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### Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

### Applications of Embedded - FPGAs

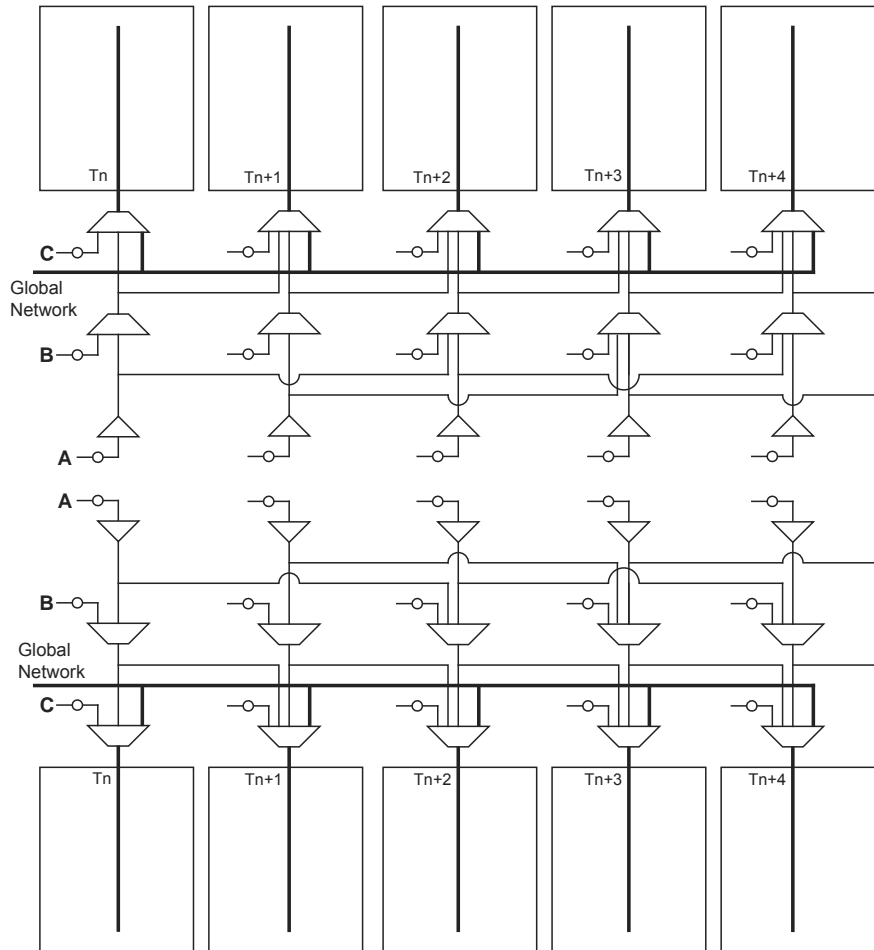
The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

#### Details

Product Status	Obsolete
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	36864
Number of I/O	71
Number of Gates	125000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	-20°C ~ 85°C (TJ)
Package / Case	100-TQFP
Supplier Device Package	100-VQFP (14x14)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/microchip-technology/a3pn125-zvq100">https://www.e-xfl.com/product-detail/microchip-technology/a3pn125-zvq100</a>

## Spine Access

The physical location of each spine is identified by the letter T (top) or B (bottom) and an accompanying number ( $Tn$  or  $Bn$ ). The number  $n$  indicates the horizontal location of the spine; 1 refers to the first spine on the left side of the die. Since there are six chip spines in each spine tree, there are up to six spines available for each combination of T (or B) and  $n$  (for example, six  $T1$  spines). Similarly, there are three quadrant spines available for each combination of T (or B) and  $n$  (for example, four  $T1$  spines), as shown in Figure 3-7.



**Figure 3-7 • Chip Global Aggregation**

A spine is also called a local clock network, and is accessed by the dedicated global MUX architecture. These MUXes define how a particular spine is driven. Refer to Figure 3-8 on page 44 for the global MUX architecture. The MUXes for each chip global spine are located in the middle of the die. Access to the top and bottom chip global spine is available from the middle of the die. There is no control dependency between the top and bottom spines. If a top spine,  $T1$ , of a chip global network is assigned to a net,  $B1$  is not wasted and can be used by the global clock network. The signal assigned only to the top or bottom spine cannot access the middle two rows of the architecture. However, if a spine is using the top and bottom at the same time ( $T1$  and  $B1$ , for instance), the previous restriction is lifted.

The MUXes for each quadrant global spine are located in the north and south sides of the die. Access to the top and bottom quadrant global spines is available from the north and south sides of the die. Since the MUXes for quadrant spines are located in the north and south sides of the die, you should not try to drive  $T1$  and  $B1$  quadrant spines from the same signal.

During Layout, Designer will assign two of the signals to quadrant global locations.

