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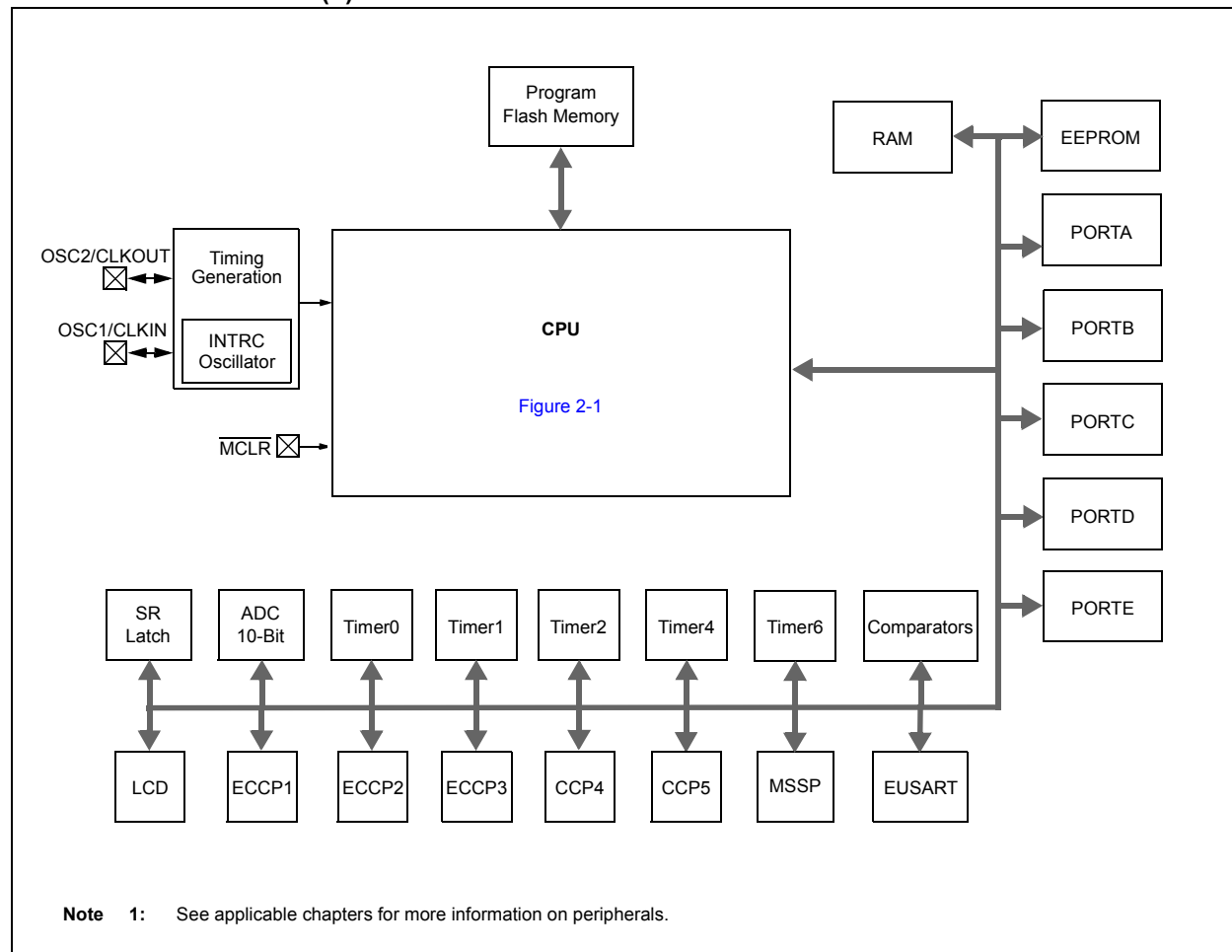
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Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	32MHz
Connectivity	I ² C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, LCD, POR, PWM, WDT
Number of I/O	25
Program Memory Size	28KB (16K x 14)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	1K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 11x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	28-UQFN Exposed Pad
Supplier Device Package	28-UQFN (4x4)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16lf1938-i-mv

PIC16(L)F1938/9

FIGURE 1-1: PIC16(L)F1938/9 BLOCK DIAGRAM



3.1.1.2 Indirect Read with FSR

The program memory can be accessed as data by setting bit 7 of the FSRxH register and reading the matching INDFx register. The `MOVIW` instruction will place the lower eight bits of the addressed word in the W register. Writes to the program memory cannot be performed via the INDF registers. Instructions that access the program memory via the FSR require one extra instruction cycle to complete. [Example 3-2](#) demonstrates accessing the program memory via an FSR.

The `HIGH` directive will set bit<7> if a label points to a location in program memory.

EXAMPLE 3-2: ACCESSING PROGRAM MEMORY VIA FSR

```
constants
    RETLW DATA0      ;Index0 data
    RETLW DATA1      ;Index1 data
    RETLW DATA2
    RETLW DATA3
my_function
    ;... LOTS OF CODE...
    MOVLW LOW constants
    MOVWF FSR1L
    MOVLW HIGH constants
    MOVWF FSR1H
    MOVIW 0[FSR1]
;THE PROGRAM MEMORY IS IN W
```

3.2.1 CORE REGISTERS

The core registers contain the registers that directly affect the basic operation of the PIC16(L)F1938/9. These registers are listed below:

- INDF0
- INDF1
- PCL
- STATUS
- FSR0 Low
- FSR0 High
- FSR1 Low
- FSR1 High
- BSR
- WREG
- PCLATH
- INTCON

Note: The core registers are the first 12 addresses of every data memory bank.

3.2 Data Memory Organization

The data memory is partitioned in 32 memory banks with 128 bytes in a bank. Each bank consists of ([Figure 3-2](#)):

- 12 core registers
- 20 Special Function Registers (SFR)
- Up to 80 bytes of General Purpose RAM (GPR)
- 16 bytes of common RAM

The active bank is selected by writing the bank number into the Bank Select Register (BSR). Unimplemented memory will read as '0'. All data memory can be accessed either directly (via instructions that use the file registers) or indirectly via the two File Select Registers (FSR). See [Section 3.6 "Indirect Addressing"](#) for more information.

Data Memory uses a 12-bit address. The upper seven bits of the address define the Bank address and the lower five bits select the registers/RAM in that bank.

