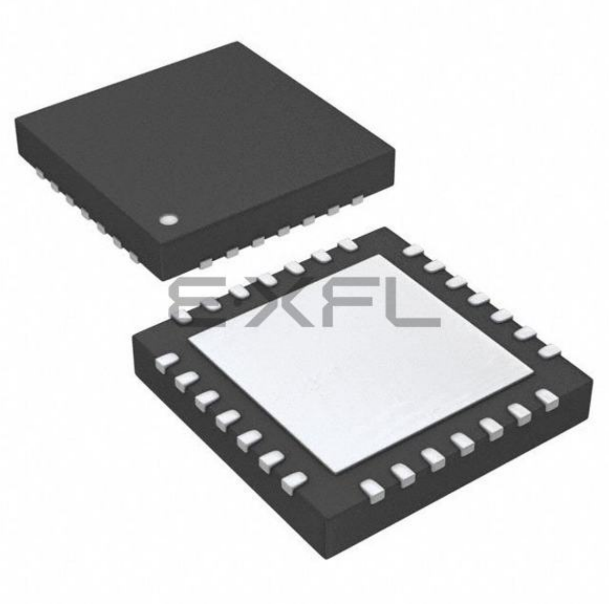

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	32MHz
Connectivity	I²C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, LCD, POR, PWM, WDT
Number of I/O	25
Program Memory Size	28KB (16K x 14)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	1K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 11x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	28-UFQFN Exposed Pad
Supplier Device Package	28-UQFN (4x4)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16lf1938t-i-mv

TABLE 1-2: PIC16(L)F1938/9 PINOUT DESCRIPTION

Name	Function	Input Type	Output Type	Description
RA0/AN0/C12IN0-/C2OUT(1)/SRNQ(1)/$\overline{SS}$(1)/VCAP(2)/SEG12	RA0	TTL	CMOS	General purpose I/O.
	AN0	AN	—	A/D Channel 0 input.
	C12IN0-	AN	—	Comparator C1 or C2 negative input.
	C2OUT	—	CMOS	Comparator C2 output.
	SRNQ	—	CMOS	SR Latch inverting output.
	$\overline{SS}$	ST	—	Slave Select input.
	VCAP	Power	Power	Filter capacitor for Voltage Regulator (PIC16F1938/9 only).
	SEG12	—	AN	LCD Analog output.
RA1/AN1/C12IN1-/SEG7	RA1	TTL	CMOS	General purpose I/O.
	AN1	AN	—	A/D Channel 1 input.
	C12IN1-	AN	—	Comparator C1 or C2 negative input.
	SEG7	—	AN	LCD Analog output.
RA2/AN2/C2IN+/VREF-/DACOUT/COM2	RA2	TTL	CMOS	General purpose I/O.
	AN2	AN	—	A/D Channel 2 input.
	C2IN+	AN	—	Comparator C2 positive input.
	VREF-	AN	—	A/D Negative Voltage Reference input.
	DACOUT	—	AN	Voltage Reference output.
	COM2	—	AN	LCD Analog output.
RA3/AN3/C1IN+/VREF+/COM3(3)/SEG15	RA3	TTL	CMOS	General purpose I/O.
	AN3	AN	—	A/D Channel 3 input.
	C1IN+	AN	—	Comparator C1 positive input.
	VREF+	AN	—	A/D Voltage Reference input.
	COM3(3)	—	AN	LCD Analog output.
	SEG15	—	AN	LCD Analog output.
RA4/C1OUT/CPS6/T0CKI/SRQ/CCP5/SEG4	RA4	TTL	CMOS	General purpose I/O.
	C1OUT	—	CMOS	Comparator C1 output.
	CPS6	AN	—	Capacitive sensing input 6.
	T0CKI	ST	—	Timer0 clock input.
	SRQ	—	CMOS	SR Latch non-inverting output.
	CCP5	ST	CMOS	Capture/Compare/PWM5.
	SEG4	—	AN	LCD Analog output.
RA5/AN4/C2OUT(1)/CPS7/SRNQ(1)/$\overline{SS}$(1)/VCAP(2)/SEG5	RA5	TTL	CMOS	General purpose I/O.
	AN4	AN	—	A/D Channel 4 input.
	C2OUT	—	CMOS	Comparator C2 output.
	CPS7	AN	—	Capacitive sensing input 7.
	SRNQ	—	CMOS	SR Latch inverting output.
	$\overline{SS}$	ST	—	Slave Select input.
	VCAP	Power	Power	Filter capacitor for Voltage Regulator (PIC16F1938/9 only).
	SEG5	—	AN	LCD Analog output.

Legend: AN = Analog input or output; CMOS = CMOS compatible input or output; OD = Open Drain
TTL = TTL compatible input; ST = Schmitt Trigger input with CMOS levels; XTAL = Crystal
HV = High Voltage; I^2C™ = Schmitt Trigger input with I^2C levels

Note 1: Pin function is selectable via the APFCON register.
2: PIC16F1938/9 devices only.
3: PIC16(L)F1938 devices only.
4: PORTD is available on PIC16(L)F1939 devices only.
5: RE<2:0> are available on PIC16(L)F1939 devices only.

TABLE 3-9: PIC16(L)F1938/9 MEMORY MAP, BANK 31

Address	Bank 31
F8Ch	Unimplemented Read as '0'
…	
FE3h	
FE4h	STATUS_SHAD
FE5h	WREG_SHAD
FE6h	BSR_SHAD
FE7h	PCLATH_SHAD
FE8h	FSR0L_SHAD
FE9h	FSR0H_SHAD
FEAh	FSR1L_SHAD
FEBh	FSR1H_SHAD
FECh	—
FEDh	STKPTR
FEEh	TOSL
FEFh	TOSH

Legend: = Unimplemented data memory locations, read as '0'.

3.3.5 SPECIAL FUNCTION REGISTERS SUMMARY

The Special Function Register Summary for the device family are as follows:

Device	Bank(s)	Page No.
PIC16(L)F1938/9	0	32
	1	33
	2	34
	3	35
	4	36
	5	37
	6	38
	7	39
	8	40
	9-14	41
	15	42
	16-30	44
	31	45

TABLE 3-10: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
Bank 5											
280h(2)	INDF0	Addressing this location uses contents of FSR0H/FSR0L to address data memory (not a physical register)								xxxx xxxx	xxxx xxxx
281h(2)	INDF1	Addressing this location uses contents of FSR1H/FSR1L to address data memory (not a physical register)								xxxx xxxx	xxxx xxxx
282h(2)	PCL	Program Counter (PC) Least Significant Byte								0000 0000	0000 0000
283h(2)	STATUS	—	—	—	$\overline{TO}$	$\overline{PD}$	Z	DC	C	---1 1000	---q quuu
284h(2)	FSR0L	Indirect Data Memory Address 0 Low Pointer								0000 0000	uuuu uuuu
285h(2)	FSR0H	Indirect Data Memory Address 0 High Pointer								0000 0000	0000 0000
286h(2)	FSR1L	Indirect Data Memory Address 1 Low Pointer								0000 0000	uuuu uuuu
287h(2)	FSR1H	Indirect Data Memory Address 1 High Pointer								0000 0000	0000 0000
288h(2)	BSR	—	—	—	BSR<4:0>					---0 0000	---0 0000
289h(2)	WREG	Working Register								0000 0000	uuuu uuuu
28Ah(1, 2)	PCLATH	—	Write Buffer for the upper 7 bits of the Program Counter							-000 0000	-000 0000
28Bh(2)	INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	0000 0000	0000 0000
28Ch	—	Unimplemented								—	—
28Dh	—	Unimplemented								—	—
28Eh	—	Unimplemented								—	—
28Fh	—	Unimplemented								—	—
290h	—	Unimplemented								—	—
291h	CCPR1L	Capture/Compare/PWM Register 1 (LSB)								xxxx xxxx	uuuu uuuu
292h	CCPR1H	Capture/Compare/PWM Register 1 (MSB)								xxxx xxxx	uuuu uuuu
293h	CCP1CON	P1M<1:0>		DC1B<1:0>		CCP1M<3:0>				0000 0000	0000 0000
294h	PWM1CON	P1RSEN	P1DC<6:0>							0000 0000	0000 0000
295h	CCP1AS	CCP1ASE	CCP1AS2	CCP1AS1	CCP1AS0	PSS1AC<1:0>		PSS1BD<1:0>		0000 0000	0000 0000
296h	PSTR1CON	—	—	—	STR1SYNC	STR1D	STR1C	STR1B	STR1A	---0 0001	---0 0001
297h	—	Unimplemented								—	—
298h	CCPR2L	Capture/Compare/PWM Register 2 (LSB)								xxxx xxxx	uuuu uuuu
299h	CCPR2H	Capture/Compare/PWM Register 2 (MSB)								xxxx xxxx	uuuu uuuu
29Ah	CCP2CON	P2M<1:0>		DC2B<1:0>		CCP2M<3:0>				0000 0000	0000 0000
29Bh	PWM2CON	P2RSEN	P2DC<6:0>							0000 0000	0000 0000
29Ch	CCP2AS	CCP2ASE	CCP2AS2	CCP2AS1	CCP2AS0	PSS2AC<1:0>		PSS2BD<1:0>		0000 0000	0000 0000
29Dh	PSTR2CON	—	—	—	STR2SYNC	STR2D	STR2C	STR2B	STR2A	---0 0001	---0 0001
29Eh	CCPTMRS0	C4TSEL1	C4TSEL0	C3TSEL1	C3TSEL0	C2TSEL1	C2TSEL0	C1TSEL1	C1TSEL0	0000 0000	0000 0000
29Fh	CCPTMRS1	—	—	—	—	—	—	C5TSEL<1:0>		---- --00	---- --00

