AMD Xilinx - XC3S500E-5PQG208C Datasheet

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	1164
Number of Logic Elements/Cells	10476
Total RAM Bits	368640
Number of I/O	158
Number of Gates	500000
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xc3s500e-5pqg208c

Input Delay Functions

Each IOB has a programmable delay block that optionally delays the input signal. In Figure 6, the signal path has a coarse delay element that can be bypassed. The input signal then feeds a 6-tap delay line. The coarse and tap delays vary; refer to timing reports for specific delay values. All six taps are available via a multiplexer for use as an asynchronous input directly into the FPGA fabric. In this way, the delay is programmable in 12 steps. Three of the six taps are also available via a multiplexer to the D inputs of the synchronous storage elements. The delay inserted in the path to the storage element can be varied in six steps. The first, coarse delay element is common to both asynchronous and synchronous paths, and must be either used or not used for both paths.

The delay values are set up in the silicon once at configuration time—they are non-modifiable in device operation.

The primary use for the input delay element is to adjust the input delay path to ensure that there is no hold time requirement when using the input flip-flop(s) with a global clock. The default value is chosen automatically by the Xilinx software tools as the value depends on device size and the specific device edge where the flip-flop resides. The value set by the Xilinx ISE software is indicated in the Map report generated by the implementation tools, and the resulting effects on input timing are reported using the Timing Analyzer tool.

If the design uses a DCM in the clock path, then the delay element can be safely set to zero because the Delay-Locked Loop (DLL) compensation automatically ensures that there is still no input hold time requirement.

Both asynchronous and synchronous values can be modified, which is useful where extra delay is required on clock or data inputs, for example, in interfaces to various types of RAM.

These delay values are defined through the IBUF_DELAY_VALUE and the IFD_DELAY_VALUE parameters. The default IBUF_DELAY_VALUE is 0, bypassing the delay elements for the asynchronous input. The user can set this parameter to 0-12. The default IFD_DELAY_VALUE is AUTO. IBUF_DELAY_VALUE and IFD_DELAY_VALUE are independent for each input. If the same input pin uses both registered and non-registered input paths, both parameters can be used, but they must both be in the same half of the total delay (both either bypassing or using the coarse delay element).

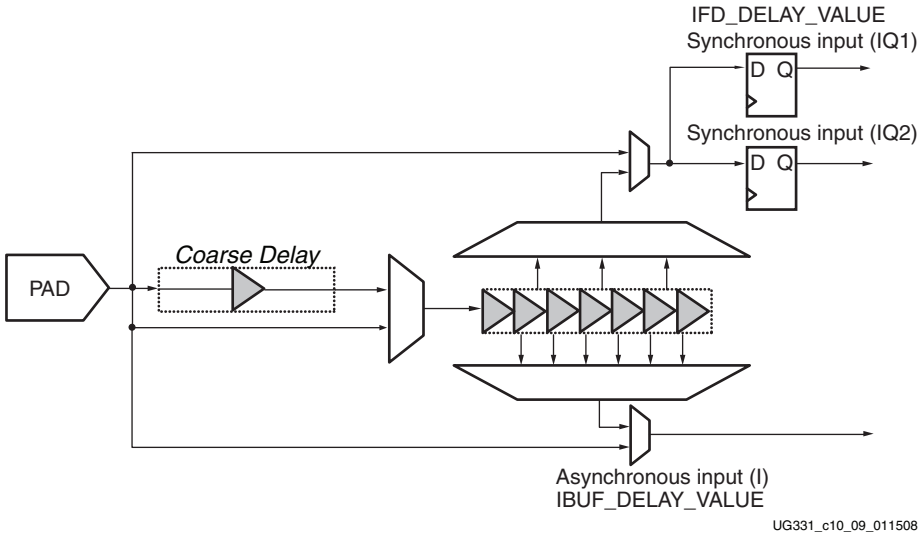

Figure 6: **Programmable Fixed Input Delay Elements**

Table 14: **Carry Logic Functions *(Cont'd)***

Function	Description
CY0G	Carry generation for top half of slice. Fixed selection of: · G1 or G2 inputs to the LUT (both equal 1 when a carry is to be generated) · GAND gate for multiplication · BY input for carry initialization · Fixed 1 or 0 input for use as a simple Boolean function
CYMUXF	Carry generation or propagation mux for bottom half of slice. Dynamic selection via CYSELF of: · CYINIT carry propagation (CYSELF = 1) · CY0F carry generation (CYSELF = 0)
CYMUXG	Carry generation or propagation mux for top half of slice. Dynamic selection via CYSELF of: · CYMUXF carry propagation (CYSELG = 1) · CY0G carry generation (CYSELG = 0)
CYSELF	Carry generation or propagation select for bottom half of slice. Fixed selection of: · F-LUT output (typically XOR result) · Fixed 1 to always propagate
CYSELG	Carry generation or propagation select for top half of slice. Fixed selection of: · G-LUT output (typically XOR result) · Fixed 1 to always propagate
XORF	Sum generation for bottom half of slice. Inputs from: · F-LUT · CYINIT carry signal from previous stage Result is sent to either the combinatorial or registered output for the top of the slice.
XORG	Sum generation for top half of slice. Inputs from: · G-LUT · CYMUXF carry signal from previous stage Result is sent to either the combinatorial or registered output for the top of the slice.
FAND	Multiplier partial product for bottom half of slice. Inputs: · F-LUT F1 input · F-LUT F2 input Result is sent through CY0F to become the carry generate signal into CYMUXF
GAND	Multiplier partial product for top half of slice. Inputs: · G-LUT G1 input · G-LUT G2 input Result is sent through CY0G to become the carry generate signal into CYMUXG

The basic usage of the carry logic is to generate a half-sum in the LUT via an XOR function, which generates or propagates a carry out COUT via the carry mux CYMUXF (or CYMUXG), and then complete the sum with the dedicated XORF (or XORG) gate and the carry input CIN. This structure allows two bits of an arithmetic function in each slice. The CYMUXF (or CYMUXG) can be instantiated using the MUXCY element, and the XORF (or XORG) can be instantiated using the XORCY element.

The FAND (or GAND) gate is used for partial product multiplication and can be instantiated using the MULT_AND component. Partial products are generated by two-input AND gates and then added. The carry logic is efficient for the adder, but one of the inputs must be outside the LUT as shown in Figure 23.

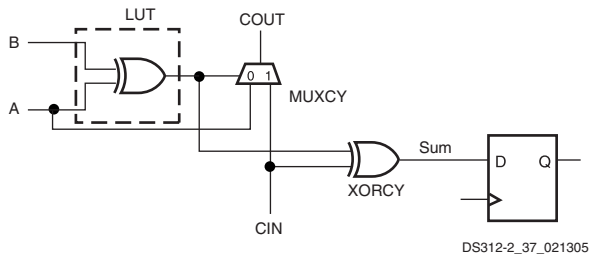

Figure 23: **Using the MUXCY and XORCY in the Carry Logic**

The FAND (or GAND) gate is used to duplicate one of the partial products, while the LUT generates both partial products and the XOR function, as shown in Figure 24.

Block RAM

For additional information, refer to the “Using Block RAM” chapter in UG331.

Spartan-3E devices incorporate 4 to 36 dedicated block RAMs, which are organized as dual-port configurable 18 Kbit blocks. Functionally, the block RAM is identical to the Spartan-3 architecture block RAM. Block RAM synchronously stores large amounts of data while distributed RAM, previously described, is better suited for buffering small amounts of data anywhere along signal paths. This section describes basic block RAM functions.

Each block RAM is configurable by setting the content’s initial values, default signal value of the output registers, port aspect ratios, and write modes. Block RAM can be used in single-port or dual-port modes.

Arrangement of RAM Blocks on Die

The block RAMs are located together with the multipliers on the die in one or two columns depending on the size of the device. The XC3S100E has one column of block RAM. The Spartan-3E devices ranging from the XC3S250E to XC3S1600E have two columns of block RAM. Table 21 shows the number of RAM blocks, the data storage capacity, and the number of columns for each device. Row(s) of CLBs are located above and below each block RAM column.

