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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	PIC
Core Size	8-Bit
Speed	33MHz
Connectivity	I²C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	66
Program Memory Size	16KB (8K x 16)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	678 x 8
Voltage - Supply (Vcc/Vdd)	4.5V ~ 5.5V
Data Converters	A/D 16x10b
Oscillator Type	External
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	80-TQFP
Supplier Device Package	80-TQFP (12x12)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic17c762-33e-pt

NOTES:

TABLE 5-4: INITIALIZATION CONDITIONS FOR SPECIAL FUNCTION REGISTERS (CONTINUED)

Register	Address	Power-on Reset Brown-out Reset	$\overline{\text{MCLR}}$ Reset WDT Reset	Wake-up from SLEEP through Interrupt
Bank 4				
PIR2	10h	000- 0010	000- 0010	uuu- uuuu(1)
PIE2	11h	000- 0000	000- 0000	uuu- uuuu
Unimplemented	12h	---- ----	---- ----	---- ----
RCSTA2	13h	0000 -00x	0000 -00u	uuuu -uuu
RCREG2	14h	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXSTA2	15h	0000 --1x	0000 --1u	uuuu --uu
TXREG2	16h	xxxx xxxx	uuuu uuuu	uuuu uuuu
SPBRG2	17h	0000 0000	0000 0000	uuuu uuuu
Bank 5				
DDRF	10h	1111 1111	1111 1111	uuuu uuuu
PORTF(4)	11h	0000 0000	0000 0000	uuuu uuuu
DDRG	12h	1111 1111	1111 1111	uuuu uuuu
PORTG(4)	13h	xxxx 0000	uuuu 0000	uuuu uuuu
ADCON0	14h	0000 -0-0	0000 -0-0	uuuu uuuu
ADCON1	15h	000- 0000	000- 0000	uuuu uuuu
ADRESL	16h	xxxx xxxx	uuuu uuuu	uuuu uuuu
ADRESH	17h	xxxx xxxx	uuuu uuuu	uuuu uuuu
Bank 6				
SSPADD	10h	0000 0000	0000 0000	uuuu uuuu
SSPCON1	11h	0000 0000	0000 0000	uuuu uuuu
SSPCON2	12h	0000 0000	0000 0000	uuuu uuuu
SSPSTAT	13h	0000 0000	0000 0000	uuuu uuuu
SSPBUF	14h	xxxx xxxx	uuuu uuuu	uuuu uuuu
Unimplemented	15h	---- ----	---- ----	---- ----
Unimplemented	16h	---- ----	---- ----	---- ----
Unimplemented	17h	---- ----	---- ----	---- ----

Legend: u = unchanged, x = unknown, - = unimplemented, read as '0', q = value depends on condition

Note 1: One or more bits in INTSTA, PIR1, PIR2 will be affected (to cause wake-up).

2: When the wake-up is due to an interrupt and the GLINTD bit is cleared, the PC is loaded with the interrupt vector.

3: See Table 5-3 for RESET value of specific condition.

4: This is the value that will be in the port output latch.

5: When the device is configured for Microprocessor or Extended Microcontroller mode, the operation of this port does not rely on these registers.

6: On any device RESET, these pins are configured as inputs.

TABLE 5-4: INITIALIZATION CONDITIONS FOR SPECIAL FUNCTION REGISTERS (CONTINUED)

Register	Address	Power-on Reset Brown-out Reset	$\overline{\text{MCLR}}$ Reset WDT Reset	Wake-up from SLEEP through Interrupt
Bank 7				
PW3DCL	10h	xx0- ----	uu0- ----	uuu- ----
PW3DCH	11h	xxxx xxxx	uuuu uuuu	uuuu uuuu
CA3L	12h	xxxx xxxx	uuuu uuuu	uuuu uuuu
CA3H	13h	xxxx xxxx	uuuu uuuu	uuuu uuuu
CA4L	14h	xxxx xxxx	uuuu uuuu	uuuu uuuu
CA4H	15h	xxxx xxxx	uuuu uuuu	uuuu uuuu
TCON3	16h	-000 0000	-000 0000	-uuu uuuu
Unimplemented	17h	---- ----	---- ----	---- ----
Bank 8				
DDRH	10h	1111 1111	1111 1111	uuuu uuuu
PORTH(4)	11h	xxxx xxxx	uuuu uuuu	uuuu uuuu
DDRJ	12h	1111 1111	1111 1111	uuuu uuuu
PORTJ(4)	13h	xxxx xxxx	uuuu uuuu	uuuu uuuu
Unbanked				
PRODL	18h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PRODH	19h	xxxx xxxx	uuuu uuuu	uuuu uuuu

Legend: u = unchanged, x = unknown, - = unimplemented, read as '0', q = value depends on condition

Note 1: One or more bits in INTSTA, PIR1, PIR2 will be affected (to cause wake-up).

2: When the wake-up is due to an interrupt and the GLINTD bit is cleared, the PC is loaded with the interrupt vector.

3: See Table 5-3 for RESET value of specific condition.

4: This is the value that will be in the port output latch.

5: When the device is configured for Microprocessor or Extended Microcontroller mode, the operation of this port does not rely on these registers.

6: On any device RESET, these pins are configured as inputs.

7.8 Bank Select Register (BSR)

The BSR is used to switch between banks in the data memory area (Figure 7-9). In the PIC17C7XX devices, the entire byte is implemented. The lower nibble is used to select the peripheral register bank. The upper nibble is used to select the general purpose memory bank.

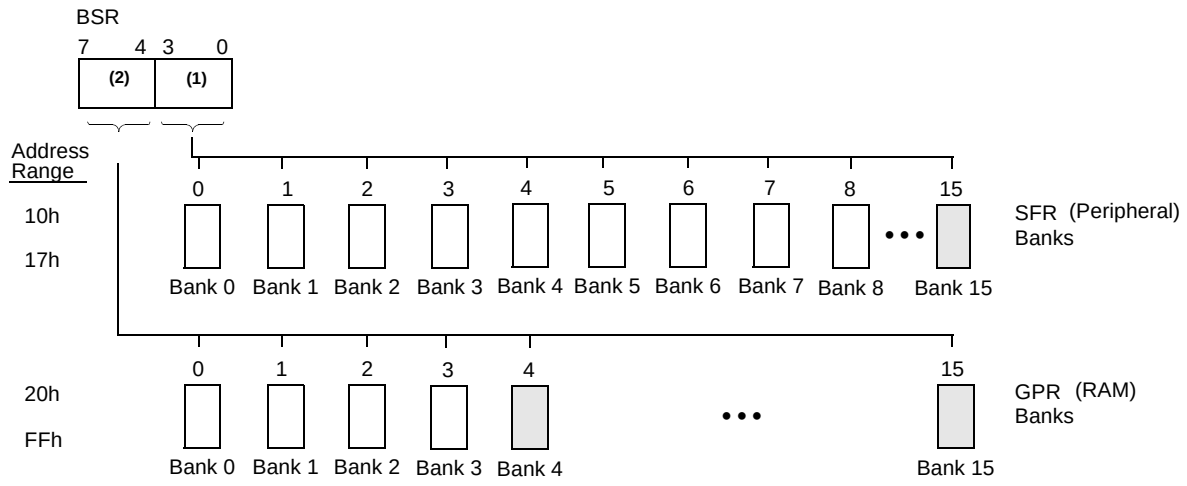
All the Special Function Registers (SFRs) are mapped into the data memory space. In order to accommodate the large number of registers, a banking scheme has been used. A segment of the SFRs, from address 10h to address 17h, is banked. The lower nibble of the bank select register (BSR) selects the currently active “peripheral bank.” Effort has been made to group the peripheral registers of related functionality in one bank. However, it will still be necessary to switch from bank to bank in order to address all peripherals related to a single task. To assist this, a `MOVLB bank` instruction has been included in the instruction set.

