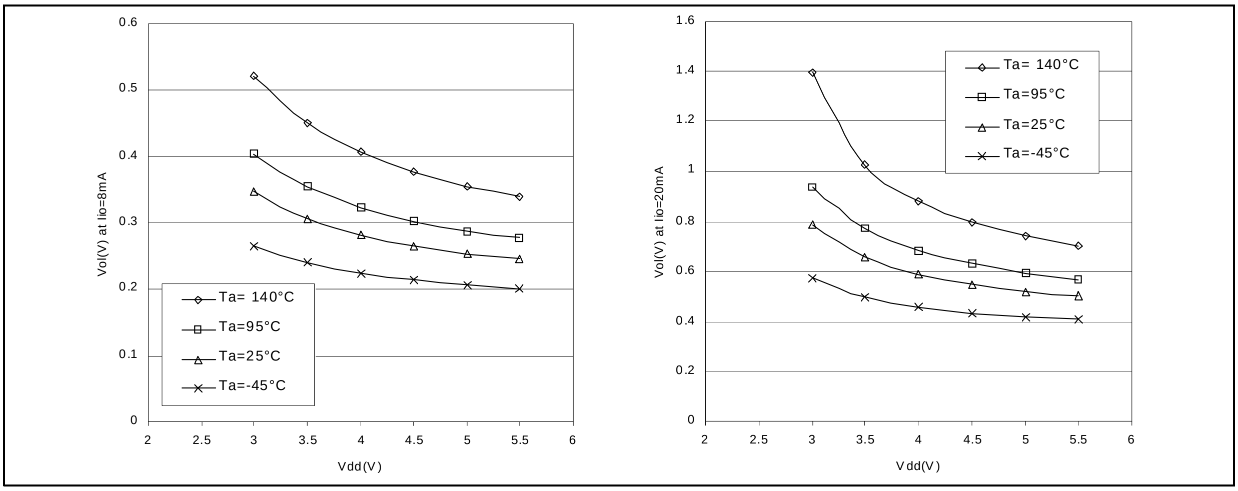
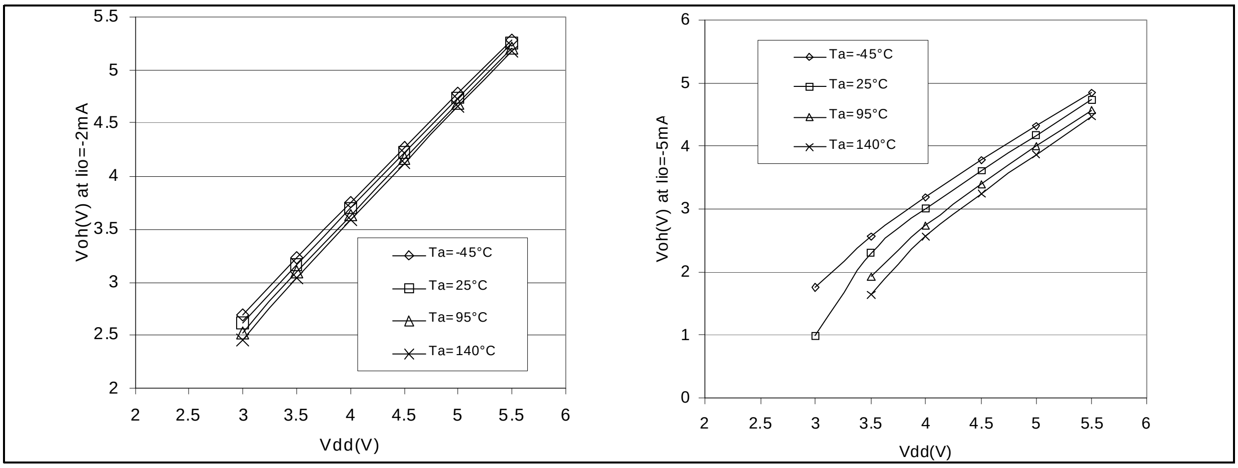


Figure 86. Typical V_{DD}-V_{OH} vs. V_{DD}



COMMUNICATION INTERFACE CHARACTERISTICS (Cont'd)

