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Understanding <u>Embedded - FPGAs (Field Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	77
Number of Gates	30000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	100-TQFP
Supplier Device Package	100-VQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/a3pn030-zvq100i

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

Global Resources in Low Power Flash Devices

Global Resource Support in Flash-Based Devices

The flash FPGAs listed in Table 3-1 support the global resources and the functions described in this document.

Table 3-1 • Flash-Based FPGAs

Series	Family*	Description
IGL00	IGL00	Ultra-low power 1.2 V to 1.5 V FPGAs with Flash*Freeze technology
	IGLOOe	Higher density IGLOO FPGAs with six PLLs and additional I/O standards
	IGLOO PLUS	IGLOO FPGAs with enhanced I/O capabilities
	IGLOO nano	The industry's lowest-power, smallest-size solution
ProASIC3	ProASIC3	Low power, high-performance 1.5 V FPGAs
	ProASIC3E	Higher density ProASIC3 FPGAs with six PLLs and additional I/O standards
	ProASIC3 nano	Lowest-cost solution with enhanced I/O capabilities
	ProASIC3L	ProASIC3 FPGAs supporting 1.2 V to 1.5 V with Flash*Freeze technology
	RT ProASIC3	Radiation-tolerant RT3PE600L and RT3PE3000L
	Military ProASIC3/EL	Military temperature A3PE600L, A3P1000, and A3PE3000L
	Automotive ProASIC3	ProASIC3 FPGAs qualified for automotive applications
Fusion	Fusion	Mixed signal FPGA integrating ProASIC3 FPGA fabric, programmable analog block, support for ARM [®] Cortex [™] -M1 soft processors, and flash memory into a monolithic device

Note: *The device names link to the appropriate datasheet, including product brief, DC and switching characteristics, and packaging information.

IGLOO Terminology

In documentation, the terms IGLOO series and IGLOO devices refer to all of the IGLOO products as listed in Table 3-1. Where the information applies to only one product line or limited devices, these exclusions will be explicitly stated.

ProASIC3 Terminology

In documentation, the terms ProASIC3 series and ProASIC3 devices refer to all of the ProASIC3 devices as listed in Table 3-1. Where the information applies to only one product line or limited devices, these exclusions will be explicitly stated.

To further understand the differences between the IGLOO and ProASIC3 devices, refer to the *Industry's Lowest Power FPGAs Portfolio*.

Global Resources in Low Power Flash Devices

External I/O or Local signal as Clock Source

External I/O refers to regular I/O pins are labeled with the I/O convention IOuxwByVz. You can allow the external I/O or internal signal to access the global. To allow the external I/O or internal signal to access the global network, you need to instantiate the CLKINT macro. Refer to Figure 3-4 on page 35 for an example illustration of the connections. Instead of using CLKINT, you can also use PDC to promote signals from external I/O or internal signal to the global network. However, it may cause layout issues because of synthesis logic replication. Refer to the "Global Promotion and Demotion Using PDC" section on page 51 for details.

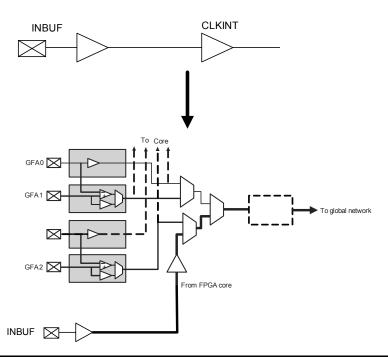


Figure 3-14 • CLKINT Macro

Using Global Macros in Synplicity

The Synplify[®] synthesis tool automatically inserts global buffers for nets with high fanout during synthesis. By default, Synplicity[®] puts six global macros (CLKBUF or CLKINT) in the netlist, including any global instantiation or PLL macro. Synplify always honors your global macro instantiation. If you have a PLL (only primary output is used) in the design, Synplify adds five more global buffers in the netlist. Synplify uses the following global counting rule to add global macros in the netlist:

- 1. CLKBUF: 1 global buffer
- 2. CLKINT: 1 global buffer
- 3. CLKDLY: 1 global buffer
- 4. PLL: 1 to 3 global buffers
 - GLA, GLB, GLC, YB, and YC are counted as 1 buffer.
 - GLB or YB is used or both are counted as 1 buffer.
 - GLC or YC is used or both are counted as 1 buffer.

Global Resources in Low Power Flash Devices

Using Spines of Occupied Global Networks

When a signal is assigned to a global network, the flash switches are programmed to set the MUX select lines (explained in the "Clock Aggregation Architecture" section on page 45) to drive the spines of that network with the global net. However, if the global net is restricted from reaching into the scope of a

For example, if you want to limit the CLK1_c signal to the left half of the chip and want to use the right side of the same global network for CLK2 c, you can add the following PDC commands:

spine, the MUX drivers of that spine are available for other high-fanout or critical signals (Figure 3-20).

```
define_region -name region1 -type inclusive 0 0 34 29
assign_net_macros region1 CLK1_c
assign_local_clock -net CLK2_c -type chip B2
```

Figure 3-20 • Design Example Using Spines of Occupied Global Networks

Conclusion

IGLOO, Fusion, and ProASIC3 devices contain 18 global networks: 6 chip global networks and 12 quadrant global networks. These global networks can be segmented into local low-skew networks called spines. The spines provide low-skew networks for the high-fanout signals of a design. These allow you up to 252 different internal/external clocks in an A3PE3000 device. This document describes the architecture for the global network, plus guidelines and methodologies in assigning signals to globals and spines.

