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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	PIC
Core Size	8-Bit
Speed	8MHz
Connectivity	I²C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	66
Program Memory Size	16KB (8K x 16)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	678 x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	A/D 16x10b
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	80-TQFP
Supplier Device Package	80-TQFP (12x12)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic17lc762t-08i-pt

TABLE 3-1: PINOUT DESCRIPTIONS (CONTINUED)

Name	PIC17C75X			PIC17C76X		I/O/P Type	Buffer Type	Description
	DIP No.	PLCC No.	TQFP No.	PLCC No.	QFP No.			
								PORTC is a bi-directional I/O Port.
RC0/AD0	2	3	58	3	72	I/O	TTL	This is also the least significant byte (LSB) of the 16-bit wide system bus in Microprocessor mode or Extended Microcontroller mode. In multiplexed system bus configuration, these pins are address output as well as data input or output.
RC1/AD1	63	67	55	83	69	I/O	TTL	
RC2/AD2	62	66	54	82	68	I/O	TTL	
RC3/AD3	61	65	53	81	67	I/O	TTL	
RC4/AD4	60	64	52	80	66	I/O	TTL	
RC5/AD5	58	63	51	79	65	I/O	TTL	
RC6/AD6	58	62	50	78	64	I/O	TTL	
RC7/AD7	57	61	49	77	63	I/O	TTL	
								PORTD is a bi-directional I/O Port.
RD0/AD8	10	11	2	15	4	I/O	TTL	This is also the most significant byte (MSB) of the 16-bit system bus in Microprocessor mode or Extended Microcontroller mode. In multiplexed system bus configuration, these pins are address output as well as data input or output.
RD1/AD9	9	10	1	14	3	I/O	TTL	
RD2/AD10	8	9	64	9	78	I/O	TTL	
RD3/AD11	7	8	63	8	77	I/O	TTL	
RD4/AD12	6	7	62	7	76	I/O	TTL	
RD5/AD13	5	6	61	6	75	I/O	TTL	
RD6/AD14	4	5	60	5	74	I/O	TTL	
RD7/AD15	3	4	59	4	73	I/O	TTL	
								PORTE is a bi-directional I/O Port.
RE0/ALE	11	12	3	16	5	I/O	TTL	In Microprocessor mode or Extended Microcontroller mode, RE0 is the Address Latch Enable (ALE) output. Address should be latched on the falling edge of ALE output.
RE1/$\overline{OE}$	12	13	4	17	6	I/O	TTL	In Microprocessor or Extended Microcontroller mode, RE1 is the Output Enable ($\overline{OE}$) control output (active low).
RE2/$\overline{WR}$	13	14	5	18	7	I/O	TTL	In Microprocessor or Extended Microcontroller mode, RE2 is the Write Enable ($\overline{WR}$) control output (active low).
RE3/CAP4	14	15	6	19	8	I/O	ST	RE3 can also be the Capture4 input pin.
								PORTF is a bi-directional I/O Port.
RF0/AN4	26	28	18	36	24	I/O	ST	RF0 can also be analog input 4.
RF1/AN5	25	27	17	35	23	I/O	ST	RF1 can also be analog input 5.
RF2/AN6	24	26	16	30	18	I/O	ST	RF2 can also be analog input 6.
RF3/AN7	23	25	15	29	17	I/O	ST	RF3 can also be analog input 7.
RF4/AN8	22	24	14	28	16	I/O	ST	RF4 can also be analog input 8.
RF5/AN9	21	23	13	27	15	I/O	ST	RF5 can also be analog input 9.
RF6/AN10	20	22	12	26	14	I/O	ST	RF6 can also be analog input 10.
RF7/AN11	19	21	11	25	13	I/O	ST	RF7 can also be analog input 11.

Legend: I = Input only; O = Output only; I/O = Input/Output;
P = Power; — = Not Used; TTL = TTL input; ST = Schmitt Trigger input

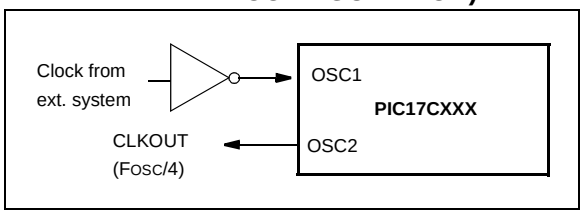
Note 1: The output is only available by the peripheral operation.

2: Open drain input/output pin. Pin forced to input upon any device RESET.

4.1.4 EXTERNAL CLOCK OSCILLATOR

In the EC oscillator mode, the OSC1 input can be driven by CMOS drivers. In this mode, the OSC1/CLKIN pin is hi-impedance and the OSC2/CLKOUT pin is the CLKOUT output (4 TOSC).

FIGURE 4-4: EXTERNAL CLOCK INPUT OPERATION (EC OSC CONFIGURATION)



4.1.5 EXTERNAL CRYSTAL OSCILLATOR CIRCUIT

Either a prepackaged oscillator can be used, or a simple oscillator circuit with TTL gates can be built. Prepackaged oscillators provide a wide operating range and better stability. A well designed crystal oscillator will provide good performance with TTL gates. Two types of crystal oscillator circuits can be used: one with series resonance, or one with parallel resonance.

Figure 4-5 shows implementation of a parallel resonant oscillator circuit. The circuit is designed to use the fundamental frequency of the crystal. The 74AS04 inverter performs the 180-degree phase shift that a parallel oscillator requires. The 4.7 kΩ resistor provides the negative feedback for stability. The 10 kΩ potentiometer biases the 74AS04 in the linear region. This could be used for external oscillator designs.

FIGURE 4-5: EXTERNAL PARALLEL RESONANT CRYSTAL OSCILLATOR CIRCUIT

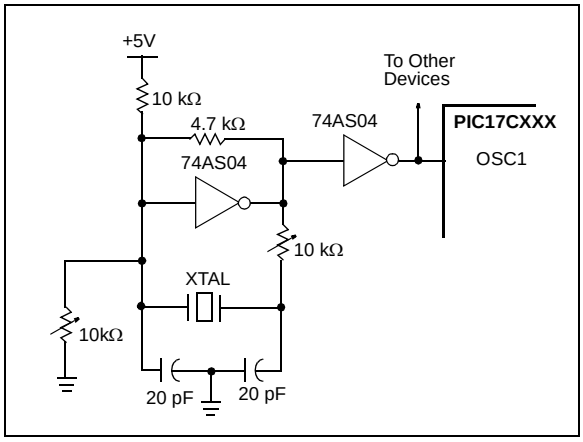


Figure 4-6 shows a series resonant oscillator circuit. This circuit is also designed to use the fundamental frequency of the crystal. The inverter performs a 180-degree phase shift in a series resonant oscillator circuit. The 330 Ω resistors provide the negative feedback to bias the inverters in their linear region.

