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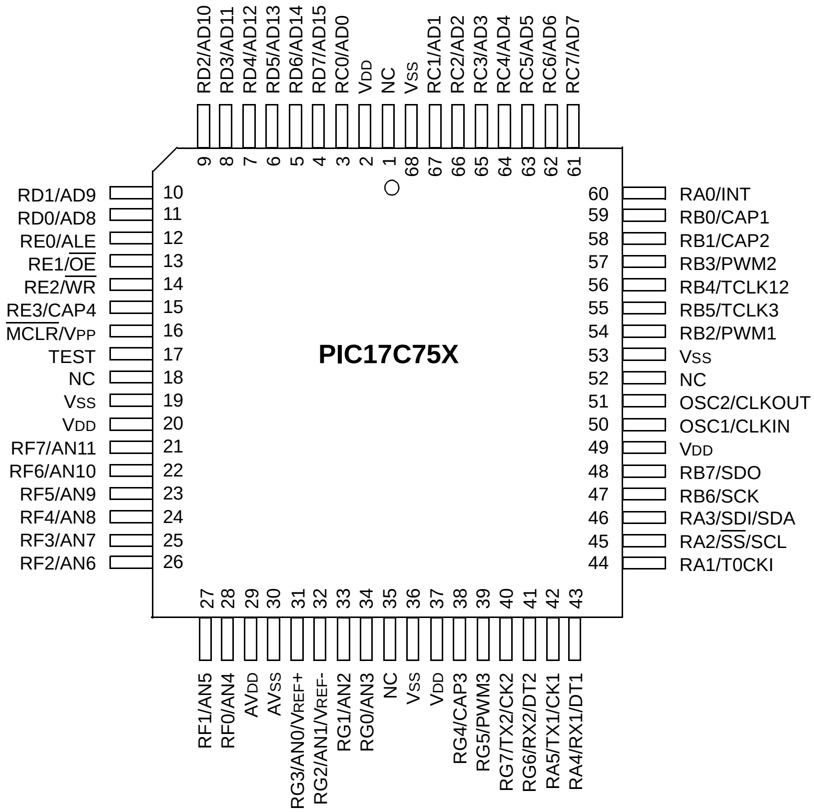
Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	33MHz
Connectivity	I²C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	66
Program Memory Size	32KB (16K x 16)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	902 x 8
Voltage - Supply (Vcc/Vdd)	4.5V ~ 5.5V
Data Converters	A/D 16x10b
Oscillator Type	External
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	84-LCC (J-Lead)
Supplier Device Package	84-PLCC (29.31x29.31)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic17c766-33-l

Pin Diagrams cont.'d

68-Pin PLCC



64-Pin TQFP

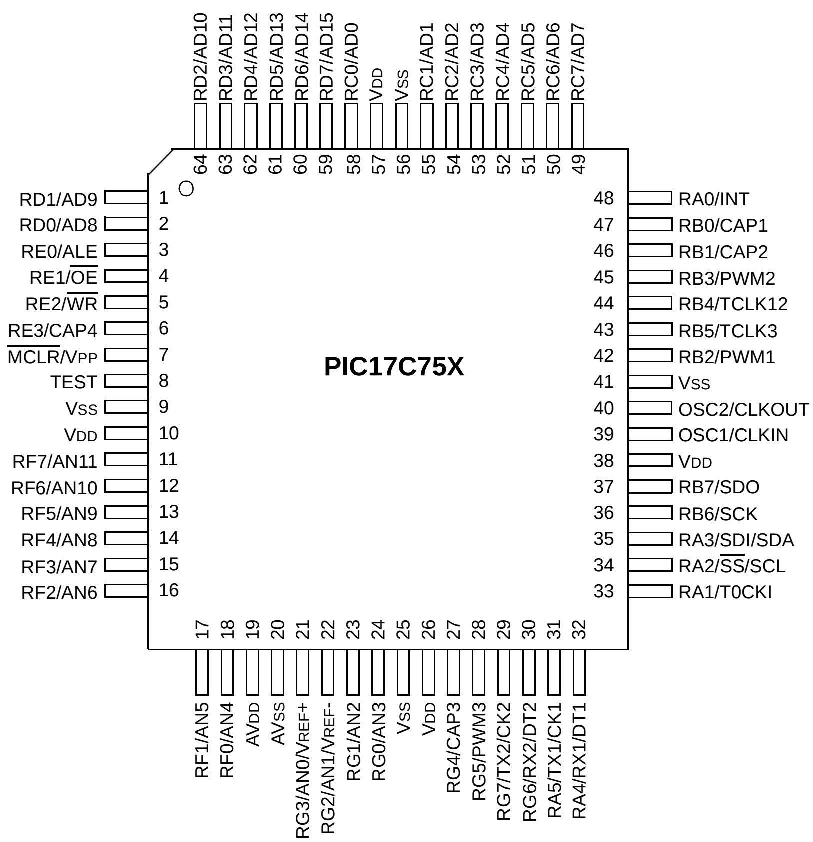
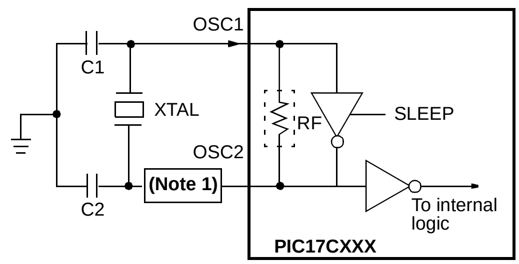


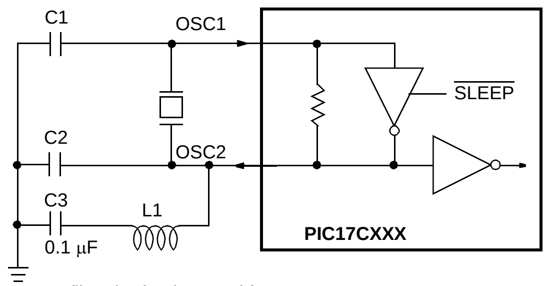
FIGURE 4-2: CRYSTAL OR CERAMIC RESONATOR OPERATION (XT OR LF OSC CONFIGURATION)



See Table 4-1 and Table 4-2 for recommended values of C1 and C2.

Note 1: A series resistor (Rs) may be required for AT strip cut crystals.

FIGURE 4-3: CRYSTAL OPERATION, OVERTONE CRYSTALS (XT OSC CONFIGURATION)



To filter the fundamental frequency:

$$\frac{1}{L1*C2} = (2\pi f)^2$$

Where f = tank circuit resonant frequency. This should be midway between the fundamental and the 3rd overtone frequencies of the crystal.
C3 blocks DC current to ground.

TABLE 4-1: CAPACITOR SELECTION FOR CERAMIC RESONATORS

Oscillator Type	Resonator Frequency	Capacitor Range C1 = C2$^{(1)}$
LF	455 kHz 2.0 MHz	15 - 68 pF 10 - 33 pF
XT	4.0 MHz 8.0 MHz 16.0 MHz	22 - 68 pF 33 - 100 pF 33 - 100 pF

Higher capacitance increases the stability of the oscillator, but also increases the start-up time. These values are for design guidance only. Since each resonator has its own characteristics, the user should consult the resonator manufacturer for appropriate values of external components.

Note 1: These values include all board capacitances on this pin. Actual capacitor value depends on board capacitance.

Resonators Used:

455 kHz	Panasonic EFO-A455K04B	± 0.3%
2.0 MHz	Murata Erie CSA2.00MG	± 0.5%
4.0 MHz	Murata Erie CSA4.00MG	± 0.5%
8.0 MHz	Murata Erie CSA8.00MT	± 0.5%
16.0 MHz	Murata Erie CSA16.00MX	± 0.5%

Resonators used did not have built-in capacitors.