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: The upper byte of the program counter is not directly accessible. PCLATH is a holding register for the PC<14:8>, whose contents are transferred to the upper byte of the program counter.

2: These registers can be addressed from any bank.

3: These registers/bits are not implemented on PIC16(L)F1938 devices, read as '0'.

4: Unimplemented, read as '1'.

NOTES:

6.1 Power-on Reset (POR)

The POR circuit holds the device in Reset until VDD has reached an acceptable level for minimum operation. Slow rising VDD, fast operating speeds or analog performance may require greater than minimum VDD. The PWRT, BOR or MCLR features can be used to extend the start-up period until all device operation conditions have been met.

6.1.1 POWER-UP TIMER (PWRT)

The Power-up Timer provides a nominal 64 ms time-out on POR or Brown-out Reset.

The device is held in Reset as long as PWRT is active. The PWRT delay allows additional time for the VDD to rise to an acceptable level. The Power-up Timer is enabled by clearing the PWRTE bit in Configuration Words.

The Power-up Timer starts after the release of the POR and BOR.

For additional information, refer to Application Note AN607, *"Power-up Trouble Shooting"* (DS00607).

6.2 Brown-Out Reset (BOR)

The BOR circuit holds the device in Reset when VDD reaches a selectable minimum level. Between the POR and BOR, complete voltage range coverage for execution protection can be implemented.

The Brown-out Reset module has four operating modes controlled by the BOREN<1:0> bits in Configuration Words. The four operating modes are:

- BOR is always on
- BOR is off when in Sleep
- BOR is controlled by software
- BOR is always off

Refer to Table 6-3 for more information.

The Brown-out Reset voltage level is selectable by configuring the BORV bit in Configuration Words.

A VDD noise rejection filter prevents the BOR from triggering on small events. If VDD falls below VBOR for a duration greater than parameter TBORDC, the device will reset. See Figure 6-2 for more information.

TABLE 6-1: BOR OPERATING MODES

BOREN<1:0>	SBOREN	Device Mode	BOR Mode	Device Operation upon release of POR	Device Operation upon wake- up from Sleep
11	X	X	Active	Waits for BOR ready(1)	
10	X	Awake	Active	Waits for BOR ready	
		Sleep	Disabled		
01	1	X	Active	Begins immediately	
	0		Disabled	Begins immediately	
00	X	X	Disabled	Begins immediately	

Note 1: In these specific cases, "Release of POR" and "Wake-up from Sleep", there is no delay in start-up. The BOR ready flag, (BORRDY = 1), will be set before the CPU is ready to execute instructions because the BOR circuit is forced on by the BOREN<1:0> bits.

6.2.1 BOR IS ALWAYS ON

When the BOREN bits of Configuration Words are programmed to '11', the BOR is always on. The device start-up will be delayed until the BOR is ready and VDD is higher than the BOR threshold.

BOR protection is active during Sleep. The BOR does not delay wake-up from Sleep.

6.2.2 BOR IS OFF IN SLEEP

When the BOREN bits of Configuration Words are programmed to '10', the BOR is on, except in Sleep. The device start-up will be delayed until the BOR is ready and VDD is higher than the BOR threshold.

BOR protection is not active during Sleep. The device wake-up will be delayed until the BOR is ready.

6.2.3 BOR CONTROLLED BY SOFTWARE

When the BOREN bits of Configuration Words are programmed to '01', the BOR is controlled by the SBOREN bit of the BORCON register. The device start-up is not delayed by the BOR ready condition or the VDD level.

BOR protection begins as soon as the BOR circuit is ready. The status of the BOR circuit is reflected in the BORRDY bit of the BORCON register.

BOR protection is unchanged by Sleep.

9.0 POWER-DOWN MODE (SLEEP)

The Power-Down mode is entered by executing a SLEEP instruction.

Upon entering Sleep mode, the following conditions exist:

1. WDT will be cleared but keeps running, if enabled for operation during Sleep.
2. $\overline{PD}$ bit of the STATUS register is cleared.
3. $\overline{TO}$ bit of the STATUS register is set.
4. CPU clock is disabled.
5. 31 kHz LFINTOSC is unaffected and peripherals that operate from it may continue operation in Sleep.
6. Timer1 oscillator is unaffected and peripherals that operate from it may continue operation in Sleep.
7. ADC is unaffected, if the dedicated FRC clock is selected.
8. Capacitive Sensing oscillator is unaffected.
9. I/O ports maintain the status they had before SLEEP was executed (driving high, low or high-impedance).
10. Resets other than WDT are not affected by Sleep mode.

Refer to individual chapters for more details on peripheral operation during Sleep.

To minimize current consumption, the following conditions should be considered:

- I/O pins should not be floating
- External circuitry sinking current from I/O pins
- Internal circuitry sourcing current from I/O pins
- Current draw from pins with internal weak pull-ups
- Modules using 31 kHz LFINTOSC
- Modules using Timer1 oscillator

I/O pins that are high-impedance inputs should be pulled to V_{DD} or V_{SS} externally to avoid switching currents caused by floating inputs.

Examples of internal circuitry that might be sourcing current include modules such as the DAC and FVR modules. See **Section 17.0 "Digital-to-Analog Converter (DAC) Module"** and **Section 14.0 "Fixed Voltage Reference (FVR)"** for more information on these modules.

9.1 Wake-up from Sleep

The device can wake-up from Sleep through one of the following events:

1. External Reset input on $\overline{MCLR}$ pin, if enabled
2. BOR Reset, if enabled
3. POR Reset
4. Watchdog Timer, if enabled
5. Any external interrupt
6. Interrupts by peripherals capable of running during Sleep (see individual peripheral for more information)

The first three events will cause a device Reset. The last three events are considered a continuation of program execution. To determine whether a device Reset or wake-up event occurred, refer to **Section 6.11 "Determining the Cause of a Reset"**.

When the SLEEP instruction is being executed, the next instruction (PC + 1) is prefetched. For the device to wake-up through an interrupt event, the corresponding interrupt enable bit must be enabled. Wake-up will occur regardless of the state of the GIE bit. If the GIE bit is disabled, the device continues execution at the instruction after the SLEEP instruction. If the GIE bit is enabled, the device executes the instruction after the SLEEP instruction, the device will then call the Interrupt Service Routine. In cases where the execution of the instruction following SLEEP is not desirable, the user should have a NOP after the SLEEP instruction.