FIGURE 4-6: EXTERNAL SERIES RESONANT CRYSTAL OSCILLATOR CIRCUIT

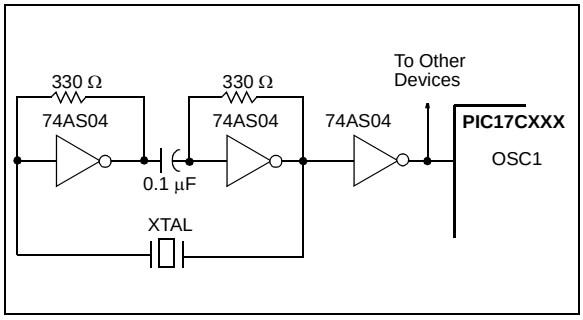


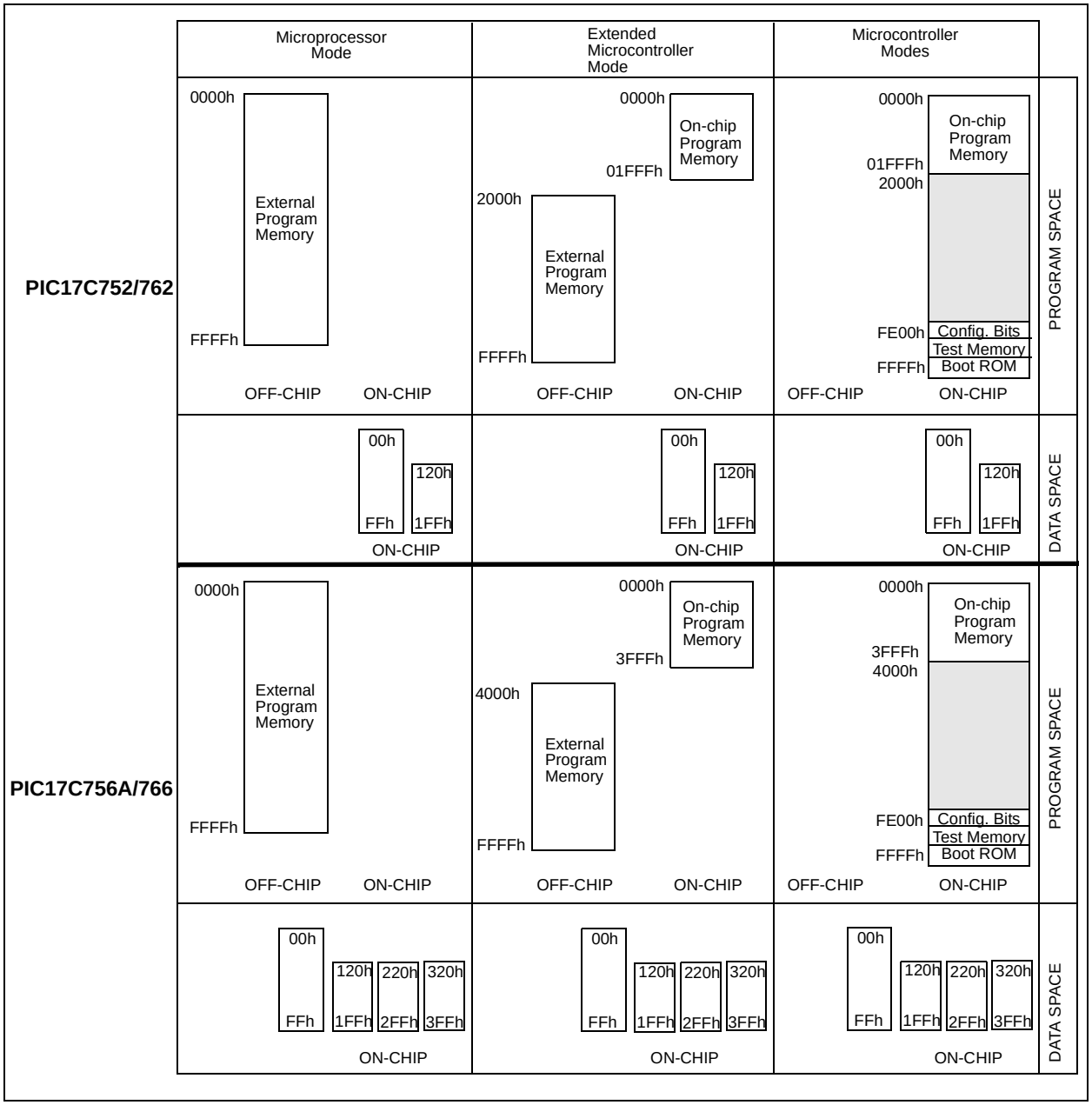
TABLE 7-1: MODE MEMORY ACCESS

Operating Mode	Internal Program Memory	Configuration Bits, Test Memory, Boot ROM
Microprocessor	No Access	No Access
Microcontroller	Access	Access
Extended Microcontroller	Access	No Access
Protected Microcontroller	Access	Access

The PIC17C7XX can operate in modes where the program memory is off-chip. They are the Microprocessor and Extended Microcontroller modes. The Microprocessor mode is the default for an unprogrammed device.

Regardless of the processor mode, data memory is always on-chip.

FIGURE 7-2: MEMORY MAP IN DIFFERENT MODES



7.3 Stack Operation

PIC17C7XX devices have a 16 x 16-bit hardware stack (Figure 7-1). The stack is not part of either the program or data memory space, and the stack pointer is neither readable nor writable. The PC (Program Counter) is "PUSH'd" onto the stack when a CALL or LCALL instruction is executed, or an interrupt is acknowledged. The stack is "POP'd" in the event of a RETURN, RETLW, or a RETFIE instruction execution. PCLATH is not affected by a "PUSH" or a "POP" operation.

The stack operates as a circular buffer, with the stack pointer initialized to '0' after all RESETS. There is a stack available bit (STKAV) to allow software to ensure that the stack will not overflow. The STKAV bit is set after a device RESET. When the stack pointer equals Fh, STKAV is cleared. When the stack pointer rolls over from Fh to 0h, the STKAV bit will be held clear until a device RESET.