TABLE 4-2: CAPACITOR SELECTION FOR CRYSTAL OSCILLATOR

Osc Type	Freq	C1$^{(2)}$	C2$^{(2)}$
LF	32 kHz 1 MHz 2 MHz	100-150 pF 10-68 pF 10-68 pF	100-150 pF 10-68 pF 10-68 pF
XT	2 MHz 4 MHz 8 MHz 16 MHz 24 MHz$^{(1)}$ 32 MHz$^{(1)}$	47-100 pF 15-68 pF 15-47 pF 15-47 pF 15-47 pF 10-47 pF	47-100 pF 15-68 pF 15-47 pF 15-47 pF 15-47 pF 10-47 pF

Higher capacitance increases the stability of the oscillator, but also increases the start-up time and the oscillator current. These values are for design guidance only. Rs may be required in XT mode to avoid overdriving the crystals with low drive level specification. Since each crystal has its own characteristics, the user should consult the crystal manufacturer for appropriate values for external components.

Note 1: Overtone crystals are used at 24 MHz and higher. The circuit in Figure 4-3 should be used to select the desired harmonic frequency.

2: These values include all board capacitances on this pin. Actual capacitor value depends on board capacitance.

Crystals Used:

32.768 kHz	Epson C-001R32.768K-A	± 20 PPM
1.0 MHz	ECS-10-13-1	± 50 PPM
2.0 MHz	ECS-20-20-1	± 50 PPM
4.0 MHz	ECS-40-20-1	± 50 PPM
8.0 MHz	ECS ECS-80-S-4 ECS-80-18-1	± 50 PPM
16.0 MHz	ECS-160-20-1	± 50 PPM
25 MHz	CTS CTS25M	± 50 PPM
32 MHz	CRYSTEK HF-2	± 50 PPM

4.1.6 RC OSCILLATOR

For timing insensitive applications, the RC device option offers additional cost savings. RC oscillator frequency is a function of the supply voltage, the resistor (REXT) and capacitor (CEXT) values, and the operating temperature. In addition to this, oscillator frequency will vary from unit to unit due to normal process parameter variation. Furthermore, the difference in lead frame capacitance between package types will also affect oscillation frequency, especially for low CEXT values. The user also needs to take into account variation due to tolerance of external R and C components used. Figure 4-7 shows how the R/C combination is connected to the PIC17CXXX. For REXT values below 2.2 kΩ, the oscillator operation may become unstable, or stop completely. For very high REXT values (e.g. 1 MΩ), the oscillator becomes sensitive to noise, humidity and leakage. Thus, we recommend to keep REXT between 3 kΩ and 100 kΩ.

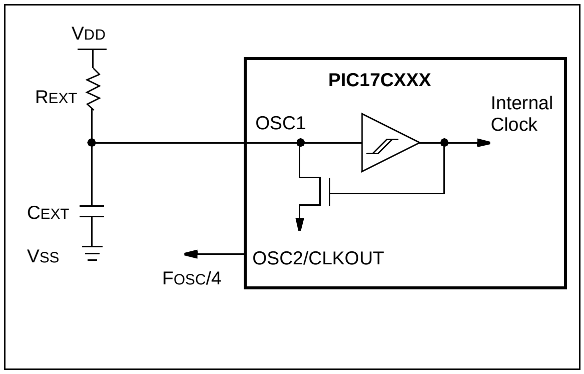
Although the oscillator will operate with no external capacitor (CEXT = 0 pF), we recommend using values above 20 pF for noise and stability reasons. With little or no external capacitance, oscillation frequency can vary dramatically due to changes in external capacitances, such as PCB trace capacitance or package lead frame capacitance.

See Section 21.0 for RC frequency variation from part to part due to normal process variation. The variation is larger for larger R (since leakage current variation will affect RC frequency more for large R) and for smaller C (since variation of input capacitance will affect RC frequency more).

See Section 21.0 for variation of oscillator frequency due to VDD for given REXT/CEXT values, as well as frequency variation due to operating temperature for given R, C, and VDD values.

The oscillator frequency, divided by 4, is available on the OSC2/CLKOUT pin and can be used for test purposes or to synchronize other logic (see Figure 4-8 for waveform).

FIGURE 4-7: RC OSCILLATOR MODE



4.1.6.1 RC Start-up

As the device voltage increases, the RC will immediately start its oscillations once the pin voltage levels meet the input threshold specifications (parameter #D032 and parameter #D042 in the electrical specification section). The time required for the RC to start oscillating depends on many factors. These include:

- Resistor value used
- Capacitor value used
- Device VDD rise time
- System temperature

5.1 Power-on Reset (POR), Power-up Timer (PWRT), Oscillator Start-up Timer (OST) and Brown-out Reset (BOR)

5.1.1 POWER-ON RESET (POR)

The Power-on Reset circuit holds the device in RESET until V_{DD} is above the trip point (in the range of 1.4V - 2.3V). The devices produce an internal RESET for both rising and falling V_{DD}. To take advantage of the POR, just tie the $\overline{MCLR}/V_{PP}$ pin directly (or through a resistor) to V_{DD}. This will eliminate external RC components usually needed to create Power-on Reset. A minimum rise time for V_{DD} is required. See Electrical Specifications for details.

Figure 5-2 and Figure 5-3 show two possible POR circuits.

FIGURE 5-2: USING ON-CHIP POR

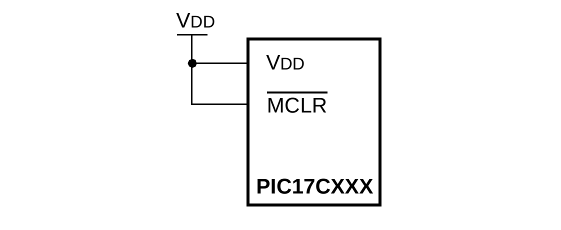
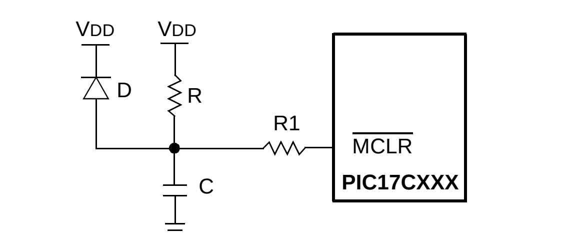


FIGURE 5-3: EXTERNAL POWER-ON RESET CIRCUIT (FOR SLOW V_{DD} POWER-UP)



Note 1: An external Power-on Reset circuit is required only if V_{DD} power-up time is too slow. The diode D helps discharge the capacitor quickly when V_{DD} powers down.

2: R < 40 kΩ is recommended to ensure that the voltage drop across R does not exceed 0.2V (max. leakage current spec. on the $\overline{MCLR}/V_{PP}$ pin is 5 μA). A larger voltage drop will degrade V_{IH} level on the $\overline{MCLR}/V_{PP}$ pin.

3: R1 = 100Ω to 1 kΩ will limit any current flowing into $\overline{MCLR}$ from external capacitor C in the event of $\overline{MCLR}/V_{PP}$ pin breakdown due to Electrostatic Discharge (ESD) or Electrical Overstress (EOS).

5.1.2 POWER-UP TIMER (PWRT)

The Power-up Timer provides a fixed 96 ms time-out (nominal) on power-up. This occurs from the rising edge of the internal POR signal if V_{DD} and $\overline{MCLR}$ are tied, or after the first rising edge of $\overline{MCLR}$ (detected high). The Power-up Timer operates on an internal RC oscillator. The chip is kept in RESET as long as the PWRT is active. In most cases, the PWRT delay allows V_{DD} to rise to an acceptable level.

The power-up time delay will vary from chip to chip and with V_{DD} and temperature. See DC parameters for details.

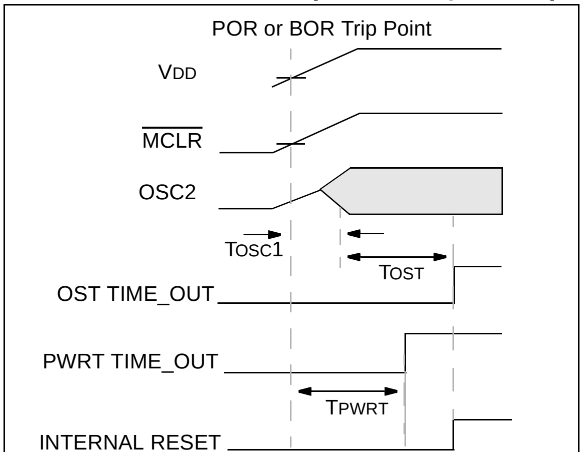
5.1.3 OSCILLATOR START-UP TIMER (OST)

The Oscillator Start-up Timer (OST) provides a 1024 oscillator cycle (1024T_{OSC}) delay whenever the PWRT is invoked, or a wake-up from SLEEP event occurs in XT or LF mode. The PWRT and OST operate in parallel.

