Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Active
Number of LABs/CLBs	612
Number of Logic Elements/Cells	5508
Total RAM Bits	221184
Number of I/O	172
Number of Gates	250000
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	256-LBGA
Supplier Device Package	256-FTBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xc3s250e-4ft256c

Double-Data-Rate Transmission

Double-Data-Rate (DDR) transmission describes the technique of synchronizing signals to both the rising and falling edges of the clock signal. Spartan-3E devices use register pairs in all three IOB paths to perform DDR operations.

The pair of storage elements on the IOB's Output path (OFF1 and OFF2), used as registers, combine with a special multiplexer to form a DDR D-type flip-flop (ODDR2). This primitive permits DDR transmission where output data bits are synchronized to both the rising and falling edges of a clock. DDR operation requires two clock signals (usually 50% duty cycle), one the inverted form of the other. These signals trigger the two registers in alternating fashion, as shown in Figure 7. The Digital Clock Manager (DCM) generates the two clock signals by mirroring an incoming signal, and then shifting it 180 degrees. This approach ensures minimal skew between the two signals. Alternatively, the inverter inside the IOB can be used to invert the clock signal, thus only using one clock line and both rising and falling edges of that clock line as the two clocks for the DDR flip-flops.

The storage-element pair on the Three-State path (TFF1 and TFF2) also can be combined with a local multiplexer to form a DDR primitive. This permits synchronizing the output enable to both the rising and falling edges of a clock. This DDR operation is realized in the same way as for the output path.

The storage-element pair on the input path (IFF1 and IFF2) allows an I/O to receive a DDR signal. An incoming DDR clock signal triggers one register, and the inverted clock signal triggers the other register. The registers take turns capturing bits of the incoming DDR data signal. The primitive to allow this functionality is called IDDR2.

Aside from high bandwidth data transfers, DDR outputs also can be used to reproduce, or *mirror*, a clock signal on the output. This approach is used to transmit clock and data signals together (source synchronously). A similar approach is used to reproduce a clock signal at multiple outputs. The advantage for both approaches is that skew across the outputs is minimal.

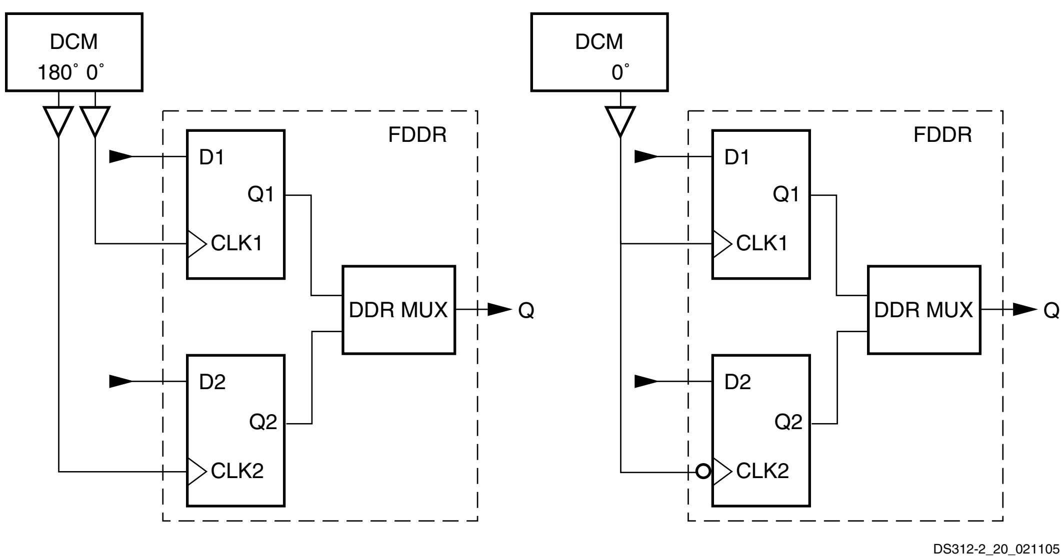

Figure 7: **Two Methods for Clocking the DDR Register**

Register Cascade Feature

In the Spartan-3E family, one of the IOBs in a differential pair can cascade its input storage elements with those in the other IOB as part of a differential pair. This is intended to make DDR operation at high speed much simpler to implement. The new DDR connections that are available are shown in Figure 5 (dashed lines), and are only available for routing between IOBs and are not accessible to the FPGA fabric. Note that this feature is only available when using the differential I/O standards LVDS, RSDS, and MINI_LVDS.

IDDR2

As a DDR input pair, the master IOB registers incoming data on the rising edge of ICLK1 (= D1) and the rising edge of ICLK2 (= D2), which is typically the same as the falling edge of ICLK1. This data is then transferred into the FPGA fabric. At some point, both signals must be brought into the same clock domain, typically ICLK1. This can be difficult at high frequencies because the available time is only one half of a clock cycle assuming a 50% duty cycle. See Figure 8 for a graphical illustration of this function.

Spartan-3E FPGAs provide additional input flexibility by allowing I/O standards to be mixed in different banks. For a particular V_{CCO} voltage, Table 6 and Table 7 list all of the IOSTANDARDs that can be combined and if the IOSTANDARD is supported as an input only or can be used for both inputs and outputs.

Table 6: **Single-Ended IOSTANDARD Bank Compatibility**

Single-Ended IOSTANDARD	V_{CCO} Supply/Compatibility					Input Requirements	
	1.2V	1.5V	1.8V	2.5V	3.3V	V_{REF}	Board Termination Voltage (V_{TT})
LVTTL	-	-	-	-	Input/ Output	N/R[1]	N/R
LVCMOS33	-	-	-	-	Input/ Output	N/R	N/R
LVCMOS25	-	-	-	Input/ Output	Input	N/R	N/R
LVCMOS18	-	-	Input/ Output	Input	Input	N/R	N/R
LVCMOS15	-	Input/ Output	Input	Input	Input	N/R	N/R
LVCMOS12	Input/ Output	Input	Input	Input	Input	N/R	N/R
PCI33_3	-	-	-	-	Input/ Output	N/R	N/R
PCI66_3	-	-	-	-	Input/ Output	N/R	N/R
HSTL_I_18	-	-	Input/ Output	Input	Input	0.9	0.9
HSTL_III_18	-	-	Input/ Output	Input	Input	1.1	1.8
SSTL18_I	-	-	Input/ Output	Input	Input	0.9	0.9
SSTL2_I	-	-	-	Input/ Output	Input	1.25	1.25

Notes:

1. N/R - Not required for input operation.

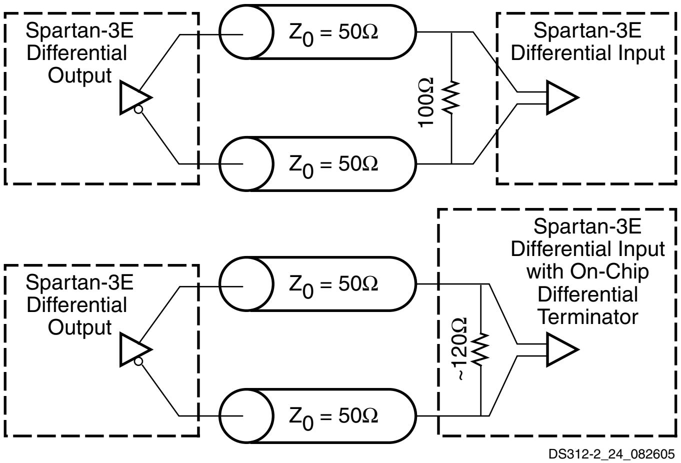

Figure 11: **Differential Inputs and Outputs**

Pull-Up and Pull-Down Resistors

Pull-up and pull-down resistors inside each IOB optionally force a floating I/O or Input-only pin to a determined state. Pull-up and pull-down resistors are commonly applied to unused I/Os, inputs, and three-state outputs, but can be used on any I/O or Input-only pin. The pull-up resistor connects an IOB to V_{CCO} through a resistor. The resistance value depends on the V_{CCO} voltage (see Module 3, DC and Switching Characteristics for the specifications). The pull-down resistor similarly connects an IOB to ground with a resistor. The PULLUP and PULLDOWN attributes and library primitives turn on these optional resistors.