Note 1: There is not a status bit for stack underflow. The STKAV bit can be used to detect the underflow which results in the stack pointer being at the Top-of-Stack.

2: There are no instruction mnemonics called PUSH or POP. These are actions that occur from the execution of the CALL, RETURN, RETLW and RETFIE instructions, or the vectoring to an interrupt vector.

3: After a RESET, if a "POP" operation occurs before a "PUSH" operation, the STKAV bit will be cleared. This will appear as if the stack is full (underflow has occurred). If a "PUSH" operation occurs next (before another "POP"), the STKAV bit will be locked clear. Only a device RESET will cause this bit to set.

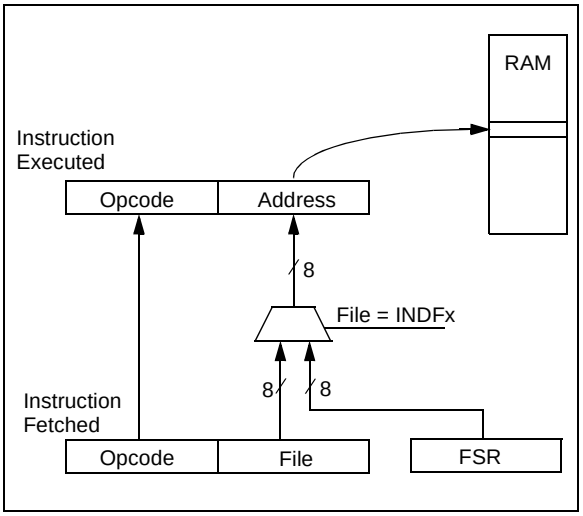
After the device is "PUSH'd" sixteen times (without a "POP"), the seventeenth push overwrites the value from the first push. The eighteenth push overwrites the second push (and so on).

7.4 Indirect Addressing

Indirect addressing is a mode of addressing data memory where the data memory address in the instruction is not fixed. That is, the register that is to be read or written can be modified by the program. This can be useful for data tables in the data memory. Figure 7-6 shows the operation of indirect addressing. This depicts the moving of the value to the data memory address specified by the value of the FSR register.

Example 7-1 shows the use of indirect addressing to clear RAM in a minimum number of instructions. A similar concept could be used to move a defined number of bytes (block) of data to the USART transmit register (TXREG). The starting address of the block of data to be transmitted could easily be modified by the program.

FIGURE 7-6: INDIRECT ADDRESSING



7.4.1 INDIRECT ADDRESSING REGISTERS

The PIC17C7XX has four registers for indirect addressing. These registers are:

- INDF0 and FSR0
- INDF1 and FSR1

Registers INDF0 and INDF1 are not physically implemented. Reading or writing to these registers activates indirect addressing, with the value in the corresponding FSR register being the address of the data. The FSR is an 8-bit register and allows addressing anywhere in the 256-byte data memory address range. For banked memory, the bank of memory accessed is specified by the value in the BSR.

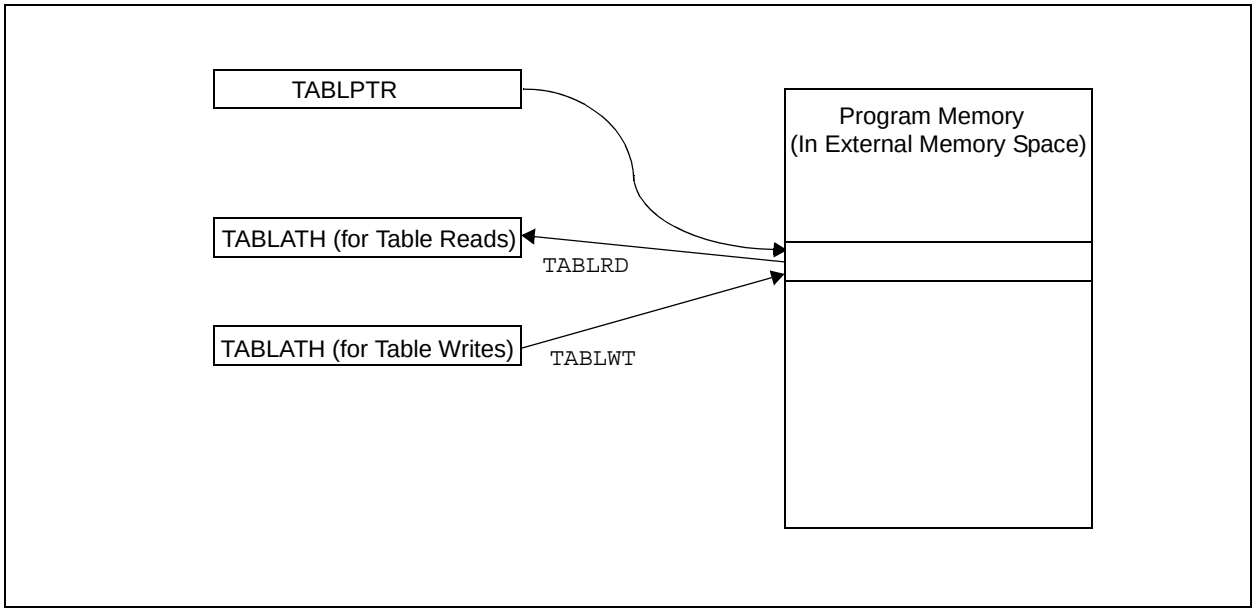
If file INDF0 (or INDF1) itself is read indirectly via an FSR, all '0's are read (Zero bit is set). Similarly, if INDF0 (or INDF1) is written to indirectly, the operation will be equivalent to a NOP, and the status bits are not affected.

8.4 Operation with External Memory Interface

When the table reads/writes are accessing external memory (via the external system interface bus), the table latch for the table reads is different from the table latch for the table writes (see Figure 8-9).

This means that you cannot do a TABLRD instruction, and use the values that were loaded into the table latches for a TABLWT instruction. Any table write sequence should use both the TLWT and then the TABLWT instructions.

FIGURE 8-9: ACCESSING EXTERNAL MEMORY WITH TABLRD AND TABLWT INSTRUCTIONS



9.0 HARDWARE MULTIPLIER

All PIC17C7XX devices have an 8 x 8 hardware multiplier included in the ALU of the device. By making the multiply a hardware operation, it completes in a single instruction cycle. This is an unsigned multiply that gives a 16-bit result. The result is stored into the 16-bit Product register (PRODH:PRODL). The multiplier does not affect any flags in the ALUSTA register.