The WDT is cleared when the device wakes up from Sleep, regardless of the source of wake-up.

12.10 Register Definitions: PORTD Control

REGISTER 12-14: PORTD: PORTD REGISTER(1)

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	

bit 7-0 **RD<7:0>**: PORTD General Purpose I/O Pin bits
1 = Port pin is > V_{IH}
0 = Port pin is < V_{IL}

Note 1: PORTD is not implemented on PIC16(L)F1938 devices, read as ‘0’.

REGISTER 12-15: TRISD: PORTD TRI-STATE REGISTER(1)

R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1
TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	

bit 7-0 **TRISD<7:0>:** PORTD Tri-State Control bits
1 = PORTD pin configured as an input (tri-stated)
0 = PORTD pin configured as an output

Note 1: TRISD is not implemented on PIC16(L)F1938 devices, read as ‘0’.
2: PORTD implemented on PIC16(L)F1939 devices only.

15.4 A/D Acquisition Requirements

For the ADC to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The Analog Input model is shown in Figure 15-4. The source impedance (RS) and the internal sampling switch (RSS) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (RSS) impedance varies over the device voltage (VDD), refer to Figure 15-4. **The maximum recommended impedance for analog sources is 10 kΩ.** As the source impedance is decreased, the acquisition time may be decreased. After the analog input channel is selected (or changed), an A/D acquisition must be done before the conversion can be started. To calculate the minimum acquisition time, Equation 15-1 may be used. This equation assumes that 1/2 LSb error is used (1,024 steps for the ADC). The 1/2 LSb error is the maximum error allowed for the ADC to meet its specified resolution.

EQUATION 15-1: ACQUISITION TIME EXAMPLE

Assumptions: $Temperature = 50°C$ *and external impedance of* $10k\Omega$ $5.0V$ V_{DD}

$$T_{ACQ} = \text{Amplifier Settling Time} + \text{Hold Capacitor Charging Time} + \text{Temperature Coefficient}$$
$$= T_{AMP} + T_C + T_{COFF}$$
$$= 2\mu s + T_C + [(Temperature - 25°C)(0.05\mu s/°C)]$$

The value for T_C *can be approximated with the following equations:*

$$V_{APPLIED}\left(1 - \frac{1}{(2^{n+1}) - 1}\right) = V_{CHOLD} \quad ;[1]\ V_{CHOLD} \text{ charged to within 1/2 lsb}$$

$$V_{APPLIED}\left(1 - e^{\frac{-T_C}{RC}}\right) = V_{CHOLD} \quad ;[2]\ V_{CHOLD} \text{ charge response to } V_{APPLIED}$$

$$V_{APPLIED}\left(1 - e^{\frac{-Tc}{RC}}\right) = V_{APPLIED}\left(1 - \frac{1}{(2^{n+1}) - 1}\right) \quad ;\text{combining [1] and [2]}$$

Note: Where n = number of bits of the ADC.

Solving for T_C:

$$T_C = -C_{HOLD}(R_{IC} + R_{SS} + R_S)\ ln(1/2047)$$
$$= -13.5pF(1k\Omega + 7k\Omega + 10k\Omega)\ ln(0.0004885)$$
$$= 1.20\mu s$$

Therefore:

$$T_{ACQ} = 2\mu s + 1.20\mu s + [(50°C - 25°C)(0.05\mu s/°C)]$$
$$= 4.45\mu s$$

Note 1: The reference voltage (VREF) has no effect on the equation, since it cancels itself out.

2: The charge holding capacitor (CHOLD) is not discharged after each conversion.

3: The maximum recommended impedance for analog sources is 10 kΩ. This is required to meet the pin leakage specification.

19.2 Latch Output

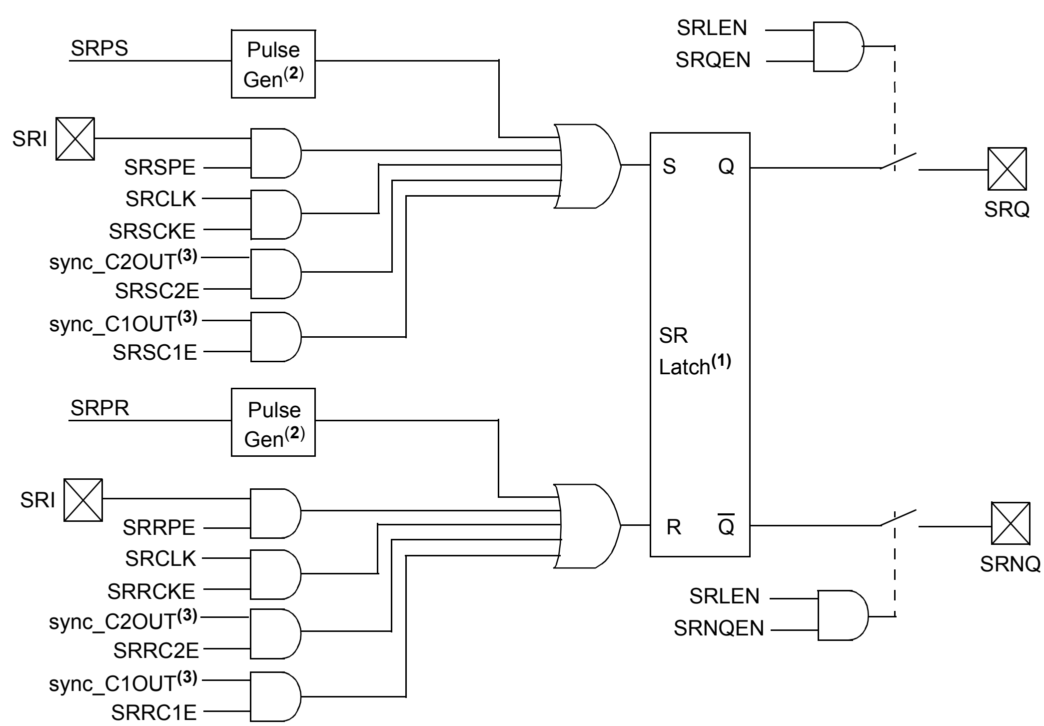
The SRQEN and SRNQEN bits of the SRCON0 register control the Q and $\overline{Q}$ latch outputs. Both of the SR Latch outputs may be directly output to an I/O pin at the same time. The $\overline{Q}$ latch output pin function can be moved to an alternate pin using the SRNQSEL bit of the APFCON register.

The applicable TRIS bit of the corresponding port must be cleared to enable the port pin output driver.

19.3 Effects of a Reset

Upon any device Reset, the SR Latch output is not initialized to a known state. The user's firmware is responsible for initializing the latch output before enabling the output pins.

FIGURE 19-1: SR LATCH SIMPLIFIED BLOCK DIAGRAM



Note 1: If R = 1 and S = 1 simultaneously, Q = 0, $\overline{Q}$ = 1

2: Pulse generator causes a 1 Q-state pulse width.

3: Name denotes the connection point at the comparator output.

21.3 Timer1 Prescaler

Timer1 has four prescaler options allowing 1, 2, 4 or 8 divisions of the clock input. The T1CKPS bits of the T1CON register control the prescale counter. The prescale counter is not directly readable or writable; however, the prescaler counter is cleared upon a write to TMR1H or TMR1L.

21.4 Timer1 Oscillator

A dedicated low-power 32.768 kHz oscillator circuit is built-in between pins T1OSI (input) and T1OSO (amplifier output). This internal circuit is to be used in conjunction with an external 32.768 kHz crystal.

The oscillator circuit is enabled by setting the T1OSCEN bit of the T1CON register. The oscillator will continue to run during Sleep.

Note:	The oscillator requires a start-up and stabilization time before use. Thus, T1OSCEN should be set and a suitable delay observed prior to using Timer1. A suitable delay similar to the OST delay can be implemented in software by clearing the TMR1IF bit then presetting the TMR1H:TMR1L register pair to FC00h. The TMR1IF flag will be set when 1024 clock cycles have elapsed, thereby indicating that the oscillator is running and reasonably stable.

