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### Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

### Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

#### Details

Product Status	Obsolete
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	34
Number of Gates	30000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	-20°C ~ 85°C (TJ)
Package / Case	48-VFQFN Exposed Pad
Supplier Device Package	48-QFN (6x6)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/microchip-technology/a3pn030-z2qng48">https://www.e-xfl.com/product-detail/microchip-technology/a3pn030-z2qng48</a>

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## Related Documents

### User's Guides

*Designer User's Guide*

[http://www.microsemi.com/soc/documents/designer\\_ug.pdf](http://www.microsemi.com/soc/documents/designer_ug.pdf)

## List of Changes

The following table lists critical changes that were made in each revision of the chapter.

Date	Changes	Page
August 2012	The "I/O State of Newly Shipped Devices" section is new (SAR 39542).	14
July 2010	This chapter is no longer published separately with its own part number and version but is now part of several FPGA fabric user's guides.	N/A
v1.4 (December 2008)	IGLOO nano and ProASIC3 nano devices were added to Table 1-1 • Flash-Based FPGAs.	10
	Figure 1-2 • IGLOO and ProASIC3 nano Device Architecture Overview with Two I/O Banks (applies to 10 k and 30 k device densities, excluding IGLOO PLUS devices) through Figure 1-5 • IGLOO, IGLOO nano, ProASIC3 nano, and ProASIC3/L Device Architecture Overview with Four I/O Banks (AGL600 device is shown) are new.	11, 12
	Table 1-4 • IGLOO nano and ProASIC3 nano Array Coordinates is new.	17
v1.3 (October 2008)	The title of this document was changed from "Core Architecture of IGLOO and ProASIC3 Devices" to "FPGA Array Architecture in Low Power Flash Devices."	9
	The "FPGA Array Architecture Support" section was revised to include new families and make the information more concise.	10
	Table 1-2 • IGLOO and ProASIC3 Array Coordinates was updated to include Military ProASIC3/EL and RT ProASIC3 devices.	16
v1.2 (June 2008)	The following changes were made to the family descriptions in Table 1-1 • Flash-Based FPGAs: <ul style="list-style-type: none"> <li>ProASIC3L was updated to include 1.5 V.</li> <li>The number of PLLs for ProASIC3E was changed from five to six.</li> </ul>	10
v1.1 (March 2008)	Table 1-1 • Flash-Based FPGAs and the accompanying text was updated to include the IGLOO PLUS family. The "IGLOO Terminology" section and "Device Overview" section are new.	10
	The "Device Overview" section was updated to note that 15 k devices do not support SRAM or FIFO.	11
	Figure 1-6 • IGLOO PLUS Device Architecture Overview with Four I/O Banks is new.	13
	Table 1-2 • IGLOO and ProASIC3 Array Coordinates was updated to add A3P015 and AGL015.	16
	Table 1-3 • IGLOO PLUS Array Coordinates is new.	16

You can control the maximum number of shared instances allowed for the legalization to take place using the Compile Option dialog box shown in Figure 3-17. Refer to Libero SoC / Designer online help for details on the Compile Option dialog box. A large number of shared instances most likely indicates a floorplanning problem that you should address.

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**Figure 3-17 • Shared Instances in the Compile Option Dialog Box**

## Designer Flow for Global Assignment

To achieve the desired result, pay special attention to global management during synthesis and place-and-route. The current Synplify tool does not insert more than six global buffers in the netlist by default. Thus, the default flow will not assign any signal to the quadrant global network. However, you can use attributes in Synplify and increase the default global macro assignment in the netlist. Designer v6.2 supports automatic quadrant global assignment, which was not available in Designer v6.1. Layout will make the choice to assign the correct signals to global. However, you can also utilize PDC and perform manual global assignment to overwrite any automatic assignment. The following step-by-step suggestions guide you in the layout of your design and help you improve timing in Designer:

1. Run Compile and check the Compile report. The Compile report has global information in the "Device Utilization" section that describes the number of chip and quadrant signals in the design. A "Net Report" section describes chip global nets, quadrant global nets, local clock nets, a list of nets listed by fanout, and net candidates for local clock assignment. Review this information. Note that YB or YC are counted as global only when they are used in isolation; if you use YB only and not GLB, this net is not shown in the global/quadrant nets report. Instead, it appears in the Global Utilization report.
2. If some signals have a very high fanout and are candidates for global promotion, promote those signals to global using the compile options or PDC commands. Figure 3-18 on page 54 shows the Globals Management section of the compile options. Select **Promote regular nets whose fanout is greater than** and enter a reasonable value for fanouts.



## CCC Support in Microsemi's Flash Devices

The flash FPGAs listed in Table 4-1 support the CCC feature and the functions described in this document.

