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Applications of "<u>Embedded -</u> <u>Microcontrollers</u>"

Details

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Product Status	Obsolete
Core Processor	PIC
Core Size	8-Bit
Speed	16MHz
Connectivity	I ² C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	50
Program Memory Size	32KB (16K x 16)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	902 x 8
Voltage - Supply (Vcc/Vdd)	4.5V ~ 5.5V
Data Converters	A/D 12x10b
Oscillator Type	External
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	68-LCC (J-Lead)
Supplier Device Package	68-PLCC (24.23x24.23)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic17c756at-16e-l

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

3.0 ARCHITECTURAL OVERVIEW

The high performance of the PIC17CXXX can be attributed to a number of architectural features, commonly found in RISC microprocessors. To begin with, the PIC17CXXX uses a modified Harvard architecture. This architecture has the program and data accessed from separate memories. So, the device has a program memory bus and a data memory bus. This improves bandwidth over traditional von Neumann architecture, where program and data are fetched from the same memory (accesses over the same bus). Separating program and data memory further allows instructions to be sized differently than the 8-bit wide data word. PIC17CXXX opcodes are 16-bits wide, enabling single word instructions. The full 16-bit wide program memory bus fetches a 16-bit instruction in a single cycle. A twostage pipeline overlaps fetch and execution of instructions. Consequently, all instructions execute in a single cycle (121 ns @ 33 MHz), except for program branches and two special instructions that transfer data between program and data memory.

The PIC17CXXX can address up to 64K x 16 of program memory space.

The **PIC17C752** and **PIC17C762** integrate 8K x 16 of EPROM program memory on-chip.

The **PIC17C756A** and **PIC17C766** integrate 16K x 16 EPROM program memory on-chip.

A simplified block diagram is shown in Figure 3-1. The descriptions of the device pins are listed in Table 3-1.

Program execution can be internal only (Microcontroller or Protected Microcontroller mode), external only (Microprocessor mode), or both (Extended Microcontroller mode). Extended Microcontroller mode does not allow code protection.

The PIC17CXXX can directly or indirectly address its register files or data memory. All special function registers, including the Program Counter (PC) and Working Register (WREG), are mapped in data memory. The PIC17CXXX has an orthogonal (symmetrical) instruction set that makes it possible to carry out any operation on any register using any addressing mode. This symmetrical nature and lack of 'special optimal situations' make programming with the PIC17CXXX simple, yet efficient. In addition, the learning curve is reduced significantly.

One of the PIC17CXXX family architectural enhancements from the PIC16CXX family, allows two file registers to be used in some two operand instructions. This allows data to be moved directly between two registers without going through the WREG register, thus increasing performance and decreasing program memory usage.

The PIC17CXXX devices contain an 8-bit ALU and working register. The ALU is a general purpose arithmetic unit. It performs arithmetic and Boolean functions between data in the working register and any register file.

The WREG register is an 8-bit working register used for ALU operations.

All PIC17CXXX devices have an 8 x 8 hardware multiplier. This multiplier generates a 16-bit result in a single cycle.

The ALU is 8-bits wide and capable of addition, subtraction, shift and logical operations. Unless otherwise mentioned, arithmetic operations are two's complement in nature.

Depending on the instruction executed, the ALU may affect the values of the Carry (C), Digit Carry (DC), Zero (Z) and Overflow (OV) bits in the ALUSTA register. The C and DC bits operate as a borrow and digit borrow out bit, respectively, in subtraction. See the SUBLW and SUBWF instructions for examples.

Signed arithmetic is comprised of a magnitude and a sign bit. The overflow bit indicates if the magnitude overflows and causes the sign bit to change state. That is, if the result of 8-bit signed operations is greater than 127 (7Fh), or less than -128 (80h).

Signed math can have greater than 7-bit values (magnitude), if more than one byte is used. The overflow bit only operates on bit6 (MSb of magnitude) and bit7 (sign bit) of each byte value in the ALU. That is, the overflow bit is not useful if trying to implement signed math where the magnitude, for example, is 11-bits.