21.5 Timer1 Operation in Asynchronous Counter Mode

If control bit $\overline{\text{T1SYNC}}$ of the T1CON register is set, the external clock input is not synchronized. The timer increments asynchronously to the internal phase clocks. If the external clock source is selected then the timer will continue to run during Sleep and can generate an interrupt on overflow, which will wake-up the processor. However, special precautions in software are needed to read/write the timer (see **Section 21.5.1 "Reading and Writing Timer1 in Asynchronous Counter Mode"**).

Note:	When switching from synchronous to asynchronous operation, it is possible to skip an increment. When switching from asynchronous to synchronous operation, it is possible to produce an additional increment.

21.5.1 READING AND WRITING TIMER1 IN ASYNCHRONOUS COUNTER MODE

Reading TMR1H or TMR1L while the timer is running from an external asynchronous clock will ensure a valid read (taken care of in hardware). However, the user should keep in mind that reading the 16-bit timer in two 8-bit values itself, poses certain problems, since the timer may overflow between the reads.

For writes, it is recommended that the user simply stop the timer and write the desired values. A write contention may occur by writing to the timer registers, while the register is incrementing. This may produce an unpredictable value in the TMR1H:TMR1L register pair.

21.6 Timer1 Gate

Timer1 can be configured to count freely or the count can be enabled and disabled using Timer1 gate circuitry. This is also referred to as Timer1 Gate Enable.

Timer1 gate can also be driven by multiple selectable sources.

21.6.1 TIMER1 GATE ENABLE

The Timer1 Gate Enable mode is enabled by setting the TMR1GE bit of the T1GCON register. The polarity of the Timer1 Gate Enable mode is configured using the T1GPOL bit of the T1GCON register.

When Timer1 Gate Enable mode is enabled, Timer1 will increment on the rising edge of the Timer1 clock source. When Timer1 Gate Enable mode is disabled, no incrementing will occur and Timer1 will hold the current count. See Figure 21-3 for timing details.

TABLE 21-3: TIMER1 GATE ENABLE SELECTIONS

T1CLK	T1GPOL	T1G	Timer1 Operation
↑	0	0	Counts
↑	0	1	Holds Count
↑	1	0	Holds Count
↑	1	1	Counts

22.0 TIMER2/4/6 MODULES

There are up to three identical Timer2-type modules available. To maintain pre-existing naming conventions, the Timers are called Timer2, Timer4 and Timer6 (also Timer2/4/6).

Note:	The 'x' variable used in this section is used to designate Timer2, Timer4, or Timer6. For example, TxCON references T2CON, T4CON, or T6CON. PRx references PR2, PR4, or PR6.

The Timer2/4/6 modules incorporate the following features:

- 8-bit Timer and Period registers (TMRx and PRx, respectively)
- Readable and writable (both registers)
- Software programmable prescaler (1:1, 1:4, 1:16, and 1:64)
- Software programmable postscaler (1:1 to 1:16)
- Interrupt on TMRx match with PRx, respectively
- Optional use as the shift clock for the MSSP module (Timer2 only)

See Figure 22-1 for a block diagram of Timer2/4/6.