Making the 8 x 8 multiplier execute in a single cycle gives the following advantages:

- Higher computational throughput
- Reduces code size requirements for multiply algorithms

The performance increase allows the device to be used in applications previously reserved for Digital Signal Processors.

Table 9-1 shows a performance comparison between PIC17CXXX devices using the single cycle hardware multiply and performing the same function without the hardware multiply.

Example 9-1 shows the sequence to do an 8 x 8 unsigned multiply. Only one instruction is required when one argument of the multiply is already loaded in the WREG register.

Example 9-2 shows the sequence to do an 8 x 8 signed multiply. To account for the sign bits of the arguments, each argument's most significant bit (MSb) is tested and the appropriate subtractions are done.

EXAMPLE 9-1: 8 x 8 UNSIGNED MULTIPLY ROUTINE

```
MOVFP    ARG1, WREG  ;
MULWF    ARG2        ; ARG1 * ARG2 ->
                     ;   PRODH:PRODL
```

EXAMPLE 9-2: 8 x 8 SIGNED MULTIPLY ROUTINE

```
MOVFP    ARG1, WREG
MULWF    ARG2        ; ARG1 * ARG2 ->
                     ;   PRODH:PRODL
BTFSC    ARG2, SB    ; Test Sign Bit
SUBWF    PRODH, F    ; PRODH = PRODH
                     ;         - ARG1
MOVFP    ARG2, WREG
BTFSC    ARG1, SB    ; Test Sign Bit
SUBWF    PRODH, F    ; PRODH = PRODH
                     ;         - ARG2
```

TABLE 9-1: PERFORMANCE COMPARISON

Routine	Multiply Method	Program Memory (Words)	Cycles (Max)	Time		
				@ 33 MHz	@ 16 MHz	@ 8 MHz
8 x 8 unsigned	Without hardware multiply	13	69	8.364 μs	17.25 μs	34.50 μs
	Hardware multiply	1	1	0.121 μs	0.25 μs	0.50 μs
8 x 8 signed	Without hardware multiply	—	—	—	—	—
	Hardware multiply	6	6	0.727 μs	1.50 μs	3.0 μs
16 x 16 unsigned	Without hardware multiply	21	242	29.333 μs	60.50 μs	121.0 μs
	Hardware multiply	24	24	2.91 μs	6.0 μs	12.0 μs
16 x 16 signed	Without hardware multiply	52	254	30.788 μs	63.50 μs	127.0 μs
	Hardware multiply	36	36	4.36 μs	9.0 μs	18.0 μs

10.2 PORTB and DDRB Registers

PORTB is an 8-bit wide, bi-directional port. The corresponding data direction register is DDRB. A '1' in DDRB configures the corresponding port pin as an input. A '0' in the DDRB register configures the corresponding port pin as an output. Reading PORTB reads the status of the pins, whereas writing to PORTB will write to the port latch.

Each of the PORTB pins has a weak internal pull-up. A single control bit can turn on all the pull-ups. This is done by clearing the $\overline{\text{RBPU}}$ (PORTA<7>) bit. The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are enabled on any RESET.

PORTB also has an interrupt-on-change feature. Only pins configured as inputs can cause this interrupt to occur (i.e., any RB7:RB0 pin configured as an output is excluded from the interrupt-on-change comparison). The input pins (of RB7:RB0) are compared with the value in the PORTB data latch. The “mismatch” outputs of RB7:RB0 are OR’d together to set the PORTB Interrupt Flag bit, RBIF (PIR1<7>).

This interrupt can wake the device from SLEEP. The user, in the Interrupt Service Routine, can clear the interrupt by:

a) Read-Write PORTB (such as: `MOVPF PORTB, PORTB`). This will end the mismatch condition.

b) Then, clear the RBIF bit.

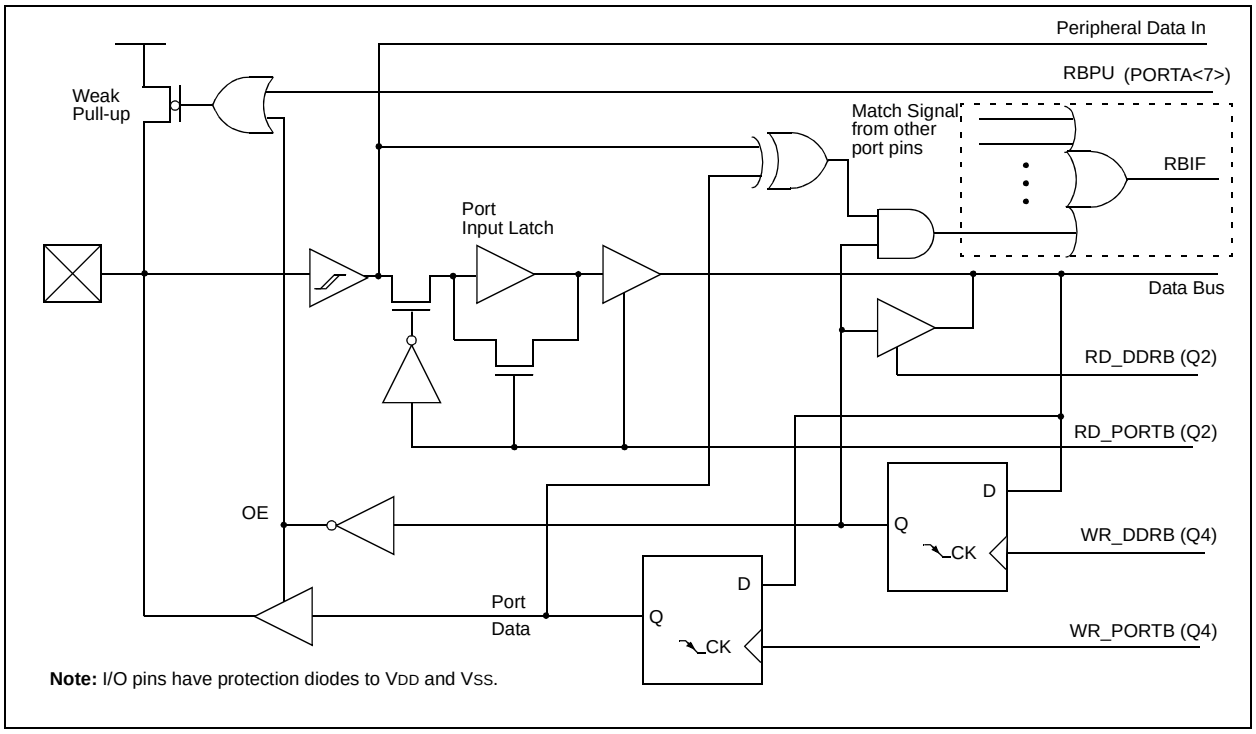
A mismatch condition will continue to set the RBIF bit. Reading, then writing PORTB, will end the mismatch condition and allow the RBIF bit to be cleared.

