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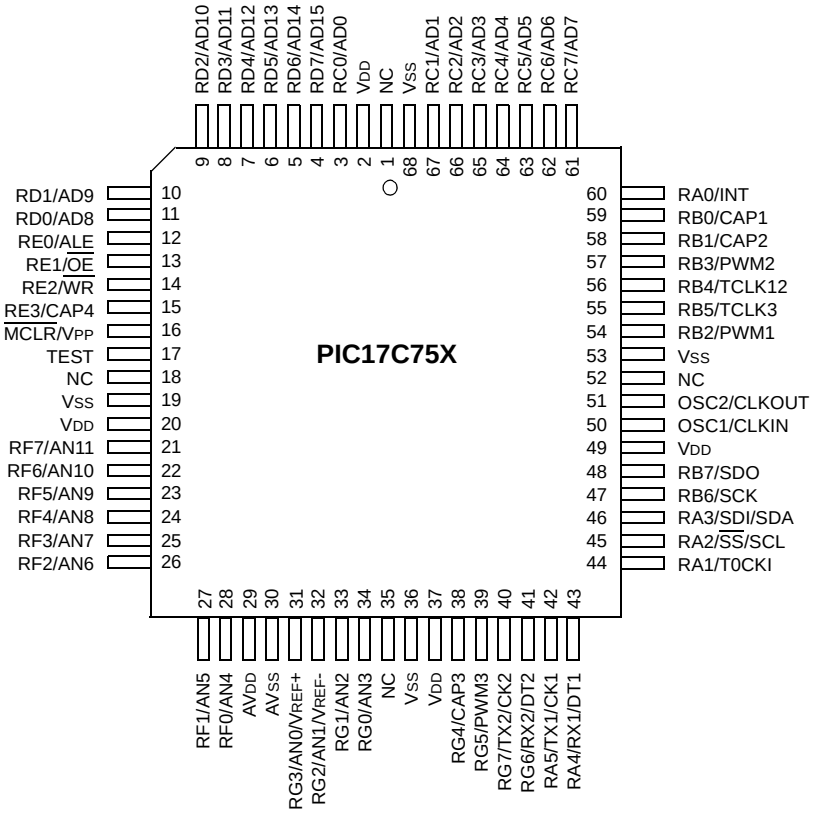
Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	PIC
Core Size	8-Bit
Speed	8MHz
Connectivity	I²C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	50
Program Memory Size	32KB (16K x 16)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	902 x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	A/D 12x10b
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	68-LCC (J-Lead)
Supplier Device Package	68-PLCC (24.23x24.23)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic17lc756at-08i-l

Pin Diagrams cont.'d

68-Pin PLCC



64-Pin TQFP

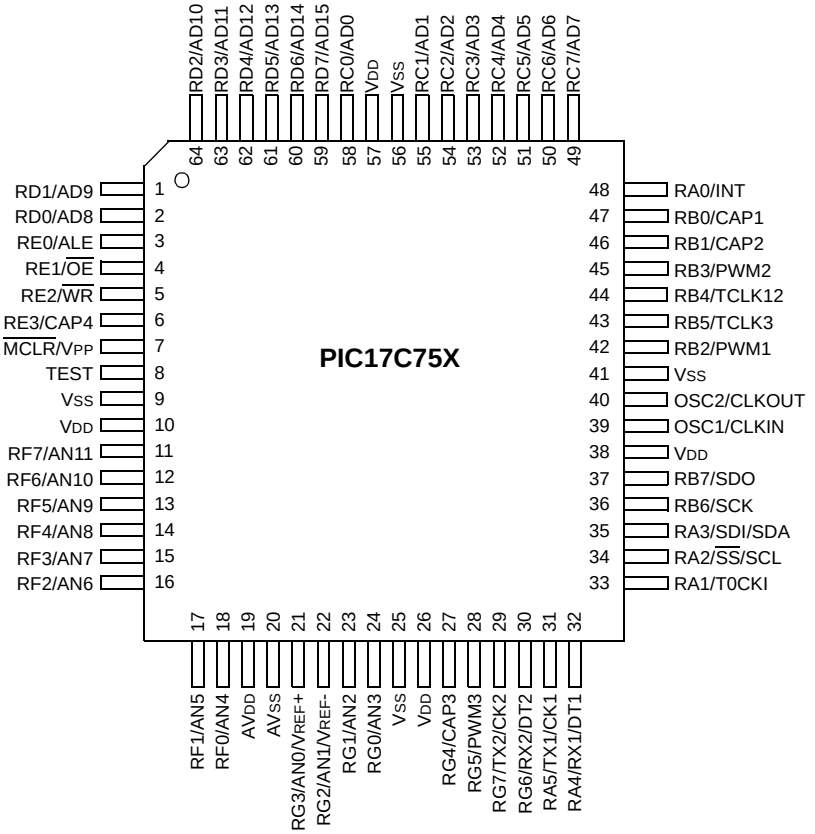


TABLE 5-4: INITIALIZATION CONDITIONS FOR SPECIAL FUNCTION REGISTERS (CONTINUED)

Register	Address	Power-on Reset Brown-out Reset	$\overline{\text{MCLR}}$ Reset WDT Reset	Wake-up from SLEEP through Interrupt
Bank 4				
PIR2	10h	000- 0010	000- 0010	uuu- uuuu(1)
PIE2	11h	000- 0000	000- 0000	uuu- uuuu
Unimplemented	12h	---- ----	---- ----	---- ----
RCSTA2	13h	0000 -00x	0000 -00u	uuuu -uuu
RCREG2	14h	xxxx xxxx	uuuu uuuu	uuuu uuuu
TXSTA2	15h	0000 --1x	0000 --1u	uuuu --uu
TXREG2	16h	xxxx xxxx	uuuu uuuu	uuuu uuuu
SPBRG2	17h	0000 0000	0000 0000	uuuu uuuu
Bank 5				
DDRF	10h	1111 1111	1111 1111	uuuu uuuu
PORTF(4)	11h	0000 0000	0000 0000	uuuu uuuu
DDRG	12h	1111 1111	1111 1111	uuuu uuuu
PORTG(4)	13h	xxxx 0000	uuuu 0000	uuuu uuuu
ADCON0	14h	0000 -0-0	0000 -0-0	uuuu uuuu
ADCON1	15h	000- 0000	000- 0000	uuuu uuuu
ADRESL	16h	xxxx xxxx	uuuu uuuu	uuuu uuuu
ADRESH	17h	xxxx xxxx	uuuu uuuu	uuuu uuuu
Bank 6				
SSPADD	10h	0000 0000	0000 0000	uuuu uuuu
SSPCON1	11h	0000 0000	0000 0000	uuuu uuuu
SSPCON2	12h	0000 0000	0000 0000	uuuu uuuu
SSPSTAT	13h	0000 0000	0000 0000	uuuu uuuu
SSPBUF	14h	xxxx xxxx	uuuu uuuu	uuuu uuuu
Unimplemented	15h	---- ----	---- ----	---- ----
Unimplemented	16h	---- ----	---- ----	---- ----
Unimplemented	17h	---- ----	---- ----	---- ----

Legend: u = unchanged, x = unknown, - = unimplemented, read as '0', q = value depends on condition

Note 1: One or more bits in INTSTA, PIR1, PIR2 will be affected (to cause wake-up).

2: When the wake-up is due to an interrupt and the GLINTD bit is cleared, the PC is loaded with the interrupt vector.

3: See Table 5-3 for RESET value of specific condition.

4: This is the value that will be in the port output latch.

5: When the device is configured for Microprocessor or Extended Microcontroller mode, the operation of this port does not rely on these registers.

6: On any device RESET, these pins are configured as inputs.

7.2.2.1 ALU Status Register (ALUSTA)

The ALUSTA register contains the status bits of the Arithmetic and Logic Unit and the mode control bits for the indirect addressing register.

As with all the other registers, the ALUSTA register can be the destination for any instruction. If the ALUSTA register is the destination for an instruction that affects the Z, DC, C, or OV bits, then the write to these three bits is disabled. These bits are set or cleared according to the device logic. Therefore, the result of an instruction with the ALUSTA register as destination may be different than intended.

For example, the `CLRF ALUSTA, F` instruction will clear the upper four bits and set the Z bit. This leaves the ALUSTA register as `0000u1uu` (where `u` = unchanged).

It is recommended, therefore, that only `BCF`, `BSF`, `SWAPF` and `MOVWF` instructions be used to alter the ALUSTA register, because these instructions do not affect any status bits. To see how other instructions affect the status bits, see the "Instruction Set Summary."

