AMD Xilinx - XC3S100E-4TQ144I Datasheet

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Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	240
Number of Logic Elements/Cells	2160
Total RAM Bits	73728
Number of I/O	108
Number of Gates	100000
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xc3s100e-4tq144i

Storage Element Functions

There are three pairs of storage elements in each IOB, one pair for each of the three paths. It is possible to configure each of these storage elements as an edge-triggered D-type flip-flop (FD) or a level-sensitive latch (LD).

The storage-element pair on either the Output path or the Three-State path can be used together with a special multiplexer to produce Double-Data-Rate (DDR) transmission. This is accomplished by taking data synchronized to the clock signal's rising edge and converting it to bits synchronized on both the rising and the falling edge. The combination of two registers and a multiplexer is referred to as a Double-Data-Rate D-type flip-flop (ODDR2).

Table 4 describes the signal paths associated with the storage element.

Table 4: **Storage Element Signal Description**

Storage Element Signal	Description	Function
D	Data input	Data at this input is stored on the active edge of CK and enabled by CE. For latch operation when the input is enabled, data passes directly to the output Q.
Q	Data output	The data on this output reflects the state of the storage element. For operation as a latch in transparent mode, Q mirrors the data at D.
CK	Clock input	Data is loaded into the storage element on this input's active edge with CE asserted.
CE	Clock Enable input	When asserted, this input enables CK. If not connected, CE defaults to the asserted state.
SR	Set/Reset input	This input forces the storage element into the state specified by the SRHIGH/SRLOW attributes. The SYNC/ASYNC attribute setting determines if the SR input is synchronized to the clock or not. If both SR and REV are active at the same time, the storage element gets a value of 0.
REV	Reverse input	This input is used together with SR. It forces the storage element into the state opposite from what SR does. The SYNC/ASYNC attribute setting determines whether the REV input is synchronized to the clock or not. If both SR and REV are active at the same time, the storage element gets a value of 0.

As shown in Figure 5, the upper registers in both the output and three-state paths share a common clock. The OTCLK1 clock signal drives the CK clock inputs of the upper registers on the output and three-state paths. Similarly, OTCLK2 drives the CK inputs for the lower registers on the output and three-state paths. The upper and lower registers on the input path have independent clock lines: ICLK1 and ICLK2.

The OCE enable line controls the CE inputs of the upper and lower registers on the output path. Similarly, TCE controls the CE inputs for the register pair on the three-state path and ICE does the same for the register pair on the input path.

The Set/Reset (SR) line entering the IOB controls all six registers, as is the Reverse (REV) line.

In addition to the signal polarity controls described in IOB Overview, each storage element additionally supports the controls described in Table 5.

Table 5: **Storage Element Options**

Option Switch	Function	Specificity
FF/Latch	Chooses between an edge-triggered flip-flop or a level-sensitive latch	Independent for each storage element
SYNC/ASYNC	Determines whether the SR set/reset control is synchronous or asynchronous	Independent for each storage element
SRHIGH/SRLOW	Determines whether SR acts as a Set, which forces the storage element to a logic 1 (SRHIGH) or a Reset, which forces a logic 0 (SRLOW)	Independent for each storage element, except when using ODDR2. In the latter case, the selection for the upper element will apply to both elements.
INIT1/INIT0	When Global Set/Reset (GSR) is asserted or after configuration this option specifies the initial state of the storage element, either set (INIT1) or reset (INIT0). By default, choosing SRLOW also selects INIT0; choosing SRHIGH also selects INIT1.	Independent for each storage element, except when using ODDR2, which uses two IOBs. In the ODDR2 case, selecting INIT0 for one IOBs applies to both elements within the IOB, although INIT1 could be selected for the elements in the other IOB.

Spartan-3E FPGAs provide additional input flexibility by allowing I/O standards to be mixed in different banks. For a particular V_{CCO} voltage, Table 6 and Table 7 list all of the IOSTANDARDs that can be combined and if the IOSTANDARD is supported as an input only or can be used for both inputs and outputs.

Table 6: **Single-Ended IOSTANDARD Bank Compatibility**

Single-Ended IOSTANDARD	V_{CCO} Supply/Compatibility					Input Requirements	
	1.2V	1.5V	1.8V	2.5V	3.3V	V_{REF}	Board Termination Voltage (V_{TT})
LVTTL	-	-	-	-	Input/ Output	N/R(1)	N/R
LVCMOS33	-	-	-	-	Input/ Output	N/R	N/R
LVCMOS25	-	-	-	Input/ Output	Input	N/R	N/R
LVCMOS18	-	-	Input/ Output	Input	Input	N/R	N/R
LVCMOS15	-	Input/ Output	Input	Input	Input	N/R	N/R
LVCMOS12	Input/ Output	Input	Input	Input	Input	N/R	N/R
PCI33_3	-	-	-	-	Input/ Output	N/R	N/R
PCI66_3	-	-	-	-	Input/ Output	N/R	N/R
HSTL_I_18	-	-	Input/ Output	Input	Input	0.9	0.9
HSTL_III_18	-	-	Input/ Output	Input	Input	1.1	1.8
SSTL18_I	-	-	Input/ Output	Input	Input	0.9	0.9
SSTL2_I	-	-	-	Input/ Output	Input	1.25	1.25

Notes:

1. N/R - Not required for input operation.

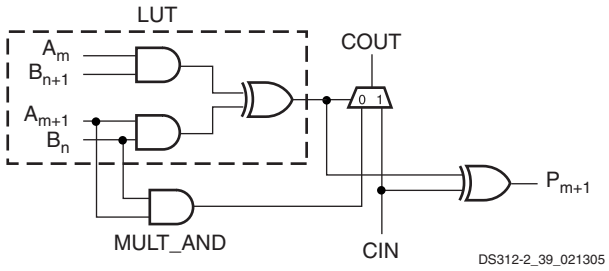


Figure 24: **Using the MULT_AND for Multiplication in Carry Logic**

The MULT_AND is useful for small multipliers. Larger multipliers can be built using the dedicated 18x18 multiplier blocks (see Dedicated Multipliers).

Storage Elements

The storage element, which is programmable as either a D-type flip-flop or a level-sensitive transparent latch, provides a means for synchronizing data to a clock signal, among other uses. The storage elements in the top and bottom portions of the slice are called FFY and FFX, respectively. FFY has a fixed multiplexer on the D input selecting either the combinatorial output Y or the bypass signal BY. FFX selects between the combinatorial output X or the bypass signal BX.

The functionality of a slice storage element is identical to that described earlier for the I/O storage elements. All signals have programmable polarity; the default active-High function is described.

Table 15: **Storage Element Signals**

Signal	Description
D	Input. For a flip-flop data on the D input is loaded when R and S (or CLR and PRE) are Low and CE is High during the Low-to-High clock transition. For a latch, Q reflects the D input while the gate (G) input and gate enable (GE) are High and R and S (or CLR and PRE) are Low. The data on the D input during the High-to-Low gate transition is stored in the latch. The data on the Q output of the latch remains unchanged as long as G or GE remains Low.
Q	Output. Toggles after the Low-to-High clock transition for a flip-flop and immediately for a latch.
C	Clock for edge-triggered flip-flops.
G	Gate for level-sensitive latches.
CE	Clock Enable for flip-flops.
GE	Gate Enable for latches.
S	Synchronous Set (Q = High). When the S input is High and R is Low, the flip-flop is set, output High, during the Low-to-High clock (C) transition. A latch output is immediately set, output High.
R	Synchronous Reset (Q = Low); has precedence over Set.
PRE	Asynchronous Preset (Q = High). When the PRE input is High and CLR is Low, the flip-flop is set, output High, during the Low-to-High clock (C) transition. A latch output is immediately set, output High.
CLR	Asynchronous Clear (Q = Low); has precedence over Preset to reset Q output Low
SR	CLB input for R, S, CLR, or PRE
REV	CLB input for opposite of SR. Must be asynchronous or synchronous to match SR.

The control inputs R, S, CE, and C are all shared between the two flip-flops in a slice.