The need for a large general purpose memory space dictated a general purpose RAM banking scheme. The upper nibble of the BSR selects the currently active general purpose RAM bank. To assist this, a `MOVLR bank` instruction has been provided in the instruction set.

If the currently selected bank is not implemented (such as Bank 13), any read will read all '0's. Any write is completed to the bit bucket and the ALU status bits will be set/cleared as appropriate.

Note:	Registers in Bank 15 in the Special Function Register area, are reserved for Microchip use. Reading of registers in this bank may cause random values to be read.

FIGURE 7-9: BSR OPERATION

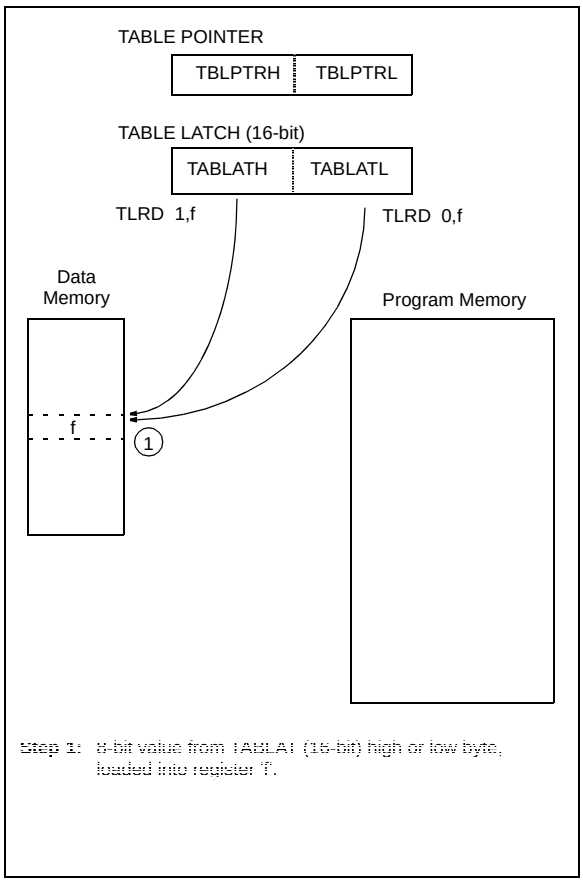


Note 1: For the SFRs only Banks 0 through 8 are implemented. Selection of an unimplemented bank is not recommended. **Bank 15 is reserved for Microchip use, reading of registers in this bank may cause random values to be read.**

2: For the GPRs, Bank 3 is unimplemented on the PIC17C752 and the PIC17C762. Selection of an unimplemented bank is not recommended.

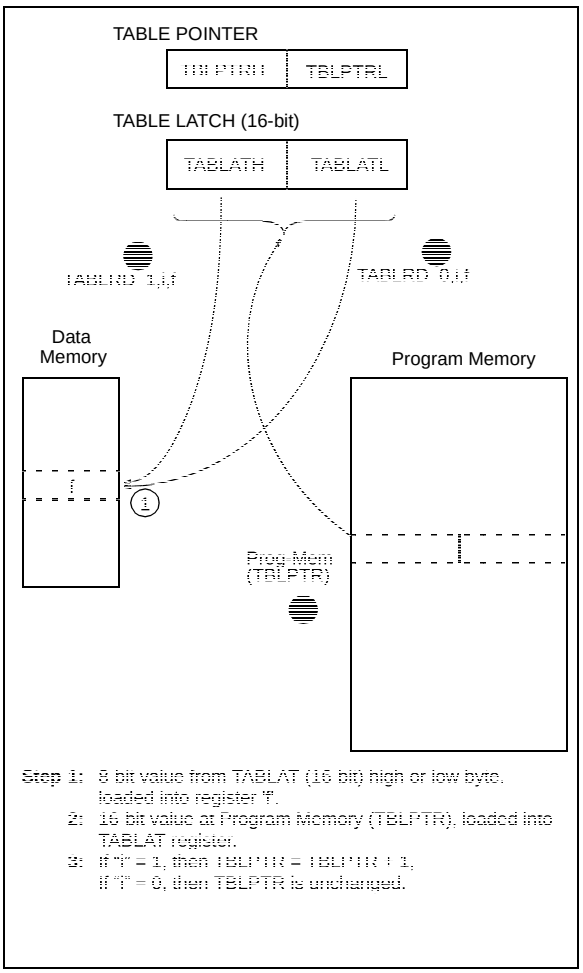
3: SFR Bank 8 is only implemented on the PIC17C76X.

FIGURE 8-3: `TLRD` INSTRUCTION OPERATION



Step 1: 8-bit value from TABLAT (16-bit) high or low byte, loaded into register 'f'.

FIGURE 8-4: `TABLRD` INSTRUCTION OPERATION



Step 1: 8 bit value from TABLAT (16 bit) high or low byte, loaded into register 'f'.
2: 16 bit value at Program Memory (TBLPTR), loaded into TABLAT register.
3: If "t" = 1, then TBLPTR = TBLPTR + 1,
If "t" = 0, then TBLPTR is unchanged.

10.10.2 SUCCESSIVE OPERATIONS ON I/O PORTS

The actual write to an I/O port happens at the end of an instruction cycle, whereas for reading, the data must be valid at the beginning of the instruction cycle (Figure 10-20). Therefore, care must be exercised if a write followed by a read operation is carried out on the same I/O port. The sequence of instructions should be such to allow the pin voltage to stabilize (load dependent) before executing the instruction that reads the values on that I/O port. Otherwise, the previous state of that pin may be read into the CPU, rather than the "new" state. When in doubt, it is better to separate these instructions with a `NOP`, or another instruction not accessing this I/O port.

Figure 10-21 shows the I/O model which causes this situation. As the effective capacitance (C) becomes larger, the rise/fall time of the I/O pin increases. As the device frequency increases, or the effective capacitance increases, the possibility of this subsequent PORTx read-modify-write instruction issue increases. This effective capacitance includes the effects of the board traces.

The best way to address this is to add a series resistor at the I/O pin. This resistor allows the I/O pin to get to the desired level before the next instruction.

The use of `NOP` instructions between the subsequent PORTx read-modify-write instructions, is a lower cost solution, but has the issue that the number of `NOP` instructions is dependent on the effective capacitance C and the frequency of the device.