Note 1: The C and DC bits operate as a $\overline{\text{borrow}}$ and digit $\overline{\text{borrow}}$ bit, respectively, in subtraction. See the `SUBLW` and `SUBWF` instructions for examples.

2: The overflow bit will be set if the 2's complement result exceeds +127, or is less than -128.

The Arithmetic and Logic Unit (ALU) is capable of carrying out arithmetic or logical operations on two operands, or a single operand. All single operand instructions operate either on the WREG register, or the given file register. For two operand instructions, one of the operands is the WREG register and the other is either a file register, or an 8-bit immediate constant.

REGISTER 7-1: ALUSTA REGISTER (ADDRESS: 04h, UNBANKED)

R/W-1	R/W-1	R/W-1	R/W-1	R/W-x	R/W-x	R/W-x	R/W-x
FS3	FS2	FS1	FS0	OV	Z	DC	C
bit 7							bit 0

bit 7-6 **FS3:FS2**: FSR1 Mode Select bits
`00` = Post auto-decrement FSR1 value
`01` = Post auto-increment FSR1 value
`1x` = FSR1 value does not change

bit 5-4 **FS1:FS0**: FSR0 Mode Select bits
`00` = Post auto-decrement FSR0 value
`01` = Post auto-increment FSR0 value
`1x` = FSR0 value does not change

bit 3 **OV**: Overflow bit
This bit is used for signed arithmetic (2's complement). It indicates an overflow of the 7-bit magnitude, which causes the sign bit (bit7) to change state.
`1` = Overflow occurred for signed arithmetic (in this arithmetic operation)
`0` = No overflow occurred

bit 2 **Z**: Zero bit
`1` = The result of an arithmetic or logic operation is zero
`0` = The result of an arithmetic or logic operation is not zero

bit 1 **DC**: Digit carry/$\overline{\text{borrow}}$ bit
For `ADDWF` and `ADDLW` instructions.
`1` = A carry-out from the 4th low order bit of the result occurred
`0` = No carry-out from the 4th low order bit of the result

Note: For borrow, the polarity is reversed.

bit 0 **C**: Carry/$\overline{\text{borrow}}$ bit

For `ADDWF` and `ADDLW` instructions. Note that a subtraction is executed by adding the two's complement of the second operand.
For rotate (`RRCF`, `RLCF`) instructions, this bit is loaded with either the high or low order bit of the source register.
`1` = A carry-out from the Most Significant bit of the result occurred
`0` = No carry-out from the Most Significant bit of the result

Note: For borrow, the polarity is reversed.

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
- n = Value at POR Reset	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

13.2 Timer3

Timer3 is a 16-bit timer consisting of the TMR3H and TMR3L registers. TMR3H is the high byte of the timer and TMR3L is the low byte. This timer has an associated 16-bit period register (PR3H/CA1H:PR3L/CA1L). This period register can be software configured to be a another 16-bit capture register.

When the TMR3CS bit (TCON1<2>) is clear, the timer increments every instruction cycle (FOSC/4). When TMR3CS is set, the counter increments on every falling edge of the RB5/TCLK3 pin. In either mode, the TMR3ON bit must be set for the timer/counter to increment. When TMR3ON is clear, the timer will not increment or set flag bit TMR3IF.

Timer3 has two modes of operation, depending on the CA1/PR3 bit (TCON2<3>). These modes are:

- Three capture and one period register mode
- Four capture register mode

The PIC17C7XX has up to four 16-bit capture registers that capture the 16-bit value of TMR3 when events are detected on capture pins. There are four capture pins (RB0/CAP1, RB1/CAP2, RG4/CAP3, and RE3/CAP4), one for each capture register pair. The capture pins are multiplexed with the I/O pins. An event can be:

- A rising edge
- A falling edge
- Every 4th rising edge
- Every 16th rising edge

Each 16-bit capture register has an interrupt flag associated with it. The flag is set when a capture is made. The capture modules are truly part of the Timer3 block. Figure 13-5 and Figure 13-6 show the block diagrams for the two modes of operation.

13.2.1 THREE CAPTURE AND ONE PERIOD REGISTER MODE

In this mode, registers PR3H/CA1H and PR3L/CA1L constitute a 16-bit period register. A block diagram is shown in Figure 13-5. The timer increments until it equals the period register and then resets to 0000h on the next timer clock. TMR3 Interrupt Flag bit (TMR3IF) is set at this point. This interrupt can be disabled by clearing the TMR3 Interrupt Enable bit (TMR3IE). TMR3IF must be cleared in software.