By default, PULLDOWN resistors terminate all unused I/O and Input-only pins. Unused I/O and Input-only pins can alternatively be set to PULLUP or FLOAT. To change the unused I/O Pad setting, set the Bitstream Generator (BitGen) option ***UnusedPin*** to PULLUP, PULLDOWN, or FLOAT. The ***UnusedPin*** option is accessed through the Properties for Generate Programming File in ISE. See Bitstream Generator (BitGen) Options.

During configuration a Low logic level on the HSWAP pin activates pull-up resistors on all I/O and Input-only pins not actively used in the selected configuration mode.

Keeper Circuit

Each I/O has an optional keeper circuit (see Figure 12) that keeps bus lines from floating when not being actively driven. The KEEPER circuit retains the last logic level on a line after all drivers have been turned off. Apply the KEEPER attribute or use the KEEPER library primitive to use the KEEPER circuitry. Pull-up and pull-down resistors override the KEEPER settings.

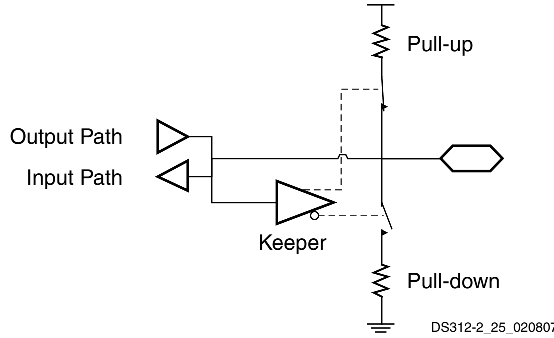

Figure 12: **Keeper Circuit**

Slew Rate Control and Drive Strength

Each IOB has a slew-rate control that sets the output switching edge-rate for LVCMOS and LVTTL outputs. The SLEW attribute controls the slew rate and can either be set to SLOW (default) or FAST.

Each LVCMOS and LVTTL output additionally supports up to six different drive current strengths as shown in Table 8. To adjust the drive strength for each output, the DRIVE attribute is set to the desired drive strength: 2, 4, 6, 8, 12, and 16. Unless otherwise specified in the FPGA application, the software default IOSTANDARD is LVCMOS25, SLOW slew rate, and 12 mA output drive.

Table 8: **Programmable Output Drive Current**

IOSTANDARD	Output Drive Current (mA)					
	2	4	6	8	12	16
LVTTL	✔	✔	✔	✔	✔	✔
LVCMOS33	✔	✔	✔	✔	✔	✔
LVCMOS25	✔	✔	✔	✔	✔	-
LVCMOS18	✔	✔	✔	✔	-	-
LVCMOS15	✔	✔	✔	-	-	-
LVCMOS12	✔	-	-	-	-	-

High output current drive strength and FAST output slew rates generally result in fastest I/O performance. However, these same settings generally also result in transmission line effects on the printed circuit board (PCB) for all but the shortest board traces. Each IOB has independent slew rate and drive strength controls. Use the slowest slew rate and lowest output drive current that meets the performance requirements for the end application.

Likewise, due to lead inductance, a given package supports a limited number of simultaneous switching outputs (SSOs) when using fast, high-drive outputs. Only use fast, high-drive outputs when required by the application.

Supply Voltages for the IOBs

The IOBs are powered by three supplies:

1. The V_{CCO} supplies, one for each of the FPGA's I/O banks, power the output drivers. The voltage on the V_{CCO} pins determines the voltage swing of the output signal.
2. V_{CCINT} is the main power supply for the FPGA's internal logic.
3. V_{CCAUX} is an auxiliary source of power, primarily to optimize the performance of various FPGA functions such as I/O switching.

I/O and Input-Only Pin Behavior During Power-On, Configuration, and User Mode

In this section, all behavior described for I/O pins also applies to input-only pins and dual-purpose I/O pins that are not actively involved in the currently-selected configuration mode.

All I/O pins have ESD clamp diodes to their respective V_{CCO} supply and from GND, as shown in Figure 5. The V_{CCINT} (1.2V), V_{CCAUX} (2.5V), and V_{CCO} supplies can be applied in any order. Before the FPGA can start its configuration process, V_{CCINT}, V_{CCO} Bank 2, and V_{CCAUX} must have reached their respective minimum recommended operating levels indicated in Table 74. At this time, all output drivers are in a high-impedance state. V_{CCO} Bank 2, V_{CCINT}, and V_{CCAUX} serve as inputs to the internal Power-On Reset circuit (POR).

A Low level applied to the HSWAP input enables pull-up resistors on user-I/O and input-only pins from power-on throughout configuration. A High level on HSWAP disables the pull-up resistors, allowing the I/Os to float. HSWAP contains an internal pull-up resistor and defaults to High if left floating. As soon as power is applied, the FPGA begins initializing its configuration memory. At the same time, the FPGA internally asserts the Global Set-Reset (GSR), which asynchronously resets all IOB storage elements to a default Low state. Also see Pin Behavior During Configuration.

Upon the completion of initialization and the beginning of configuration, INIT_B goes High, sampling the M0, M1, and M2 inputs to determine the configuration mode. Configuration data is then loaded into the FPGA. The I/O drivers remain in a high-impedance state (with or without pull-up resistors, as determined by the HSWAP input) throughout configuration.

At the end of configuration, the GSR net is released, placing the IOB registers in a Low state by default, unless the loaded design reverses the polarity of their respective SR inputs.

The Global Three State (GTS) net is released during Start-Up, marking the end of configuration and the beginning of design operation in the User mode. After the GTS net is released, all user I/Os go active while all unused I/Os are pulled down (PULLDOWN). The designer can control how the unused I/Os are terminated after GTS is released by setting the Bitstream Generator (BitGen) option UnusedPin to PULLUP, PULLDOWN, or FLOAT.

One clock cycle later (default), the Global Write Enable (GWE) net is released allowing the RAM and registers to change states. Once in User mode, any pull-up resistors enabled by HSWAP revert to the user settings and HSWAP is available as a general-purpose I/O. For more information on PULLUP and PULLDOWN, see Pull-Up and Pull-Down Resistors.

Behavior of Unused I/O Pins After Configuration

By default, the Xilinx ISE development software automatically configures all unused I/O pins as input pins with individual internal pull-down resistors to GND.

This default behavior is controlled by the ***UnusedPin*** bitstream generator (BitGen) option, as described in Table 69.

JTAG Boundary-Scan Capability

All Spartan-3E IOBs support boundary-scan testing compatible with IEEE 1149.1/1532 standards. During boundary-scan operations such as EXTEST and HIGHZ the pull-down resistor is active. See JTAG Mode for more information on programming via JTAG.

The connections for the bottom-edge BUFGMUX elements are similar to the top-edge connections (see Figure 46).

On the left and right edges, only two clock inputs feed each pair of BUFGMUX elements.

