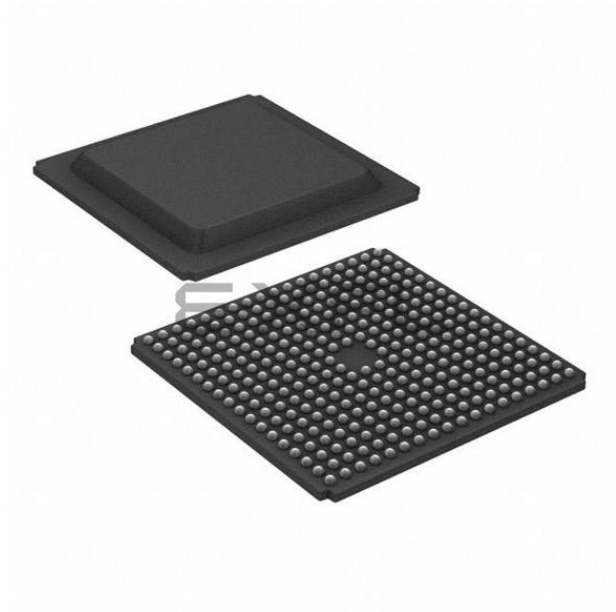
AMD Xilinx - XC3S500E-4FG320C Datasheet



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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	1164
Number of Logic Elements/Cells	10476
Total RAM Bits	368640
Number of I/O	232
Number of Gates	500000
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	320-BGA
Supplier Device Package	320-FBGA (19x19)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xc3s500e-4fg320c

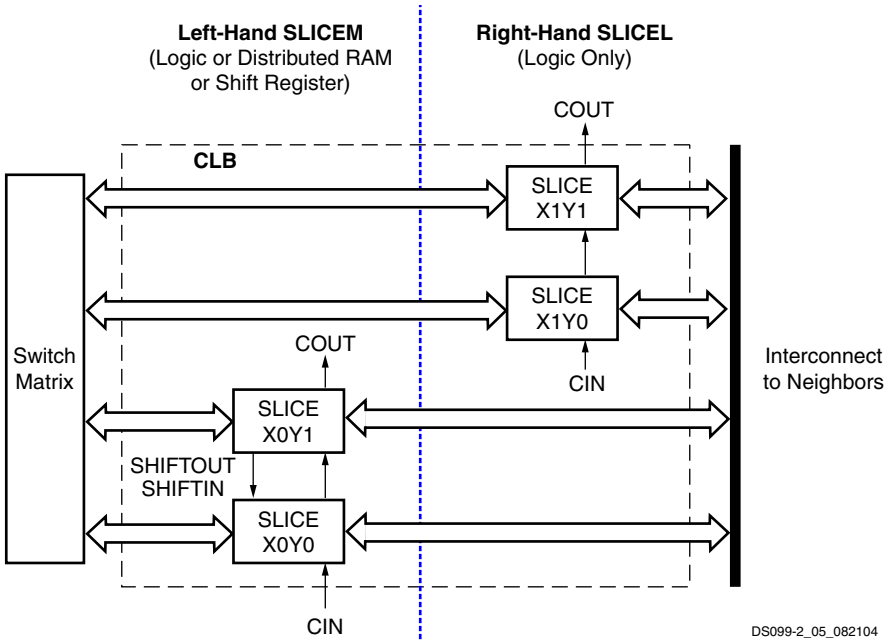


Figure 16: **Arrangement of Slices within the CLB**

Slice Location Designations

The Xilinx development software designates the location of a slice according to its X and Y coordinates, starting in the bottom left corner, as shown in Figure 14. The letter 'X' followed by a number identifies columns of slices, incrementing from the left side of the die to the right. The letter 'Y' followed by a number identifies the position of each slice in a pair as well as indicating the CLB row, incrementing from the bottom of the die. Figure 16 shows the CLB located in the lower left-hand corner of the die. The SLICEM always has an even 'X' number, and the SLICEL always has an odd 'X' number.

Slice Overview

A slice includes two LUT function generators and two storage elements, along with additional logic, as shown in Figure 17.

Both SLICEM and SLICEL have the following elements in common to provide logic, arithmetic, and ROM functions:

- Two 4-input LUT function generators, F and G
- Two storage elements
- Two wide-function multiplexers, F5MUX and FiMUX
- Carry and arithmetic logic

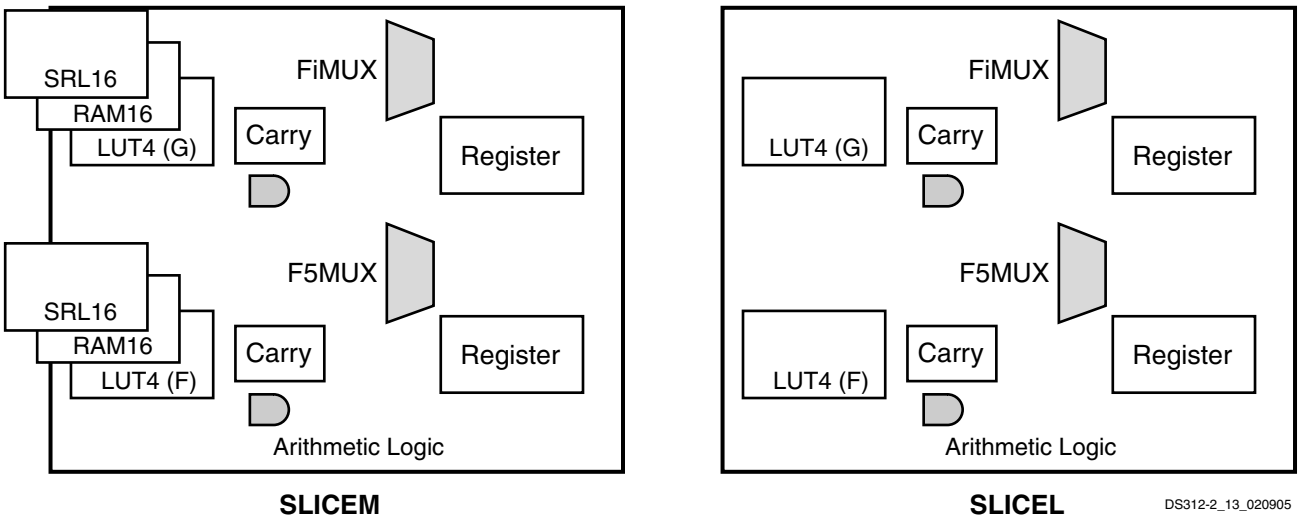


Figure 17: **Resources in a Slice**

Block RAM

For additional information, refer to the “Using Block RAM” chapter in UG331.

Spartan-3E devices incorporate 4 to 36 dedicated block RAMs, which are organized as dual-port configurable 18 Kbit blocks. Functionally, the block RAM is identical to the Spartan-3 architecture block RAM. Block RAM synchronously stores large amounts of data while distributed RAM, previously described, is better suited for buffering small amounts of data anywhere along signal paths. This section describes basic block RAM functions.

Each block RAM is configurable by setting the content’s initial values, default signal value of the output registers, port aspect ratios, and write modes. Block RAM can be used in single-port or dual-port modes.

Arrangement of RAM Blocks on Die

The block RAMs are located together with the multipliers on the die in one or two columns depending on the size of the device. The XC3S100E has one column of block RAM. The Spartan-3E devices ranging from the XC3S250E to XC3S1600E have two columns of block RAM. Table 21 shows the number of RAM blocks, the data storage capacity, and the number of columns for each device. Row(s) of CLBs are located above and below each block RAM column.

Table 21: **Number of RAM Blocks by Device**

Device	Total Number of RAM Blocks	Total Addressable Locations (bits)	Number of Columns
XC3S100E	4	73,728	1
XC3S250E	12	221,184	2
XC3S500E	20	368,640	2
XC3S1200E	28	516,096	2
XC3S1600E	36	663,552	2

Immediately adjacent to each block RAM is an embedded 18x18 hardware multiplier. The upper 16 bits of the block RAM's Port A Data input bus are shared with the upper 16 bits of the A multiplicand input bus of the multiplier. Similarly, the upper 16 bits of Port B's data input bus are shared with the B multiplicand input bus of the multiplier.

The Internal Structure of the Block RAM

The block RAM has a dual port structure. The two identical data ports called A and B permit independent access to the common block RAM, which has a maximum capacity of 18,432 bits, or 16,384 bits with no parity bits (see parity bits description in Table 22). Each port has its own dedicated set of data, control, and clock lines for synchronous read and write operations. There are four basic data paths, as shown in Figure 30:

1. Write to and read from Port A
2. Write to and read from Port B
3. Data transfer from Port A to Port B
4. Data transfer from Port B to Port A

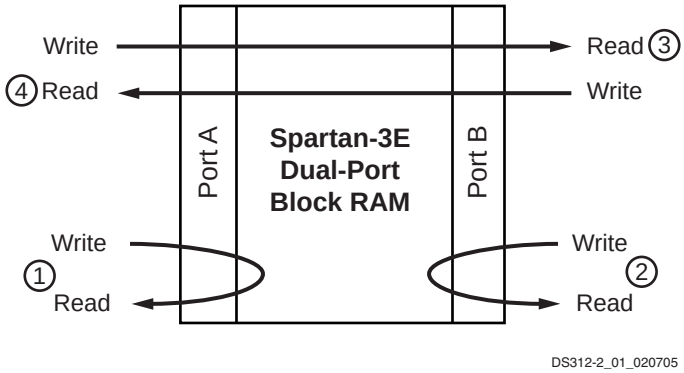


Figure 30: **Block RAM Data Paths**

Number of Ports

A choice among primitives determines whether the block RAM functions as dual- or single-port memory. A name of the form RAMB16_S[w_A]_S[w_B] calls out the dual-port primitive, where the integers w_A and w_B specify the total data path width at ports A and B, respectively. Thus, a RAMB16_S9_S18 is a dual-port RAM with a 9-bit Port A and an 18-bit Port B. A name of the form RAMB16_S[w] identifies the single-port primitive, where the integer w specifies the total data path width of the lone port A. A RAMB16_S18 is a single-port RAM with an 18-bit port.