FIGURE 13-5: TIMER3 WITH THREE CAPTURE AND ONE PERIOD REGISTER BLOCK DIAGRAM

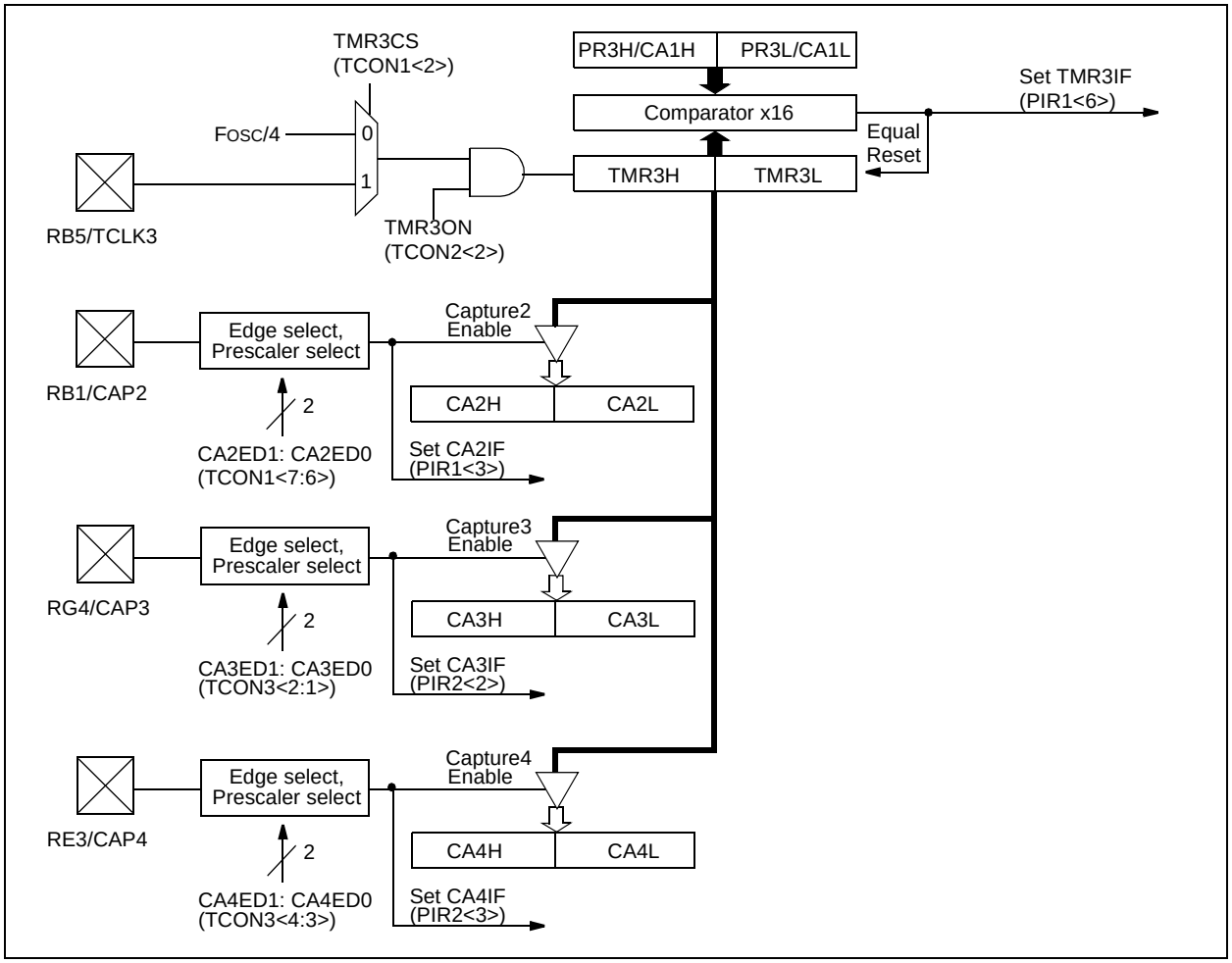


TABLE 14-8: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	$\overline{\text{MCLR}}$, WDT
16h, Bank 1	PIR1	RBIF	TMR3IF	TMR2IF	TMR1IF	CA2IF	CA1IF	TX1IF	RC1IF	x000 0010	u000 0010
17h, Bank 1	PIE1	RBIE	TMR3IE	TMR2IE	TMR1IE	CA2IE	CA1IE	TX1IE	RC1IE	0000 0000	0000 0000
13h, Bank 0	RCSTA1	SPEN	RX9	SREN	CREN	—	FERR	OERR	RX9D	0000 -00x	0000 -00u
16h, Bank 0	TXREG1	TX7	TX6	TX5	TX4	TX3	TX2	TX1	TX0	xxxx xxxx	uuuu uuuu
15h, Bank 0	TXSTA1	CSRC	TX9	TXEN	SYNC	—	—	TRMT	TX9D	0000 --1x	0000 --1u
17h, Bank 0	SPBRG1	Baud Rate Generator Register								0000 0000	0000 0000
10h, Bank 4	PIR2	SSPIF	BCLIF	ADIF	—	CA4IF	CA3IF	TX2IF	RC2IF	000- 0010	000- 0010
11h, Bank 4	PIE2	SSPIE	BCLIE	ADIE	—	CA4IE	CA3IE	TX2IE	RC2IE	000- 0000	000- 0000
13h, Bank 4	RCSTA2	SPEN	RX9	SREN	CREN	—	FERR	OERR	RX9D	0000 -00x	0000 -00u
16h, Bank 4	TXREG2	TX7	TX6	TX5	TX4	TX3	TX2	TX1	TX0	xxxx xxxx	uuuu uuuu
15h, Bank 4	TXSTA2	CSRC	TX9	TXEN	SYNC	—	—	TRMT	TX9D	0000 --1x	0000 --1u
17h, Bank 4	SPBRG2	Baud Rate Generator Register								0000 0000	0000 0000

Legend: x = unknown, u = unchanged, - = unimplemented, read as a '0'. Shaded cells are not used for synchronous master transmission.

FIGURE 14-8: SYNCHRONOUS TRANSMISSION

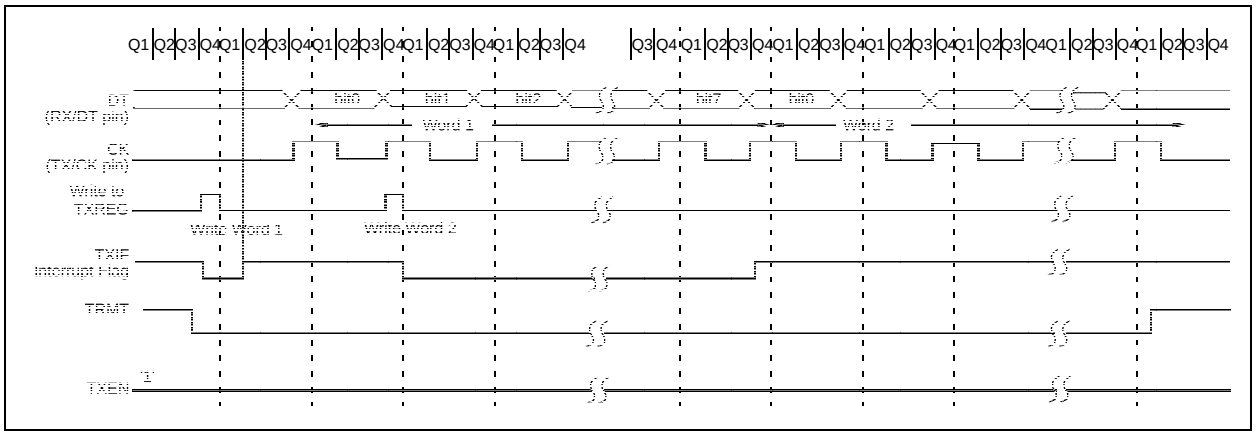
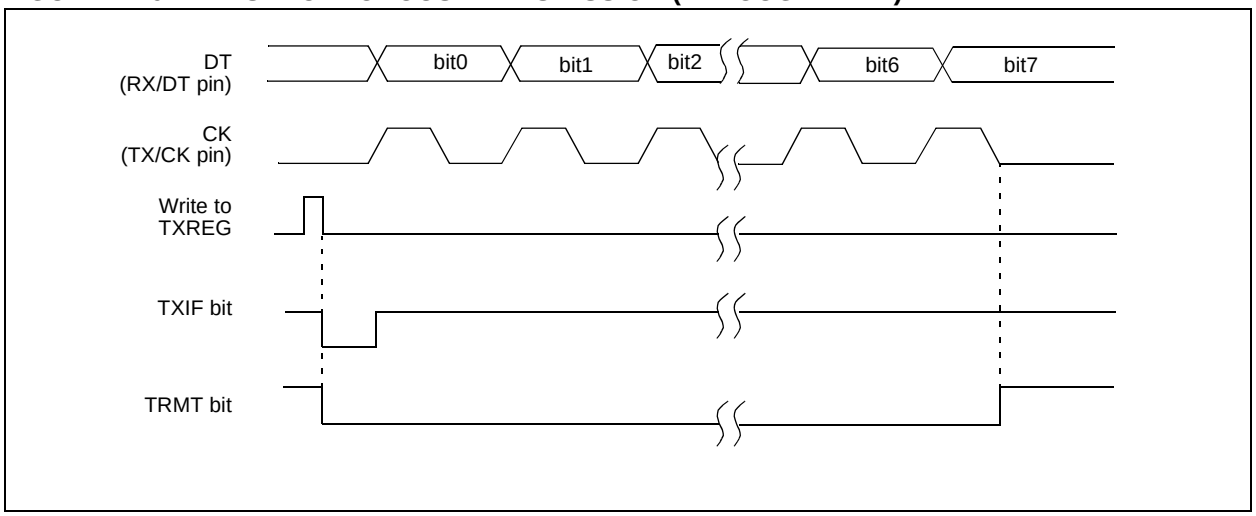


FIGURE 14-9: SYNCHRONOUS TRANSMISSION (THROUGH TXEN)



15.0 MASTER SYNCHRONOUS SERIAL PORT (MSSP) MODULE

The Master Synchronous Serial Port (MSSP) module is a serial interface useful for communicating with other peripheral or microcontroller devices. These peripheral devices may be serial EEPROMs, shift registers, display drivers, A/D converters, etc. The MSSP module can operate in one of two modes:

- Serial Peripheral Interface (SPI)
- Inter-Integrated Circuit™ (I^2C)

Figure 15-1 shows a block diagram for the SPI mode, while Figure 15-2 and Figure 15-3 show the block diagrams for the two different I^2C modes of operation.