If the signed math values are greater than 7-bits (such as 15-, 24-, or 31-bit), the algorithm must ensure that the low order bytes of the signed value ignore the overflow status bit.

Example 3-1 shows two cases of doing signed arithmetic. The Carry (C) bit and the Overflow (OV) bit are the most important status bits for signed math operations.

EXAMPLE 3-1: 8-BIT MATH ADDITION

Hex Value	Signed Values	Unsigned Values
FFh + 01h = 00h	-1 + 1 = 0 (FEh)	255 $\frac{+ 1}{= 256} \rightarrow 00h$
C bit = 1	C bit = 1	C bit = 1
OV bit = 0	OV bit = 0	OV bit = 0
DC bit = 1	DC bit = 1	DC bit = 1
Z bit = 1	Z bit = 1	Z bit = 1
Hex Value	Signed Values	Unsigned Values
7Fh <u>+ 01h</u> = 80h	127 $+ 1$ $= 128 \rightarrow 00h$	127 <u>+ 1</u> = 128
C bit = 0	C bit = 0	C bit = 0
OV bit = 1	OV bit = 1	OV bit = 1
DC bit = 1	DC bit = 1	DC bit = 1
Z bit = 0	Z bit = 0	Z bit = 0

4.1.6 RC OSCILLATOR

For timing insensitive applications, the RC device option offers additional cost savings. RC oscillator frequency is a function of the supply voltage, the resistor (REXT) and capacitor (CEXT) values, and the operating temperature. In addition to this, oscillator frequency will vary from unit to unit due to normal process parameter variation. Furthermore, the difference in lead frame capacitance between package types will also affect oscillation frequency, especially for low CEXT values. The user also needs to take into account variation due to tolerance of external R and C components used. Figure 4-7 shows how the R/C combination is connected to the PIC17CXXX. For REXT values below 2.2 k Ω , the oscillator operation may become unstable, or stop completely. For very high REXT values (e.g. 1 M Ω), the oscillator becomes sensitive to noise, humidity and leakage. Thus, we recommend to keep REXT between 3 k Ω and 100 k Ω .

Although the oscillator will operate with no external capacitor (CExT = 0 pF), we recommend using values above 20 pF for noise and stability reasons. With little or no external capacitance, oscillation frequency can vary dramatically due to changes in external capacitances, such as PCB trace capacitance or package lead frame capacitance.

See Section 21.0 for RC frequency variation from part to part due to normal process variation. The variation is larger for larger R (since leakage current variation will affect RC frequency more for large R) and for smaller C (since variation of input capacitance will affect RC frequency more).

See Section 21.0 for variation of oscillator frequency due to VDD for given REXT/CEXT values, as well as frequency variation due to operating temperature for given R, C, and VDD values.

The oscillator frequency, divided by 4, is available on the OSC2/CLKOUT pin and can be used for test purposes or to synchronize other logic (see Figure 4-8 for waveform).

FIGURE 4-7: RC OSCILLATOR MODE



4.1.6.1 RC Start-up

As the device voltage increases, the RC will immediately start its oscillations once the pin voltage levels meet the input threshold specifications (parameter #D032 and parameter #D042 in the electrical specification section). The time required for the RC to start oscillating depends on many factors. These include:

- Resistor value used
- · Capacitor value used
- Device VDD rise time
- System temperature

NOTES:

In Figure 5-5, Figure 5-6 and Figure 5-7, the TPWRT timer time-out is greater then the TOST timer time-out, as would be the case in higher frequency crystals. For lower frequency crystals (i.e., 32 kHz), TOST may be greater.



FIGURE 5-5: TIME-OUT SEQUENCE ON POWER-UP (MCLR TIED TO VDD)









7.0 MEMORY ORGANIZATION

There are two memory blocks in the PIC17C7XX; program memory and data memory. Each block has its own bus, so that access to each block can occur during the same oscillator cycle.