The OST counts the oscillator pulses on the OSC1/CLKIN pin. The counter only starts incrementing after the amplitude of the signal reaches the oscillator input thresholds. This delay allows the crystal oscillator or resonator to stabilize before the device exits RESET. The length of the time-out is a function of the crystal/resonator frequency.

Figure 5-4 shows the operation of the OST circuit. In this figure, the oscillator is of such a low frequency that although enabled simultaneously, the OST does not time-out until after the Power-up Timer time-out.

FIGURE 5-4: OSCILLATOR START-UP TIME (LOW FREQUENCY)



This figure shows in greater detail the timings involved with the oscillator start-up timer. In this example, the low frequency crystal start-up time is larger than power-up time (T_{PWRT}).

T_{OSC1} = time for the crystal oscillator to react to an oscillation level detectable by the Oscillator Start-up Timer (OST).

T_{OST} = 1024T_{OSC}.

TABLE 5-4: INITIALIZATION CONDITIONS FOR SPECIAL FUNCTION REGISTERS (CONTINUED)

Register	Address	Power-on Reset Brown-out Reset	$\overline{\text{MCLR}}$ Reset WDT Reset	Wake-up from SLEEP through Interrupt
Bank 1				
DDRC(5)	10h	1111 1111	1111 1111	uuuu uuuu
PORTC(4,5)	11h	xxxx xxxx	uuuu uuuu	uuuu uuuu
DDRD(5)	12h	1111 1111	1111 1111	uuuu uuuu
PORTD(4,5)	13h	xxxx xxxx	uuuu uuuu	uuuu uuuu
DDRE(5)	14h	---- 1111	---- 1111	---- uuuu
PORTE(4,5)	15h	---- xxxx	---- uuuu	---- uuuu
PIR1	16h	x000 0010	u000 0010	uuuu uuuu(1)
PIE1	17h	0000 0000	0000 0000	uuuu uuuu
Bank 2				
TMR1	10h	xxxx xxxx	uuuu uuuu	uuuu uuuu
TMR2	11h	xxxx xxxx	uuuu uuuu	uuuu uuuu
TMR3L	12h	xxxx xxxx	uuuu uuuu	uuuu uuuu
TMR3H	13h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PR1	14h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PR2	15h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PR3/CA1L	16h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PR3/CA1H	17h	xxxx xxxx	uuuu uuuu	uuuu uuuu
Bank 3				
PW1DCL	10h	xx-- ----	uu-- ----	uu-- ----
PW2DCL	11h	xx0- ----	uu0- ----	uuu- ----
PW1DCH	12h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PW2DCH	13h	xxxx xxxx	uuuu uuuu	uuuu uuuu
CA2L	14h	xxxx xxxx	uuuu uuuu	uuuu uuuu
CA2H	15h	xxxx xxxx	uuuu uuuu	uuuu uuuu
TCON1	16h	0000 0000	0000 0000	uuuu uuuu
TCON2	17h	0000 0000	0000 0000	uuuu uuuu

Legend: u = unchanged, x = unknown, - = unimplemented, read as '0', q = value depends on condition

Note 1: One or more bits in INTSTA, PIR1, PIR2 will be affected (to cause wake-up).

2: When the wake-up is due to an interrupt and the GLINTD bit is cleared, the PC is loaded with the interrupt vector.

3: See Table 5-3 for RESET value of specific condition.

4: This is the value that will be in the port output latch.

5: When the device is configured for Microprocessor or Extended Microcontroller mode, the operation of this port does not rely on these registers.

6: On any device RESET, these pins are configured as inputs.

7.2.2.2 CPU Status Register (CPUSTA)

The CPUSTA register contains the status and control bits for the CPU. This register has a bit that is used to globally enable/disable interrupts. If only a specific interrupt is desired to be enabled/disabled, please refer to the Interrupt Status (INTSTA) register and the Peripheral Interrupt Enable (PIE) registers. The CPUSTA register also indicates if the stack is available and contains the Power-down ($\overline{\text{PD}}$) and Time-out ($\overline{\text{TO}}$) bits. The $\overline{\text{TO}}$, $\overline{\text{PD}}$, and STKAV bits are not writable. These bits are set and cleared according to device logic. Therefore, the result of an instruction with the CPUSTA register as destination may be different than intended.

The $\overline{\text{POR}}$ bit allows the differentiation between a Power-on Reset, external $\overline{\text{MCLR}}$ Reset, or a WDT Reset. The $\overline{\text{BOR}}$ bit indicates if a Brown-out Reset occurred.

> **Note 1:** The $\overline{\text{BOR}}$ status bit is a don't care and is not necessarily predictable if the Brown-out circuit is disabled (when the BODEN bit in the Configuration word is programmed).

REGISTER 7-2: CPUSTA REGISTER (ADDRESS: 06h, UNBANKED)

U-0	U-0	R-1	R/W-1	R-1	R-1	R/W-0	R/W-1
—	—	STKAV	GLINTD	$\overline{\text{TO}}$	$\overline{\text{PD}}$	$\overline{\text{POR}}$	$\overline{\text{BOR}}$
bit 7							bit 0

bit 7-6 **Unimplemented**: Read as '0'

bit 5 **STKAV**: Stack Available bit
This bit indicates that the 4-bit stack pointer value is Fh, or has rolled over from Fh → 0h (stack overflow).
`1` = Stack is available
`0` = Stack is full, or a stack overflow may have occurred (once this bit has been cleared by a stack overflow, only a device RESET will set this bit)

bit 4 **GLINTD**: Global Interrupt Disable bit
This bit disables all interrupts. When enabling interrupts, only the sources with their enable bits set can cause an interrupt.
`1` = Disable all interrupts
`0` = Enables all unmasked interrupts

bit 3 **$\overline{\text{TO}}$**: WDT Time-out Status bit
`1` = After power-up, by a `CLRWDT` instruction, or by a `SLEEP` instruction
`0` = A Watchdog Timer time-out occurred

bit 2 **$\overline{\text{PD}}$**: Power-down Status bit
`1` = After power-up or by the `CLRWDT` instruction
`0` = By execution of the `SLEEP` instruction

bit 1 **$\overline{\text{POR}}$**: Power-on Reset Status bit
`1` = No Power-on Reset occurred
`0` = A Power-on Reset occurred (must be set by software)

bit 0 **$\overline{\text{BOR}}$**: Brown-out Reset Status bit

When BODEN Configuration bit is set (enabled):
`1` = No Brown-out Reset occurred
`0` = A Brown-out Reset occurred (must be set by software)

When BODEN Configuration bit is clear (disabled):
Don't care

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
- n = Value at POR Reset	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

REGISTER 13-2: TCON2 REGISTER (ADDRESS: 17h, BANK 3)

R-0	R-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
CA2OVF	CA1OVF	PWM2ON	PWM1ON	CA1/$\overline{PR3}$	TMR3ON	TMR2ON	TMR1ON
bit 7							bit 0

bit 7 **CA2OVF**: Capture2 Overflow Status bit
This bit indicates that the capture value had not been read from the capture register pair (CA2H:CA2L) before the next capture event occurred. The capture register retains the oldest unread capture value (last capture before overflow). Subsequent capture events will not update the capture register with the TMR3 value until the capture register has been read (both bytes).
1 = Overflow occurred on Capture2 register
0 = No overflow occurred on Capture2 register

bit 6 **CA1OVF**: Capture1 Overflow Status bit
This bit indicates that the capture value had not been read from the capture register pair (PR3H/CA1H:PR3L/CA1L), before the next capture event occurred. The capture register retains the oldest unread capture value (last capture before overflow). Subsequent capture events will not update the capture register with the TMR3 value until the capture register has been read (both bytes).
1 = Overflow occurred on Capture1 register
0 = No overflow occurred on Capture1 register

bit 5 **PWM2ON**: PWM2 On bit
1 = PWM2 is enabled
(The RB3/PWM2 pin ignores the state of the DDRB<3> bit.)
0 = PWM2 is disabled
(The RB3/PWM2 pin uses the state of the DDRB<3> bit for data direction.)