FIGURE 15-1: SPI MODE BLOCK DIAGRAM

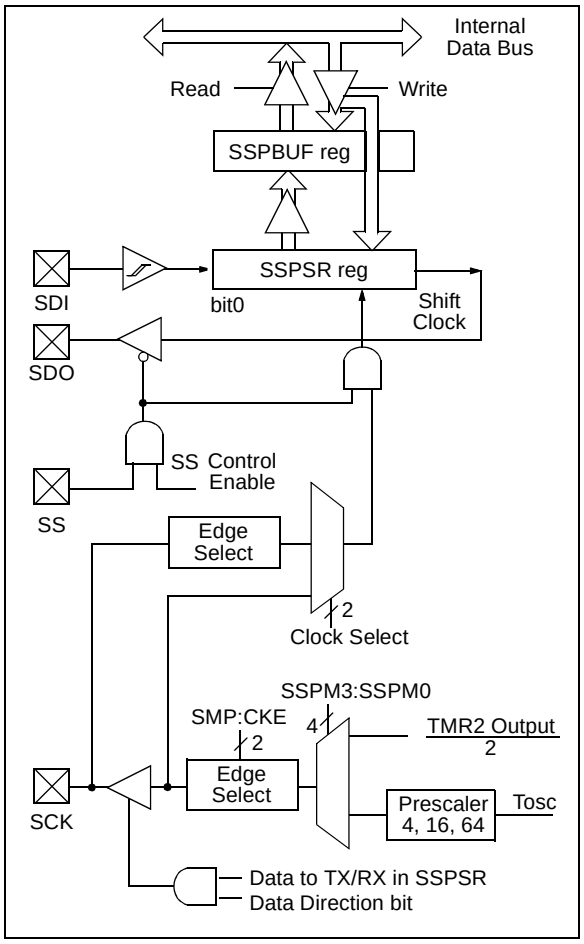


FIGURE 15-2: I^2C SLAVE MODE BLOCK DIAGRAM

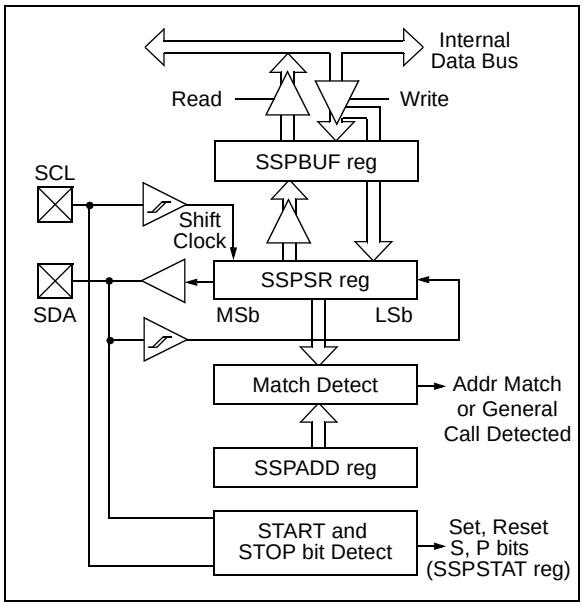
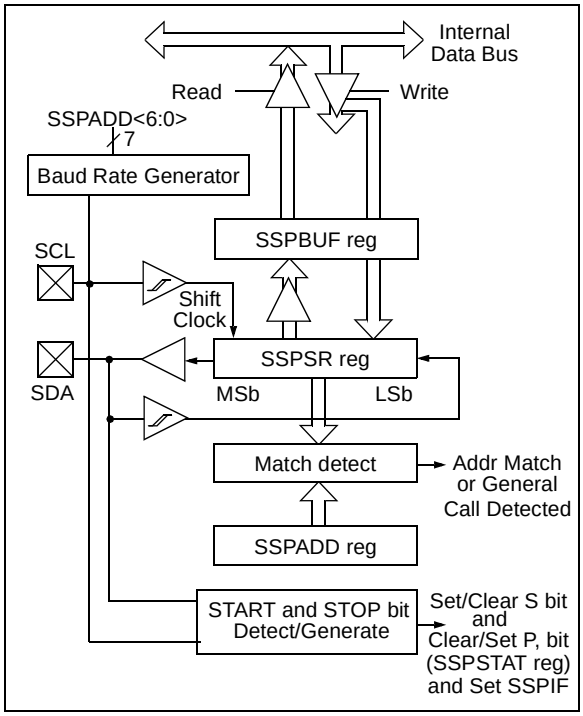


FIGURE 15-3: I^2C MASTER MODE BLOCK DIAGRAM



15.2.1.1 Addressing

Once the MSSP module has been enabled, it waits for a START condition to occur. Following the START condition, the 8-bits are shifted into the SSPSR register. All incoming bits are sampled with the rising edge of the clock (SCL) line. The value of register SSPSR<7:1> is compared to the value of the SSPADD register. The address is compared on the falling edge of the eighth clock (SCL) pulse. If the addresses match and the BF and SSPOV bits are clear, the following events occur:

a) The SSPSR register value is loaded into the SSPBUF register on the falling edge of the 8th SCL pulse.
b) The buffer full bit, BF, is set on the falling edge of the 8th SCL pulse.
c) An $\overline{ACK}$ pulse is generated.
d) SSP interrupt flag bit, SSPIF (PIR2<7>), is set (interrupt is generated if enabled) - on the falling edge of the 9th SCL pulse.

In 10-bit address mode, two address bytes need to be received by the slave. The five Most Significant bits (MSbs) of the first address byte specify if this is a 10-bit address. Bit R/$\overline{W}$ (SSPSTAT<2>) must specify a write so the slave device will receive the second address byte. For a 10-bit address, the first byte would equal '1111 0 A9 A8 0', where A9 and A8 are the two MSbs of the address. The sequence of events for a 10-bit address is as follows, with steps 7- 9 for slave-transmitter:

1. Receive first (high) byte of Address (bits SSPIF, BF and bit UA (SSPSTAT<1>) are set).
2. Update the SSPADD register with second (low) byte of Address (clears bit UA and releases the SCL line).
3. Read the SSPBUF register (clears bit BF) and clear flag bit SSPIF.
4. Receive second (low) byte of Address (bits SSPIF, BF and UA are set).
5. Update the SSPADD register with the first (high) byte of Address. This will clear bit UA and release the SCL line.
6. Read the SSPBUF register (clears bit BF) and clear flag bit SSPIF.
7. Receive Repeated Start condition.
8. Receive first (high) byte of Address (bits SSPIF and BF are set).
9. Read the SSPBUF register (clears bit BF) and clear flag bit SSPIF.

Note:	Following the Repeated Start condition (step 7) in 10-bit mode, the user only needs to match the first 7-bit address. The user does not update the SSPADD for the second half of the address.

15.2.1.2 Slave Reception

When the R/$\overline{W}$ bit of the address byte is clear and an address match occurs, the R/$\overline{W}$ bit of the SSPSTAT register is cleared. The received address is loaded into the SSPBUF register.

When the address byte overflow condition exists, then no acknowledge ($\overline{ACK}$) pulse is given. An overflow condition is defined as either bit BF (SSPSTAT<0>) is set, or bit SSPOV (SSPCON1<6>) is set.

An SSP interrupt is generated for each data transfer byte. Flag bit SSPIF (PIR2<7>) must be cleared in software. The SSPSTAT register is used to determine the status of the received byte.

Note:	The SSPBUF will be loaded if the SSPOV bit is set and the BF flag is cleared. If a read of the SSPBUF was performed, but the user did not clear the state of the SSPOV bit before the next receive occurred, the $\overline{ACK}$ is not sent and the SSPBUF is updated.

TABLE 15-2: DATA TRANSFER RECEIVED BYTE ACTIONS

Status Bits as Data Transfer is Received		SSPSR → SSPBUF	Generate $\overline{ACK}$ Pulse	Set bit SSPIF (SSP Interrupt occurs if enabled)
BF	SSPOV			
0	0	Yes	Yes	Yes
1	0	No	No	Yes
1	1	No	No	Yes
0	1	Yes	No	Yes

Note 1: Shaded cells show the conditions where the user software did not properly clear the overflow condition.

15.2.18.3 Bus Collision During a STOP Condition

Bus collision occurs during a STOP condition if:

a) After the SDA pin has been de-asserted and allowed to float high, SDA is sampled low after the BRG has timed out.

b) After the SCL pin is de-asserted, SCL is sampled low before SDA goes high.

The STOP condition begins with SDA asserted low. When SDA is sampled low, the SCL pin is allowed to float. When the pin is sampled high (clock arbitration), the baud rate generator is loaded with SSPADD<6:0> and counts down to ‘0’. After the BRG times out, SDA is sampled. If SDA is sampled low, a bus collision has occurred. This is due to another master attempting to drive a data '0'. If the SCL pin is sampled low before SDA is allowed to float high, a bus collision occurs. This is another case of another master attempting to drive a data '0' (Figure 15-40).