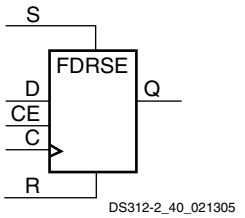


Figure 25: **FD Flip-Flop Component with Synchronous Reset, Set, and Clock Enable**

Table 16: **FD Flip-Flop Functionality with Synchronous Reset, Set, and Clock Enable**

Inputs					Outputs
R	S	CE	D	C	Q
1	X	X	X	↑	0
0	1	X	X	↑	1
0	0	0	X	X	No Change
0	0	1	1	↑	1
0	0	1	0	↑	0

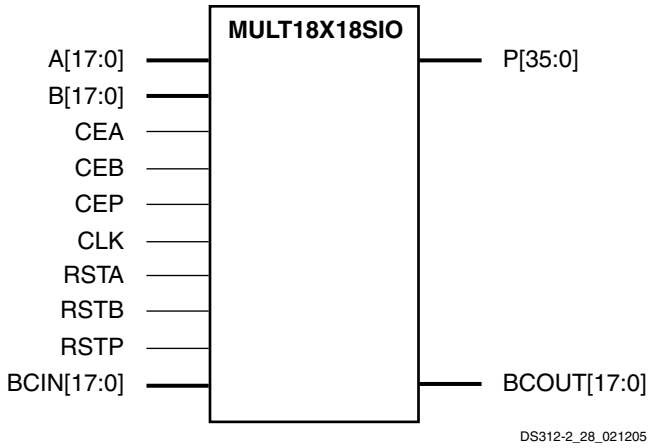


Figure 37: **MULT18X18SIO Primitive**

Cascading Multipliers

The MULT18X18SIO primitive has two additional ports called BCIN and BCOUT to cascade or share the multiplier's 'B' input among several multiplier bocks. The 18-bit BCIN "cascade" input port offers an alternate input source from the more typical 'B' input. The B_INPUT attribute specifies whether the specific implementation uses the BCIN or 'B' input path. Setting B_INPUT to DIRECT chooses the 'B' input. Setting B_INPUT to CASCADE selects the alternate BCIN input. The BREG register then optionally holds the selected input value, if required.

BCOUT is an 18-bit output port that always reflects the value that is applied to the multiplier's second input, which is either the 'B' input, the cascaded value from the BCIN input, or the output of the BREG if it is inserted.

Figure 38 illustrates the four possible configurations using different settings for the B_INPUT attribute and the BREG attribute.

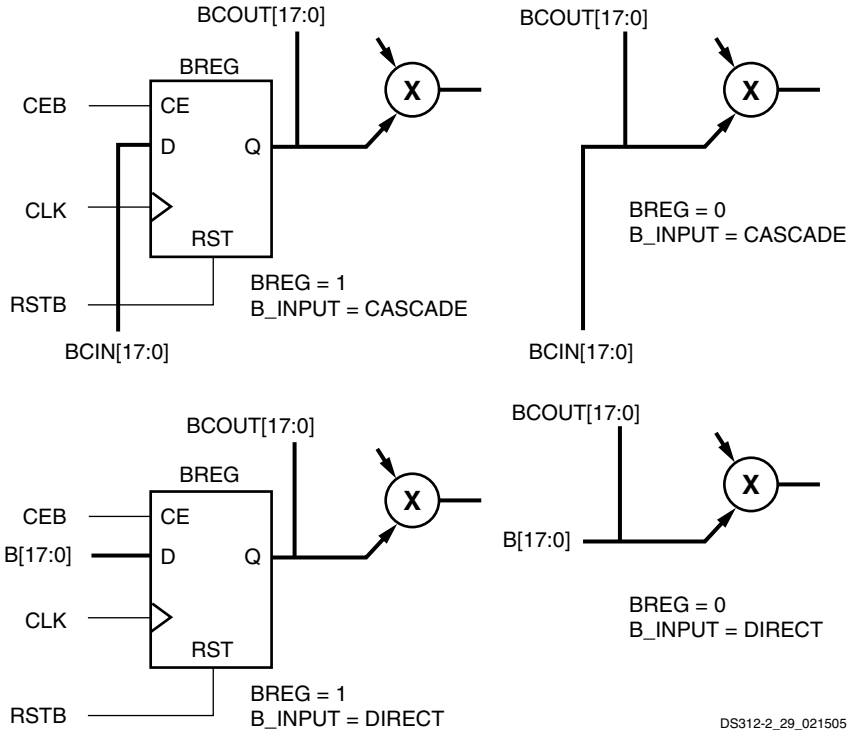


Figure 38: **Four Configurations of the B Input**

The BCIN and BCOUT ports have associated dedicated routing that connects adjacent multipliers within the same column. Via the cascade connection, the BCOUT port of one multiplier block drives the BCIN port of the multiplier block directly above it. There is no connection to the BCIN port of the bottom-most multiplier block in a column or a connection from the BCOUT port of the top-most block in a column. As an example, Figure 39 shows the multiplier cascade capability within the XC3S100E FPGA, which has a single column of multiplier, four blocks tall. For clarity, the figure omits the register control inputs.

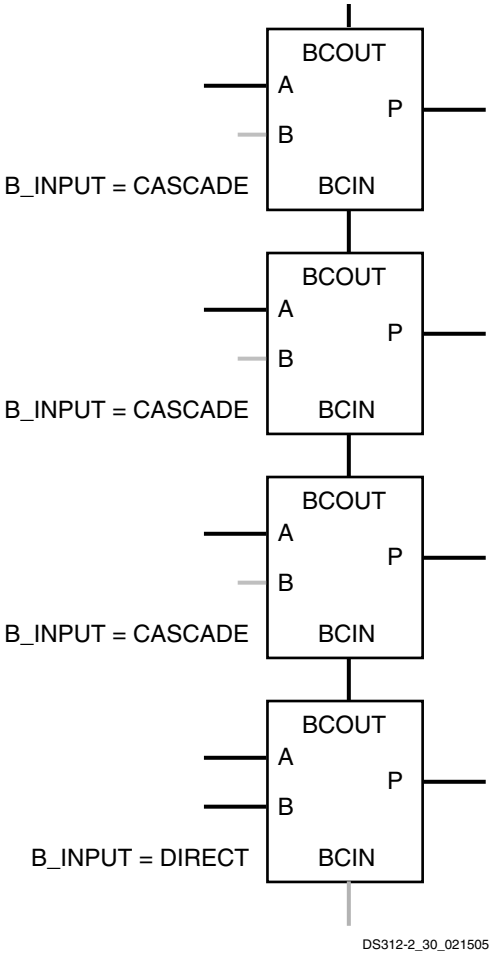


Figure 39: **Multiplier Cascade Connection**

When using the BREG register, the cascade connection forms a shift register structure typically used in DSP algorithms such as direct-form FIR filters. When the BREG register is omitted, the cascade structure essentially feeds the same input value to more than one multiplier. This parallel connection serves to create wide-input multipliers, implement transpose FIR filters, and is used in any application that requires that several multipliers have the same input value.

cancel out the clock skew. When the DLL phase-aligns the CLK0 signal with the CLKIN signal, it asserts the LOCKED output, indicating a lock on to the CLKIN signal.

DLL Attributes and Related Functions

The DLL unit has a variety of associated attributes as described in Table 29. Each attribute is described in detail in the sections that follow.

Table 29: **DLL Attributes**

Attribute	Description	Values
CLK_FEEDBACK	Chooses either the CLK0 or CLK2X output to drive the CLKFB input	NONE, ***1X***, 2X
CLKIN_DIVIDE_BY_2	Halves the frequency of the CLKIN signal just as it enters the DCM	***FALSE***, TRUE
CLKDV_DIVIDE	Selects the constant used to divide the CLKIN input frequency to generate the CLKDV output frequency	1.5, ***2***, 2.5, 3, 3.5, 4, 4.5, 5, 5.5, 6.0, 6.5, 7.0, 7.5, 8, 9, 10, 11, 12, 13, 14, 15, and 16
CLKIN_PERIOD	Additional information that allows the DLL to operate with the most efficient lock time and the best jitter tolerance	Floating-point value representing the CLKIN period in nanoseconds

DLL Clock Input Connections

For best results, an external clock source enters the FPGA via a Global Clock Input (GCLK). Each specific DCM has four possible direct, optimal GCLK inputs that feed the DCM's CLKIN input, as shown in Table 30. Table 30 also provides the specific pin numbers by package for each GCLK input. The two additional DCM's on the XC3S1200E and XC3S1600E have similar optimal connections from the left-edge LHCLK and the right-edge RHCLK inputs, as described in Table 31 and Table 32.