bit 4 **PWM1ON**: PWM1 On bit
1 = PWM1 is enabled
(The RB2/PWM1 pin ignores the state of the DDRB<2> bit.)
0 = PWM1 is disabled
(The RB2/PWM1 pin uses the state of the DDRB<2> bit for data direction.)

bit 3 **CA1/$\overline{PR3}$**: CA1/$\overline{PR3}$ Register Mode Select bit
1 = Enables Capture1
(PR3H/CA1H:PR3L/CA1L is the Capture1 register. Timer3 runs without a period register.)
0 = Enables the Period register
(PR3H/CA1H:PR3L/CA1L is the Period register for Timer3.)

bit 2 **TMR3ON**: Timer3 On bit
1 = Starts Timer3
0 = Stops Timer3

bit 1 **TMR2ON**: Timer2 On bit
This bit controls the incrementing of the TMR2 register. When TMR2:TMR1 form the 16-bit timer (T16 is set), TMR2ON must be set. This allows the MSB of the timer to increment.
1 = Starts Timer2 (must be enabled if the T16 bit (TCON1<3>) is set)
0 = Stops Timer2

bit 0 **TMR1ON**: Timer1 On bit

When T16 is set (in 16-bit Timer mode):
1 = Starts 16-bit TMR2:TMR1
0 = Stops 16-bit TMR2:TMR1

When T16 is clear (in 8-bit Timer mode:
1 = Starts 8-bit Timer1
0 = Stops 8-bit Timer1

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
- n = Value at POR Reset	’1’ = Bit is set	’0’ = Bit is cleared	x = Bit is unknown

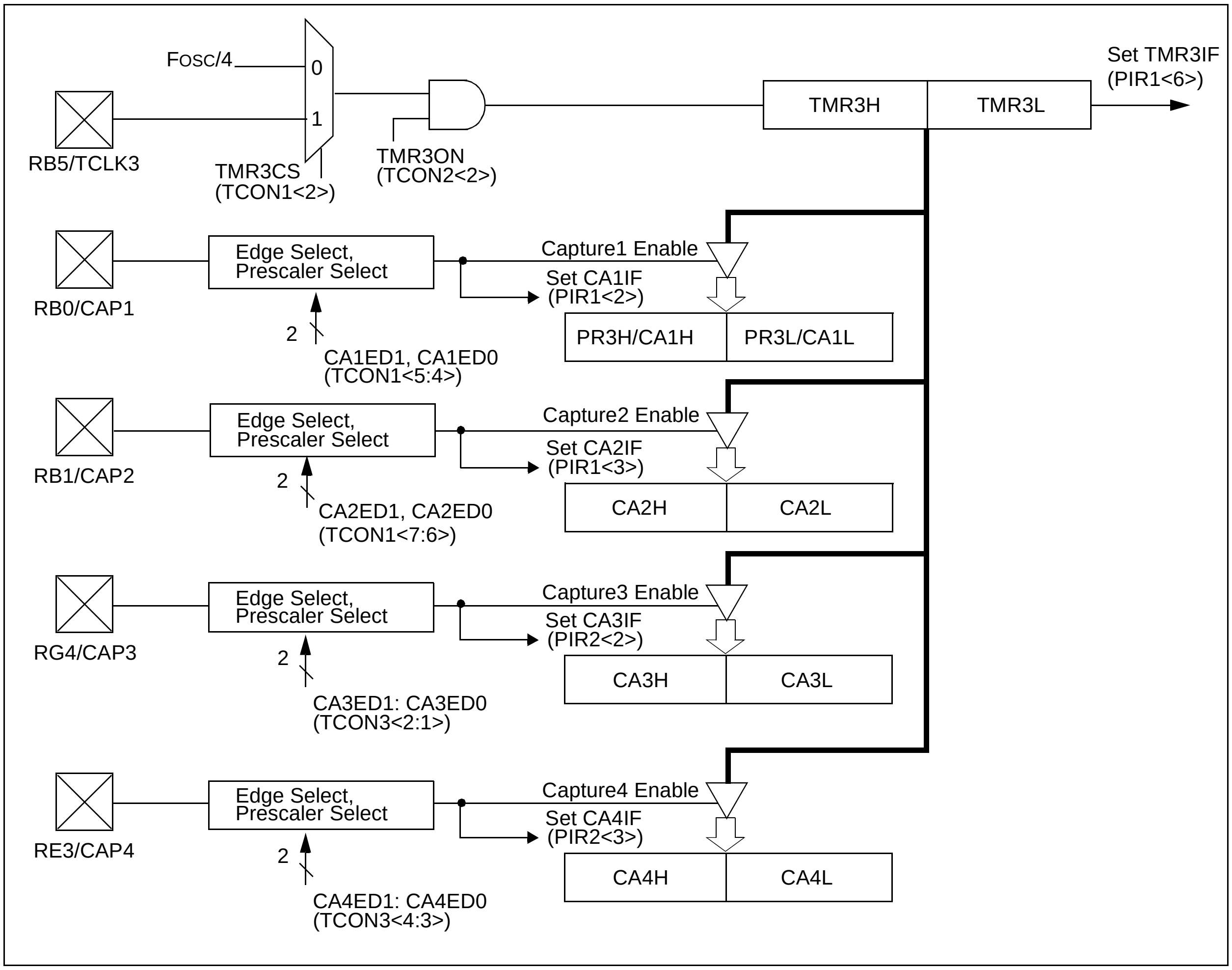
13.2.2 FOUR CAPTURE MODE

This mode is selected by setting bit CA1/$\overline{PR3}$. A block diagram is shown in Figure 13-6. In this mode, TMR3 runs without a period register and increments from 0000h to FFFFh and rolls over to 0000h. The TMR3 interrupt Flag (TMR3IF) is set on this rollover. The TMR3IF bit must be cleared in software.

Registers PR3H/CA1H and PR3L/CA1L make a 16-bit capture register (Capture1). It captures events on pin RB0/CAP1. Capture mode is configured by the CA1ED1 and CA1ED0 bits. Capture1 Interrupt Flag bit (CA1IF) is set upon detection of the capture event. The corresponding interrupt mask bit is CA1IE. The Capture1 Overflow Status bit is CA1OVF.

All the captures operate in the same manner. Refer to Section 13.2.1 for the operation of capture.

FIGURE 13-6: TIMER3 WITH FOUR CAPTURES BLOCK DIAGRAM



14.1 USART Baud Rate Generator (BRG)

The BRG supports both the Asynchronous and Synchronous modes of the USART. It is a dedicated 8-bit baud rate generator. The SPBRG register controls the period of a free running 8-bit timer. Table 14-2 shows the formula for computation of the baud rate for different USART modes. These only apply when the USART is in Synchronous Master mode (internal clock) and Asynchronous mode.

Given the desired baud rate and Fosc, the nearest integer value between 0 and 255 can be calculated using the formula below. The error in baud rate can then be determined.

TABLE 14-2: BAUD RATE FORMULA

SYNC	Mode	Baud Rate
0	Asynchronous	FOSC/(64(X+1))
1	Synchronous	FOSC/(4(X+1))

X = value in SPBRG (0 to 255)

Example 14-1 shows the calculation of the baud rate error for the following conditions:

FOSC = 16 MHz
Desired Baud Rate = 9600
SYNC = 0

EXAMPLE 14-1: CALCULATING BAUD RATE ERROR

Desired Baud Rate = FOSC / (64 (X + 1))

9600 = 16000000 /(64 (X + 1))

X = 25.042 $\rightarrow$ 25

Calculated Baud Rate = 16000000 / (64 (25 + 1))

= 9615

Error = $\frac{\text{(Calculated Baud Rate - Desired Baud Rate)}}{\text{Desired Baud Rate}}$

= (9615 - 9600) / 9600

= 0.16%

Writing a new value to the SPBRG, causes the BRG timer to be reset (or cleared). This ensures that the BRG does not wait for a timer overflow before outputting the new baud rate.

Effects of Reset

After any device RESET, the SPBRG register is cleared. The SPBRG register will need to be loaded with the desired value after each RESET.