### Step 3 (optional)

You can also assign the QCLK1\_c and QCLK2\_c nets to quadrant regions using the following PDC commands:

```
assign_local_clock -net QCLK1_c -type quadrant UL
assign_local_clock -net QCLK2_c -type quadrant LL
```

### Step 4

Import this PDC with the netlist and run Compile again. You will see the following in the Compile report:

The following nets have been assigned to a global resource:

Fanout	Type	Name
1536	INT_NET	Net : EN_ALL_c Driver: EN_ALL_pad_CLKINT Source: AUTO PROMOTED
1536	SET/RESET_NET	Net : ACLR_c Driver: ACLR_pad_CLKINT Source: AUTO PROMOTED
256	CLK_NET	Net : QCLK3_c Driver: QCLK3_pad_CLKINT Source: AUTO PROMOTED
256	CLK_NET	Net : \$1N14 Driver: \$1I5/Core Source: ESSENTIAL
256	CLK_NET	Net : \$1N12 Driver: \$1I6/Core Source: ESSENTIAL
256	CLK_NET	Net : \$1N10 Driver: \$1I6/Core Source: ESSENTIAL

The following nets have been assigned to a quadrant clock resource using PDC:

Fanout	Type	Name
256	CLK_NET	Net : QCLK1_c Driver: QCLK1_pad_CLKINT Region: quadrant_UL
256	CLK_NET	Net : QCLK2_c Driver: QCLK2_pad_CLKINT Region: quadrant_LL

### Step 5

Run Layout.

## Global Management in PLL Design

This section describes the legal global network connections to PLLs in the low power flash devices. For detailed information on using PLLs, refer to "Clock Conditioning Circuits in Low Power Flash Devices and Mixed Signal FPGAs" section on page 61. Microsemi recommends that you use the dedicated global pins to directly drive the reference clock input of the associated PLL for reduced propagation delays and clock distortion. However, low power flash devices offer the flexibility to connect other signals to reference clock inputs. Each PLL is associated with three global networks (Figure 3-5 on page 36). There are some limitations, such as when trying to use the global and PLL at the same time:

- If you use a PLL with only primary output, you can still use the remaining two free global networks.
- If you use three globals associated with a PLL location, you cannot use the PLL on that location.
- If the YB or YC output is used standalone, it will occupy one global, even though this signal does not go to the global network.

## 4 – Clock Conditioning Circuits in Low Power Flash Devices and Mixed Signal FPGAs

### Introduction

This document outlines the following device information: Clock Conditioning Circuit (CCC) features, PLL core specifications, functional descriptions, software configuration information, detailed usage information, recommended board-level considerations, and other considerations concerning clock conditioning circuits and global networks in low power flash devices or mixed signal FPGAs.

### Overview of Clock Conditioning Circuitry

In Fusion, IGLOO, and ProASIC3 devices, the CCCs are used to implement frequency division, frequency multiplication, phase shifting, and delay operations. The CCCs are available in six chip locations—each of the four chip corners and the middle of the east and west chip sides. For device-specific variations, refer to the "Device-Specific Layout" section on page 78.

The CCC is composed of the following:

- PLL core
- 3 phase selectors
- 6 programmable delays and 1 fixed delay that advances/delays phase
- 5 programmable frequency dividers that provide frequency multiplication/division (not shown in Figure 4-6 on page 71 because they are automatically configured based on the user's required frequencies)
- 1 dynamic shift register that provides CCC dynamic reconfiguration capability

Figure 4-1 provides a simplified block diagram of the physical implementation of the building blocks in each of the CCCs.

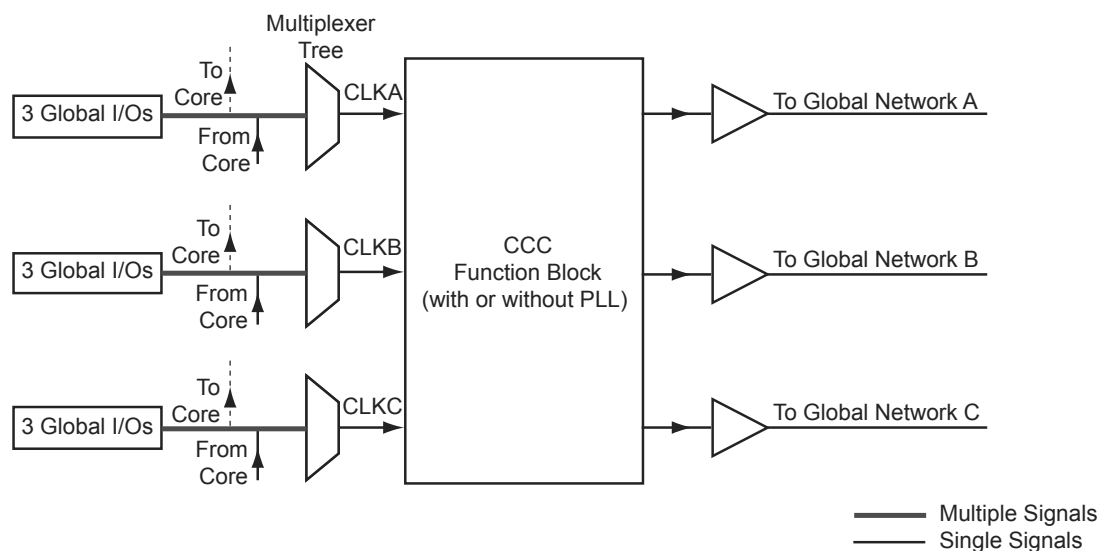


Figure 4-1 • Overview of the CCCs Offered in Fusion, IGLOO, and ProASIC3



YB and YC are identical to GLB and GLC, respectively, with the exception of a higher selectable final output delay. The SmartGen PLL Wizard will configure these outputs according to user specifications and can enable these signals with or without the enabling of Global Output Clocks.