FIGURE 10-20: SUCCESSIVE I/O OPERATION

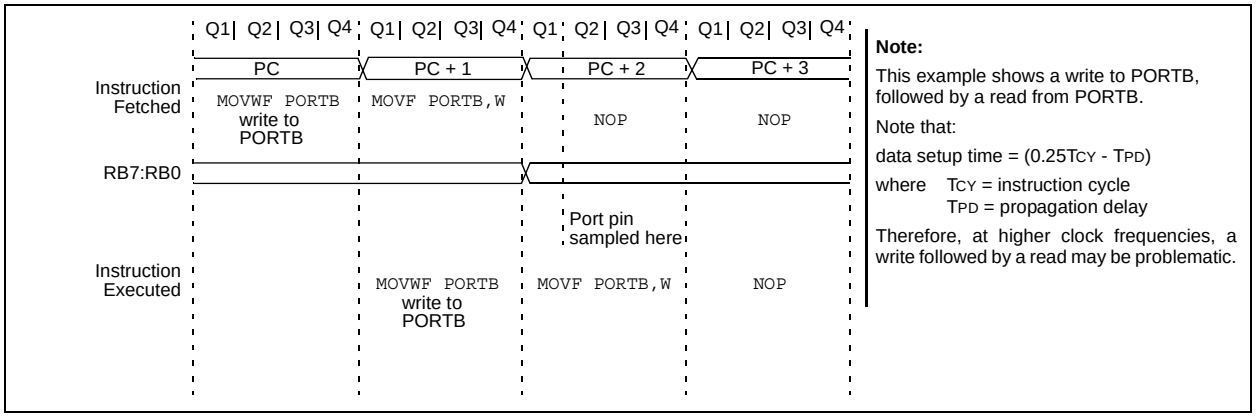
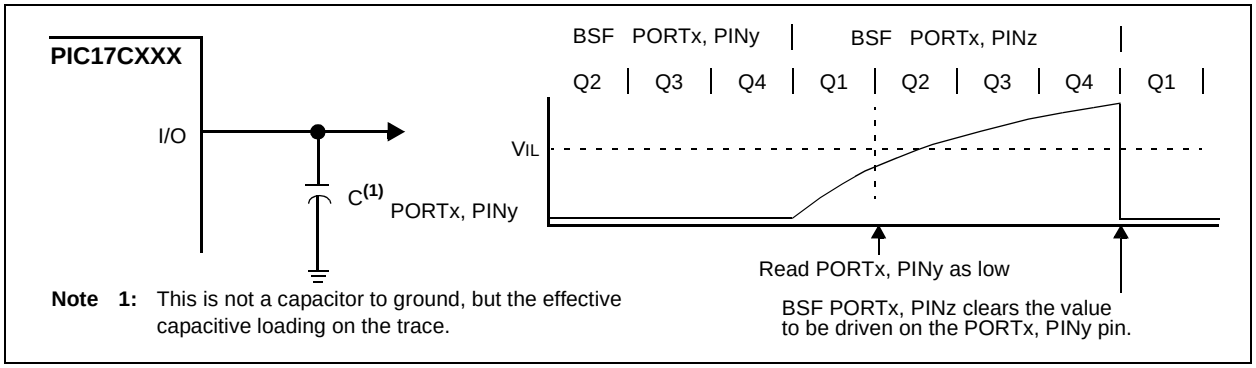


FIGURE 10-21: I/O CONNECTION ISSUES



Note 1: This is not a capacitor to ground, but the effective capacitive loading on the trace.

13.0 TIMER1, TIMER2, TIMER3, PWMS AND CAPTURES

The PIC17C7XX has a wealth of timers and time based functions to ease the implementation of control applications. These time base functions include three PWM outputs and four Capture inputs.

Timer1 and Timer2 are two 8-bit incrementing timers, each with an 8-bit period register (PR1 and PR2, respectively) and separate overflow interrupt flags. Timer1 and Timer2 can operate either as timers (increment on internal FOSC/4 clock), or as counters (increment on falling edge of external clock on pin RB4/TCLK12). They are also software configurable to operate as a single 16-bit timer/counter. These timers are also used as the time base for the PWM (Pulse Width Modulation) modules.

Timer3 is a 16-bit timer/counter which uses the TMR3H and TMR3L registers. Timer3 also has two additional registers (PR3H/CA1H:PR3L/CA1L) that are configurable as a 16-bit period register or a 16-bit capture register. TMR3 can be software configured to increment from the internal system clock (FOSC/4), or from an external signal on the RB5/TCLK3 pin. Timer3 is the time base for all of the 16-bit captures.

Six other registers comprise the Capture2, Capture3, and Capture4 registers (CA2H:CA2L, CA3H:CA3L, and CA4H:CA4L).

Figure 13-1, Figure 13-2 and Figure 13-3 are the control registers for the operation of Timer1, Timer2 and Timer3, as well as PWM1, PWM2, PWM3, Capture1, Capture2, Capture3 and Capture4.

Table 13-1 shows the Timer resource requirements for these time base functions. Each timer is an open resource so that multiple functions may operate with it.

TABLE 13-1: TIME-BASE FUNCTION/ RESOURCE REQUIREMENTS

Time Base Function	Timer Resource
PWM1	Timer1
PWM2	Timer1 or Timer2
PWM3	Timer1 or Timer2
Capture1	Timer3
Capture2	Timer3
Capture3	Timer3
Capture4	Timer3

REGISTER 13-1: TCON1 REGISTER (ADDRESS: 16h, BANK 3)

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
CA2ED1	CA2ED0	CA1ED1	CA1ED0	T16	TMR3CS	TMR2CS	TMR1CS
bit 7							bit 0

bit 7-6 **CA2ED1:CA2ED0**: Capture2 Mode Select bits
00 = Capture on every falling edge
01 = Capture on every rising edge
10 = Capture on every 4th rising edge
11 = Capture on every 16th rising edge

bit 5-4 **CA1ED1:CA1ED0**: Capture1 Mode Select bits
00 = Capture on every falling edge
01 = Capture on every rising edge
10 = Capture on every 4th rising edge
11 = Capture on every 16th rising edge

bit 3 **T16**: Timer2:Timer1 Mode Select bit
1 = Timer2 and Timer1 form a 16-bit timer
0 = Timer2 and Timer1 are two 8-bit timers

bit 2 **TMR3CS**: Timer3 Clock Source Select bit
1 = TMR3 increments off the falling edge of the RB5/TCLK3 pin
0 = TMR3 increments off the internal clock

bit 1 **TMR2CS**: Timer2 Clock Source Select bit
1 = TMR2 increments off the falling edge of the RB4/TCLK12 pin
0 = TMR2 increments off the internal clock

bit 0 **TMR1CS**: Timer1 Clock Source Select bit
1 = TMR1 increments off the falling edge of the RB4/TCLK12 pin
0 = TMR1 increments off the internal clock

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
- n = Value at POR Reset	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

13.1.3.3 External Clock Source

The PWMs will operate, regardless of the clock source of the timer. The use of an external clock has ramifications that must be understood. Because the external TCLK12 input is synchronized internally (sampled once per instruction cycle), the time TCLK12 changes to the time the timer increments, will vary by as much as 1TCY (one instruction cycle). This will cause jitter in the duty cycle as well as the period of the PWM output.

This jitter will be ±1TCY, unless the external clock is synchronized with the processor clock. Use of one of the PWM outputs as the clock source to the TCLK12 input, will supply a synchronized clock.

In general, when using an external clock source for PWM, its frequency should be much less than the device frequency (FOSC).