**Table 4-1 • Flash-Based FPGAs**

Series	Family*	Description
IGLOO	IGLOO	Ultra-low power 1.2 V to 1.5 V FPGAs with Flash*Freeze technology
	IGLOOe	Higher density IGLOO FPGAs with six PLLs and additional I/O standards
	IGLOO PLUS	IGLOO FPGAs with enhanced I/O capabilities
	IGLOO nano	The industry's lowest-power, smallest-size solution
ProASIC3	ProASIC3	Low power, high-performance 1.5 V FPGAs
	ProASIC3E	Higher density ProASIC3 FPGAs with six PLLs and additional I/O standards
	ProASIC3 nano	Lowest-cost solution with enhanced I/O capabilities
	ProASIC3L	ProASIC3 FPGAs supporting 1.2 V to 1.5 V with Flash*Freeze technology
	RT ProASIC3	Radiation-tolerant RT3PE600L and RT3PE3000L
	Military ProASIC3/EL	Military temperature A3PE600L, A3P1000, and A3PE3000L
	Automotive ProASIC3	ProASIC3 FPGAs qualified for automotive applications
Fusion	Fusion	Mixed signal FPGA integrating ProASIC3 FPGA fabric, programmable analog block, support for ARM® Cortex™-M1 soft processors, and flash memory into a monolithic device

*Note:* \*The device names link to the appropriate datasheet, including product brief, DC and switching characteristics, and packaging information.

### **IGLOO Terminology**

In documentation, the terms IGLOO series and IGLOO devices refer to all of the IGLOO devices as listed in Table 4-1. Where the information applies to only one product line or limited devices, these exclusions will be explicitly stated.

### **ProASIC3 Terminology**

In documentation, the terms ProASIC3 series and ProASIC3 devices refer to all of the ProASIC3 devices as listed in Table 4-1. Where the information applies to only one product line or limited devices, these exclusions will be explicitly stated.

To further understand the differences between the IGLOO and ProASIC3 devices, refer to the *Industry's Lowest Power FPGAs Portfolio*.

YB and YC are identical to GLB and GLC, respectively, with the exception of a higher selectable final output delay. The SmartGen PLL Wizard will configure these outputs according to user specifications and can enable these signals with or without the enabling of Global Output Clocks.

The above signals can be enabled in the following output groupings in both internal and external feedback configurations of the static PLL:

- One output – GLA only
- Two outputs – GLA + (GLB and/or YB)
- Three outputs – GLA + (GLB and/or YB) + (GLC and/or YC)

## PLL Macro Block Diagram

As illustrated, the PLL supports three distinct output frequencies from a given input clock. Two of these (GLB and GLC) can be routed to the B and C global network access, respectively, and/or routed to the device core (YB and YC).

There are five delay elements to support phase control on all five outputs (GLA, GLB, GLC, YB, and YC).

There are delay elements in the feedback loop that can be used to advance the clock relative to the reference clock.

The PLL macro reference clock can be driven in the following ways:

1. By an INBUF\* macro to create a composite macro, where the I/O macro drives the global buffer (with programmable delay) using a hardwired connection. In this case, the I/O must be placed in one of the dedicated global I/O locations.
2. Directly from the FPGA core.
3. From an I/O that is routed through the FPGA regular routing fabric. In this case, users must instantiate a special macro, PLLINT, to differentiate from the hardwired I/O connection described earlier.

During power-up, the PLL outputs will toggle around the maximum frequency of the voltage-controlled oscillator (VCO) gear selected. Toggle frequencies can range from 40 MHz to 250 MHz. This will continue as long as the clock input (CLKA) is constant (HIGH or LOW). This can be prevented by LOW assertion of the POWERDOWN signal.

The visual PLL configuration in SmartGen, a component of the Libero SoC and Designer tools, will derive the necessary internal divider ratios based on the input frequency and desired output frequencies selected by the user.

Primary Clock Output Delay from CLKA -3.020

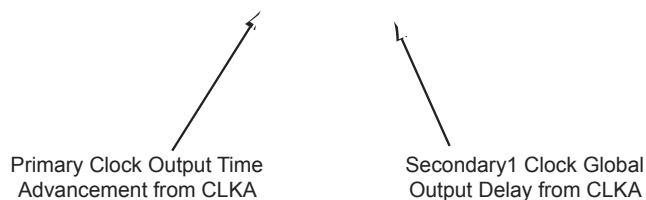
Secondary1 Clock frequency 40.000

Secondary1 Clock Phase Shift 0.000

Secondary1 Clock Global Output Delay from CLKA 2.515

Next, perform simulation in ModelSim to verify the correct delays. Figure 4-30 shows the simulation results. The delay values match those reported in the SmartGen PLL Wizard.

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**Figure 4-30 • ModelSim Simulation Results**

The timing can also be analyzed using SmartTime in Designer. The user should import the synthesized netlist to Designer, perform Compile and Layout, and then invoke SmartTime. Go to **Tools > Options** and change the maximum delay operating conditions to **Typical Case**. Then expand the Clock-to-Out paths of GLA and GLB and the individual components of the path delays are shown. The path of GLA is shown in Figure 4-31 on page 107 displaying the same delay value.

---

**Figure 4-31 • Static Timing Analysis Using SmartTime**

**Place-and-Route Stage Considerations**

Several considerations must be noted to properly place the CCC macros for layout.

For CCCs with clock inputs configured with the Hardwired I/O–Driven option:

- PLL macros must have the clock input pad coming from one of the GmA\* locations.
- CLKDLY macros must have the clock input pad coming from one of the Global I/Os.

If a PLL with a Hardwired I/O input is used at a CCC location and a Hardwired I/O–Driven CLKDLY macro is used at the same CCC location, the clock input of the CLKDLY macro must be chosen from one of the GmB\* or GmC\* pin locations. If the PLL is not used or is an External I/O–Driven or Core Logic–Driven PLL, the clock input of the CLKDLY macro can be sourced from the GmA\*, GmB\*, or GmC\* pin locations.