FIGURE 22-1: TIMER2/4/6 BLOCK DIAGRAM

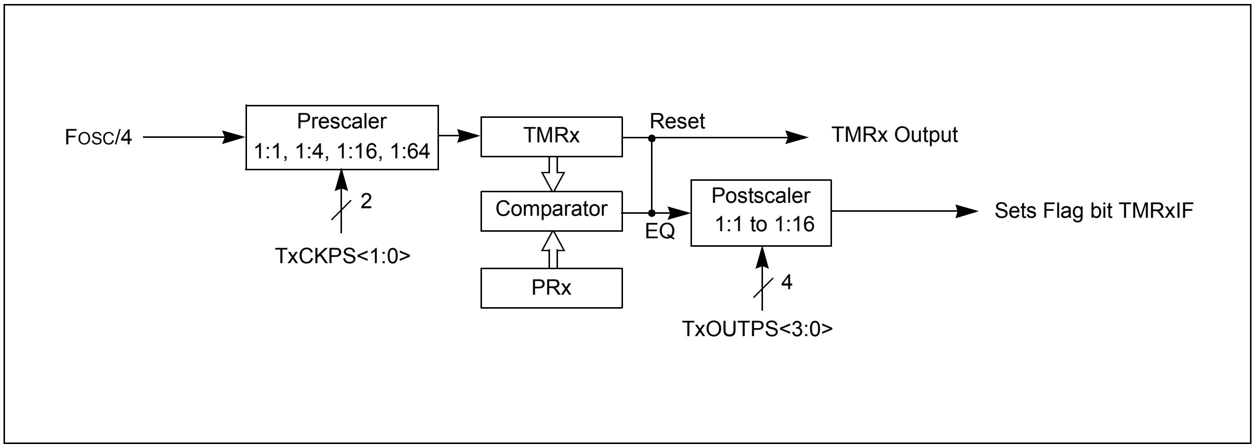


FIGURE 24-7: SPI DAISY-CHAIN CONNECTION

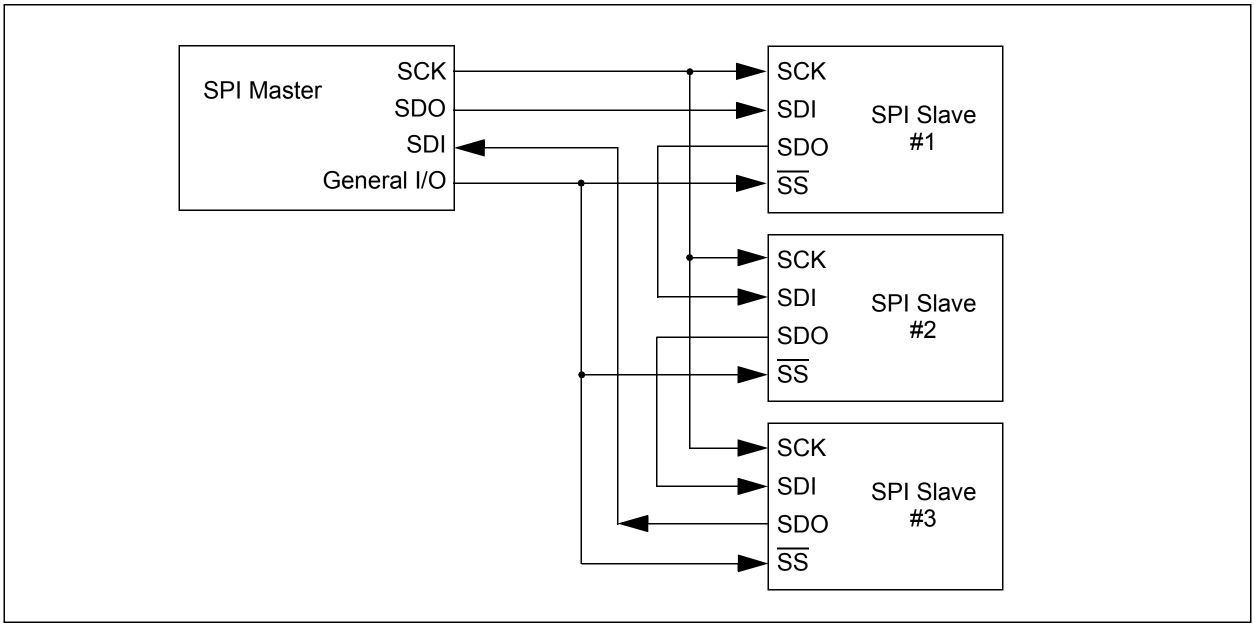
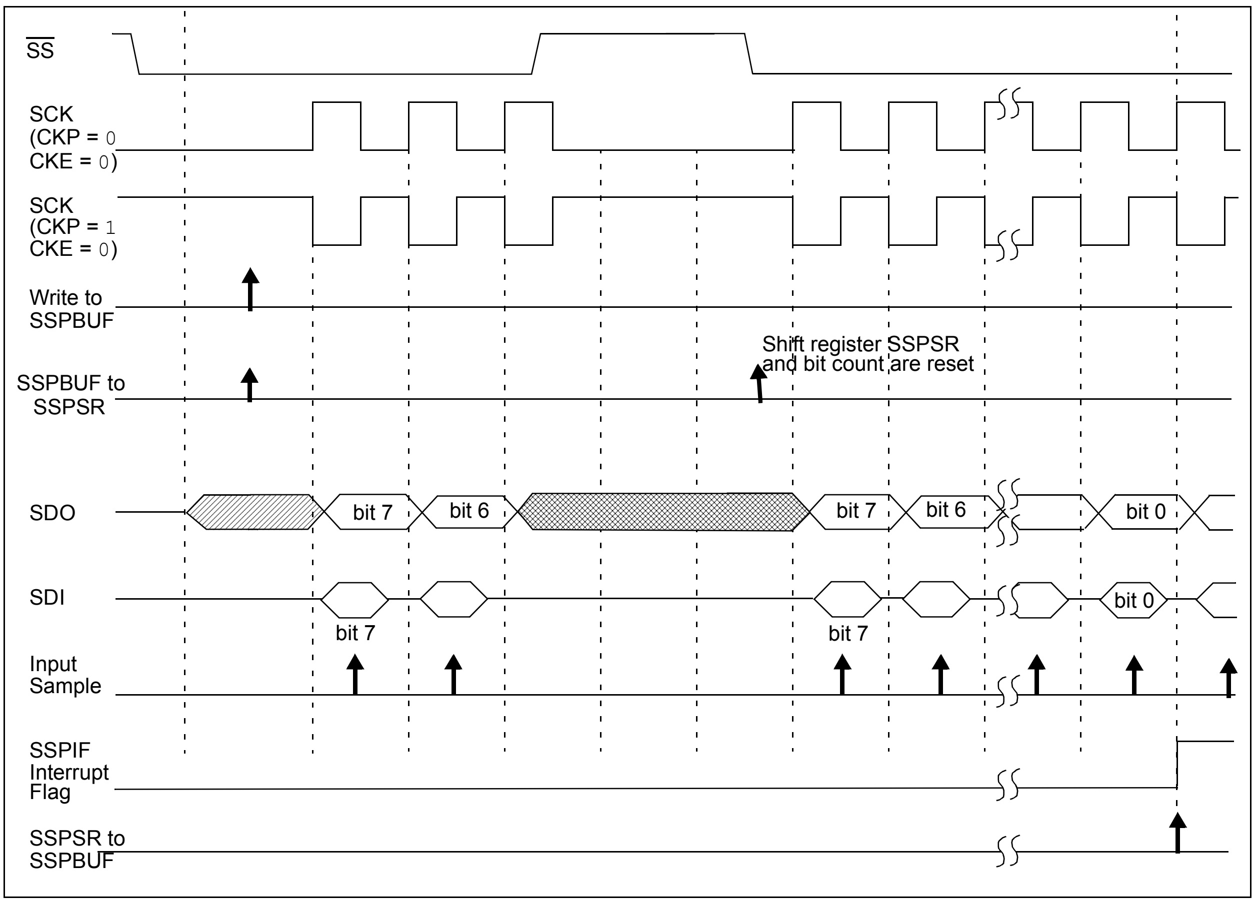


FIGURE 24-8: SLAVE SELECT SYNCHRONOUS WAVEFORM



24.4.9 ACKNOWLEDGE SEQUENCE

The ninth SCL pulse for any transferred byte in I²C is dedicated as an Acknowledge. It allows receiving devices to respond back to the transmitter by pulling the SDA line low. The transmitter must release control of the line during this time to shift in the response. The Acknowledge ($\overline{\text{ACK}}$) is an active-low signal, pulling the SDA line low indicated to the transmitter that the device has received the transmitted data and is ready to receive more.

The result of an $\overline{\text{ACK}}$ is placed in the ACKSTAT bit of the SSPCON2 register.

Slave software, when the AHEN and DHEN bits are set, allow the user to set the $\overline{\text{ACK}}$ value sent back to the transmitter. The ACKDT bit of the SSPCON2 register is set/cleared to determine the response.

Slave hardware will generate an $\overline{\text{ACK}}$ response if the AHEN and DHEN bits of the SSPCON3 register are clear.

There are certain conditions where an $\overline{\text{ACK}}$ will not be sent by the slave. If the BF bit of the SSPSTAT register or the SSPOV bit of the SSPCON1 register are set when a byte is received.

When the module is addressed, after the eighth falling edge of SCL on the bus, the ACKTIM bit of the SSPCON3 register is set. The ACKTIM bit indicates the acknowledge time of the active bus. The ACKTIM Status bit is only active when the AHEN bit or DHEN bit is enabled.

24.5 I²C Slave Mode Operation

The MSSP Slave mode operates in one of four modes selected in the SSPM bits of SSPCON1 register. The modes can be divided into 7-bit and 10-bit Addressing mode. 10-bit Addressing modes operate the same as 7-bit with some additional overhead for handling the larger addresses.

Modes with Start and Stop bit interrupts operated the same as the other modes with SSPIF additionally getting set upon detection of a Start, Restart, or Stop condition.

24.5.1 SLAVE MODE ADDRESSES

The SSPADD register (Register 24-6) contains the Slave mode address. The first byte received after a Start or Restart condition is compared against the value stored in this register. If the byte matches, the value is loaded into the SSPBUF register and an interrupt is generated. If the value does not match, the module goes idle and no indication is given to the software that anything happened.

The SSP Mask register (Register 24-5) affects the address matching process. See **Section 24.5.9 "SSP Mask Register"** for more information.

24.5.1.1 I²C Slave 7-bit Addressing Mode

In 7-bit Addressing mode, the LSb of the received data byte is ignored when determining if there is an address match.

24.5.1.2 I²C Slave 10-bit Addressing Mode

In 10-bit Addressing mode, the first received byte is compared to the binary value of '1 1 1 1 0 A9 A8 0'. A9 and A8 are the two MSb of the 10-bit address and stored in bits 2 and 1 of the SSPADD register.

After the acknowledge of the high byte the UA bit is set and SCL is held low until the user updates SSPADD with the low address. The low address byte is clocked in and all eight bits are compared to the low address value in SSPADD. Even if there is not an address match; SSPIF and UA are set, and SCL is held low until SSPADD is updated to receive a high byte again. When SSPADD is updated the UA bit is cleared. This ensures the module is ready to receive the high address byte on the next communication.

A high and low address match as a write request is required at the start of all 10-bit addressing communication. A transmission can be initiated by issuing a Restart once the slave is addressed, and clocking in the high address with the R/$\overline{\text{W}}$ bit set. The slave hardware will then acknowledge the read request and prepare to clock out data. This is only valid for a slave after it has received a complete high and low address byte match.

24.6 I²C Master Mode

Master mode is enabled by setting and clearing the appropriate SSPM bits in the SSPCON1 register and by setting the SSPEN bit. In Master mode, the SDA and SCK pins must be configured as inputs. The MSSP peripheral hardware will override the output driver TRIS controls when necessary to drive the pins low.

Master mode of operation is supported by interrupt generation on the detection of the Start and Stop conditions. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the I²C bus may be taken when the P bit is set, or the bus is Idle.

In Firmware Controlled Master mode, user code conducts all I²C bus operations based on Start and Stop bit condition detection. Start and Stop condition detection is the only active circuitry in this mode. All other communication is done by the user software directly manipulating the SDA and SCL lines.

The following events will cause the SSP Interrupt Flag bit, SSPIF, to be set (SSP interrupt, if enabled):

- Start condition detected
- Stop condition detected
- Data transfer byte transmitted/received
- Acknowledge transmitted/received
- Repeated Start generated

> **Note 1:** The MSSP module, when configured in I²C Master mode, does not allow queueing of events. For instance, the user is not allowed to initiate a Start condition and immediately write the SSPBUF register to initiate transmission before the Start condition is complete. In this case, the SSPBUF will not be written to and the WCOL bit will be set, indicating that a write to the SSPBUF did not occur
>
> **2:** When in Master mode, Start/Stop detection is masked and an interrupt is generated when the SEN/PEN bit is cleared and the generation is complete.