This interrupt-on-mismatch feature, together with software configurable pull-ups on this port, allows easy interface to a keypad and makes it possible for wake-up on key depression. For an example, refer to Application Note AN552, “Implementing Wake-up on Keystroke.”

The interrupt-on-change feature is recommended for wake-up on operations, where PORTB is only used for the interrupt-on-change feature and key depression operations.

Note:	On a device RESET, the RBIF bit is indeterminate, since the value in the latch may be different than the pin.

FIGURE 10-5: BLOCK DIAGRAM OF RB5:RB4 AND RB1:RB0 PORT PINS



13.2 Timer3

Timer3 is a 16-bit timer consisting of the TMR3H and TMR3L registers. TMR3H is the high byte of the timer and TMR3L is the low byte. This timer has an associated 16-bit period register (PR3H/CA1H:PR3L/CA1L). This period register can be software configured to be a another 16-bit capture register.

When the TMR3CS bit (TCON1<2>) is clear, the timer increments every instruction cycle (FOSC/4). When TMR3CS is set, the counter increments on every falling edge of the RB5/TCLK3 pin. In either mode, the TMR3ON bit must be set for the timer/counter to increment. When TMR3ON is clear, the timer will not increment or set flag bit TMR3IF.

Timer3 has two modes of operation, depending on the CA1/PR3 bit (TCON2<3>). These modes are:

- Three capture and one period register mode
- Four capture register mode

The PIC17C7XX has up to four 16-bit capture registers that capture the 16-bit value of TMR3 when events are detected on capture pins. There are four capture pins (RB0/CAP1, RB1/CAP2, RG4/CAP3, and RE3/CAP4), one for each capture register pair. The capture pins are multiplexed with the I/O pins. An event can be:

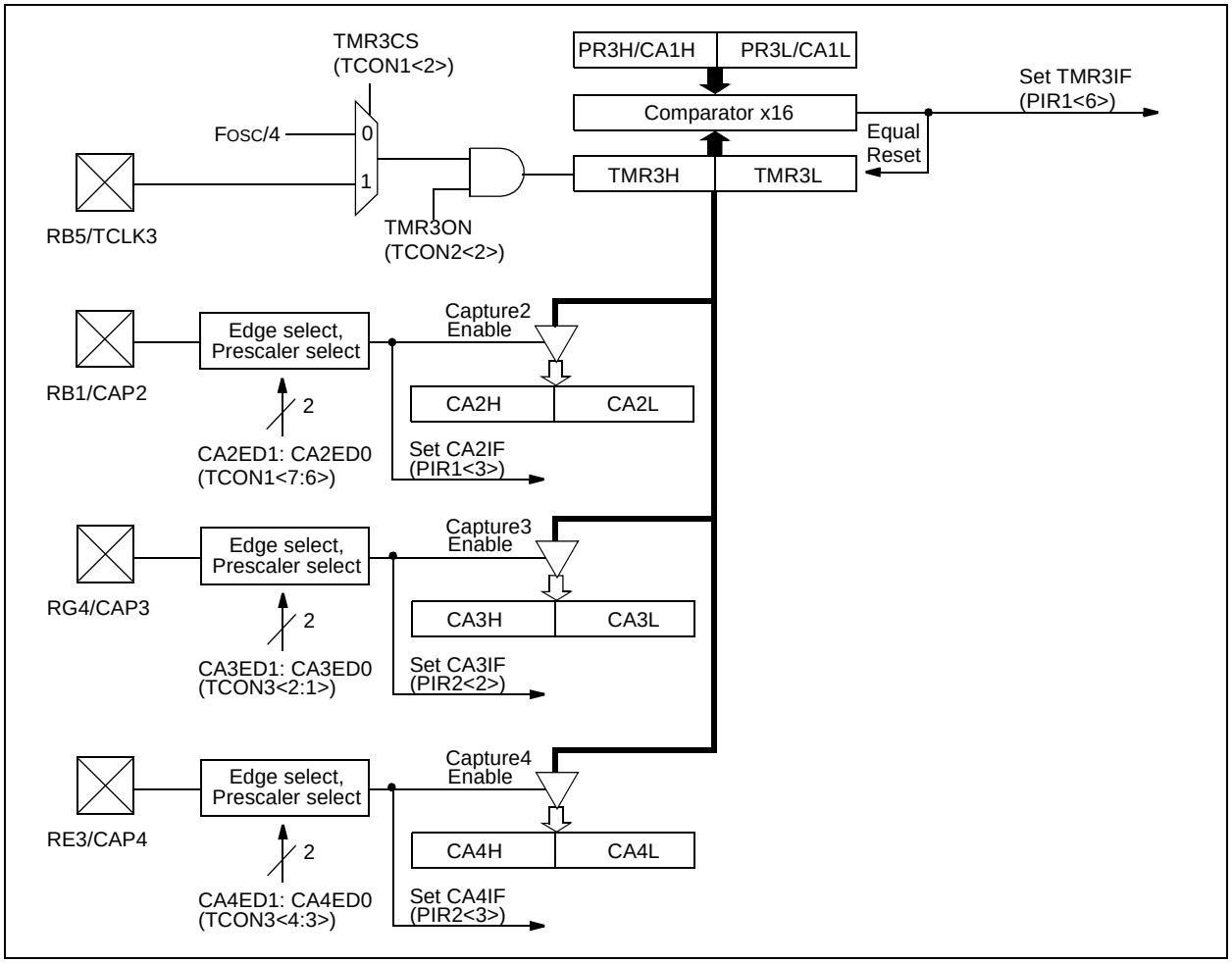
- A rising edge
- A falling edge
- Every 4th rising edge
- Every 16th rising edge

Each 16-bit capture register has an interrupt flag associated with it. The flag is set when a capture is made. The capture modules are truly part of the Timer3 block. Figure 13-5 and Figure 13-6 show the block diagrams for the two modes of operation.