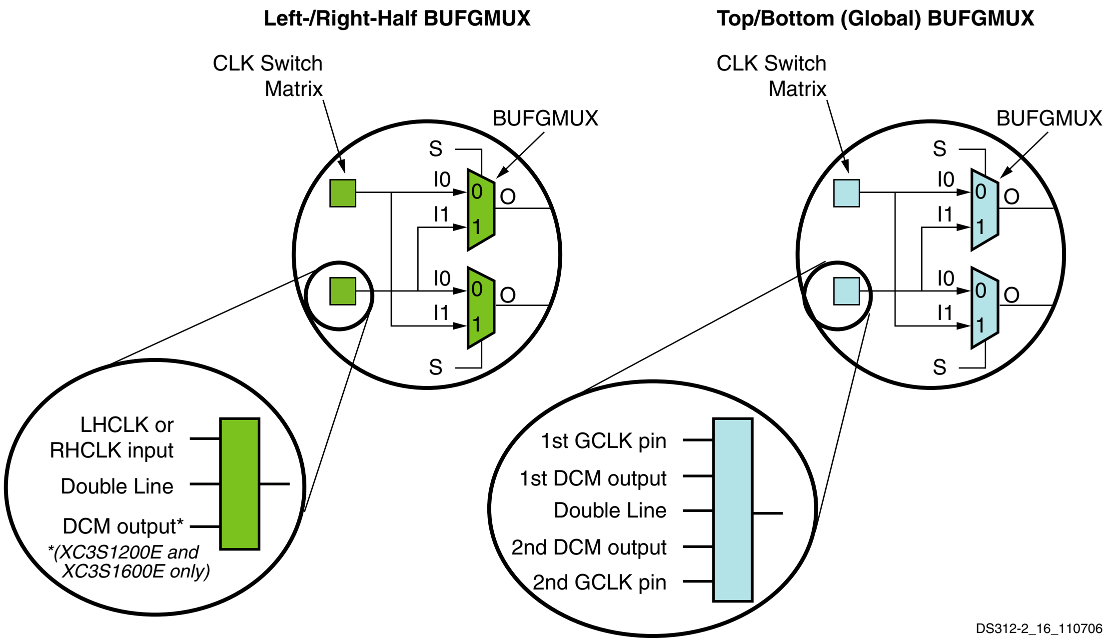

Figure 46: **Clock Switch Matrix to BUFGMUX Pair Connectivity**

Quadrant Clock Routing

The clock routing within the FPGA is quadrant-based, as shown in Figure 45. Each clock quadrant supports eight total clock signals, labeled 'A' through 'H' in Table 41 and Figure 47. The clock source for an individual clock line originates either from a global BUFGMUX element along the top and bottom edges or from a BUFGMUX element along the associated edge, as shown in Figure 47. The clock lines feed the synchronous resource elements (CLBs, IOBs, block RAM, multipliers, and DCMs) within the quadrant.

The four quadrants of the device are:

- Top Right (TR)
- Bottom Right (BR)
- Bottom Left (BL)
- Top Left (TL)

Note that the quadrant clock notation (TR, BR, BL, TL) is separate from that used for similar IOB placement constraints.

To estimate the quadrant location for a particular I/O, see the footprint diagrams in Module 4, Pinout Descriptions. For exact quadrant locations, use the floorplanning tool. In the QFP packages (VQ100, TQ144 and PQ208) the quadrant borders fall in the middle of each side of the package, at a GND pin. The clock inputs fall on the quadrant boundaries, as indicated in Table 42.

Table 42: **QFP Package Clock Quadrant Locations**

Clock Pins	Quadrant
GCLK[3:0]	BR
GCLK[7:4]	TR
GCLK[11:8]	TL
GCLK[15:12]	BL
RHCLK[3:0]	BR
RHCLK[7:4]	TR
LHCLK[3:0]	TL
LHCLK[7:4]	BL

In a few cases, a dedicated input is physically in one quadrant of the device but connects to a different clock quadrant:

- FT256, H16 is in clock quadrant BR
- FG320, K2 is in clock quadrant BL
- FG400, L8 is in clock quadrant TL and the I/O at N11 is in clock quadrant BL
- FG484, M2 is in clock quadrant TL and L15 is in clock quadrant BR

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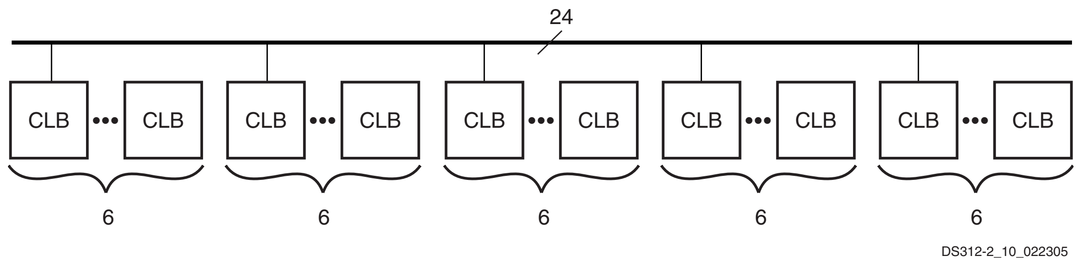
(horizontal channel shown as an example)</td><td>
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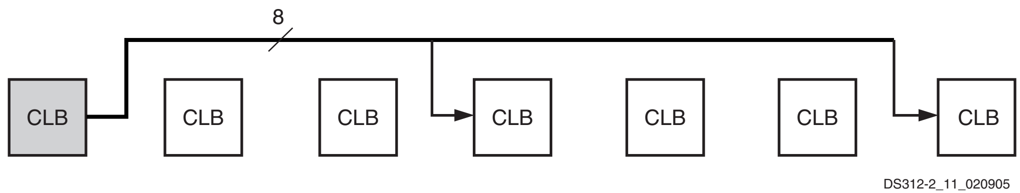
(horizontal channel shown as an example)</td><td>
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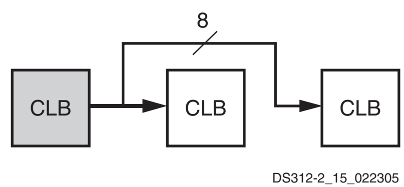
(horizontal channel shown as an example)</td><td>
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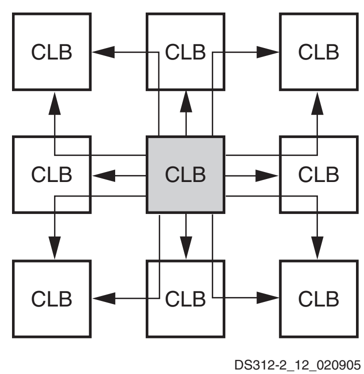
<tr><td>Direct Connections</td><td>
</td></tr>
</table>

Figure 50: **Interconnect Types between Two Adjacent Interconnect Tiles**

The four types of general-purpose interconnect available in each channel, shown in Figure 50, are described below.

Long Lines

Each set of 24 long line signals spans the die both horizontally and vertically and connects to one out of every six interconnect tiles. At any tile, four of the long lines drive or receive signals from a switch matrix. Because of their low capacitance, these lines are well-suited for carrying high-frequency signals with minimal loading effects (e.g. skew). If all global clock lines are already committed and additional clock signals remain to be assigned, long lines serve as a good alternative.

Hex Lines

Each set of eight hex lines are connected to one out of every three tiles, both horizontally and vertically. Thirty-two hex lines are available between any given interconnect tile. Hex lines are only driven from one end of the route.

Double Lines

Each set of eight double lines are connected to every other tile, both horizontally and vertically. in all four directions. Thirty-two double lines available between any given interconnect tile. Double lines are more connections and more flexibility, compared to long line and hex lines.

Voltage Compatibility

The PROM's V_{CCINT} supply must be either 3.3V for the serial XCFxxS Platform Flash PROMs or 1.8V for the serial/parallel XCFxxP PROMs.

(V) The FPGA's VCCO_2 supply input and the Platform Flash PROM's V_{CCO} supply input must be the same voltage, ideally +2.5V. Both devices also support 1.8V and 3.3V interfaces but the FPGA's PROG_B and DONE pins require special attention as they are powered by the FPGA's V_{CCAUX} supply, nominally 2.5V. See application note XAPP453: *The 3.3V Configuration of Spartan-3 FPGAs* for additional information.

Supported Platform Flash PROMs

Table 51 shows the smallest available Platform Flash PROM to program one Spartan-3E FPGA. A multiple-FPGA daisy-chain application requires a Platform Flash PROM large enough to contain the sum of the various FPGA file sizes.

Table 51: **Number of Bits to Program a Spartan-3E FPGA and Smallest Platform Flash PROM**

Spartan-3E FPGA	Number of Configuration Bits	Smallest Available Platform Flash
XC3S100E	581,344	XCF01S
XC3S250E	1,353,728	XCF02S
XC3S500E	2,270,208	XCF04S
XC3S1200E	3,841,184	XCF04S
XC3S1600E	5,969,696	XCF08P or 2 x XCF04S

The XC3S1600E requires an 8 Mbit PROM. Two solutions are possible: either a single 8 Mbit XCF08P parallel/serial PROM or two 4 Mbit XCF04S serial PROMs cascaded. The two XCF04S PROMs use a 3.3V V_{CCINT} supply while the XCF08P requires a 1.8V V_{CCINT} supply. If the board does not already have a 1.8V supply available, the two cascaded XCF04S PROM solution is recommended.