Related Documents

User's Guides

IGLOO, ProASIC3, SmartFusion, and Fusion Macro Library Guide http://www.microsemi.com/soc/documents/pa3_libguide_ug.pdf



Available I/O Standards

Table 4-4 • Available I/O Standards within CLKBUF and CLKBUF LVDS/LVPECL Macros

CLKBUF_LVCMOS5
CLKBUF_LVCMOS33 ¹
CLKBUF_LVCMOS25 ²
CLKBUF_LVCMOS18
CLKBUF_LVCMOS15
CLKBUF_PCI
CLKBUF_PCIX ³
CLKBUF_GTL25 ^{2,3}
CLKBUF_GTL33 ^{2,3}
CLKBUF_GTLP25 ^{2,3}
CLKBUF_GTLP33 ^{2,3}
CLKBUF_HSTL_I ^{2,3}
CLKBUF_HSTL_II 2,3
CLKBUF_SSTL3_I ^{2,3}
CLKBUF_SSTL3_II ^{2,3}
CLKBUF_SSTL2_I ^{2,3}
CLKBUF_SSTL2_II ^{2,3}
CLKBUF_LVDS 4,5
CLKBUF_LVPECL ⁵
AL /

Notes:

- 1. By default, the CLKBUF macro uses 3.3 V LVTTL I/O technology. For more details, refer to the IGLOO, ProASIC3, SmartFusion, and Fusion Macro Library Guide.
- 2. I/O standards only supported in ProASIC3E and IGLOOe families.
- 3. I/O standards only supported in the following Fusion devices: AFS600 and AFS1500.
- 4. B-LVDS and M-LVDS standards are supported by CLKBUF_LVDS.
- 5. Not supported for IGLOO nano and ProASIC3 nano devices.

Global Synthesis Constraints

The Synplify[®] synthesis tool, by default, allows six clocks in a design for Fusion, IGLOO, and ProASIC3. When more than six clocks are needed in the design, a user synthesis constraint attribute, syn_global_buffers, can be used to control the maximum number of clocks (up to 18) that can be inferred by the synthesis engine.

High-fanout nets will be inferred with clock buffers and/or internal clock buffers. If the design consists of CCC global buffers, they are included in the count of clocks in the design.

The subsections below discuss the clock input source (global buffers with no programmable delays) and the clock conditioning functional block (global buffers with programmable delays and/or PLL function) in detail.



ProASIC3 nano FPGA Fabric User's Guide

This section outlines the following device information: CCC features, PLL core specifications, functional descriptions, software configuration information, detailed usage information, recommended board-level considerations, and other considerations concerning global networks in low power flash devices.

Clock Conditioning Circuits with Integrated PLLs

Each of the CCCs with integrated PLLs includes the following:

- 1 PLL core, which consists of a phase detector, a low-pass filter, and a four-phase voltagecontrolled oscillator
- 3 global multiplexer blocks that steer signals from the global pads and the PLL core onto the global networks
- · 6 programmable delays and 1 fixed delay for time advance/delay adjustments
- 5 programmable frequency divider blocks to provide frequency synthesis (automatically configured by the SmartGen macro builder tool)

Clock Conditioning Circuits without Integrated PLLs

There are two types of simplified CCCs without integrated PLLs in low power flash devices.

- 1. The simplified CCC with programmable delays, which is composed of the following:
 - 3 global multiplexer blocks that steer signals from the global pads and the programmable delay elements onto the global networks
 - 3 programmable delay elements to provide time delay adjustments
- 2. The simplified CCC (referred to as CCC-GL) without programmable delay elements, which is composed of the following:
 - A global multiplexer block that steer signals from the global pads onto the global networks

Clock Conditioning Circuits in Low Power Flash Devices and Mixed Signal FPGAs

CCC Locations

CCCs located in the middle of the east and west sides of the device access the three VersaNet global networks on each side (six total networks), while the four CCCs located in the four corners access three quadrant global networks (twelve total networks). See Figure 4-13.