FIGURE 15-40: BUS COLLISION DURING A STOP CONDITION (CASE 1)

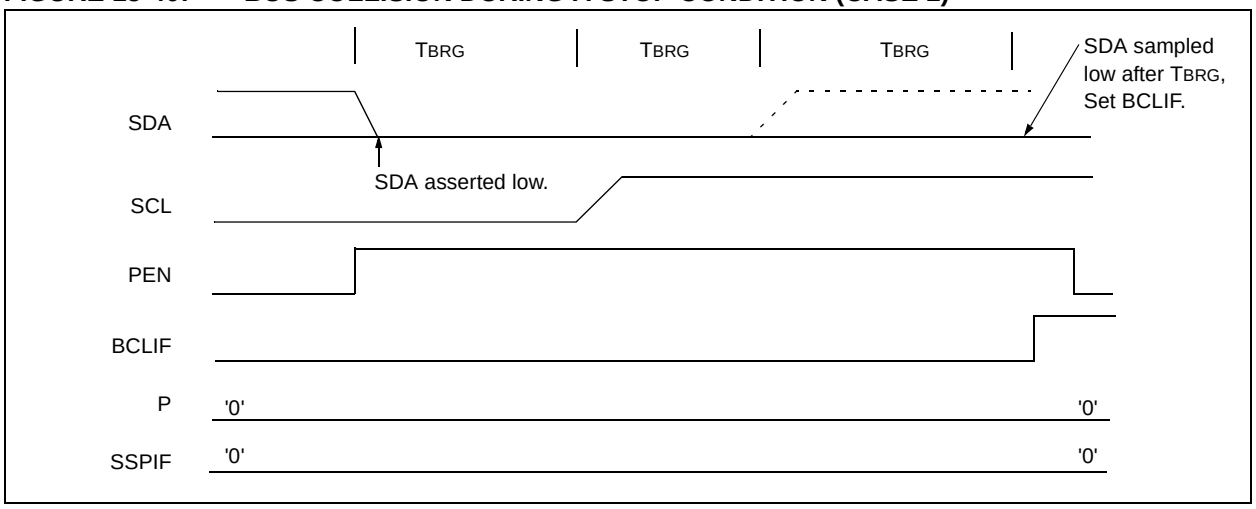
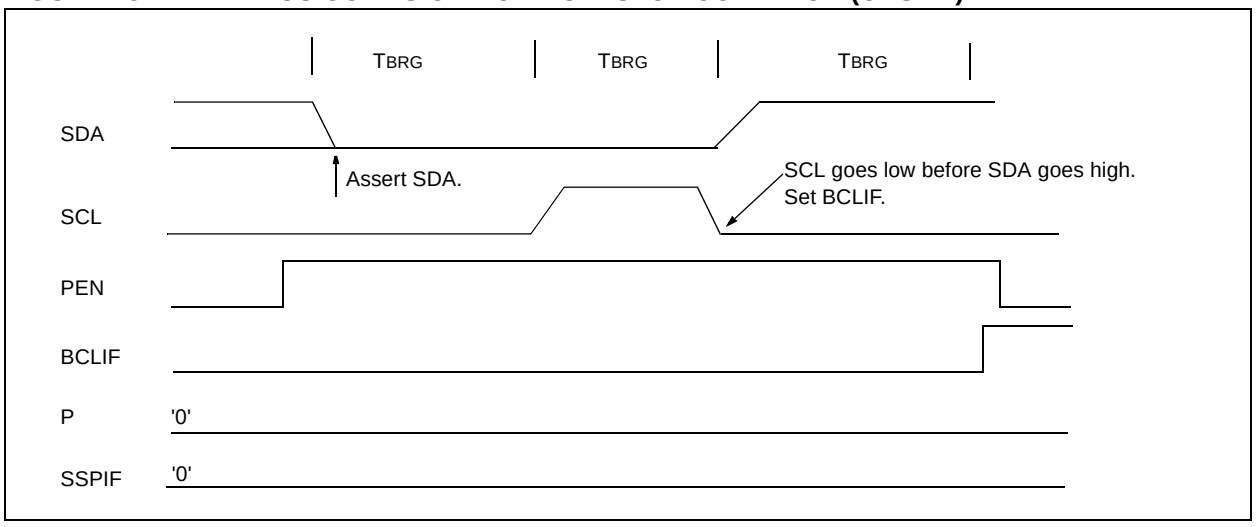


FIGURE 15-41: BUS COLLISION DURING A STOP CONDITION (CASE 2)



15.4 Example Program

Example 15-2 shows MPLAB® C17 'C' code for using the I^2C module in Master mode to communicate with a 24LC01B serial EEPROM. This example uses the PIC® MCU 'C' libraries included with MPLAB C17.

EXAMPLE 15-2: INTERFACING TO A 24LC01B SERIAL EEPROM (USING MPLAB C17)

```
// Include necessary header files
#include <p17c756.h>     // Processor header file
#include <delays.h>      // Delay routines header file
#include <stdlib.h>      // Standard Library header file
#include <i2c16.h>       // I2C routines header file

#define CONTROL 0xa0     // Control byte definition for 24LC01B

// Function declarations
void main(void);
void WritePORTD(static unsigned char data);
void ByteWrite(static unsigned char address,static unsigned char data);
unsigned char ByteRead(static unsigned char address);
void ACKPoll(void);

// Main program
void main(void)
{
static unsigned char address;    // I2C address of 24LC01B
static unsigned char datao;      // Data written to 24LC01B
static unsigned char datai;      // Data read from 24LC01B

    address = 0;                 // Preset address to 0
    OpenI2C(MASTER,SLEW_ON);     // Configure I2C Module Master mode, Slew rate control on
    SSPADD = 39;                 // Configure clock for 100KHz

    while(address<128)           // Loop 128 times, 24LC01B is 128x8
    {
        datao = PORTB;
        do
        {
            ByteWrite(address,datao);   // Write data to EEPROM
            ACKPoll();                  // Poll the 24LC01B for state
            datai = ByteRead(address);  // Read data from EEPROM into SSPBUF
        } while(datai != datao);        // Loop as long as data not correctly
                                        //    written to 24LC01B

        address++;                      // Increment address
    }
    while(1)                            // Done writing 128 bytes to 24LC01B, Loop forever
    {
        Nop();
    }
}
```

16.2 Selecting the A/D Conversion Clock

The A/D conversion time per bit is defined as TAD. The A/D conversion requires a minimum 12TAD per 10-bit conversion. The source of the A/D conversion clock is software selected. The four possible options for TAD are:

- 8TOSC
- 32TOSC
- 64TOSC
- Internal RC oscillator

For correct A/D conversions, the A/D conversion clock (TAD) must be selected to ensure a minimum TAD time of 1.6 μs.

Table 16-1 and Table 16-2 show the resultant TAD times derived from the device operating frequencies and the A/D clock source selected. These times are for standard voltage range devices.

TABLE 16-1: TAD vs. DEVICE OPERATING FREQUENCIES (STANDARD DEVICES (C))

AD Clock Source (TAD)		Max FOSC (MHz)
Operation	ADCS1:ADCS0	
8TOSC	00	5
32TOSC	01	20
64TOSC	10	33
RC	11	—

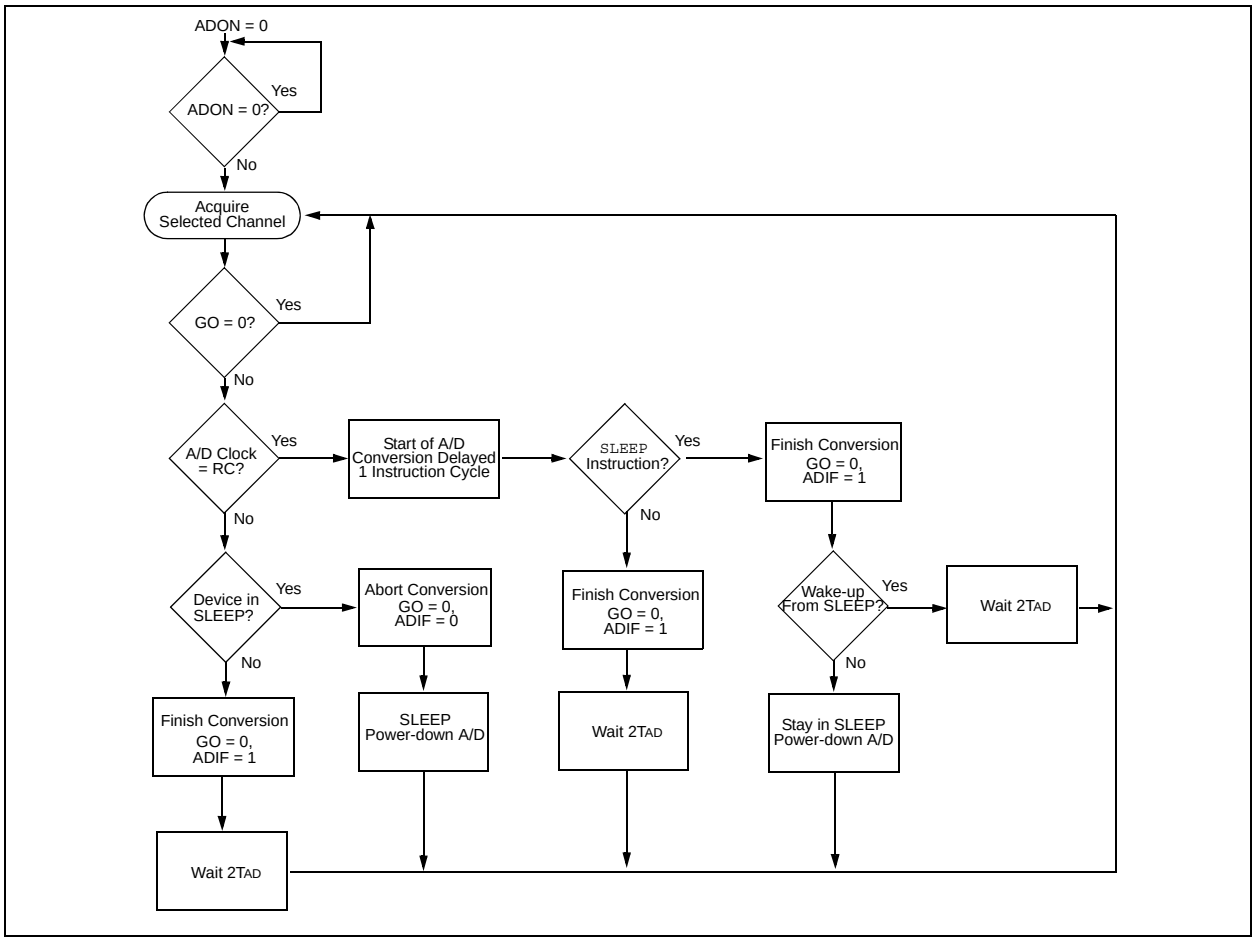
Note: When the device frequency is greater than 1 MHz, the RC A/D conversion clock source is only recommended for SLEEP operation.