Port Aspect Ratios

Each port of the block RAM can be configured independently to select a number of different possible widths for the data input (DI) and data output (DO) signals as shown in Table 22.

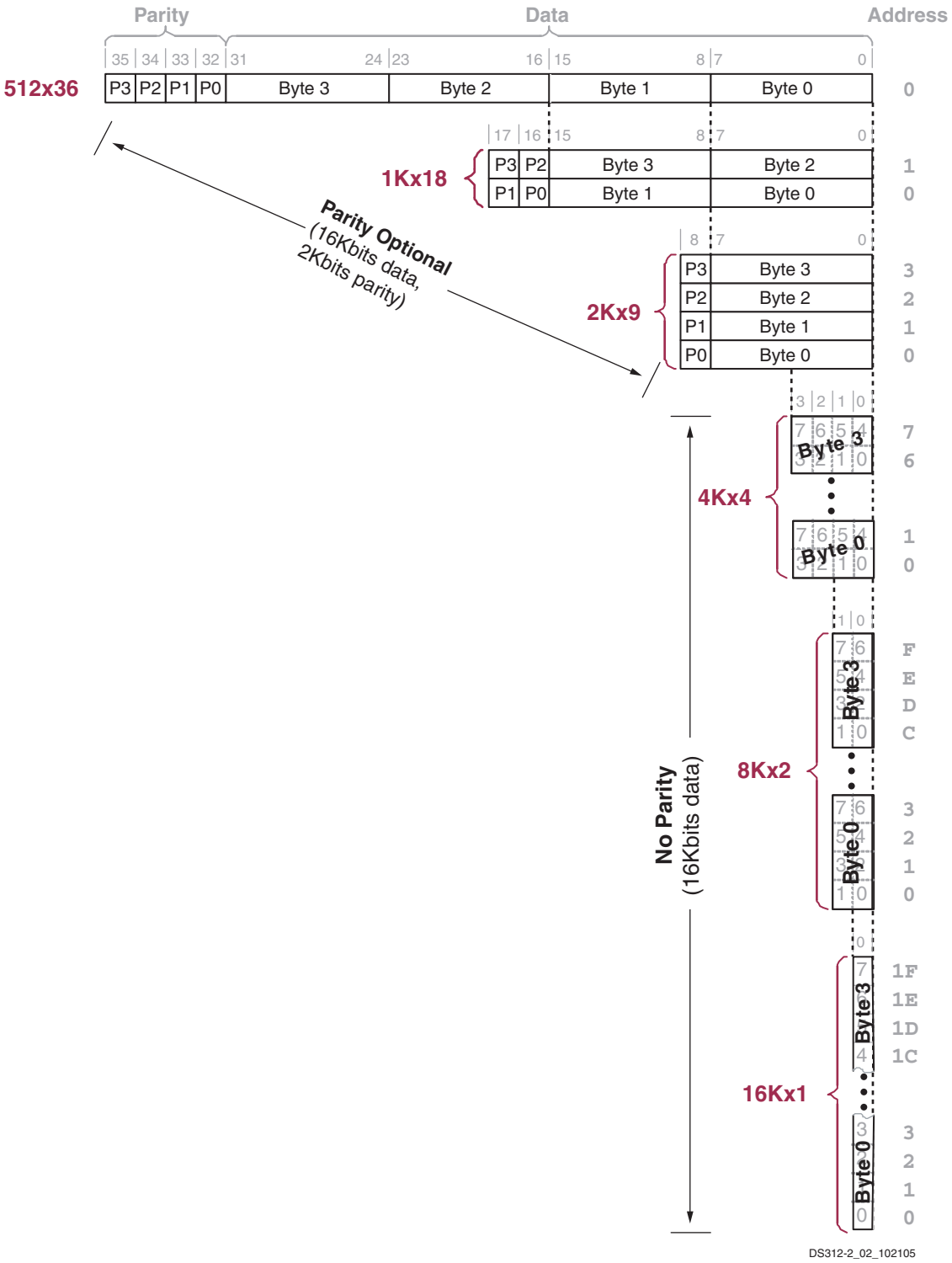


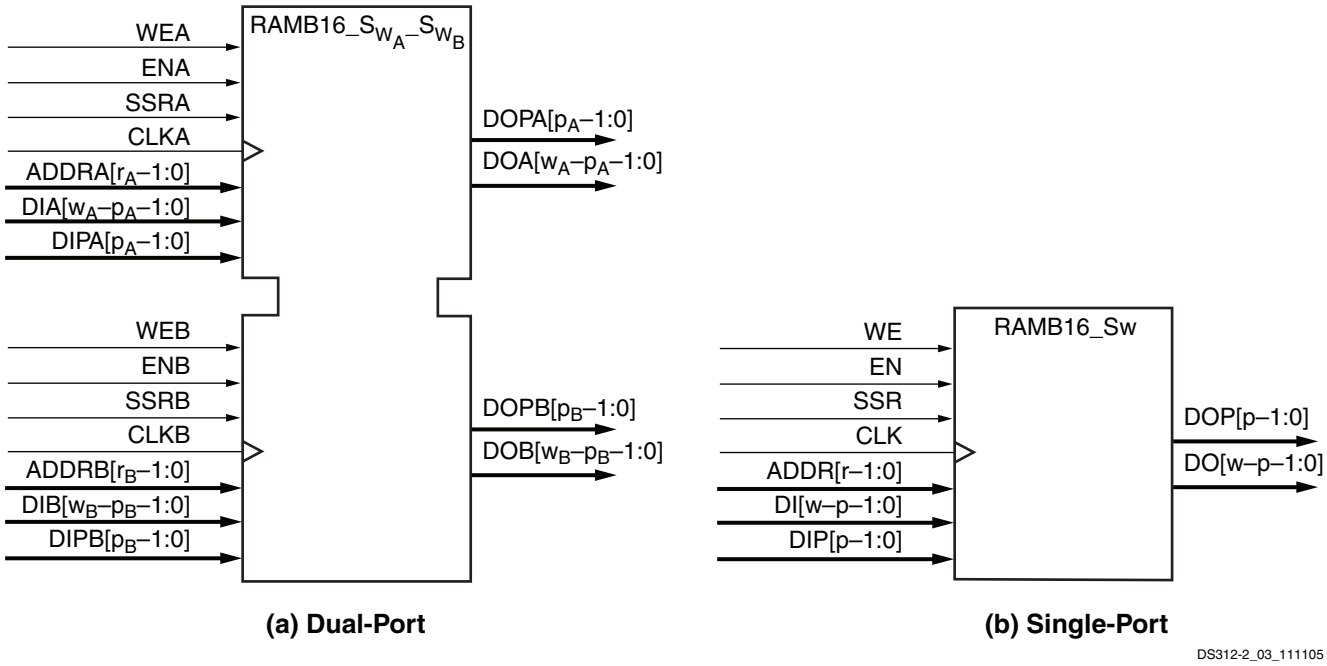
Figure 31: **Data Organization and Bus-matching Operation with Different Port Widths on Port A and Port B**

Block RAM Port Signal Definitions

Representations of the dual-port primitive RAMB16_S[w_A]_S[w_B] and the single-port primitive RAMB16_S[w] with their associated signals are shown in Figure 32a and Figure 32b, respectively. These signals are defined in Table 23. The control signals (WE, EN, CLK, and SSR) on the block RAM are active High. However, optional inverters on the control signals change the polarity of the active edge to active Low.

Design Note

Whenever a block RAM port is enabled (ENA or ENB = High), all address transitions must meet the data sheet setup and hold times with respect to the port clock (CLKA or CLKB), as shown in Table 103, page 138.This requirement must be met even if the RAM read output is of no interest.



Notes:

1. w_A and w_B are integers representing the total data path width (i.e., data bits plus parity bits) at Ports A and B, respectively.
2. p_A and p_B are integers that indicate the number of data path lines serving as parity bits.
3. r_A and r_B are integers representing the address bus width at ports A and B, respectively.
4. The control signals CLK, WE, EN, and SSR on both ports have the option of inverted polarity.

Figure 32: **Block RAM Primitives**

DLL Clock Output and Feedback Connections

As many as four of the nine DCM clock outputs can simultaneously drive four of the BUFGMUX buffers on the same die edge. All DCM clock outputs can simultaneously drive general routing resources, including interconnect leading to OBUF buffers.

The feedback loop is essential for DLL operation. Either the CLK0 or CLK2X outputs feed back to the CLKFB input via a BUFGMUX global buffer to eliminate the clock distribution delay. The specific BUFGMUX buffer used to feed back the CLK0 or CLK2X signal is ideally one of the BUFGMUX buffers associated with a specific DCM, as shown in Table 30, Table 31, and Table 32.

The feedback path also phase-aligns the other seven DLL outputs: CLK0, CLK90, CLK180, CLK270, CLKDV, CLK2X, or CLK2X180. The CLK_FEEDBACK attribute value must agree with the physical feedback connection. Use “1X” for CLK0 feedback and “2X” for CLK2X feedback. If the DFS unit is used stand-alone, without the DLL, then no feedback is required and set the CLK_FEEDBACK attribute to “NONE”.