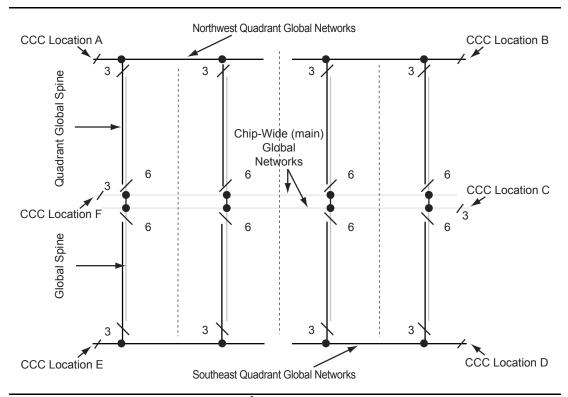


Figure 4-13 • Global Network Architecture for 60 k Gate Devices and Above

The following explains the locations of the CCCs in IGLOO and ProASIC3 devices:

In Figure 4-15 on page 82 through Figure 4-16 on page 82, CCCs with integrated PLLs are indicated in red, and simplified CCCs are indicated in yellow. There is a letter associated with each location of the CCC, in clockwise order. The upper left corner CCC is named "A," the upper right is named "B," and so on. These names finish up at the middle left with letter "F."



Table 4-9 to Table 4-15 on page 94 provide descriptions of the configuration data for the configuration bits.

Table 4-9 • Input Clock Divider, FINDIV[6:0] (/n)

FINDIV<6:0> State	Divisor	New Frequency Factor	
0	1	1.00000	
1	2	0.50000	
:	:	:	
127	128	0.0078125	

Table 4-10 • Feedback Clock Divider, FBDIV[6:0] (/m)

FBDIV<6:0> State	Divisor	New Frequency Factor	
0	1	1	
1	2	2	
:	:	:	
127	128	128	

Table 4-11 • Output Frequency Dividers

A Output Divider, OADIV <4:0> (/u);

B Output Divider, OBDIV <4:0> (/v);

C Output Divider, OCDIV <4:0> (/w)

OADIV<4:0>; OBDIV<4:0>; CDIV<4:0> State	Divisor	New Frequency Factor
0	1	1.00000
1	2	0.50000
:	i i	i i
31	32	0.03125

Table 4-12 • MUXA, MUXB, MUXC

OAMUX<2:0>; OBMUX<2:0>; OCMUX<2:0> State	MUX Input Selected
0	None. Six-input MUX and PLL are bypassed. Clock passes only through global MUX and goes directly into HC ribs.
1	Not available
2	PLL feedback delay line output
3	Not used
4	PLL VCO 0° phase shift
5	PLL VCO 270° phase shift
6	PLL VCO 180° phase shift
7	PLL VCO 90° phase shift

Figure 4-31 • Static Timing Analysis Using SmartTime

Place-and-Route Stage Considerations

Several considerations must be noted to properly place the CCC macros for layout.

For CCCs with clock inputs configured with the Hardwired I/O-Driven option:

- PLL macros must have the clock input pad coming from one of the GmA* locations.
- CLKDLY macros must have the clock input pad coming from one of the Global I/Os.

If a PLL with a Hardwired I/O input is used at a CCC location and a Hardwired I/O–Driven CLKDLY macro is used at the same CCC location, the clock input of the CLKDLY macro must be chosen from one of the GmB* or GmC* pin locations. If the PLL is not used or is an External I/O–Driven or Core Logic–Driven PLL, the clock input of the CLKDLY macro can be sourced from the GmA*, GmB*, or GmC* pin locations.

For CCCs with clock inputs configured with the External I/O–Driven option, the clock input pad can be assigned to any regular I/O location (IO************* pins). Note that since global I/O pins can also be used as regular I/Os, regardless of CCC function (CLKDLY or PLL), clock inputs can also be placed in any of these I/O locations.

By default, the Designer layout engine will place global nets in the design at one of the six chip globals. When the number of globals in the design is greater than six, the Designer layout engine will automatically assign additional globals to the quadrant global networks of the low power flash devices. If the user wishes to decide which global signals should be assigned to chip globals (six available) and which to the quadrant globals (three per quadrant for a total of 12 available), the assignment can be achieved with PinEditor, ChipPlanner, or by importing a placement constraint file. Layout will fail if the



FlashROM in Microsemi's Low Power Flash Devices

Figure 5-12 shows the programming file generator, which enables different STAPL file generation methods. When you select **Program FlashROM** and choose the UFC file, the FlashROM Settings window appears, as shown in Figure 5-13. In this window, you can select the FlashROM page you want to program and the data value for the configured regions. This enables you to use a different page for different programming files.

Figure 5-12 • Programming File Generator

Figure 5-13 • Setting FlashROM during Programming File Generation

The programming hardware and software can load the FlashROM with the appropriate STAPL file. Programming software handles the single STAPL file that contains multiple FlashROM contents for multiple devices, and programs the FlashROM in sequential order (e.g., for device serialization). This feature is supported in the programming software. After programming with the STAPL file, you can run DEVICE_INFO to check the FlashROM content.



Conclusion

Fusion, IGLOO, and ProASIC3 devices provide users with extremely flexible SRAM blocks for most design needs, with the ability to choose between an easy-to-use dual-port memory or a wide-word twoport memory. Used with the built-in FIFO controllers, these memory blocks also serve as highly efficient FIFOs that do not consume user gates when implemented. The SmartGen core generator provides a fast and easy way to configure these memory elements for use in designs.