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4.6 Device ID and Revision ID

The memory location 8006h is where the Device ID and Revision ID are stored. The upper nine bits hold the Device ID. The lower five bits hold the Revision ID. See [Section 11.5 “User ID, Device ID and Configuration Word Access”](#) for more information on accessing these memory locations.

Development tools, such as device programmers and debuggers, may be used to read the Device ID and Revision ID.

4.7 Register Definitions: Device ID

REGISTER 4-3: DEVIDEID: DEVICE ID REGISTER⁽¹⁾

R	R	R	R	R	R
DEV8	DEV7	DEV6	DEV5	DEV4	DEV3
bit 13			bit 8		

R	R	R	R	R	R	R	R
DEV2	DEV1	DEV0	REV4	REV3	REV2	REV1	REV0
bit 7			bit 0				

Legend:

R = Readable bit

'0' = Bit is cleared

'1' = Bit is set

-n = Value when blank or after Bulk Erase

bit 13-5 **DEV<8:0>**: Device ID bits

DEVICE	DEVIDEID<13:0>	
	DEV<8:0>	REV<4:0>
PIC16F1938	10 0011 101	X XXXX
PIC16F1939	10 0011 110	X XXXX
PIC16LF1938	10 0100 101	X XXXX
PIC16LF1939	10 0100 110	X XXXX

bit 4-0 **REV<4:0>**: Revision ID bits

These bits are used to identify the revision (see the table under DEV<8:0> above).

Note 1: This location cannot be written.

5.2.2.6 32 MHz Internal Oscillator Frequency Selection

The Internal Oscillator Block can be used with the 4X PLL associated with the External Oscillator Block to produce a 32 MHz internal system clock source. The following settings are required to use the 32 MHz internal clock source:

- The FOSC bits in Configuration Words must be set to use the INTOSC source as the device system clock (FOSC<2:0> = 100).
- The SCS bits in the OSCCON register must be cleared to use the clock determined by FOSC<2:0> in Configuration Words (SCS<1:0> = 00).
- The IRCF bits in the OSCCON register must be set to the 8 MHz HFINTOSC set to use (IRCF<3:0> = 1110).
- The SPLLEN bit in the OSCCON register must be set to enable the 4xPLL, or the PLEN bit of the Configuration Words must be programmed to a '1'.

Note: When using the PLEN bit of the Configuration Words, the 4xPLL cannot be disabled by software and the 8 MHz HFINTOSC option will no longer be available.

The 4xPLL is not available for use with the internal oscillator when the SCS bits of the OSCCON register are set to '1x'. The SCS bits must be set to '00' to use the 4xPLL with the internal oscillator.

5.2.2.7 Internal Oscillator Clock Switch Timing

When switching between the HFINTOSC, MFINTOSC and the LFINTOSC, the new oscillator may already be shut down to save power (see [Figure 5-7](#)). If this is the case, there is a delay after the IRCF<3:0> bits of the OSCCON register are modified before the frequency selection takes place. The OSCSTAT register will reflect the current active status of the HFINTOSC, MFINTOSC and LFINTOSC oscillators. The sequence of a frequency selection is as follows:

1. IRCF<3:0> bits of the OSCCON register are modified.
2. If the new clock is shut down, a clock start-up delay is started.
3. Clock switch circuitry waits for a falling edge of the current clock.
4. The current clock is held low and the clock switch circuitry waits for a rising edge in the new clock.
5. The new clock is now active.
6. The OSCSTAT register is updated as required.
7. Clock switch is complete.

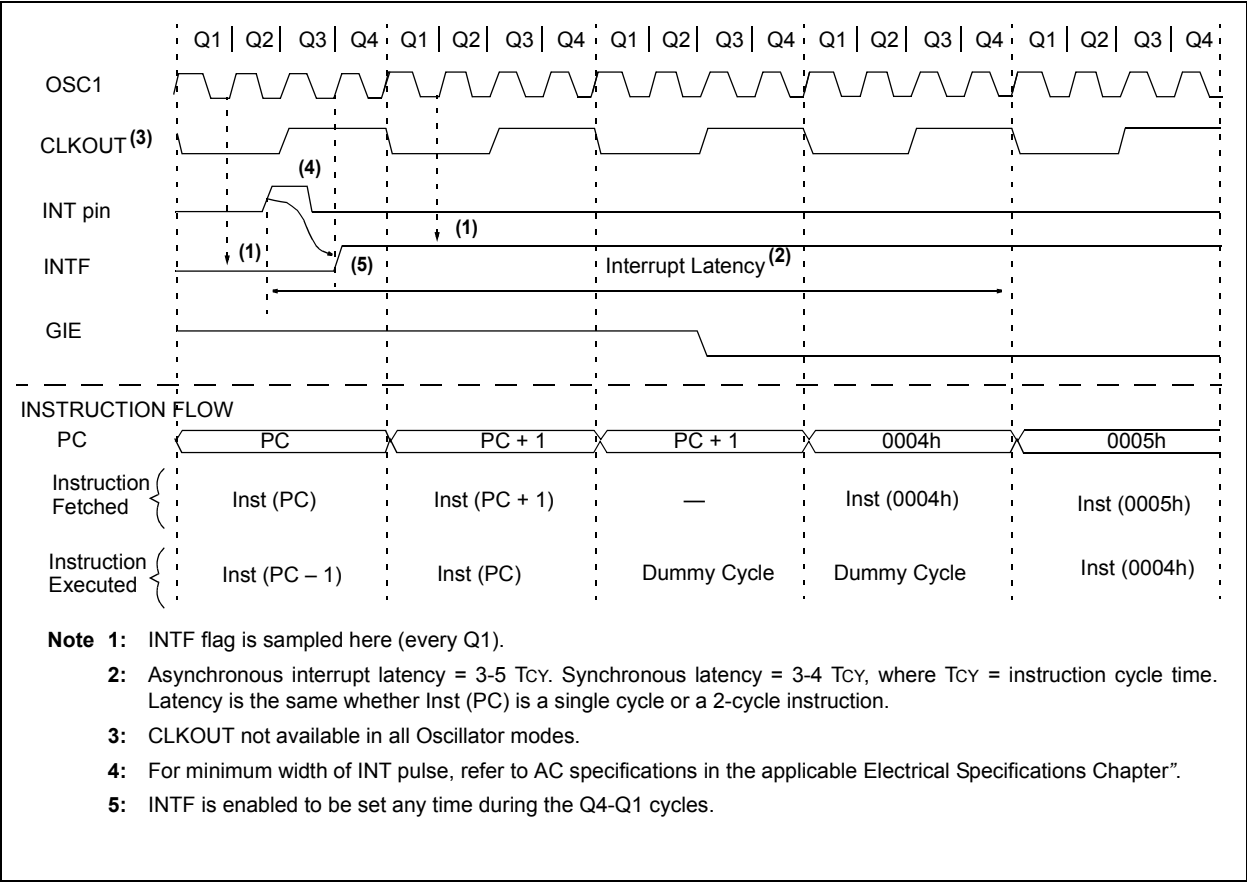
See [Figure 5-7](#) for more details.