Table 21: **Number of RAM Blocks by Device**

Device	Total Number of RAM Blocks	Total Addressable Locations (bits)	Number of Columns
XC3S100E	4	73,728	1
XC3S250E	12	221,184	2
XC3S500E	20	368,640	2
XC3S1200E	28	516,096	2
XC3S1600E	36	663,552	2

Immediately adjacent to each block RAM is an embedded 18x18 hardware multiplier. The upper 16 bits of the block RAM's Port A Data input bus are shared with the upper 16 bits of the A multiplicand input bus of the multiplier. Similarly, the upper 16 bits of Port B's data input bus are shared with the B multiplicand input bus of the multiplier.

The Internal Structure of the Block RAM

The block RAM has a dual port structure. The two identical data ports called A and B permit independent access to the common block RAM, which has a maximum capacity of 18,432 bits, or 16,384 bits with no parity bits (see parity bits description in Table 22). Each port has its own dedicated set of data, control, and clock lines for synchronous read and write operations. There are four basic data paths, as shown in Figure 30:

1. Write to and read from Port A
2. Write to and read from Port B
3. Data transfer from Port A to Port B
4. Data transfer from Port B to Port A

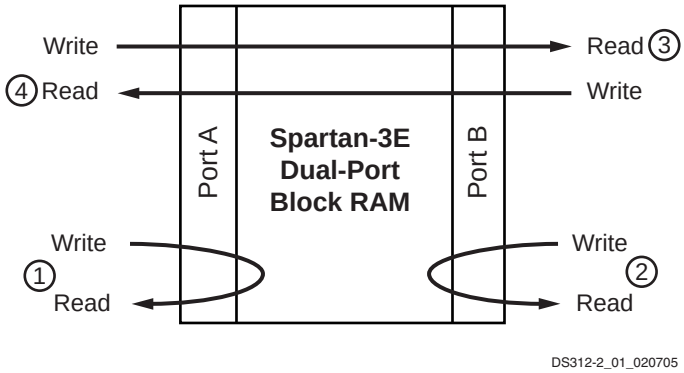

Figure 30: **Block RAM Data Paths**

Number of Ports

A choice among primitives determines whether the block RAM functions as dual- or single-port memory. A name of the form RAMB16_S[w_A]_S[w_B] calls out the dual-port primitive, where the integers w_A and w_B specify the total data path width at ports A and B, respectively. Thus, a RAMB16_S9_S18 is a dual-port RAM with a 9-bit Port A and an 18-bit Port B. A name of the form RAMB16_S[w] identifies the single-port primitive, where the integer w specifies the total data path width of the lone port A. A RAMB16_S18 is a single-port RAM with an 18-bit port.

Port Aspect Ratios

Each port of the block RAM can be configured independently to select a number of different possible widths for the data input (DI) and data output (DO) signals as shown in Table 22.

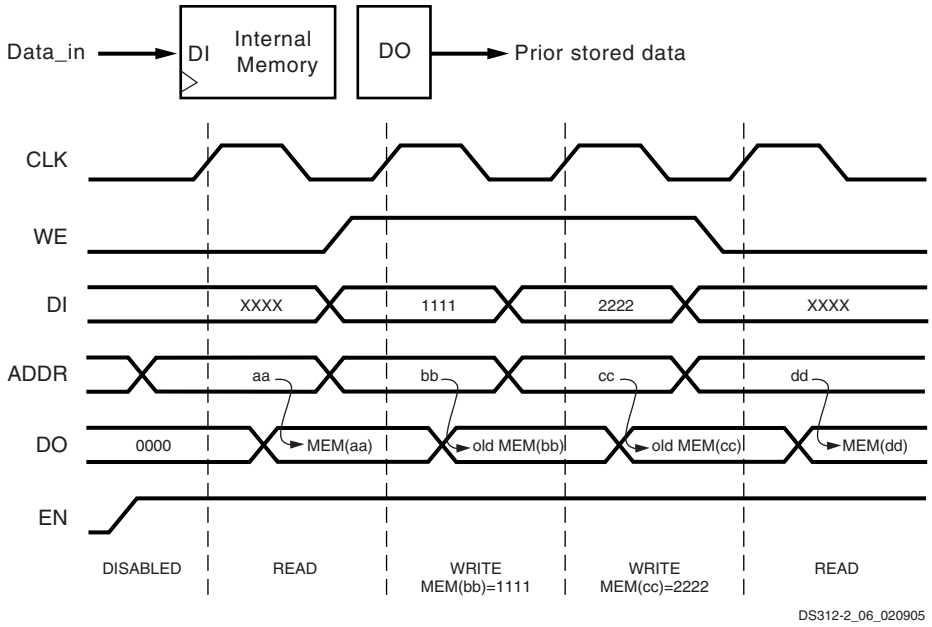

Figure 34: **Waveforms of Block RAM Data Operations with READ_FIRST Selected**

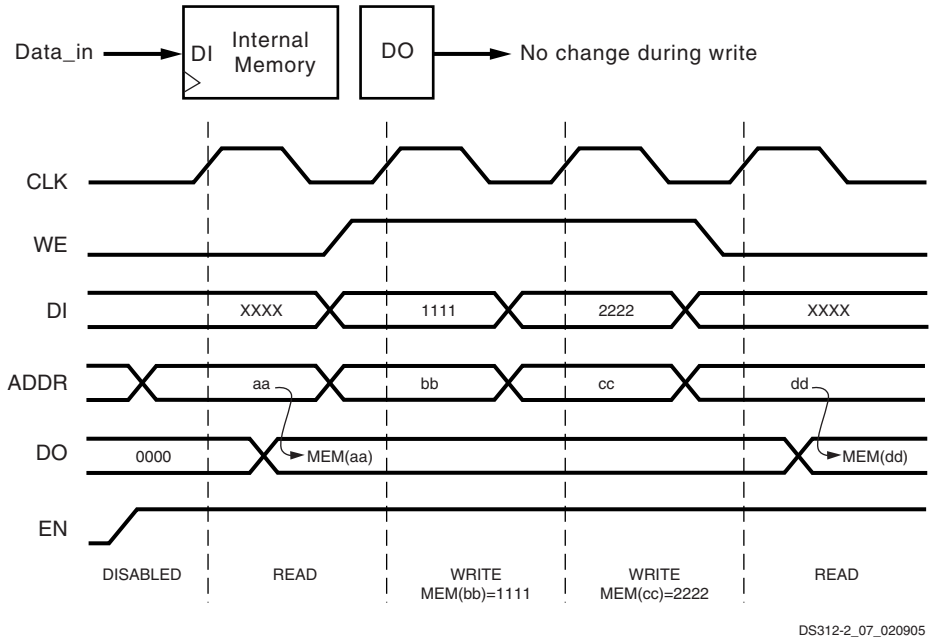

Figure 35: **Waveforms of Block RAM Data Operations with NO_CHANGE Selected**

Setting the WRITE_MODE attribute to a value of NO_CHANGE, puts the DO outputs in a latched state when asserting WE. Under this condition, the DO outputs retain the data driven just before WE is asserted. NO_CHANGE timing is shown in the portion of Figure 35 during which WE is High.

By contrast, the clock switch matrixes on the top and bottom edges receive signals from any of the five following sources: two GCLK pins, two DCM outputs, or one Double-Line interconnect.

Table 41 indicates permissible connections between clock inputs and BUFGMUX elements. The I0-input provides the best input path to a clock buffer. The I1-input provides the secondary input for the clock multiplexer function.

The four BUFGMUX elements on the top edge are paired together and share inputs from the eight global clock inputs along the top edge. Each BUFGMUX pair connects to four of the eight global clock inputs, as shown in Figure 45. This optionally allows differential inputs to the global clock inputs without wasting a BUFGMUX element.