The data memory can further be broken down into General Purpose RAM and the Special Function Registers (SFRs). The operation of the SFRs that control the "core" are described here. The SFRs used to control the peripheral modules are described in the section discussing each individual peripheral module.

7.1 Program Memory Organization

PIC17C7XX devices have a 16-bit program counter capable of addressing a 64K x 16 program memory space. The RESET vector is at 0000h and the interrupt vectors are at 0008h, 0010h, 0018h, and 0020h (Figure 7-1).

7.1.1 PROGRAM MEMORY OPERATION

The PIC17C7XX can operate in one of four possible program memory configurations. The configuration is selected by configuration bits. The possible modes are:

- Microprocessor
- Microcontroller
- Extended Microcontroller
- Protected Microcontroller

The **Microcontroller** and **Protected Microcontroller** modes only allow internal execution. Any access beyond the program memory reads unknown data. The Protected Microcontroller mode also enables the code protection feature.

The **Extended Microcontroller** mode accesses both the internal program memory, as well as external program memory. Execution automatically switches between internal and external memory. The 16-bits of address allow a program memory range of 64K-words.

The **Microprocessor** mode only accesses the external program memory. The on-chip program memory is ignored. The 16-bits of address allow a program memory range of 64K-words. Microprocessor mode is the default mode of an unprogrammed device.

The different modes allow different access to the configuration bits, test memory and boot ROM. Table 7-1 lists which modes can access which areas in memory. Test Memory and Boot Memory are not required for normal operation of the device. Care should be taken to ensure that no unintended branches occur to these areas.

FIGURE 7-1:

PROGRAM MEMORY MAP AND STACK



7.3 Stack Operation

PIC17C7XX devices have a 16 x 16-bit hardware stack (Figure 7-1). The stack is not part of either the program or data memory space, and the stack pointer is neither readable nor writable. The PC (Program Counter) is "PUSH'd" onto the stack when a CALL or LCALL instruction is executed, or an interrupt is acknowledged. The stack is "POP'd" in the event of a RETURN, RETLW, or a RETFIE instruction execution. PCLATH is not affected by a "PUSH" or a "POP" operation.

The stack operates as a circular buffer, with the stack pointer initialized to '0' after all RESETS. There is a stack available bit (STKAV) to allow software to ensure that the stack will not overflow. The STKAV bit is set after a device RESET. When the stack pointer equals Fh, STKAV is cleared. When the stack pointer rolls over from Fh to 0h, the STKAV bit will be held clear until a device RESET.

- **Note 1:** There is not a status bit for stack underflow. The STKAV bit can be used to detect the underflow which results in the stack pointer being at the Top-of-Stack.
 - 2: There are no instruction mnemonics called PUSH or POP. These are actions that occur from the execution of the CALL, RETURN, RETLW and RETFIE instructions, or the vectoring to an interrupt vector.
 - 3: After a RESET, if a "POP" operation occurs before a "PUSH" operation, the STKAV bit will be cleared. This will appear as if the stack is full (underflow has occurred). If a "PUSH" operation occurs next (before another "POP"), the STKAV bit will be locked clear. Only a device RESET will cause this bit to set.

After the device is "PUSH'd" sixteen times (without a "POP"), the seventeenth push overwrites the value from the first push. The eighteenth push overwrites the second push (and so on).

7.4 Indirect Addressing

Indirect addressing is a mode of addressing data memory where the data memory address in the instruction is not fixed. That is, the register that is to be read or written can be modified by the program. This can be useful for data tables in the data memory. Figure 7-6 shows the operation of indirect addressing. This depicts the moving of the value to the data memory address specified by the value of the FSR register.

Example 7-1 shows the use of indirect addressing to clear RAM in a minimum number of instructions. A similar concept could be used to move a defined number of bytes (block) of data to the USART transmit register (TXREG). The starting address of the block of data to be transmitted could easily be modified by the program.