If the internal oscillator speed is switched between two clocks of the same source, there is no start-up delay before the new frequency is selected. Clock switching time delays are shown in [Table 5-1](#).

Start-up delay specifications are located in the oscillator tables in the applicable Electrical Specifications Chapter.

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FIGURE 7-3: INT PIN INTERRUPT TIMING



EXAMPLE 11-5: WRITING TO FLASH PROGRAM MEMORY

```

; This write routine assumes the following:
; 1. The 16 bytes of data are loaded, starting at the address in DATA_ADDR
; 2. Each word of data to be written is made up of two adjacent bytes in DATA_ADDR,
;    stored in little endian format
; 3. A valid starting address (the least significant bits = 000) is loaded in ADDRH:ADDRL
; 4. ADDRH and ADDRL are located in shared data memory 0x70 - 0x7F (common RAM)
;
    BCF      INTCON,GIE      ; Disable ints so required sequences will execute properly
    BANKSEL  EEADRH          ; Bank 3
    MOVF     ADDRH,W         ; Load initial address
    MOVWF    EEADRH          ;
    MOVF     ADDRL,W         ;
    MOVWF    EEADRL         ;
    MOVLW    LOW DATA_ADDR  ; Load initial data address
    MOVWF    FSR0L           ;
    MOVLW    HIGH DATA_ADDR ; Load initial data address
    MOVWF    FSR0H           ;
    BSF      EECON1,EEPGD    ; Point to program memory
    BCF      EECON1,CFGSR    ; Not configuration space
    BSF      EECON1,WREN     ; Enable writes
    BSF      EECON1,LWLO     ; Only Load Write Latches

LOOP
    MOVIW    FSR0++          ; Load first data byte into lower
    MOVWF    EEDATL          ;
    MOVIW    FSR0++          ; Load second data byte into upper
    MOVWF    EEDATH          ;

    MOVF     EEADRL,W         ; Check if lower bits of address are '000'
    XORLW    0x07            ; Check if we're on the last of 8 addresses
    ANDLW    0x07            ;
    BTFSC    STATUS,Z        ; Exit if last of eight words,
    GOTO     START_WRITE     ;

    Required Sequence
    MOVLW    55h              ; Start of required write sequence:
    MOVWF    EECON2           ; Write 55h
    MOVLW    0AAh            ;
    MOVWF    EECON2           ; Write AAh
    BSF      EECON1,WR        ; Set WR bit to begin write
    NOP                      ; Any instructions here are ignored as processor
                                ; halts to begin write sequence
    NOP                      ; Processor will stop here and wait for write to complete.

                                ; After write processor continues with 3rd instruction.

    INCF     EEADRL,F         ; Still loading latches Increment address
    GOTO     LOOP            ; Write next latches

START_WRITE
    BCF      EECON1,LWLO     ; No more loading latches - Actually start Flash program
                                ; memory write

    Required Sequence
    MOVLW    55h              ; Start of required write sequence:
    MOVWF    EECON2           ; Write 55h
    MOVLW    0AAh            ;
    MOVWF    EECON2           ; Write AAh
    BSF      EECON1,WR        ; Set WR bit to begin write
    NOP                      ; Any instructions here are ignored as processor
                                ; halts to begin write sequence
    NOP                      ; Processor will stop here and wait for write complete.

                                ; after write processor continues with 3rd instruction
    BCF      EECON1,WREN     ; Disable writes
    BSF      INTCON,GIE      ; Enable interrupts

```



15.2 ADC Operation

15.2.1 STARTING A CONVERSION

To enable the ADC module, the ADON bit of the ADCON0 register must be set to a '1'. Setting the GO/DONE bit of the ADCON0 register to a '1' will start the Analog-to-Digital conversion.

Note: The GO/DONE bit should not be set in the same instruction that turns on the ADC. Refer to [Section 15.2.6 “A/D Conversion Procedure”](#).

15.2.2 COMPLETION OF A CONVERSION

When the conversion is complete, the ADC module will:

- Clear the GO/DONE bit
- Set the ADIF Interrupt Flag bit
- Update the ADRESH and ADRESL registers with new conversion result

15.2.3 TERMINATING A CONVERSION

If a conversion must be terminated before completion, the GO/DONE bit can be cleared in software. The ADRESH and ADRESL registers will be updated with the partially complete Analog-to-Digital conversion sample. Incomplete bits will match the last bit converted.

Note: A device Reset forces all registers to their Reset state. Thus, the ADC module is turned off and any pending conversion is terminated.

15.2.4 ADC OPERATION DURING SLEEP

The ADC module can operate during Sleep. This requires the ADC clock source to be set to the FRC option. When the FRC clock source is selected, the ADC waits one additional instruction before starting the conversion. This allows the SLEEP instruction to be executed, which can reduce system noise during the conversion. If the ADC interrupt is enabled, the device will wake-up from Sleep when the conversion completes. If the ADC interrupt is disabled, the ADC module is turned off after the conversion completes, although the ADON bit remains set.

When the ADC clock source is something other than FRC, a SLEEP instruction causes the present conversion to be aborted and the ADC module is turned off, although the ADON bit remains set.

15.2.5 SPECIAL EVENT TRIGGER

The Special Event Trigger of the CCPx/ECCPx module allows periodic ADC measurements without software intervention. When this trigger occurs, the GO/DONE bit is set by hardware and the Timer1 counter resets to zero.

TABLE 15-2: SPECIAL EVENT TRIGGER

Device	CCPx/ECCPx
PIC16(L)F193X	CCP5

Using the Special Event Trigger does not assure proper ADC timing. It is the user's responsibility to ensure that the ADC timing requirements are met.