Table 41: **Connections from Clock Inputs to BUFGMUX Elements and Associated Quadrant Clock**

Quadrant Clock Line[1]	Left-Half BUFGMUX			Top or Bottom BUFGMUX			Right-Half BUFGMUX		
	Location[2]	I0 Input	I1 Input	Location[2]	I0 Input	I1 Input	Location[2]	I0 Input	I1 Input
H	*X0Y9*	LHCLK7	LHCLK6	*X1Y10*	GCLK7 or GCLK11	GCLK6 or GCLK10	*X3Y9*	RHCLK3	RHCLK2
G	*X0Y8*	LHCLK6	LHCLK7	*X1Y11*	GCLK6 or GCLK10	GCLK7 or GCLK11	*X3Y8*	RHCLK2	RHCLK3
F	*X0Y7*	LHCLK5	LHCLK4	*X2Y10*	GCLK5 or GCLK9	GCLK4 or GCLK8	*X3Y7*	RHCLK1	RHCLK0
E	*X0Y6*	LHCLK4	LHCLK5	*X2Y11*	GCLK4 or GCLK8	GCLK5 or GCLK9	*X3Y6*	RHCLK0	RHCLK1
D	*X0Y5*	LHCLK3	LHCLK2	*X1Y0*	GCLK3 or GCLK15	GCLK2 or GCLK14	*X3Y5*	RHCLK7	RHCLK6
C	*X0Y4*	LHCLK2	LHCLK3	*X1Y1*	GCLK2 or GCLK14	GCLK3 or GCLK15	*X3Y4*	RHCLK6	RHCLK7
B	*X0Y3*	LHCLK1	LHCLK0	*X2Y0*	GCLK1 or GCLK13	GCLK0 or GCLK12	*X3Y3*	RHCLK5	RHCLK4
A	*X0Y2*	LHCLK0	LHCLK1	*X2Y1*	GCLK0 or GCLK12	GCLK1 or GCLK13	*X3Y2*	RHCLK4	RHCLK5

Notes:

1. See Quadrant Clock Routing for connectivity details for the eight quadrant clocks.
2. See Figure 45 for specific BUFGMUX locations, and Figure 47 for information on how BUFGMUX elements drive onto a specific clock line within a quadrant.

Interconnect

For additional information, refer to the “Using Interconnect” chapter in UG331.

Interconnect is the programmable network of signal pathways between the inputs and outputs of functional elements within the FPGA, such as IOBs, CLBs, DCMs, and block RAM.

Overview

Interconnect, also called routing, is segmented for optimal connectivity. Functionally, interconnect resources are identical to that of the Spartan-3 architecture. There are four kinds of interconnects: long lines, hex lines, double lines, and direct lines. The Xilinx Place and Route (PAR) software exploits the rich interconnect array to deliver optimal system performance and the fastest compile times.

Switch Matrix

The switch matrix connects to the different kinds of interconnects across the device. An interconnect tile, shown in Figure 48, is defined as a single switch matrix connected to a functional element, such as a CLB, IOB, or DCM. If a functional element spans across multiple switch matrices such as the block RAM or multipliers, then an interconnect tile is defined by the number of switch matrices connected to that functional element. A Spartan-3E device can be represented as an array of interconnect tiles where interconnect resources are for the channel between any two adjacent interconnect tile rows or columns as shown in Figure 49.

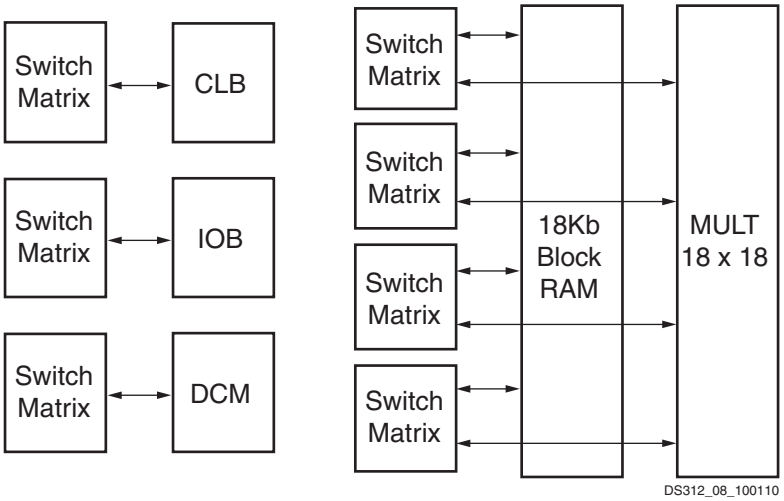

Figure 48: **Four Types of Interconnect Tiles (CLBs, IOBs, DCMs, and Block RAM/Multiplier)**

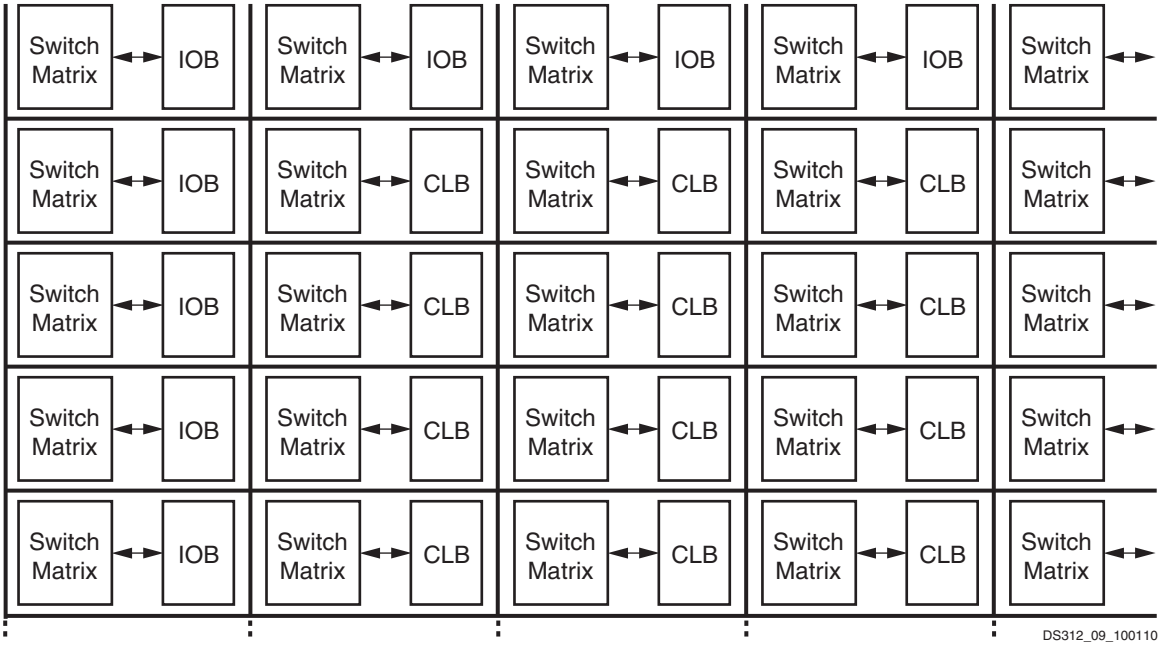

Figure 49: **Array of Interconnect Tiles in Spartan-3E FPGA**

Configuration

For additional information on configuration, refer to UG332: *Spartan-3 Generation Configuration User Guide*.

Differences from Spartan-3 FPGAs

In general, Spartan-3E FPGA configuration modes are a superset to those available in Spartan-3 FPGAs. Two new modes added in Spartan-3E FPGAs provide a glueless configuration interface to industry-standard parallel NOR Flash and SPI serial Flash memories.

Configuration Process

The function of a Spartan-3E FPGA is defined by loading application-specific configuration data into the FPGA's internal, reprogrammable CMOS configuration latches (CCLs), similar to the way a microprocessor's function is defined by its application program. For FPGAs, this configuration process uses a subset of the device pins, some of which are dedicated to configuration; other pins are merely borrowed and returned to the application as general-purpose user I/Os after configuration completes.

Spartan-3E FPGAs offer several configuration options to minimize the impact of configuration on the overall system design. In some configuration modes, the FPGA generates a clock and loads itself from an external memory source, either serially or via a byte-wide data path. Alternatively, an external host such as a microprocessor downloads the FPGA's configuration data using a simple synchronous serial interface or via a byte-wide peripheral-style interface. Furthermore, multiple-FPGA designs share a single configuration memory source, creating a structure called a daisy chain.

Three FPGA pins—M2, M1, and M0—select the desired configuration mode. The mode pin settings appear in Table 44. The mode pin values are sampled during the start of configuration when the FPGA's INIT_B output goes High. After the FPGA completes configuration, the mode pins are available as user I/Os.