List of Changes

The following table lists critical changes that were made in each revision of the chapter.

Date	Changes	Page
August 2012	The note connected with Figure 6-3 • Supported Basic RAM Macros, regarding RAM4K9, was revised to explain that it applies only to part numbers of certain revisions and earlier (SAR 29574).	
July 2010	This chapter is no longer published separately with its own part number and version but is now part of several FPGA fabric user's guides.	N/A
v1.5 (December 2008)	IGLOO nano and ProASIC3 nano devices were added to Table 6-1 • Flash-Based FPGAs.	134
	IGLOO nano and ProASIC3 nano devices were added to Figure 6-8 • Interfacing TAP Ports and SRAM Blocks.	148
v1.4 (October 2008)	The "SRAM/FIFO Support in Flash-Based Devices" section was revised to include new families and make the information more concise.	134
	The "SRAM and FIFO Architecture" section was modified to remove "IGLOO and ProASIC3E" from the description of what the memory block includes, as this statement applies to all memory blocks.	135
	Wording in the "Clocking" section was revised to change "IGLOO and ProASIC3 devices support inversion" to "Low power flash devices support inversion." The reference to IGLOO and ProASIC3 development tools in the last paragraph of the section was changed to refer to development tools in general.	141
	The "ESTOP and FSTOP Usage" section was updated to refer to FIFO counters in devices in general rather than only IGLOO and ProASIC3E devices.	144
v1.3 (August 2008)	The note was removed from Figure 6-7 • RAM Block with Embedded FIFO Controller and placed in the WCLK and RCLK description.	142
	The "WCLK and RCLK" description was revised.	143
v1.2 (June 2008)	The following changes were made to the family descriptions in Table 6-1 • Flash-Based FPGAs:	134
	ProASIC3L was updated to include 1.5 V.	
	The number of PLLs for ProASIC3E was changed from five to six.	
v1.1 (March 2008)	The "Introduction" section was updated to include the IGLOO PLUS family.	131
	The "Device Architecture" section was updated to state that 15 k gate devices do not support SRAM and FIFO.	131
	The first note in Figure 6-1 • IGLOO and ProASIC3 Device Architecture Overview was updated to include mention of 15 k gate devices, and IGLOO PLUS was added to the second note.	133



Power-Up Behavior

Low power flash devices are power-up/-down friendly; i.e., no particular sequencing is required for power-up and power-down. This eliminates extra board components for power-up sequencing, such as a power-up sequencer.

During power-up, all I/Os are tristated, irrespective of I/O macro type (input buffers, output buffers, I/O buffers with weak pull-ups or weak pull-downs, etc.). Once I/Os become activated, they are set to the user-selected I/O macros. Refer to the "Power-Up/-Down Behavior of Low Power Flash Devices" section on page 307 for details.

Drive Strength

Low power flash devices have up to four programmable output drive strengths. The user can select the drive strength of a particular output in the I/O Attribute Editor or can instantiate a specialized I/O macro, such as OUTBUF_S_8 (slew = low, out_drive = 8 mA).

The maximum available drive strength is 8 mA per I/O. Though no I/O should be forced to source or sink more than 8 mA indefinitely, I/Os may handle a higher amount of current (refer to the device IBIS model for maximum source/sink current) during signal transition (AC current). Every device package has its own power dissipation limit; hence, power calculation must be performed accurately to determine how much current can be tolerated per I/O within that limit.

I/O Interfacing

Low power flash devices are 5 V-input- and 5 V-output-tolerant without adding any extra circuitry. Along with other low-voltage I/O macros, this 5 V tolerance makes these devices suitable for many types of board component interfacing.

Table 7-17 shows some high-level interfacing examples using low power flash devices.

Table 7-17 • nano High-Level Interface

	Clock		I/O			
Interface	Туре	Frequency	Туре	Signals In	Signals Out	Data I/O
GM	Src Sync	125 MHz	LVTTL	8	8	125 Mbps
TBI	Src Sync	125 MHz	LVTTL	10	10	125 Mbps

Conclusion

IGLOO nano and ProASIC3 nano device support for multiple I/O standards minimizes board-level components and makes possible a wide variety of applications. The Microsemi Designer software, integrated with Libero SoC, presents a clear visual display of I/O assignments, allowing users to verify I/O and board-level design requirements before programming the device. The nano device I/O features and functionalities ensure board designers can produce low-cost and low power FPGA applications fulfilling the complexities of contemporary design needs.

I/O Software Control in Low Power Flash Devices

List of Changes

The following table lists critical changes that were made in each revision of the document.