The above signals can be enabled in the following output groupings in both internal and external feedback configurations of the static PLL:

- One output – GLA only
- Two outputs – GLA + (GLB and/or YB)
- Three outputs – GLA + (GLB and/or YB) + (GLC and/or YC)

## PLL Macro Block Diagram

As illustrated, the PLL supports three distinct output frequencies from a given input clock. Two of these (GLB and GLC) can be routed to the B and C global network access, respectively, and/or routed to the device core (YB and YC).

There are five delay elements to support phase control on all five outputs (GLA, GLB, GLC, YB, and YC).

There are delay elements in the feedback loop that can be used to advance the clock relative to the reference clock.

The PLL macro reference clock can be driven in the following ways:

1. By an INBUF\* macro to create a composite macro, where the I/O macro drives the global buffer (with programmable delay) using a hardwired connection. In this case, the I/O must be placed in one of the dedicated global I/O locations.
2. Directly from the FPGA core.
3. From an I/O that is routed through the FPGA regular routing fabric. In this case, users must instantiate a special macro, PLLINT, to differentiate from the hardwired I/O connection described earlier.

During power-up, the PLL outputs will toggle around the maximum frequency of the voltage-controlled oscillator (VCO) gear selected. Toggle frequencies can range from 40 MHz to 250 MHz. This will continue as long as the clock input (CLKA) is constant (HIGH or LOW). This can be prevented by LOW assertion of the POWERDOWN signal.

The visual PLL configuration in SmartGen, a component of the Libero SoC and Designer tools, will derive the necessary internal divider ratios based on the input frequency and desired output frequencies selected by the user.

When the external feedback (EXTFB) signal of the PLL in the ProASIC3/E devices is implemented, the phase detector of the PLL core receives the reference clock (CLKA) and EXTFB as inputs. EXTFB must be sourced as an INBUF macro and located at the global/chip clock location associated with the target PLL by Designer software. EXTFB cannot be sourced from the FPGA fabric.

The reference clock, CLKA, can be assigned on GxA0 or GxA1.

External Feedback (EXTFB) signal is assigned on GxB1 by Designer automatically.

To Core

Source for CCC (CLKA or CLKB or CLKC)

Routed Clock (from FPGA core)

To Core

Source for CCC (CLKA or CLKB or CLKC)

Routed Clock (from FPGA core)

x represents global location; can be A, B, C, D, E, or F

**Figure 4-5 • CLKA and EXTFB Assigned to Global I/Os**

## Phase Adjustment

The four phases available (0, 90, 180, 270) are phases with respect to VCO (PLL output). The VCO is divided to achieve the user's CCC required output frequency (GLA, YB/GLB, YC/GLC). The division happens after the selection of the VCO phase. The effective phase shift is actually the VCO phase shift divided by the output divider. This is why the visual CCC shows both the actual achievable phase and more importantly the actual delay that is equivalent to the phase shift that can be achieved.

## Dynamic PLL Configuration

The CCCs can be configured both statically and dynamically.

In addition to the ports available in the Static CCC, the Dynamic CCC has the dynamic shift register signals that enable dynamic reconfiguration of the CCC. With the Dynamic CCC, the ports CLKB and CLKC are also exposed. All three clocks (CLKA, CLKB, and CLKC) can be configured independently.

The CCC block is fully configurable. The following two sources can act as the CCC configuration bits.

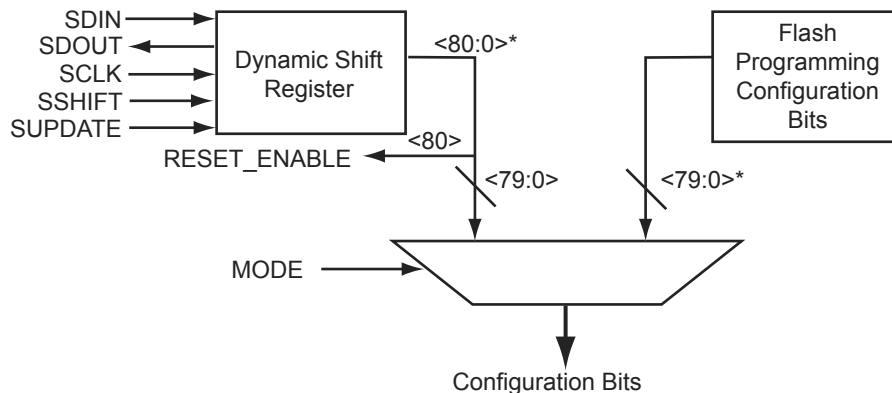
### Flash Configuration Bits

The flash configuration bits are the configuration bits associated with programmed flash switches. These bits are used when the CCC is in static configuration mode. Once the device is programmed, these bits cannot be modified. They provide the default operating state of the CCC.

### Dynamic Shift Register Outputs

This source does not require core reprogramming and allows core-driven dynamic CCC reconfiguration. When the dynamic register drives the configuration bits, the user-defined core circuit takes full control over SDIN, SDOUT, SCLK, SSHIFT, and SUPDATE. The configuration bits can consequently be dynamically changed through shift and update operations in the serial register interface. Access to the logic core is accomplished via the dynamic bits in the specific tiles assigned to the PLLs.

Figure 4-21 illustrates a simplified block diagram of the MUX architecture in the CCCs.



*Note: \*For Fusion, bit <88:81> is also needed.*

**Figure 4-21 • The CCC Configuration MUX Architecture**

The selection between the flash configuration bits and the bits from the configuration register is made using the MODE signal shown in Figure 4-21. If the MODE signal is logic HIGH, the dynamic shift register configuration bits are selected. There are 81 control bits to configure the different functions of the CCC.