TABLE 16-2: TAD vs. DEVICE OPERATING FREQUENCIES (EXTENDED VOLTAGE DEVICES (LC))

AD Clock Source (TAD)		Max FOSC (MHz)
Operation	ADCS1:ADCS0	
8TOSC	00	2.67
32TOSC	01	10.67
64TOSC	10	21.33
RC	11	—

Note: When the device frequency is greater than 1 MHz, the RC A/D conversion clock source is only recommended for SLEEP operation.

FIGURE 16-5: FLOW CHART OF A/D OPERATION



16.7 A/D Accuracy/Error

In systems where the device frequency is low, use of the A/D RC clock is preferred. At moderate to high frequencies, TAD should be derived from the device oscillator.

The absolute accuracy specified for the A/D converter includes the sum of all contributions for quantization error, integral error, differential error, full scale error, offset error, and monotonicity. It is defined as the maximum deviation from an actual transition versus an ideal transition for any code. The absolute error of the A/D converter is specified at < ±1 LSb for VDD = VREF (over the device's specified operating range). However, the accuracy of the A/D converter will degrade as VREF diverges from VDD.

For a given range of analog inputs, the output digital code will be the same. This is due to the quantization of the analog input to a digital code. Quantization error is typically ± 1/2 LSb and is inherent in the analog to digital conversion process. The only way to reduce quantization error is to increase the resolution of the A/D converter or oversample.

Offset error measures the first actual transition of a code versus the first ideal transition of a code. Offset error shifts the entire transfer function. Offset error can be calibrated out of a system or introduced into a system through the interaction of the total leakage current and source impedance at the analog input.

Gain error measures the maximum deviation of the last actual transition and the last ideal transition adjusted for offset error. This error appears as a change in slope of the transfer function. The difference in gain error to full scale error is that full scale does not take offset error into account. Gain error can be calibrated out in software.

Linearity error refers to the uniformity of the code changes. Linearity errors cannot be calibrated out of the system. Integral non-linearity error measures the actual code transition versus the ideal code transition, adjusted by the gain error for each code.

Differential non-linearity measures the maximum actual code width versus the ideal code width. This measure is unadjusted.

The maximum pin leakage current is specified in the Device Data Sheet electrical specification (Table 20-2, parameter #D060).

In systems where the device frequency is low, use of the A/D RC clock is preferred. At moderate to high frequencies, TAD should be derived from the device oscillator. TAD must not violate the minimum and should be minimized to reduce inaccuracies due to noise and sampling capacitor bleed off.

In systems where the device will enter SLEEP mode after the start of the A/D conversion, the RC clock source selection is required. In this mode, the digital noise from the modules in SLEEP are stopped. This method gives high accuracy.

16.8 Connection Considerations

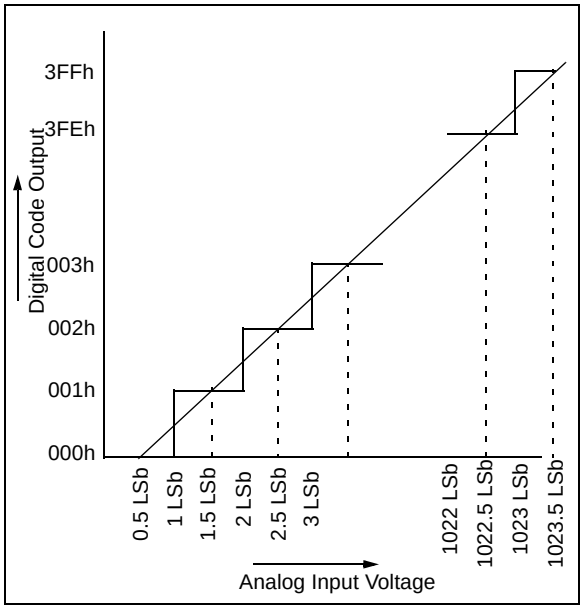
If the input voltage exceeds the rail values (VSS or VDD) by greater than 0.3V, then the accuracy of the conversion is out of specification.

An external RC filter is sometimes added for anti-aliasing of the input signal. The R component should be selected to ensure that the total source impedance is kept under the 10 kΩ recommended specification. Any external components connected (via hi-impedance) to an analog input pin (capacitor, zener diode, etc.) should have very little leakage current at the pin.

16.9 Transfer Function

The transfer function of the A/D converter is as follows: the first transition occurs when the analog input voltage (VAIN) equals Analog VREF / 1024 (Figure 16-7).

FIGURE 16-7: A/D TRANSFER FUNCTION



17.4.2 MINIMIZING CURRENT CONSUMPTION

To minimize current consumption, all I/O pins should be either at VDD, or VSS, with no external circuitry drawing current from the I/O pin. I/O pins that are hi-impedance inputs should be pulled high or low externally to avoid switching currents caused by floating inputs. The T0CKI input should be at VDD or VSS. The contributions from on-chip pull-ups on PORTB should also be considered and disabled, when possible.

17.5 Code Protection

The code in the program memory can be protected by selecting the microcontroller in Code Protected mode (PM2:PM0 = '000').

In this mode, instructions that are in the on-chip program memory space, can continue to read or write the program memory. An instruction that is executed outside of the internal program memory range will be inhibited from writing to, or reading from, program memory.

Note:	Microchip does not recommend code protecting windowed devices.

If the code protection bit(s) have not been programmed, the on-chip program memory can be read out for verification purposes.

DECF **Decrement f**

Syntax: [*label*] DECF f,d

Operands: $0 \le f \le 255$
$d \in [0,1]$

Operation: (f) – 1 → (dest)

Status Affected: OV, C, DC, Z

Encoding:

0000	011d	ffff	ffff

Description: Decrement register 'f'. If 'd' is 0, the result is stored in WREG. If 'd' is 1, the result is stored back in register 'f'.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

Example: `DECF CNT, 1`

Before Instruction
CNT = 0x01
Z = 0

After Instruction
CNT = 0x00
Z = 1

DECFSZ **Decrement f, skip if 0**

Syntax: [*label*] DECFSZ f,d

Operands: $0 \le f \le 255$
$d \in [0,1]$

Operation: (f) – 1 → (dest);
skip if result = 0

Status Affected: None

Encoding:

0001	011d	ffff	ffff

Description: The contents of register 'f' are decremented. If 'd' is 0, the result is placed in WREG. If 'd' is 1, the result is placed back in register 'f'.

If the result is 0, the next instruction, which is already fetched is discarded and a `NOP` is executed instead, making it a two-cycle instruction.