13.1.3.4 Maximum Resolution/Frequency for External Clock Input

The use of an external clock for the PWM time base (Timer1 or Timer2) limits the PWM output to a maximum resolution of 8-bits. The PWxDCL<7:6> bits must be kept cleared. Use of any other value will distort the PWM output. All resolutions are supported when internal clock mode is selected. The maximum attainable frequency is also lower. This is a result of the timing requirements of an external clock input for a timer (see the Electrical Specification section). The maximum PWM frequency, when the timers clock source is the RB4/TCLK12 pin, is shown in Table 13-4 (Standard Resolution mode).

TABLE 13-5: REGISTERS/BITS ASSOCIATED WITH PWM

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	$\overline{\text{MCLR}}$, WDT
16h, Bank 3	TCON1	CA2ED1	CA2ED0	CA1ED1	CA1ED0	T16	TMR3CS	TMR2CS	TMR1CS	0000 0000	0000 0000
17h, Bank 3	TCON2	CA2OVF	CA1OVF	PWM2ON	PWM1ON	CA1/$\overline{\text{PR3}}$	TMR3ON	TMR2ON	TMR1ON	0000 0000	0000 0000
16h, Bank 7	TCON3	—	CA4OVF	CA3OVF	CA4ED1	CA4ED0	CA3ED1	CA3ED0	PWM3ON	-000 0000	-000 0000
10h, Bank 2	TMR1	Timer1's Register								xxxx xxxx	uuuu uuuu
11h, Bank 2	TMR2	Timer2's Register								xxxx xxxx	uuuu uuuu
16h, Bank 1	PIR1	RBIF	TMR3IF	TMR2IF	TMR1IF	CA2IF	CA1IF	TX1IF	RC1IF	x000 0010	u000 0010
17h, Bank 1	PIE1	RBIE	TMR3IE	TMR2IE	TMR1IE	CA2IE	CA1IE	TX1IE	RC1IE	0000 0000	0000 0000
07h, Unbanked	INTSTA	PEIF	T0CKIF	T0IF	INTF	PEIE	T0CKIE	T0IE	INTE	0000 0000	0000 0000
06h, Unbanked	CPUSTA	—	—	STKAV	GLINTD	$\overline{\text{TO}}$	$\overline{\text{PD}}$	POR	BOR	--11 11qq	--11 qquu
14h, Bank 2	PR1	Timer1 Period Register								xxxx xxxx	uuuu uuuu
15h, Bank 2	PR2	Timer2 Period Register								xxxx xxxx	uuuu uuuu
10h, Bank 3	PW1DCL	DC1	DC0	—	—	—	—	—	—	xx-- ----	uu-- ----
11h, Bank 3	PW2DCL	DC1	DC0	TM2PW2	—	—	—	—	—	xx0- ----	uu0- ----
10h, Bank 7	PW3DCL	DC1	DC0	TM2PW3	—	—	—	—	—	xx0- ----	uu0- ----
12h, Bank 3	PW1DCH	DC9	DC8	DC7	DC6	DC5	DC4	DC3	DC2	xxxx xxxx	uuuu uuuu
13h, Bank 3	PW2DCH	DC9	DC8	DC7	DC6	DC5	DC4	DC3	DC2	xxxx xxxx	uuuu uuuu
11h, Bank 7	PW3DCH	DC9	DC8	DC7	DC6	DC5	DC4	DC3	DC2	xxxx xxxx	uuuu uuuu

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0', q = value depends on conditions.
Shaded cells are not used by PWM Module.

This mode (3 Capture, 1 Period) is selected if control bit CA1/PR3 is clear. In this mode, the Capture1 register, consisting of high byte (PR3H/CA1H) and low byte (PR3L/CA1L), is configured as the period control register for TMR3. Capture1 is disabled in this mode and the corresponding interrupt bit, CA1IF, is never set. TMR3 increments until it equals the value in the period register and then resets to 0000h on the next timer clock.

All other Captures are active in this mode.

13.2.1.1 Capture Operation

The CAxED1 and CAxED0 bits determine the event on which capture will occur. The possible events are:

- Capture on every falling edge
- Capture on every rising edge
- Capture every 4th rising edge
- Capture every 16th rising edge

When a capture takes place, an interrupt flag is latched into the CAxIF bit. This interrupt can be enabled by setting the corresponding mask bit CAxIE. The Peripheral Interrupt Enable bit (PEIE) must be set and the Global Interrupt Disable bit (GLINTD) must be cleared for the interrupt to be acknowledged. The CAxIF interrupt flag bit is cleared in software.

When the capture prescale select is changed, the prescaler is not reset and an event may be generated. Therefore, the first capture after such a change will be ambiguous. However, it sets the time-base for the next capture. The prescaler is reset upon chip RESET.

The capture pin, CAPx, is a multiplexed pin. When used as a port pin, the capture is not disabled. However, the user can simply disable the Capture interrupt by clearing CAxIE. If the CAPx pin is used as an output pin, the user can activate a capture by writing to the port pin. This may be useful during development phase to emulate a capture interrupt.

The input on the capture pin CAPx is synchronized internally to internal phase clocks. This imposes certain restrictions on the input waveform (see the Electrical Specification section for timing).

The capture overflow status flag bit is double buffered. The master bit is set if one captured word is already residing in the Capture register (CAxH:CAxL) and another “event” has occurred on the CAPx pin. The new event will not transfer the TMR3 value to the capture register, protecting the previous unread capture value. When the user reads both the high and the low bytes (in any order) of the Capture register, the master overflow bit is transferred to the slave overflow bit (CAxOVF) and then the master bit is reset. The user can then read TCONx to determine the value of CAxOVF.

The recommended sequence to read capture registers and capture overflow flag bits is shown in Example 13-1.