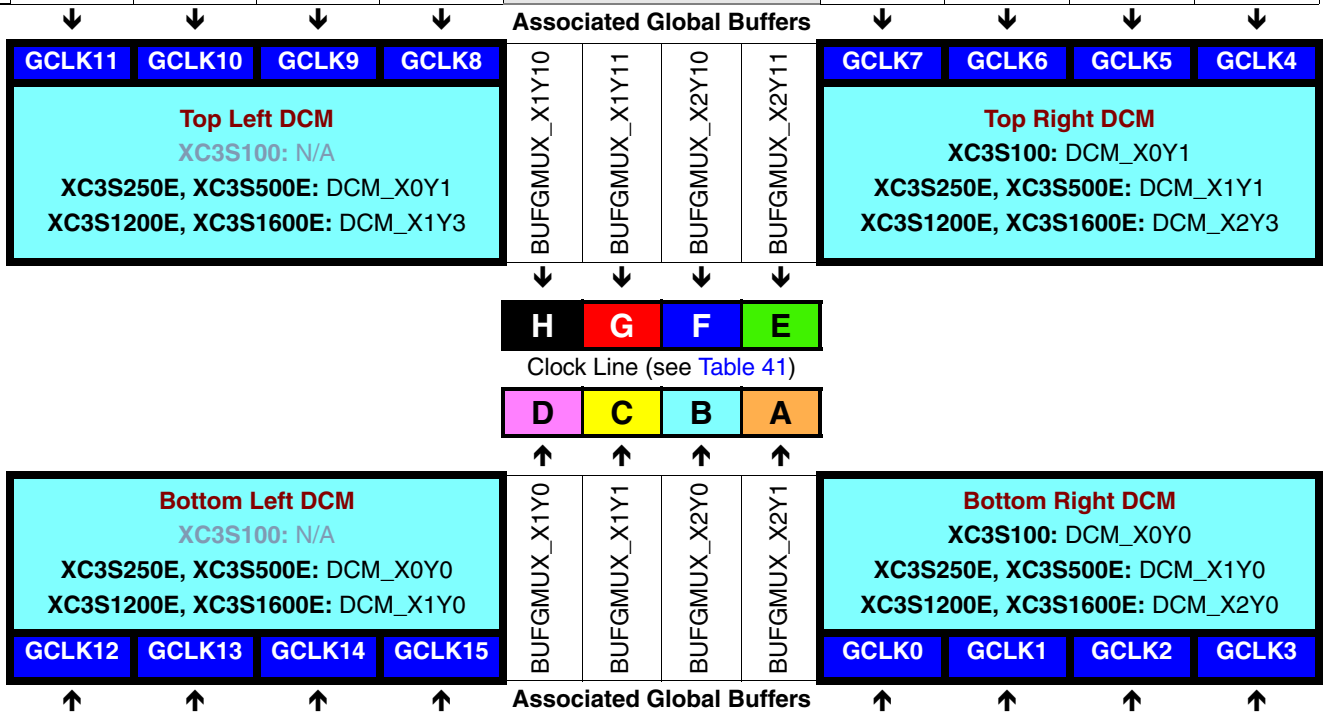
- The DCM supports differential clock inputs (for example, LVDS, LVPECL_25) via a pair of GCLK inputs that feed an internal single-ended signal to the DCM's CLKIN input.

Design Note

Avoid using global clock input GCLK1 as it is always shared with the M2 mode select pin. Global clock inputs GCLK0, GCLK2, GCLK3, GCLK12, GCLK13, GCLK14, and GCLK15 have shared functionality in some configuration modes.

Table 30: **Direct Clock Input Connections and Optional External Feedback to Associated DCMs**

Package	Differential Pair		Differential Pair			Differential Pair		Differential Pair	
	N	P	N	P		N	P	N	P
	Pin Number for Single-Ended Input					Pin Number for Single-Ended Input			
VQ100	P91	P90	P89	P88		P86	P85	P84	P83
CP132	B7	A7	C8	B8		A9	B9	C9	A10
TQ144	P131	P130	P129	P128		P126	P125	P123	P122
PQ208	P186	P185	P184	P183		P181	P180	P178	P177
FT256	D8	C8	B8	A8		A9	A10	F9	E9
FG320	D9	C9	B9	B8		A10	B10	E10	D10
FG400	A9	A10	G10	H10		E10	E11	G11	F11
FG484	B11	C11	H11	H12		C12	B12	E12	F12



Package	Differential Pair		Differential Pair			Differential Pair		Differential Pair	
	P	N	P	N		P	N	P	N
	Pin Number for Single-Ended Input					Pin Number for Single-Ended Input			
VQ100	P32	P33	P35	P36		P38	P39	P40	P41
CP132	M4	N4	M5	N5		M6	N6	P6	P7
TQ144	P50	P51	P53	P54		P56	P57	P58	P59
PQ208	P74	P75	P77	P78		P80	P81	P82	P83
FT256	M8	L8	N8	P8		T9	R9	P9	N9
FG320	N9	M9	U9	V9		U10	T10	R10	P10
FG400	W9	W10	R10	P10		P11	P12	V10	V11
FG484	V11	U11	R11	T11		R12	P12	Y12	W12

The CLKFX_DIVIDE is an integer ranging from 1 to 32, inclusive and forms the denominator in Equation 1. For example, if CLKFX_MULTIPLY = 5 and CLKFX_DIVIDE = 3, the frequency of the output clock signal is 5/3 that of the input clock signal. These attributes and their acceptable ranges are described in Table 34.

Table 34: **DFS Attributes**

Attribute	Description	Values
CLKFX_MULTIPLY	Frequency multiplier constant	Integer from 2 to 32, inclusive
CLKFX_DIVIDE	Frequency divisor constant	Integer from 1 to 32, inclusive

Any combination of integer values can be assigned to the CLKFX_MULTIPLY and CLKFX_DIVIDE attributes, provided that two conditions are met:

1. The two values fall within their corresponding ranges, as specified in Table 34.
2. The f_{CLKFX} output frequency calculated in Equation 1 falls within the DCM's operating frequency specifications (see Table 107 in Module 3).

DFS With or Without the DLL

Although the CLKIN input is shared with both units, the DFS unit functions with or separately from the DLL unit. Separate from the DLL, the DFS generates an output frequency from the CLKIN frequency according to the respective CLKFX_MULTIPLY and CLKFX_DIVIDE values. Frequency synthesis does not require a feedback loop. Furthermore, without the DLL, the DFS unit supports a broader operating frequency range.

With the DLL, the DFS unit operates as described above, only with the additional benefit of eliminating the clock distribution delay. In this case, a feedback loop from the CLK0 or CLK2X output to the CLKFB input must be present.

When operating with the DLL unit, the DFS's CLKFX and CLKFX180 outputs are phase-aligned with the CLKIN input every CLKFX_DIVIDE cycles of CLKIN and every CLKFX_MULTIPLY cycles of CLKFX. For example, when CLKFX_MULTIPLY = 5 and CLKFX_DIVIDE = 3, the input and output clock edges coincide every three CLKIN input periods, which is equivalent in time to five CLKFX output periods.

Smaller CLKFX_MULTIPLY and CLKFX_DIVIDE values result in faster lock times. Therefore, CLKFX_MULTIPLY and CLKFX_DIVIDE must be factored to reduce their values wherever possible. For example, given CLKFX_MULTIPLY = 9 and CLKFX_DIVIDE = 6, removing a factor of three yields CLKFX_MULTIPLY = 3 and CLKFX_DIVIDE = 2. While both value-pairs result in the multiplication of clock frequency by 3/2, the latter value-pair enables the DLL to lock more quickly.