Words: 1

Cycles: 1(2)

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

If skip:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation

Example:
```
HERE    DECFSZ  CNT, 1
        GOTO    HERE
NZERO
ZERO
```

Before Instruction
PC = Address `(HERE)`

After Instruction
CNT = CNT - 1
If CNT = 0;
PC = Address `(HERE)`
If CNT ≠ 0;
PC = Address `(NZERO)`

MOVLR — **Move Literal to high nibble in BSR**

Syntax: [*label*] MOVLR k

Operands: $0 \le k \le 15$

Operation: $k \rightarrow (BSR<7:4>)$

Status Affected: None

Encoding:

1011	101x	kkkk	uuuu

Description: The 4-bit literal 'k' is loaded into the most significant 4-bits of the Bank Select Register (BSR). Only the high 4-bits of the Bank Select Register are affected. The lower half of the BSR is unchanged. The assembler will encode the “u” fields as 0.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'k'	Process Data	Write literal 'k' to BSR<7:4>

Example: `MOVLR 5`

Before Instruction
BSR register = 0x22

After Instruction
BSR register = 0x52

MOVLW — **Move Literal to WREG**

Syntax: [*label*] MOVLW k

Operands: $0 \le k \le 255$

Operation: $k \rightarrow (WREG)$

Status Affected: None

Encoding:

1011	0000	kkkk	kkkk

Description: The eight-bit literal 'k' is loaded into WREG.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'k'	Process Data	Write to WREG

Example: `MOVLW 0x5A`

After Instruction
WREG = 0x5A

RETFIE **Return from Interrupt**

Syntax: [*label*] RETFIE

Operands: None

Operation: TOS → (PC);
0 → GLINTD;
PCLATH is unchanged.

Status Affected: GLINTD

Encoding:

0000	0000	0000	0101

Description: Return from Interrupt. Stack is POP'ed and Top-of-Stack (TOS) is loaded in the PC. Interrupts are enabled by clearing the GLINTD bit. GLINTD is the global interrupt disable bit (CPUSTA<4>).

Words: 1

Cycles: 2

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	No operation	Clear GLINTD	POP PC from stack
No operation	No operation	No operation	No operation

Example: `RETFIE`

After Interrupt
PC = TOS
GLINTD = 0

RETLW **Return Literal to WREG**

Syntax: [*label*] RETLW k

Operands: $0 \le k \le 255$

Operation: k → (WREG); TOS → (PC);
PCLATH is unchanged

Status Affected: None

Encoding:

1011	0110	kkkk	kkkk

Description: WREG is loaded with the eight-bit literal 'k'. The program counter is loaded from the top of the stack (the return address). The high address latch (PCLATH) remains unchanged.

Words: 1

Cycles: 2

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'k'	Process Data	POP PC from stack, Write to WREG
No operation	No operation	No operation	No operation

Example:

```
    CALL TABLE ; WREG contains table
               ;  offset value
               ;  WREG now has
               ;  table value
    :
TABLE
    ADDWF PC   ; WREG = offset
    RETLW k0   ; Begin table
    RETLW k1   ;
    :
    :
    RETLW kn   ; End of table
```

Before Instruction
WREG = 0x07

After Instruction
WREG = value of k7

SUBWF **Subtract WREG from f**

Syntax: [*label*] SUBWF f,d

Operands: $0 \le f \le 255$
$d \in [0,1]$

Operation: (f) – (W) → (dest)

Status Affected: OV, C, DC, Z

Encoding:

0000	010d	ffff	ffff

Description: Subtract WREG from register 'f' (2's complement method). If 'd' is 0, the result is stored in WREG. If 'd' is 1, the result is stored back in register 'f'.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

Example 1: `SUBWF REG1, 1`

Before Instruction
REG1 = 3
WREG = 2
C = ?

After Instruction
REG1 = 1
WREG = 2
C = 1 ; result is positive
Z = 0

Example 2:

Before Instruction
REG1 = 2
WREG = 2
C = ?

After Instruction
REG1 = 0
WREG = 2
C = 1 ; result is zero
Z = 1

Example 3:

Before Instruction
REG1 = 1
WREG = 2
C = ?

After Instruction
REG1 = FF
WREG = 2
C = 0 ; result is negative
Z = 0

SUBWFB **Subtract WREG from f with Borrow**

Syntax: [*label*] SUBWFB f,d

Operands: $0 \le f \le 255$
$d \in [0,1]$

Operation: (f) – (W) – $\overline{C}$ → (dest)

Status Affected: OV, C, DC, Z

Encoding:

0000	001d	ffff	ffff

Description: Subtract WREG and the carry flag (borrow) from register 'f' (2's complement method). If 'd' is 0, the result is stored in WREG. If 'd' is 1, the result is stored back in register 'f'.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

Example 1: `SUBWFB REG1, 1`

Before Instruction
REG1 = 0x19 (0001 1001)
WREG = 0x0D (0000 1101)
C = 1

After Instruction
REG1 = 0x0C (0000 1011)
WREG = 0x0D (0000 1101)
C = 1 ; result is positive
Z = 0

Example2: `SUBWFB REG1,0`

Before Instruction
REG1 = 0x1B (0001 1011)
WREG = 0x1A (0001 1010)
C = 0

After Instruction
REG1 = 0x1B (0001 1011)
WREG = 0x00
C = 1 ; result is zero
Z = 1

Example3: `SUBWFB REG1,1`

Before Instruction
REG1 = 0x03 (0000 0011)
WREG = 0x0E (0000 1101)
C = 1

After Instruction
REG1 = 0xF5 (1111 0100) [2's comp]
WREG = 0x0E (0000 1101)
C = 0 ; result is negative
Z = 0

TABLE 20-18: A/D CONVERTER CHARACTERISTICS

Param. No.	Sym	Characteristic		Min	Typ†	Max	Units	Conditions
A01	NR	Resolution		—	—	10	bit	VREF+ = VDD = 5.12V, VSS ≤ VAIN ≤ VREF+
				—	—	10	bit	(VREF+ — VREF-) ≥ 3.0V, VREF- ≤ VAIN ≤ VREF+
A02	EABS	Absolute error		—	—	< ±1	LSb	VREF+ = VDD = 5.12V, VSS ≤ VAIN ≤ VREF+
				—	—	< ±1	LSb	(VREF+ — VREF-) ≥ 3.0V, VREF- ≤ VAIN ≤ VREF+
A03	EIL	Integral linearity error		—	—	< ±1	LSb	VREF+ = VDD = 5.12V, VSS ≤ VAIN ≤ VREF+
				—	—	< ±1	LSb	(VREF+ — VREF-) ≥ 3.0V, VREF- ≤ VAIN ≤ VREF+
A04	EDL	Differential linearity error		—	—	< ±1	LSb	VREF+ = VDD = 5.12V, VSS ≤ VAIN ≤ VREF+
				—	—	< ±1	LSb	(VREF+ — VREF-) ≥ 3.0V, VREF- ≤ VAIN ≤ VREF+
A05	EFS	Full scale error		—	—	< ±1	LSb	VREF+ = VDD = 5.12V, VSS ≤ VAIN ≤ VREF+
				—	—	< ±1	LSb	(VREF+ — VREF-) ≥ 3.0V, VREF- ≤ VAIN ≤ VREF+
A06	EOFF	Offset error		—	—	< ±1	LSb	VREF+ = VDD = 5.12V, VSS ≤ VAIN ≤ VREF+
				—	—	< ±1	LSb	(VREF+ — VREF-) ≥ 3.0V, VREF- ≤ VAIN ≤ VREF+
A10	—	Monotonicity		—	guaranteed(3)	—	—	VSS ≤ VAIN ≤ VREF
A20	VREF	Reference voltage (VREF+ — VREF-)		0V	—	—	V	VREF delta when changing voltage levels on VREF inputs
A20A				3V	—	—	V	Absolute minimum electrical spec. to ensure 10-bit accuracy
A21	VREF+	Reference voltage high		AVSS + 3.0V	—	AVDD + 0.3V	V	
A22	VREF-	Reference voltage low		Avss - 0.3V	—	AVDD - 3.0V	V	
A25	VAIN	Analog input voltage		AVSS - 0.3V	—	Vref + 0.3V	V	
A30	ZAIN	Recommended impedance of analog voltage source		—	—	10.0	kΩ	
A40	IAD	A/D conversion current (VDD)	PIC17**C**XXX	—	180	—	μA	Average current consumption when A/D is on **(Note 1)**
			PIC17**LC**XXX	—	90	—	μA	
A50	IREF	VREF input current **(Note 2)**		10	—	1000	μA	During VAIN acquisition. Based on differential of VHOLD to VAIN
				—	—	10	μA	During A/D conversion cycle

† Data in "Typ" column is at 5V, 25°C unless otherwise stated.