FIGURE 14-1: USART TRANSMIT

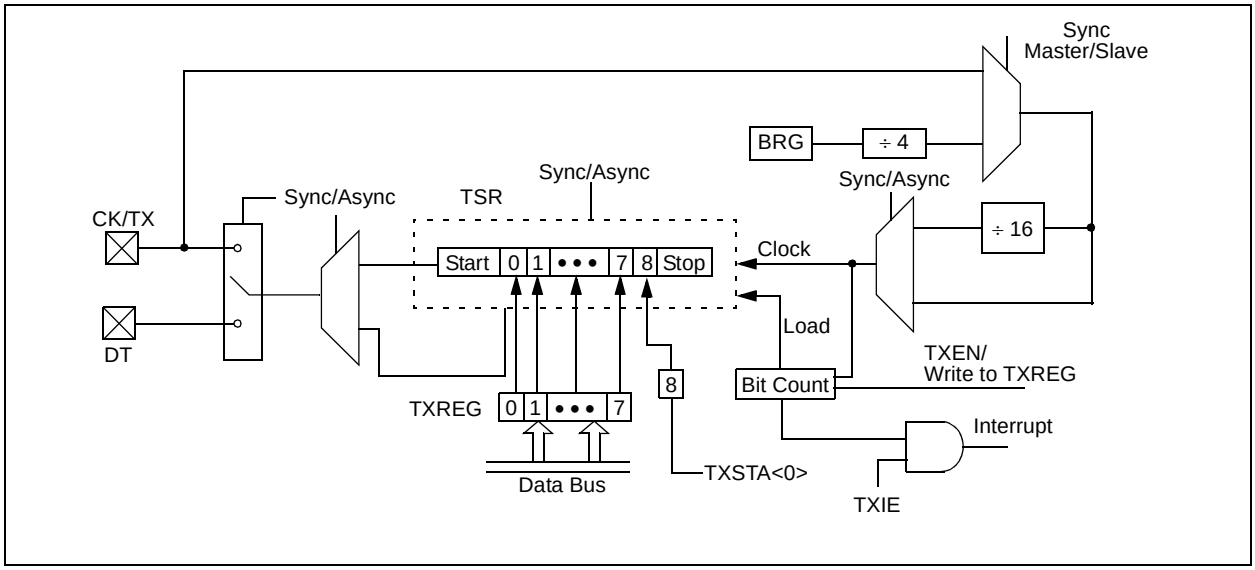


FIGURE 14-2: USART RECEIVE

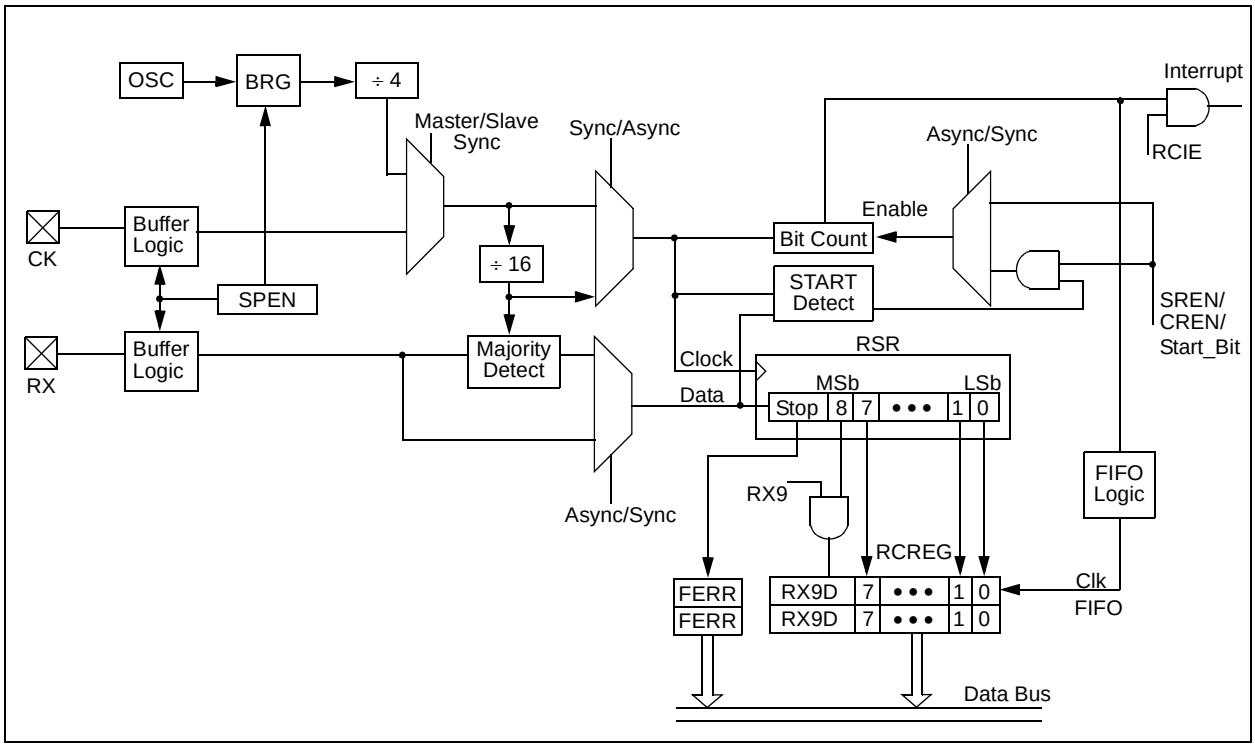


TABLE 14-9: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER RECEPTION

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	$\overline{\text{MCLR}}$, WDT
16h, Bank 1	PIR1	RBIF	TMR3IF	TMR2IF	TMR1IF	CA2IF	CA1IF	TX1IF	RC1IF	x000 0010	u000 0010
17h, Bank 1	PIE1	RBIE	TMR3IE	TMR2IE	TMR1IE	CA2IE	CA1IE	TX1IE	RC1IE	0000 0000	0000 0000
13h, Bank 0	RCSTA1	SPEN	RX9	SREN	CREN	—	FERR	OERR	RX9D	0000 -00x	0000 -00u
14h, Bank 0	RCREG1	RX7	RX6	RX5	RX4	RX3	RX2	RX1	RX0	xxxx xxxx	uuuu uuuu
15h, Bank 0	TXSTA1	CSRC	TX9	TXEN	SYNC	—	—	TRMT	TX9D	0000 --1x	0000 --1u
17h, Bank 0	SPBRG1	Baud Rate Generator Register								0000 0000	0000 0000
10h, Bank 4	PIR2	SSPIF	BCLIF	ADIF	—	CA4IF	CA3IF	TX2IF	RC2IF	000- 0010	000- 0010
11h, Bank 4	PIE2	SSPIE	BCLIE	ADIE	—	CA4IE	CA3IE	TX2IE	RC2IE	000- 0000	000- 0000
13h, Bank 4	RCSTA2	SPEN	RX9	SREN	CREN	—	FERR	OERR	RX9D	0000 -00x	0000 -00u
14h, Bank 4	RCREG2	RX7	RX6	RX5	RX4	RX3	RX2	RX1	RX0	xxxx xxxx	uuuu uuuu
15h, Bank 4	TXSTA2	CSRC	TX9	TXEN	SYNC	—	—	TRMT	TX9D	0000 --1x	0000 --1u
17h, Bank 4	SPBRG2	Baud Rate Generator Register								0000 0000	0000 0000

Legend: x = unknown, u = unchanged, - = unimplemented, read as a '0'. Shaded cells are not used for synchronous master reception.

When the application software is expecting to receive valid data, the SSPBUF should be read before the next byte of data to transfer is written to the SSPBUF. Buffer full bit, BF (SSPSTAT<0>), indicates when SSPBUF has been loaded with the received data (transmission is complete). When the SSPBUF is read, bit BF is cleared. This data may be irrelevant if the SPI is only a transmitter. Generally the MSSP interrupt is used to determine when the transmission/reception has completed. The SSPBUF must be read and/or written. If the interrupt method is not going to be used, then software polling can be done to ensure that a write collision does not occur. Example 15-1 shows the loading of the SSPBUF (SSPSR) for data transmission.

EXAMPLE 15-1: LOADING THE SSPBUF (SSPSR) REGISTER

```
     MOVLB 6                ; Bank 6
LOOP BTFSS SSPSTAT, BF      ; Has data been
                            ;   received
                            ;   (transmit
                            ;   complete)?
     GOTO  LOOP             ; No
     MOVPF SSPBUF, RXDATA   ; Save in user RAM
     MOVFP TXDATA, SSPBUF   ; New data to xmit
```

The SSPSR is not directly readable, or writable and can only be accessed by addressing the SSPBUF register. Additionally, the MSSP status register (SSPSTAT) indicates the various status conditions.