CCLK Frequency

In Master Serial mode, the FPGA's internal oscillator generates the configuration clock frequency. The FPGA provides this clock on its CCLK output pin, driving the PROM's CLK input pin. The FPGA starts configuration at its lowest frequency and increases its frequency for the remainder of the configuration process if so specified in the configuration bitstream. The maximum frequency is specified using the ***ConfigRate*** bitstream generator option. Table 52 shows the maximum ***ConfigRate*** settings, approximately equal to MHz, for various Platform Flash devices and I/O voltages. For the serial XCFxxS PROMs, the maximum frequency also depends on the interface voltage.

Table 52: **Maximum ConfigRate Settings for Platform Flash**

Platform Flash Part Number	I/O Voltage (VCCO_2, V_{CCO})	Maximum *ConfigRate* Setting
XCF01S XCF02S XCF04S	3.3V or 2.5V	25
	1.8V	12
XCF08P XCF16P XCF32P	3.3V, 2.5V, or 1.8V	25

Table 55: **Serial Peripheral Interface (SPI) Connections *(Cont'd)***

Pin Name	FPGA Direction	Description	During Configuration	After Configuration
VS[2:0] Ⓢ	Input	**Variant Select**. Instructs the FPGA how to communicate with the attached SPI Flash PROM. See Design Considerations for the HSWAP, M[2:0], and VS[2:0] Pins.	Must be at the logic levels shown in Table 53. Sampled when INIT_B goes High.	User I/O
MOSI	Output	**Serial Data Output**.	FPGA sends SPI Flash memory read commands and starting address to the PROM's serial data input.	User I/O
DIN	Input	**Serial Data Input**.	FPGA receives serial data from PROM's serial data output.	User I/O
CSO_B	Output	**Chip Select Output**. Active Low.	Connects to the SPI Flash PROM's chip-select input. If HSWAP = 1, connect this signal to a 4.7 kΩ pull-up resistor to 3.3V.	Drive CSO_B High after configuration to disable the SPI Flash and reclaim the MOSI, DIN, and CCLK pins. Optionally, re-use this pin and MOSI, DIN, and CCLK to continue communicating with SPI Flash.
CCLK	Output	**Configuration Clock**. Generated by FPGA internal oscillator. Frequency controlled by ***ConfigRate*** bitstream generator option. If CCLK PCB trace is long or has multiple connections, terminate this output to maintain signal integrity. See CCLK Design Considerations.	Drives PROM's clock input.	User I/O
DOUT	Output	**Serial Data Output**.	Actively drives. Not used in single-FPGA designs. In a daisy-chain configuration, this pin connects to DIN input of the next FPGA in the chain.	User I/O
INIT_B	Open-drain bidirectional I/O	**Initialization Indicator**. Active Low. Goes Low at start of configuration during Initialization memory clearing process. Released at end of memory clearing, when mode select pins are sampled. In daisy-chain applications, this signal requires an external 4.7 kΩ pull-up resistor to VCCO_2.	Active during configuration. If SPI Flash PROM requires > 2 ms to awake after powering on, hold INIT_B Low until PROM is ready. If CRC error detected during configuration, FPGA drives INIT_B Low.	User I/O. If unused in the application, drive INIT_B High.
DONE	Open-drain bidirectional I/O	**FPGA Configuration Done**. Low during configuration. Goes High when FPGA successfully completes configuration. Requires external 330 Ω pull-up resistor to 2.5V.	Low indicates that the FPGA is not yet configured.	Pulled High via external pull-up. When High, indicates that the FPGA successfully configured.
PROG_B	Input	**Program FPGA**. Active Low. When asserted Low for 500 ns or longer, forces the FPGA to restart its configuration process by clearing configuration memory and resetting the DONE and INIT_B pins once PROG_B returns High. Recommend external 4.7 kΩ pull-up resistor to 2.5V. Internal pull-up value may be weaker (see Table 78). If driving externally with a 3.3V output, use an open-drain or open-collector driver or use a current limiting series resistor.	Must be High to allow configuration to start.	Drive PROG_B Low and release to reprogram FPGA. Hold PROG_B to force FPGA I/O pins into Hi-Z, allowing direct programming access to SPI Flash PROM pins.

This addressing flexibility allows the FPGA to share the parallel Flash PROM with an external or embedded processor. Depending on the specific processor architecture, the processor boots either from the top or bottom of memory. The FPGA is flexible and boots from the opposite end of memory from the processor. Only the processor or the FPGA can boot at any given time. The FPGA can configure first, holding the processor in reset or the processor can boot first, asserting the FPGA's PROG_B pin.

The mode select pins, M[2:0], are sampled when the FPGA's INIT_B output goes High and must be at defined logic levels during this time. After configuration, when the FPGA's DONE output goes High, the mode pins are available as full-featured user-I/O pins.

(P) Similarly, the FPGA's HSWAP pin must be Low to enable pull-up resistors on all user-I/O pins or High to disable the pull-up resistors. The HSWAP control must remain at a constant logic level throughout FPGA configuration. After configuration, when the FPGA's DONE output goes High, the HSWAP pin is available as full-featured user-I/O pin and is powered by the VCCO_0 supply.

The RDWR_B and CSI_B must be Low throughout the configuration process. After configuration, these pins also become user I/O.

In a single-FPGA application, the FPGA's CSO_B and CCLK pins are not used but are actively driving during the configuration process. The BUSY pin is not used but also actively drives during configuration and is available as a user I/O after configuration.

After configuration, all of the interface pins except DONE and PROG_B are available as user I/Os. Furthermore, the bidirectional SelectMAP configuration peripheral interface (see Slave Parallel Mode) is available after configuration. To continue using SelectMAP mode, set the ***Persist*** bitstream generator option to ***Yes***. An external host can then read and verify configuration data.

The Persist option will maintain A20-A23 as configuration pins although they are not used in SelectMAP mode.

Table 59: **Byte-Wide Peripheral Interface (BPI) Connections**

Pin Name	FPGA Direction	Description	During Configuration	After Configuration
HSWAP (P)	Input	**User I/O Pull-Up Control**. When Low during configuration, enables pull-up resistors in all I/O pins to respective I/O bank V_{CCO} input. 0: Pull-ups during configuration 1: No pull-ups	Drive at valid logic level throughout configuration.	User I/O
M[2:0] (A)	Input	**Mode Select**. Selects the FPGA configuration mode. See Design Considerations for the HSWAP, M[2:0], and VS[2:0] Pins.	M2 = 0, M1 = 1. Set M0 = 0 to start at address 0, increment addresses. Set M0 = 1 to start at address 0xFFFFFF and decrement addresses. Sampled when INIT_B goes High.	User I/O
CSI_B	Input	**Chip Select Input**. Active Low.	Must be Low throughout configuration.	User I/O. If bitstream option ***Persist=Yes***, becomes part of SelectMap parallel peripheral interface.
RDWR_B	Input	**Read/Write Control**. Active Low write enable. Read functionality typically only used after configuration, if bitstream option ***Persist=Yes***.	Must be Low throughout configuration.	User I/O. If bitstream option ***Persist=Yes***, becomes part of SelectMap parallel peripheral interface.
LDC0	Output	PROM Chip Enable	Connect to PROM chip-select input (CE#). FPGA drives this signal Low throughout configuration.	User I/O. If the FPGA does not access the PROM after configuration, drive this pin High to deselect the PROM. A[23:0], D[7:0], LDC[2:1], and HDC then become available as user I/O.
LDC1	Output	PROM Output Enable	Connect to the PROM output-enable input (OE#). The FPGA drives this signal Low throughout configuration.	User I/O

Table 65: **Slave Parallel Mode Connections *(Cont'd)***

Pin Name	FPGA Direction	Description	During Configuration	After Configuration
INIT_B	Open-drain bidirectional I/O	**Initialization Indicator**. Active Low. Goes Low at the start of configuration during the Initialization memory clearing process. Released at the end of memory clearing, when mode select pins are sampled. In daisy-chain applications, this signal requires an external 4.7 kΩ pull-up resistor to VCCO_2.	Active during configuration. If CRC error detected during configuration, FPGA drives INIT_B Low.	User I/O. If unused in the application, drive INIT_B High.
DONE	Open-drain bidirectional I/O	**FPGA Configuration Done**. Low during configuration. Goes High when FPGA successfully completes configuration. Requires external 330 Ω pull-up resistor to 2.5V.	Low indicates that the FPGA is not yet configured.	Pulled High via external pull-up. When High, indicates that the FPGA successfully configured.
PROG_B	Input	**Program FPGA**. Active Low. When asserted Low for 500 ns or longer, forces the FPGA to restart its configuration process by clearing configuration memory and resetting the DONE and INIT_B pins once PROG_B returns High. Recommend external 4.7 kΩ pull-up resistor to 2.5V. Internal pull-up value may be weaker (see Table 78). If driving externally with a 3.3V output, use an open-drain or open-collector driver or use a current limiting series resistor.	Must be High to allow configuration to start.	Drive PROG_B Low and release to reprogram FPGA.