Refer to [Section 23.0 “Capture/Compare/PWM Modules”](#) for more information.

TABLE 23-10: SUMMARY OF REGISTERS ASSOCIATED WITH ENHANCED PWM

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
CCPxCON	PxM<1:0> ⁽¹⁾		DCxB<1:0>		CCPxM<3:0>				228
CCPxAS	CCPxASE	CCPxAS2	CCPxAS1	CCPxAS0	PSSxAC<1:0>		PSSxBD<1:0>		231
CCPTMRS0	C4TSEL<1:0>		C3TSEL<1:0>		C2TSEL<1:0>		C1TSEL<1:0>		229
CCPTMRS1	—	—	—	—	—	—	C5TSEL<1:0>		230
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	90
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	91
PIE2	OSFIE	C2IE	C1IE	EEIE	BCLIE	LCDIE	—	CCP2IE	92
PIE3	—	CCP5IE	CCP4IE	CCP3IE	TMR6IE	—	TMR4IE	—	93
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	94
PIR2	OSFIF	C2IF	C1IF	EEIF	BCLIF	LCDIF	—	CCP2IF	95
PIR3	—	CCP5IF	CCP4IF	CCP3IF	TMR6IF	—	TMR4IF	—	96
PRx	Timer2/4/6 Period Register								201*
PSTRxCON	—	—	—	STRxSYNC	STRxD	STRxC	STRxB	STRxA	233
PWMxCON	PxRSEN	PxDC<6:0>							232
TxCON	—	TxOUTPS<3:0>				TMRxON	TxCKPS<:0>1		203
TMRx	Timer2/4/6 Module Register								201
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	125
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	130
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	134
TRISD ⁽²⁾	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0	137
TRISE	—	—	—	—	— ⁽³⁾	TRISE2 ⁽²⁾	TRISE1 ⁽²⁾	TRISE0 ⁽²⁾	140

Legend: — = Unimplemented location, read as '0'. Shaded cells are not used by the PWM.

Note 1: Applies to ECCP modules only.

2: These registers/bits are not implemented on PIC16(L)F1938 devices, read as '0'.

3: Unimplemented, read as '1'.

* Page provides register information.

24.5.4 SLAVE MODE 10-BIT ADDRESS RECEPTION

This section describes a standard sequence of events for the MSSP module configured as an I²C Slave in 10-bit Addressing mode.

Figure 24-20 is used as a visual reference for this description.

This is a step by step process of what must be done by slave software to accomplish I²C communication.

1. Bus starts Idle.
2. Master sends Start condition; S bit of SSPSTAT is set; SSPIF is set if interrupt on Start detect is enabled.
3. Master sends matching high address with R/W bit clear; UA bit of the SSPSTAT register is set.
4. Slave sends $\overline{\text{ACK}}$ and SSPIF is set.
5. Software clears the SSPIF bit.
6. Software reads received address from SSPBUF clearing the BF flag.
7. Slave loads low address into SSPADD, releasing SCL.
8. Master sends matching low address byte to the Slave; UA bit is set.

Note: Updates to the SSPADD register are not allowed until after the $\overline{\text{ACK}}$ sequence.

9. Slave sends $\overline{\text{ACK}}$ and SSPIF is set.

Note: If the low address does not match, SSPIF and UA are still set so that the slave software can set SSPADD back to the high address. BF is not set because there is no match. CKP is unaffected.

10. Slave clears SSPIF.
11. Slave reads the received matching address from SSPBUF clearing BF.
12. Slave loads high address into SSPADD.
13. Master clocks a data byte to the slave and clocks out the slaves $\overline{\text{ACK}}$ on the ninth SCL pulse; SSPIF is set.
14. If SEN bit of SSPCON2 is set, CKP is cleared by hardware and the clock is stretched.
15. Slave clears SSPIF.
16. Slave reads the received byte from SSPBUF clearing BF.
17. If SEN is set the slave sets CKP to release the SCL.
18. Steps 13-17 repeat for each received byte.
19. Master sends Stop to end the transmission.

24.5.5 10-BIT ADDRESSING WITH ADDRESS OR DATA HOLD

Reception using 10-bit addressing with AHEN or DHEN set is the same as with 7-bit modes. The only difference is the need to update the SSPADD register using the UA bit. All functionality, specifically when the CKP bit is cleared and SCL line is held low are the same. Figure 24-21 can be used as a reference of a slave in 10-bit addressing with AHEN set.

Figure 24-22 shows a standard waveform for a slave transmitter in 10-bit Addressing mode.

24.6 I²C Master Mode

Master mode is enabled by setting and clearing the appropriate SSPM bits in the SSPCON1 register and by setting the SSPEN bit. In Master mode, the SDA and SCK pins must be configured as inputs. The MSSP peripheral hardware will override the output driver TRIS controls when necessary to drive the pins low.

Master mode of operation is supported by interrupt generation on the detection of the Start and Stop conditions. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the I²C bus may be taken when the P bit is set, or the bus is Idle.

In Firmware Controlled Master mode, user code conducts all I²C bus operations based on Start and Stop bit condition detection. Start and Stop condition detection is the only active circuitry in this mode. All other communication is done by the user software directly manipulating the SDA and SCL lines.

The following events will cause the SSP Interrupt Flag bit, SSPIF, to be set (SSP interrupt, if enabled):

- Start condition detected
- Stop condition detected
- Data transfer byte transmitted/received
- Acknowledge transmitted/received
- Repeated Start generated

Note 1: The MSSP module, when configured in I²C Master mode, does not allow queueing of events. For instance, the user is not allowed to initiate a Start condition and immediately write the SSPBUF register to initiate transmission before the Start condition is complete. In this case, the SSPBUF will not be written to and the WCOL bit will be set, indicating that a write to the SSPBUF did not occur

2: When in Master mode, Start/Stop detection is masked and an interrupt is generated when the SEN/PEN bit is cleared and the generation is complete.

24.6.1 I²C MASTER MODE OPERATION

The master device generates all of the serial clock pulses and the Start and Stop conditions. A transfer is ended with a Stop condition or with a Repeated Start condition. Since the Repeated Start condition is also the beginning of the next serial transfer, the I²C bus will not be released.