24.6.1 I²C MASTER MODE OPERATION

The master device generates all of the serial clock pulses and the Start and Stop conditions. A transfer is ended with a Stop condition or with a Repeated Start condition. Since the Repeated Start condition is also the beginning of the next serial transfer, the I²C bus will not be released.

In Master Transmitter mode, serial data is output through SDA, while SCL outputs the serial clock. The first byte transmitted contains the slave address of the receiving device (7 bits) and the Read/$\overline{\text{Write}}$ (R/$\overline{\text{W}}$) bit. In this case, the R/$\overline{\text{W}}$ bit will be logic '0'. Serial data is transmitted eight bits at a time. After each byte is transmitted, an Acknowledge bit is received. Start and Stop conditions are output to indicate the beginning and the end of a serial transfer.

In Master Receive mode, the first byte transmitted contains the slave address of the transmitting device (7 bits) and the R/$\overline{\text{W}}$ bit. In this case, the R/$\overline{\text{W}}$ bit will be logic '1'. Thus, the first byte transmitted is a 7-bit slave address followed by a '1' to indicate the receive bit. Serial data is received via SDA, while SCL outputs the serial clock. Serial data is received eight bits at a time. After each byte is received, an Acknowledge bit is transmitted. Start and Stop conditions indicate the beginning and end of transmission.

A Baud Rate Generator is used to set the clock frequency output on SCL. See **Section 24.7 "Baud Rate Generator"** for more detail.

REGISTER 24-5: SSPMSK: SSP MASK REGISTER

R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1
MSK<7:0>							
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	

bit 7-1 **MSK<7:1>:** Mask bits
`1` = The received address bit n is compared to SSPADD<n> to detect I^{2}C address match
`0` = The received address bit n is not used to detect I^{2}C address match

bit 0 **MSK<0>:** Mask bit for I^{2}C Slave mode, 10-bit Address
I^{2}C Slave mode, 10-bit address (SSPM<3:0> = `0111` or `1111`):
`1` = The received address bit 0 is compared to SSPADD<0> to detect I^{2}C address match
`0` = The received address bit 0 is not used to detect I^{2}C address match
I^{2}C Slave mode, 7-bit address, the bit is ignored

REGISTER 24-6: SSPADD: MSSP ADDRESS AND BAUD RATE REGISTER (I^{2}C MODE)

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
ADD<7:0>							
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	

Master mode:

bit 7-0 **ADD<7:0>:** Baud Rate Clock Divider bits
SCL pin clock period = ((ADD<7:0> + 1) *4)/FOSC

10-Bit Slave mode – Most Significant Address Byte:

bit 7-3 **Not used:** Unused for Most Significant Address byte. Bit state of this register is a “don’t care”. Bit pattern sent by master is fixed by I^{2}C specification and must be equal to ‘`11110`’. However, those bits are compared by hardware and are not affected by the value in this register.

bit 2-1 **ADD<2:1>:** Two Most Significant bits of 10-bit address

bit 0 **Not used:** Unused in this mode. Bit state is a “don’t care”.

10-Bit Slave mode – Least Significant Address Byte:

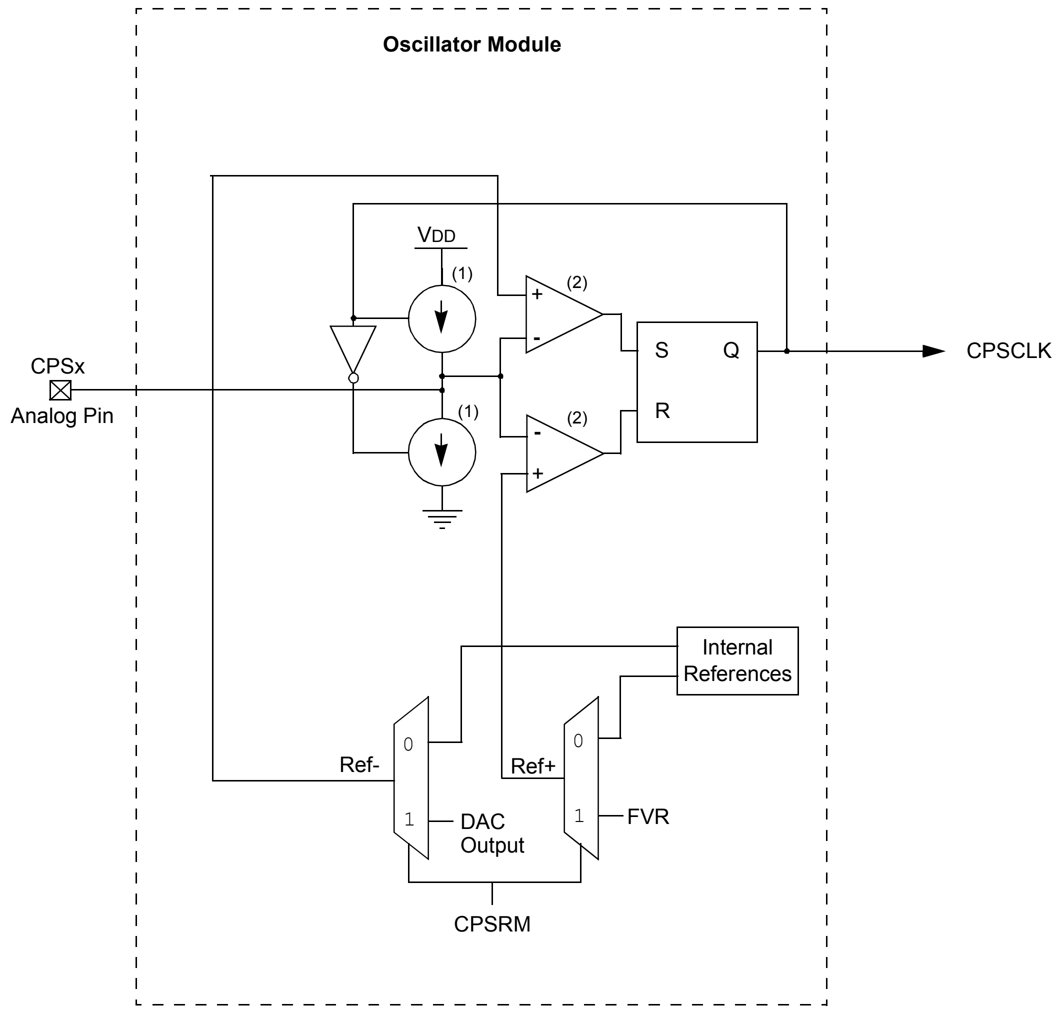
bit 7-0 **ADD<7:0>:** Eight Least Significant bits of 10-bit address

7-Bit Slave mode:

bit 7-1 **ADD<7:1>:** 7-bit address

bit 0 **Not used:** Unused in this mode. Bit state is a “don’t care”.

FIGURE 26-2: CAPACITIVE SENSING OSCILLATOR BLOCK DIAGRAM



Note 1: Module Enable, Voltage Reference mode and Current Range selections are not shown.
2: Comparators remain active in Noise Detection mode.