Date	Changes		
August 2012	The notes in Table 8-2 • Designer State (resulting from I/O attribute modification) were revised to clarify which device families support programmable input delay (SAR 39666).		
June 2011	e 2011 Figure 8-2 • SmartGen Catalog was updated (SAR 24310). Figure 8-3 • Expanded I/O Section and the step associated with it were deleted to reflect changes in the software.		
	The following rule was added to the "VREF Rules for the Implementation of Voltage-Referenced I/O Standards" section:	199	
	Only minibanks that contain input or bidirectional I/Os require a VREF. A VREF is not needed for minibanks composed of output or tristated I/Os (SAR 24310).		
July 2010	Notes were added where appropriate to point out that IGLOO nano and ProASIC3 nano devices do not support differential inputs (SAR 21449).		
v1.4 (December 2008)	IGLOO nano and ProASIC3 nano devices were added to Table 8-1 • Flash-Based FPGAs.	186	
	The notes for Table 8-2 • Designer State (resulting from I/O attribute modification) were revised to indicate that skew control and input delay do not apply to nano devices.	187	
v1.3 (October 2008)	The "Flash FPGAs I/O Support" section was revised to include new families and make the information more concise.	186	
v1.2 (June 2008)	The following changes were made to the family descriptions in Table 8-1 • Flash-Based FPGAs: • ProASIC3L was updated to include 1.5 V. • The number of PLLs for ProASIC3E was changed from five to six.	186	
v1.1 (March 2008)	This document was previously part of the I/O Structures in IGLOO and ProASIC3 Devices document. The content was separated and made into a new document.	N/A	
	Table 8-2 • Designer State (resulting from I/O attribute modification) was updated to include note 2 for IGLOO PLUS.	187	

I/O Cell Architecture

Low power flash devices support DDR in the I/O cells in four different modes: Input, Output, Tristate, and Bidirectional pins. For each mode, different I/O standards are supported, with most I/O standards having special sub-options. For the ProASIC3 nano and IGLOO nano devices, DDR is supported only in the 60 k, 125 k, and 250 k logic densities. Refer to Table 9-2 for a sample of the available I/O options. Additional I/O options can be found in the relevant family datasheet.

Table 9-2 • DDR I/O Options

DDR Register Type	I/O Type	I/O Standard	Sub-Options	Comments
Receive Register	Input	Normal	None	3.3 V TTL (default)
		LVCMOS	Voltage	1.5 V, 1.8 V, 2.5 V, 5 V (1.5 V default)
			Pull-Up	None (default)
		PCI/PCI-X	None	
		GTL/GTL+	Voltage	2.5 V, 3.3 V (3.3 V default)
		HSTL	Class	I / II (I default)
		SSTL2/SSTL3	Class	I / II (I default)
		LVPECL	None	
		LVDS	None	
Transmit Register	Output	Normal	None	3.3 V TTL (default)
		LVTTL	Output Drive	2, 4, 6, 8, 12, 16, 24, 36 mA (8 mA default)
			Slew Rate	Low/high (high default)
		LVCMOS	Voltage	1.5 V, 1.8 V, 2.5 V, 5 V (1.5 V default)
		PCI/PCI-X	None	
		GTL/GTL+	Voltage	1.8 V, 2.5 V, 3.3 V (3.3 V default)
		HSTL	Class	I / II (I default)
		SSTL2/SSTL3	Class	I / II (I default)
		LVPECL*	None	
		LVDS*	None	

Note: *IGLOO nano and ProASIC3 nano devices do not support differential inputs.



DDR for Microsemi's Low Power Flash Devices

Instantiating DDR Registers

Using SmartGen is the simplest way to generate the appropriate RTL files for use in the design. Figure 9-4 shows an example of using SmartGen to generate a DDR SSTL2 Class I input register. SmartGen provides the capability to generate all of the DDR I/O cells as described. The user, through the graphical user interface, can select from among the many supported I/O standards. The output formats supported are Verilog, VHDL, and EDIF.

Figure 9-5 on page 211 through Figure 9-8 on page 214 show the I/O cell configured for DDR using SSTL2 Class I technology. For each I/O standard, the I/O pad is buffered by a special primitive that indicates the I/O standard type.

Figure 9-4 • Example of Using SmartGen to Generate a DDR SSTL2 Class I Input Register



Types of Programming for Flash Devices

The number of devices to be programmed will influence the optimal programming methodology. Those available are listed below:

- · In-system programming
 - Using a programmer
 - Using a microprocessor or microcontroller
- · Device programmers
 - Single-site programmers
 - Multi-site programmers, batch programmers, or gang programmers
 - Automated production (robotic) programmers
- Volume programming services
 - Microsemi in-house programming
 - Programming centers

In-System Programming

Device Type Supported: Flash

ISP refers to programming the FPGA after it has been mounted on the system printed circuit board. The FPGA may be preprogrammed and later reprogrammed using ISP.

The advantage of using ISP is the ability to update the FPGA design many times without any changes to the board. This eliminates the requirement of using a socket for the FPGA, saving cost and improving reliability. It also reduces programming hardware expenses, as the ISP methodology is die-/package-independent.

There are two methods of in-system programming: external and internal.