TABLE 14-3: REGISTERS ASSOCIATED WITH BAUD RATE GENERATOR

	Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on $\overline{\text{MCLR}}$, WDT
USART1	13h, Bank 0	RCSTA1	SPEN	RX9	SREN	CREN	—	FERR	OERR	RX9D	0000 -00x	0000 -00u
	15h, Bank 0	TXSTA1	CSRC	TX9	TXEN	SYNC	—	—	TRMT	TX9D	0000 --1x	0000 --1u
	17h, Bank 0	SPBRG1	Baud Rate Generator Register								0000 0000	0000 0000
USART2	13h, Bank 4	RCSTA2	SPEN	RX9	SREN	CREN	—	FERR	OERR	RX9D	0000 -00x	0000 -00u
	15h, Bank 4	TXSTA2	CSRC	TX9	TXEN	SYNC	—	—	TRMT	TX9D	0000 --1x	0000 --1u
	17h, Bank 4	SPBRG2	Baud Rate Generator Register								0000 0000	0000 0000

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by the Baud Rate Generator.

15.1.4 MASTER MODE

The master can initiate the data transfer at any time because it controls the SCK. The master determines when the slave (Processor 2, Figure 15-5) is to broadcast data by the software protocol.

In Master mode, the data is transmitted/received as soon as the SSPBUF register is written to. If the SPI is only going to receive, the SDO output could be disabled (programmed as an input). The SSPSR register will continue to shift in the signal present on the SDI pin at the programmed clock rate. As each byte is received, it will be loaded into the SSPBUF register as if a normal received byte (interrupts and status bits appropriately set). This could be useful in receiver applications as a “Line Activity Monitor” mode.

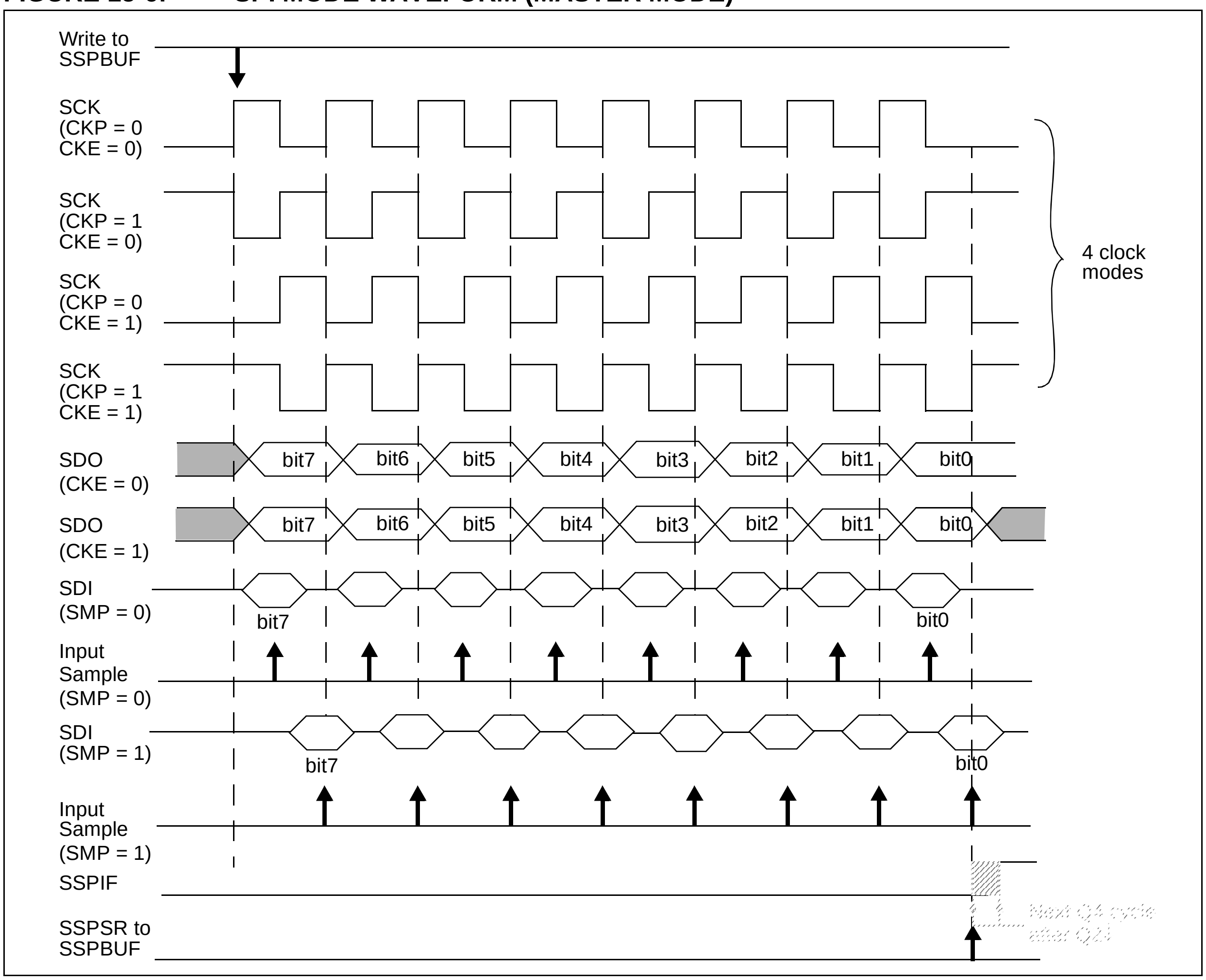
The clock polarity is selected by appropriately programming bit CKP (SSPCON1<4>). This then, would give waveforms for SPI communication as shown in Figure 15-6, Figure 15-8 and Figure 15-9, where the MSb is transmitted first. In Master mode, the SPI clock rate (bit rate) is user programmable to be one of the following:

- FOSC/4 (or TCY)
- FOSC/16 (or 4 • TCY)
- FOSC/64 (or 16 • TCY)
- Timer2 output/2

This allows a maximum bit clock frequency (at 33 MHz) of 8.25 MHz.

Figure 15-6 shows the waveforms for Master mode. When CKE = 1, the SDO data is valid before there is a clock edge on SCK. The change of the input sample is shown based on the state of the SMP bit. The time when the SSPBUF is loaded with the received data is shown.

FIGURE 15-6: SPI MODE WAVEFORM (MASTER MODE)



15.1.5 SLAVE MODE

In Slave mode, the data is transmitted and received as the external clock pulses appear on SCK. When the last bit is latched, the interrupt flag bit SSPIF (PIR2<7>) is set.

While in Slave mode, the external clock is supplied by the external clock source on the SCK pin. This external clock must meet the minimum high and low times as specified in the electrical specifications.

While in SLEEP mode, the slave can transmit/receive data. When a byte is received, the device will wake-up from SLEEP.

15.1.6 SLAVE SELECT SYNCHRONIZATION

The $\overline{SS}$ pin allows a Synchronous Slave mode. The SPI must be in Slave mode with $\overline{SS}$ pin control enabled (SSPCON1<3:0> = 04h). The pin must not be driven low for the $\overline{SS}$ pin to function as an input. The RA2 Data Latch must be high. When the $\overline{SS}$ pin is low, transmission and reception are enabled and the SDO pin is driven. When the $\overline{SS}$ pin goes high, the SDO pin is no longer driven, even if in the middle of a transmitted byte and becomes a floating output. External pull-up/pull-down resistors may be desirable, depending on the application.

> **Note 1:** When the SPI is in Slave mode with $\overline{SS}$ pin control enabled (SSPCON<3:0> = `0100`), the SPI module will reset if the $\overline{SS}$ pin is set to V_{DD}.
>
> **2:** If the SPI is used in Slave mode with CKE = '1', then the $\overline{SS}$ pin control must be enabled.

When the SPI module resets, the bit counter is forced to 0. This can be done by either forcing the $\overline{SS}$ pin to a high level, or clearing the SSPEN bit.

To emulate two-wire communication, the SDO pin can be connected to the SDI pin. When the SPI needs to operate as a receiver, the SDO pin can be configured as an input. This disables transmissions from the SDO. The SDI can always be left as an input (SDI function), since it cannot create a bus conflict.