For CCCs with clock inputs configured with the External I/O–Driven option, the clock input pad can be assigned to any regular I/O location (IO\*\*\*\*\* pins). Note that since global I/O pins can also be used as regular I/Os, regardless of CCC function (CLKDLY or PLL), clock inputs can also be placed in any of these I/O locations.

By default, the Designer layout engine will place global nets in the design at one of the six chip globals. When the number of globals in the design is greater than six, the Designer layout engine will automatically assign additional globals to the quadrant global networks of the low power flash devices. If the user wishes to decide which global signals should be assigned to chip globals (six available) and which to the quadrant globals (three per quadrant for a total of 12 available), the assignment can be achieved with PinEditor, ChipPlanner, or by importing a placement constraint file. Layout will fail if the

## Conclusion

The Fusion, IGLOO, and ProASIC3 families are the only FPGAs that offer on-chip FlashROM support. This document presents information on the FlashROM architecture, possible applications, programming, access through the JTAG and UJTAG interface, and integration into your design. In addition, the Libero tool set enables easy creation and modification of the FlashROM content.

The nonvolatile FlashROM block in the FPGA can be customized, enabling multiple applications.

Additionally, the security offered by the low power flash devices keeps both the contents of FlashROM and the FPGA design safe from system over-builders, system cloners, and IP thieves.

## Related Documents

### User's Guides

*FlashPro User's Guide*

[http://www.microsemi.com/documents/FlashPro\\_UG.pdf](http://www.microsemi.com/documents/FlashPro_UG.pdf)

## List of Changes

The following table lists critical changes that were made in each revision of the chapter.

Date	Changes	Page
July 2010	This chapter is no longer published separately with its own part number and version but is now part of several FPGA fabric user's guides.	N/A
v1.4 (December 2008)	IGLOO nano and ProASIC3 nano devices were added to Table 5-1 • Flash-Based FPGAs.	118
v1.3 (October 2008)	The "FlashROM Support in Flash-Based Devices" section was revised to include new families and make the information more concise.	118
	Figure 5-2 • Fusion Device Architecture Overview (AFS600) was replaced. Figure 5-5 • Programming FlashROM Using AES was revised to change "Fusion" to "Flash Device."	119, 121
	The <i>FlashPoint User's Guide</i> was removed from the "User's Guides" section, as its content is now part of the <i>FlashPro User's Guide</i> .	130
v1.2 (June 2008)	The following changes were made to the family descriptions in Table 5-1 • Flash-Based FPGAs: <ul style="list-style-type: none"> <li>ProASIC3L was updated to include 1.5 V.</li> <li>The number of PLLs for ProASIC3E was changed from five to six.</li> </ul>	118
v1.1 (March 2008)	The chapter was updated to include the IGLOO PLUS family and information regarding 15 k gate devices. The "IGLOO Terminology" section and "ProASIC3 Terminology" section are new.	N/A

## RD

This is the output data bus and is 18 bits wide. Not all 18 bits are valid in all configurations. Like the WD bus, high-order bits become unusable if the data width is less than 18. The output data on unused pins is undefined (Table 6-7).

**Table 6-7 • Input Data Signal Usage for Different Aspect Ratios**

D×W	WD/RD Unused
4k×1	WD[17:1], RD[17:1]
2k×2	WD[17:2], RD[17:2]
1k×4	WD[17:4], RD[17:4]
512×9	WD[17:9], RD[17:9]
256×18	—

## ESTOP, FSTOP

ESTOP is used to stop the FIFO read counter from further counting once the FIFO is empty (i.e., the EMPTY flag goes HIGH). A HIGH on this signal inhibits the counting.

FSTOP is used to stop the FIFO write counter from further counting once the FIFO is full (i.e., the FULL flag goes HIGH). A HIGH on this signal inhibits the counting.

For more information on these signals, refer to the "ESTOP and FSTOP Usage" section.

## FULL, EMPTY

When the FIFO is full and no more data can be written, the FULL flag asserts HIGH. The FULL flag is synchronous to WCLK to inhibit writing immediately upon detection of a full condition and to prevent overflows. Since the write address is compared to a resynchronized (and thus time-delayed) version of the read address, the FULL flag will remain asserted until two WCLK active edges after a read operation eliminates the full condition.

When the FIFO is empty and no more data can be read, the EMPTY flag asserts HIGH. The EMPTY flag is synchronous to RCLK to inhibit reading immediately upon detection of an empty condition and to prevent underflows. Since the read address is compared to a resynchronized (and thus time-delayed) version of the write address, the EMPTY flag will remain asserted until two RCLK active edges after a write operation removes the empty condition.

For more information on these signals, refer to the "FIFO Flag Usage Considerations" section on page 145.

## AFULL, AEMPTY

These are programmable flags and will be asserted on the threshold specified by AFVAL and AEVAL, respectively.

When the number of words stored in the FIFO reaches the amount specified by AEVAL while reading, the AEMPTY output will go HIGH. Likewise, when the number of words stored in the FIFO reaches the amount specified by AFVAL while writing, the AFULL output will go HIGH.