FIGURE 7-6: INDIRECT ADDRESSING



7.4.1 INDIRECT ADDRESSING REGISTERS

The PIC17C7XX has four registers for indirect addressing. These registers are:

- INDF0 and FSR0
- INDF1 and FSR1

Registers INDF0 and INDF1 are not physically implemented. Reading or writing to these registers activates indirect addressing, with the value in the corresponding FSR register being the address of the data. The FSR is an 8-bit register and allows addressing anywhere in the 256-byte data memory address range. For banked memory, the bank of memory accessed is specified by the value in the BSR.

If file INDF0 (or INDF1) itself is read indirectly via an FSR, all '0's are read (Zero bit is set). Similarly, if INDF0 (or INDF1) is written to indirectly, the operation will be equivalent to a NOP, and the status bits are not affected.

TABLE 10-11: PORTF FUNCTIONS

Name	Bit	Buffer Type	Function	
RF0/AN4	bit0	ST	Input/output or analog input 4.	
RF1/AN5	bit1	ST	Input/output or analog input 5.	
RF2/AN6	bit2	ST	Input/output or analog input 6.	
RF3/AN7	bit3	ST	Input/output or analog input 7.	
RF4/AN8	bit4	ST	Input/output or analog input 8.	
RF5/AN9	bit5	ST	Input/output or analog input 9.	
RF6/AN10	bit6	ST	Input/output or analog input 10.	
RF7/AN11	bit7	ST	Input/output or analog input 11.	

Legend: ST = Schmitt Trigger input

TABLE 10-12: REGISTERS/BITS ASSOCIATED WITH PORTF

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	MCLR, WDT
10h, Bank 5	DDRF	Data Dir	ection Reg	gister for P	ORTF					1111 1111	1111 1111
11h, Bank 5	PORTF	RF7/ AN11	RF6/ AN10	RF5/ AN9	RF4/ AN8	RF3/ AN7	RF2/ AN6	RF1/ AN5	RF0/ AN4	0000 0000	0000 0000
15h, Bank 5	ADCON1	ADCS1	ADCS0	ADFM		PCFG3	PCFG2	PCFG1	PCFG0	000- 0000	000- 0000

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by PORTF.

13.1.2 TIMER1 AND TIMER2 IN 16-BIT MODE

To select 16-bit mode, set the T16 bit. In this mode, TMR2 and TMR1 are concatenated to form a 16-bit timer (TMR2:TMR1). The 16-bit timer increments until it matches the 16-bit period register (PR2:PR1). On the following timer clock, the timer value is reset to 0h, and the TMR1IF bit is set.

When selecting the clock source for the 16-bit timer, the TMR1CS bit controls the entire 16-bit timer and TMR2CS is a "don't care", however, ensure that TMR2ON is set (allows TMR2 to increment). When TMR1CS is clear, the timer increments once every instruction cycle (Fosc/4). When TMR1CS is set, the timer increments on every falling edge of the RB4/TCLK12 pin. For the 16-bit timer to increment, both TMR1ON and TMR2ON bits must be set (Table 13-2).

TABLE 13-2: TURNING ON 16-BIT TIMER

T16	TMR2ON	TMR1ON	Result
1	1	1	16-bit timer (TMR2:TMR1) ON
1	0	1	Only TMR1 increments
1	х	0	16-bit timer OFF
0	1	1	Timers in 8-bit mode



13.1.2.1 External Clock Input for TMR2:TMR1

When TMR1CS is set, the 16-bit TMR2:TMR1 increments on the falling edge of clock input TCLK12. The input on the RB4/TCLK12 pin is sampled and synchronized by the internal phase clocks twice every instruction cycle. This causes a delay from the time a falling edge appears on RB4/TCLK12 to the time TMR2:TMR1 is actually incremented. For the external clock input timing requirements, see the Electrical Specification section.

14.3 USART Synchronous Master Mode

In Master Synchronous mode, the data is transmitted in a half-duplex manner; i.e., transmission and reception do not occur at the same time: when transmitting data, the reception is inhibited and vice versa. The synchronous mode is entered by setting the SYNC (TXSTA<4>) bit. In addition, the SPEN (RCSTA<7>) bit is set in order to configure the I/O pins to CK (clock) and DT (data) lines, respectively. The Master mode indicates that the processor transmits the master clock on the CK line. The Master mode is entered by setting the CSRC (TXSTA<7>) bit.