FIGURE 15-7: SLAVE SYNCHRONIZATION WAVEFORM

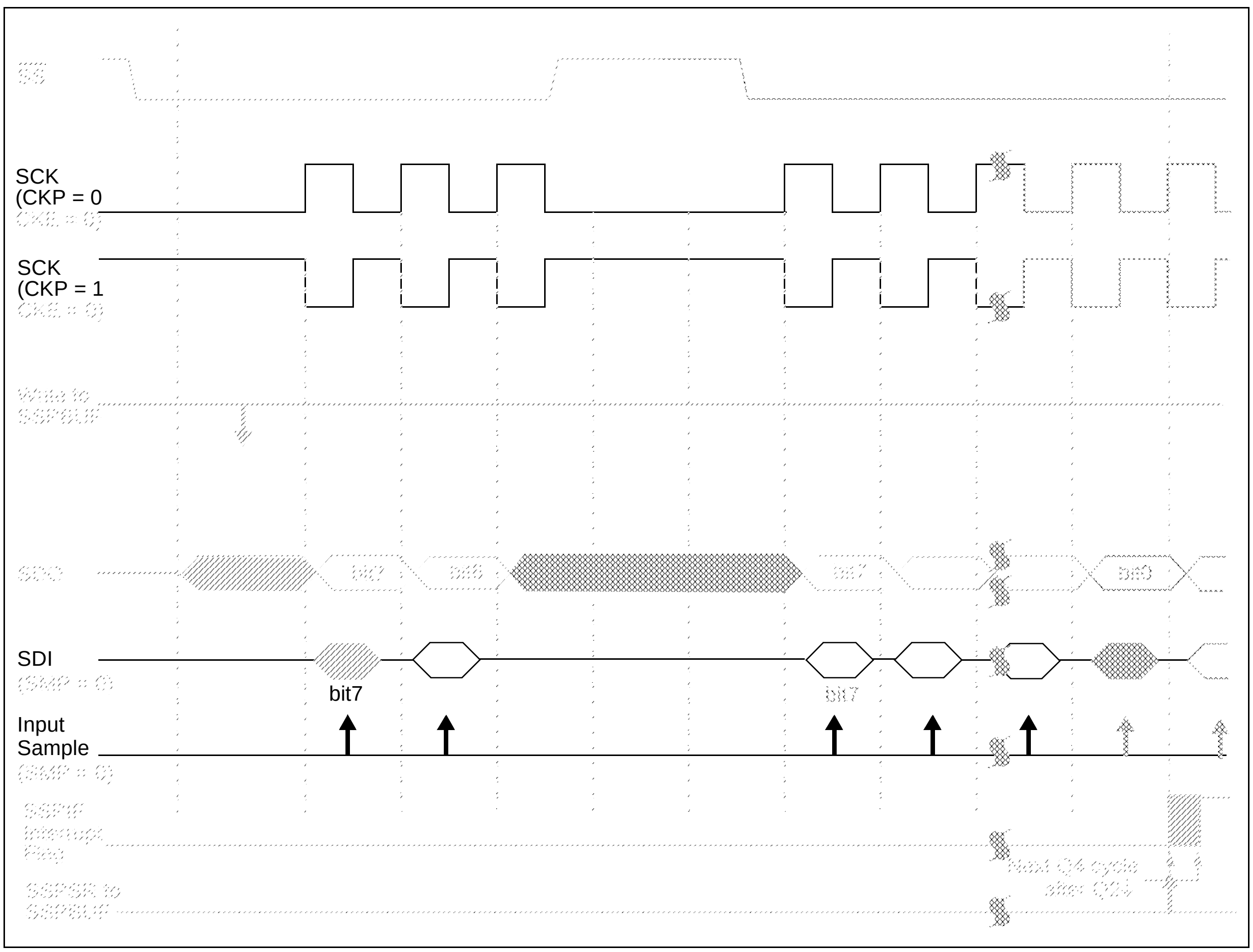


FIGURE 15-8: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 0)

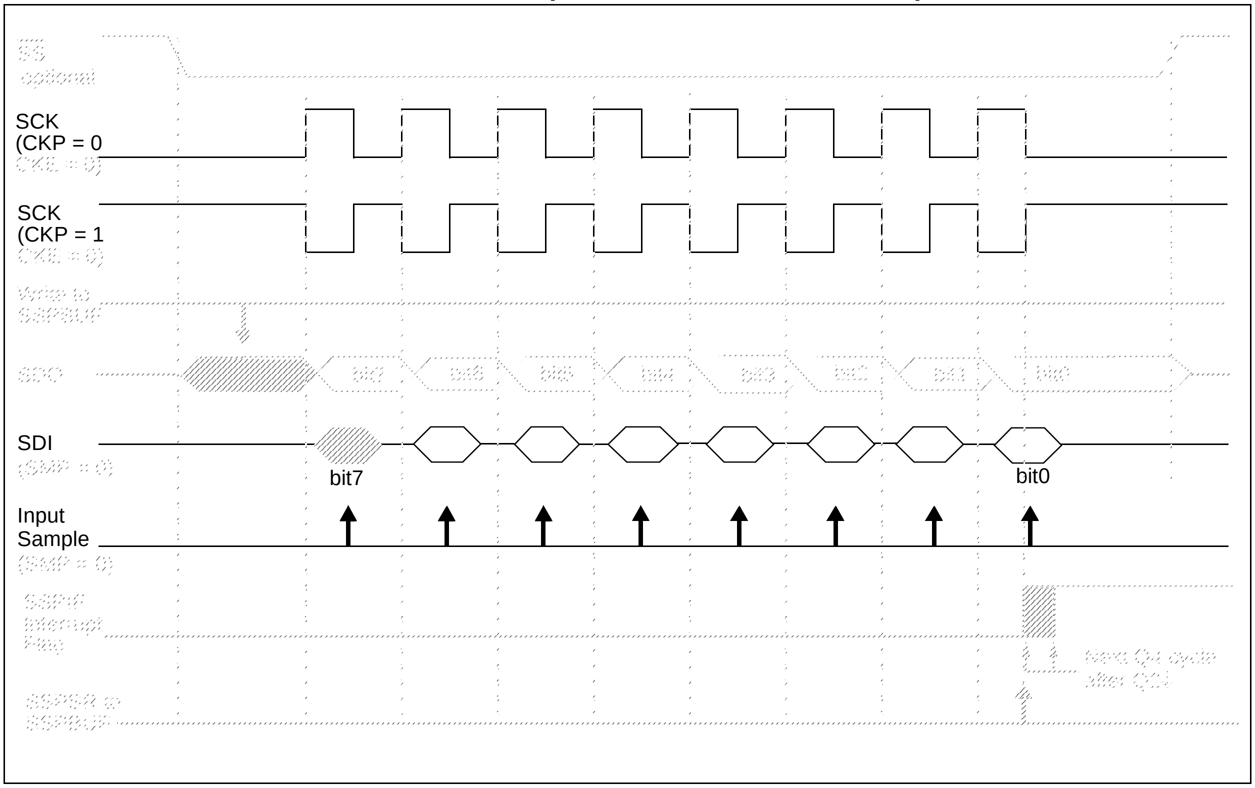
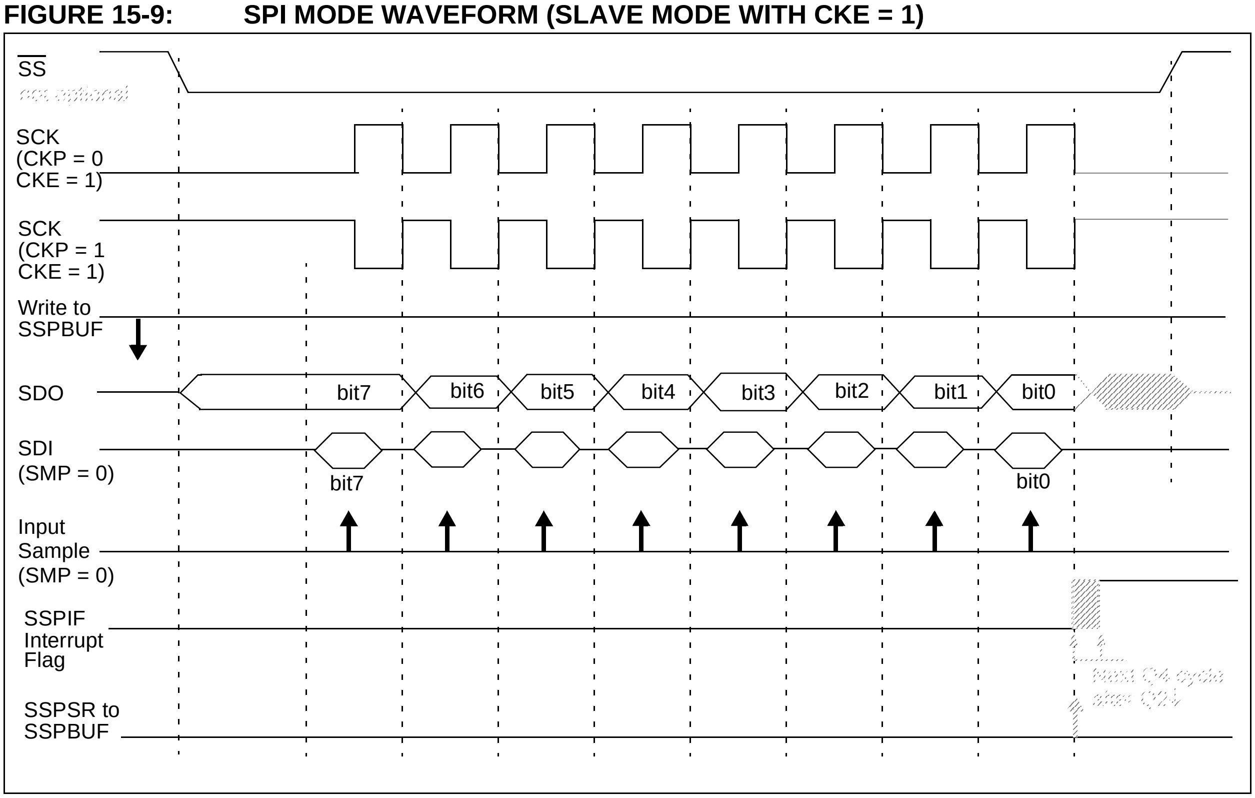