In Master Transmitter mode, serial data is output through SDA, while SCL outputs the serial clock. The first byte transmitted contains the slave address of the receiving device (7 bits) and the Read/Write (R/W) bit. In this case, the R/W bit will be logic '0'. Serial data is transmitted eight bits at a time. After each byte is transmitted, an Acknowledge bit is received. Start and Stop conditions are output to indicate the beginning and the end of a serial transfer.

In Master Receive mode, the first byte transmitted contains the slave address of the transmitting device (7 bits) and the R/W bit. In this case, the R/W bit will be logic '1'. Thus, the first byte transmitted is a 7-bit slave address followed by a '1' to indicate the receive bit. Serial data is received via SDA, while SCL outputs the serial clock. Serial data is received eight bits at a time. After each byte is received, an Acknowledge bit is transmitted. Start and Stop conditions indicate the beginning and end of transmission.

A Baud Rate Generator is used to set the clock frequency output on SCL. See [Section 24.7 “Baud Rate Generator”](#) for more detail.

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REGISTER 25-3: BAUDCON: BAUD RATE CONTROL REGISTER

R-0/0	R-1/1	U-0	R/W-0/0	R/W-0/0	U-0	R/W-0/0	R/W-0/0
ABDOVF	RCIDL	—	SCKP	BRG16	—	WUE	ABDEN
bit 7						bit 0	

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

bit 7 **ABDOVF:** Auto-Baud Detect Overflow bit

Asynchronous mode:

1 = Auto-baud timer overflowed

0 = Auto-baud timer did not overflow

Synchronous mode:

Don't care

bit 6 **RCIDL:** Receive Idle Flag bit

Asynchronous mode:

1 = Receiver is Idle

0 = Start bit has been received and the receiver is receiving

Synchronous mode:

Don't care

bit 5 **Unimplemented:** Read as '0'

bit 4 **SCKP:** Synchronous Clock Polarity Select bit

Asynchronous mode:

1 = Transmit inverted data to the TX/CK pin

0 = Transmit non-inverted data to the TX/CK pin

Synchronous mode:

1 = Data is clocked on rising edge of the clock

0 = Data is clocked on falling edge of the clock

bit 3 **BRG16:** 16-bit Baud Rate Generator bit

1 = 16-bit Baud Rate Generator is used

0 = 8-bit Baud Rate Generator is used

bit 2 **Unimplemented:** Read as '0'

bit 1 **WUE:** Wake-up Enable bit

Asynchronous mode:

1 = Receiver is waiting for a falling edge. No character will be received, byte RCIF will be set. WUE will automatically clear after RCIF is set.

0 = Receiver is operating normally

Synchronous mode:

Don't care

bit 0 **ABDEN:** Auto-Baud Detect Enable bit

Asynchronous mode:

1 = Auto-Baud Detect mode is enabled (clears when auto-baud is complete)

0 = Auto-Baud Detect mode is disabled

Synchronous mode:

Don't care

25.4 EUSART Baud Rate Generator (BRG)

The Baud Rate Generator (BRG) is an 8-bit or 16-bit timer that is dedicated to the support of both the asynchronous and synchronous EUSART operation. By default, the BRG operates in 8-bit mode. Setting the BRG16 bit of the BAUDCON register selects 16-bit mode.

The SPBRGH, SPBRGL register pair determines the period of the free running baud rate timer. In Asynchronous mode the multiplier of the baud rate period is determined by both the BRGH bit of the TXSTA register and the BRG16 bit of the BAUDCON register. In Synchronous mode, the BRGH bit is ignored.

Table 25-3 contains the formulas for determining the baud rate. Example 25-1 provides a sample calculation for determining the baud rate and baud rate error.

Typical baud rates and error values for various Asynchronous modes have been computed for your convenience and are shown in Table 25-3. It may be advantageous to use the high baud rate (BRGH = 1), or the 16-bit BRG (BRG16 = 1) to reduce the baud rate error. The 16-bit BRG mode is used to achieve slow baud rates for fast oscillator frequencies.

Writing a new value to the SPBRGH, SPBRGL register pair causes the BRG timer to be reset (or cleared). This ensures that the BRG does not wait for a timer overflow before outputting the new baud rate.

If the system clock is changed during an active receive operation, a receive error or data loss may result. To avoid this problem, check the status of the RCIDL bit to make sure that the receive operation is Idle before changing the system clock.

EXAMPLE 25-1: CALCULATING BAUD RATE ERROR

For a device with FOSC of 16 MHz, desired baud rate of 9600, Asynchronous mode, 8-bit BRG:

$$\text{Desired Baud Rate} = \frac{F_{OSC}}{64([SPBRGH:SPBRG] + 1)}$$

Solving for SPBRGH:SPBRGL:

$$X = \frac{\frac{F_{OSC}}{\text{Desired Baud Rate}}}{64} - 1$$

$$= \frac{\frac{16000000}{9600}}{64} - 1$$

$$= [25.042] = 25$$

$$\text{Calculated Baud Rate} = \frac{16000000}{64(25 + 1)}$$

$$= 9615$$

$$\text{Error} = \frac{\text{Calc. Baud Rate} - \text{Desired Baud Rate}}{\text{Desired Baud Rate}}$$

$$= \frac{(9615 - 9600)}{9600} = 0.16\%$$

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REGISTER 26-2: CPSCON1: CAPACITIVE SENSING CONTROL REGISTER 1

U-0	U-0	U-0	U-0	R/W-0/0 ⁽²⁾	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—	—	CPSCH<3:0>			
bit 7				bit 0			

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

bit 7-4 **Unimplemented:** Read as '0'

bit 3-0 **CPSCH<3:0>:** Capacitive Sensing Channel Select bits

If CPSON = 0:

These bits are ignored. No channel is selected.