Phase Shifter (PS)

The DCM provides two approaches to controlling the phase of a DCM clock output signal relative to the CLKIN signal: First, eight of the nine DCM clock outputs — CLK0, CLK90, CLK180, CLK270, CLK2X, CLK2X180, CLKFX, and CLKFX180 — provide either quadrant or half-period phase shifting of the input clock.

Second, the PS unit provides additional fine phase shift control of all nine DCM outputs. The PS unit accomplishes this by introducing a "fine phase shift" delay (T_{PS}) between the CLKFB and CLKIN signals inside the DLL unit. In FIXED phase shift mode, the fine phase shift is specified at design time with a resolution down to $^{1}/_{256}^{th}$ of a CLKIN cycle or one delay step (DCM_DELAY_STEP), whichever is greater. This fine phase shift value is relative to the coarser quadrant or half-period phase shift of the DCM clock output. When used, the PS unit shifts the phase of all nine DCM clock output signals.

Enabling Phase Shifting and Selecting an Operating Mode

The CLKOUT_PHASE_SHIFT attribute controls the PS unit for the specific DCM instantiation. As described in Table 35, this attribute has three possible values: NONE, FIXED, and VARIABLE. When CLKOUT_PHASE_SHIFT = NONE, the PS unit is disabled and the DCM output clocks are phase-aligned to the CLKIN input via the CLKFB feedback path. Figure 44a shows this case.

The PS unit is enabled when the CLKOUT_PHASE_SHIFT attribute is set to FIXED or VARIABLE modes. These two modes are described in the sections that follow.

Table 35: **PS Attributes**

Attribute	Description	Values
CLKOUT_PHASE_SHIFT	Disables the PS component or chooses between Fixed Phase and Variable Phase modes.	***NONE***, FIXED, VARIABLE
PHASE_SHIFT	Determines size and direction of initial fine phase shift.	Integers from –255 to +255

FIXED Phase Shift Mode

The FIXED phase shift mode shifts the DCM outputs by a fixed amount (T_{PS}), controlled by the user-specified PHASE_SHIFT attribute. The PHASE_SHIFT value (shown as P in Figure 44) must be an integer ranging from –255 to +255. PHASE_SHIFT specifies a phase shift delay as a fraction of the T_{CLKIN}. The phase shift behavior is different between ISE 8.1, Service Pack 3 and prior software versions, as described below.

Design Note

Prior to ISE 8.1i, Service Pack 3, the FIXED phase shift feature operated differently than the Spartan-3 DCM design primitive and simulation model. Designs using software prior to ISE 8.1i, Service Pack 3 require recompilation using the latest ISE software release. The following Answer Record contains additional information:

http://www.xilinx.com/support/answers/23153.htm.

FIXED Phase Shift using ISE 8.1i, Service Pack 3 and later: See Equation 2. The value corresponds to a phase shift range of –360° to +360°, which matches behavior of the Spartan-3 DCM design primitive and simulation model.

$$t_{PS} = \left(\frac{PHASESHIFT}{256}\right) \bullet T_{CLKIN} \qquad Eq\ 2$$

FIXED Phase Shift prior to ISE 8.1i, Service Pack 3: See Equation 3. The value corresponds to a phase shift range of –180° to +180° degrees, which is different from the Spartan-3 DCM design primitive and simulation model. Designs created prior to ISE 8.1i, Service Pack 3 must be recompiled using the most recent ISE development software.

$$t_{PS} = \left(\frac{PHASESHIFT}{512}\right) \bullet T_{CLKIN} \qquad Eq\ 3$$

When the PHASE_SHIFT value is zero, CLKFB and CLKIN are in phase, the same as when the PS unit is disabled. When the PHASE_SHIFT value is positive, the DCM outputs are shifted later in time with respect to CLKIN input. When the attribute value is negative, the DCM outputs are shifted earlier in time with respect to CLKIN.

Figure 44b illustrates the relationship between CLKFB and CLKIN in the Fixed Phase mode. In the Fixed Phase mode, the PSEN, PSCLK, and PSINCDEC inputs are not used and must be tied to GND.

Equation 2 or Equation 3 applies only to FIXED phase shift mode. The VARIABLE phase shift mode operates differently.

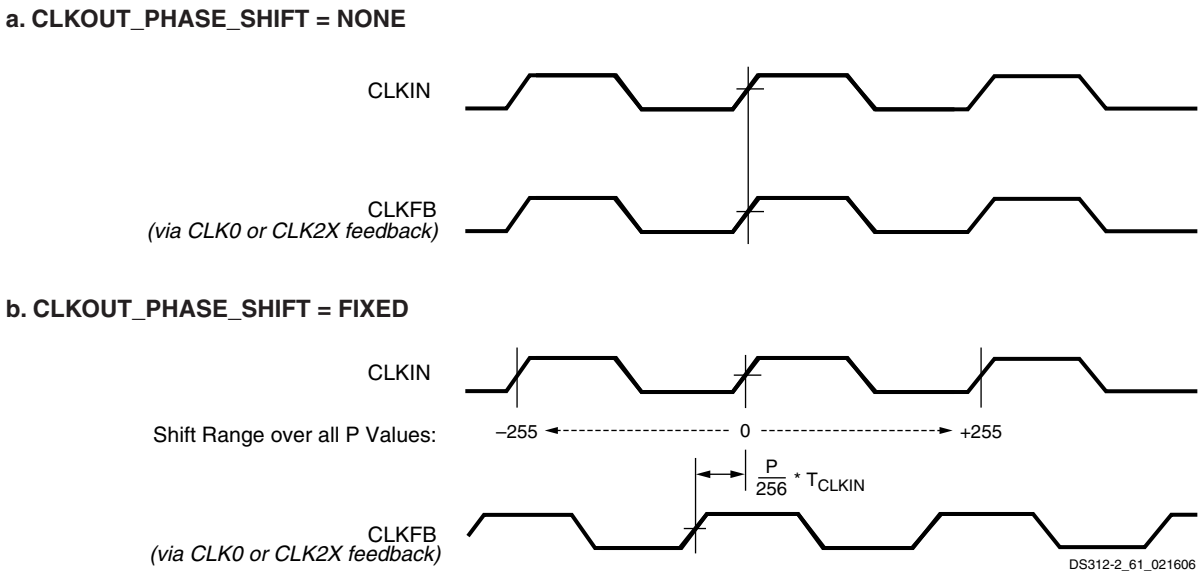
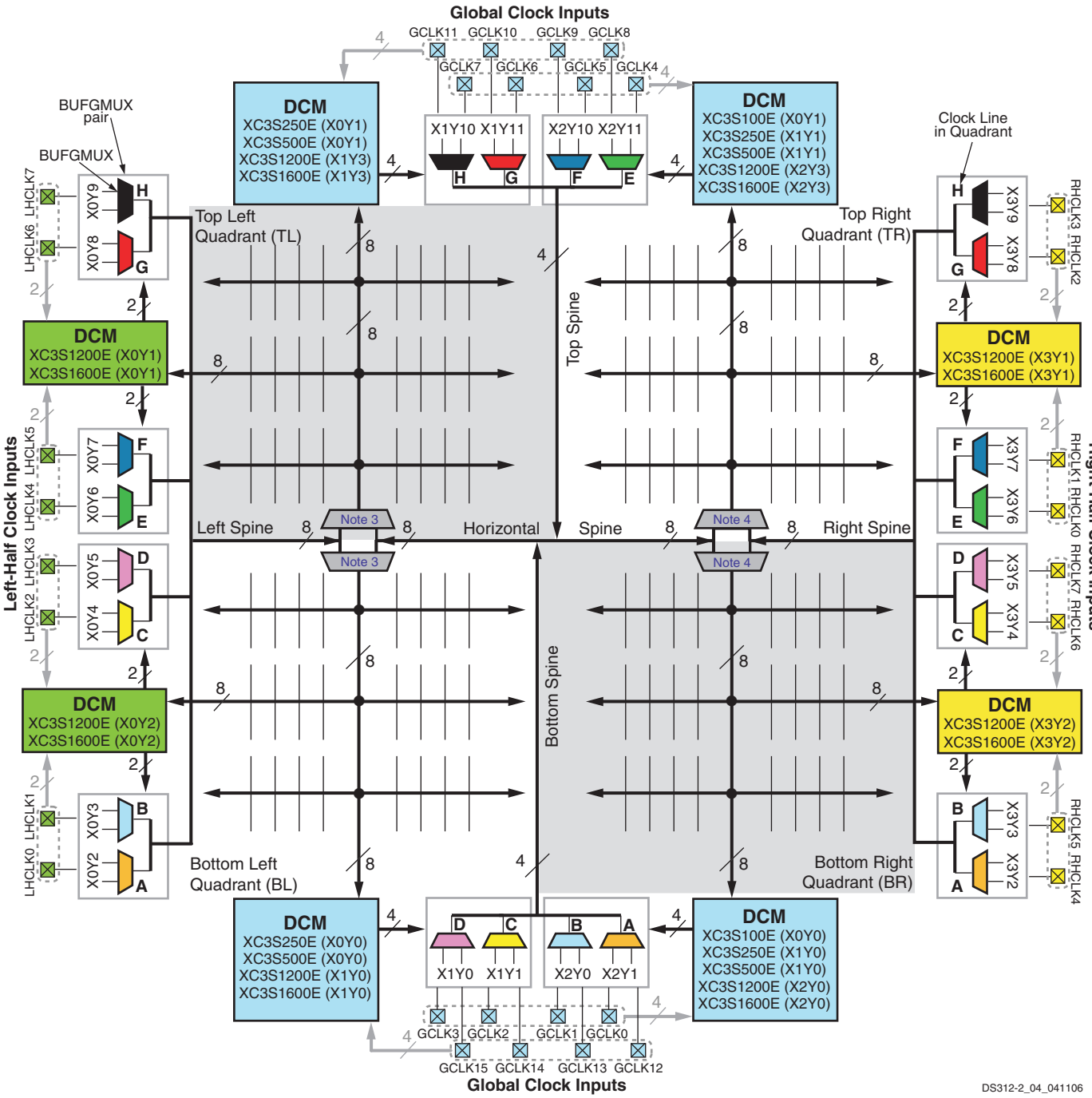