14.3.1 USART SYNCHRONOUS MASTER TRANSMISSION

The USART transmitter block diagram is shown in Figure 14-1. The heart of the transmitter is the transmit (serial) shift register (TSR). The shift register obtains its data from the read/write transmit buffer TXREG. TXREG is loaded with data in software. The TSR is not loaded until the last bit has been transmitted from the previous load. As soon as the last bit is transmitted, the TSR is loaded with new data from TXREG (if available). Once TXREG transfers the data to the TSR (occurs in one TCY at the end of the current BRG cycle), TXREG is empty and the TXIF bit is set. This interrupt can be enabled/disabled by setting/clearing the TXIE bit. TXIF will be set regardless of the state of bit TXIE and cannot be cleared in software. It will reset only when new data is loaded into TXREG. While TXIF indicates the status of TXREG, TRMT (TXSTA<1>) shows the status of the TSR. TRMT is a read only bit which is set when the TSR is empty. No interrupt logic is tied to this bit, so the user has to poll this bit in order to determine if the TSR is empty. The TSR is not mapped in data memory, so it is not available to the user.

Transmission is enabled by setting the TXEN (TXSTA<5>) bit. The actual transmission will not occur until TXREG has been loaded with data. The first data bit will be shifted out on the next available rising edge of the clock on the TX/CK pin. Data out is stable around the falling edge of the synchronous clock (Figure 14-9). The transmission can also be started by first loading TXREG and then setting TXEN. This is advantageous when slow baud rates are selected, since BRG is kept in RESET when the TXEN, CREN, and SREN bits are clear. Setting the TXEN bit will start the BRG, creating a shift clock immediately. Normally when transmission is first started, the TSR is empty, so a transfer to TXREG will result in an immediate transfer to the TSR, resulting in an empty TXREG. Back-to-back transfers are possible.

Clearing TXEN during a transmission will cause the transmission to be aborted and will reset the transmitter. The RX/DT and TX/CK pins will revert to hi-impedance. If either CREN or SREN are set during a transmission, the transmission is aborted and the RX/ DT pin reverts to a hi-impedance state (for a reception). The TX/CK pin will remain an output if the CSRC bit is set (internal clock). The transmitter logic is not reset, although it is disconnected from the pins. In order to reset the transmitter, the user has to clear the TXEN bit. If the SREN bit is set (to interrupt an ongoing transmission and receive a single word), then after the single word is received, SREN will be cleared and the serial port will revert back to transmitting, since the TXEN bit is still set. The DT line will immediately switch from hiimpedance Receive mode to transmit and start driving. To avoid this, TXEN should be cleared.

In order to select 9-bit transmission, the TX9 (TXSTA<6>) bit should be set and the ninth bit should be written to TX9D (TXSTA<0>). The ninth bit must be written before writing the 8-bit data to TXREG. This is because a data write to TXREG can result in an immediate transfer of the data to the TSR (if the TSR is empty). If the TSR was empty and TXREG was written before writing the "new" TX9D, the "present" value of TX9D is loaded.

Steps to follow when setting up a Synchronous Master Transmission:

- 1. Initialize the SPBRG register for the appropriate baud rate (see Baud Rate Generator Section for details).
- 2. Enable the synchronous master serial port by setting the SYNC, SPEN, and CSRC bits.
- 3. Ensure that the CREN and SREN bits are clear (these bits override transmission when set).
- If interrupts are desired, then set the TXIE bit (the GLINTD bit must be clear and the PEIE bit must be set).
- 5. If 9-bit transmission is desired, then set the TX9 bit.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in TX9D.
- 7. Start transmission by loading data to the TXREG register.
- 8. Enable the transmission by setting TXEN.

Writing the transmit data to the TXREG, then enabling the transmit (setting TXEN), allows transmission to start sooner than doing these two events in the reverse order.