## AFVAL, AEVAL

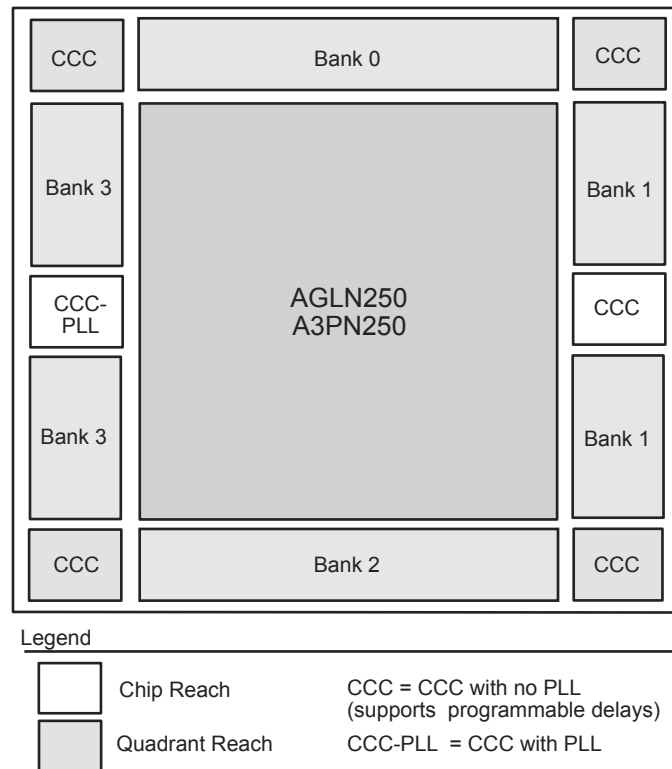
The AEVAL and AFVAL pins are used to specify the almost-empty and almost-full threshold values. They are 12-bit signals. For more information on these signals, refer to the "FIFO Flag Usage Considerations" section on page 145.

## FIFO Usage

### ESTOP and FSTOP Usage

The ESTOP pin is used to stop the read counter from counting any further once the FIFO is empty (i.e., the EMPTY flag goes HIGH). Likewise, the FSTOP pin is used to stop the write counter from counting any further once the FIFO is full (i.e., the FULL flag goes HIGH).

The FIFO counters in the device start the count at zero, reach the maximum depth for the configuration (e.g., 511 for a 512×9 configuration), and then restart at zero. An example application for ESTOP, where the read counter keeps counting, would be writing to the FIFO once and reading the same content over and over without doing another write.



**Figure 7-12 • I/O Bank Architecture of AGLN250/A3PN250 Devices**

## Board-Level Considerations

Low power flash devices have robust I/O features that can help in reducing board-level components. The devices offer single-chip solutions, which makes the board layout simpler and more immune to signal integrity issues. Although, in many cases, these devices resolve board-level issues, special attention should always be given to overall signal integrity. This section covers important board-level considerations to facilitate optimum device performance.

### Termination

Proper termination of all signals is essential for good signal quality. Nonterminated signals, especially clock signals, can cause malfunctioning of the device.

For general termination guidelines, refer to the *Board-Level Considerations* application note for Microsemi FPGAs. Also refer to the "Pin Descriptions and Packaging" chapter of the appropriate device datasheet for termination requirements for specific pins.

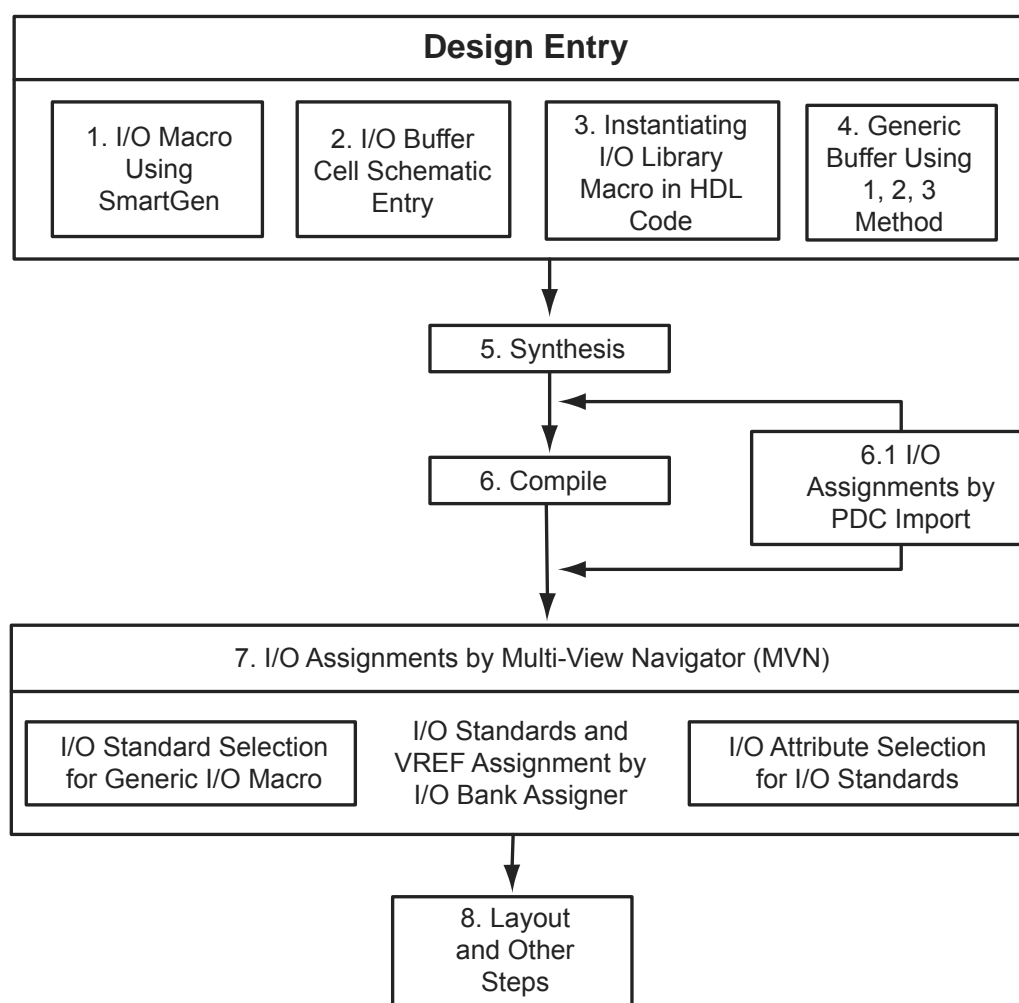
Low power flash I/Os are equipped with on-chip pull-up/-down resistors. The user can enable these resistors by instantiating them either in the top level of the design (refer to the *IGLOO*, *ProASIC3*, *SmartFusion*, and *Fusion Macro Library Guide* for the available I/O macros with pull-up/-down) or in the I/O Attribute Editor in Designer if generic input or output buffers are instantiated in the top level. Unused I/O pins are configured as inputs with pull-up resistors.