Figure 44: **NONE and FIXED Phase Shifter Waveforms (ISE 8.1i, Service Pack 3 and later)**



Notes:

1. The diagram presents electrical connectivity. The diagram locations do not necessarily match the physical location on the device, although the coordinate locations shown are correct.
2. Number of DCMs and locations of these DCM varies for different device densities. The left and right DCMs are only in the XC3S1200E and XC3S1600E. The XC3S100E has only two DCMs, one on the top right and one on the bottom right of the die.
3. See Figure 47a, which shows how the eight clock lines are multiplexed on the left-hand side of the device.
4. See Figure 47b, which shows how the eight clock lines are multiplexed on the right-hand side of the device.
5. For best direct clock inputs to a particular clock buffer, not a DCM, see Table 41.
6. For best direct clock inputs to a particular DCM, not a BUFGMUX, see Table 30, Table 31, and Table 32. Direct pin inputs to a DCM are shown in gray.

Figure 45: **Spartan-3E Internal Quadrant-Based Clock Network (Electrical Connectivity View)**

(P) Similarly, the FPGA's HSWAP pin must be Low to enable pull-up resistors on all user-I/O pins during configuration or High to disable the pull-up resistors. The HSWAP control must remain at a constant logic level throughout FPGA configuration. After configuration, when the FPGA's DONE output goes High, the HSWAP pin is available as full-featured user-I/O pin and is powered by the VCCO_0 supply.

The FPGA's DOUT pin is used in daisy-chain applications, described later. In a single-FPGA application, the FPGA's DOUT pin is not used but is actively driving during the configuration process.

Table 50: **Serial Master Mode Connections**

Pin Name	FPGA Direction	Description	During Configuration	After Configuration
HSWAP (P)	Input	**User I/O Pull-Up Control**. When Low during configuration, enables pull-up resistors in all I/O pins to respective I/O bank V_{CCO} input. 0: Pull-ups during configuration 1: No pull-ups	Drive at valid logic level throughout configuration.	User I/O
M[2:0]	Input	**Mode Select**. Selects the FPGA configuration mode. See Design Considerations for the HSWAP, M[2:0], and VS[2:0] Pins.	M2 = 0, M1 = 0, M0 = 0. Sampled when INIT_B goes High.	User I/O
DIN	Input	**Serial Data Input**.	Receives serial data from PROM's D0 output.	User I/O
CCLK	Output	**Configuration Clock**. Generated by FPGA internal oscillator. Frequency controlled by ***ConfigRate*** bitstream generator option. If CCLK PCB trace is long or has multiple connections, terminate this output to maintain signal integrity. See CCLK Design Considerations.	Drives PROM's CLK clock input.	User I/O
DOUT	Output	**Serial Data Output**.	Actively drives. Not used in single-FPGA designs. In a daisy-chain configuration, this pin connects to DIN input of the next FPGA in the chain.	User I/O
INIT_B	Open-drain bidirectional I/O	**Initialization Indicator**. Active Low. Goes Low at start of configuration during Initialization memory clearing process. Released at end of memory clearing, when mode select pins are sampled. Requires external 4.7 kΩ pull-up resistor to VCCO_2.	Connects to PROM's OE/RESET input. FPGA clears PROM's address counter at start of configuration, enables outputs during configuration. PROM also holds FPGA in Initialization state until PROM reaches Power-On Reset (POR) state. If CRC error detected during configuration, FPGA drives INIT_B Low.	User I/O. If unused in the application, drive INIT_B High.
DONE	Open-drain bidirectional I/O	**FPGA Configuration Done**. Low during configuration. Goes High when FPGA successfully completes configuration. Requires external 330 Ω pull-up resistor to 2.5V.	Connects to PROM's chip-enable (CE) input. Enables PROM during configuration. Disables PROM after configuration.	Pulled High via external pull-up. When High, indicates that the FPGA successfully configured.
PROG_B	Input	**Program FPGA**. Active Low. When asserted Low for 500 ns or longer, forces the FPGA to restart its configuration process by clearing configuration memory and resetting the DONE and INIT_B pins once PROG_B returns High. Recommend external 4.7 kΩ pull-up resistor to 2.5V. Internal pull-up value may be weaker (see Table 78). If driving externally with a 3.3V output, use an open-drain or open-collector driver or use a current limiting series resistor.	Must be High during configuration to allow configuration to start. Connects to PROM's CF pin, allowing JTAG PROM programming algorithm to reprogram the FPGA.	Drive PROG_B Low and release to reprogram FPGA.

read operations at this time. Spartan-3E FPGAs issue the read command just once. If the SPI Flash is not ready, then the FPGA does not properly configure.

If the 3.3V supply is last in the sequence and does not ramp fast enough, or if the SPI Flash PROM cannot be ready when required by the FPGA, delay the FPGA configuration process by holding either the FPGA's PROG_B input or INIT_B input Low, as highlighted in Figure 54. Release the FPGA when the SPI Flash PROM is ready. For example, a simple R-C delay circuit attached to the INIT_B pin forces the FPGA to wait for a preselected amount of time. Alternately, a Power Good signal from the 3.3V supply or a system reset signal accomplishes the same purpose. Use an open-drain or open-collector output when driving PROG_B or INIT_B.

SPI Flash PROM Density Requirements

Table 57 shows the smallest usable SPI Flash PROM to program a single Spartan-3E FPGA. Commercially available SPI Flash PROMs range in density from 1 Mbit to 128 Mbits. A multiple-FPGA daisy-chained application requires a SPI Flash PROM large enough to contain the sum of the FPGA file sizes. An application can also use a larger-density SPI Flash PROM to hold additional data beyond just FPGA configuration data. For example, the SPI Flash PROM can also store application code for a MicroBlaze™ RISC processor core integrated in the Spartan-3E FPGA. See Using the SPI Flash Interface after Configuration.