**Table 4-13 • 2-Bit Feedback MUX**

FBSEL<1:0> State	MUX Input Selected
0	Ground. Used for power-down mode in power-down logic block.
1	PLL VCO 0° phase shift
2	PLL delayed VCO 0° phase shift
3	N/A

**Table 4-14 • Programmable Delay Selection for Feedback Delay and Secondary Core Output Delays**

FBDLY<4:0>; DLYYB<4:0>; DLYYC<4:0> State	Delay Value
0	Typical delay = 600 ps
1	Typical delay = 760 ps
2	Typical delay = 920 ps
⋮	⋮
31	Typical delay = 5.56 ns

**Table 4-15 • Programmable Delay Selection for Global Clock Output Delays**

DLYGLA<4:0>; DLYGLB<4:0>; DLYGLC<4:0> State	Delay Value
0	Typical delay = 225 ps
1	Typical delay = 760 ps
2	Typical delay = 920 ps
⋮	⋮
31	Typical delay = 5.56 ns

**Table 4-16 • Fusion Dynamic CCC Clock Source Selection**

RXASEL	DYNASEL	Source of CLKA
1	0	RC Oscillator
1	1	Crystal Oscillator
RXBSEL	DYNBSEL	Source of CLKB
1	0	RC Oscillator
1	1	Crystal Oscillator
RXCSEL	DYNCSEL	Source of CLKC
1	0	RC Oscillator
1	1	Crystal Oscillator

**Table 4-17 • Fusion Dynamic CCC NGMUX Configuration**

GLMUXCFG<1:0>	NGMUX Select Signal	Supported Input Clocks to NGMUX
00	0	GLA
	1	GLC
01	0	GLA
	1	GLINT
10	0	GLC
	1	GLINT

The following is an example of a PLL configuration utilizing the clock frequency synthesis and clock delay adjustment features. The steps include generating the PLL core with SmartGen, performing simulation for verification with ModelSim, and performing static timing analysis with SmartTime in Designer.

Parameters of the example PLL configuration:

Input Frequency – 20 MHz

Primary Output Requirement – 20 MHz with clock advancement of 3.02 ns

Secondary 1 Output Requirement – 40 MHz with clock delay of 2.515 ns

Figure 4-29 shows the SmartGen settings. Notice that the overall delays are calculated automatically, allowing the user to adjust the delay elements appropriately to obtain the desired delays.

#### Figure 4-29 • SmartGen Settings

After confirming the correct settings, generate a structural netlist of the PLL and verify PLL core settings by checking the log file:

```
Name                : test_pll_delays
Family              : ProASIC3E
Output Format       : VHDL
Type               : Static PLL
Input Freq(MHz)    : 20.000
CLKA Source        : Hardwired I/O
Feedback Delay Value Index : 21
Feedback Mux Select : 2
XDLY Mux Select    : No
Primary Freq(MHz)  : 20.000
Primary PhaseShift : 0
Primary Delay Value Index : 1
Primary Mux Select : 4
Secondary1 Freq(MHz) : 40.000
Use GLB            : YES
Use YB             : NO
...
...
...
Primary Clock frequency 20.000
Primary Clock Phase Shift 0.000
```

## Example of RAM Initialization

This section of the document presents a sample design in which a 4×4 RAM block is being initialized through the JTAG port. A test feature has been implemented in the design to read back the contents of the RAM after initialization to verify the procedure.

The interface block of this example performs two major functions: initialization of the RAM block and running a test procedure to read back the contents. The clock output of the interface is either the write clock (for initialization) or the read clock (for reading back the contents). The Verilog code for the interface block is included in the "Sample Verilog Code" section on page 151.

For simulation purposes, users can declare the input ports of the UJTAG macro for easier assignment in the testbench. However, the UJTAG input ports should not be declared on the top level during synthesis. If the input ports of the UJTAG are declared during synthesis, the synthesis tool will instantiate input buffers on these ports. The input buffers on the ports will cause Compile to fail in Designer.

Figure 6-10 shows the simulation results for the initialization step of the example design.

The CLK\_OUT signal, which is the clock output of the interface block, is the inverted DR\_UPDATE output of the UJTAG macro. It is clear that it gives sufficient time (while the TAP Controller is in the Data Register Update state) for the write address and data to become stable before loading them into the RAM block.

Figure 6-11 presents the test procedure of the example. The data read back from the memory block matches the written data, thus verifying the design functionality.

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**Figure 6-10 • Simulation of Initialization Step**

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**Figure 6-11 • Simulation of the Test Procedure of the Example**

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## User I/O Naming Convention

Due to the comprehensive and flexible nature of nano Standard I/Os, a naming scheme is used to show the details of each I/O (Figure 7-8). The name identifies to which I/O bank it belongs.

I/O Nomenclature = FF/Gmn/IOuxwBy

Gmn is only used for I/Os that also have CCC access—i.e., global pins.

FF = Indicates the I/O dedicated for the Flash\*Freeze mode activation pin

G = Global

m = Global pin location associated with each CCC on the device: A (northwest corner), B (northeast corner), C (east middle), D (southeast corner), E (southwest corner), and F (west middle)

n = Global input MUX and pin number of the associated Global location m—either A0, A1, A2, B0, B1, B2, C0, C1, or C2. Refer to the "Global Resources in Low Power Flash Devices" section on page 31 for information about the three input pins per clock source MUX at CCC location m.