Two basic cases determine how to connect the DLL clock outputs and feedback connections: on-chip synchronization and off-chip synchronization, which are illustrated in Figure 42a through Figure 42d.

In the on-chip synchronization case in Figure 42a and Figure 42b, it is possible to connect any of the DLL’s seven output clock signals through general routing resources to the FPGA’s internal registers. Either a Global Clock Buffer (BUFG) or a BUFGMUX affords access to the global clock network. As shown in Figure 42a, the feedback loop is created by routing CLK0 (or CLK2X) in Figure 42b to a global clock net, which in turn drives the CLKFB input.

In the off-chip synchronization case in Figure 42c and Figure 42d, CLK0 (or CLK2X) plus any of the DLL’s other output clock signals exit the FPGA using output buffers (OBUF) to drive an external clock network plus registers on the board. As shown in Figure 42c, the feedback loop is formed by feeding CLK0 (or CLK2X) in Figure 42d back into the FPGA, then to the DCM’s CLKFB input via a Global Buffer Input, specified in Table 30.

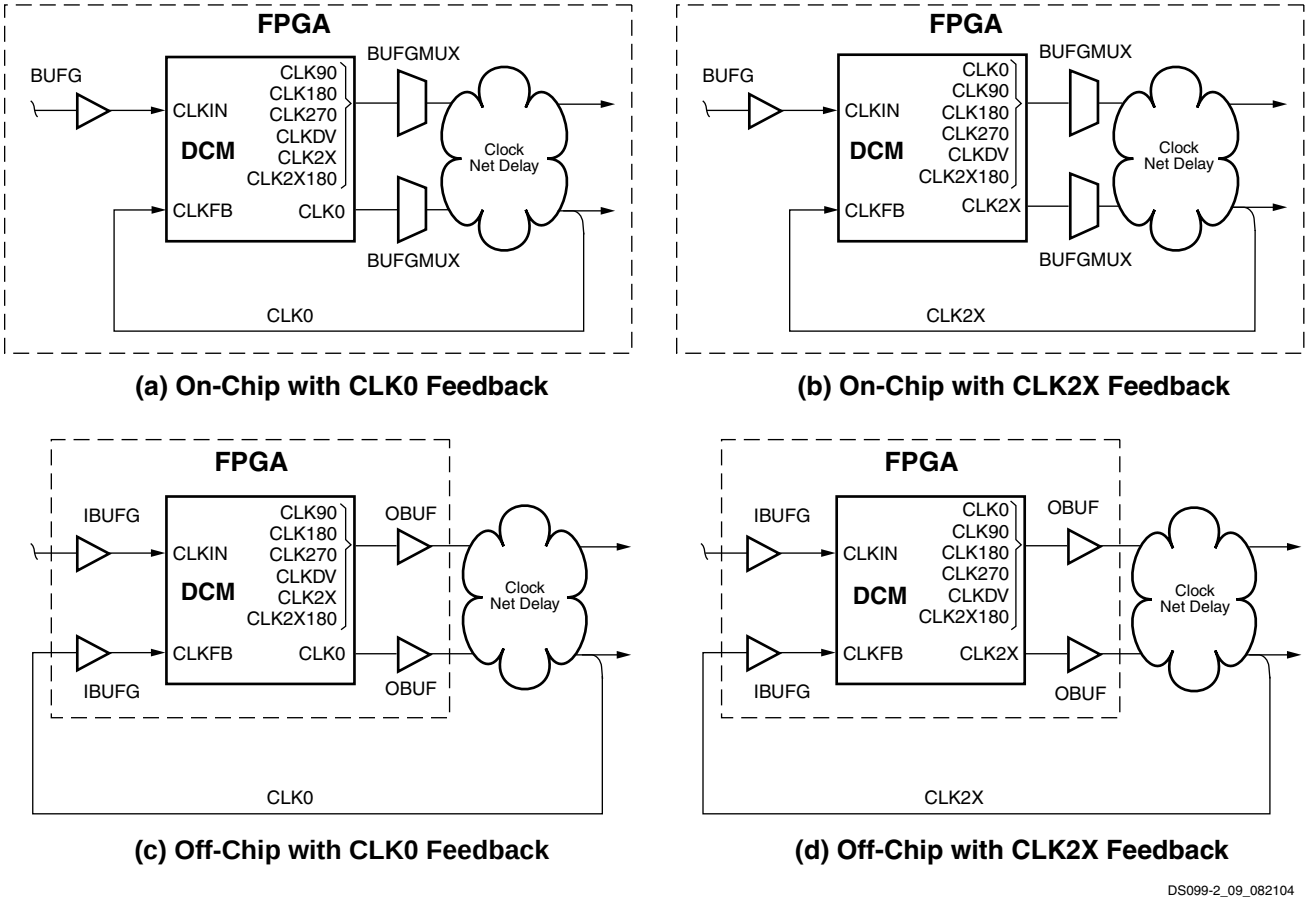


Figure 42: **Input Clock, Output Clock, and Feedback Connections for the DLL**

Accommodating Input Frequencies Beyond Specified Maximums

If the CLKIN input frequency exceeds the maximum permitted, divide it down to an acceptable value using the CLKIN_DIVIDE_BY_2 attribute. When this attribute is set to "TRUE", the CLKIN frequency is divided by a factor of two as it enters the DCM. In addition, the CLKIN_DIVIDE_BY_2 option produces a 50% duty-cycle on the input clock, although at half the CLKIN frequency.

Quadrant and Half-Period Phase Shift Outputs

In addition to CLK0 for zero-phase alignment to the CLKIN signal, the DLL also provides the CLK90, CLK180, and CLK270 outputs for 90°, 180°, and 270° phase-shifted signals, respectively. These signals are described in Table 28, page 47 and their relative timing is shown in Figure 43. For control in finer increments than 90°, see Phase Shifter (PS).

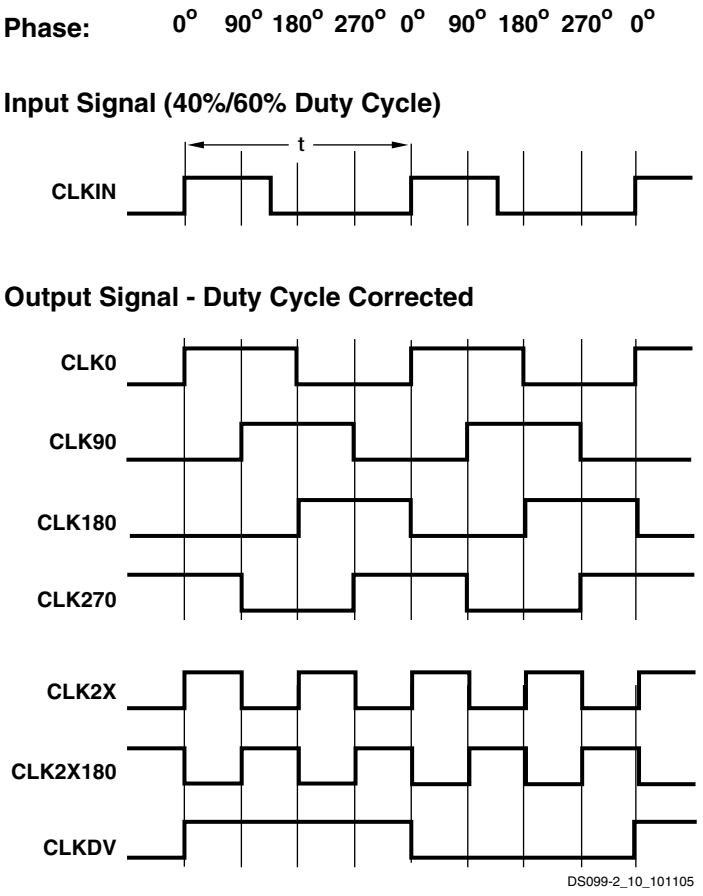


Figure 43: **Characteristics of the DLL Clock Outputs**

Basic Frequency Synthesis Outputs

The DLL component provides basic options for frequency multiplication and division in addition to the more flexible synthesis capability of the DFS component, described in a later section. These operations result in output clock signals with frequencies that are either a fraction (for division) or a multiple (for multiplication) of the incoming clock frequency. The CLK2X output produces an in-phase signal that is twice the frequency of CLKIN. The CLK2X180 output also doubles the frequency, but is 180° out-of-phase with respect to CLKIN. The CLKDIV output generates a clock frequency that is a predetermined fraction of the CLKIN frequency. The CLKDV_DIVIDE attribute determines the factor used to divide the CLKIN frequency. The attribute can be set to various values as described in Table 29. The basic frequency synthesis outputs are described in Table 28.

Duty Cycle Correction of DLL Clock Outputs

The DLL output signals exhibit a 50% duty cycle, even if the incoming CLKIN signal has a different duty cycle. Fifty-percent duty cycle means that the High and Low times of each clock cycle are equal.

DLL Performance Differences Between Steppings

As indicated in Digital Clock Manager (DCM) Timing (Module 3), the Stepping 1 revision silicon supports higher maximum input and output frequencies. Stepping 1 devices are backwards compatible with Stepping 0 devices.

Digital Frequency Synthesizer (DFS)

The DFS unit generates clock signals where the output frequency is a product of the CLKIN input clock frequency and a ratio of two user-specified integers. The two dedicated outputs from the DFS unit, CLKFX and CLKFX180, are defined in Table 33.