- Programmer ISP—Refer to the "In-System Programming (ISP) of Microsemi's Low Power Flash Devices Using FlashPro4/3/3X" section on page 261 for more information.
 - Using an external programmer and a cable, the device can be programmed through a header on the system board. In Microsemi SoC Products Group documentation, this is referred to as external ISP. Microsemi provides FlashPro4, FlashPro3, FlashPro Lite, or Silicon Sculptor 3 to perform external ISP. Note that Silicon Sculptor II and Silicon Sculptor 3 can only provide ISP for ProASIC and ProASICPLUS® families, not for SmartFusion, Fusion, IGLOO, or ProASIC3. Silicon Sculptor II and Silicon Sculptor 3 can be used for programming ProASIC and ProASICPLUS devices by using an adapter module (part number SMPA-ISP-ACTEL-3).
 - Advantages: Allows local control of programming and data files for maximum security. The
 programming algorithms and hardware are available from Microsemi. The only hardware
 required on the board is a programming header.
 - Limitations: A negligible board space requirement for the programming header and JTAG signal routing
- Microprocessor ISP—Refer to the "Microprocessor Programming of Microsemi's Low Power Flash Devices" chapter of an appropriate FPGA fabric user's guide for more information.
 - Using a microprocessor and an external or internal memory, you can store the program in memory and use the microprocessor to perform the programming. In Microsemi documentation, this is referred to as internal ISP. Both the code for the programming algorithm and the FPGA programming file must be stored in memory on the board. Programming voltages must also be generated on the board.
 - Advantages: The programming code is stored in the system memory. An external programmer is not required during programming.
 - Limitations: This is the approach that requires the most design work, since some way of getting and/or storing the data is needed; a system interface to the device must be designed; and the low-level API to the programming firmware must be written and linked into the code provided by Microsemi. While there are benefits to this methodology, serious thought and planning should go into the decision.

Boundary Scan in Low Power Flash Devices

Microsemi's Flash Devices Support the JTAG Feature

The flash-based FPGAs listed in Table 15-1 support the JTAG feature and the functions described in this document.

Table 15-1 • Flash-Based FPGAs

Series	Family*	Description
IGLOO IGLOO		Ultra-low power 1.2 V to 1.5 V FPGAs with Flash*Freeze technology
	IGLO0e	Higher density IGLOO FPGAs with six PLLs and additional I/O standards
	IGLOO nano	The industry's lowest-power, smallest-size solution
	IGLOO PLUS	IGLOO FPGAs with enhanced I/O capabilities
ProASIC3	ProASIC3	Low power, high-performance 1.5 V FPGAs
	ProASIC3E	Higher density ProASIC3 FPGAs with six PLLs and additional I/O standards
	ProASIC3 nano	Lowest-cost solution with enhanced I/O capabilities
	ProASIC3L	ProASIC3 FPGAs supporting 1.2 V to 1.5 V with Flash*Freeze technology
	RT ProASIC3	Radiation-tolerant RT3PE600L and RT3PE3000L
	Military ProASIC3/EL	Military temperature A3PE600L, A3P1000, and A3PE3000L
	Automotive ProASIC3	ProASIC3 FPGAs qualified for automotive applications
Fusion	Fusion	Mixed signal FPGA integrating ProASIC [®] 3 FPGA fabric, programmable analog block, support for ARM [®] Cortex [™] -M1 soft processors, and flash memory into a monolithic device

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In documentation, the terms ProASIC3 series and ProASIC3 devices refer to all of the ProASIC3 devices as listed in Table 15-1. Where the information applies to only one product line or limited devices, these exclusions will be explicitly stated.

To further understand the differences between the IGLOO and ProASIC3 devices, refer to the *Industry's Lowest Power FPGAs Portfolio*.

Power-Up/-Down Sequence and Transient Current

Microsemi's low power flash devices use the following main voltage pins during normal operation:²

- VCCPLX
- VJTAG
- · VCC: Voltage supply to the FPGA core
 - $-\,$ VCC is 1.5 V \pm 0.075 V for IGLOO, IGLOO nano, IGLOO PLUS, and ProASIC3 devices operating at 1.5 V.
 - VCC is 1.2 V ± 0.06 V for IGLOO, IGLOO nano, IGLOO PLUS, and ProASIC3L devices operating at 1.2 V.
 - V5 devices will require a 1.5 V VCC supply, whereas V2 devices can utilize either a 1.2 V or 1.5 V VCC.
- VCCIBx: Supply voltage to the bank's I/O output buffers and I/O logic. Bx is the I/O bank number.
- VMVx: Quiet supply voltage to the input buffers of each I/O bank. x is the bank number. (Note: IGLOO nano, IGLOO PLUS, and ProASIC3 nano devices do not have VMVx supply pins.)