Table 57: **Number of Bits to Program a Spartan-3E FPGA and Smallest SPI Flash PROM**

Device	Number of Configuration Bits	Smallest Usable SPI Flash PROM
XC3S100E	581,344	1 Mbit
XC3S250E	1,353,728	2 Mbit
XC3S500E	2,270,208	4 Mbit
XC3S1200E	3,841,184	4 Mbit
XC3S1600E	5,969,696	8 Mbit

CCLK Frequency

In SPI Flash mode, the FPGA's internal oscillator generates the configuration clock frequency. The FPGA provides this clock on its CCLK output pin, driving the PROM's clock input pin. The FPGA starts configuration at its lowest frequency and increases its frequency for the remainder of the configuration process if so specified in the configuration bitstream. The maximum frequency is specified using the ***ConfigRate*** bitstream generator option. The maximum frequency supported by the FPGA configuration logic depends on the timing for the SPI Flash device. Without examining the timing for a specific SPI Flash PROM, use ***ConfigRate*** = 12 or lower. SPI Flash PROMs that support the FAST READ command support higher data rates. Some such PROMs support up to ***ConfigRate*** = 25 and beyond but require careful data sheet analysis. See Serial Peripheral Interface (SPI) Configuration Timing for more detailed timing analysis.

Using the SPI Flash Interface after Configuration

After the FPGA successfully completes configuration, all of the pins connected to the SPI Flash PROM are available as user-I/O pins.

If not using the SPI Flash PROM after configuration, drive CSO_B High to disable the PROM. The MOSI, DIN, and CCLK pins are then available to the FPGA application.

Because all the interface pins are user I/O after configuration, the FPGA application can continue to use the SPI Flash interface pins to communicate with the SPI Flash PROM, as shown in Figure 56. SPI Flash PROMs offer random-accessible, byte-addressable, read/write, non-volatile storage to the FPGA application.

SPI Flash PROMs are available in densities ranging from 1 Mbit up to 128 Mbits. However, a single Spartan-3E FPGA requires less than 6 Mbits. If desired, use a larger SPI Flash PROM to contain additional non-volatile application data, such as MicroBlaze processor code, or other user data such as serial numbers and Ethernet MAC IDs. In the example shown in Figure 56, the FPGA configures from SPI Flash PROM. Then using FPGA logic after configuration, the FPGA copies MicroBlaze code from SPI Flash into external DDR SDRAM for code execution. Similarly, the FPGA application can store non-volatile application data within the SPI Flash PROM.

The FPGA configuration data is stored starting at location 0. Store any additional data beginning in the next available SPI Flash PROM sector or page. Do not mix configuration data and user data in the same sector or page.

Similarly, the SPI bus can be expanded to additional SPI peripherals. Because SPI is a common industry-standard interface, various SPI-based peripherals are available, such as analog-to-digital (A/D) converters, digital-to-analog (D/A) converters, CAN controllers, and temperature sensors. However, if sufficient I/O pins are available in the application, Xilinx recommends creating a separate SPI bus to control peripherals. Creating a second port reduces the loading on the CCLK and DIN pins, which are crucial for configuration.

The MOSI, DIN, and CCLK pins are common to all SPI peripherals. Connect the select input on each additional SPI peripheral to one of the FPGA user I/O pins. If HSWAP = 0 during configuration, the FPGA holds the select line High. If HSWAP = 1, connect the select line to +3.3V via an external 4.7 kΩ pull-up resistor to avoid spurious read or write operations. After configuration, drive the select line Low to select the desired SPI peripheral.

Power Supply Specifications

Table 74: **Supply Voltage Thresholds for Power-On Reset**

Symbol	Description	Min	Max	Units
V_{CCINTT}	Threshold for the V_{CCINT} supply	0.4	1.0	V
V_{CCAUXT}	Threshold for the V_{CCAUX} supply	0.8	2.0	V
V_{CCO2T}	Threshold for the V_{CCO} Bank 2 supply	0.4	1.0	V

Notes:

1. V_{CCINT}, V_{CCAUX}, and V_{CCO} supplies to the FPGA can be applied in any order. However, the FPGA's configuration source (Platform Flash, SPI Flash, parallel NOR Flash, microcontroller) might have specific requirements. Check the data sheet for the attached configuration source. In Step 0 devices using the HSWAP internal pull-up, V_{CCINT} must be applied before V_{CCAUX}.
2. To ensure successful power-on, V_{CCINT}, V_{CCO} Bank 2, and V_{CCAUX} supplies must rise through their respective threshold-voltage ranges with no dips at any point.

Table 75: **Supply Voltage Ramp Rate**

Symbol	Description	Min	Max	Units
V_{CCINTR}	Ramp rate from GND to valid V_{CCINT} supply level	0.2	50	ms
V_{CCAUXR}	Ramp rate from GND to valid V_{CCAUX} supply level	0.2	50	ms
V_{CCO2R}	Ramp rate from GND to valid V_{CCO} Bank 2 supply level	0.2	50	ms

Notes:

1. V_{CCINT}, V_{CCAUX}, and V_{CCO} supplies to the FPGA can be applied in any order. However, the FPGA's configuration source (Platform Flash, SPI Flash, parallel NOR Flash, microcontroller) might have specific requirements. Check the data sheet for the attached configuration source. In Step 0 devices using the HSWAP internal pull-up, V_{CCINT} must be applied before V_{CCAUX}.
2. To ensure successful power-on, V_{CCINT}, V_{CCO} Bank 2, and V_{CCAUX} supplies must rise through their respective threshold-voltage ranges with no dips at any point.

Table 76: **Supply Voltage Levels Necessary for Preserving RAM Contents**

Symbol	Description	Min	Units
V_{DRINT}	V_{CCINT} level required to retain RAM data	1.0	V
V_{DRAUX}	V_{CCAUX} level required to retain RAM data	2.0	V

Notes:

1. RAM contents include configuration data.

Quiescent Current Requirements

Table 79: **Quiescent Supply Current Characteristics**

Symbol	Description	Device	Typical	Commercial Maximum(1)	Industrial Maximum(1)	Units
I_{CCINTQ}	Quiescent V_{CCINT} supply current	XC3S100E	8	27	36	mA
		XC3S250E	15	78	104	mA
		XC3S500E	25	106	145	mA
		XC3S1200E	50	259	324	mA
		XC3S1600E	65	366	457	mA
I_{CCOQ}	Quiescent V_{CCO} supply current	XC3S100E	0.8	1.0	1.5	mA
		XC3S250E	0.8	1.0	1.5	mA
		XC3S500E	0.8	1.0	1.5	mA
		XC3S1200E	1.5	2.0	2.5	mA
		XC3S1600E	1.5	2.0	2.5	mA
I_{CCAUXQ}	Quiescent V_{CCAUX} supply current	XC3S100E	8	12	13	mA
		XC3S250E	12	22	26	mA
		XC3S500E	18	31	34	mA
		XC3S1200E	35	52	59	mA
		XC3S1600E	45	76	86	mA

Notes:

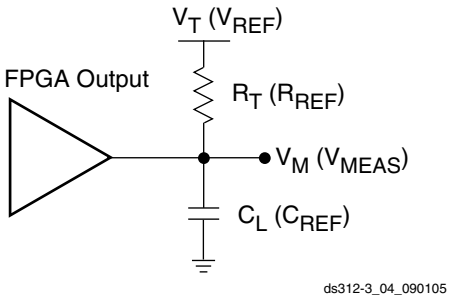
1. The maximum numbers in this table indicate the minimum current each power rail requires in order for the FPGA to power-on successfully.
2. The numbers in this table are based on the conditions set forth in Table 77.
3. Quiescent supply current is measured with all I/O drivers in a high-impedance state and with all pull-up/pull-down resistors at the I/O pads disabled. Typical values are characterized using typical devices at room temperature (T_J of 25°C at V_{CCINT} = 1.2 V, V_{CCO} = 3.3V, and V_{CCAUX} = 2.5V). The maximum limits are tested for each device at the respective maximum specified junction temperature and at maximum voltage limits with V_{CCINT} = 1.26V, V_{CCO} = 3.465V, and V_{CCAUX} = 2.625V. The FPGA is programmed with a "blank" configuration data file (i.e., a design with no functional elements instantiated). For conditions other than those described above, (e.g., a design including functional elements), measured quiescent current levels may be different than the values in the table. For more accurate estimates for a specific design, use the Xilinx® XPower tools.
4. There are two recommended ways to estimate the total power consumption (quiescent plus dynamic) for a specific design: a) The Spartan-3E XPower Estimator provides quick, approximate, typical estimates, and does not require a netlist of the design. b) XPower Analyzer uses a netlist as input to provide maximum estimates as well as more accurate typical estimates.