Note: To terminate a transmission, either clear the SPEN bit, or the TXEN bit. This will reset the transmit logic, so that it will be in the proper state when transmit is reenabled.

14.4 USART Synchronous Slave Mode

The Synchronous Slave mode differs from the Master mode, in the fact that the shift clock is supplied externally at the TX/CK pin (instead of being supplied internally in the Master mode). This allows the device to transfer or receive data in the SLEEP mode. The Slave mode is entered by clearing the CSRC (TXSTA<7>) bit.

14.4.1 USART SYNCHRONOUS SLAVE TRANSMIT

The operation of the SYNC Master and Slave modes are identical except in the case of the SLEEP mode.

If two words are written to TXREG and then the SLEEP instruction executes, the following will occur. The first word will immediately transfer to the TSR and will transmit as the shift clock is supplied. The second word will remain in TXREG. TXIF will not be set. When the first word has been shifted out of TSR, TXREG will transfer the second word to the TSR and the TXIF flag will now be set. If TXIE is enabled, the interrupt will wake the chip from SLEEP and if the global interrupt is enabled, then the program will branch to the interrupt vector (0020h).

Steps to follow when setting up a Synchronous Slave Transmission:

- 1. Enable the synchronous slave serial port by setting the SYNC and SPEN bits and clearing the CSRC bit.
- 2. Clear the CREN bit.
- 3. If interrupts are desired, then set the TXIE bit.
- 4. If 9-bit transmission is desired, then set the TX9 bit.
- 5. If 9-bit transmission is selected, the ninth bit should be loaded in TX9D.
- 6. Start transmission by loading data to TXREG.
- 7. Enable the transmission by setting TXEN.

Writing the transmit data to the TXREG, then enabling the transmit (setting TXEN), allows transmission to start sooner than doing these two events in the reverse order.



14.4.2 USART SYNCHRONOUS SLAVE RECEPTION

Operation of the Synchronous Master and Slave modes are identical except in the case of the SLEEP mode. Also, SREN is a "don't care" in Slave mode.

If receive is enabled (CREN) prior to the SLEEP instruction, then a word may be received during SLEEP. On completely receiving the word, the RSR will transfer the data to RCREG (setting RCIF) and if the RCIE bit is set, the interrupt generated will wake the chip from SLEEP. If the global interrupt is enabled, the program will branch to the interrupt vector (0020h).

Steps to follow when setting up a Synchronous Slave Reception:

- 1. Enable the synchronous master serial port by setting the SYNC and SPEN bits and clearing the CSRC bit.
- 2. If interrupts are desired, then set the RCIE bit.
- 3. If 9-bit reception is desired, then set the RX9 bit.
- 4. To enable reception, set the CREN bit.
- The RCIF bit will be set when reception is complete and an interrupt will be generated if the RCIE bit was set.
- 6. Read RCSTA to get the ninth bit (if enabled) and determine if any error occurred during reception.
- 7. Read the 8-bit received data by reading RCREG.
- 8. If any error occurred, clear the error by clearing the CREN bit.

Note: To abort reception, either clear the SPEN bit, or the CREN bit (when in Continuous Receive mode). This will reset the receive logic, so that it will be in the proper state when receive is re-enabled.

When the application software is expecting to receive valid data, the SSPBUF should be read before the next byte of data to transfer is written to the SSPBUF. Buffer full bit, BF (SSPSTAT<0>), indicates when SSPBUF has been loaded with the received data (transmission is complete). When the SSPBUF is read, bit BF is cleared. This data may be irrelevant if the SPI is only a transmitter. Generally the MSSP interrupt is used to determine when the transmission/reception has completed. The SSPBUF must be read and/or written. If the interrupt method is not going to be used, then software polling can be done to ensure that a write collision does not occur. Example 15-1 shows the loading of the SSPBUF (SSPSR) for data transmission.