13.2.1 THREE CAPTURE AND ONE PERIOD REGISTER MODE

In this mode, registers PR3H/CA1H and PR3L/CA1L constitute a 16-bit period register. A block diagram is shown in Figure 13-5. The timer increments until it equals the period register and then resets to 0000h on the next timer clock. TMR3 Interrupt Flag bit (TMR3IF) is set at this point. This interrupt can be disabled by clearing the TMR3 Interrupt Enable bit (TMR3IE). TMR3IF must be cleared in software.

FIGURE 13-5: TIMER3 WITH THREE CAPTURE AND ONE PERIOD REGISTER BLOCK DIAGRAM



14.4 USART Synchronous Slave Mode

The Synchronous Slave mode differs from the Master mode, in the fact that the shift clock is supplied externally at the TX/CK pin (instead of being supplied internally in the Master mode). This allows the device to transfer or receive data in the SLEEP mode. The Slave mode is entered by clearing the CSRC (TXSTA<7>) bit.

14.4.1 USART SYNCHRONOUS SLAVE TRANSMIT

The operation of the SYNC Master and Slave modes are identical except in the case of the SLEEP mode.

If two words are written to TXREG and then the SLEEP instruction executes, the following will occur. The first word will immediately transfer to the TSR and will transmit as the shift clock is supplied. The second word will remain in TXREG. TXIF will not be set. When the first word has been shifted out of TSR, TXREG will transfer the second word to the TSR and the TXIF flag will now be set. If TXIE is enabled, the interrupt will wake the chip from SLEEP and if the global interrupt is enabled, then the program will branch to the interrupt vector (0020h).

Steps to follow when setting up a Synchronous Slave Transmission:

1. Enable the synchronous slave serial port by setting the SYNC and SPEN bits and clearing the CSRC bit.
2. Clear the CREN bit.
3. If interrupts are desired, then set the TXIE bit.
4. If 9-bit transmission is desired, then set the TX9 bit.
5. If 9-bit transmission is selected, the ninth bit should be loaded in TX9D.
6. Start transmission by loading data to TXREG.
7. Enable the transmission by setting TXEN.

Writing the transmit data to the TXREG, then enabling the transmit (setting TXEN), allows transmission to start sooner than doing these two events in the reverse order.

Note:	To terminate a transmission, either clear the SPEN bit, or the TXEN bit. This will reset the transmit logic, so that it will be in the proper state when transmit is re-enabled.

14.4.2 USART SYNCHRONOUS SLAVE RECEPTION

Operation of the Synchronous Master and Slave modes are identical except in the case of the SLEEP mode. Also, SREN is a “don't care” in Slave mode.

If receive is enabled (CREN) prior to the SLEEP instruction, then a word may be received during SLEEP. On completely receiving the word, the RSR will transfer the data to RCREG (setting RCIF) and if the RCIE bit is set, the interrupt generated will wake the chip from SLEEP. If the global interrupt is enabled, the program will branch to the interrupt vector (0020h).

Steps to follow when setting up a Synchronous Slave Reception:

1. Enable the synchronous master serial port by setting the SYNC and SPEN bits and clearing the CSRC bit.
2. If interrupts are desired, then set the RCIE bit.
3. If 9-bit reception is desired, then set the RX9 bit.
4. To enable reception, set the CREN bit.
5. The RCIF bit will be set when reception is complete and an interrupt will be generated if the RCIE bit was set.
6. Read RCSTA to get the ninth bit (if enabled) and determine if any error occurred during reception.
7. Read the 8-bit received data by reading RCREG.
8. If any error occurred, clear the error by clearing the CREN bit.

Note:	To abort reception, either clear the SPEN bit, or the CREN bit (when in Continuous Receive mode). This will reset the receive logic, so that it will be in the proper state when receive is re-enabled.

15.1.7 SLEEP OPERATION

In Master mode, all module clocks are halted, and the transmission/reception will remain in that state until the device wakes from SLEEP. After the device returns to normal mode, the module will continue to transmit/ receive data.

In Slave mode, the SPI transmit/receive shift register operates asynchronously to the device. This allows the device to be placed in SLEEP mode and data to be shifted into the SPI transmit/receive shift register. When all 8-bits have been received, the MSSP interrupt flag bit will be set and if enabled, will wake the device from SLEEP.

15.1.8 EFFECTS OF A RESET

A RESET disables the MSSP module and terminates the current transfer.

TABLE 15-1: REGISTERS ASSOCIATED WITH SPI OPERATION

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	POR, BOR	$\overline{\text{MCLR}}$, WDT
07h, Unbanked	INTSTA	PEIF	T0CKIF	T0IF	INTF	PEIE	T0CKIE	T0IE	INTE	0000 0000	0000 0000
10h, Bank 4	PIR2	SSPIF	BCLIF	ADIF	—	CA4IF	CA3IF	TX2IF	RC2IF	000- 0010	000- 0010
11h, Bank 4	PIE2	SSPIE	BCLIE	ADIE	—	CA4IE	CA3IE	TX2IE	RC2IE	000- 0000	000- 0000
14h, Bank 6	SSPBUF	Synchronous Serial Port Receive Buffer/Transmit Register								xxxx xxxx	uuuu uuuu
11h, Bank 6	SSPCON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000	0000 0000
13h, Bank 6	SSPSTAT	SMP	CKE	D/$\overline{\text{A}}$	P	S	R/$\overline{\text{W}}$	UA	BF	0000 0000	0000 0000

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by the SSP in SPI mode.

EXAMPLE 15-2: INTERFACING TO A 24LC01B SERIAL EEPROM (USING MPLAB C17)

```
// Writes the byte data to 24LC01B at the specified address
void ByteWrite(static unsigned char address, static unsigned char data)
{
    StartI2C();                      // Send start bit
    IdleI2C();                       // Wait for idle condition
    WriteI2C(CONTROL);               // Send control byte
    IdleI2C();                       // Wait for idle condition
    if (!SSPCON2bits.ACKSTAT)        // If 24LC01B ACKs
    {
        WriteI2C(address);           // Send control byte
        IdleI2C();                   // Wait for idle condition

        if (!SSPCON2bits.ACKSTAT)    // If 24LC01B ACKs
            WriteI2C(data);          // Send data
    }
    IdleI2C();                       // Wait for idle condition
    StopI2C();                       // Send stop bit
    IdleI2C();                       // Wait for idle condition
    return;
}


// Reads a byte of data from 24LC01B at the specified address
unsigned char ByteRead(static unsigned char address)
{
    StartI2C();                      // Send start bit
    IdleI2C();                       // Wait for idle condition
    WriteI2C(CONTROL);               // Send control byte
    IdleI2C();                       // Wait for idle condition
    if (!SSPCON2bits.ACKSTAT)        // If the 24LC01B ACKs
    {
        WriteI2C(address);           // Send address
        IdleI2C();                   // Wait for idle condition
        if (!SSPCON2bits.ACKSTAT)    // If the 24LC01B ACKs
        {
            RestartI2C();            // Send restart
            IdleI2C();               // Wait for idle condition
            WriteI2C(CONTROL+1);     // Send control byte with R/W set
            IdleI2C();               // Wait for idle condition
            if (!SSPCON2bits.ACKSTAT)    // If the 24LC01B ACKs
            {
                getcI2C();           // Read a byte of data from 24LC01B
                IdleI2C();           // Wait for idle condition
                NotAckI2C();         // Send a NACK to 24LC01B
                IdleI2C();           // Wait for idle condition
                StopI2C();           // Send stop bit
                IdleI2C();           // Wait for idle condition
             }
        }
    }
    return(SSPBUF);
}
```