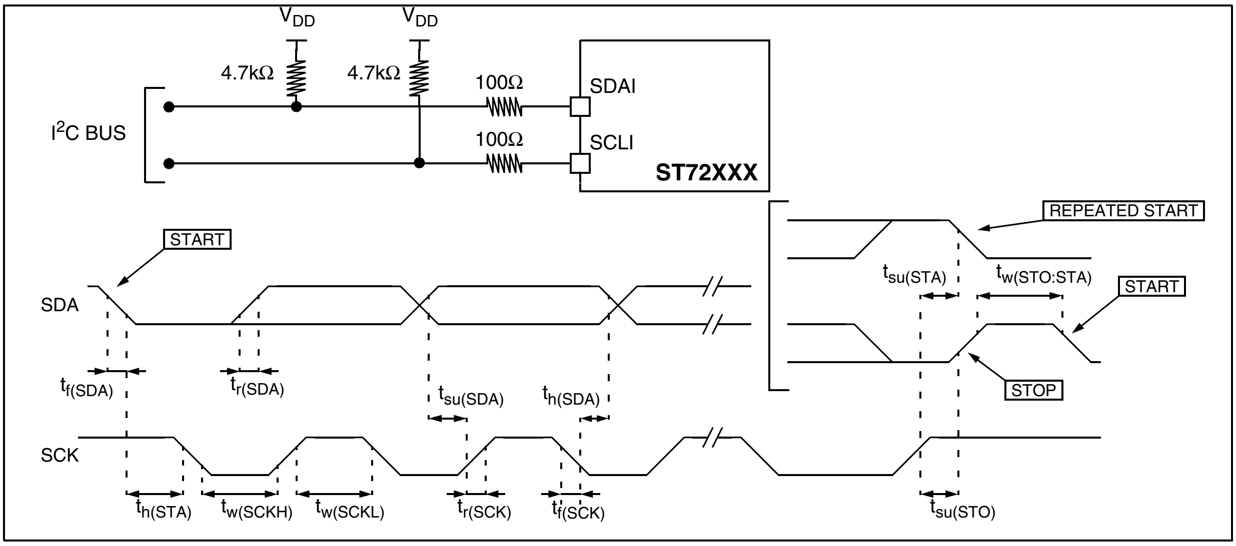
12.11.2 I²C - Inter IC Control Interface

Subject to general operating conditions for V_{DD}, f_{CPU}, and T_A unless otherwise specified.

Refer to I/O port characteristics for more details on the input/output alternate function characteristics (SDAI and SCLI). The ST7 I²C interface meets the requirements of the Standard I²C communication protocol described in the following table.

Symbol	Parameter	Standard mode I²C		Fast mode I²C[5]		Unit
		Min[1]	Max[1]	Min[1]	Max[1]	
$t_{w(SCLL)}$	SCL clock low time	4.7		1.3		µs
$t_{w(SCLH)}$	SCL clock high time	4.0		0.6		µs
$t_{su(SDA)}$	SDA setup time	250		100		ns
$t_{h(SDA)}$	SDA data hold time	0[3]		0[2]	900[3]	ns
$t_{r(SDA)}$ $t_{r(SCL)}$	SDA and SCL rise time		1000	$20+0.1C_b$	300	ns
$t_{f(SDA)}$ $t_{f(SCL)}$	SDA and SCL fall time		300	$20+0.1C_b$	300	ns
$t_{h(STA)}$	START condition hold time	4.0		0.6		µs
$t_{su(STA)}$	Repeated START condition setup time	4.7		0.6		µs
$t_{su(STO)}$	STOP condition setup time	4.0		0.6		µs
$t_{w(STO:STA)}$	STOP to START condition time (bus free)	4.7		1.3		µs
C_b	Capacitive load for each bus line		400		400	pF

Figure 93. Typical Application with I²C Bus and Timing Diagram [4]



Notes:

1. Data based on standard I²C protocol requirement, not tested in production.
2. The device must internally provide a hold time of at least 300ns for the SDA signal in order to bridge the undefined region of the falling edge of SCL.
3. The maximum hold time of the START condition has only to be met if the interface does not stretch the low period of SCL signal.
4. Measurement points are done at CMOS levels: $0.3xV_{DD}$ and $0.7xV_{DD}$.
5. At 4MHz f_{CPU}, max.I²C speed (400kHz) is not achievable. In this case, max. I²C speed will be approximately 260KHz.

ADC CHARACTERISTICS (Cont'd)

12.12.1 Analog Power Supply and Reference Pins

Depending on the MCU pin count, the package may feature separate V_{AREF} and V_{SSA} analog power supply pins. These pins supply power to the A/D converter cell and function as the high and low reference voltages for the conversion.

Separation of the digital and analog power pins allow board designers to improve A/D performance. Conversion accuracy can be impacted by voltage drops and noise in the event of heavily loaded or badly decoupled power supply lines (see Section 12.12.2 General PCB Design Guidelines).

12.12.2 General PCB Design Guidelines

To obtain best results, some general design and layout rules should be followed when designing the application PCB to shield the noise-sensitive, analog physical interface from noise-generating CMOS logic signals.