Voltage Compatibility

(V) Most Slave Parallel interface signals are within the FPGA's I/O Bank 2, supplied by the VCCO_2 supply input. The VCCO_2 voltage can be 1.8V, 2.5V, or 3.3V to match the requirements of the external host, ideally 2.5V. Using 1.8V or 3.3V requires additional design considerations as the DONE and PROG_B pins are powered by the FPGA's 2.5V V_{CCAUX} supply. See XAPP453: *The 3.3V Configuration of Spartan-3 FPGAs* for additional information.

Daisy-Chaining

If the application requires multiple FPGAs with different configurations, then configure the FPGAs using a daisy chain. Use Slave Parallel mode (M[2:0] = <1:1:0>) for all FPGAs in the daisy-chain. The schematic in Figure 62 is optimized for FPGA downloading and does not support the SelectMAP read interface. The FPGA's RDWR_B pin must be Low during configuration.

After the lead FPGA is filled with its configuration data, the lead FPGA enables the next FPGA in the daisy-chain by asserting is chip-select output, CSO_B.

Table 93: **Timing for the IOB Three-State Path**

Symbol	Description	Conditions	Device	Speed Grade -5 Max	Speed Grade -4 Max	Units
Synchronous Output Enable/Disable Times						
T_{IOCKHZ}	Time from the active transition at the OTCLK input of the Three-state Flip-Flop (TFF) to when the Output pin enters the high-impedance state	LVCMOS25, 12 mA output drive, Fast slew rate	All	1.49	1.71	ns
T_{IOCKON}[(2)]	Time from the active transition at TFF's OTCLK input to when the Output pin drives valid data		All	2.70	3.10	ns
Asynchronous Output Enable/Disable Times						
T_{GTS}	Time from asserting the Global Three State (GTS) input on the STARTUP_SPARTAN3E primitive to when the Output pin enters the high-impedance state	LVCMOS25, 12 mA output drive, Fast slew rate	All	8.52	9.79	ns
Set/Reset Times						
T_{IOSRHZ}	Time from asserting TFF's SR input to when the Output pin enters a high-impedance state	LVCMOS25, 12 mA output drive, Fast slew rate	All	2.11	2.43	ns
T_{IOSRON}[(2)]	Time from asserting TFF's SR input at TFF to when the Output pin drives valid data		All	3.32	3.82	ns

Notes:

1. The numbers in this table are tested using the methodology presented in Table 95 and are based on the operating conditions set forth in Table 77 and Table 80.
2. This time requires adjustment whenever a signal standard other than LVCMOS25 with 12 mA drive and Fast slew rate is assigned to the data Output. When this is true, *add* the appropriate Output adjustment from Table 94.
3. For minimum delays use the values reported by the Timing Analyzer.

Table 95: **Test Methods for Timing Measurement at I/Os *(Cont'd)***

Signal Standard (IOSTANDARD)	Inputs			Outputs		Inputs and Outputs
	V_{REF} (V)	V_L (V)	V_H (V)	R_T (Ω)	V_T (V)	V_M (V)
DIFF_HSTL_I_18	-	$V_{REF} - 0.5$	$V_{REF} + 0.5$	50	0.9	V_{ICM}
DIFF_HSTL_III_18	-	$V_{REF} - 0.5$	$V_{REF} + 0.5$	50	1.8	V_{ICM}
DIFF_SSTL18_I	-	$V_{REF} - 0.5$	$V_{REF} + 0.5$	50	0.9	V_{ICM}
DIFF_SSTL2_I	-	$V_{REF} - 0.5$	$V_{REF} + 0.5$	50	1.25	V_{ICM}

Notes:

1. Descriptions of the relevant symbols are as follows:
   V_{REF} – The reference voltage for setting the input switching threshold
   V_{ICM} – The common mode input voltage
   V_M – Voltage of measurement point on signal transition
   V_L – Low-level test voltage at Input pin
   V_H – High-level test voltage at Input pin
   R_T – Effective termination resistance, which takes on a value of 1MΩ when no parallel termination is required
   V_T – Termination voltage
2. The load capacitance (C_L) at the Output pin is 0 pF for all signal standards.
3. According to the PCI specification.

The capacitive load (C_L) is connected between the output and GND. *The Output timing for all standards, as published in the speed files and the data sheet, is always based on a C_L value of zero.* High-impedance probes (less than 1 pF) are used for all measurements. Any delay that the test fixture might contribute to test measurements is subtracted from those measurements to produce the final timing numbers as published in the speed files and data sheet.

Using IBIS Models to Simulate Load Conditions in Application

IBIS models permit the most accurate prediction of timing delays for a given application. The parameters found in the IBIS model (V_{REF}, R_{REF}, and V_{MEAS}) correspond directly with the parameters used in Table 95 (V_T, R_T, and V_M). Do not confuse V_{REF} (the termination voltage) from the IBIS model with V_{REF} (the input-switching threshold) from the table. A fourth parameter, C_{REF}, is always zero. The four parameters describe all relevant output test conditions. IBIS models are found in the Xilinx development software as well as at the following link:

http://www.xilinx.com/support/download/index.htm

Delays for a given application are simulated according to its specific load conditions as follows:

1. Simulate the desired signal standard with the output driver connected to the test setup shown in Figure 72. Use parameter values V_T, R_T, and V_M from Table 95. C_{REF} is zero.
2. Record the time to V_M.
3. Simulate the same signal standard with the output driver connected to the PCB trace with load. Use the appropriate IBIS model (including V_{REF}, R_{REF}, C_{REF}, and V_{MEAS} values) or capacitive value to represent the load.
4. Record the time to V_{MEAS}.
5. Compare the results of steps 2 and 4. Add (or subtract) the increase (or decrease) in delay to (or from) the appropriate Output standard adjustment (Table 94) to yield the worst-case delay of the PCB trace.

Simultaneously Switching Output Guidelines

This section provides guidelines for the recommended maximum allowable number of Simultaneous Switching Outputs (SSOs). These guidelines describe the maximum number of user I/O pins of a given output signal standard that should simultaneously switch in the same direction, while maintaining a safe level of switching noise. Meeting these guidelines for the stated test conditions ensures that the FPGA operates free from the adverse effects of ground and power bounce.

Ground or power bounce occurs when a large number of outputs simultaneously switch in the same direction. The output drive transistors all conduct current to a common voltage rail. Low-to-High transitions conduct to the V_{CCO} rail; High-to-Low transitions conduct to the GND rail. The resulting cumulative current transient induces a voltage difference across the inductance that exists between the die pad and the power supply or ground return. The inductance is associated with bonding wires, the package lead frame, and any other signal routing inside the package. Other variables contribute to SSO noise levels, including stray inductance on the PCB as well as capacitive loading at receivers. Any SSO-induced voltage consequently affects internal switching noise margins and ultimately signal quality.

Table 96 and Table 97 provide the essential SSO guidelines. For each device/package combination, Table 96 provides the number of equivalent V_{CCO}/GND pairs. The equivalent number of pairs is based on characterization and might not match the physical number of pairs. For each output signal standard and drive strength, Table 97 recommends the maximum number of SSOs, switching in the same direction, allowed per V_{CCO}/GND pair within an I/O bank. The guidelines in Table 97 are categorized by package style. Multiply the appropriate numbers from Table 96 and Table 97 to calculate the maximum number of SSOs allowed within an I/O bank. Exceeding these SSO guidelines might result in increased power or ground bounce, degraded signal integrity, or increased system jitter.