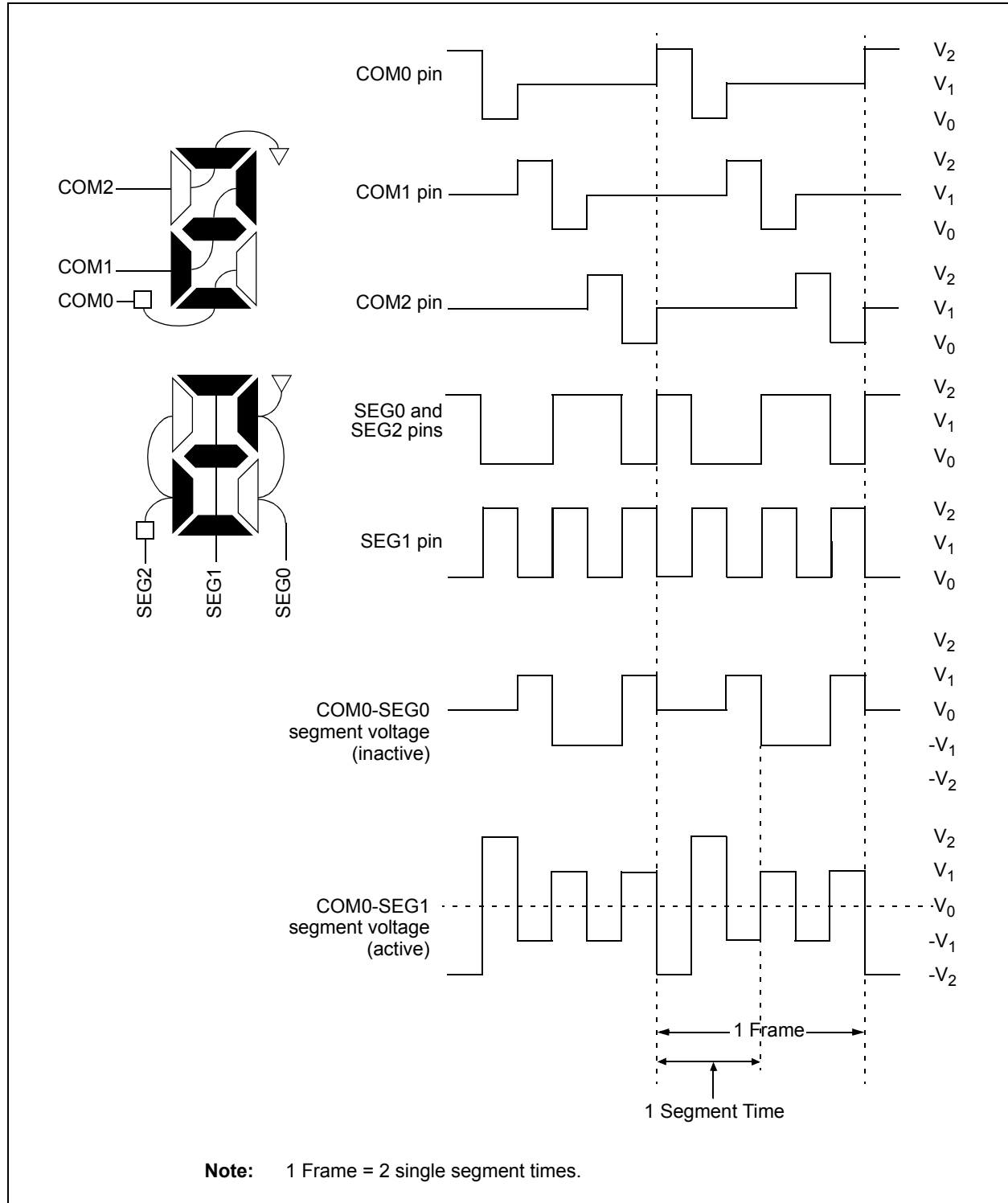
If CPSON = 1:

0000 = channel 0, (CPS0)
0001 = channel 1, (CPS1)
0010 = channel 2, (CPS2)
0011 = channel 3, (CPS3)
0100 = channel 4, (CPS4)
0101 = channel 5, (CPS5)
0110 = channel 6, (CPS6)
0111 = channel 7, (CPS7)
1000 = channel 8, (CPS8⁽¹⁾)
1001 = channel 9, (CPS9⁽¹⁾)
1010 = channel 10, (CPS10⁽¹⁾)
1011 = channel 11, (CPS11⁽¹⁾)
1100 = channel 12, (CPS12⁽¹⁾)
1101 = channel 13, (CPS13⁽¹⁾)
1110 = channel 14, (CPS14⁽¹⁾)
1111 = channel 15, (CPS15⁽¹⁾)

Note 1: These channels are not implemented on the PIC16(L)F1938.

Note 2: This bit is not implemented on PIC16(L)F1938, read as '0'

FIGURE 27-13: TYPE-A WAVEFORMS IN 1/3 MUX, 1/2 BIAS DRIVE



27.13 Configuring the LCD Module

The following is the sequence of steps to configure the LCD module.

1. Select the frame clock prescale using bits LP<3:0> of the LCDPS register.
2. Configure the appropriate pins to function as segment drivers using the LCDSEn registers.
3. Configure the LCD module for the following using the LCDCON register:
 - Multiplex and Bias mode, bits LMUX<1:0>
 - Timing source, bits CS<1:0>
 - Sleep mode, bit SLPEN
4. Write initial values to pixel data registers, LCD-DATA0 through LCDDATA11 (LCDDATA23 on PIC16F1938).
5. Clear LCD Interrupt Flag, LCDIF bit of the PIR2 register and if desired, enable the interrupt by setting bit LCDIE of the PIE2 register.
6. Configure bias voltages by setting the LCDRL, LCDREF and the associated ANSELx registers as needed.
7. Enable the LCD module by setting bit LCDEN of the LCDCON register.

27.14 Disabling the LCD Module

To disable the LCD module, write all '0's to the LCDCON register.

27.15 LCD Current Consumption

When using the LCD module the current consumption consists of the following three factors:

- Oscillator Selection
- LCD Bias Source
- Capacitance of the LCD segments

The current consumption of just the LCD module can be considered negligible compared to these other factors.

27.15.1 OSCILLATOR SELECTION

The current consumed by the clock source selected must be considered when using the LCD module. See the applicable Electrical Specifications Chapter for oscillator current consumption information.

27.15.2 LCD BIAS SOURCE

The LCD bias source, internal or external, can contribute significantly to the current consumption. Use the highest possible resistor values while maintaining contrast to minimize current.

27.15.3 CAPACITANCE OF THE LCD SEGMENTS

The LCD segments which can be modeled as capacitors which must be both charged and discharged every frame. The size of the LCD segment and its technology determines the segment's capacitance.

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BCF Bit Clear f

Syntax: [*label*] BCF f,b
Operands: $0 \leq f \leq 127$
 $0 \leq b \leq 7$
Operation: $0 \rightarrow (f)$
Status Affected: None
Description: Bit 'b' in register 'f' is cleared.