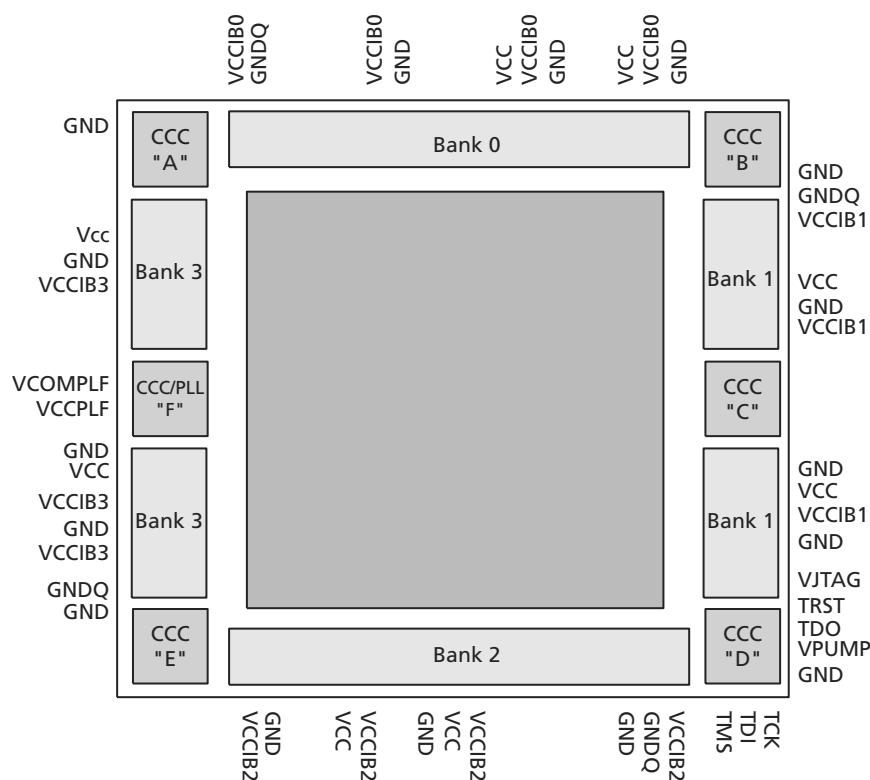
u = I/O pair number in the bank, starting at 00 from the northwest I/O bank and proceeding in a clockwise direction

x = R (Regular—single-ended) for the I/Os that support single-ended standards.

w = S (Single-Ended)

B = Bank

y = Bank number (0–3). The Bank number starts at 0 from the northwest I/O bank and proceeds in a clockwise direction.