SSO_{MAX}/IO Bank = Table 96 x Table 97

The recommended maximum SSO values assumes that the FPGA is soldered on the printed circuit board and that the board uses sound design practices. The SSO values do not apply for FPGAs mounted in sockets, due to the lead inductance introduced by the socket.

The number of SSOs allowed for quad-flat packages (VQ, TQ, PQ) is lower than for ball grid array packages (FG) due to the larger lead inductance of the quad-flat packages. The results for chip-scale packaging (CP132) are better than quad-flat packaging but not as high as for ball grid array packaging. Ball grid array packages are recommended for applications with a large number of simultaneously switching outputs.

Table 96: **Equivalent V_{CCO}/GND Pairs per Bank**

Device	Package Style *(including Pb-free)*							
	VQ100	**CP132**	**TQ144**	**PQ208**	**FT256**	**FG320**	**FG400**	**FG484**
XC3S100E	2	2	2	-	-	-	-	-
XC3S250E	2	2	2	3	4	-	-	-
XC3S500E	2	2	-	3	4	5	-	-
XC3S1200E	-	-	-	-	4	5	6	-
XC3S1600E	-	-	-	-	-	5	6	7

Table 117: **Timing for the Slave Parallel Configuration Mode *(Cont'd)***

Symbol	Description			All Speed Grades Min	All Speed Grades Max	Units
Clock Timing						
T_{CCH}	The High pulse width at the CCLK input pin			5	-	ns
T_{CCL}	The Low pulse width at the CCLK input pin			5	-	ns
F_{CCPAR}	Frequency of the clock signal at the CCLK input pin	No bitstream compression	Not using the BUSY pin(2)	0	50	MHz
			Using the BUSY pin	0	66	MHz
		With bitstream compression		0	20	MHz

Notes:

1. The numbers in this table are based on the operating conditions set forth in Table 77.
2. In the Slave Parallel mode, it is necessary to use the BUSY pin when the CCLK frequency exceeds this maximum specification.
3. Some Xilinx documents refer to Parallel modes as "SelectMAP" modes.

Table 119: **Configuration Timing Requirements for Attached SPI Serial Flash**

Symbol	Description	Requirement	Units
T_{CCS}	SPI serial Flash PROM chip-select time	$T_{CCS} \leq T_{MCCL1} - T_{CCO}$	ns
T_{DSU}	SPI serial Flash PROM data input setup time	$T_{DSU} \leq T_{MCCL1} - T_{CCO}$	ns
T_{DH}	SPI serial Flash PROM data input hold time	$T_{DH} \leq T_{MCCH1}$	ns
T_V	SPI serial Flash PROM data clock-to-output time	$T_V \leq T_{MCCLn} - T_{DCC}$	ns
f_C or f_R	Maximum SPI serial Flash PROM clock frequency (also depends on specific read command used)	$f_C \geq \frac{1}{T_{CCLKn(min)}}$	MHz

Notes:

1. These requirements are for successful FPGA configuration in SPI mode, where the FPGA provides the CCLK frequency. The post configuration timing can be different to support the specific needs of the application loaded into the FPGA and the resulting clock source.
2. Subtract additional printed circuit board routing delay as required by the application.

CP132 Footprint

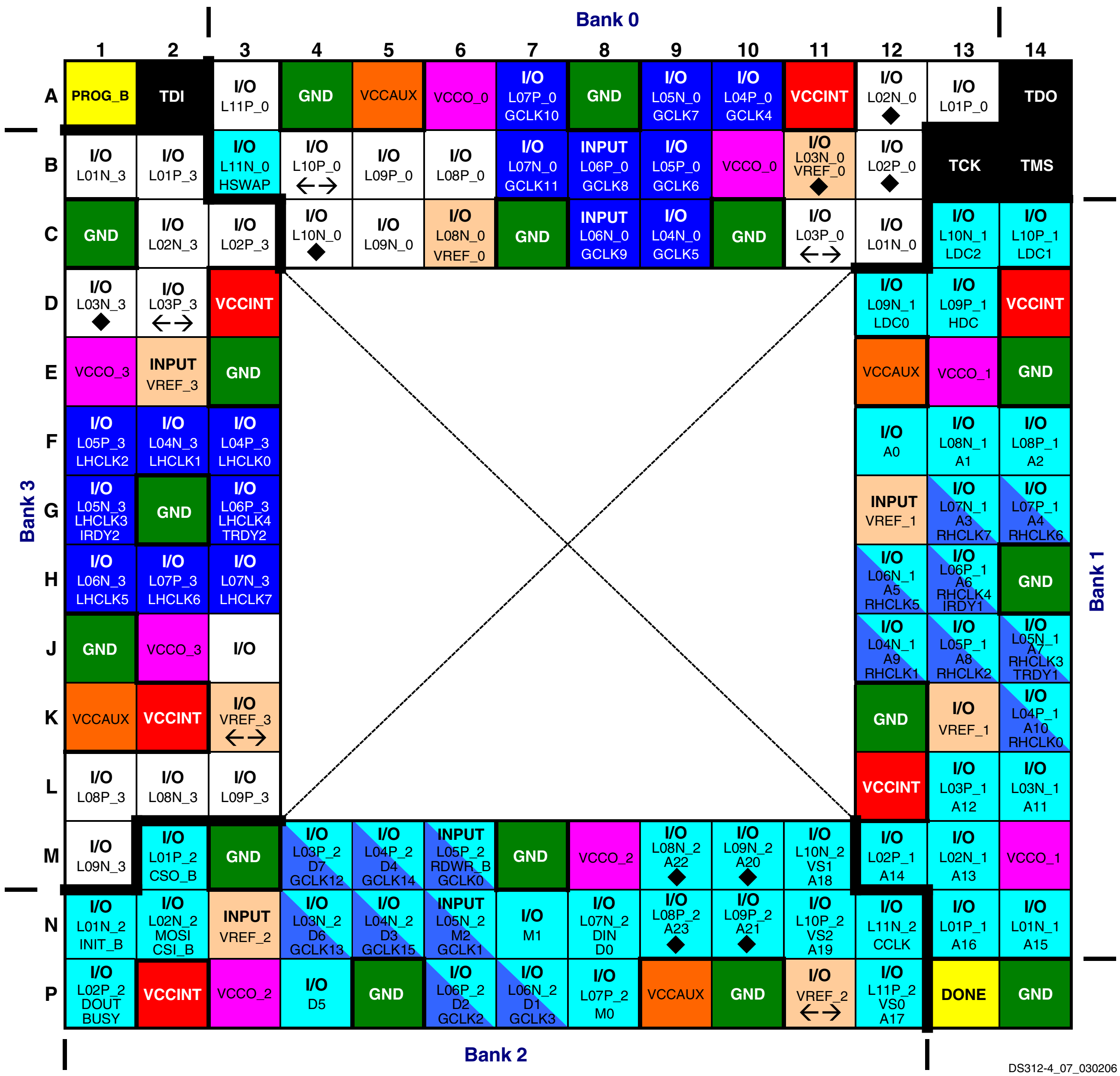


Figure 81: **CP132 Package Footprint (top view)**

16-22	**I/O:** Unrestricted, general-purpose user I/O	42-46	**DUAL:** Configuration pin, then possible user I/O	7-8	**VREF:** User I/O or input voltage reference for bank
0-2	**INPUT:** Unrestricted, general-purpose input pin	16	**CLK:** User I/O, input, or global buffer input	8	**VCCO:** Output voltage supply for bank
2	**CONFIG:** Dedicated configuration pins	4	**JTAG:** Dedicated JTAG port pins	6	**VCCINT:** Internal core supply voltage (+1.2V)
9	**N.C.:** Unconnected balls on the XC3S100E FPGA (◆)	16	**GND:** Ground	4	**VCCAUX:** Auxiliary supply voltage (+2.5V)

PQ208: 208-pin Plastic Quad Flat Package

The 208-pin plastic quad flat package, PQ208, supports two different Spartan-3E FPGAs, including the XC3S250E and the XC3S500E.

Table 141 lists all the PQ208 package pins. They are sorted by bank number and then by pin name. Pairs of pins that form a differential I/O pair appear together in the table. The table also shows the pin number for each pin and the pin type, as defined earlier.