BTFSC Bit Test f, Skip if Clear

Syntax: [*label*] BTFSC f,b
Operands: $0 \leq f \leq 127$
 $0 \leq b \leq 7$
Operation: skip if $(f) = 0$
Status Affected: None
Description: If bit 'b' in register 'f' is '1', the next instruction is executed.
If bit 'b', in register 'f', is '0', the next instruction is discarded, and a NOP is executed instead, making this a 2-cycle instruction.

BRA Relative Branch

Syntax: [*label*] BRA label
[*label*] BRA \$+k
Operands: $-256 \leq \text{label} - \text{PC} + 1 \leq 255$
 $-256 \leq k \leq 255$
Operation: $(\text{PC}) + 1 + k \rightarrow \text{PC}$
Status Affected: None
Description: Add the signed 9-bit literal 'k' to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be $\text{PC} + 1 + k$. This instruction is a 2-cycle instruction. This branch has a limited range.

BTFSS Bit Test f, Skip if Set

Syntax: [*label*] BTFSS f,b
Operands: $0 \leq f \leq 127$
 $0 \leq b < 7$
Operation: skip if $(f) = 1$
Status Affected: None
Description: If bit 'b' in register 'f' is '0', the next instruction is executed.
If bit 'b' is '1', then the next instruction is discarded and a NOP is executed instead, making this a 2-cycle instruction.

BRW Relative Branch with W

Syntax: [*label*] BRW
Operands: None
Operation: $(\text{PC}) + (W) \rightarrow \text{PC}$
Status Affected: None
Description: Add the contents of W (unsigned) to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be $\text{PC} + 1 + (W)$. This instruction is a 2-cycle instruction.

BSF Bit Set f

Syntax: [*label*] BSF f,b
Operands: $0 \leq f \leq 127$
 $0 \leq b \leq 7$
Operation: $1 \rightarrow (f)$
Status Affected: None
Description: Bit 'b' in register 'f' is set.

FIGURE 30-4: POR AND POR REARM WITH SLOW RISING V_{DD}

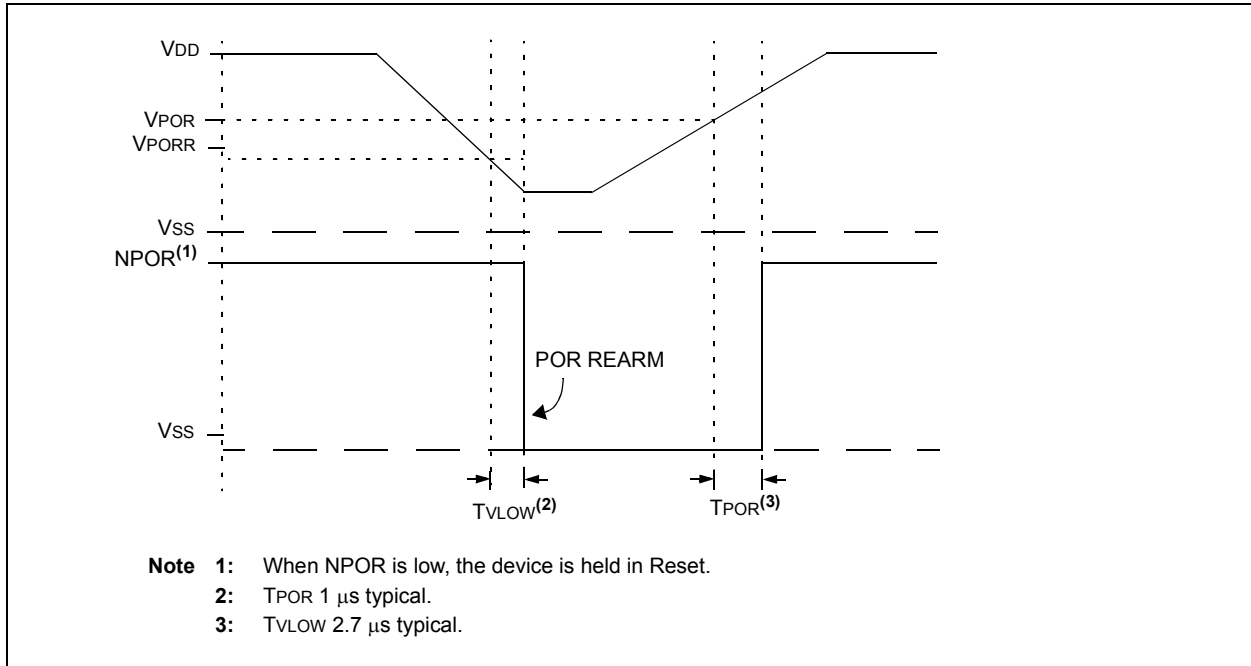


TABLE 30-2: OSCILLATOR PARAMETERS

Standard Operating Conditions (unless otherwise stated)								
Operating Temperature $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$								
Param No.	Sym.	Characteristic	Freq. Tolerance	Min.	Typ†	Max.	Units	Conditions
OS08	HFosc	Internal Calibrated HFINTOSC Frequency ⁽¹⁾	$\pm 2\%$	—	16.0	—	MHz	$0^{\circ}\text{C} \leq T_A \leq +60^{\circ}\text{C}$, $V_{DD} \geq 2.5\text{V}$
			$\pm 3\%$	—	16.0	—	MHz	$60^{\circ}\text{C} \leq T_A \leq 85^{\circ}\text{C}$, $V_{DD} \geq 2.5\text{V}$
			$\pm 5\%$	—	16.0	—	MHz	$-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$
OS08A	MFosc	Internal Calibrated MFINTOSC Frequency ⁽¹⁾	$\pm 2\%$	—	500	—	kHz	$0^{\circ}\text{C} \leq T_A \leq +60^{\circ}\text{C}$, $V_{DD} \geq 2.5\text{V}$
			$\pm 3\%$	—	500	—	kHz	$60^{\circ}\text{C} \leq T_A \leq 85^{\circ}\text{C}$, $V_{DD} \geq 2.5\text{V}$
			$\pm 5\%$	—	500	—	kHz	$-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$
OS09	LFosc	Internal LFINTOSC Frequency	—	—	31	—	kHz	$-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$
OS10*	Tiosc ST	HFINTOSC Wake-up from Sleep Start-up Time	—	—	3.2	8	μs	
		MFINTOSC Wake-up from Sleep Start-up Time	—	—	24	35	μs	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: To ensure these oscillator frequency tolerances, V_{DD} and V_{SS} must be capacitively decoupled as close to the device as possible. 0.1 μF and 0.01 μF values in parallel are recommended.