As mentioned earlier, low power flash devices have multiple programmable drive strengths, and the user can eliminate unwanted overshoot and undershoot by adjusting the drive strengths.

## 8 – I/O Software Control in Low Power Flash Devices

Fusion, IGLOO, and ProASIC3 I/Os provide more design flexibility, allowing the user to control specific features by enabling certain I/O standards. Some features are selectable only for certain I/O standards, whereas others are available for all I/O standards. For example, slew control is not supported by differential I/O standards. Conversely, I/O register combining is supported by all I/O standards. For detailed information about which I/O standards and features are available on each device and each I/O type, refer to the I/O Structures section of the handbook for the device you are using.

Figure 8-1 shows the various points in the software design flow where a user can provide input or control of the I/O selection and parameters. A detailed description is provided throughout this document.



**Figure 8-1 • User I/O Assignment Flow Chart**



## **I/O Function**

Figure 8-8 shows an example of the I/O Function table included in the I/O bank report:

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---

### **Figure 8-8 • I/O Function Table**

This table lists the number of input I/Os, output I/Os, bidirectional I/Os, and differential input and output I/O pairs that use I/O and DDR registers.

Note: IGLOO nano and ProASIC3 nano devices do not support differential inputs.

Certain rules must be met to implement registered and DDR I/O functions (refer to the I/O Structures section of the handbook for the device you are using and the "DDR" section on page 190).

## **I/O Technology**

The I/O Technology table (shown in Figure 8-9) gives the values of VCCI and VREF (reference voltage) for all the I/O standards used in the design. The user should assign these voltages appropriately.

---

---

### **Figure 8-9 • I/O Technology Table**

those banks, the user does not need to assign the same VCCI voltage to another bank. The user needs to assign the other three VCCI voltages to three more banks.

## Assigning Technologies and VREF to I/O Banks

Low power flash devices offer a wide variety of I/O standards, including voltage-referenced standards. Before proceeding to Layout, each bank must have the required VCCI voltage assigned for the corresponding I/O technologies used for that bank. The voltage-referenced standards require the use of a reference voltage (VREF). This assignment can be done manually or automatically. The following sections describe this in detail.

### Manually Assigning Technologies to I/O Banks

The user can import the PDC at this point and resolve this requirement. The PDC command is

```
set_iobank [bank name] -vcci [vcci value]
```

Another method is to use the I/O Bank Settings dialog box (**MVN > Edit > I/O Bank Settings**) to set up the  $V_{CCI}$  voltage for the bank (Figure 8-12).

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**Figure 8-12 • Setting VCCI for a Bank**

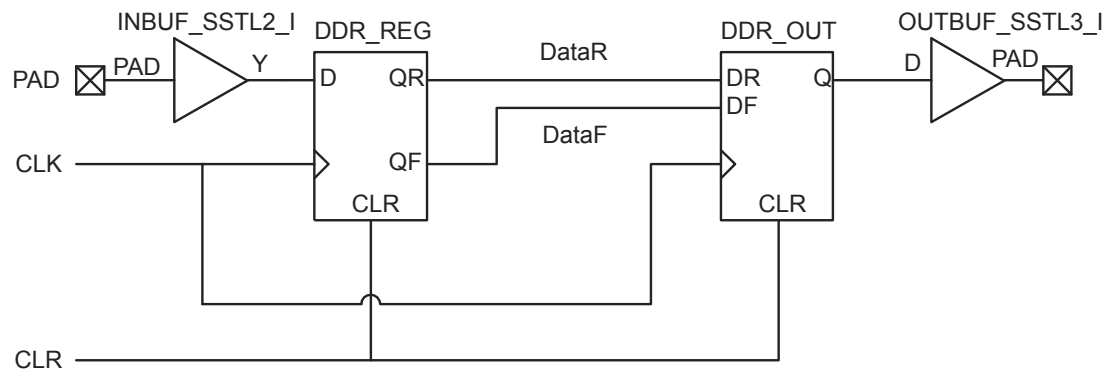
## 9 – DDR for Microsemi’s Low Power Flash Devices