An electronic version of this package pinout table and footprint diagram is available for download from the Xilinx website at:

http://www.xilinx.com/support/documentation/data_sheets/s3e_pin.zip

Pinout Table

Table 141: **PQ208 Package Pinout**

Bank	XC3S250E XC3S500E Pin Name	PQ208 Pin	Type
0	IO	P187	I/O
0	IO/VREF_0	P179	VREF
0	IO_L01N_0	P161	I/O
0	IO_L01P_0	P160	I/O
0	IO_L02N_0/VREF_0	P163	VREF
0	IO_L02P_0	P162	I/O
0	IO_L03N_0	P165	I/O
0	IO_L03P_0	P164	I/O
0	IO_L04N_0/VREF_0	P168	VREF
0	IO_L04P_0	P167	I/O
0	IO_L05N_0	P172	I/O
0	IO_L05P_0	P171	I/O
0	IO_L07N_0/GCLK5	P178	GCLK
0	IO_L07P_0/GCLK4	P177	GCLK
0	IO_L08N_0/GCLK7	P181	GCLK
0	IO_L08P_0/GCLK6	P180	GCLK
0	IO_L10N_0/GCLK11	P186	GCLK
0	IO_L10P_0/GCLK10	P185	GCLK
0	IO_L11N_0	P190	I/O
0	IO_L11P_0	P189	I/O
0	IO_L12N_0/VREF_0	P193	VREF
0	IO_L12P_0	P192	I/O
0	IO_L13N_0	P197	I/O
0	IO_L13P_0	P196	I/O
0	IO_L14N_0/VREF_0	P200	VREF
0	IO_L14P_0	P199	I/O
0	IO_L15N_0	P203	I/O

Table 141: **PQ208 Package Pinout *(Cont'd)***

Bank	XC3S250E XC3S500E Pin Name	PQ208 Pin	Type
0	IO_L15P_0	P202	I/O
0	IO_L16N_0/HSWAP	P206	DUAL
0	IO_L16P_0	P205	I/O
0	IP	P159	INPUT
0	IP	P169	INPUT
0	IP	P194	INPUT
0	IP	P204	INPUT
0	IP_L06N_0	P175	INPUT
0	IP_L06P_0	P174	INPUT
0	IP_L09N_0/GCLK9	P184	GCLK
0	IP_L09P_0/GCLK8	P183	GCLK
0	VCCO_0	P176	VCCO
0	VCCO_0	P191	VCCO
0	VCCO_0	P201	VCCO
1	IO_L01N_1/A15	P107	DUAL
1	IO_L01P_1/A16	P106	DUAL
1	IO_L02N_1/A13	P109	DUAL
1	IO_L02P_1/A14	P108	DUAL
1	IO_L03N_1/VREF_1	P113	VREF
1	IO_L03P_1	P112	I/O
1	IO_L04N_1	P116	I/O
1	IO_L04P_1	P115	I/O
1	IO_L05N_1/A11	P120	DUAL
1	IO_L05P_1/A12	P119	DUAL
1	IO_L06N_1/VREF_1	P123	VREF
1	IO_L06P_1	P122	I/O
1	IO_L07N_1/A9/RHCLK1	P127	RHCLK/DUAL
1	IO_L07P_1/A10/RHCLK0	P126	RHCLK/DUAL
1	IO_L08N_1/A7/RHCLK3	P129	RHCLK/DUAL
1	IO_L08P_1/A8/RHCLK2	P128	RHCLK/DUAL
1	IO_L09N_1/A5/RHCLK5	P133	RHCLK/DUAL
1	IO_L09P_1/A6/RHCLK4	P132	RHCLK/DUAL
1	IO_L10N_1/A3/RHCLK7	P135	RHCLK/DUAL
1	IO_L10P_1/A4/RHCLK6	P134	RHCLK/DUAL
1	IO_L11N_1/A1	P138	DUAL
1	IO_L11P_1/A2	P137	DUAL
1	IO_L12N_1/A0	P140	DUAL
1	IO_L12P_1	P139	I/O
1	IO_L13N_1	P145	I/O
1	IO_L13P_1	P144	I/O
1	IO_L14N_1	P147	I/O
1	IO_L14P_1	P146	I/O

Table 143: **FT256 Package Pinout *(Cont'd)***

Bank	XC3S250E Pin Name	XC3S500E Pin Name	XC3S1200E Pin Name	FT256 Ball	Type
1	N.C. (◆)	IO_L05P_1	IO_L05P_1	L12	***250E:*** N.C. ***500E:*** I/O ***1200E:*** I/O
1	IO_L06N_1	IO_L06N_1	IO_L06N_1	L15	I/O
1	IO_L06P_1	IO_L06P_1	IO_L06P_1	L14	I/O
1	IO_L07N_1/A11	IO_L07N_1/A11	IO_L07N_1/A11	K12	DUAL
1	IO_L07P_1/A12	IO_L07P_1/A12	IO_L07P_1/A12	K13	DUAL
1	IO_L08N_1/VREF_1	IO_L08N_1/VREF_1	IO_L08N_1/VREF_1	K14	VREF
1	IO_L08P_1	IO_L08P_1	IO_L08P_1	K15	I/O
1	IO_L09N_1/A9/RHCLK1	IO_L09N_1/A9/RHCLK1	IO_L09N_1/A9/RHCLK1	J16	RHCLK/DUAL
1	IO_L09P_1/A10/RHCLK0	IO_L09P_1/A10/RHCLK0	IO_L09P_1/A10/RHCLK0	K16	RHCLK/DUAL
1	IO_L10N_1/A7/RHCLK3/ TRDY1	IO_L10N_1/A7/RHCLK3/ TRDY1	IO_L10N_1/A7/RHCLK3/ TRDY1	J13	RHCLK/DUAL
1	IO_L10P_1/A8/RHCLK2	IO_L10P_1/A8/RHCLK2	IO_L10P_1/A8/RHCLK2	J14	RHCLK/DUAL
1	IO_L11N_1/A5/RHCLK5	IO_L11N_1/A5/RHCLK5	IO_L11N_1/A5/RHCLK5	H14	RHCLK/DUAL
1	IO_L11P_1/A6/RHCLK4/ IRDY1	IO_L11P_1/A6/RHCLK4/ IRDY1	IO_L11P_1/A6/RHCLK4/ IRDY1	H15	RHCLK/DUAL
1	IO_L12N_1/A3/RHCLK7	IO_L12N_1/A3/RHCLK7	IO_L12N_1/A3/RHCLK7	H11	RHCLK/DUAL
1	IO_L12P_1/A4/RHCLK6	IO_L12P_1/A4/RHCLK6	IO_L12P_1/A4/RHCLK6	H12	RHCLK/DUAL
1	IO_L13N_1/A1	IO_L13N_1/A1	IO_L13N_1/A1	G16	DUAL
1	IO_L13P_1/A2	IO_L13P_1/A2	IO_L13P_1/A2	G15	DUAL
1	IO_L14N_1/A0	IO_L14N_1/A0	IO_L14N_1/A0	G14	DUAL
1	IO_L14P_1	IO_L14P_1	IO_L14P_1	G13	I/O
1	IO_L15N_1	IO_L15N_1	IO_L15N_1	F15	I/O
1	IO_L15P_1	IO_L15P_1	IO_L15P_1	F14	I/O
1	IO_L16N_1	IO_L16N_1	IO_L16N_1	F12	I/O
1	IO_L16P_1	IO_L16P_1	IO_L16P_1	F13	I/O
1	N.C. (◆)	IO_L17N_1	IO_L17N_1	E16	***250E:*** N.C. ***500E:*** I/O ***1200E:*** I/O
1	N.C. (◆).	IO_L17P_1	IO_L17P_1	E13	***250E:*** N.C. ***500E:*** I/O ***1200E:*** I/O
1	IO_L18N_1/LDC0	IO_L18N_1/LDC0	IO_L18N_1/LDC0	D14	DUAL
1	IO_L18P_1/HDC	IO_L18P_1/HDC	IO_L18P_1/HDC	D15	DUAL
1	IO_L19N_1/LDC2	IO_L19N_1/LDC2	IO_L19N_1/LDC2	C15	DUAL
1	IO_L19P_1/LDC1	IO_L19P_1/LDC1	IO_L19P_1/LDC1	C16	DUAL
1	IP	IP	IP	B16	INPUT
1	IP	IP	IP	E14	INPUT
1	IP	IP	IP	G12	INPUT
1	IP	IP	IP	H16	INPUT
1	IP	IP	IP	J11	INPUT
1	IP	IP	IP	J12	INPUT
1	IP	IP	IP	M13	INPUT

Footprint Migration Differences

Table 151 summarizes any footprint and functionality differences between the XC3S500E, the XC3S1200E, and the XC3S1600E FPGAs that may affect easy migration between devices available in the FG320 package. There are 26 such balls. All other pins not listed in Table 151 unconditionally migrate between Spartan-3E devices available in the FG320 package.