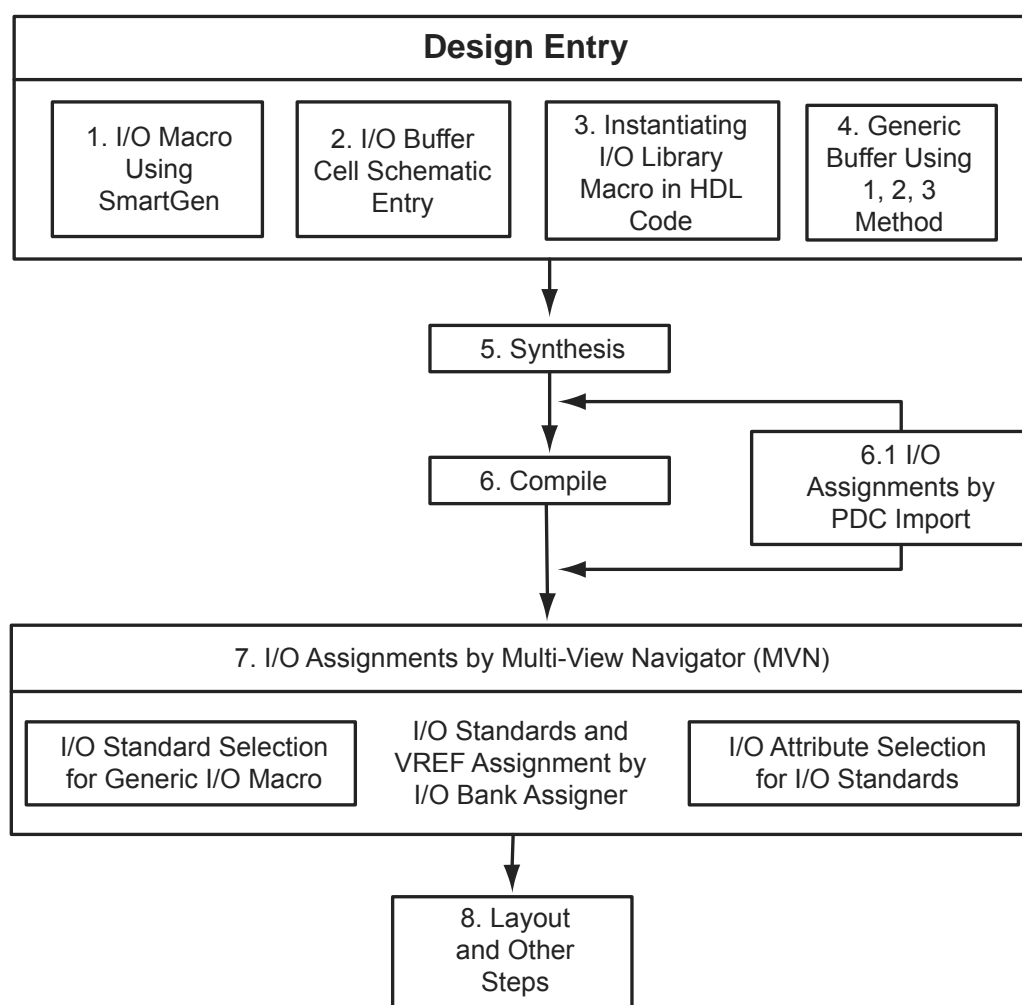


**Figure 7-8 • I/O Naming Conventions for nano Devices – Top View**

## 8 – I/O Software Control in Low Power Flash Devices

Fusion, IGLOO, and ProASIC3 I/Os provide more design flexibility, allowing the user to control specific features by enabling certain I/O standards. Some features are selectable only for certain I/O standards, whereas others are available for all I/O standards. For example, slew control is not supported by differential I/O standards. Conversely, I/O register combining is supported by all I/O standards. For detailed information about which I/O standards and features are available on each device and each I/O type, refer to the I/O Structures section of the handbook for the device you are using.

Figure 8-1 shows the various points in the software design flow where a user can provide input or control of the I/O selection and parameters. A detailed description is provided throughout this document.



**Figure 8-1 • User I/O Assignment Flow Chart**



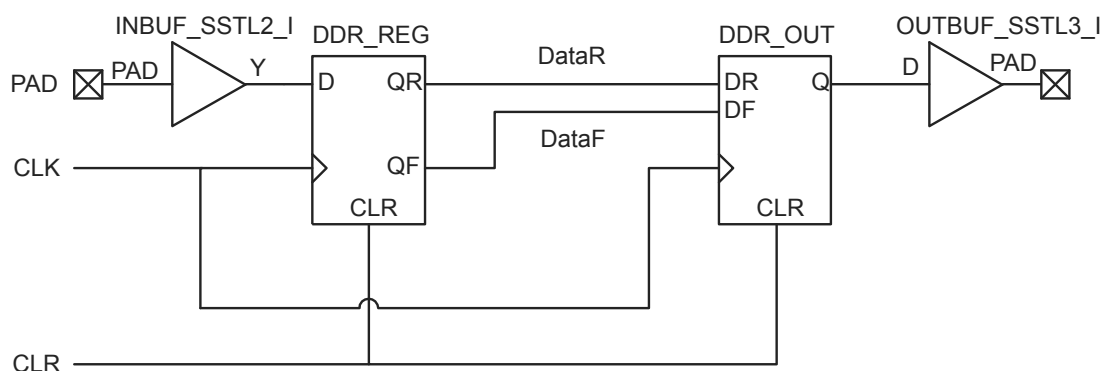
## 9 – DDR for Microsemi's Low Power Flash Devices

### Introduction

The I/Os in Fusion, IGLOO, and ProASIC3 devices support Double Data Rate (DDR) mode. In this mode, new data is present on every transition (or clock edge) of the clock signal. This mode doubles the data transfer rate compared with Single Data Rate (SDR) mode, where new data is present on one transition (or clock edge) of the clock signal. Low power flash devices have DDR circuitry built into the I/O tiles. I/Os are configured to be DDR receivers or transmitters by instantiating the appropriate special macros (examples shown in Figure 9-4 on page 210 and Figure 9-5 on page 211) and buffers (DDR\_OUT or DDR\_REG) in the RTL design. This document discusses the options the user can choose to configure the I/Os in this mode and how to instantiate them in the design.

### Double Data Rate (DDR) Architecture

Low power flash devices support 350 MHz DDR inputs and outputs. In DDR mode, new data is present on every transition of the clock signal. Clock and data lines have identical bandwidths and signal integrity requirements, making them very efficient for implementing very high-speed systems. High-speed DDR interfaces can be implemented using LVDS (not applicable for IGLOO nano and ProASIC3 nano devices). In IGLOOe, ProASIC3E, AFS600, and AFS1500 devices, DDR interfaces can also be implemented using the HSTL, SSTL, and LVPECL I/O standards. The DDR feature is primarily implemented in the FPGA core periphery and is not tied to a specific I/O technology or limited to any I/O standard.



**Figure 9-1 • DDR Support in Low Power Flash Devices**

### ***Signal Integrity While Using ISP***

For ISP of flash devices, customers are expected to follow the board-level guidelines provided on the Microsemi SoC Products Group website. These guidelines are discussed in the datasheets and application notes (refer to the “Related Documents” section of the datasheet for application note links). Customers are also expected to troubleshoot board-level signal integrity issues by measuring voltages and taking oscilloscope plots.

### **Programming Failure Allowances**

Microsemi has strict policies regarding programming failure allowances. Please refer to *Programming and Functional Failure Guidelines* on the Microsemi SoC Products Group website for details.

### **Contacting the Customer Support Group**

Highly skilled engineers staff the Customer Applications Center from 7:00 A.M. to 6:00 P.M., Pacific time, Monday through Friday. You can contact the center by one of the following methods:

#### ***Electronic Mail***

You can communicate your technical questions to our email address and receive answers back by email, fax, or phone. Also, if you have design problems, you can email your design files to receive assistance. Microsemi monitors the email account throughout the day. When sending your request to us, please be sure to include your full name, company name, and contact information for efficient processing of your request. The technical support email address is [soc\\_tech@microsemi.com](mailto:soc_tech@microsemi.com).