EXAMPLE 15-2: INTERFACING TO A 24LC01B SERIAL EEPROM (USING MPLAB C17)

```
void ACKPoll(void)
{
        StartI2C();                     // Send start bit
        IdleI2C();                      // Wait for idle condition
        WriteI2C(CONTROL);              // Send control byte
        IdleI2C();                      // Wait for idle condition
        // Poll the ACK bit coming from the 24LC01B
        // Loop as long as the 24LC01B NACKs
        while (SSPCON2bits.ACKSTAT)
        {
                RestartI2C();           // Send a restart bit
                IdleI2C();              // Wait for idle condition
                WriteI2C(CONTROL);      // Send control byte
                IdleI2C();              // Wait for idle condition
        }
        IdleI2C();                      // Wait for idle condition
        StopI2C();                      // Send stop bit
        IdleI2C();                      // Wait for idle condition
        return;
}
```

16.4.1 A/D RESULT REGISTERS

The ADRESH:ADRESL register pair is the location where the 10-bit A/D result is loaded at the completion of the A/D conversion. This register pair is 16-bits wide. The A/D module gives the flexibility to left or right justify the 10-bit result in the 16-bit result register. The A/D Format Select bit (ADFM) controls this justification. Figure 16-6 shows the operation of the A/D result justification. The extra bits are loaded with '0's'. When an A/D result will not overwrite these locations (A/D disable), these registers may be used as two general purpose 8-bit registers.

16.5 A/D Operation During SLEEP

The A/D module can operate during SLEEP mode. This requires that the A/D clock source be set to RC (ADCS1:ADCS0 = 11). When the RC clock source is selected, the A/D module waits one instruction cycle before starting the conversion. This allows the SLEEP instruction to be executed, which eliminates all digital switching noise from the conversion. When the conversion is completed, the GO/DONE bit will be cleared, and the result loaded into the ADRES register. If the A/D interrupt is enabled, the device will wake-up from SLEEP. If the A/D interrupt is not enabled, the A/D module will then be turned off, although the ADON bit will remain set.

When the A/D clock source is another clock option (not RC), a SLEEP instruction will cause the present conversion to be aborted and the A/D module to be turned off, though the ADON bit will remain set.

Turning off the A/D places the A/D module in its lowest current consumption state.

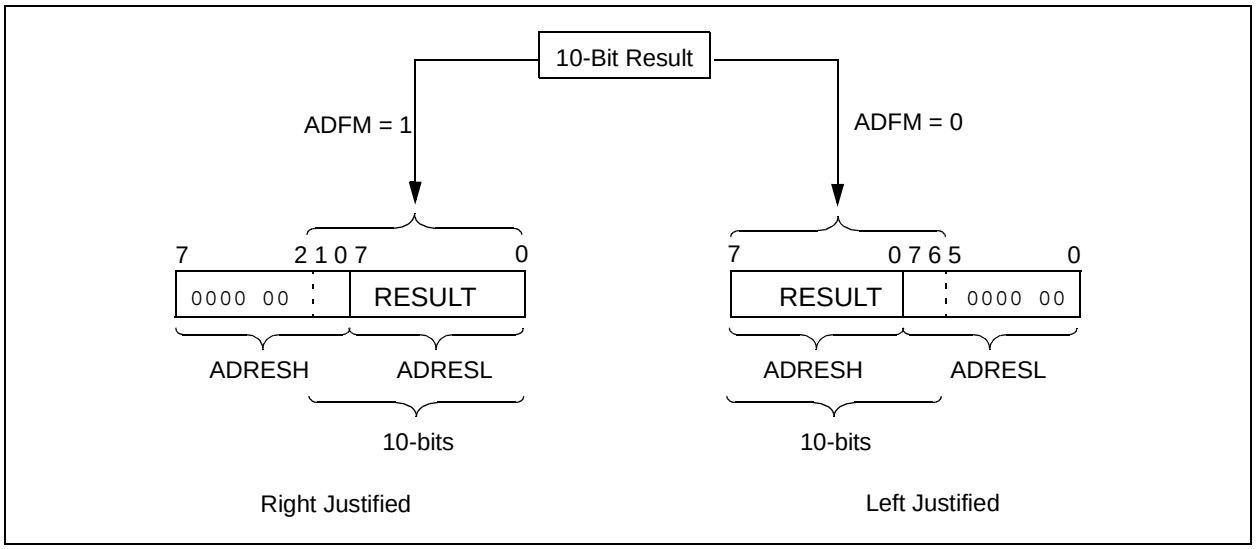
Note:	For the A/D module to operate in SLEEP, the A/D clock source must be set to RC (ADCS1:ADCS0 = 11). To allow the conversion to occur during SLEEP, ensure the SLEEP instruction immediately follows the instruction that sets the GO/DONE bit.

16.6 Effects of a RESET

A device RESET forces all registers to their RESET state. This forces the A/D module to be turned off, and any conversion is aborted.

The value that is in the ADRESH:ADRESL registers is not modified for a Power-on Reset. The ADRESH:ADRESL registers will contain unknown data after a Power-on Reset.