FIGURE 15-9: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 1)



15.2.11 I²C MASTER MODE TRANSMISSION

Transmission of a data byte, a 7-bit address, or either half of a 10-bit address, is accomplished by simply writing a value to SSPBUF register. This action will set the buffer full flag (BF) and allow the baud rate generator to begin counting and start the next transmission. Each bit of address/data will be shifted out onto the SDA pin after the falling edge of SCL is asserted (see data hold time spec). SCL is held low for one baud rate generator roll over count (TBRG). Data should be valid before SCL is released high (see Data setup time spec). When the SCL pin is released high, it is held that way for TBRG, the data on the SDA pin must remain stable for that duration and some hold time after the next falling edge of SCL. After the eighth bit is shifted out (the falling edge of the eighth clock), the BF flag is cleared and the master releases SDA, allowing the slave device being addressed to respond with an $\overline{\text{ACK}}$ bit during the ninth bit time, if an address match occurs or if data was received properly. The status of $\overline{\text{ACK}}$ is read into the ACKDT on the falling edge of the ninth clock. If the master receives an acknowledge, the acknowledge status bit (AKSTAT) is cleared. If not, the bit is set. After the ninth clock, the SSPIF is set and the master clock (baud rate generator) is suspended until the next data byte is loaded into the SSPBUF, leaving SCL low and SDA unchanged (Figure 15-26).

After the write to the SSPBUF, each bit of address will be shifted out on the falling edge of SCL until all seven address bits and the R/$\overline{\text{W}}$ bit are completed. On the falling edge of the eighth clock, the master will de-assert the SDA pin, allowing the slave to respond with an acknowledge. On the falling edge of the ninth clock, the master will sample the SDA pin to see if the address was recognized by a slave. The status of the ACK bit is loaded into the ACKSTAT status bit (SSPCON2<6>). Following the falling edge of the ninth clock transmission of the address, the SSPIF is set, the BF flag is cleared and the baud rate generator is turned off until another write to the SSPBUF takes place, holding SCL low and allowing SDA to float.

15.2.11.1 BF Status Flag

In Transmit mode, the BF bit (SSPSTAT<0>) is set when the CPU writes to SSPBUF and is cleared when all 8 bits are shifted out.

15.2.11.2 WCOL Status Flag

If the user writes the SSPBUF when a transmit is already in progress (i.e., SSPSR is still shifting out a data byte), then WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

WCOL must be cleared in software.

15.2.11.3 AKSTAT Status Flag

In Transmit mode, the AKSTAT bit (SSPCON2<6>) is cleared when the slave has sent an acknowledge ($\overline{\text{ACK}}$ = 0) and is set when the slave does not acknowledge ($\overline{\text{ACK}}$ = 1). A slave sends an acknowledge when it has recognized its address (including a general call), or when the slave has properly received its data.

EXAMPLE 15-2: INTERFACING TO A 24LC01B SERIAL EEPROM (USING MPLAB C17)

```
void ACKPoll(void)
{
        StartI2C();                     // Send start bit
        IdleI2C();                      // Wait for idle condition
        WriteI2C(CONTROL);              // Send control byte
        IdleI2C();                      // Wait for idle condition
        // Poll the ACK bit coming from the 24LC01B
        // Loop as long as the 24LC01B NACKs
        while (SSPCON2bits.ACKSTAT)
        {
                RestartI2C();           // Send a restart bit
                IdleI2C();              // Wait for idle condition
                WriteI2C(CONTROL);      // Send control byte
                IdleI2C();              // Wait for idle condition
        }
        IdleI2C();                      // Wait for idle condition
        StopI2C();                      // Send stop bit
        IdleI2C();                      // Wait for idle condition
        return;
}
```


16.0 ANALOG-TO-DIGITAL CONVERTER (A/D) MODULE

The analog-to-digital (A/D) converter module has twelve analog inputs for the PIC17C75X devices and sixteen for the PIC17C76X devices.

The analog input charges a sample and hold capacitor. The output of the sample and hold capacitor is the input into the converter. The converter then generates a digital result of this analog level via successive approximation. This A/D conversion of the analog input signal, results in a corresponding 10-bit digital number.

The analog reference voltages (positive and negative supply) are software selectable to either the device's supply voltages (AVDD, AVss), or the voltage level on the RG3/AN0/VREF+ and RG2/AN1/VREF- pins.

The A/D converter has a unique feature of being able to operate while the device is in SLEEP mode. To operate in SLEEP, the A/D clock must be derived from the A/D's internal RC oscillator.

The A/D module has four registers. These registers are:

- A/D Result High Register (ADRESH)
- A/D Result Low Register (ADRESL)
- A/D Control Register0 (ADCON0)
- A/D Control Register1 (ADCON1)

The ADCON0 register, shown in Register 16-1, controls the operation of the A/D module. The ADCON1 register, shown in Register 16-2, configures the functions of the port pins. The port pins can be configured as analog inputs (RG3 and RG2 can also be the voltage references), or as digital I/O.

REGISTER 16-1: ADCON0 REGISTER (ADDRESS: 14h, BANK 5)

R/W-0	R/W-0	R/W-0	R/W-0	U-0	R/W-0	U-0	R/W-0
CHS3	CHS2	CHS1	CHS0	—	GO/$\overline{\text{DONE}}$	—	$\overline{\text{ADON}}$
bit 7							bit 0

bit 7-4 **CHS3:CHS0**: Analog Channel Select bits
`0000` = channel 0, (AN0)
`0001` = channel 1, (AN1)
`0010` = channel 2, (AN2)
`0011` = channel 3, (AN3)
`0100` = channel 4, (AN4)
`0101` = channel 5, (AN5)
`0110` = channel 6, (AN6)
`0111` = channel 7, (AN7)
`1000` = channel 8, (AN8)
`1001` = channel 9, (AN9)
`1010` = channel 10, (AN10)
`1011` = channel 11, (AN11)
`1100` = channel 12, (AN12) (PIC17C76X only)
`1101` = channel 13, (AN13) (PIC17C76X only)
`1110` = channel 14, (AN14) (PIC17C76X only)
`1111` = channel 15, (AN15) (PIC17C76X only)
`11xx` = **RESERVED**, do not select (PIC17C75X only)

bit 3 **Unimplemented**: Read as '0'

bit 2 **GO/$\overline{\text{DONE}}$:** A/D Conversion Status bit
If ADON = 1:
`1` = A/D conversion in progress (setting this bit starts the A/D conversion, which is automatically cleared by hardware when the A/D conversion is complete)
`0` = A/D conversion not in progress

bit 1 **Unimplemented**: Read as '0'

bit 0 **ADON**: A/D On bit
`1` = A/D converter module is operating
`0` = A/D converter module is shut-off and consumes no operating current