#### ***Telephone***

Our Technical Support Hotline answers all calls. The center retrieves information, such as your name, company name, telephone number, and question. Once this is done, a case number is assigned. Then the center forwards the information to a queue where the first available applications engineer receives the data and returns your call. The phone hours are from 7:00 A.M. to 6:00 P.M., Pacific time, Monday through Friday.

The Customer Applications Center number is (800) 262-1060.

European customers can call +44 (0) 1256 305 600.

## Generating Programming Files

### Generation of the Programming File in a Trusted Environment— Application 1

As discussed in the "Application 1: Trusted Environment" section on page 243, in a trusted environment, the user can choose to program the device with plaintext bitstream content. It is possible to use plaintext for programming even when the FlashLock Pass Key option has been selected. In this application, it is not necessary to employ AES encryption protection. For AES encryption settings, refer to the next sections.

The generated programming file will include the security setting (if selected) and the plaintext programming file content for the FPGA array, FlashROM, and/or FBs. These options are indicated in Table 11-2 and Table 11-3.

**Table 11-2 • IGLOO and ProASIC3 Plaintext Security Options, No AES**

Security Protection	FlashROM Only	FPGA Core Only	Both FlashROM and FPGA
No AES / no FlashLock	✓	✓	✓
FlashLock only	✓	✓	✓
AES and FlashLock	–	–	–

**Table 11-3 • Fusion Plaintext Security Options**

Security Protection	FlashROM Only	FPGA Core Only	FB Core Only	All
No AES / no FlashLock	✓	✓	✓	✓
FlashLock	✓	✓	✓	✓
AES and FlashLock	–	–	–	–

*Note: For all instructions, the programming of Flash Blocks refers to Fusion only.*

For this scenario, generate the programming file as follows:

1. Select the **Silicon features to be programmed** (Security Settings, FPGA Array, FlashROM, Flash Memory Blocks), as shown in Figure 11-10 on page 248 and Figure 11-11 on page 248. Click **Next**.

If **Security Settings** is selected (i.e., the FlashLock security Pass Key feature), an additional dialog will be displayed to prompt you to select the security level setting. If no security setting is selected, you will be directed to Step 3.

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**Figure 11-10 • All Silicon Features Selected for IGLOO and ProASIC3 Devices**

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**Figure 11-11 • All Silicon Features Selected for Fusion**

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## Security in ARM-Enabled Low Power Flash Devices

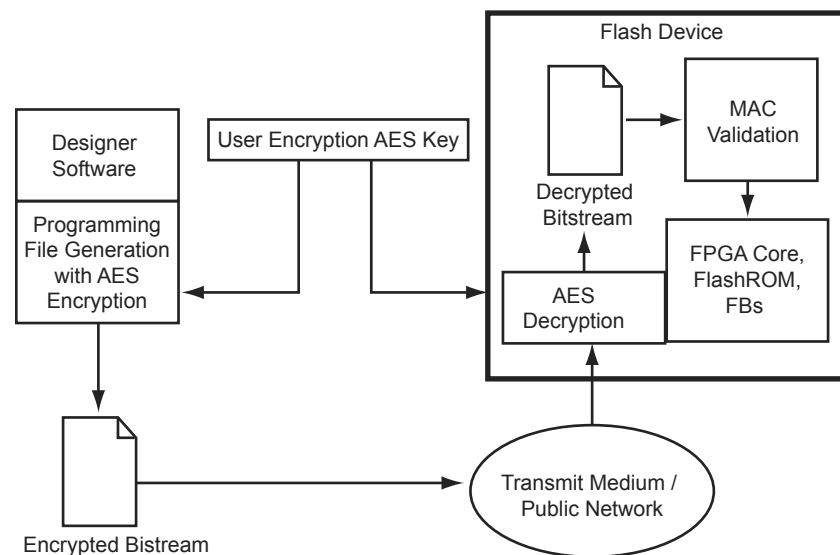
There are slight differences between the regular flash device and the ARM-enabled flash devices, which have the M1 prefix.

The AES key is used by Microsemi and preprogrammed into the device to protect the ARM IP. As a result, the design will be encrypted along with the ARM IP, according to the details below.

### Cortex-M1 and Cortex-M3 Device Security

Cortex-M1-enabled and Cortex-M3 devices are shipped with the following security features:

- FPGA array enabled for AES-encrypted programming and verification
- FlashROM enabled for AES-encrypted write and verify
- Embedded Flash Memory enabled for AES encrypted write



**Figure 12-1 • AES-128 Security Features**

## FlashROM and Programming Files

Each low power flash device has 1 kbit of on-chip, nonvolatile flash memory that can be accessed from the FPGA core. This nonvolatile FlashROM is arranged in eight pages of 128 bits (Figure 12-3). Each page can be programmed independently, with or without the 128-bit AES encryption. The FlashROM can only be programmed via the IEEE 1532 JTAG port and cannot be programmed from the FPGA core. In addition, during programming of the FlashROM, the FPGA core is powered down automatically by the on-chip programming control logic.