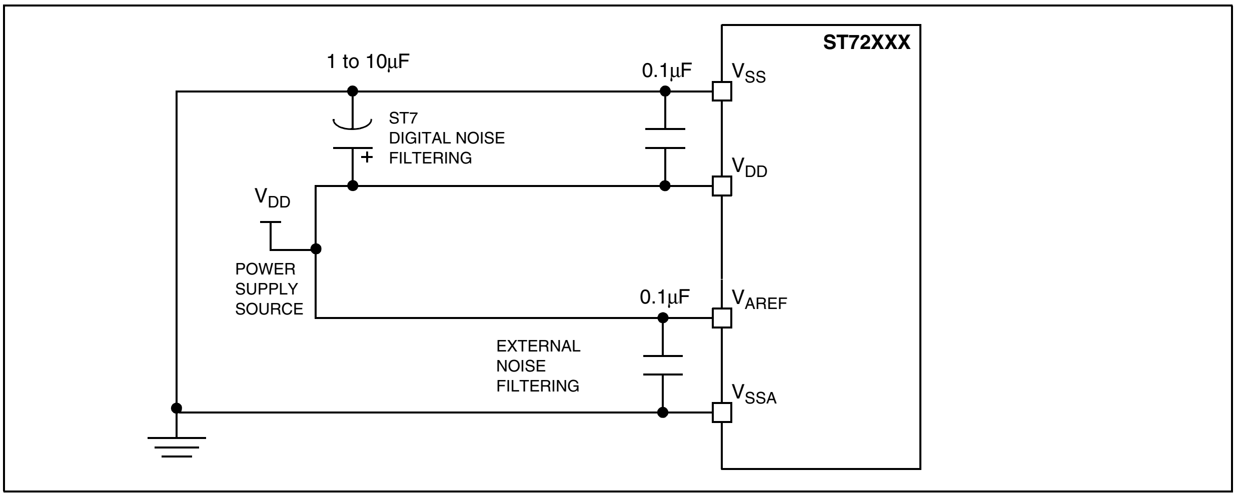
– Use separate digital and analog planes. The analog ground plane should be connected to the digital ground plane via a single point on the PCB.

– Filter power to the analog power planes. It is recommended to connect capacitors, with good high frequency characteristics, between the power and ground lines, placing 0.1µF and optionally, if needed 10pF capacitors as close as possible to the ST7 power supply pins and a 1 to 10µF capacitor close to the power source (see Figure 97).

– The analog and digital power supplies should be connected in a star network. Do not use a resistor, as V_{AREF} is used as a reference voltage by the A/D converter and any resistance would cause a voltage drop and a loss of accuracy.

– Properly place components and route the signal traces on the PCB to shield the analog inputs. Analog signals paths should run over the analog ground plane and be as short as possible. Isolate analog signals from digital signals that may switch while the analog inputs are being sampled by the A/D converter. Do not toggle digital outputs on the same I/O port as the A/D input being converted.

Figure 97. Power Supply Filtering



```
LD sema,A
IRET
```

Case 2: Writing to PxOR or PxDDR with Global Interrupts Disabled:

```
SIM          ; set the interrupt mask
LD A,PFDR
AND A,#$02
LD X,A       ; store the level before writing to PxOR/PxDDR
LD A,#$90
LD PFDDR,A; Write into PFDDR
LD A,#$ff
LD PFOR,A          ; Write to PFOR
LD A,PFDR
AND A,#$02
LD Y,A       ; store the level after writing to PxOR/PxDDR
LD A,X       ; check for falling edge
cp A,#$02
jrne OUT
TNZ Y
jrne OUT
LD A,#$01
LD sema,A    ; set the semaphore to '1' if edge is detected
RIM           ; reset the interrupt mask
LD A,sema    ; check the semaphore status
CP A,#$01
jrne OUT
call call_routine; call the interrupt routine
RIM
OUT:         RIM
JP while_loop
.call_routine ; entry to call_routine
PUSH A
PUSH X
PUSH CC
.ext1_rt     ; entry to interrupt routine
LD A,#$00
LD sema,A
IRET
```

15.1.3 Clearing active interrupts outside interrupt routine

When an active interrupt request occurs at the same time as the related flag is being cleared, an unwanted reset may occur.

Note: clearing the related interrupt mask will not generate an unwanted reset

Concurrent interrupt context

The symptom does not occur when the interrupts are handled normally, i.e.

when:

– The interrupt flag is cleared within its own interrupt routine

– The interrupt flag is cleared within any interrupt routine

– The interrupt flag is cleared in any part of the code while this interrupt is disabled

If these conditions are not met, the symptom can be avoided by implementing the following sequence:

Perform SIM and RIM operation before and after resetting an active interrupt request.

Example:

```
SIM
reset interrupt flag
RIM
```

Nested interrupt context:

The symptom does not occur when the interrupts are handled normally, i.e.

when:

– The interrupt flag is cleared within its own interrupt routine

– The interrupt flag is cleared within any interrupt routine with higher or identical priority level

– The interrupt flag is cleared in any part of the code while this interrupt is disabled

If these conditions are not met, the symptom can be avoided by implementing the following sequence:

```
PUSH CC
SIM
reset interrupt flag
POP CC
```