27.0 LIQUID CRYSTAL DISPLAY (LCD) DRIVER MODULE

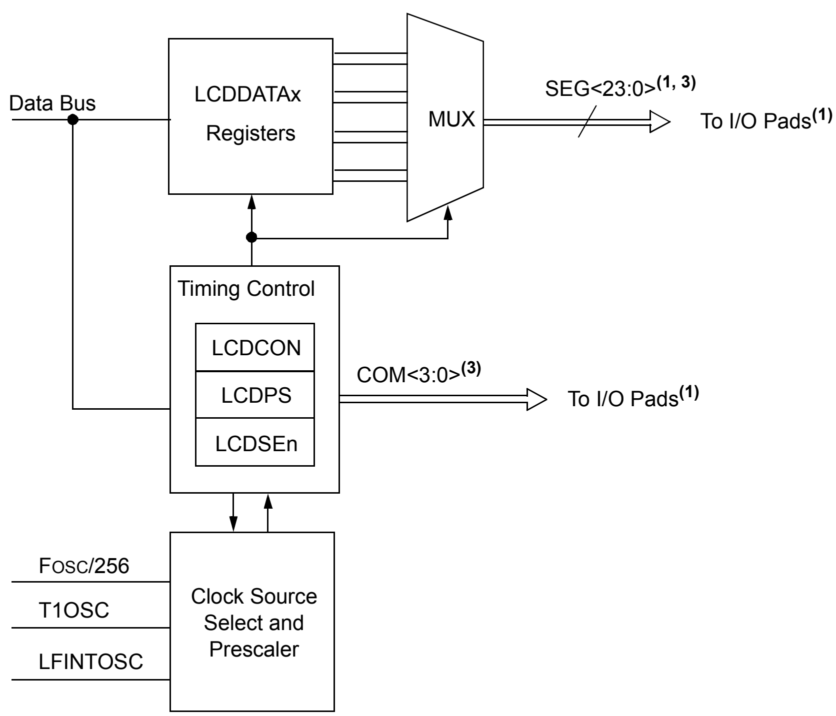
The Liquid Crystal Display (LCD) Driver module generates the timing control to drive a static or multiplexed LCD panel. In the PIC16(L)F193X device, the module drives the panels of up to four commons and up to 24 segments. The LCD module also provides control of the LCD pixel data.

The LCD Driver module supports:

- Direct driving of LCD panel
- Three LCD clock sources with selectable prescaler
- Up to four common pins:
  - Static (1 common)
  - 1/2 multiplex (2 commons)
  - 1/3 multiplex (3 commons)
  - 1/4 multiplex (4 commons)
- Segment pins up to:
  - 16 (PIC16(L)F1938)
  - 24 (PIC16(L)F1939)
- Static, 1/2 or 1/3 LCD Bias

Note:	COM3 and SEG15 share the same physical pin on the PIC16(L)F1938, therefore SEG15 is not available when using 1/4 multiplex displays.

FIGURE 27-1: LCD DRIVER MODULE BLOCK DIAGRAM



Note 1: These are not directly connected to the I/O pads, but may be tri-stated, depending on the configuration of the LCD module.

2: SEG<23:0> on PIC16(L)F1939, SEG<15:0> on PIC16(L)F1938.

3: COM3 and SEG15 share the same physical pin on the PIC16(L)F1938, therefore SEG15 is not available when using 1/4 multiplex displays.

27.10 LCD Waveform Generation

LCD waveforms are generated so that the net AC voltage across the dark pixel should be maximized and the net AC voltage across the clear pixel should be minimized. The net DC voltage across any pixel should be zero.

The COM signal represents the time slice for each common, while the SEG contains the pixel data.

The pixel signal (COM-SEG) will have no DC component and it can take only one of the two RMS values. The higher RMS value will create a dark pixel and a lower RMS value will create a clear pixel.

As the number of commons increases, the delta between the two RMS values decreases. The delta represents the maximum contrast that the display can have.

The LCDs can be driven by two types of waveform: Type-A and Type-B. In Type-A waveform, the phase changes within each common type, whereas in Type-B waveform, the phase changes on each frame boundary. Thus, Type-A waveform maintains 0 VDC over a single frame, whereas Type-B waveform takes two frames.

> **Note 1:** If Sleep has to be executed with LCD Sleep disabled (LCDCON<SLPEN> is '1'), then care must be taken to execute Sleep only when VDC on all the pixels is '0'.
>
> **2:** When the LCD clock source is FOSC/256, if Sleep is executed, irrespective of the LCDCON<SLPEN> setting, the LCD immediately goes into Sleep. Thus, take care to see that VDC on all pixels is '0' when Sleep is executed.

Figure 27-8 through Figure 27-18 provide waveforms for static, half-multiplex, 1/3-multiplex and 1/4-multiplex drives for Type-A and Type-B waveforms.