Timing Measurement Methodology

When measuring timing parameters at the programmable I/Os, different signal standards call for different test conditions. Table 95 lists the conditions to use for each standard.

The method for measuring Input timing is as follows: A signal that swings between a Low logic level of V_L and a High logic level of V_H is applied to the Input under test. Some standards also require the application of a bias voltage to the V_{REF} pins of a given bank to properly set the input-switching threshold. The measurement point of the Input signal (V_M) is commonly located halfway between V_L and V_H.

The Output test setup is shown in Figure 72. A termination voltage V_T is applied to the termination resistor R_T, the other end of which is connected to the Output. For each standard, R_T and V_T generally take on the standard values recommended for minimizing signal reflections. If the standard does not ordinarily use terminations (e.g., LVCMOS, LVTTL), then R_T is set to 1MΩ to indicate an open connection, and V_T is set to zero. The same measurement point (V_M) that was used at the Input is also used at the Output.



Notes:

1. The names shown in parentheses are used in the IBIS file.

Figure 72: **Output Test Setup**

Table 95: **Test Methods for Timing Measurement at I/Os**

Signal Standard (IOSTANDARD)		Inputs			Outputs		Inputs and Outputs
		V_{REF} (V)	V_L (V)	V_H (V)	R_T (Ω)	V_T (V)	V_M (V)
Single-Ended							
LVTTL		-	0	3.3	1M	0	1.4
LVCMOS33		-	0	3.3	1M	0	1.65
LVCMOS25		-	0	2.5	1M	0	1.25
LVCMOS18		-	0	1.8	1M	0	0.9
LVCMOS15		-	0	1.5	1M	0	0.75
LVCMOS12		-	0	1.2	1M	0	0.6
PCI33_3	Rising	-	Note 3	Note 3	25	0	0.94
	Falling				25	3.3	2.03
PCI66_3	Rising	-	Note 3	Note 3	25	0	0.94
	Falling				25	3.3	2.03
HSTL_I_18		0.9	V_{REF} – 0.5	V_{REF} + 0.5	50	0.9	V_{REF}
HSTL_III_18		1.1	V_{REF} – 0.5	V_{REF} + 0.5	50	1.8	V_{REF}
SSTL18_I		0.9	V_{REF} – 0.5	V_{REF} + 0.5	50	0.9	V_{REF}
SSTL2_I		1.25	V_{REF} – 0.75	V_{REF} + 0.75	50	1.25	V_{REF}
Differential							
LVDS_25		-	V_{ICM} – 0.125	V_{ICM} + 0.125	50	1.2	V_{ICM}
BLVDS_25		-	V_{ICM} – 0.125	V_{ICM} + 0.125	1M	0	V_{ICM}
MINI_LVDS_25		-	V_{ICM} – 0.125	V_{ICM} + 0.125	50	1.2	V_{ICM}
LVPECL_25		-	V_{ICM} – 0.3	V_{ICM} + 0.3	1M	0	V_{ICM}
RSDS_25		-	V_{ICM} – 0.1	V_{ICM} + 0.1	50	1.2	V_{ICM}

IEEE 1149.1/1532 JTAG Test Access Port Timing

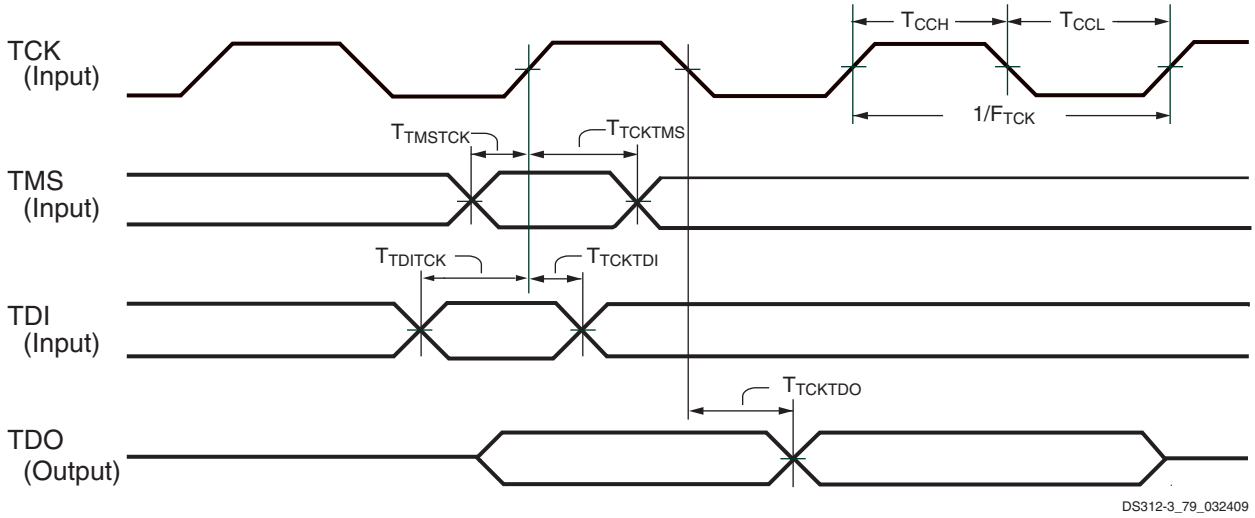


Figure 78: **JTAG Waveforms**

Table 123: **Timing for the JTAG Test Access Port**

Symbol	Description	All Speed Grades		Units
		Min	Max	
Clock-to-Output Times				
T_{TCKTDO}	The time from the falling transition on the TCK pin to data appearing at the TDO pin	1.0	11.0	ns
Setup Times				
T_{TDITCK}	The time from the setup of data at the TDI pin to the rising transition at the TCK pin	7.0	-	ns
T_{TMSTCK}	The time from the setup of a logic level at the TMS pin to the rising transition at the TCK pin	7.0	-	ns
Hold Times				
T_{TCKTDI}	The time from the rising transition at the TCK pin to the point when data is last held at the TDI pin	0	-	ns
T_{TCKTMS}	The time from the rising transition at the TCK pin to the point when a logic level is last held at the TMS pin	0	-	ns
Clock Timing				
T_{CCH}	The High pulse width at the TCK pin	5	-	ns
T_{CCL}	The Low pulse width at the TCK pin	5	-	ns
F_{TCK}	Frequency of the TCK signal	-	30	MHz

Notes:

1. The numbers in this table are based on the operating conditions set forth in Table 77.

Table 124: **Types of Pins on Spartan-3E FPGAs *(Cont'd)***

Type / Color Code	Description	Pin Name(s) in Type[1]
VCCAUX	Dedicated auxiliary power supply pin. The number of VCCAUX pins depends on the package used. All must be connected to +2.5V. See the Powering Spartan-3E FPGAs section in Module 2 for details.	VCCAUX
VCCINT	Dedicated internal core logic power supply pin. The number of VCCINT pins depends on the package used. All must be connected to +1.2V. See the Powering Spartan-3E FPGAs section in Module 2 for details.	VCCINT
VCCO	Along with all the other VCCO pins in the same bank, this pin supplies power to the output buffers within the I/O bank and sets the input threshold voltage for some I/O standards. See the Powering Spartan-3E FPGAs section in Module 2 for details.	VCCO_#
N.C.	This package pin is not connected in this specific device/package combination but may be connected in larger devices in the same package.	N.C.