EXAMPLE 15-1: LOADING THE SSPBUF (SSPSR) REGISTER

	MOVLB	6		;	Bank 6
LOOP	BTFSS	SSPSTAT	, BF	;	Has data been
				;	received
				;	(transmit
				;	complete)?
	GOTO	LOOP		;	No
	MOVPF	SSPBUF,	RXDATA	;	Save in user RAM
	MOVFP	TXDATA,	SSPBUF	;	New data to xmit

The SSPSR is not directly readable, or writable and can only be accessed by addressing the SSPBUF register. Additionally, the MSSP status register (SSPSTAT) indicates the various status conditions.

15.1.2 ENABLING SPI I/O

To enable the serial port, MSSP Enable bit, SSPEN (SSPCON1<5>), must be set. To reset or reconfigure SPI mode, clear bit SSPEN, re-initialize the SSPCON registers and then set bit SSPEN. This configures the SDI, SDO, SCK and SS pins as serial port pins. For the pins to behave as the serial port function, some must have their data direction bits (in the DDR register) appropriately programmed. That is:

- SDI is automatically controlled by the SPI module
- SDO must have DDRB<7> cleared
- SCK (Master mode) must have DDRB<6> cleared
- SCK (Slave mode) must have DDRB<6> set
- SS must have PORTA<2> set

Any serial port function that is not desired may be overridden by programming the corresponding data direction (DDR) register to the opposite value.

15.1.3 TYPICAL CONNECTION

Figure 15-5 shows a typical connection between two microcontrollers. The master controller (Processor 1) initiates the data transfer by sending the SCK signal. Data is shifted out of both shift registers on their programmed clock edge and latched on the opposite edge of the clock. Both processors should be programmed to same Clock Polarity (CKP), then both controllers would send and receive data at the same time. Whether the data is meaningful (or dummy data) depends on the application software. This leads to three scenarios for data transmission:

- Master sends data Slave sends dummy data
- Master sends data Slave sends data
- Master sends dummy data Slave sends data



FIGURE 15-5: SPI MASTER/SLAVE CONNECTION



FIGURE 15-9: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 1)



15.2.14 STOP CONDITION TIMING

A STOP bit is asserted on the SDA pin at the end of a receive/transmit by setting the Stop Sequence Enable bit PEN (SSPCON2<2>). At the end of a receive/ transmit the SCL line is held low after the falling edge of the ninth clock. When the PEN bit is set, the master will assert the SDA line low. When the SDA line is sampled low, the baud rate generator is reloaded and counts down to '0'. When the baud rate generator times out, the SCL pin will be brought high and one TBRG (baud rate generator rollover count) later, the SDA pin will be de-asserted. When the SDA pin is sampled high while SCL is high, the P bit (SSPSTAT<4>) is set. A TBRG later, the PEN bit is cleared and the SSPIF bit is set (Figure 15-31).

Whenever the firmware decides to take control of the bus, it will first determine if the bus is busy by checking the S and P bits in the SSPSTAT register. If the bus is busy, then the CPU can be interrupted (notified) when a STOP bit is detected (i.e., bus is free).

15.2.14.1 WCOL Status Flag

If the user writes the SSPBUF when a STOP sequence is in progress, then WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).



FIGURE 15-31: STOP CONDITION RECEIVE OR TRANSMIT MODE

15.4 Example Program

Example 15-2 shows MPLAB[®] C17 'C' code for using the I²C module in Master mode to communicate with a 24LC01B serial EEPROM. This example uses the PIC[®] MCU 'C' libraries included with MPLAB C17.