Table 33: **DFS Signals**

Signal	Direction	Description
CLKFX	Output	Multiplies the CLKIN frequency by the attribute-value ratio (CLKFX_MULTIPLY/ CLKFX_DIVIDE) to generate a clock signal with a new target frequency.
CLKFX180	Output	Generates a clock signal with the same frequency as CLKFX, but shifted 180° out-of-phase.

The signal at the CLKFX180 output is essentially an inversion of the CLKFX signal. These two outputs always exhibit a 50% duty cycle, even when the CLKIN signal does not. The DFS clock outputs are active coincident with the seven DLL outputs and their output phase is controlled by the Phase Shifter unit (PS).

The output frequency (f_{CLKFX}) of the DFS is a function of the incoming clock frequency (f_{CLKIN}) and two integer attributes, as follows.

$$f_{CLKFX} = f_{CLKIN} \bullet \left(\frac{\text{CLKFX\_MULTIPLY}}{\text{CLKFX\_DIVIDE}} \right) \qquad \text{Eq 1}$$

The CLKFX_MULTIPLY attribute is an integer ranging from 2 to 32, inclusive, and forms the numerator in Equation 1.

By contrast, the clock switch matrixes on the top and bottom edges receive signals from any of the five following sources: two GCLK pins, two DCM outputs, or one Double-Line interconnect.

Table 41 indicates permissible connections between clock inputs and BUFGMUX elements. The I0-input provides the best input path to a clock buffer. The I1-input provides the secondary input for the clock multiplexer function.

The four BUFGMUX elements on the top edge are paired together and share inputs from the eight global clock inputs along the top edge. Each BUFGMUX pair connects to four of the eight global clock inputs, as shown in Figure 45. This optionally allows differential inputs to the global clock inputs without wasting a BUFGMUX element.

Table 41: **Connections from Clock Inputs to BUFGMUX Elements and Associated Quadrant Clock**

Quadrant Clock Line[1]	Left-Half BUFGMUX			Top or Bottom BUFGMUX			Right-Half BUFGMUX		
	Location[2]	I0 Input	I1 Input	Location[2]	I0 Input	I1 Input	Location[2]	I0 Input	I1 Input
H	*X0Y9*	LHCLK7	LHCLK6	*X1Y10*	GCLK7 or GCLK11	GCLK6 or GCLK10	*X3Y9*	RHCLK3	RHCLK2
G	*X0Y8*	LHCLK6	LHCLK7	*X1Y11*	GCLK6 or GCLK10	GCLK7 or GCLK11	*X3Y8*	RHCLK2	RHCLK3
F	*X0Y7*	LHCLK5	LHCLK4	*X2Y10*	GCLK5 or GCLK9	GCLK4 or GCLK8	*X3Y7*	RHCLK1	RHCLK0
E	*X0Y6*	LHCLK4	LHCLK5	*X2Y11*	GCLK4 or GCLK8	GCLK5 or GCLK9	*X3Y6*	RHCLK0	RHCLK1
D	*X0Y5*	LHCLK3	LHCLK2	*X1Y0*	GCLK3 or GCLK15	GCLK2 or GCLK14	*X3Y5*	RHCLK7	RHCLK6
C	*X0Y4*	LHCLK2	LHCLK3	*X1Y1*	GCLK2 or GCLK14	GCLK3 or GCLK15	*X3Y4*	RHCLK6	RHCLK7
B	*X0Y3*	LHCLK1	LHCLK0	*X2Y0*	GCLK1 or GCLK13	GCLK0 or GCLK12	*X3Y3*	RHCLK5	RHCLK4
A	*X0Y2*	LHCLK0	LHCLK1	*X2Y1*	GCLK0 or GCLK12	GCLK1 or GCLK13	*X3Y2*	RHCLK4	RHCLK5

Notes:

1. See Quadrant Clock Routing for connectivity details for the eight quadrant clocks.
2. See Figure 45 for specific BUFGMUX locations, and Figure 47 for information on how BUFGMUX elements drive onto a specific clock line within a quadrant.

The connections for the bottom-edge BUFGMUX elements are similar to the top-edge connections (see Figure 46).

On the left and right edges, only two clock inputs feed each pair of BUFGMUX elements.

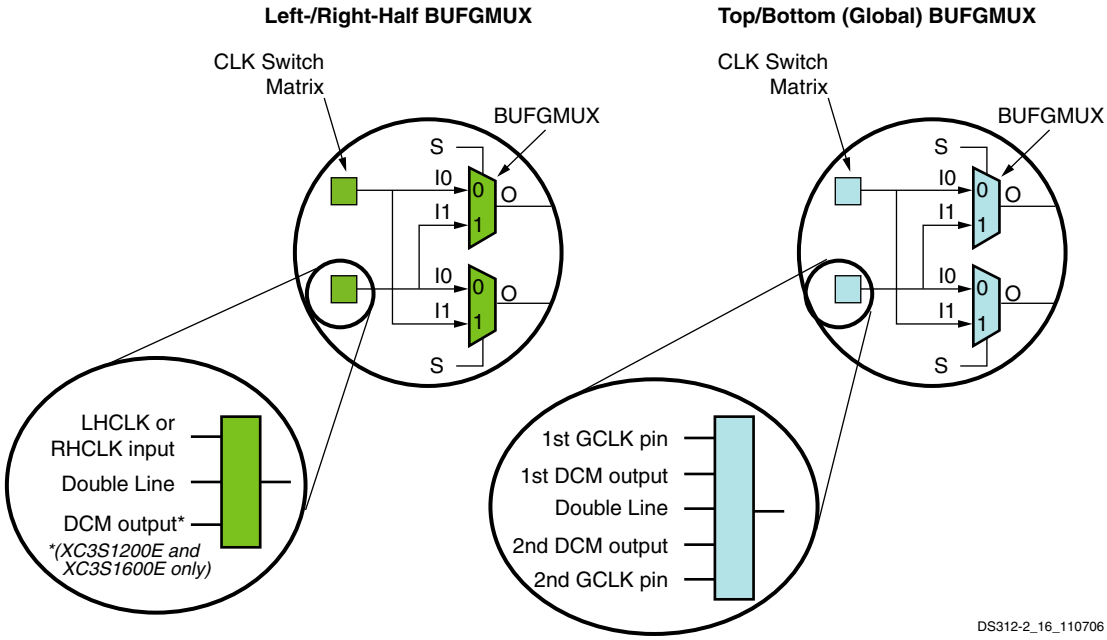


Figure 46: **Clock Switch Matrix to BUFGMUX Pair Connectivity**

Quadrant Clock Routing

The clock routing within the FPGA is quadrant-based, as shown in Figure 45. Each clock quadrant supports eight total clock signals, labeled 'A' through 'H' in Table 41 and Figure 47. The clock source for an individual clock line originates either from a global BUFGMUX element along the top and bottom edges or from a BUFGMUX element along the associated edge, as shown in Figure 47. The clock lines feed the synchronous resource elements (CLBs, IOBs, block RAM, multipliers, and DCMs) within the quadrant.

The four quadrants of the device are:

- Top Right (TR)
- Bottom Right (BR)
- Bottom Left (BL)
- Top Left (TL)

Note that the quadrant clock notation (TR, BR, BL, TL) is separate from that used for similar IOB placement constraints.

To estimate the quadrant location for a particular I/O, see the footprint diagrams in Module 4, Pinout Descriptions. For exact quadrant locations, use the floorplanning tool. In the QFP packages (VQ100, TQ144 and PQ208) the quadrant borders fall in the middle of each side of the package, at a GND pin. The clock inputs fall on the quadrant boundaries, as indicated in Table 42.

Table 42: **QFP Package Clock Quadrant Locations**

Clock Pins	Quadrant
GCLK[3:0]	BR
GCLK[7:4]	TR
GCLK[11:8]	TL
GCLK[15:12]	BL
RHCLK[3:0]	BR
RHCLK[7:4]	TR
LHCLK[3:0]	TL
LHCLK[7:4]	BL

In a few cases, a dedicated input is physically in one quadrant of the device but connects to a different clock quadrant:

- FT256, H16 is in clock quadrant BR
- FG320, K2 is in clock quadrant BL
- FG400, L8 is in clock quadrant TL and the I/O at N11 is in clock quadrant BL
- FG484, M2 is in clock quadrant TL and L15 is in clock quadrant BR