Table 44: **Spartan-3E Configuration Mode Options and Pin Settings**

	Master Serial	SPI	BPI	Slave Parallel	Slave Serial	JTAG
M[2:0] mode pin settings	<0:0:0>	<0:0:1>	<0:1:0>=Up <0:1:1>=Down	<1:1:0>	<1:1:1>	<1:0:1>
Data width	Serial	Serial	Byte-wide	Byte-wide	Serial	Serial
Configuration memory source	Xilinx Platform Flash	Industry-standard SPI serial Flash	Industry-standard parallel NOR Flash or Xilinx parallel Platform Flash	Any source via microcontroller, CPU, Xilinx parallel Platform Flash, etc.	Any source via microcontroller, CPU, Xilinx Platform Flash, etc.	Any source via microcontroller, CPU, System ACE™ CF, etc.
Clock source	Internal oscillator	Internal oscillator	Internal oscillator	External clock on CCLK pin	External clock on CCLK pin	External clock on TCK pin
Total I/O pins borrowed during configuration	8	13	46	21	8	0
Configuration mode for downstream daisy-chained FPGAs	Slave Serial	Slave Serial	Slave Parallel	Slave Parallel or Memory Mapped	Slave Serial	JTAG
Stand-alone FPGA applications (no external download host)	✓	✓	✓	Possible using XCFxxP Platform Flash, which optionally generates CCLK	Possible using XCFxxP Platform Flash, which optionally generates CCLK	
Uses low-cost, industry-standard Flash		✓	✓			
Supports optional MultiBoot, multi-configuration mode			✓			

Voltage Compatibility

Available SPI Flash PROMs use a single 3.3V supply voltage. All of the FPGA's SPI Flash interface signals are within I/O Bank 2. Consequently, the FPGA's VCCO_2 supply voltage must also be 3.3V to match the SPI Flash PROM.

Power-On Precautions if 3.3V Supply is Last in Sequence

Spartan-3E FPGAs have a built-in power-on reset (POR) circuit, as shown in Figure 66, page 102. The FPGA waits for its three power supplies — V_{CCINT}, V_{CCAUX}, and V_{CCO} to I/O Bank 2 (VCCO_2) — to reach their respective power-on thresholds before beginning the configuration process.

The SPI Flash PROM is powered by the same voltage supply feeding the FPGA's VCCO_2 voltage input, typically 3.3V. SPI Flash PROMs specify that they cannot be accessed until their V_{CC} supply reaches its minimum data sheet voltage, followed by an additional delay. For some devices, this additional delay is as little as 10 µs as shown in Table 56. For other vendors, this delay is as much as 20 ms.

Table 56: **Example Minimum Power-On to Select Times for Various SPI Flash PROMs**

Vendor	SPI Flash PROM Part Number	Data Sheet Minimum Time from V_{CC} min to Select = Low		
		Symbol	Value	Units
STMicroelectronics	M25Pxx	T_{VSL}	10	µs
Spansion	S25FLxxxA	t_{PU}	10	ms
NexFlash	NX25xx	T_{VSL}	10	µs
Macronix	MX25Lxxxx	t_{VSL}	10	µs
Silicon Storage Technology	SST25LFxx	$T_{PU-READ}$	10	µs
Programmable Microelectronics Corporation	Pm25LVxxx	T_{VCS}	50	µs
Atmel Corporation	AT45DBxxxD	t_{VCSL}	30	µs
	AT45DBxxxB		20	ms

In many systems, the 3.3V supply feeding the FPGA's VCCO_2 input is valid before the FPGA's other V_{CCINT} and V_{CCAUX} supplies, and consequently, there is no issue. However, if the 3.3V supply feeding the FPGA's VCCO_2 supply is last in the sequence, a potential race occurs between the FPGA and the SPI Flash PROM, as shown in Figure 55.

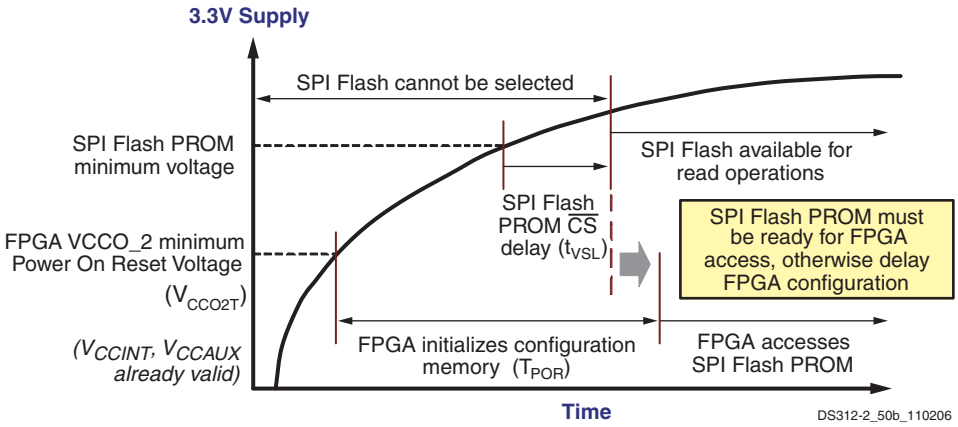

Figure 55: **SPI Flash PROM/FPGA Power-On Timing if 3.3V Supply is Last in Power-On Sequence**

If the FPGA's V_{CCINT} and V_{CCAUX} supplies are already valid, then the FPGA waits for VCCO_2 to reach its minimum threshold voltage before starting configuration. This threshold voltage is labeled as V_{CCO2T} in Table 74 of Module 3 and ranges from approximately 0.4V to 1.0V, substantially lower than the SPI Flash PROM's minimum voltage. Once all three FPGA supplies reach their respective Power On Reset (POR) thresholds, the FPGA starts the configuration process and begins initializing its internal configuration memory. Initialization requires approximately 1 ms (T_{POR}, minimum in Table 111 of Module 3, after which the FPGA de-asserts INIT_B, selects the SPI Flash PROM, and starts sending the appropriate read command. The SPI Flash PROM must be ready for

Readback

FPGA configuration data can be read back using either the Slave Parallel or JTAG mode. This function is disabled if the Bitstream Generator ***Security*** option is set to either ***Level1*** or ***Level2***.

Along with the configuration data, it is possible to read back the contents of all registers and distributed RAM.

To synchronously control when register values are captured for readback, use the CAPTURE_SPARTAN3 library primitive, which applies for both Spartan-3 and Spartan-3E FPGA families.

The Readback feature is available in most Spartan-3E FPGA product options, as indicated in Table 68. The Readback feature is not available in the XC3S1200E and XC3S1600E FPGAs when using the -4 speed grade in the Commercial temperature grade. Similarly, block RAM Readback support is not available in the -4 speed grade, Commercial temperature devices. If Readback is required in an XC3S1200E or XC3S1600E FPGA, or if block RAM Readback is required on any Spartan-3E FPGA, upgrade to either the Industrial temperature grade version or the -5 speed grade.

The Xilinx iMPACT programming software uses the Readback feature for its optional Verify and Readback operations. The Xilinx ChipScope™ software presently does not use Readback but may in future updates.

Table 68: **Readback Support in Spartan-3E FPGAs**

Temperature Range	Commercial		Industrial
Speed Grade	**-4**	**-5**	**-4**
Block RAM Readback			
All Spartan-3E FPGAs	No	Yes	Yes
General Readback (registers, distributed RAM)			
XC3S100E	Yes	Yes	Yes
XC3S250E	Yes	Yes	Yes
XC3S500E	Yes	Yes	Yes
XC3S1200E	No	Yes	Yes
XC3S1600E	No	Yes	Yes

Voltage Regulators

Various power supply manufacturers offer complete power solutions for Xilinx FPGAs including some with integrated three-rail regulators specifically designed for Spartan-3 and Spartan-3E FPGAs. The Xilinx Power Corner website provides links to vendor solution guides and Xilinx power estimation and analysis tools.

Power Distribution System (PDS) Design and Decoupling/Bypass Capacitors

Good power distribution system (PDS) design is important for all FPGA designs, but especially so for high performance applications, greater than 100 MHz. Proper design results in better overall performance, lower clock and DCM jitter, and a generally more robust system. Before designing the printed circuit board (PCB) for the FPGA design, please review XAPP623: *Power Distribution System (PDS) Design: Using Bypass/Decoupling Capacitors*.

Power-On Behavior

For additional power-on behavior information, including I/O behavior before and during configuration, refer to the "Sequence of Events" chapter in UG332.

Spartan-3E FPGAs have a built-in Power-On Reset (POR) circuit that monitors the three power rails required to successfully configure the FPGA. At power-up, the POR circuit holds the FPGA in a reset state until the V_{CCINT}, V_{CCAUX}, and V_{CCO} Bank 2 supplies reach their respective input threshold levels (see Table 74 in Module 3). After all three supplies reach their respective thresholds, the POR reset is released and the FPGA begins its configuration process.