The I/O bank VMV pin must be tied to the VCCI pin within the same bank. Therefore, the supplies that need to be powered up/down during normal operation are VCC and VCCI. These power supplies can be powered up/down in any sequence during normal operation of IGLOO, IGLOO nano, IGLOO PLUS, ProASIC3L, ProASIC3, and ProASIC3 nano FPGAs. During power-up, I/Os in each bank will remain tristated until the last supply (either VCCIBx or VCC) reaches its functional activation voltage. Similarly, during power-down, I/Os of each bank are tristated once the first supply reaches its brownout deactivation voltage.

Although Microsemi's low power flash devices have no power-up or power-down sequencing requirements, Microsemi identifies the following power conditions that will result in higher than normal transient current. Use this information to help maximize power savings:

Microsemi recommends tying VCCPLX to VCC and using proper filtering circuits to decouple VCC noise from the PLL.

- a. If VCCPLX is powered up before VCC, a static current of up to 5 mA (typical) per PLL may be measured on VCCPLX.
 - The current vanishes as soon as VCC reaches the VCCPLX voltage level.
 - The same current is observed at power-down (VCC before VCCPLX).
- b. If VCCPLX is powered up simultaneously or after VCC:
 - Microsemi's low power flash devices exhibit very low transient current on VCC. For ProASIC3 devices, the maximum transient current on V_{CC} does not exceed the maximum standby current specified in the device datasheet.

The source of transient current, also known as inrush current, varies depending on the FPGA technology. Due to their volatile technology, the internal registers in SRAM FPGAs must be initialized before configuration can start. This initialization is the source of significant inrush current in SRAM FPGAs during power-up. Due to the nonvolatile nature of flash technology, low power flash devices do not require any initialization at power-up, and there is very little or no crossbar current through PMOS and NMOS devices. Therefore, the transient current at power-up is significantly less than for SRAM FPGAs. Figure 17-1 on page 310 illustrates the types of power consumption by SRAM FPGAs compared to Microsemi's antifuse and flash FPGAs.

For more information on Microsemi FPGA voltage supplies, refer to the appropriate datasheet located at http://www.microsemi.com/soc/techdocs/ds.

Brownout Voltage

Brownout is a condition in which the voltage supplies are lower than normal, causing the device to malfunction as a result of insufficient power. In general, Microsemi does not guarantee the functionality of the design inside the flash FPGA if voltage supplies are below their minimum recommended operating condition. Microsemi has performed measurements to characterize the brownout levels of FPGA power supplies. Refer to Table 17-3 for device-specific brownout deactivation levels. For the purpose of characterization, a direct path from the device input to output is monitored while voltage supplies are lowered gradually. The brownout point is defined as the voltage level at which the output stops following the input. Characterization tests performed on several IGLOO, ProASIC3L, and ProASIC3 devices in typical operating conditions showed the brownout voltage levels to be within the specification.

During device power-down, the device I/Os become tristated once the first supply in the power-down sequence drops below its brownout deactivation voltage.

Table 17-3 • Brownout Deactivation Levels for VCC and VCCI

Devices	VCC Brownout Deactivation Level (V)	VCCI Brownout Deactivation Level (V)
ProASIC3, ProASIC3 nano, IGLOO, IGLOO nano, IGLOO PLUS and ProASIC3L devices running at VCC = 1.5 V	0.75 V ± 0.25 V	0.8 V ± 0.3 V
IGLOO, IGLOO nano, IGLOO PLUS, and ProASIC3L devices running at VCC = 1.2 V	0.75 V ± 0.2 V	0.8 V ± 0.15 V

PLL Behavior at Brownout Condition

When PLL power supply voltage and/or V_{CC} levels drop below the V_{CC} brownout levels mentioned above for 1.5 V and 1.2 V devices, the PLL output lock signal goes LOW and/or the output clock is lost. The following sections explain PLL behavior during and after the brownout condition.

VCCPLL and VCC Tied Together

In this condition, both VCC and VCCPLL drop below the 0.75 V (\pm 0.25 V or \pm 0.2 V) brownout level. During the brownout recovery, once VCCPLL and VCC reach the activation point (0.85 \pm 0.25 V or \pm 0.2 V) again, the PLL output lock signal may still remain LOW with the PLL output clock signal toggling. If this condition occurs, there are two ways to recover the PLL output lock signal:

- 1. Cycle the power supplies of the PLL (power off and on) by using the PLL POWERDOWN signal.
- 2. Turn off the input reference clock to the PLL and then turn it back on.

Only VCCPLL Is at Brownout

In this case, only VCCPLL drops below the 0.75 V (\pm 0.25 V or \pm 0.2 V) brownout level and the VCC supply remains at nominal recommended operating voltage (1.5 V \pm 0.075 V for 1.5 V devices and 1.2 V \pm 0.06 V for 1.2 V devices). In this condition, the PLL behavior after brownout recovery is similar to initial power-up condition, and the PLL will regain lock automatically after VCCPLL is ramped up above the activation level (0.85 \pm 0.25 V or \pm 0.2 V). No intervention is necessary in this case.