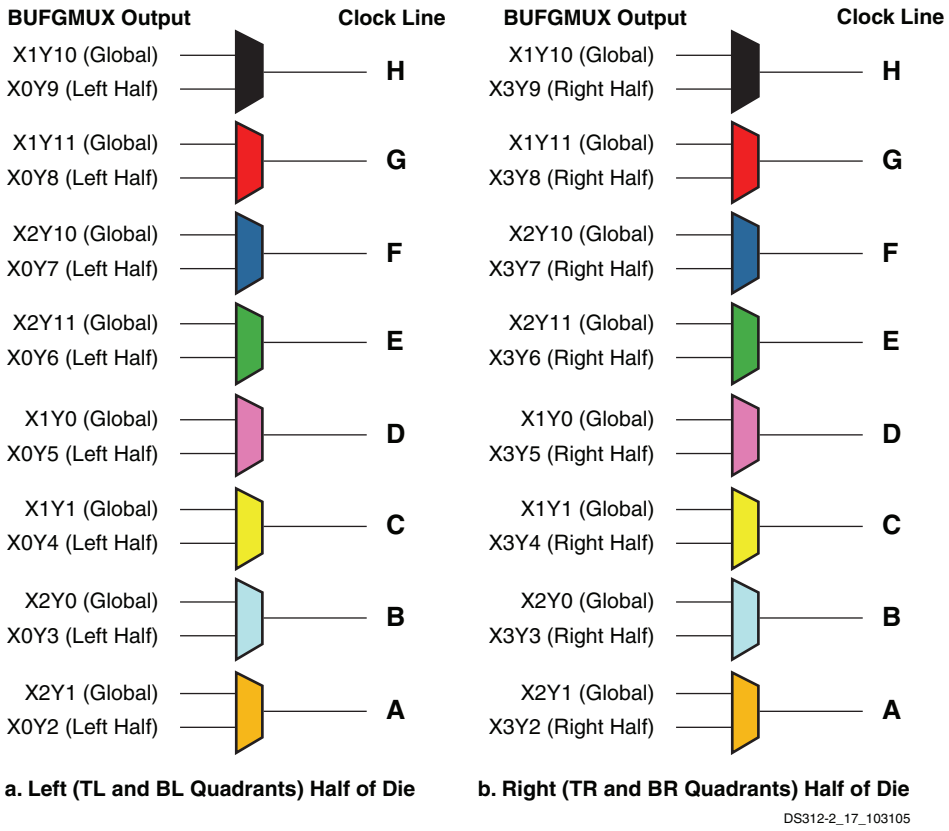


Figure 47: **Clock Sources for the Eight Clock Lines within a Clock Quadrant**

The outputs of the top or bottom BUFGMUX elements connect to two vertical spines, each comprising four vertical clock lines as shown in Figure 45. At the center of the die, these clock signals connect to the eight-line horizontal clock spine.

Outputs of the left and right BUFGMUX elements are routed onto the left or right horizontal spines, each comprising eight horizontal clock lines.

Each of the eight clock signals in a clock quadrant derives either from a global clock signal or a half clock signal. In other words, there are up to 24 total potential clock inputs to the FPGA, eight of which can connect to clocked elements in a single clock quadrant. Figure 47 shows how the clock lines in each quadrant are selected from associated BUFGMUX sources. For example, if quadrant clock ‘A’ in the bottom left (BL) quadrant originates from BUFGMUX_X2Y1, then the clock signal from BUFGMUX_X0Y2 is unavailable in the bottom left quadrant. However, the top left (TL) quadrant clock ‘A’ can still solely use the output from either BUFGMUX_X2Y1 or BUFGMUX_X0Y2 as the source.

To minimize the dynamic power dissipation of the clock network, the Xilinx development software automatically disables all clock segments not in use.

Table 46: **Pin Behavior during Configuration *(Cont'd)***

Pin Name	Master Serial	SPI (Serial Flash)	BPI (Parallel NOR Flash)	JTAG	Slave Parallel	Slave Serial	I/O Bank[3]
D0/DIN	DIN	DIN	D0		D0	DIN	2
RDWR_B			RDWR_B		RDWR_B		2
A23			A23				2
A22			A22				2
A21			A21				2
A20			A20				2
A19/VS2		VS2	A19				2
A18/VS1		VS1	A18				2
A17/VS0		VS0	A17				2
A16			A16				1
A15			A15				1
A14			A14				1
A13			A13				1
A12			A12				1
A11			A11				1
A10			A10				1
A9			A9				1
A8			A8				1
A7			A7				1
A6			A6				1
A5			A5				1
A4			A4				1
A3			A3				1
A2			A2				1
A1			A1				1
A0			A0				1
LDC0			LDC0				1
LDC1			LDC1				1
LDC2			LDC2				1
HDC			HDC				1

Notes:
1. Gray shaded cells represent pins that are in a high-impedance state (Hi-Z, floating) during configuration. These pins have an optional internal pull-up resistor to their respective V_{CCO} supply pin that is active throughout configuration if the HSWAP input is Low.
2. Yellow shaded cells represent pins with an internal pull-up resistor to its respective voltage supply rail that is active during configuration, regardless of the HSWAP pin.
3. Note that dual-purpose outputs are supplied by V_{CCO}, and configuration inputs are supplied by V_{CCAUX}.

The HSWAP pin itself has a pull-up resistor enabled during configuration. However, the VCCO_0 supply voltage must be applied before the pull-up resistor becomes active. If the VCCO_0 supply ramps after the VCCO_2 power supply, do not let HSWAP float; tie HSWAP to the desired logic level externally.

Spartan-3E FPGAs have only six dedicated configuration pins, including the DONE and PROG_B pins, and the four JTAG boundary-scan pins: TDI, TDO, TMS, and TCK. All other configuration pins are dual-purpose I/O pins and are available to the FPGA application after the DONE pin goes High. See Start-Up for additional information.

Table 47 shows the default I/O standard setting for the various configuration pins during the configuration process. The configuration interface is designed primarily for 2.5V operation when the VCCO_2 (and VCCO_1 in BPI mode) connects to 2.5V.

Table 47: **Default I/O Standard Setting During Configuration (VCCO_2 = 2.5V)**

Pin(s)	I/O Standard	Output Drive	Slew Rate
All, including CCLK	LVCMOS25	8 mA	Slow

(P) Similarly, the FPGA's HSWAP pin must be Low to enable pull-up resistors on all user-I/O pins during configuration or High to disable the pull-up resistors. The HSWAP control must remain at a constant logic level throughout FPGA configuration. After configuration, when the FPGA's DONE output goes High, the HSWAP pin is available as full-featured user-I/O pin and is powered by the VCCO_0 supply.

The FPGA's DOUT pin is used in daisy-chain applications, described later. In a single-FPGA application, the FPGA's DOUT pin is not used but is actively driving during the configuration process.

Table 50: **Serial Master Mode Connections**

Pin Name	FPGA Direction	Description	During Configuration	After Configuration
HSWAP (P)	Input	**User I/O Pull-Up Control**. When Low during configuration, enables pull-up resistors in all I/O pins to respective I/O bank V_{CCO} input. 0: Pull-ups during configuration 1: No pull-ups	Drive at valid logic level throughout configuration.	User I/O
M[2:0]	Input	**Mode Select**. Selects the FPGA configuration mode. See Design Considerations for the HSWAP, M[2:0], and VS[2:0] Pins.	M2 = 0, M1 = 0, M0 = 0. Sampled when INIT_B goes High.	User I/O
DIN	Input	**Serial Data Input**.	Receives serial data from PROM's D0 output.	User I/O
CCLK	Output	**Configuration Clock**. Generated by FPGA internal oscillator. Frequency controlled by ***ConfigRate*** bitstream generator option. If CCLK PCB trace is long or has multiple connections, terminate this output to maintain signal integrity. See CCLK Design Considerations.	Drives PROM's CLK clock input.	User I/O
DOUT	Output	**Serial Data Output**.	Actively drives. Not used in single-FPGA designs. In a daisy-chain configuration, this pin connects to DIN input of the next FPGA in the chain.	User I/O
INIT_B	Open-drain bidirectional I/O	**Initialization Indicator**. Active Low. Goes Low at start of configuration during Initialization memory clearing process. Released at end of memory clearing, when mode select pins are sampled. Requires external 4.7 kΩ pull-up resistor to VCCO_2.	Connects to PROM's OE/RESET input. FPGA clears PROM's address counter at start of configuration, enables outputs during configuration. PROM also holds FPGA in Initialization state until PROM reaches Power-On Reset (POR) state. If CRC error detected during configuration, FPGA drives INIT_B Low.	User I/O. If unused in the application, drive INIT_B High.
DONE	Open-drain bidirectional I/O	**FPGA Configuration Done**. Low during configuration. Goes High when FPGA successfully completes configuration. Requires external 330 Ω pull-up resistor to 2.5V.	Connects to PROM's chip-enable (CE) input. Enables PROM during configuration. Disables PROM after configuration.	Pulled High via external pull-up. When High, indicates that the FPGA successfully configured.
PROG_B	Input	**Program FPGA**. Active Low. When asserted Low for 500 ns or longer, forces the FPGA to restart its configuration process by clearing configuration memory and resetting the DONE and INIT_B pins once PROG_B returns High. Recommend external 4.7 kΩ pull-up resistor to 2.5V. Internal pull-up value may be weaker (see Table 78). If driving externally with a 3.3V output, use an open-drain or open-collector driver or use a current limiting series resistor.	Must be High during configuration to allow configuration to start. Connects to PROM's CF pin, allowing JTAG PROM programming algorithm to reprogram the FPGA.	Drive PROG_B Low and release to reprogram FPGA.

Table 53: **Variant Select Codes for Various SPI Serial Flash PROMs**

VS2	VS1	VS0	SPI Read Command	Dummy Bytes	SPI Serial Flash Vendor	SPI Flash Family	iMPACT Programming Support
1	1	1	FAST READ (0x0B) (see Figure 53)	1	STMicroelectronics (ST)	M25Pxx M25PExx/M45PExx	Yes
					Atmel	AT45DB 'D'-Series Data Flash	Yes
						AT26 / AT25(1)	
					Intel	S33	
					Spansion *(AMD, Fujitsu)*	S25FLxxxA	
					Winbond *(NexFlash)*	NX25 / W25	
					Macronix	MX25Lxxxx	
					Silicon Storage Technology (SST)	SST25LFxxxA SST25VFxxxA	
					Programmable Microelectronics Corp. (PMC)	Pm25LVxxx	
					AMIC Technology	A25L	
					Eon Silicon Solution, Inc.	EN25	
1	0	1	READ (0x03) (see Figure 53)	0	STMicroelectronics (ST)	M25Pxx M25PExx/M45PExx	Yes
					Spansion *(AMD, Fujitsu)*	S25FLxxxA	
					Winbond *(NexFlash)*	NX25 / W25	
					Macronix	MX25Lxxxx	
					Silicon Storage Technology (SST)	SST25LFxxxA SST25VFxxxA SST25VFxxx	
					Programmable Microelectronics Corp. (PMC)	Pm25LVxxx	
1	1	0	READ ARRAY (0xE8) (see Figure 54)	4	Atmel Corporation	AT45DB DataFlash *(use only 'C' or 'D' Series for Industrial temperature range)*	Yes
Others			Reserved				

Notes:
1. See iMPACT documentation for specific device support.