Supply Sequencing

Because the three FPGA supply inputs must be valid to release the POR reset and can be supplied in any order, there are no FPGA-specific voltage sequencing requirements. Applying the FPGA's V_{CCAUX} supply before the V_{CCINT} supply uses the least I_{CCINT} current.

Although the FPGA has no specific voltage sequence requirements, be sure to consider any potential sequencing requirement of the configuration device attached to the FPGA, such as an SPI serial Flash PROM, a parallel NOR Flash PROM, or a microcontroller. For example, Flash PROMs have a minimum time requirement before the PROM can be selected and this must be considered if the 3.3V supply is the last in the sequence. See Power-On Precautions if 3.3V Supply is Last in Sequence for more details.

When all three supplies are valid, the minimum current required to power-on the FPGA equals the worst-case quiescent current, specified in Table 79. Spartan-3E FPGAs do not require Power-On Surge (POS) current to successfully configure.

Surplus I_{CCINT} if V_{CCINT} Applied before V_{CCAUX}

If the V_{CCINT} supply is applied before the V_{CCAUX} supply, the FPGA might draw a surplus I_{CCINT} current in addition to the I_{CCINT} quiescent current levels specified in Table 79, page 118. The momentary additional I_{CCINT} surplus current might be a few hundred milliamperes under nominal conditions, significantly less than the instantaneous current consumed by the bypass capacitors at power-on. However, the surplus current immediately disappears when the V_{CCAUX} supply is applied, and, in response, the FPGA's I_{CCINT} quiescent current demand drops to the levels specified in Table 79. The FPGA does not use or require the surplus current to successfully power-on and configure. If applying V_{CCINT} before V_{CCAUX}, ensure that the regulator does not have a foldback feature that could inadvertently shut down in the presence of the surplus current.

Configuration Data Retention, Brown-Out

The FPGA's configuration data is stored in robust CMOS configuration latches. The data in these latches is retained even when the voltages drop to the minimum levels necessary to preserve RAM contents, as specified in Table 76.

If, after configuration, the V_{CCAUX} or V_{CCINT} supply drops below its data retention voltage, the current device configuration must be cleared using one of the following methods:

- Force the V_{CCAUX} or V_{CCINT} supply voltage below the minimum Power On Reset (POR) voltage threshold (Table 74).
- Assert PROG_B Low.

The POR circuit does not monitor the VCCO_2 supply after configuration. Consequently, dropping the VCCO_2 voltage does not reset the device by triggering a Power-On Reset (POR) event.

No Internal Charge Pumps or Free-Running Oscillators

Some system applications are sensitive to sources of analog noise. Spartan-3E FPGA circuitry is fully static and does not employ internal charge pumps.

The CCLK configuration clock is active during the FPGA configuration process. After configuration completes, the CCLK oscillator is automatically disabled unless the Bitstream Generator (BitGen) option ***Persist=Yes***.

General Recommended Operating Conditions

Table 77: **General Recommended Operating Conditions**

Symbol	Description			Min	Nominal	Max	Units
T_J	Junction temperature	Commercial		0	–	85	°C
		Industrial		–40	–	100	°C
V_{CCINT}	Internal supply voltage			1.140	1.200	1.260	V
V_{CCO} (1)	Output driver supply voltage			1.100	-	3.465	V
V_{CCAUX}	Auxiliary supply voltage			2.375	2.500	2.625	V
V_{IN} (2,3)	Input voltage extremes to avoid turning on I/O protection diodes	I/O, Input-only, and Dual-Purpose pins (4)	IP or IO_#	–0.5	–	V_{CCO} + 0.5	V
			IO_Lxxy_# (5)	–0.5	–	V_{CCO} + 0.5	V
		Dedicated pins (6)		–0.5	–	V_{CCAUX} + 0.5	V
T_{IN}	Input signal transition time (7)			–	–	500	ns

Notes:

1. This V_{CCO} range spans the lowest and highest operating voltages for all supported I/O standards. Table 80 lists the recommended V_{CCO} range specific to each of the single-ended I/O standards, and Table 82 lists that specific to the differential standards.
2. Input voltages outside the recommended range require the I_{IK} input clamp diode rating is met and no more than 100 pins exceed the range simultaneously. Refer to Table 73.
3. See XAPP459: *Eliminating I/O Coupling Effects when Interfacing Large-Swing Single-Ended Signals to User I/O Pins on Spartan-3 Families*.
4. Each of the User I/O and Dual-Purpose pins is associated with one of the four banks' V_{CCO} rails. Meeting the V_{IN} limit ensures that the internal diode junctions that exist between these pins and their associated V_{CCO} and GND rails do not turn on. The absolute maximum rating is provided in Table 73.
5. For single-ended signals that are placed on a differential-capable I/O, V_{IN} of –0.2V to –0.5V is supported but can cause increased leakage between the two pins. See *Parasitic Leakage* in UG331, *Spartan-3 Generation FPGA User Guide*.
6. All Dedicated pins (PROG_B, DONE, TCK, TDI, TDO, and TMS) draw power from the V_{CCAUX} rail (2.5V). Meeting the V_{IN} max limit ensures that the internal diode junctions that exist between each of these pins and the V_{CCAUX} and GND rails do not turn on.
7. Measured between 10% and 90% V_{CCO}. Follow Signal Integrity recommendations.

Digital Clock Manager (DCM) Timing

For specification purposes, the DCM consists of three key components: the Delay-Locked Loop (DLL), the Digital Frequency Synthesizer (DFS), and the Phase Shifter (PS).

Aspects of DLL operation play a role in all DCM applications. All such applications inevitably use the CLKIN and the CLKFB inputs connected to either the CLK0 or the CLK2X feedback, respectively. Thus, specifications in the DLL tables (Table 104 and Table 105) apply to any application that only employs the DLL component. When the DFS and/or the PS components are used together with the DLL, then the specifications listed in the DFS and PS tables (Table 106 through Table 109) supersede any corresponding ones in the DLL tables. DLL specifications that do not change with the addition of DFS or PS functions are presented in Table 104 and Table 105.

Period jitter and cycle-cycle jitter are two of many different ways of specifying clock jitter. Both specifications describe statistical variation from a mean value.

Period jitter is the worst-case deviation from the ideal clock period over a collection of millions of samples. In a histogram of period jitter, the mean value is the clock period.

Cycle-cycle jitter is the worst-case difference in clock period between adjacent clock cycles in the collection of clock periods sampled. In a histogram of cycle-cycle jitter, the mean value is zero.

Spread Spectrum

DCMs accept typical spread spectrum clocks as long as they meet the input requirements. The DLL will track the frequency changes created by the spread spectrum clock to drive the global clocks to the FPGA logic. See XAPP469, *Spread-Spectrum Clocking Reception for Displays* for details.

Delay-Locked Loop (DLL)

Table 104: **Recommended Operating Conditions for the DLL**

Symbol		Description			Speed Grade				Units
					-5		-4		
					Min	Max	Min	Max	
Input Frequency Ranges									
F_{CLKIN}	CLKIN_FREQ_DLL	Frequency of the CLKIN clock input	Stepping 0	XC3S100E XC3S250E XC3S500E XC3S1600E	N/A	N/A	5(2)	90(3)	MHz
				XC3S1200E(3)				200(3)	MHz
			Stepping 1	All	5(2)	275(3)		240(3)	MHz
Input Pulse Requirements									
CLKIN_PULSE		CLKIN pulse width as a percentage of the CLKIN period	$F_{CLKIN} \leq 150$ MHz		40%	60%	40%	60%	-
			$F_{CLKIN} > 150$ MHz		45%	55%	45%	55%	-
Input Clock Jitter Tolerance and Delay Path Variation(4)									
CLKIN_CYC_JITT_DLL_LF		Cycle-to-cycle jitter at the CLKIN input	$F_{CLKIN} \leq 150$ MHz		-	±300	-	±300	ps
CLKIN_CYC_JITT_DLL_HF			$F_{CLKIN} > 150$ MHz		-	±150	-	±150	ps
CLKIN_PER_JITT_DLL		Period jitter at the CLKIN input			-	±1	-	±1	ns
CLKFB_DELAY_VAR_EXT		Allowable variation of off-chip feedback delay from the DCM output to the CLKFB input			-	±1	-	±1	ns

Notes:

1. DLL specifications apply when any of the DLL outputs (CLK0, CLK90, CLK180, CLK270, CLK2X, CLK2X180, or CLKDV) are in use.
2. The DFS, when operating independently of the DLL, supports lower FCLKIN frequencies. See Table 106.
3. To support double the maximum effective FCLKIN limit, set the CLKIN_DIVIDE_BY_2 attribute to TRUE. This attribute divides the incoming clock frequency by two as it enters the DCM. The CLK2X output reproduces the clock frequency provided on the CLKIN input.
4. CLKIN input jitter beyond these limits might cause the DCM to lose lock.