TABLE 30-3: PLL CLOCK TIMING SPECIFICATIONS ($V_{DD} = 2.7\text{V TO } 5.5\text{V}$)

Param No.	Sym.	Characteristic	Min.	Typ†	Max.	Units	Conditions
F10	FOSC	Oscillator Frequency Range	4	—	8	MHz	
F11	FSYS	On-Chip VCO System Frequency	16	—	32	MHz	
F12	TRC	PLL Start-up Time (Lock Time)	—	—	2	ms	
F13*	ΔCLK	CLKOUT Stability (Jitter)	-0.25%	—	+0.25%	%	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

FIGURE 30-7: CLKOUT AND I/O TIMING

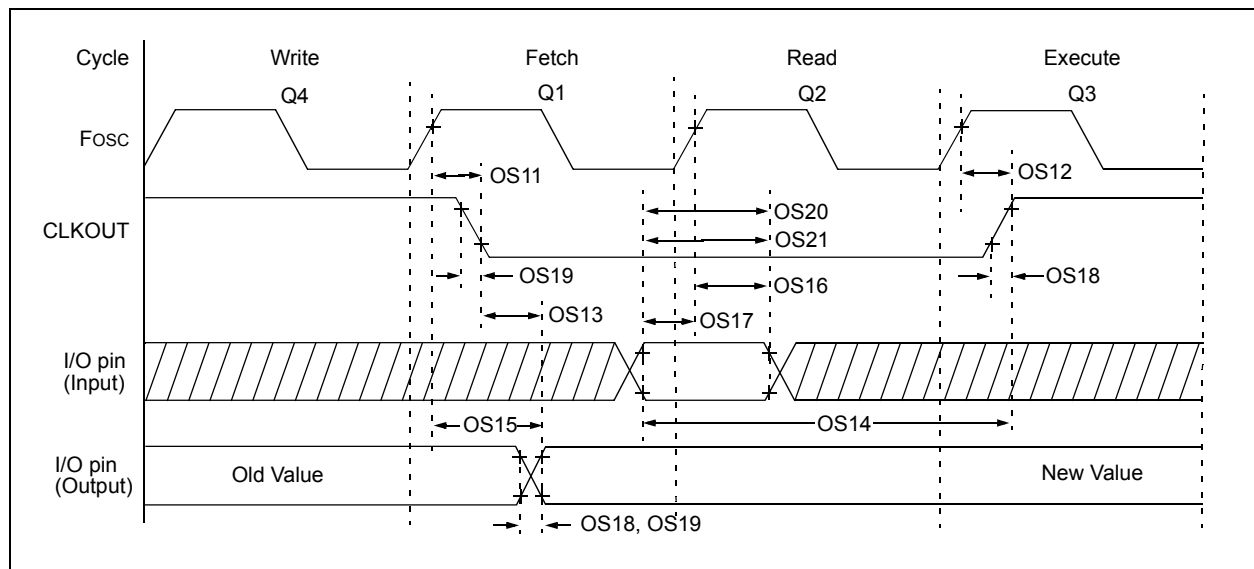


FIGURE 31-19: I_{DD} , LFINTOSC, $F_{osc} = 31$ kHz, PIC16LF1938/9 ONLY

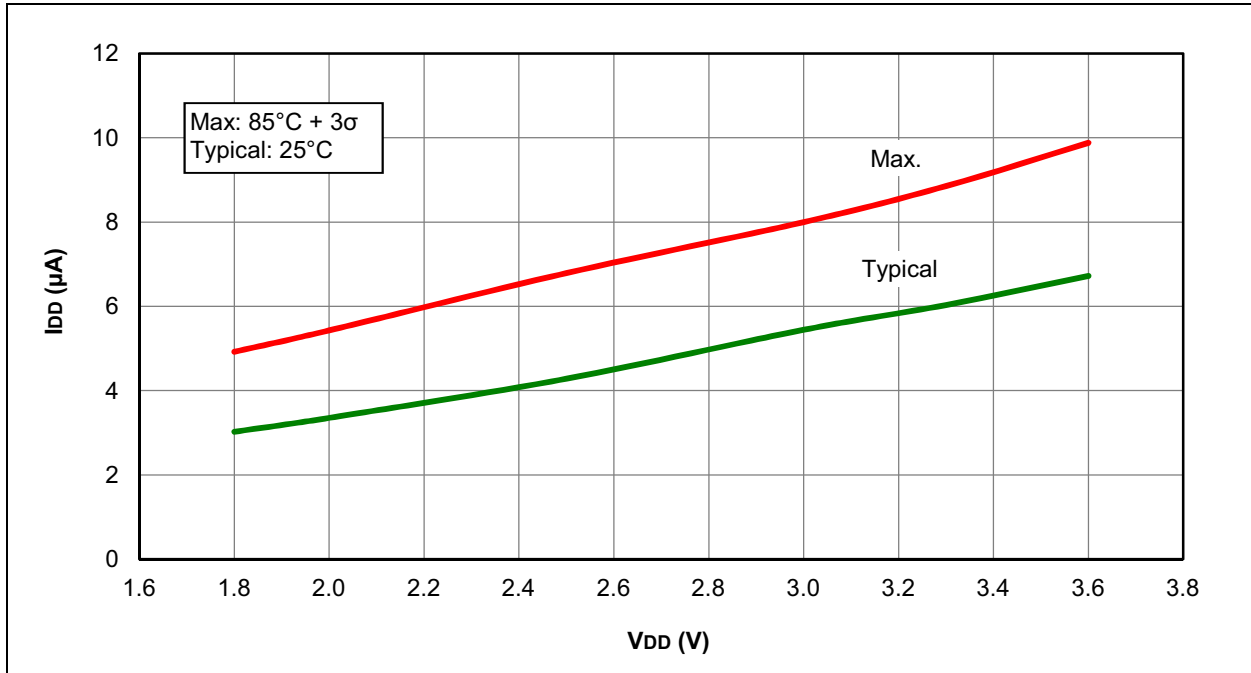


FIGURE 31-20: I_{DD} , LFINTOSC, $F_{osc} = 31$ kHz, PIC16F1938/9 ONLY