Note 1: When A/D is off, it will not consume any current other than minor leakage current. The power-down current spec includes any such leakage from the A/D module.

2: VREF current is from RG0 and RG1 pins or AVDD and AVSS pins, whichever is selected as reference input.

3: The A/D conversion result never decreases with an increase in the Input Voltage and has no missing codes.

21.0 PIC17C7XX DC AND AC CHARACTERISTICS

The graphs and tables provided in this section are for design guidance and are not tested nor guaranteed. In some graphs or tables the data presented is outside specified operating range (e.g., outside specified VDD range). This is for information only and devices are ensured to operate properly only within the specified range.

The data presented in this section is a statistical summary of data collected on units from different lots over a period of time.

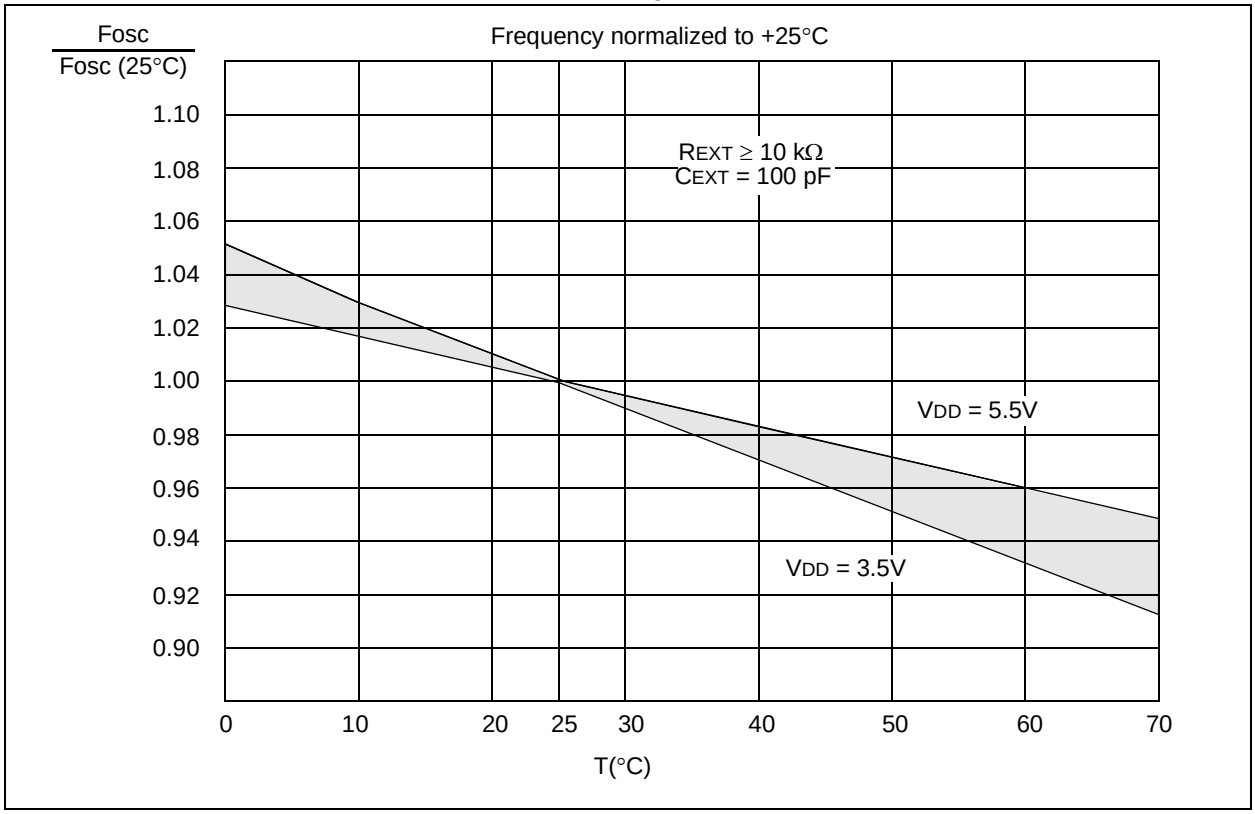
- **Typ** or **Typical** represents the mean of the distribution at 25°C.
- **Max** or **Maximum** represents (mean + 3σ) over the temperature range of -40°C to 85°C.
- **Min** or **Minimum** represents (mean - 3σ) over the temperature range of -40°C to 85°C.

Note: Standard deviation is denoted by sigma (σ).

TABLE 21-1: PIN CAPACITANCE PER PACKAGE TYPE

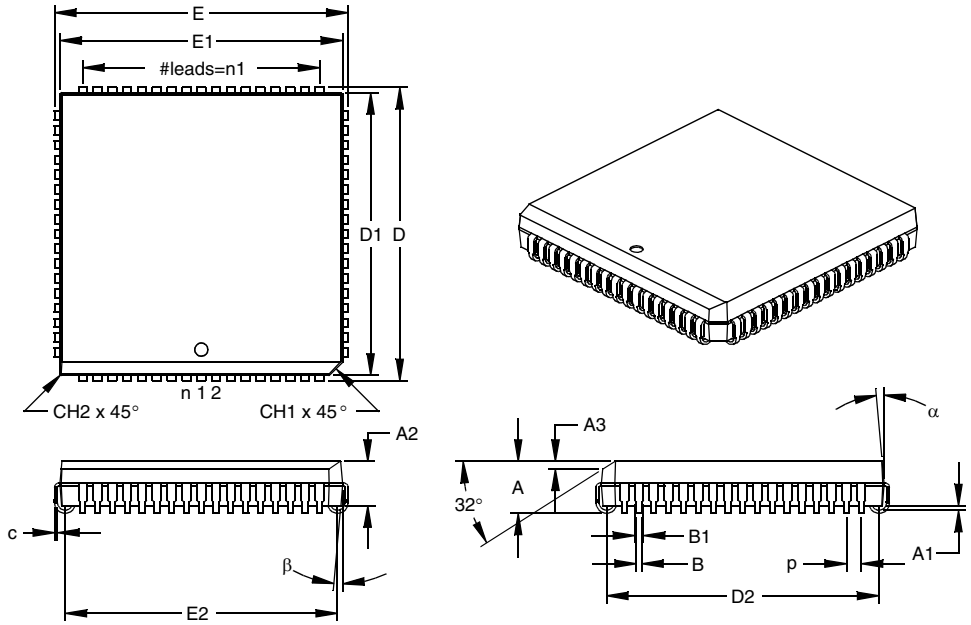
Pin Name	Typical Capacitance (pF)	
	68-pin PLCC	64-pin TQFP
All pins, except $\overline{MCLR}$, VDD, and VSS	10	10
$\overline{MCLR}$ pin	20	20

FIGURE 21-1: TYPICAL RC OSCILLATOR FREQUENCY vs. TEMPERATURE



84-Lead Plastic Leaded Chip Carrier (L) – Square (PLCC)

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		INCHES*			MILLIMETERS		
Dimension Limits		MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		68			68	
Pitch	p		.050			1.27	
Pins per Side	n1		17			17	
Overall Height	A	.165	.173	.180	4.19	4.39	4.57
Molded Package Thickness	A2	.145	.153	.160	3.68	3.87	4.06
Standoff §	A1	.020	.028	.035	0.51	0.71	0.89
Side 1 Chamfer Height	A3	.024	.029	.034	0.61	0.74	0.86
Corner Chamfer 1	CH1	.040	.045	.050	1.02	1.14	1.27
Corner Chamfer (others)	CH2	.000	.005	.010	0.00	0.13	0.25
Overall Width	E	.985	.990	.995	25.02	25.15	25.27
Overall Length	D	.985	.990	.995	25.02	25.15	25.27
Molded Package Width	E1	.950	.954	.958	24.13	24.23	24.33
Molded Package Length	D1	.950	.954	.958	24.13	24.23	24.33
Footprint Width	E2	.890	.920	.930	22.61	23.37	23.62
Footprint Length	D2	.890	.920	.930	22.61	23.37	23.62
Lead Thickness	c	.008	.011	.013	0.20	0.27	0.33
Upper Lead Width	B1	.026	.029	.032	0.66	0.74	0.81
Lower Lead Width	B	.013	.020	.021	0.33	0.51	0.53
Mold Draft Angle Top	α	0	5	10	0	5	10
Mold Draft Angle Bottom	β	0	5	10	0	5	10

* Controlling Parameter
§ Significant Characteristic
Notes:
Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side.
JEDEC Equivalent: MO-047
Drawing No. C04-093