Phase Shifter (PS)

Table 108: **Recommended Operating Conditions for the PS in Variable Phase Mode**

Symbol	Description	Speed Grade				Units
		-5		-4		
		Min	Max	Min	Max	
Operating Frequency Ranges						
PSCLK_FREQ (F_{PSCLK})	Frequency for the PSCLK input	1	167	1	167	MHz
Input Pulse Requirements						
PSCLK_PULSE	PSCLK pulse width as a percentage of the PSCLK period	40%	60%	40%	60%	-

Table 109: **Switching Characteristics for the PS in Variable Phase Mode**

Symbol	Description	Equation		Units
Phase Shifting Range				
MAX_STEPS(2)	Maximum allowed number of DCM_DELAY_STEP steps for a given CLKIN clock period, where T = CLKIN clock period in ns. If using CLKIN_DIVIDE_BY_2 = TRUE, double the effective clock period.(3)	CLKIN < 60 MHz	±[INTEGER(10 • (T_{CLKIN} – 3 ns))]	steps
		CLKIN ≥ 60 MHz	±[INTEGER(15 • (T_{CLKIN} – 3 ns))]	steps
FINE_SHIFT_RANGE_MIN	Minimum guaranteed delay for variable phase shifting	±[MAX_STEPS • DCM_DELAY_STEP_MIN]		ns
FINE_SHIFT_RANGE_MAX	Maximum guaranteed delay for variable phase shifting	±[MAX_STEPS • DCM_DELAY_STEP_MAX]		ns

Notes:

1. The numbers in this table are based on the operating conditions set forth in Table 77 and Table 108.
2. The maximum variable phase shift range, MAX_STEPS, is only valid when the DCM is has no initial fixed phase shifting, i.e., the PHASE_SHIFT attribute is set to 0.
3. The DCM_DELAY_STEP values are provided at the bottom of Table 105.

Miscellaneous DCM Timing

Table 110: **Miscellaneous DCM Timing**

Symbol	Description	Min	Max	Units
DCM_RST_PW_MIN(1)	Minimum duration of a RST pulse width	3	-	CLKIN cycles
DCM_RST_PW_MAX(2)	Maximum duration of a RST pulse width	N/A	N/A	seconds
DCM_CONFIG_LAG_TIME(3)	Maximum duration from V_{CCINT} applied to FPGA configuration successfully completed (DONE pin goes High) and clocks applied to DCM DLL	N/A	N/A	minutes

Notes:

1. This limit only applies to applications that use the DCM DLL outputs (CLK0, CLK90, CLK180, CLK270, CLK2X, CLK2X180, and CLKDV). The DCM DFS outputs (CLKFX, CLKFX180) are unaffected.
2. This specification is equivalent to the Virtex-4 DCM_RESET specfication.This specification does not apply for Spartan-3E FPGAs.
3. This specification is equivalent to the Virtex-4 TCONFIG specification. This specification does not apply for Spartan-3E FPGAs.

Table 117: **Timing for the Slave Parallel Configuration Mode *(Cont'd)***

Symbol	Description			All Speed Grades		Units
				Min	Max	
Clock Timing						
T_{CCH}	The High pulse width at the CCLK input pin			5	-	ns
T_{CCL}	The Low pulse width at the CCLK input pin			5	-	ns
F_{CCPAR}	Frequency of the clock signal at the CCLK input pin	No bitstream compression	Not using the BUSY pin(2)	0	50	MHz
			Using the BUSY pin	0	66	MHz
		With bitstream compression		0	20	MHz

Notes:

1. The numbers in this table are based on the operating conditions set forth in Table 77.
2. In the Slave Parallel mode, it is necessary to use the BUSY pin when the CCLK frequency exceeds this maximum specification.
3. Some Xilinx documents refer to Parallel modes as "SelectMAP" modes.

Revision History

The following table shows the revision history for this document.

Date	Version	Revision
03/01/05	1.0	Initial Xilinx release.
11/23/05	2.0	Added AC timing information and additional DC specifications.
03/22/06	3.0	Upgraded data sheet status to Preliminary. Finalized production timing parameters. All speed grades for all Spartan-3E FPGAs are now Production status using the v1.21 speed files, as shown in Table 84. Expanded description in Note 2, Table 78. Updated pin-to-pin and clock-to-output timing based on final characterization, shown in Table 86. Updated system-synchronous input setup and hold times based on final characterization, shown in Table 87 and Table 88. Updated other I/O timing in Table 90. Provided input and output adjustments for LVPECL_25, DIFF_SSTL and DIFF_HSTL I/O standards that supersede the v1.21 speed file values, in Table 91 and Table 94. Reduced I/O three-state and set/reset delays in Table 93. Added XC3S100E FPGA in CP132 package to Table 96. Increased T_{AS} slice flip-flop timing by 100 ps in Table 98. Updated distributed RAM timing in Table 99 and SRL16 timing in Table 100. Updated global clock timing, removed left/right clock buffer limits in Table 101. Updated block RAM timing in Table 103. Added DCM parameters for remainder of Step 0 device; added improved Step 1 DCM performance to Table 104, Table 105, Table 106, and Table 107. Added minimum INIT_B pulse width specification, T_{INIT}, in Table 111. Increased data hold time for Slave Parallel mode to 1.0 ns (T_{SMCCD}) in Table 117. Improved the DCM performance for the XC3S1200E, Stepping 0 in Table 104, Table 105, Table 106, and Table 107. Corrected links in Table 118 and Table 120. Added MultiBoot timing specifications to Table 122.
04/07/06	3.1	Improved SSO limits for LVDS_25, MINI_LVDS_25, and RSDS_25 I/O standards in the QFP packages (Table 97). Removed potentially confusing Note 2 from Table 78.
05/19/06	3.2	Clarified that 100 mV of hysteresis applies to LVCMOS33 and LVCMOS25 I/O standards (Note 4, Table 80). Other minor edits.
05/30/06	3.2.1	Corrected various typos and incorrect links.
11/09/06	3.4	Improved absolute maximum voltage specifications in Table 73, providing additional overshoot allowance. Widened the recommended voltage range for PCI and PCI-X standards in Table 80. Clarified Note 2, Table 83. Improved various timing specifications for v1.26 speed file. Added Table 85 to summarize the history of speed file releases after which time all devices became Production status. Added absolute minimum values for Table 86, Table 92, and Table 93. Updated pin-to-pin setup and hold timing based on default IFD_DELAY_VALUE settings in Table 87, Table 88, and Table 90. Added Table 89 about source-synchronous input capture sample window. Promoted Module 3 to Production status. Synchronized all modules to v3.4.
03/16/07	3.5	Based on extensive 90 nm production data, improved (reduced) the maximum quiescent current limits for the I_{CCINTQ}, I_{CCAUXQ}, and I_{CCOQ} specifications in Table 79 by an average of 50%.
05/29/07	3.6	Added note to Table 74 and Table 75 regarding HSWAP in step 0 devices. Updated t_{RPW_CLB} in Table 98 to match value in speed file. Improved CLKOUT_FREQ_CLK90 to 200 MHz for Stepping 1 in Table 105.
04/18/08	3.7	Clarified that Stepping 0 was offered only for -4C and removed Stepping 0 -5 specifications. Added reference to XAPP459 in Table 73 and Table 77. Improved recommended max V_{CCO} to 3.465V (3.3V + 5%) in Table 77. Removed minimum input capacitance from Table 78. Updated Recommended Operating Conditions for LVCMOS and PCI I/O standards in Table 80. Removed Absolute Minimums from Table 86, Table 92 and Table 93 and added footnote recommending use of Timing Analyzer for minimum values. Updated T_{PSFD} and T_{PHFD} in Table 87 to match current speed file. Update T_{RPW_IOB} in Table 88 to match current speed file and CLB equivalent spec. Added XC3S500E VQG100 to Table 96. Replaced $T_{MULCKID}$ with T_{MSCKD} for A, B, and P registers in Table 102. Updated CLKOUT_PER_JITT_FX in Table 107. Updated MAX_STEPS equation in Table 109. Updated Figure 77 and Table 120 to correct CCLK active edge. Updated links.