Notes:

1. # = I/O bank number, an integer between 0 and 3.
2. IRDY/TRDY designations are for PCI designs; refer to PCI documentation for details.

Differential Pair Labeling

I/Os with Lxxy_# are part of a differential pair. 'L' indicates differential capability. The 'xx' field is a two-digit integer, unique to each bank that identifies a differential pin-pair. The 'y' field is either 'P' for the true signal or 'N' for the inverted signal in the differential pair. The '#' field is the I/O bank number.

The pin name suffix has the following significance. Figure 79 provides a specific example showing a differential input to and a differential output from Bank 1.

'L' indicates that the pin is part of a differential pair.

'xx' is a two-digit integer, unique for each bank, that identifies a differential pin-pair.

'y' is replaced by 'P' for the true signal or 'N' for the inverted. These two pins form one differential pin-pair.

'#' is an integer, 0 through 3, indicating the associated I/O bank.

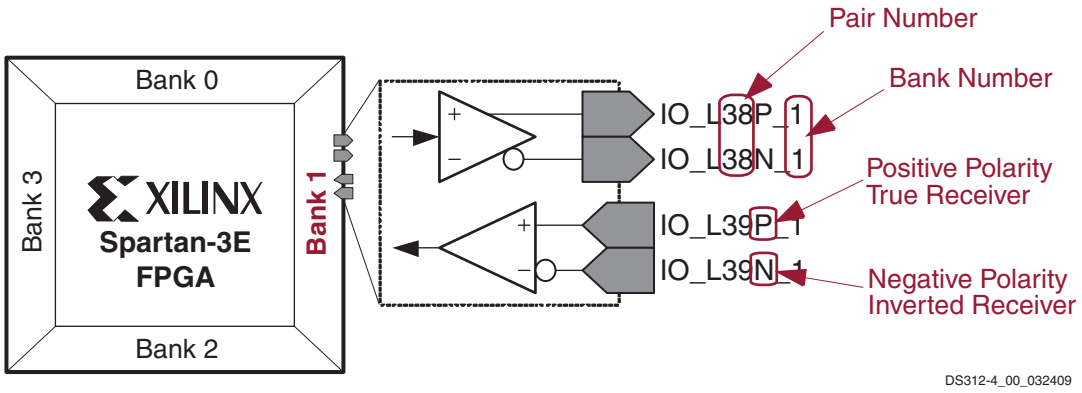


Figure 79: **Differential Pair Labeling**

User I/Os by Bank

Table 134 shows how the 83 available user-I/O pins are distributed on the XC3S100E FPGA packaged in the CP132 package. Table 135 indicates how the 92 available user-I/O pins are distributed on the XC3S250E and the XC3S500E FPGAs in the CP132 package.

Table 134: **User I/Os Per Bank for the XC3S100E in the CP132 Package**

Package Edge	I/O Bank	Maximum I/O	All Possible I/O Pins by Type				
			I/O	INPUT	DUAL	VREF(1)	CLK(2)
Top	0	18	6	2	1	1	8
Right	1	23	0	0	21	2	0(2)
Bottom	2	22	0	0	20	2	0(2)
Left	3	20	10	0	0	2	8
TOTAL		**83**	**16**	**2**	**42**	**7**	**16**

Notes:

1. Some VREF and CLK pins are on INPUT pins.
2. The eight global clock pins in this bank have optional functionality during configuration and are counted in the DUAL column.

Table 135: **User I/Os Per Bank for the XC3S250E and XC3S500E in the CP132 Package**

Package Edge	I/O Bank	Maximum I/O	All Possible I/O Pins by Type				
			I/O	INPUT	DUAL	VREF(1)	CLK(2)
Top	0	22	11	0	1	2	8
Right	1	23	0	0	21	2	0(2)
Bottom	2	26	0	0	24	2	0(2)
Left	3	21	11	0	0	2	8
TOTAL		**92**	**22**	**0**	**46**	**8**	**16**

Notes:

1. Some VREF and CLK pins are on INPUT pins.
2. The eight global clock pins in this bank have optional functionality during configuration and are counted in the DUAL column.

FT256: 256-ball Fine-pitch, Thin Ball Grid Array

The 256-ball fine-pitch, thin ball grid array package, FT256, supports three different Spartan-3E FPGAs, including the XC3S250E, the XC3S500E, and the XC3S1200E.

Table 143 lists all the package pins. They are sorted by bank number and then by pin name of the largest device. Pins that form a differential I/O pair appear together in the table. The table also shows the pin number for each pin and the pin type, as defined earlier.

The highlighted rows indicate pinout differences between the XC3S250E, the XC3S500E, and the XC3S1200E FPGAs. The XC3S250E has 18 unconnected balls, indicated as N.C. (No Connection) in Table 143 and with the black diamond character (◆) in Table 143 and Figure 83.

If the table row is highlighted in tan, then this is an instance where an unconnected pin on the XC3S250E FPGA maps to a VREF pin on the XC3S500E and XC3S1200E FPGA. If the FPGA application uses an I/O standard that requires a VREF voltage reference, connect the highlighted pin to the VREF voltage supply, even though this does not actually connect to the XC3S250E FPGA. This VREF connection on the board allows future migration to the larger devices without modifying the printed-circuit board.

All other balls have nearly identical functionality on all three devices. Table 147 summarizes the Spartan-3E footprint migration differences for the FT256 package.

An electronic version of this package pinout table and footprint diagram is available for download from the Xilinx web site at:

http://www.xilinx.com/support/documentation/data_sheets/s3e_pin.zip

Pinout Table

Table 143: **FT256 Package Pinout**

Bank	XC3S250E Pin Name	XC3S500E Pin Name	XC3S1200E Pin Name	FT256 Ball	Type
0	IO	IO	IO	A7	I/O
0	IO	IO	IO	A12	I/O
0	IO	IO	IO	B4	I/O
0	IP	IP	IO	B6	***250E:*** INPUT ***500E:*** INPUT ***1200E:*** I/O
0	IP	IP	IO	B10	***250E:*** INPUT ***500E:*** INPUT ***1200E:*** I/O
0	IO/VREF_0	IO/VREF_0	IO/VREF_0	D9	VREF
0	IO_L01N_0	IO_L01N_0	IO_L01N_0	A14	I/O
0	IO_L01P_0	IO_L01P_0	IO_L01P_0	B14	I/O
0	IO_L03N_0/VREF_0	IO_L03N_0/VREF_0	IO_L03N_0/VREF_0	A13	VREF
0	IO_L03P_0	IO_L03P_0	IO_L03P_0	B13	I/O
0	IO_L04N_0	IO_L04N_0	IO_L04N_0	E11	I/O
0	IO_L04P_0	IO_L04P_0	IO_L04P_0	D11	I/O
0	IO_L05N_0/VREF_0	IO_L05N_0/VREF_0	IO_L05N_0/VREF_0	B11	VREF
0	IO_L05P_0	IO_L05P_0	IO_L05P_0	C11	I/O
0	IO_L06N_0	IO_L06N_0	IO_L06N_0	E10	I/O
0	IO_L06P_0	IO_L06P_0	IO_L06P_0	D10	I/O
0	IO_L08N_0/GCLK5	IO_L08N_0/GCLK5	IO_L08N_0/GCLK5	F9	GCLK
0	IO_L08P_0/GCLK4	IO_L08P_0/GCLK4	IO_L08P_0/GCLK4	E9	GCLK
0	IO_L09N_0/GCLK7	IO_L09N_0/GCLK7	IO_L09N_0/GCLK7	A9	GCLK
0	IO_L09P_0/GCLK6	IO_L09P_0/GCLK6	IO_L09P_0/GCLK6	A10	GCLK
0	IO_L11N_0/GCLK11	IO_L11N_0/GCLK11	IO_L11N_0/GCLK11	D8	GCLK
0	IO_L11P_0/GCLK10	IO_L11P_0/GCLK10	IO_L11P_0/GCLK10	C8	GCLK
0	IO_L12N_0	IO_L12N_0	IO_L12N_0	F8	I/O