The XC3S500E is duplicated on both the left and right sides of the table to show migrations to and from the XC3S1200E and the XC3S1600E. The arrows indicate the direction for easy migration. A double-ended arrow (←→) indicates that the two pins have identical functionality. A left-facing arrow (←) indicates that the pin on the device on the right unconditionally migrates to the pin on the device on the left. It may be possible to migrate the opposite direction depending on the I/O configuration. For example, an I/O pin (Type = I/O) can migrate to an input-only pin (Type = INPUT) if the I/O pin is configured as an input.

Table 151: **FG320 Footprint Migration Differences**

Pin	Bank	XC3S500E	Migration	XC3S1200E	Migration	XC3S1600E	Migration	XC3S500E
A7	0	INPUT	→	I/O	←→	I/O	←	INPUT
A12	0	N.C.	→	I/O	←→	I/O	←	N.C.
D4	3	N.C.	→	I/O	←→	I/O	←	N.C.
D6	0	N.C.	→	I/O	←→	I/O	←	N.C.
D13	0	INPUT	→	I/O	←→	I/O	←	INPUT
E3	3	N.C.	→	I/O	←→	I/O	←	N.C.
E4	3	N.C.	→	I/O	←→	I/O	←	N.C.
E6	0	N.C.	→	I/O	←→	I/O	←	N.C.
E15	1	N.C.	→	I/O	←→	I/O	←	N.C.
E16	1	N.C.	→	I/O	←→	I/O	←	N.C.
E17	1	I/O	←	INPUT	←→	INPUT	→	I/O
F4	3	I/O	←	INPUT	←→	INPUT	→	I/O
N12	2	N.C.	→	I/O	←→	I/O	←	N.C.
N14	1	N.C.	→	I/O	←→	I/O	←	N.C.
N15	1	N.C.	→	I/O	←→	I/O	←	N.C.
P3	3	N.C.	→	I/O	←→	I/O	←	N.C.
P4	3	N.C.	→	I/O	←→	I/O	←	N.C.
P12	2	N.C.	→	I/O	←→	I/O	←	N.C.
P15	1	I/O	←	INPUT	←→	INPUT	→	I/O
P16	1	N.C.	→	I/O	←→	I/O	←	N.C.
R4	3	VREF(I/O)	←	VREF(INPUT)	←→	VREF(INPUT)	→	VREF(I/O)
U6	2	INPUT	→	I/O	←→	I/O	←	INPUT
U13	2	INPUT	→	I/O	←→	I/O	←	INPUT
V5	2	N.C.	→	I/O	←→	I/O	←	N.C.
V6	2	N.C.	→	VREF	←→	VREF	←	N.C.
V7	2	N.C.	→	I/O	←→	I/O	←	N.C.
DIFFERENCES			**26**		**0**		**26**	

Legend:

←→ This pin is identical on the device on the left and the right.

→ This pin can unconditionally migrate from the device on the left to the device on the right. Migration in the other direction may be possible depending on how the pin is configured for the device on the right.

← This pin can unconditionally migrate from the device on the right to the device on the left. Migration in the other direction may be possible depending on how the pin is configured for the device on the left.

Table 152: **FG400 Package Pinout *(Cont'd)***

Bank	XC3S1200E XC3S1600E Pin Name	FG400 Ball	Type
2	IP_L14P_2	T9	INPUT
2	IP_L17N_2/M2/GCLK1	P12	DUAL/ GCLK
2	IP_L17P_2/RDWR_B/ GCLK0	P11	DUAL/ GCLK
2	IP_L20N_2	T12	INPUT
2	IP_L20P_2	R12	INPUT
2	IP_L23N_2/VREF_2	T13	VREF
2	IP_L23P_2	T14	INPUT
2	IP_L26N_2	V14	INPUT
2	IP_L26P_2	V15	INPUT
2	IP_L29N_2	W16	INPUT
2	IP_L29P_2	Y16	INPUT
2	VCCO_2	R11	VCCO
2	VCCO_2	U8	VCCO
2	VCCO_2	U14	VCCO
2	VCCO_2	W5	VCCO
2	VCCO_2	W11	VCCO
2	VCCO_2	W17	VCCO
3	IO_L01N_3	D2	I/O
3	IO_L01P_3	D3	I/O
3	IO_L02N_3/VREF_3	E3	VREF
3	IO_L02P_3	E4	I/O
3	IO_L03N_3	C1	I/O
3	IO_L03P_3	B1	I/O
3	IO_L04N_3	E1	I/O
3	IO_L04P_3	D1	I/O
3	IO_L05N_3	F3	I/O
3	IO_L05P_3	F4	I/O
3	IO_L06N_3	F1	I/O
3	IO_L06P_3	F2	I/O
3	IO_L07N_3	G4	I/O
3	IO_L07P_3	G3	I/O
3	IO_L08N_3	G5	I/O
3	IO_L08P_3	H5	I/O
3	IO_L09N_3/VREF_3	H3	VREF
3	IO_L09P_3	H2	I/O
3	IO_L10N_3	H7	I/O
3	IO_L10P_3	H6	I/O
3	IO_L11N_3	J4	I/O
3	IO_L11P_3	J3	I/O
3	IO_L12N_3	J1	I/O
3	IO_L12P_3	J2	I/O
3	IO_L13N_3	J6	I/O

Table 152: **FG400 Package Pinout *(Cont'd)***

Bank	XC3S1200E XC3S1600E Pin Name	FG400 Ball	Type
3	IO_L13P_3	K6	I/O
3	IO_L14N_3/LHCLK1	K2	LHCLK
3	IO_L14P_3/LHCLK0	K3	LHCLK
3	IO_L15N_3/LHCLK3/IRDY2	L7	LHCLK
3	IO_L15P_3/LHCLK2	K7	LHCLK
3	IO_L16N_3/LHCLK5	L1	LHCLK
3	IO_L16P_3/LHCLK4/TRDY2	M1	LHCLK
3	IO_L17N_3/LHCLK7	L3	LHCLK
3	IO_L17P_3/LHCLK6	M3	LHCLK
3	IO_L18N_3	M7	I/O
3	IO_L18P_3	M8	I/O
3	IO_L19N_3	M4	I/O
3	IO_L19P_3	M5	I/O
3	IO_L20N_3/VREF_3	N6	VREF
3	IO_L20P_3	M6	I/O
3	IO_L21N_3	N2	I/O
3	IO_L21P_3	N1	I/O
3	IO_L22N_3	P7	I/O
3	IO_L22P_3	N7	I/O
3	IO_L23N_3	N4	I/O
3	IO_L23P_3	N3	I/O
3	IO_L24N_3	R1	I/O
3	IO_L24P_3	P1	I/O
3	IO_L25N_3	R5	I/O
3	IO_L25P_3	P5	I/O
3	IO_L26N_3	T2	I/O
3	IO_L26P_3	R2	I/O
3	IO_L27N_3	R4	I/O
3	IO_L27P_3	R3	I/O
3	IO_L28N_3/VREF_3	T1	VREF
3	IO_L28P_3	U1	I/O
3	IO_L29N_3	T3	I/O
3	IO_L29P_3	U3	I/O
3	IO_L30N_3	V1	I/O
3	IO_L30P_3	V2	I/O
3	IP	F5	INPUT
3	IP	G1	INPUT
3	IP	G6	INPUT
3	IP	H1	INPUT
3	IP	J5	INPUT
3	IP	L5	INPUT
3	IP	L8	INPUT
3	IP	M2	INPUT