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
- n = Value at POR Reset	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

REGISTER 16-2: ADCON1 REGISTER (ADDRESS 15h, BANK 5)

R/W-0	R/W-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0
ADCS1	ADCS0	ADFM	—	PCFG3	PCFG2	PCFG1	PCFG0
bit 7							bit 0

bit 7-6 **ADCS1:ADCS0**: A/D Conversion Clock Select bits
00 = FOSC/8
01 = FOSC/32
10 = FOSC/64
11 = FRC (clock derived from an internal RC oscillator)

bit 5 **ADFM**: A/D Result Format Select
1 = Right justified. 6 Most Significant bits of ADRESH are read as ’0’.
0 = Left justified. 6 Least Significant bits of ADRESL are read as ’0’.

bit 4 **Unimplemented**: Read as '0'

bit 3-1 **PCFG3:PCFG1**: A/D Port Configuration Control bits

bit 0 **PCFG0**: A/D Voltage Reference Select bit
1 = A/D reference is the VREF+ and VREF- pins
0 = A/D reference is AVDD and AVSS

Note: When this bit is set, ensure that the A/D voltage reference specifications are met.

PCFG3:PCFG0	AN15	AN14	AN13	AN12	AN11	AN10	AN9	AN8	AN7	AN6	AN5	AN4	AN3	AN2	AN1	AN0
000x	A	A	A	A	A	A	A	A	A	A	A	A	A	A	A	A
001x	D	A	A	A	A	A	A	A	D	A	A	A	A	A	A	A
010x	D	D	A	A	A	A	A	A	D	D	A	A	A	A	A	A
011x	D	D	D	A	A	A	A	A	D	D	D	A	A	A	A	A
100x	D	D	D	D	A	A	A	A	D	D	D	D	A	A	A	A
101x	D	D	D	D	D	A	A	A	D	D	D	D	D	A	A	A
110x	D	D	D	D	D	D	A	A	D	D	D	D	D	D	A	A
111x	D	D	D	D	D	D	D	D	D	D	D	D	D	D	D	D

A = Analog input D = Digital I/O

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
- n = Value at POR Reset	’1’ = Bit is set	’0’ = Bit is cleared	x = Bit is unknown

ANDWF — **AND WREG with f**

Syntax:	[*label*] ANDWF f,d
Operands:	$0 \leq f \leq 255$ $d \in [0,1]$
Operation:	(WREG) .AND. (f) $\rightarrow$ (dest)
Status Affected:	Z
Encoding:	`0000` `101d` `ffff` `ffff`
Description:	The contents of WREG are AND'ed with register 'f'. If 'd' is 0 the result is stored in WREG. If 'd' is 1 the result is stored back in register 'f'.
Words:	1
Cycles:	1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

Example: `ANDWF REG, 1`

Before Instruction
WREG = 0x17
REG = 0xC2

After Instruction
WREG = 0x17
REG = 0x02

BCF — **Bit Clear f**

Syntax:	[*label*] BCF f,b
Operands:	$0 \leq f \leq 255$ $0 \leq b \leq 7$
Operation:	$0 \rightarrow$ (f<b>)
Status Affected:	None
Encoding:	`1000` `1bbb` `ffff` `ffff`
Description:	Bit 'b' in register 'f' is cleared.
Words:	1
Cycles:	1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write register 'f'

Example: `BCF FLAG_REG, 7`

Before Instruction
FLAG_REG = 0xC7

After Instruction
FLAG_REG = 0x47

RETURN — Return from Subroutine

Syntax:	[*label*] RETURN
Operands:	None
Operation:	TOS $\rightarrow$ PC;
Status Affected:	None
Encoding:	`0000 0000 0000 0010`
Description:	Return from subroutine. The stack is popped and the top of the stack (TOS) is loaded into the program counter.
Words:	1
Cycles:	2

Q Cycle Activity:

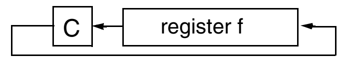
Q1	Q2	Q3	Q4
Decode	No operation	Process Data	POP PC from stack
No operation	No operation	No operation	No operation

Example: `RETURN`

After Interrupt
PC = TOS

RLCF — Rotate Left f through Carry

Syntax:	[*label*] RLCF f,d
Operands:	$0 \le f \le 255$ $d \in [0,1]$
Operation:	f<n> $\rightarrow$ d<n+1>; f<7> $\rightarrow$ C; C $\rightarrow$ d<0>
Status Affected:	C
Encoding:	`0001 101d ffff ffff`
Description:	The contents of register 'f' are rotated one bit to the left through the Carry Flag. If 'd' is 0, the result is placed in WREG. If 'd' is 1, the result is stored back in register 'f'.



Words:	1
Cycles:	1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

Example: `RLCF REG,0`

Before Instruction
REG = `1110 0110`
C = `0`

After Instruction
REG = `1110 0110`
WREG = `1100 1100`
C = `1`

RRNCF — Rotate Right f (no carry)

Syntax:	[*label*] RRNCF f,d
Operands:	$0 \le f \le 255$ $d \in [0,1]$
Operation:	f<n> $\rightarrow$ d<n-1>; f<0> $\rightarrow$ d<7>
Status Affected:	None
Encoding:	`0010` `000d` `ffff` `ffff`
Description:	The contents of register 'f' are rotated one bit to the right. If 'd' is 0, the result is placed in WREG. If 'd' is 1, the result is placed back in register 'f'.

Words:	1
Cycles:	1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

<u>Example 1</u>: `RRNCF REG, 1`

Before Instruction
- WREG = ?
- REG = `1101 0111`

After Instruction
- WREG = `0`
- REG = `1110 1011`

<u>Example 2</u>: `RRNCF REG, 0`

Before Instruction
- WREG = ?
- REG = `1101 0111`

After Instruction
- WREG = `1110 1011`
- REG = `1101 0111`

SETF — Set f

Syntax:	[*label*] SETF f,s
Operands:	$0 \le f \le 255$ $s \in [0,1]$
Operation:	FFh $\rightarrow$ f; FFh $\rightarrow$ d
Status Affected:	None
Encoding:	`0010` `101s` `ffff` `ffff`
Description:	If 's' is 0, both the data memory location 'f' and WREG are set to FFh. If 's' is 1, only the data memory location 'f' is set to FFh.
Words:	1
Cycles:	1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write register 'f' and other specified register

<u>Example1</u>: `SETF REG, 0`

Before Instruction
- REG = 0xDA
- WREG = 0x05

After Instruction
- REG = 0xFF
- WREG = 0xFF

<u>Example2</u>: `SETF REG, 1`

Before Instruction
- REG = 0xDA
- WREG = 0x05

After Instruction
- REG = 0xFF
- WREG = 0x05

20.3 Timing Parameter Symbology

The timing parameter symbols have been created following one of the following formats:

1. TppS2ppS		3. TCC:ST	(I^2C specifications only)
2. TppS		4. Ts	(I^2C specifications only)

T			
F	Frequency	T	Time

Lowercase symbols (pp) and their meanings:

pp			
ad	Address/Data	ost	Oscillator Start-Up Timer
al	ALE	pwrt	Power-Up Timer
cc	Capture1 and Capture2	rb	PORTB
ck	CLKOUT or clock	rd	$\overline{RD}$
dt	Data in	rw	$\overline{RD}$ or $\overline{WR}$
in	INT pin	t0	T0CKI
io	I/O port	t123	TCLK12 and TCLK3
mc	$\overline{MCLR}$	wdt	Watchdog Timer
oe	$\overline{OE}$	wr	$\overline{WR}$
os	OSC1		

Uppercase symbols and their meanings:

S			
D	Driven	L	Low
E	Edge	P	Period
F	Fall	R	Rise
H	High	V	Valid
I	Invalid (Hi-impedance)	Z	Hi-impedance