Only VCC Is at Brownout

In this condition, VCC drops below the 0.75 V (\pm 0.25 V or \pm 0.2 V) brownout level and VCCPLL remains at nominal recommended operating voltage (1.5 V \pm 0.075 V for 1.5 V devices and 1.2 V \pm 0.06 V for 1.2 V devices). During the brownout recovery, once VCC reaches the activation point again (0.85 \pm 0.25 V or \pm 0.2 V), the PLL output lock signal may still remain LOW with the PLL output clock signal toggling. If this condition occurs, there are two ways to recover the PLL output lock signal:

- 1. Cycle the power supplies of the PLL (power off and on) by using the PLL POWERDOWN signal.
- 2. Turn off the input reference clock to the PLL and then turn it back on.

It is important to note that Microsemi recommends using a monotonic power supply or voltage regulator to ensure proper power-up behavior.

Microsemi

Index

architecture of user nonvolatile 117 configuration 120 custom serialization 129 design flow 124 generation 125 programming and accessing 122 programming file 127 programming files 267 SmartGen 126 FlashROM read-back 305	ISP 223, 224 architecture 261 board-level considerations 271 circuit 277 microprocessor 283 J JTAG 1532 261 JTAG interface 285	
G	L	
global architecture 31	layout	
global buffers	device-specific 78	
no programmable delays 64	LTC3025 linear voltage regulator 277	
with PLL function 67		
with programmable delays 64	M	
global macros	MAC validation/authentication 288	
Synplicity 50	macros	
globals	CLKBUF 77	
designer flow 53	CLKBUF_LVDS/LVPECL 77	
networks 58	CLKDLY 65, 73	
spines and rows 41	FIFO4KX18 141	
·	PLL 73	
Н	PLL macro signal descriptions 68	
HLD code	RAM4K9 137	
instantiating 192	RAM512X18 139	
hot-swap 167	supported basic RAM macros 136 UJTAG 299	
hot-swapping 317		
	MCU FPGA programming model 286 memory availability 146	
I	memory blocks 135	
I/O banks	microprocessor programming 283	
standards 40	Microsemi SoC Products Group	
standards compatibility 162	email 321	
I/O standards 77	web-based technical support 321	
global macros 46	website 321	
single-ended 166	WODONG 021	
I/Os	0	
assigning technologies 198	OTP 223	
assignments defined in PDC file 193	output slew rate 175	
automatically assigning 202	output siew rate 175	
behavior at power-up/-down 311	D	
board-level considerations 181	P	
buffer schematic cell 191	PDC	
cell architecture 207	global promotion and demotion 51	
configuration with SmartGen 188	place-and-route 193	
features 163, 164, 167	PLL	
global, naming 35	behavior at brownout condition 315 configuration bits 90	
manually assigning technologies 198 nano standard 162	corniguration bits 90 core specifications 84	
register combining 174	dynamic PLL configuration 87	
simplified buffer circuitry 165	functional description 85	
software support 177	power supply decoupling scheme 112	
software-controlled attributes 187	PLL block signals 68	
user I/O assignment flow chart 185	PLL macro block diagram 69	
user naming convention 178	product support	
wide range support 166	customer service 321	

email 321 My Cases 322 outside the U.S. 322	spine assignment 52 SRAM features 137	
technical support 321 website 321	initializing 148 software support 154	
programmers 225	usage 141	
device support 228	SSOs 176	
programming	STAPL player 285	
AES encryption 253	STAPL vs. DirectC 287	
basics 223	switching circuit 278	
features 223	verification 278	
file header definition 257	synthesizing 192	
flash and antifuse 225	Synthesizing 192	
flash devices 223	T	
glossary 258	Τ	
guidelines for flash programming 229	TAP controller state machine 291, 300	
header pin numbers 270	tech support	
microprocessor 283	ITAR 322	
power supplies 263	My Cases 322	
security 247	outside the U.S. 322	
solution 268	technical support 321	
solutions 227	transient current	
voltage 263	VCC 310	
volume services 226	VCCI 310	
programming support 221	transient current, power-up/-down 309	
R	U	
RAM	UJTAG	
memory block consumption 147	CCC dynamic reconfiguration 302	
remote upgrade via TCP/IP 288	fine tuning 303	
routing structure 18	macro 299	
rouning of dotars 10	operation 300	
S	port usage 301	
	use to read FlashROM contents 297	
Schmitt trigger 174	ultra-fast local lines 18	
security 264		
architecture 237	V	
encrypted programming 288	variable aspect ratio and cascading 145	
examples 242 features 238	VersaNet global networks 33	
FlashLock 241	VersaTile 15	
FlashROM 121	very-long-line resources 19	
FlashROM use models 245	ViewDraw 191	
in programmable logic 235	VREF pins	
overview 235	manually assigning 199	
signal integrity problem 271	manusmy seesgrang ree	
silicon testing 304	W	
single tile designs 159	weak pull-down 175	
SmartGen 154	weak pull-down 175 weak pull-up 175	
spine architecture 41	weak pull-up 175 web-based technical support 321	
	web-based technical support 321	