### Introduction

The I/Os in Fusion, IGLOO, and ProASIC3 devices support Double Data Rate (DDR) mode. In this mode, new data is present on every transition (or clock edge) of the clock signal. This mode doubles the data transfer rate compared with Single Data Rate (SDR) mode, where new data is present on one transition (or clock edge) of the clock signal. Low power flash devices have DDR circuitry built into the I/O tiles. I/Os are configured to be DDR receivers or transmitters by instantiating the appropriate special macros (examples shown in Figure 9-4 on page 210 and Figure 9-5 on page 211) and buffers (DDR\_OUT or DDR\_REG) in the RTL design. This document discusses the options the user can choose to configure the I/Os in this mode and how to instantiate them in the design.

### Double Data Rate (DDR) Architecture

Low power flash devices support 350 MHz DDR inputs and outputs. In DDR mode, new data is present on every transition of the clock signal. Clock and data lines have identical bandwidths and signal integrity requirements, making them very efficient for implementing very high-speed systems. High-speed DDR interfaces can be implemented using LVDS (not applicable for IGLOO nano and ProASIC3 nano devices). In IGLOOe, ProASIC3E, AFS600, and AFS1500 devices, DDR interfaces can also be implemented using the HSTL, SSTL, and LVPECL I/O standards. The DDR feature is primarily implemented in the FPGA core periphery and is not tied to a specific I/O technology or limited to any I/O standard.



**Figure 9-1 • DDR Support in Low Power Flash Devices**

## Generating Programming Files

### Generation of the Programming File in a Trusted Environment— Application 1

As discussed in the "Application 1: Trusted Environment" section on page 243, in a trusted environment, the user can choose to program the device with plaintext bitstream content. It is possible to use plaintext for programming even when the FlashLock Pass Key option has been selected. In this application, it is not necessary to employ AES encryption protection. For AES encryption settings, refer to the next sections.

The generated programming file will include the security setting (if selected) and the plaintext programming file content for the FPGA array, FlashROM, and/or FBs. These options are indicated in Table 11-2 and Table 11-3.

**Table 11-2 • IGLOO and ProASIC3 Plaintext Security Options, No AES**

Security Protection	FlashROM Only	FPGA Core Only	Both FlashROM and FPGA
No AES / no FlashLock	✓	✓	✓
FlashLock only	✓	✓	✓
AES and FlashLock	–	–	–

**Table 11-3 • Fusion Plaintext Security Options**

Security Protection	FlashROM Only	FPGA Core Only	FB Core Only	All
No AES / no FlashLock	✓	✓	✓	✓
FlashLock	✓	✓	✓	✓
AES and FlashLock	–	–	–	–

*Note: For all instructions, the programming of Flash Blocks refers to Fusion only.*

For this scenario, generate the programming file as follows:

1. Select the **Silicon features to be programmed** (Security Settings, FPGA Array, FlashROM, Flash Memory Blocks), as shown in Figure 11-10 on page 248 and Figure 11-11 on page 248. Click **Next**.

If **Security Settings** is selected (i.e., the FlashLock security Pass Key feature), an additional dialog will be displayed to prompt you to select the security level setting. If no security setting is selected, you will be directed to Step 3.

*Note: The settings in this figure are used to show the generation of an AES-encrypted programming file for the FPGA array, FlashROM, and FB contents. One or all locations may be selected for encryption.*

---

**Figure 11-17 • Settings to Program a Device Secured with FlashLock and using AES Encryption**

Choose the **High** security level to reprogram devices using both the FlashLock Pass Key and AES key protection (Figure 11-18 on page 255). Enter the AES key and click **Next**.

A device that has already been secured with FlashLock and has an AES key loaded must recognize the AES key to program the device and generate a valid bitstream in authentication. The FlashLock Key is only required to unlock the device and change the security settings.

This is what makes it possible to program in an untrusted environment. The AES key is protected inside the device by the FlashLock Key, so you can only program if you have the correct AES key. In fact, the AES key is not in the programming file either. It is the key used to encrypt the data in the file. The same key previously programmed with the FlashLock Key matches to decrypt the file.

An AES-encrypted file programmed to a device without FlashLock would not be secure, since without FlashLock to protect the AES key, someone could simply reprogram the AES key first, then program with any AES key desired or no AES key at all. This option is therefore not available in the software.

## STAPL vs. DirectC

Programming the low power flash devices is performed using DirectC or the STAPL player. Both tools use the STAPL file as an input. DirectC is a compiled language, whereas STAPL is an interpreted language. Microprocessors will be able to load the FPGA using DirectC much more quickly than STAPL. This speed advantage becomes more apparent when lower clock speeds of 8- or 16-bit microprocessors are used. DirectC also requires less memory than STAPL, since the programming algorithm is directly implemented. STAPL does have one advantage over DirectC—the ability to upgrade. When a new programming algorithm is required, the STAPL user simply needs to regenerate a STAPL file using the latest version of the Designer software and download it to the system. The DirectC user must download the latest version of DirectC from Microsemi, compile everything, and download the result into the system (Figure 14-4).