VQ100: 100-lead Very-thin Quad Flat Package

The XC3S100E, XC3S250E, and the XC3S500E devices are available in the 100-lead very-thin quad flat package, VQ100. All devices share a common footprint for this package as shown in Table 131 and Figure 80.

Table 131 lists all the package pins. They are sorted by bank number and then by pin name. Pins that form a differential I/O pair appear together in the table. The table also shows the pin number for each pin and the pin type, as defined earlier.

The VQ100 package does not support the Byte-wide Peripheral Interface (BPI) configuration mode. Consequently, the VQ100 footprint has fewer DUAL-type pins than other packages.

An electronic version of this package pinout table and footprint diagram is available for download from the Xilinx web site at:

http://www.xilinx.com/support/documentation/data_sheets/s3e_pin.zip

Pinout Table

Table 131 shows the pinout for production Spartan-3E FPGAs in the VQ100 package.

Table 131: **VQ100 Package Pinout**

Bank	XC3S100E XC3S250E XC3S500E Pin Name	VQ100 Pin Number	Type
0	IO	P92	I/O
0	IO_L01N_0	P79	I/O
0	IO_L01P_0	P78	I/O
0	IO_L02N_0/GCLK5	P84	GCLK
0	IO_L02P_0/GCLK4	P83	GCLK
0	IO_L03N_0/GCLK7	P86	GCLK
0	IO_L03P_0/GCLK6	P85	GCLK
0	IO_L05N_0/GCLK11	P91	GCLK
0	IO_L05P_0/GCLK10	P90	GCLK
0	IO_L06N_0/VREF_0	P95	VREF
0	IO_L06P_0	P94	I/O
0	IO_L07N_0/HSWAP	P99	DUAL
0	IO_L07P_0	P98	I/O
0	IP_L04N_0/GCLK9	P89	GCLK
0	IP_L04P_0/GCLK8	P88	GCLK
0	VCCO_0	P82	VCCO
0	VCCO_0	P97	VCCO
1	IO_L01N_1	P54	I/O
1	IO_L01P_1	P53	I/O
1	IO_L02N_1	P58	I/O
1	IO_L02P_1	P57	I/O
1	IO_L03N_1/RHCLK1	P61	RHCLK
1	IO_L03P_1/RHCLK0	P60	RHCLK
1	IO_L04N_1/RHCLK3	P63	RHCLK
1	IO_L04P_1/RHCLK2	P62	RHCLK
1	IO_L05N_1/RHCLK5	P66	RHCLK
1	IO_L05P_1/RHCLK4	P65	RHCLK
1	IO_L06N_1/RHCLK7	P68	RHCLK
1	IO_L06P_1/RHCLK6	P67	RHCLK
1	IO_L07N_1	P71	I/O
1	IO_L07P_1	P70	I/O
1	IP/VREF_1	P69	VREF
1	VCCO_1	P55	VCCO
1	VCCO_1	P73	VCCO
2	IO/D5	P34	DUAL
2	IO/M1	P42	DUAL
2	IO_L01N_2/INIT_B	P25	DUAL
2	IO_L01P_2/CSO_B	P24	DUAL
2	IO_L02N_2/MOSI/CSI_B	P27	DUAL
2	IO_L02P_2/DOUT/BUSY	P26	DUAL
2	IO_L03N_2/D6/GCLK13	P33	DUAL/GCLK
2	IO_L03P_2/D7/GCLK12	P32	DUAL/GCLK
2	IO_L04N_2/D3/GCLK15	P36	DUAL/GCLK
2	IO_L04P_2/D4/GCLK14	P35	DUAL/GCLK
2	IO_L06N_2/D1/GCLK3	P41	DUAL/GCLK
2	IO_L06P_2/D2/GCLK2	P40	DUAL/GCLK
2	IO_L07N_2/DIN/D0	P44	DUAL
2	IO_L07P_2/M0	P43	DUAL
2	IO_L08N_2/VS1	P48	DUAL
2	IO_L08P_2/VS2	P47	DUAL
2	IO_L09N_2/CCLK	P50	DUAL
2	IO_L09P_2/VS0	P49	DUAL
2	IP/VREF_2	P30	VREF
2	IP_L05N_2/M2/GCLK1	P39	DUAL/GCLK
2	IP_L05P_2/RDWR_B/ GCLK0	P38	DUAL/GCLK
2	VCCO_2	P31	VCCO
2	VCCO_2	P45	VCCO
3	IO_L01N_3	P3	I/O
3	IO_L01P_3	P2	I/O
3	IO_L02N_3/VREF_3	P5	VREF

Table 137: **TQ144 Package Pinout *(Cont'd)***

Bank	XC3S100E Pin Name	XC3S250E Pin Name	TQ144 Pin	Type
2	IP	IP	P38	INPUT
2	IP	IP	P41	INPUT
2	IP	IP	P69	INPUT
2	IP_L03N_2/VREF_2	IP_L03N_2/VREF_2	P48	VREF
2	IP_L03P_2	IP_L03P_2	P47	INPUT
2	IP_L06N_2/M2/GCLK1	IP_L06N_2/M2/GCLK1	P57	DUAL/GCLK
2	IP_L06P_2/RDWR_B/GCLK0	IP_L06P_2/RDWR_B/GCLK0	P56	DUAL/GCLK
2	VCCO_2	VCCO_2	P42	VCCO
2	VCCO_2	VCCO_2	P49	VCCO
2	VCCO_2	VCCO_2	P64	VCCO
3	IP/VREF_3	IO/VREF_3	P31	***100E:*** VREF(INPUT) ***250E:*** VREF(I/O)
3	IO_L01N_3	IO_L01N_3	P3	I/O
3	IO_L01P_3	IO_L01P_3	P2	I/O
3	IO_L02N_3/VREF_3	IO_L02N_3/VREF_3	P5	VREF
3	IO_L02P_3	IO_L02P_3	P4	I/O
3	IO_L03N_3	IO_L03N_3	P8	I/O
3	IO_L03P_3	IO_L03P_3	P7	I/O
3	IO_L04N_3/LHCLK1	IO_L04N_3/LHCLK1	P15	LHCLK
3	IO_L04P_3/LHCLK0	IO_L04P_3/LHCLK0	P14	LHCLK
3	IO_L05N_3/LHCLK3/IRDY2	IO_L05N_3/LHCLK3	P17	LHCLK
3	IO_L05P_3/LHCLK2	IO_L05P_3/LHCLK2	P16	LHCLK
3	IO_L06N_3/LHCLK5	IO_L06N_3/LHCLK5	P21	LHCLK
3	IO_L06P_3/LHCLK4/TRDY2	IO_L06P_3/LHCLK4	P20	LHCLK
3	IO_L07N_3/LHCLK7	IO_L07N_3/LHCLK7	P23	LHCLK
3	IO_L07P_3/LHCLK6	IO_L07P_3/LHCLK6	P22	LHCLK
3	IO_L08N_3	IO_L08N_3	P26	I/O
3	IO_L08P_3	IO_L08P_3	P25	I/O
3	IO_L09N_3	IO_L09N_3	P33	I/O
3	IO_L09P_3	IO_L09P_3	P32	I/O
3	IO_L10N_3	IO_L10N_3	P35	I/O
3	IO_L10P_3	IO_L10P_3	P34	I/O
3	IP	IP	P6	INPUT
3	IO	IP	P10	***100E:*** I/O ***250E:*** INPUT
3	IP	IP	P18	INPUT
3	IP	IP	P24	INPUT
3	IO	IP	P29	***100E:*** I/O ***250E:*** INPUT
3	IP	IP	P36	INPUT
3	IP/VREF_3	IP/VREF_3	P12	VREF
3	VCCO_3	VCCO_3	P13	VCCO
3	VCCO_3	VCCO_3	P28	VCCO
GND	GND	GND	P11	GND
GND	GND	GND	P19	GND