WRITER NOTE: Many of the URLs in this table are obsolete or otherwise broken.

During the configuration process, CCLK is controlled by the FPGA and limited to the frequencies generated by the FPGA. After configuration, the FPGA application can use other clock signals to drive the CCLK pin and can further optimize SPI-based communication.

Refer to the individual SPI peripheral data sheet for specific interface and communication protocol requirements.

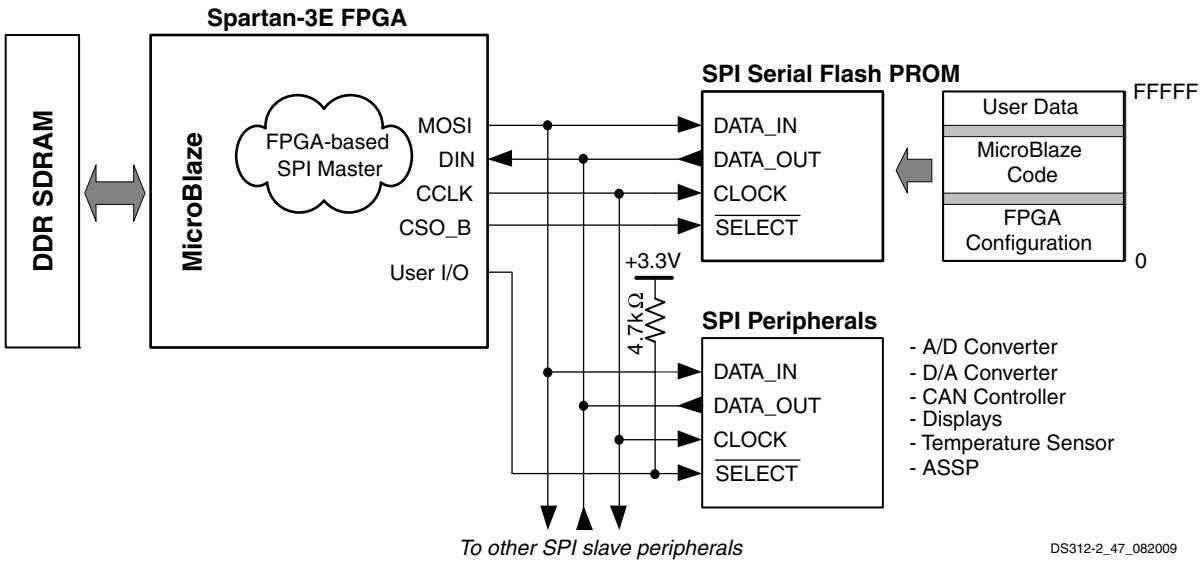


Figure 56: **Using the SPI Flash Interface After Configuration**

Powering Spartan-3E FPGAs

For additional information, refer to the “Powering Spartan-3 Generation FPGAs” chapter in UG331.

Voltage Supplies

Like Spartan-3 FPGAs, Spartan-3E FPGAs have multiple voltage supply inputs, as shown in Table 70. There are two supply inputs for internal logic functions, V_{CCINT} and V_{CCAUX}. Each of the four I/O banks has a separate V_{CCO} supply input that powers the output buffers within the associated I/O bank. All of the V_{CCO} connections to a specific I/O bank must be connected and must connect to the same voltage.

Table 70: **Spartan-3E Voltage Supplies**

Supply Input	Description	Nominal Supply Voltage
V_{CCINT}	Internal core supply voltage. Supplies all internal logic functions, such as CLBs, block RAM, and multipliers. Input to Power-On Reset (POR) circuit.	1.2V
V_{CCAUX}	Auxiliary supply voltage. Supplies Digital Clock Managers (DCMs), differential drivers, dedicated configuration pins, JTAG interface. Input to Power-On Reset (POR) circuit.	2.5V
VCCO_0	Supplies the output buffers in I/O Bank 0, the bank along the top edge of the FPGA.	Selectable, 3.3V, 2.5V, 1.8, 1.5V, or 1.2V
VCCO_1	Supplies the output buffers in I/O Bank 1, the bank along the right edge of the FPGA. In Byte-Wide Peripheral Interface (BPI) Parallel Flash Mode, connects to the same voltage as the Flash PROM.	Selectable, 3.3V, 2.5V, 1.8, 1.5V, or 1.2V
VCCO_2	Supplies the output buffers in I/O Bank 2, the bank along the bottom edge of the FPGA. Connects to the same voltage as the FPGA configuration source. Input to Power-On Reset (POR) circuit.	Selectable, 3.3V, 2.5V, 1.8, 1.5V, or 1.2V
VCCO_3	Supplies the output buffers in I/O Bank 3, the bank along the left edge of the FPGA.	Selectable, 3.3V, 2.5V, 1.8, 1.5V, or 1.2V

In a 3.3V-only application, all four V_{CCO} supplies connect to 3.3V. However, Spartan-3E FPGAs provide the ability to bridge between different I/O voltages and standards by applying different voltages to the V_{CCO} inputs of different banks. Refer to I/O Banking Rules for which I/O standards can be intermixed within a single I/O bank.

Each I/O bank also has an separate, optional input voltage reference supply, called V_{REF}. If the I/O bank includes an I/O standard that requires a voltage reference such as HSTL or SSTL, then all V_{REF} pins within the I/O bank must be connected to the same voltage.

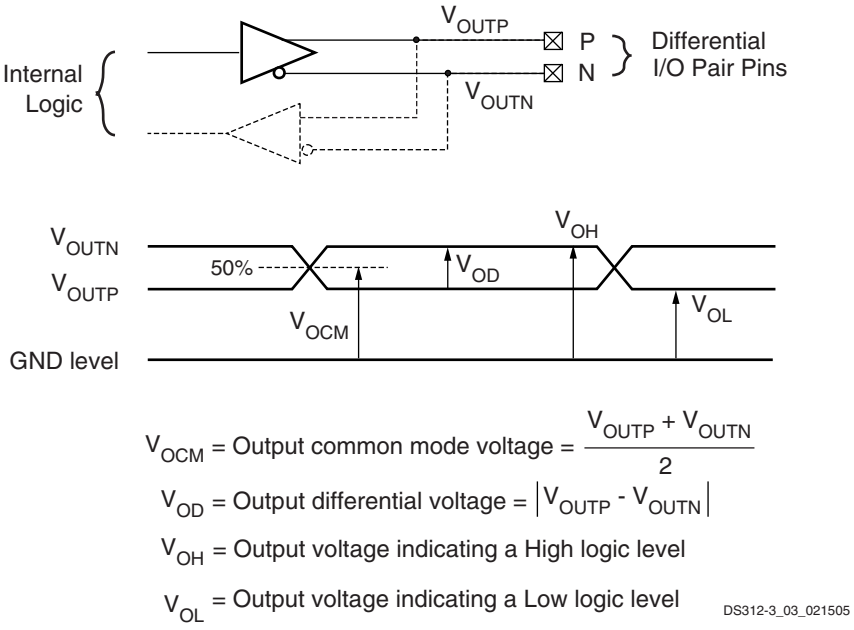


Figure 70: **Differential Output Voltages**

Table 83: **DC Characteristics of User I/Os Using Differential Signal Standards**

IOSTANDARD Attribute	V_{OD}			ΔV_{OD}		V_{OCM}			ΔV_{OCM}		V_{OH}	V_{OL}
	Min (mV)	Typ (mV)	Max (mV)	Min (mV)	Max (mV)	Min (V)	Typ (V)	Max (V)	Min (mV)	Max (mV)	Min (V)	Max (V)
LVDS_25	250	350	450	–	–	1.125	–	1.375	–	–	–	–
BLVDS_25	250	350	450	–	–	–	1.20	–	–	–	–	–
MINI_LVDS_25	300	–	600	–	50	1.0	–	1.4	–	50	–	–
RSDS_25	100	–	400	–	–	1.1	–	1.4	–	–	–	–
DIFF_HSTL_I_18	–	–	–	–	–	–	–	–	–	–	V_{CCO} – 0.4	0.4
DIFF_HSTL_III_18	–	–	–	–	–	–	–	–	–	–	V_{CCO} – 0.4	0.4
DIFF_SSTL18_I	–	–	–	–	–	–	–	–	–	–	V_{TT} + 0.475	V_{TT} – 0.475
DIFF_SSTL2_I	–	–	–	–	–	–	–	–	–	–	V_{TT} + 0.61	V_{TT} – 0.61

Notes:

1. The numbers in this table are based on the conditions set forth in Table 77 and Table 82.
2. Output voltage measurements for all differential standards are made with a termination resistor (R_T) of 100Ω across the N and P pins of the differential signal pair. The exception is for BLVDS, shown in Figure 71.
3. At any given time, no more than two of the following differential output standards may be assigned to an I/O bank: LVDS_25, RSDS_25, MINI_LVDS_25

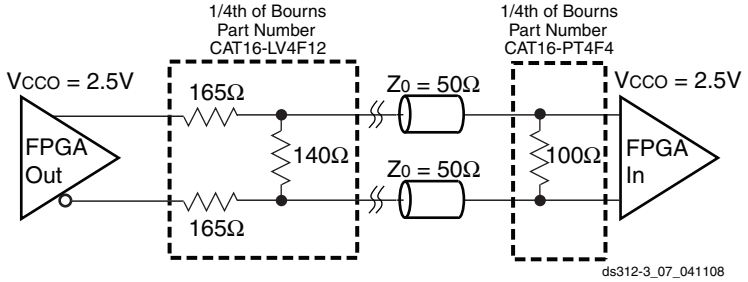


Figure 71: **External Termination Resistors for BLVDS Transmitter and BLVDS Receiver**

Switching Characteristics

All Spartan-3E FPGAs ship in two speed grades: -4 and the higher performance -5. Switching characteristics in this document may be designated as Advance, Preliminary, or Production, as shown in Table 84. Each category is defined as follows:

Advance: These specifications are based on simulations only and are typically available soon after establishing FPGA specifications. Although speed grades with this designation are considered relatively stable and conservative, some under-reporting might still occur.