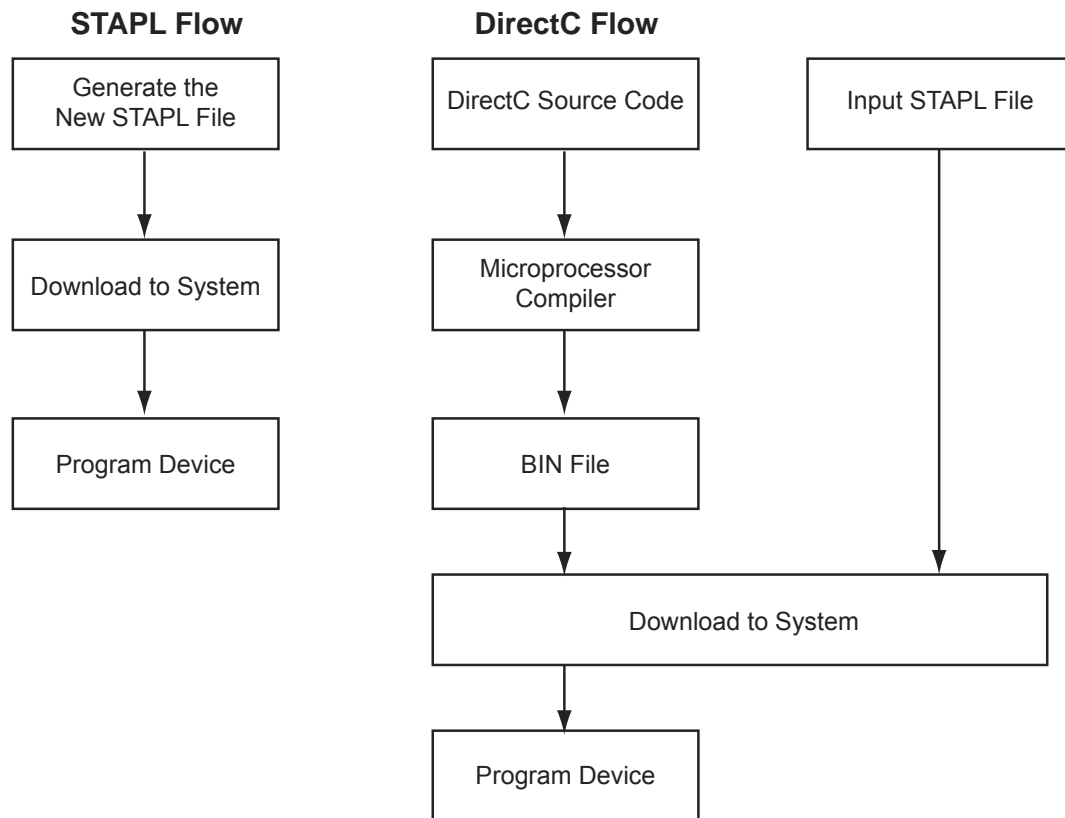


Figure 14-4 • STAPL vs. DirectC

## 15 – Boundary Scan in Low Power Flash Devices

### Boundary Scan

Low power flash devices are compatible with IEEE Standard 1149.1, which defines a hardware architecture and the set of mechanisms for boundary scan testing. JTAG operations are used during boundary scan testing.

The basic boundary scan logic circuit is composed of the TAP controller, test data registers, and instruction register (Figure 15-2 on page 294).

Low power flash devices support three types of test data registers: bypass, device identification, and boundary scan. The bypass register is selected when no other register needs to be accessed in a device. This speeds up test data transfer to other devices in a test data path. The 32-bit device identification register is a shift register with four fields (LSB, ID number, part number, and version). The boundary scan register observes and controls the state of each I/O pin. Each I/O cell has three boundary scan register cells, each with serial-in, serial-out, parallel-in, and parallel-out pins.

### TAP Controller State Machine

The TAP controller is a 4-bit state machine (16 states) that operates as shown in Figure 15-1.

The 1s and 0s represent the values that must be present on TMS at a rising edge of TCK for the given state transition to occur. IR and DR indicate that the instruction register or the data register is operating in that state.

The TAP controller receives two control inputs (TMS and TCK) and generates control and clock signals for the rest of the test logic architecture. On power-up, the TAP controller enters the Test-Logic-Reset state. To guarantee a reset of the controller from any of the possible states, TMS must remain HIGH for five TCK cycles. The TRST pin can also be used to asynchronously place the TAP controller in the Test-Logic-Reset state.