		Byte Number in Page															
		15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Page Number	7																
	6																
	5																
	4																
	3																
	2																
	1																
	0																

**Figure 12-3 • FlashROM Architecture**

When using FlashROM combined with AES, many subscription-based applications or device serialization applications are possible. The FROM configurator found in the Libero SoC Catalog supports easy management of the FlashROM contents, even over large numbers of devices. The FROM configurator can support FlashROM contents that contain the following:

- Static values
- Random numbers
- Values read from a file
- Independent updates of each page

In addition, auto-incrementing of fields is possible. In applications where the FlashROM content is different for each device, you have the option to generate a single STAPL file for all the devices or individual serialization files for each device. For more information on how to generate the FlashROM content for device serialization, refer to the "FlashROM in Microsemi's Low Power Flash Devices" section on page 117.

Libero SoC includes a unique tool to support the generation and management of FlashROM and FPGA programming files. This tool is called FlashPoint.

Depending on the applications, designers can use the FlashPoint software to generate a STAPL file with different contents. In each case, optional AES encryption and/or different security settings can be set.

In Designer, when you click the Programming File icon, FlashPoint launches, and you can generate STAPL file(s) with four different cases (Figure 12-4 on page 268). When the serialization feature is used during the configuration of FlashROM, you can generate a single STAPL file that will program all the devices or an individual STAPL file for each device.

The following cases present the FPGA core and FlashROM programming file combinations that can be used for different applications. In each case, you can set the optional security settings (FlashLock Pass Key and/or AES Key) depending on the application.

1. A single STAPL file or multiple STAPL files with multiple FlashROM contents and the FPGA core content. A single STAPL file will be generated if the device serialization feature is not used. You can program the whole FlashROM or selectively program individual pages.
2. A single STAPL file for the FPGA core content

## List of Changes

The following table lists critical changes that were made in each revision of the chapter.

Date	Changes	Page
September 2012	The "Security" section was modified to clarify that Microsemi does not support read-back of FPGA core-programmed data (SAR 41235).	288
July 2010	This chapter is no longer published separately with its own part number and version but is now part of several FPGA fabric user's guides.	N/A
v1.4 (December 2008)	IGLOO nano and ProASIC3 nano devices were added to Table 14-1 • Flash-Based FPGAs.	284
v1.3 (October 2008)	The "Microprocessor Programming Support in Flash Devices" section was revised to include new families and make the information more concise.	284
v1.2 (June 2008)	The following changes were made to the family descriptions in Table 14-1 • Flash-Based FPGAs: <ul style="list-style-type: none"> <li>ProASIC3L was updated to include 1.5 V.</li> <li>The number of PLLs for ProASIC3E was changed from five to six.</li> </ul>	284
v1.1 (March 2008)	The "Microprocessor Programming Support in Flash Devices" section was updated to include information on the IGLOO PLUS family. The "IGLOO Terminology" section and "ProASIC3 Terminology" section are new.	284

## **B – Product Support**

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Microsemi SoC Products Group backs its products with various support services, including Customer Service, Customer Technical Support Center, a website, electronic mail, and worldwide sales offices. This appendix contains information about contacting Microsemi SoC Products Group and using these support services.

### **Customer Service**

Contact Customer Service for non-technical product support, such as product pricing, product upgrades, update information, order status, and authorization.

From North America, call 800.262.1060

From the rest of the world, call 650.318.4460

Fax, from anywhere in the world, 650.318.8044

### **Customer Technical Support Center**

Microsemi SoC Products Group staffs its Customer Technical Support Center with highly skilled engineers who can help answer your hardware, software, and design questions about Microsemi SoC Products. The Customer Technical Support Center spends a great deal of time creating application notes, answers to common design cycle questions, documentation of known issues, and various FAQs. So, before you contact us, please visit our online resources. It is very likely we have already answered your questions.

### **Technical Support**

Visit the Customer Support website ([www.microsemi.com/soc/support/search/default.aspx](http://www.microsemi.com/soc/support/search/default.aspx)) for more information and support. Many answers available on the searchable web resource include diagrams, illustrations, and links to other resources on the website.

### **Website**

You can browse a variety of technical and non-technical information on the SoC home page, at [www.microsemi.com/soc](http://www.microsemi.com/soc).

### **Contacting the Customer Technical Support Center**

Highly skilled engineers staff the Technical Support Center. The Technical Support Center can be contacted by email or through the Microsemi SoC Products Group website.

#### **Email**

You can communicate your technical questions to our email address and receive answers back by email, fax, or phone. Also, if you have design problems, you can email your design files to receive assistance. We constantly monitor the email account throughout the day. When sending your request to us, please be sure to include your full name, company name, and your contact information for efficient processing of your request.

The technical support email address is [soc\\_tech@microsemi.com](mailto:soc_tech@microsemi.com).



## My Cases

Microsemi SoC Products Group customers may submit and track technical cases online by going to My Cases.

## Outside the U.S.

Customers needing assistance outside the US time zones can either contact technical support via email ([soc\\_tech@microsemi.com](mailto:soc_tech@microsemi.com)) or contact a local sales office. Sales office listings can be found at [www.microsemi.com/soc/company/contact/default.aspx](http://www.microsemi.com/soc/company/contact/default.aspx).

## ITAR Technical Support

For technical support on RH and RT FPGAs that are regulated by International Traffic in Arms Regulations (ITAR), contact us via [soc\\_tech\\_itar@microsemi.com](mailto:soc_tech_itar@microsemi.com). Alternatively, within My Cases, select **Yes** in the ITAR drop-down list. For a complete list of ITAR-regulated Microsemi FPGAs, visit the ITAR web page.