Preliminary: These specifications are based on complete early silicon characterization. Devices and speed grades with this designation are intended to give a better indication of the expected performance of production silicon. The probability of under-reporting preliminary delays is greatly reduced compared to Advance data.

Production: These specifications are approved once enough production silicon of a particular device family member has been characterized to provide full correlation between speed files and devices over numerous production lots. There is no under-reporting of delays, and customers receive formal notification of any subsequent changes. Typically, the slowest speed grades transition to Production before faster speed grades.

Software Version Requirements

Production-quality systems must use FPGA designs compiled using a speed file designated as PRODUCTION status. FPGAs designs using a less mature speed file designation should only be used during system prototyping or pre-production qualification. FPGA designs with speed files designated as Advance or Preliminary should not be used in a production-quality system.

Whenever a speed file designation changes, as a device matures toward Production status, rerun the latest Xilinx ISE software on the FPGA design to ensure that the FPGA design incorporates the latest timing information and software updates.

All parameter limits are representative of worst-case supply voltage and junction temperature conditions. **Unless otherwise noted, the published parameter values apply to all Spartan-3E devices. AC and DC characteristics are specified using the same numbers for both commercial and industrial grades.**

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Timing parameters and their representative values are selected for inclusion below either because they are important as general design requirements or they indicate fundamental device performance characteristics. The Spartan-3E speed files (v1.27), part of the Xilinx Development Software, are the original source for many but not all of the values. The speed grade designations for these files are shown in Table 84. For more complete, more precise, and worst-case data, use the values reported by the Xilinx static timing analyzer (TRACE in the Xilinx development software) and back-annotated to the simulation netlist.

Table 84: **Spartan-3E v1.27 Speed Grade Designations**

Device	Advance	Preliminary	Production
XC3S100E			-MIN, -4, -5
XC3S250E			-MIN, -4, -5
XC3S500E			-MIN, -4, -5
XC3S1200E			-MIN, -4, -5
XC3S1600E			-MIN, -4, -5

Table 85 provides the history of the Spartan-3E speed files since all devices reached Production status.

Table 85: **Spartan-3E Speed File Version History**

Version	ISE Release	Description
1.27	9.2.03i	Added XA Automotive.
1.26	8.2.02i	Added -0/-MIN speed grade, which includes minimum values.
1.25	8.2.01i	Added XA Automotive devices to speed file. Improved model for left and right DCMs.
1.23	8.2i	Updated input setup/hold values based on default IFD_DELAY_VALUE settings.
1.21	8.1.03i	All Spartan-3E FPGAs and all speed grades elevated to Production status.

Table 88: **Setup and Hold Times for the IOB Input Path**

Symbol	Description	Conditions	IFD_DELAY_VALUE=	Device	Speed Grade		Units
					-5	-4	
					Min	Min	
Setup Times							
T_{IOPICK}	Time from the setup of data at the Input pin to the active transition at the ICLK input of the Input Flip-Flop (IFF). No Input Delay is programmed.	LVCMOS25(2), IFD_DELAY_VALUE = 0	0	All	1.84	2.12	ns
$T_{IOPICKD}$	Time from the setup of data at the Input pin to the active transition at the IFF's ICLK input. The Input Delay is programmed.	LVCMOS25(2), IFD_DELAY_VALUE = default software setting	2	XC3S100E	6.12	7.01	ns
			3	All Others	6.76	7.72	
Hold Times							
T_{IOICKP}	Time from the active transition at the IFF's ICLK input to the point where data must be held at the Input pin. No Input Delay is programmed.	LVCMOS25(3), IFD_DELAY_VALUE = 0	0	All	–0.76	–0.76	ns
$T_{IOICKPD}$	Time from the active transition at the IFF's ICLK input to the point where data must be held at the Input pin. The Input Delay is programmed.	LVCMOS25(3), IFD_DELAY_VALUE = default software setting	2	XC3S100E	–3.93	–3.93	ns
			3	All Others	–3.50	–3.50	
Set/Reset Pulse Width							
T_{RPW_IOB}	Minimum pulse width to SR control input on IOB			All	1.57	1.80	ns

Notes:

1. The numbers in this table are tested using the methodology presented in Table 95 and are based on the operating conditions set forth in Table 77 and Table 80.
2. This setup time requires adjustment whenever a signal standard other than LVCMOS25 is assigned to the data Input. If this is true, add the appropriate Input adjustment from Table 91.
3. These hold times require adjustment whenever a signal standard other than LVCMOS25 is assigned to the data Input. If this is true, subtract the appropriate Input adjustment from Table 91. When the hold time is negative, it is possible to change the data before the clock's active edge.

Table 89: **Sample Window (Source Synchronous)**

Symbol	Description	Max	Units
T_{SAMP}	Setup and hold capture window of an IOB input flip-flop	The input capture sample window value is highly specific to a particular application, device, package, I/O standard, I/O placement, DCM usage, and clock buffer. Please consult the appropriate Xilinx application note for application-specific values. • XAPP485: *1:7 Deserialization in Spartan-3E FPGAs at Speeds Up to 666 Mbps*	ps

Table 137: **TQ144 Package Pinout *(Cont'd)***

Bank	XC3S100E Pin Name	XC3S250E Pin Name	TQ144 Pin	Type
1	IO_L02P_1/A14	IO_L02P_1/A14	P76	DUAL
1	IO_L03N_1/A11	IO_L03N_1/A11	P82	DUAL
1	IO_L03P_1/A12	IO_L03P_1/A12	P81	DUAL
1	IO_L04N_1/A9/RHCLK1	IO_L04N_1/A9/RHCLK1	P86	RHCLK/DUAL
1	IO_L04P_1/A10/RHCLK0	IO_L04P_1/A10/RHCLK0	P85	RHCLK/DUAL
1	IO_L05N_1/A7/RHCLK3/TRDY1	IO_L05N_1/A7/RHCLK3	P88	RHCLK/DUAL
1	IO_L05P_1/A8/RHCLK2	IO_L05P_1/A8/RHCLK2	P87	RHCLK/DUAL
1	IO_L06N_1/A5/RHCLK5	IO_L06N_1/A5/RHCLK5	P92	RHCLK/DUAL
1	IO_L06P_1/A6/RHCLK4/IRDY1	IO_L06P_1/A6/RHCLK4	P91	RHCLK/DUAL
1	IO_L07N_1/A3/RHCLK7	IO_L07N_1/A3/RHCLK7	P94	RHCLK/DUAL
1	IO_L07P_1/A4/RHCLK6	IO_L07P_1/A4/RHCLK6	P93	RHCLK/DUAL
1	IO_L08N_1/A1	IO_L08N_1/A1	P97	DUAL
1	IO_L08P_1/A2	IO_L08P_1/A2	P96	DUAL
1	IO_L09N_1/LDC0	IO_L09N_1/LDC0	P104	DUAL
1	IO_L09P_1/HDC	IO_L09P_1/HDC	P103	DUAL
1	IO_L10N_1/LDC2	IO_L10N_1/LDC2	P106	DUAL
1	IO_L10P_1/LDC1	IO_L10P_1/LDC1	P105	DUAL
1	IP	IP	P78	INPUT
1	IP	IP	P84	INPUT
1	IP	IP	P89	INPUT
1	IP	IP	P101	INPUT
1	IP	IP	P107	INPUT
1	IP/VREF_1	IP/VREF_1	P95	VREF
1	VCCO_1	VCCO_1	P79	VCCO
1	VCCO_1	VCCO_1	P100	VCCO
2	IO/D5	IO/D5	P52	DUAL
2	IO/M1	IO/M1	P60	DUAL
2	IP/VREF_2	IO/VREF_2	P66	***100E:*** VREF(INPUT) ***250E:*** VREF(I/O)
2	IO_L01N_2/INIT_B	IO_L01N_2/INIT_B	P40	DUAL
2	IO_L01P_2/CSO_B	IO_L01P_2/CSO_B	P39	DUAL
2	IO_L02N_2/MOSI/CSI_B	IO_L02N_2/MOSI/CSI_B	P44	DUAL
2	IO_L02P_2/DOUT/BUSY	IO_L02P_2/DOUT/BUSY	P43	DUAL
2	IO_L04N_2/D6/GCLK13	IO_L04N_2/D6/GCLK13	P51	DUAL/GCLK
2	IO_L04P_2/D7/GCLK12	IO_L04P_2/D7/GCLK12	P50	DUAL/GCLK
2	IO_L05N_2/D3/GCLK15	IO_L05N_2/D3/GCLK15	P54	DUAL/GCLK
2	IO_L05P_2/D4/GCLK14	IO_L05P_2/D4/GCLK14	P53	DUAL/GCLK
2	IO_L07N_2/D1/GCLK3	IO_L07N_2/D1/GCLK3	P59	DUAL/GCLK
2	IO_L07P_2/D2/GCLK2	IO_L07P_2/D2/GCLK2	P58	DUAL/GCLK
2	IO_L08N_2/DIN/D0	IO_L08N_2/DIN/D0	P63	DUAL
2	IO_L08P_2/M0	IO_L08P_2/M0	P62	DUAL
2	IO_L09N_2/VS1/A18	IO_L09N_2/VS1/A18	P68	DUAL
2	IO_L09P_2/VS2/A19	IO_L09P_2/VS2/A19	P67	DUAL
2	IO_L10N_2/CCLK	IO_L10N_2/CCLK	P71	DUAL
2	IO_L10P_2/VS0/A17	IO_L10P_2/VS0/A17	P70	DUAL

FG400: 400-ball Fine-pitch Ball Grid Array

The 400-ball fine-pitch ball grid array, FG400, supports two different Spartan-3E FPGAs, including the XC3S1200E and the XC3S1600E. Both devices share a common footprint for this package as shown in Table 152 and Figure 87.