15.1.2 ENABLING SPI I/O

To enable the serial port, MSSP Enable bit, SSPEN (SSPCON1<5>), must be set. To reset or reconfigure SPI mode, clear bit SSPEN, re-initialize the SSPCON registers and then set bit SSPEN. This configures the SDI, SDO, SCK and $\overline{SS}$ pins as serial port pins. For the pins to behave as the serial port function, some must have their data direction bits (in the DDR register) appropriately programmed. That is:

- SDI is automatically controlled by the SPI module
- SDO must have DDRB<7> cleared
- SCK (Master mode) must have DDRB<6> cleared
- SCK (Slave mode) must have DDRB<6> set
- $\overline{SS}$ must have PORTA<2> set

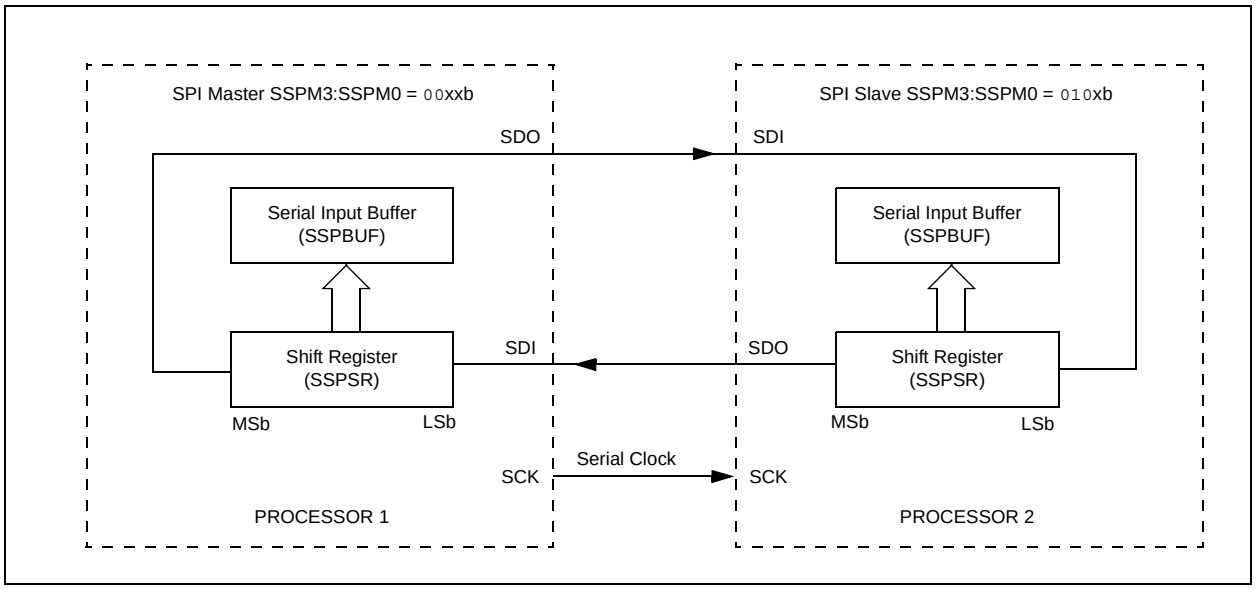
Any serial port function that is not desired may be overridden by programming the corresponding data direction (DDR) register to the opposite value.

15.1.3 TYPICAL CONNECTION

Figure 15-5 shows a typical connection between two microcontrollers. The master controller (Processor 1) initiates the data transfer by sending the SCK signal. Data is shifted out of both shift registers on their programmed clock edge and latched on the opposite edge of the clock. Both processors should be programmed to same Clock Polarity (CKP), then both controllers would send and receive data at the same time. Whether the data is meaningful (or dummy data) depends on the application software. This leads to three scenarios for data transmission:

- Master sends data — Slave sends dummy data
- Master sends data — Slave sends data
- Master sends dummy data — Slave sends data

FIGURE 15-5: SPI MASTER/SLAVE CONNECTION



15.2.1.1 Addressing

Once the MSSP module has been enabled, it waits for a START condition to occur. Following the START condition, the 8-bits are shifted into the SSPSR register. All incoming bits are sampled with the rising edge of the clock (SCL) line. The value of register SSPSR<7:1> is compared to the value of the SSPADD register. The address is compared on the falling edge of the eighth clock (SCL) pulse. If the addresses match and the BF and SSPOV bits are clear, the following events occur:

a) The SSPSR register value is loaded into the SSPBUF register on the falling edge of the 8th SCL pulse.
b) The buffer full bit, BF, is set on the falling edge of the 8th SCL pulse.
c) An $\overline{ACK}$ pulse is generated.
d) SSP interrupt flag bit, SSPIF (PIR2<7>), is set (interrupt is generated if enabled) - on the falling edge of the 9th SCL pulse.

In 10-bit address mode, two address bytes need to be received by the slave. The five Most Significant bits (MSbs) of the first address byte specify if this is a 10-bit address. Bit R/$\overline{W}$ (SSPSTAT<2>) must specify a write so the slave device will receive the second address byte. For a 10-bit address, the first byte would equal '1111 0 A9 A8 0', where A9 and A8 are the two MSbs of the address. The sequence of events for a 10-bit address is as follows, with steps 7- 9 for slave-transmitter:

1. Receive first (high) byte of Address (bits SSPIF, BF and bit UA (SSPSTAT<1>) are set).
2. Update the SSPADD register with second (low) byte of Address (clears bit UA and releases the SCL line).
3. Read the SSPBUF register (clears bit BF) and clear flag bit SSPIF.
4. Receive second (low) byte of Address (bits SSPIF, BF and UA are set).
5. Update the SSPADD register with the first (high) byte of Address. This will clear bit UA and release the SCL line.
6. Read the SSPBUF register (clears bit BF) and clear flag bit SSPIF.
7. Receive Repeated Start condition.
8. Receive first (high) byte of Address (bits SSPIF and BF are set).
9. Read the SSPBUF register (clears bit BF) and clear flag bit SSPIF.

Note:	Following the Repeated Start condition (step 7) in 10-bit mode, the user only needs to match the first 7-bit address. The user does not update the SSPADD for the second half of the address.

15.2.1.2 Slave Reception

When the R/$\overline{W}$ bit of the address byte is clear and an address match occurs, the R/$\overline{W}$ bit of the SSPSTAT register is cleared. The received address is loaded into the SSPBUF register.

When the address byte overflow condition exists, then no acknowledge ($\overline{ACK}$) pulse is given. An overflow condition is defined as either bit BF (SSPSTAT<0>) is set, or bit SSPOV (SSPCON1<6>) is set.

An SSP interrupt is generated for each data transfer byte. Flag bit SSPIF (PIR2<7>) must be cleared in software. The SSPSTAT register is used to determine the status of the received byte.

Note:	The SSPBUF will be loaded if the SSPOV bit is set and the BF flag is cleared. If a read of the SSPBUF was performed, but the user did not clear the state of the SSPOV bit before the next receive occurred, the $\overline{ACK}$ is not sent and the SSPBUF is updated.