Table 143: **FT256 Package Pinout *(Cont'd)***

Bank	XC3S250E Pin Name	XC3S500E Pin Name	XC3S1200E Pin Name	FT256 Ball	Type
3	IP	IP	IP	N3	INPUT
3	IP/VREF_3	IP/VREF_3	IP/VREF_3	G1	VREF
3	IO/VREF_3	IO/VREF_3	IP/VREF_3	N2	***250E:*** VREF(I/O) ***500E:*** VREF(I/O) ***1200E:*** VREF(INPUT)
3	VCCO_3	VCCO_3	VCCO_3	E2	VCCO
3	VCCO_3	VCCO_3	VCCO_3	G6	VCCO
3	VCCO_3	VCCO_3	VCCO_3	K6	VCCO
3	VCCO_3	VCCO_3	VCCO_3	M2	VCCO
GND	GND	GND	GND	A1	GND
GND	GND	GND	GND	A16	GND
GND	GND	GND	GND	B9	GND
GND	GND	GND	GND	F6	GND
GND	GND	GND	GND	F11	GND
GND	GND	GND	GND	G7	GND
GND	GND	GND	GND	G8	GND
GND	GND	GND	GND	G9	GND
GND	GND	GND	GND	G10	GND
GND	GND	GND	GND	H2	GND
GND	GND	GND	GND	H7	GND
GND	GND	GND	GND	H8	GND
GND	GND	GND	GND	H9	GND
GND	GND	GND	GND	H10	GND
GND	GND	GND	GND	J7	GND
GND	GND	GND	GND	J8	GND
GND	GND	GND	GND	J9	GND
GND	GND	GND	GND	J10	GND
GND	GND	GND	GND	J15	GND
GND	GND	GND	GND	K7	GND
GND	GND	GND	GND	K8	GND
GND	GND	GND	GND	K9	GND
GND	GND	GND	GND	K10	GND
GND	GND	GND	GND	L6	GND
GND	GND	GND	GND	L11	GND
GND	GND	GND	GND	R8	GND
GND	GND	GND	GND	T1	GND
GND	GND	GND	GND	T16	GND
VCCAUX	DONE	DONE	DONE	T15	CONFIG
VCCAUX	PROG_B	PROG_B	PROG_B	D3	CONFIG
VCCAUX	TCK	TCK	TCK	A15	JTAG
VCCAUX	TDI	TDI	TDI	A2	JTAG
VCCAUX	TDO	TDO	TDO	C14	JTAG
VCCAUX	TMS	TMS	TMS	B15	JTAG
VCCAUX	VCCAUX	VCCAUX	VCCAUX	A6	VCCAUX

Table 143: **FT256 Package Pinout *(Cont'd)***

Bank	XC3S250E Pin Name	XC3S500E Pin Name	XC3S1200E Pin Name	FT256 Ball	Type
VCCAUX	VCCAUX	VCCAUX	VCCAUX	A11	VCCAUX
VCCAUX	VCCAUX	VCCAUX	VCCAUX	F1	VCCAUX
VCCAUX	VCCAUX	VCCAUX	VCCAUX	F16	VCCAUX
VCCAUX	VCCAUX	VCCAUX	VCCAUX	L1	VCCAUX
VCCAUX	VCCAUX	VCCAUX	VCCAUX	L16	VCCAUX
VCCAUX	VCCAUX	VCCAUX	VCCAUX	T6	VCCAUX
VCCAUX	VCCAUX	VCCAUX	VCCAUX	T11	VCCAUX
VCCINT	VCCINT	VCCINT	VCCINT	D4	VCCINT
VCCINT	VCCINT	VCCINT	VCCINT	D13	VCCINT
VCCINT	VCCINT	VCCINT	VCCINT	E5	VCCINT
VCCINT	VCCINT	VCCINT	VCCINT	E12	VCCINT
VCCINT	VCCINT	VCCINT	VCCINT	M5	VCCINT
VCCINT	VCCINT	VCCINT	VCCINT	M12	VCCINT
VCCINT	VCCINT	VCCINT	VCCINT	N4	VCCINT
VCCINT	VCCINT	VCCINT	VCCINT	N13	VCCINT

Table 148: **FG320 Package Pinout *(Cont'd)***

Bank	XC3S500E Pin Name	XC3S1200E Pin Name	XC3S1600E Pin Name	FG320 Ball	Type
3	N.C. (◆)	IO_L04N_3	IO_L04N_3	E3	**500E:** N.C. **1200E:** I/O **1600E:** I/O
3	N.C. (◆)	IO_L04P_3	IO_L04P_3	E4	**500E:** N.C. **1200E:** I/O **1600E:** I/O
3	IO_L05N_3	IO_L05N_3	IO_L05N_3	F2	I/O
3	IO_L05P_3	IO_L05P_3	IO_L05P_3	F1	I/O
3	IO_L06N_3/VREF_3	IO_L06N_3/VREF_3	IO_L06N_3/VREF_3	G4	VREF
3	IO_L06P_3	IO_L06P_3	IO_L06P_3	G3	I/O
3	IO_L07N_3	IO_L07N_3	IO_L07N_3	G5	I/O
3	IO_L07P_3	IO_L07P_3	IO_L07P_3	G6	I/O
3	IO_L08N_3	IO_L08N_3	IO_L08N_3	H5	I/O
3	IO_L08P_3	IO_L08P_3	IO_L08P_3	H6	I/O
3	IO_L09N_3	IO_L09N_3	IO_L09N_3	H3	I/O
3	IO_L09P_3	IO_L09P_3	IO_L09P_3	H4	I/O
3	IO_L10N_3	IO_L10N_3	IO_L10N_3	H1	I/O
3	IO_L10P_3	IO_L10P_3	IO_L10P_3	H2	I/O
3	IO_L11N_3/LHCLK1	IO_L11N_3/LHCLK1	IO_L11N_3/LHCLK1	J4	LHCLK
3	IO_L11P_3/LHCLK0	IO_L11P_3/LHCLK0	IO_L11P_3/LHCLK0	J5	LHCLK
3	IO_L12N_3/LHCLK3/ IRDY2	IO_L12N_3/LHCLK3/ IRDY2	IO_L12N_3/LHCLK3/ IRDY2	J2	LHCLK
3	IO_L12P_3/LHCLK2	IO_L12P_3/LHCLK2	IO_L12P_3/LHCLK2	J1	LHCLK
3	IO_L13N_3/LHCLK5	IO_L13N_3/LHCLK5	IO_L13N_3/LHCLK5	K4	LHCLK
3	IO_L13P_3/LHCLK4/ TRDY2	IO_L13P_3/LHCLK4/ TRDY2	IO_L13P_3/LHCLK4/ TRDY2	K3	LHCLK
3	IO_L14N_3/LHCLK7	IO_L14N_3/LHCLK7	IO_L14N_3/LHCLK7	K5	LHCLK
3	IO_L14P_3/LHCLK6	IO_L14P_3/LHCLK6	IO_L14P_3/LHCLK6	K6	LHCLK
3	IO_L15N_3	IO_L15N_3	IO_L15N_3	L2	I/O
3	IO_L15P_3	IO_L15P_3	IO_L15P_3	L1	I/O
3	IO_L16N_3	IO_L16N_3	IO_L16N_3	L4	I/O
3	IO_L16P_3	IO_L16P_3	IO_L16P_3	L3	I/O
3	IO_L17N_3/VREF_3	IO_L17N_3/VREF_3	IO_L17N_3/VREF_3	L5	VREF
3	IO_L17P_3	IO_L17P_3	IO_L17P_3	L6	I/O
3	IO_L18N_3	IO_L18N_3	IO_L18N_3	M3	I/O
3	IO_L18P_3	IO_L18P_3	IO_L18P_3	M4	I/O
3	IO_L19N_3	IO_L19N_3	IO_L19N_3	M6	I/O
3	IO_L19P_3	IO_L19P_3	IO_L19P_3	M5	I/O
3	IO_L20N_3	IO_L20N_3	IO_L20N_3	N5	I/O
3	IO_L20P_3	IO_L20P_3	IO_L20P_3	N4	I/O
3	IO_L21N_3	IO_L21N_3	IO_L21N_3	P1	I/O
3	IO_L21P_3	IO_L21P_3	IO_L21P_3	P2	I/O
3	N.C. (◆)	IO_L22N_3	IO_L22N_3	P4	**500E:** N.C. **1200E:** I/O **1600E:** I/O