Table 152 lists all the FG400 package pins. They are sorted by bank number and then by pin name. Pairs of pins that form a differential I/O pair appear together in the table. The table also shows the pin number for each pin and the pin type, as defined earlier.

An electronic version of this package pinout table and footprint diagram is available for download from the Xilinx website at:

http://www.xilinx.com/support/documentation/data_sheets/s3e_pin.zip

Pinout Table

Table 152: **FG400 Package Pinout**

Bank	XC3S1200E XC3S1600E Pin Name	FG400 Ball	Type
0	IO	A3	I/O
0	IO	A8	I/O
0	IO	A12	I/O
0	IO	C7	I/O
0	IO	C10	I/O
0	IO	E8	I/O
0	IO	E13	I/O
0	IO	E16	I/O
0	IO	F13	I/O
0	IO	F14	I/O
0	IO	G7	I/O
0	IO/VREF_0	C11	VREF
0	IO_L01N_0	B17	I/O
0	IO_L01P_0	C17	I/O
0	IO_L03N_0/VREF_0	A18	VREF
0	IO_L03P_0	A19	I/O
0	IO_L04N_0	A17	I/O
0	IO_L04P_0	A16	I/O
0	IO_L06N_0	A15	I/O
0	IO_L06P_0	B15	I/O
0	IO_L07N_0	C14	I/O
0	IO_L07P_0	D14	I/O
0	IO_L09N_0/VREF_0	A13	VREF
0	IO_L09P_0	A14	I/O
0	IO_L10N_0	B13	I/O
0	IO_L10P_0	C13	I/O
0	IO_L12N_0	C12	I/O
0	IO_L12P_0	D12	I/O
0	IO_L13N_0	E12	I/O
0	IO_L13P_0	F12	I/O
0	IO_L15N_0/GCLK5	G11	GCLK
0	IO_L15P_0/GCLK4	F11	GCLK
0	IO_L16N_0/GCLK7	E10	GCLK
0	IO_L16P_0/GCLK6	E11	GCLK
0	IO_L18N_0/GCLK11	A9	GCLK
0	IO_L18P_0/GCLK10	A10	GCLK
0	IO_L19N_0	F9	I/O
0	IO_L19P_0	E9	I/O
0	IO_L21N_0	C9	I/O
0	IO_L21P_0	D9	I/O
0	IO_L22N_0/VREF_0	B8	VREF
0	IO_L22P_0	B9	I/O
0	IO_L24N_0/VREF_0	F7	VREF
0	IO_L24P_0	F8	I/O
0	IO_L25N_0	A6	I/O
0	IO_L25P_0	A7	I/O
0	IO_L27N_0	B5	I/O
0	IO_L27P_0	B6	I/O
0	IO_L28N_0	D6	I/O
0	IO_L28P_0	C6	I/O
0	IO_L30N_0/VREF_0	C5	VREF
0	IO_L30P_0	D5	I/O
0	IO_L31N_0	A2	I/O
0	IO_L31P_0	B2	I/O
0	IO_L32N_0/HSWAP	D4	DUAL
0	IO_L32P_0	C4	I/O
0	IP	B18	INPUT
0	IP	E5	INPUT
0	IP_L02N_0	C16	INPUT
0	IP_L02P_0	D16	INPUT
0	IP_L05N_0	D15	INPUT
0	IP_L05P_0	C15	INPUT
0	IP_L08N_0	E14	INPUT
0	IP_L08P_0	E15	INPUT
0	IP_L11N_0	G14	INPUT
0	IP_L11P_0	G13	INPUT
0	IP_L14N_0	B11	INPUT
0	IP_L14P_0	B12	INPUT
0	IP_L17N_0/GCLK9	G10	GCLK

Table 154: **FG484 Package Pinout *(Cont'd)***

Bank	XC3S1600E Pin Name	FG484 Ball	Type
0	IP	E6	INPUT
0	IP_L02N_0	D17	INPUT
0	IP_L02P_0	D18	INPUT
0	IP_L05N_0	C17	INPUT
0	IP_L05P_0	B17	INPUT
0	IP_L08N_0	E15	INPUT
0	IP_L08P_0	D15	INPUT
0	IP_L14N_0	D13	INPUT
0	IP_L14P_0	C13	INPUT
0	IP_L17N_0	A12	INPUT
0	IP_L17P_0	A13	INPUT
0	IP_L20N_0/GCLK9	H11	GCLK
0	IP_L20P_0/GCLK8	H12	GCLK
0	IP_L23N_0	F10	INPUT
0	IP_L23P_0	F11	INPUT
0	IP_L26N_0	G9	INPUT
0	IP_L26P_0	G10	INPUT
0	IP_L31N_0	C8	INPUT
0	IP_L31P_0	D8	INPUT
0	IP_L34N_0	C7	INPUT
0	IP_L34P_0	C6	INPUT
0	IP_L37N_0	A3	INPUT
0	IP_L37P_0	A2	INPUT
0	VCCO_0	B5	VCCO
0	VCCO_0	B10	VCCO
0	VCCO_0	B14	VCCO
0	VCCO_0	B18	VCCO
0	VCCO_0	E8	VCCO
0	VCCO_0	F14	VCCO
0	VCCO_0	G11	VCCO
1	IO_L01N_1/A15	Y22	DUAL
1	IO_L01P_1/A16	AA22	DUAL
1	IO_L02N_1/A13	W21	DUAL
1	IO_L02P_1/A14	Y21	DUAL
1	IO_L03N_1/VREF_1	W20	VREF
1	IO_L03P_1	V20	I/O
1	IO_L04N_1	U19	I/O
1	IO_L04P_1	V19	I/O
1	IO_L05N_1	V22	I/O
1	IO_L05P_1	W22	I/O
1	IO_L06N_1	T19	I/O
1	IO_L06P_1	T18	I/O
1	IO_L07N_1/VREF_1	U20	VREF
1	IO_L07P_1	U21	I/O

Table 154: **FG484 Package Pinout *(Cont'd)***

Bank	XC3S1600E Pin Name	FG484 Ball	Type
1	IO_L08N_1	T22	I/O
1	IO_L08P_1	U22	I/O
1	IO_L09N_1	R19	I/O
1	IO_L09P_1	R18	I/O
1	IO_L10N_1	R16	I/O
1	IO_L10P_1	T16	I/O
1	IO_L11N_1	R21	I/O
1	IO_L11P_1	R20	I/O
1	IO_L12N_1/VREF_1	P18	VREF
1	IO_L12P_1	P17	I/O
1	IO_L13N_1	P22	I/O
1	IO_L13P_1	R22	I/O
1	IO_L14N_1	P15	I/O
1	IO_L14P_1	P16	I/O
1	IO_L15N_1	N18	I/O
1	IO_L15P_1	N19	I/O
1	IO_L16N_1/A11	N16	DUAL
1	IO_L16P_1/A12	N17	DUAL
1	IO_L17N_1/VREF_1	M20	VREF
1	IO_L17P_1	N20	I/O
1	IO_L18N_1/A9/RHCLK1	M22	RHCLK/ DUAL
1	IO_L18P_1/A10/RHCLK0	N22	RHCLK/ DUAL
1	IO_L19N_1/A7/RHCLK3/ TRDY1	M16	RHCLK/ DUAL
1	IO_L19P_1/A8/RHCLK2	M15	RHCLK/ DUAL
1	IO_L20N_1/A5/RHCLK5	L21	RHCLK/ DUAL
1	IO_L20P_1/A6/RHCLK4/ IRDY1	L20	RHCLK/ DUAL
1	IO_L21N_1/A3/RHCLK7	L19	RHCLK/ DUAL
1	IO_L21P_1/A4/RHCLK6	L18	RHCLK/ DUAL
1	IO_L22N_1/A1	K22	DUAL
1	IO_L22P_1/A2	L22	DUAL
1	IO_L23N_1/A0	K17	DUAL
1	IO_L23P_1	K16	I/O
1	IO_L24N_1	K19	I/O
1	IO_L24P_1	K18	I/O
1	IO_L25N_1	K15	I/O
1	IO_L25P_1	J15	I/O
1	IO_L26N_1	J20	I/O
1	IO_L26P_1	J21	I/O