FIGURE 27-8: TYPE-A/TYPE-B WAVEFORMS IN STATIC DRIVE

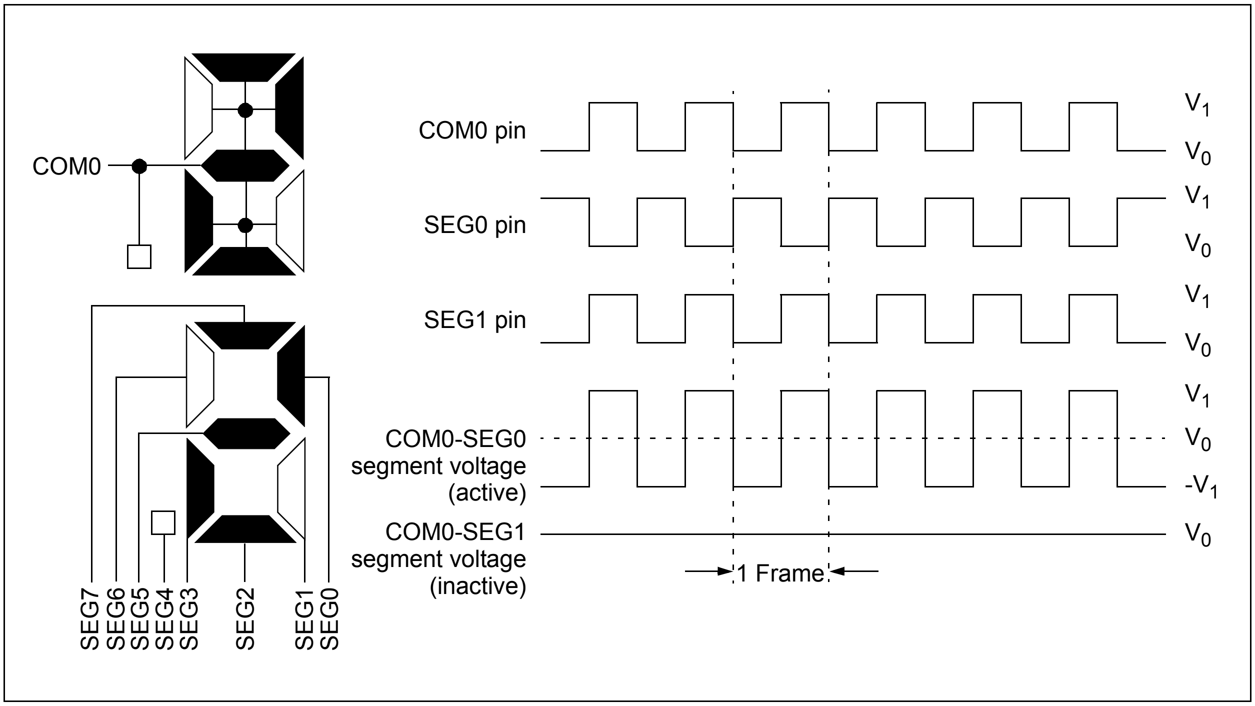


FIGURE 31-27: IDD TYPICAL, HS OSCILLATOR, PIC16LF1938/9 ONLY

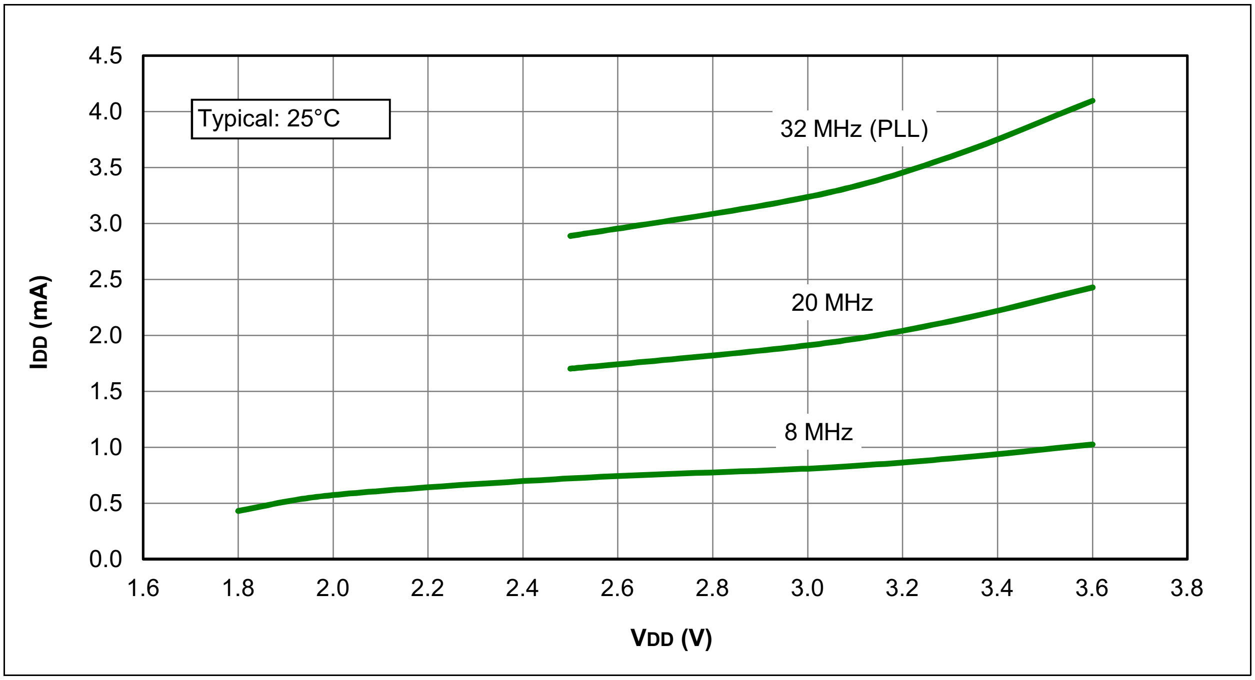


FIGURE 31-28: IDD MAXIMUM, HS OSCILLATOR, PIC16LF1938/9 ONLY

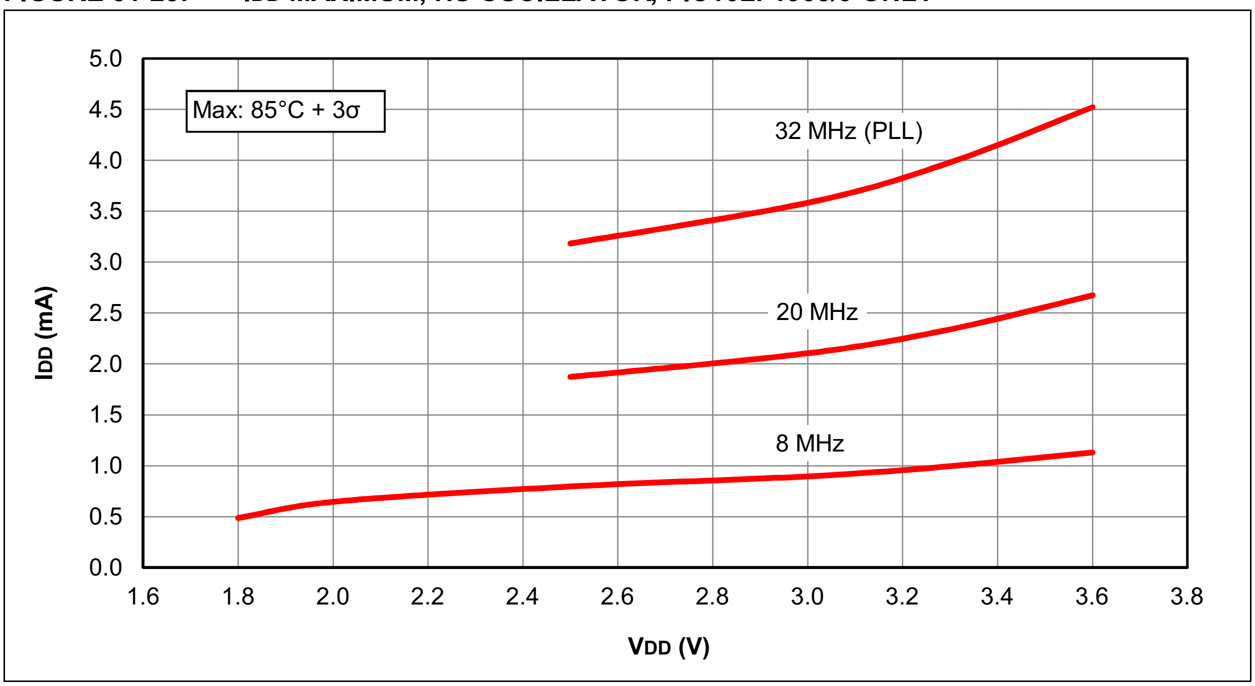


FIGURE 31-67: COMPARATOR RESPONSE TIME, NORMAL-POWER MODE, (CxSP = 1)

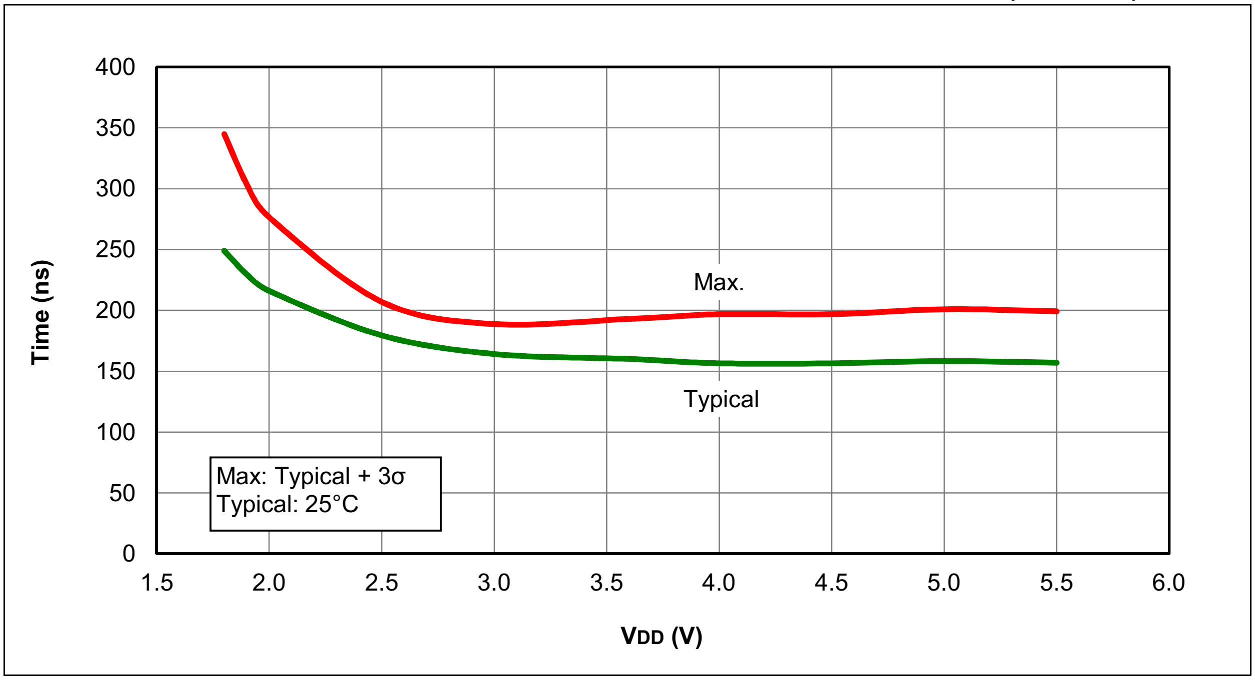


FIGURE 31-68: COMPARATOR RESPONSE TIME OVER TEMPERATURE, NORMAL-POWER MODE (CxSP = 1)