EXAMPLE 15-2: INTERFACING TO A 24LC01B SERIAL EEPROM (USING MPLAB C17)

```
// Include necessary header files
#include <p17c756.h>
                       // Processor header file
                       // Delay routines header file
// Standard Library header file
#include <delays.h>
#include <stdlib.h>
                       // Standard Lizzard
// I2C routines header file
#include <i2c16.h>
#define CONTROL 0xa0
                         // Control byte definition for 24LC01B
// Function declarations
void main(void);
void WritePORTD(static unsigned char data);
void ByteWrite(static unsigned char address, static unsigned char data);
unsigned char ByteRead(static unsigned char address);
void ACKPoll(void);
// Main program
void main(void)
{
static unsigned char address; // I2C address of 24LC01B
static unsigned char datao; // Data written to 24LC01B
static unsigned char datai;
                                 // Data read from 24LC01B
                                  // Preset address to 0
    address = 0;
   OpenI2C(MASTER,SLEW_ON);
                                  \ensuremath{//} Configure I2C Module Master mode, Slew rate control on
   SSPADD = 39;
                                 // Configure clock for 100KHz
    while(address<128)
                                 // Loop 128 times, 24LC01B is 128x8
    {
        datao = PORTB;
        do
        {
            ByteWrite(address,datao); // Write data to EEPROM
            ACKPoll();
                                         // Poll the 24LC01B for state
            datai = ByteRead(address); // Read data from EEPROM into SSPBUF
        while(datai != datao);
                                        // Loop as long as data not correctly
                                         11
                                              written to 24LC01B
        address++;
                                         // Increment address
    }
    while(1)
                                         // Done writing 128 bytes to 24LC01B, Loop forever
    {
        Nop();
    }
```

FIGURE 16-3: ANALOG INPUT MODEL



NOTES:

20.0 PIC17C7XX ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings †

Ambient temperature under bias	55°C to +125°C
Storage temperature	65°C to +150°C
Voltage on VDD with respect to Vss	0 V to +7.5 V
Voltage on MCLR with respect to Vss (Note 2)	0.3 V to +14 V
Voltage on RA2 and RA3 with respect to Vss	0.3 V to +8.5 V
Voltage on all other pins with respect to Vss	0.3 V to VDD + 0.3 V
Total power dissipation (Note 1)	1.0 W
Maximum current out of Vss pin(s) - total (@ 70°C)	500 mA
Maximum current into VDD pin(s) - total(@ 70°C)	500 mA
Input clamp current, Iк (Vi < 0 or Vi > VDD)	±20 mA
Output clamp current, Iок (Vo < 0 or Vo > Voo)	±20 mA
Maximum output current sunk by any I/O pin (except RA2 and RA3)	35 mA
Maximum output current sunk by RA2 or RA3 pins	60 mA
Maximum output current sourced by any I/O pin	20 mA
Maximum current sunk by PORTA and PORTB (combined)	150 mA
Maximum current sourced by PORTA and PORTB (combined)	100 mA
Maximum current sunk by PORTC, PORTD and PORTE (combined)	150 mA
Maximum current sourced by PORTC, PORTD and PORTE (combined)	100 mA
Maximum current sunk by PORTF and PORTG (combined)	150 mA
Maximum current sourced by PORTF and PORTG (combined)	100 mA
Maximum current sunk by PORTH and PORTJ (combined)	150 mA
Maximum current sourced by PORTH and PORTJ (combined)	100 mA
Note 1: Power dissipation is calculated as follows: Pdis = VDD x {IDD - \sum IOH} + \sum {(VD	D-VOH) x IOH} + Σ (VOL x IOL)

2: Voltage spikes below Vss at the $\overline{\text{MCLR}}$ pin, inducing currents greater than 80 mA, may cause latch-up. Thus, a series resistor of 50-100 Ω should be used when applying a "low" level to the $\overline{\text{MCLR}}$ pin, rather than pulling this pin directly to Vss.

† NOTICE: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.









22.0 PACKAGING INFORMATION

22.1 Package Marking Information



Legend	: XXX Y YY WW NNN (©3) *	Customer-specific information Year code (last digit of calendar year) Year code (last 2 digits of calendar year) Week code (week of January 1 is week '01') Alphanumeric traceability code Pb-free JEDEC designator for Matte Tin (Sn) This package is Pb-free. The Pb-free JEDEC designator ((e3)) can be found on the outer packaging for this package.
Note:	In the even be carried characters	nt the full Microchip part number cannot be marked on one line, it will d over to the next line, thus limiting the number of available s for customer-specific information.

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