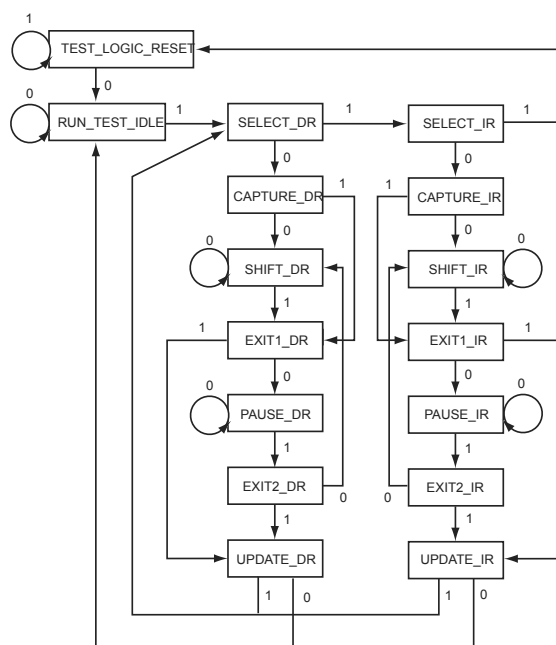


Figure 15-1 • TAP Controller State Machine

## Boundary Scan Support in Low Power Devices

The information in this document applies to all Fusion, IGLOO, and ProASIC3 devices. For IGLOO, IGLOO PLUS, and ProASIC3L devices, the Flash\*Freeze pin must be deasserted for successful boundary scan operations. Devices cannot enter JTAG mode directly from Flash\*Freeze mode.

## Boundary Scan Opcodes

Low power flash devices support all mandatory IEEE 1149.1 instructions (EXTEST, SAMPLE/PRELOAD, and BYPASS) and the optional IDCODE instruction (Table 15-2).

**Table 15-2 • Boundary Scan Opcodes**

	Hex Opcode
EXTEST	00
HIGHZ	07
USERCODE	0E
SAMPLE/PRELOAD	01
IDCODE	0F
CLAMP	05
BYPASS	FF

## Boundary Scan Chain

The serial pins are used to serially connect all the boundary scan register cells in a device into a boundary scan register chain (Figure 15-2 on page 294), which starts at the TDI pin and ends at the TDO pin. The parallel ports are connected to the internal core logic I/O tile and the input, output, and control ports of an I/O buffer to capture and load data into the register to control or observe the logic state of each I/O.

Each test section is accessed through the TAP, which has five associated pins: TCK (test clock input), TDI, TDO (test data input and output), TMS (test mode selector), and TRST (test reset input). TMS, TDI, and TRST are equipped with pull-up resistors to ensure proper operation when no input data is supplied to them. These pins are dedicated for boundary scan test usage. Refer to the "JTAG Pins" section in the "Pin Descriptions and Packaging" chapter of the appropriate device datasheet for pull-up/-down recommendations for TCK and TRST pins. Pull-down recommendations are also given in Table 15-3 on page 294

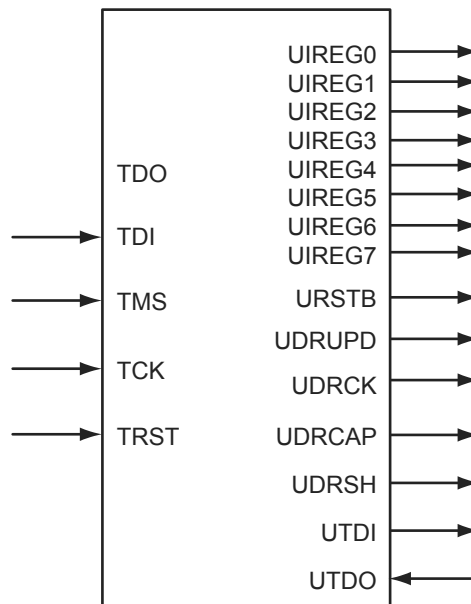


## UJTAG Macro

The UJTAG tiles can be instantiated in a design using the UJTAG macro from the Fusion, IGLOO, or ProASIC3 macro library. Note that "UJTAG" is a reserved name and cannot be used for any other user-defined blocks. A block symbol of the UJTAG tile macro is presented in Figure 16-2. In this figure, the ports on the left side of the block are connected to the JTAG TAP Controller, and the right-side ports are accessible by the FPGA core VersaTiles. The TDI, TMS, TDO, TCK, and TRST ports of UJTAG are only provided for design simulation purposes and should be treated as external signals in the design netlist. However, these ports must NOT be connected to any I/O buffer in the netlist. Figure 16-3 on page 300 illustrates the correct connection of the UJTAG macro to the user design netlist. Microsemi Designer software will automatically connect these ports to the TAP during place-and-route. Table 16-2 gives the port descriptions for the rest of the UJTAG ports:

**Table 16-2 • UJTAG Port Descriptions**

Port	Description
UIREG [7:0]	This 8-bit bus carries the contents of the JTAG Instruction Register of each device. Instruction Register values 16 to 127 are not reserved and can be employed as user-defined instructions.
URSTB	URSTB is an active-low signal and will be asserted when the TAP Controller is in Test-Logic-Reset mode. URSTB is asserted at power-up, and a power-on reset signal resets the TAP Controller. URSTB will stay asserted until an external TAP access changes the TAP Controller state.
UTDI	This port is directly connected to the TAP's TDI signal.
UTDO	This port is the user TDO output. Inputs to the UTDO port are sent to the TAP TDO output MUX when the IR address is in user range.
UDRSH	Active-high signal enabled in the ShiftDR TAP state
UDRCAP	Active-high signal enabled in the CaptureDR TAP state
UDRCK	This port is directly connected to the TAP's TCK signal.
UDRUPD	Active-high signal enabled in the UpdateDR TAP state



**Figure 16-2 • UJTAG Tile Block Symbol**