TABLE 15-2: DATA TRANSFER RECEIVED BYTE ACTIONS

Status Bits as Data Transfer is Received		SSPSR → SSPBUF	Generate $\overline{ACK}$ Pulse	Set bit SSPIF (SSP Interrupt occurs if enabled)
BF	SSPOV			
0	0	Yes	Yes	Yes
1	0	No	No	Yes
1	1	No	No	Yes
0	1	Yes	No	Yes

Note 1: Shaded cells show the conditions where the user software did not properly clear the overflow condition.

15.2.10 I²C MASTER MODE REPEATED START CONDITION TIMING

A Repeated Start condition occurs when the RSEN bit (SSPCON2<1>) is programmed high and the I²C module is in the idle state. When the RSEN bit is set, the SCL pin is asserted low. When the SCL pin is sampled low, the baud rate generator is loaded with the contents of SSPADD<6:0> and begins counting. The SDA pin is released (brought high) for one baud rate generator count (TBRG). When the baud rate generator times out, if SDA is sampled high, the SCL pin will be de-asserted (brought high). When SCL is sampled high the baud rate generator is reloaded with the contents of SSPADD<6:0> and begins counting. SDA and SCL must be sampled high for one TBRG. This action is then followed by assertion of the SDA pin (SDA is low) for one TBRG while SCL is high. Following this, the RSEN bit in the SSPCON2 register will be automatically cleared and the baud rate generator is not reloaded, leaving the SDA pin held low. As soon as a START condition is detected on the SDA and SCL pins, the S bit (SSPSTAT<3>) will be set. The SSPIF bit will not be set until the baud rate generator has timed out.

Note 1: If the RSEN is programmed while any other event is in progress, it will not take effect.

2: A bus collision during the Repeated Start condition occurs if:

- SDA is sampled low when SCL goes from low to high.
- SCL goes low before SDA is asserted low. This may indicate that another master is attempting to transmit a data "1".

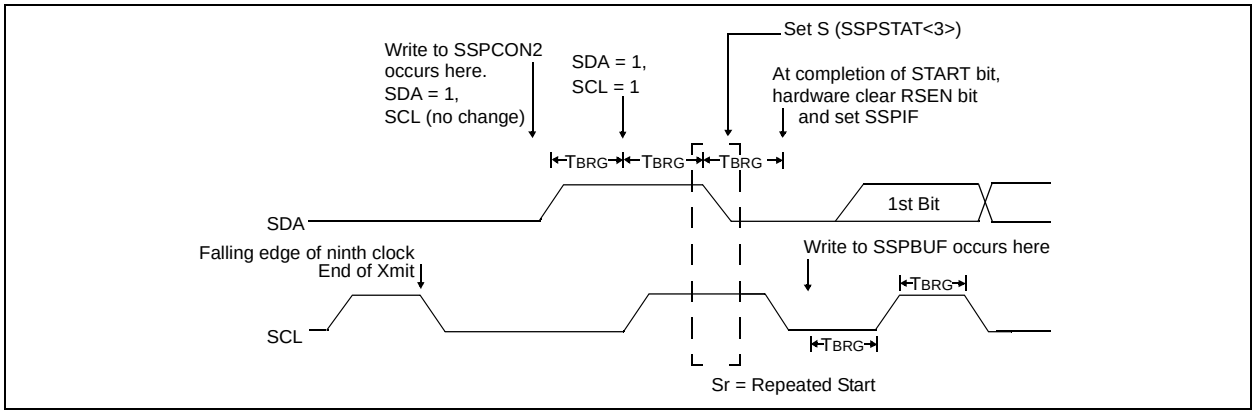
Immediately following the SSPIF bit getting set, the user may write the SSPBUF with the 7-bit address in 7-bit mode, or the default first address in 10-bit mode. After the first eight bits are transmitted and an ACK is received, the user may then transmit an additional eight bits of address (10-bit mode), or eight bits of data (7-bit mode).

15.2.10.1 WCOL status flag

If the user writes the SSPBUF when a Repeated Start sequence is in progress, then WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

Note: Because queueing of events is not allowed, writing of the lower 5 bits of SSPCON2 is disabled until the Repeated Start condition is complete.

FIGURE 15-22: REPEAT START CONDITION WAVEFORM



15.2.15 CLOCK ARBITRATION

Clock arbitration occurs when the master, during any receive, transmit, or Repeated Start/Stop condition, deasserts the SCL pin (SCL allowed to float high). When the SCL pin is allowed to float high, the baud rate generator (BRG) is suspended from counting until the SCL pin is actually sampled high. When the SCL pin is sampled high, the baud rate generator is reloaded with the contents of SSPADD<6:0> and begins counting. This ensures that the SCL high time will always be at least one BRG rollover count, in the event that the clock is held low by an external device (Figure 15-33).

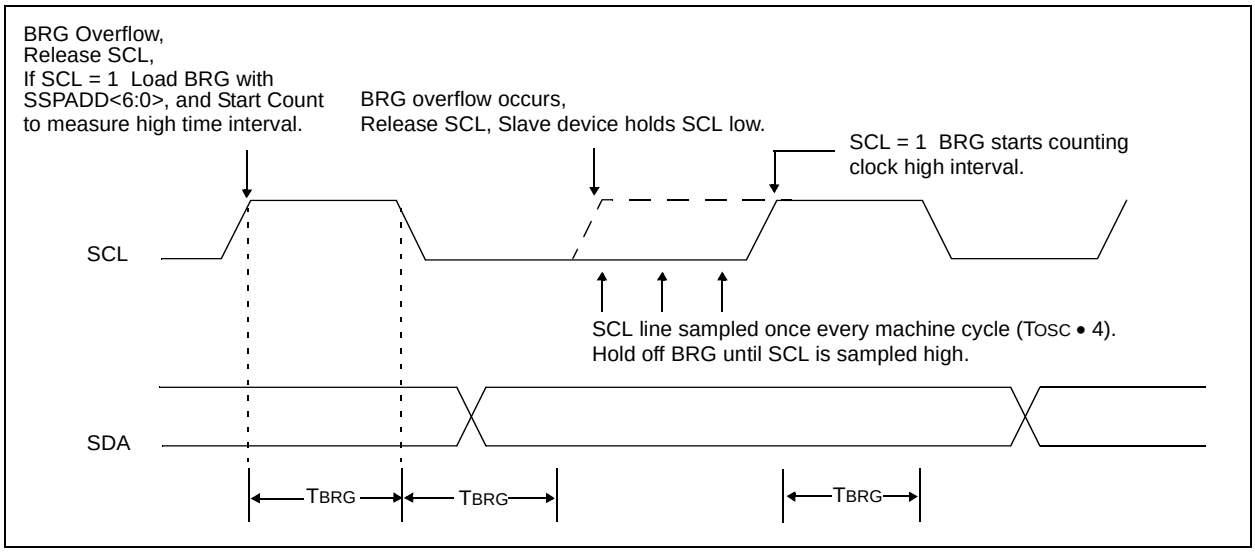
15.2.16 SLEEP OPERATION

While in SLEEP mode, the I^2C module can receive addresses or data and when an address match or complete byte transfer occurs, wake the processor from SLEEP (if the SSP interrupt is enabled).

15.2.17 EFFECTS OF A RESET

A RESET disables the SSP module and terminates the current transfer.

FIGURE 15-33: CLOCK ARBITRATION TIMING IN MASTER TRANSMIT MODE



15.3 Connection Considerations for I²C Bus

For standard mode I²C bus devices, the values of resistors R_p R_s in Figure 15-42 depends on the following parameters:

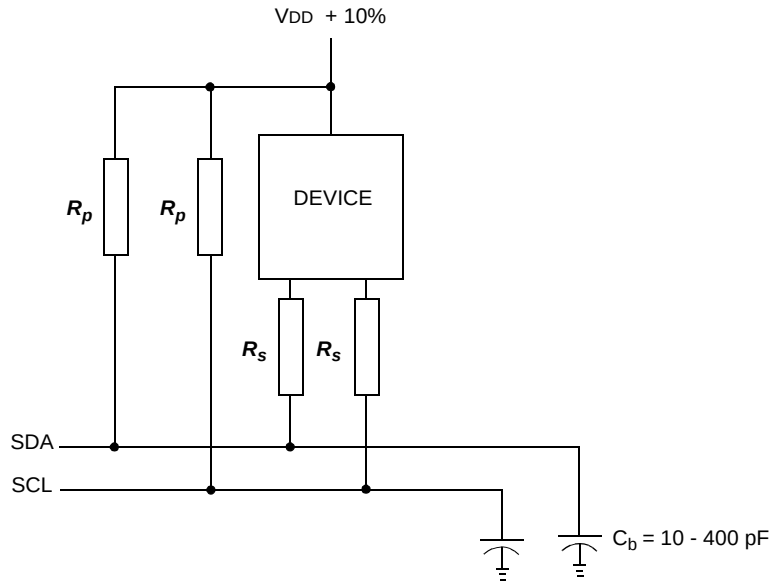
- Supply voltage
- Bus capacitance
- Number of connected devices (input current + leakage current)

The supply voltage limits the minimum value of resistor R_p due to the specified minimum sink current of 3 mA at VOL max = 0.4V for the specified output stages. For example, with a supply voltage of VDD = 5V ±10% and VOL max = 0.4V at 3 mA, R_p min = (5.5-0.4)/0.003 = 1.7 kΩ. VDD as a function of R_p is shown in Figure 15-42. The desired noise margin of 0.1 VDD for the low level, limits the maximum value of R_s. Series resistors are optional and used to improve ESD susceptibility.

The bus capacitance is the total capacitance of wire, connections and pins. This capacitance limits the maximum value of R_p due to the specified rise time (Figure 15-42).

The SMP bit is the slew rate control enabled bit. This bit is in the SSPSTAT register and controls the slew rate of the I/O pins when in I²C mode (master or slave).

FIGURE 15-42: SAMPLE DEVICE CONFIGURATION FOR I²C BUS



Note: I²C devices with input levels related to VDD must have one common supply line to which the pull-up resistor is also connected.

CLRWDT **Clear Watchdog Timer**

Syntax: [*label*] CLRWDT

Operands: None

Operation: 00h → WDT
0 → $\overline{\text{WDT}}$ postscaler,
1 → $\overline{\text{TO}}$
1 → $\overline{\text{PD}}$

Status Affected: $\overline{\text{TO}}$, $\overline{\text{PD}}$

Encoding:

0000	0000	0000	0100

Description: CLRWDT instruction resets the Watchdog Timer. It also resets the postscaler of the WDT. Status bits $\overline{\text{TO}}$ and $\overline{\text{PD}}$ are set.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	No operation	Process Data	No operation

Example: CLRWDT

Before Instruction
WDT counter = ?

After Instruction
WDT counter = 0x00
$\overline{\text{WDT}}$ Postscaler = 0
$\overline{\text{TO}}$ = 1
$\overline{\text{PD}}$ = 1

COMF **Complement f**

Syntax: [*label*] COMF f,d

Operands: $0 \le f \le 255$
$d \in [0,1]$

Operation: $(\bar{f}) \rightarrow$ (dest)

Status Affected: Z

Encoding:

0001	001d	ffff	ffff

Description: The contents of register 'f' are complemented. If 'd' is 0 the result is stored in WREG. If 'd' is 1 the result is stored back in register 'f'.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

Example: COMF REG1,0

Before Instruction
REG1 = 0x13

After Instruction
REG1 = 0x13
WREG = 0xEC

DECF **Decrement f**

Syntax: [*label*] DECF f,d

Operands: $0 \le f \le 255$
$d \in [0,1]$

Operation: (f) – 1 → (dest)

Status Affected: OV, C, DC, Z

Encoding:

0000	011d	ffff	ffff

Description: Decrement register 'f'. If 'd' is 0, the result is stored in WREG. If 'd' is 1, the result is stored back in register 'f'.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

Example: `DECF CNT, 1`

Before Instruction
CNT = 0x01
Z = 0

After Instruction
CNT = 0x00
Z = 1

DECFSZ **Decrement f, skip if 0**

Syntax: [*label*] DECFSZ f,d

Operands: $0 \le f \le 255$
$d \in [0,1]$

Operation: (f) – 1 → (dest);
skip if result = 0

Status Affected: None

Encoding:

0001	011d	ffff	ffff

Description: The contents of register 'f' are decremented. If 'd' is 0, the result is placed in WREG. If 'd' is 1, the result is placed back in register 'f'.

If the result is 0, the next instruction, which is already fetched is discarded and a `NOP` is executed instead, making it a two-cycle instruction.

Words: 1

Cycles: 1(2)

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

If skip:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation

Example:
```
HERE    DECFSZ  CNT, 1
        GOTO    HERE
NZERO
ZERO
```

Before Instruction
PC = Address `(HERE)`

After Instruction
CNT = CNT - 1
If CNT = 0;
PC = Address `(HERE)`
If CNT ≠ 0;
PC = Address `(NZERO)`

MOVFP **Move f to p**

Syntax: [*label*] MOVFP f,p

Operands: $0 \le f \le 255$
$0 \le p \le 31$

Operation: (f) → (p)

Status Affected: None

Encoding:

011p	pppp	ffff	ffff

Description: Move data from data memory location 'f' to data memory location 'p'. Location 'f' can be anywhere in the 256 byte data space (00h to FFh), while 'p' can be 00h to 1Fh.

Either 'p' or 'f' can be WREG (a useful, special situation).

MOVFP is particularly useful for transferring a data memory location to a peripheral register (such as the transmit buffer or an I/O port). Both 'f' and 'p' can be indirectly addressed.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write register 'p'

Example: MOVFP REG1, REG2

Before Instruction
REG1 = 0x33,
REG2 = 0x11

After Instruction
REG1 = 0x33,
REG2 = 0x33

MOVLB **Move Literal to low nibble in BSR**

Syntax: [*label*] MOVLB k

Operands: $0 \le k \le 15$

Operation: k → (BSR<3:0>)

Status Affected: None

Encoding:

1011	1000	uuuu	kkkk

Description: The four-bit literal 'k' is loaded in the Bank Select Register (BSR). Only the low 4-bits of the Bank Select Register are affected. The upper half of the BSR is unchanged. The assembler will encode the “u” fields as '0'.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'k'	Process Data	Write literal 'k' to BSR<3:0>

Example: MOVLB 5

Before Instruction
BSR register = 0x22

After Instruction
BSR register = 0x25 (Bank 5)

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