FIGURE 16-6: A/D RESULT JUSTIFICATION



CLRWDT **Clear Watchdog Timer**

Syntax: [*label*] CLRWDT

Operands: None

Operation: 00h → WDT
0 → $\overline{\text{WDT}}$ postscaler,
1 → $\overline{\text{TO}}$
1 → $\overline{\text{PD}}$

Status Affected: $\overline{\text{TO}}$, $\overline{\text{PD}}$

Encoding:

0000	0000	0000	0100

Description: CLRWDT instruction resets the Watchdog Timer. It also resets the postscaler of the WDT. Status bits $\overline{\text{TO}}$ and $\overline{\text{PD}}$ are set.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	No operation	Process Data	No operation

Example: CLRWDT

Before Instruction
WDT counter = ?

After Instruction
WDT counter = 0x00
$\overline{\text{WDT}}$ Postscaler = 0
$\overline{\text{TO}}$ = 1
$\overline{\text{PD}}$ = 1

COMF **Complement f**

Syntax: [*label*] COMF f,d

Operands: $0 \le f \le 255$
$d \in [0,1]$

Operation: $(\bar{f}) \rightarrow$ (dest)

Status Affected: Z

Encoding:

0001	001d	ffff	ffff

Description: The contents of register 'f' are complemented. If 'd' is 0 the result is stored in WREG. If 'd' is 1 the result is stored back in register 'f'.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

Example: COMF REG1,0

Before Instruction
REG1 = 0x13

After Instruction
REG1 = 0x13
WREG = 0xEC

TABLWT **Table Write**

Example1: `TABLWT 1, 1, REG`

Before Instruction

REG	=	0x53
TBLATH	=	0xAA
TBLATL	=	0x55
TBLPTR	=	0xA356
MEMORY(TBLPTR)	=	0xFFFF

After Instruction (table write completion)

REG	=	0x53
TBLATH	=	0x53
TBLATL	=	0x55
TBLPTR	=	0xA357
MEMORY(TBLPTR - 1)	=	0x5355

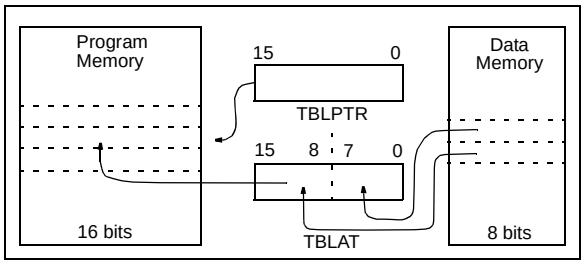
Example 2: `TABLWT 0, 0, REG`

Before Instruction

REG	=	0x53
TBLATH	=	0xAA
TBLATL	=	0x55
TBLPTR	=	0xA356
MEMORY(TBLPTR)	=	0xFFFF

After Instruction (table write completion)

REG	=	0x53
TBLATH	=	0xAA
TBLATL	=	0x53
TBLPTR	=	0xA356
MEMORY(TBLPTR)	=	0xAA53



TLRD **Table Latch Read**

Syntax: [*label*] TLRD t,f

Operands: $0 \le f \le 255$
$t \in [0,1]$

Operation: If t = 0,
TBLATL → f;
If t = 1,
TBLATH → f

Status Affected: None

Encoding:

1010	00tx	ffff	ffff

Description: Read data from 16-bit table latch (TBLAT) into file register 'f'. Table Latch is unaffected.
If t = 1; high byte is read
If t = 0; low byte is read
This instruction is used in conjunction with `TABLRD` to transfer data from program memory to data memory.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register TBLATH or TBLATL	Process Data	Write register 'f'

Example: `TLRD t, RAM`

Before Instruction

t	=	0	
RAM	=	?	
TBLAT	=	0x00AF	(TBLATH = 0x00) (TBLATL = 0xAF)

After Instruction

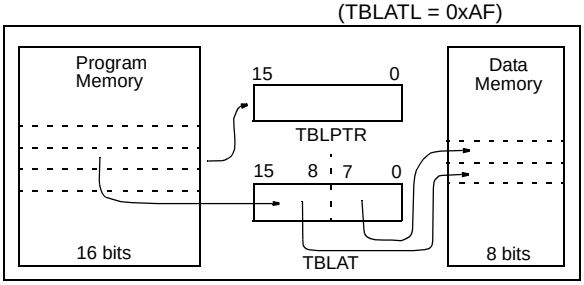
RAM	=	0xAF	
TBLAT	=	0x00AF	(TBLATH = 0x00) (TBLATL = 0xAF)

Before Instruction

t	=	1	
RAM	=	?	
TBLAT	=	0x00AF	(TBLATH = 0x00) (TBLATL = 0xAF)

After Instruction

RAM	=	0x00	
TBLAT	=	0x00AF	(TBLATH = 0x00) (TBLATL = 0xAF)



TLWT **Table Latch Write**

Syntax: [*label*] TLWT t,f

Operands: $0 \le f \le 255$
$t \in [0,1]$

Operation: If t = 0,
f → TBLATL;
If t = 1,
f → TBLATH

Status Affected: None

Encoding:

1010	01tx	ffff	ffff

Description: Data from file register 'f' is written into the 16-bit table latch (TBLAT).
If t = 1; high byte is written
If t = 0; low byte is written
This instruction is used in conjunction with TABLWT to transfer data from data memory to program memory.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write register TBLATH or TBLATL

Example: TLWT t, RAM

Before Instruction
t = 0
RAM = 0xB7
TBLAT = 0x0000 (TBLATH = 0x00) (TBLATL = 0x00)

After Instruction
RAM = 0xB7
TBLAT = 0x00B7 (TBLATH = 0x00) (TBLATL = 0xB7)

Before Instruction
t = 1
RAM = 0xB7
TBLAT = 0x0000 (TBLATH = 0x00) (TBLATL = 0x00)

After Instruction
RAM = 0xB7
TBLAT = 0xB700 (TBLATH = 0xB7) (TBLATL = 0x00)

TSTFSZ **Test f, skip if 0**

Syntax: [*label*] TSTFSZ f

Operands: $0 \le f \le 255$

Operation: skip if f = 0

Status Affected: None

Encoding:

0011	0011	ffff	ffff

Description: If 'f' = 0, the next instruction, fetched during the current instruction execution, is discarded and a NOP is executed, making this a two-cycle instruction.

Words: 1

Cycles: 1 (2)

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	No operation

If skip:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation

Example:
```
HERE    TSTFSZ  CNT
NZERO     :
ZERO      :
```

Before Instruction
PC = Address (HERE)

After Instruction
If CNT = 0x00,
PC = Address (ZERO)
If CNT ¼ 0x00,
PC = Address (NZERO)

FIGURE 20-13: SPI MASTER MODE TIMING (CKE = 0)

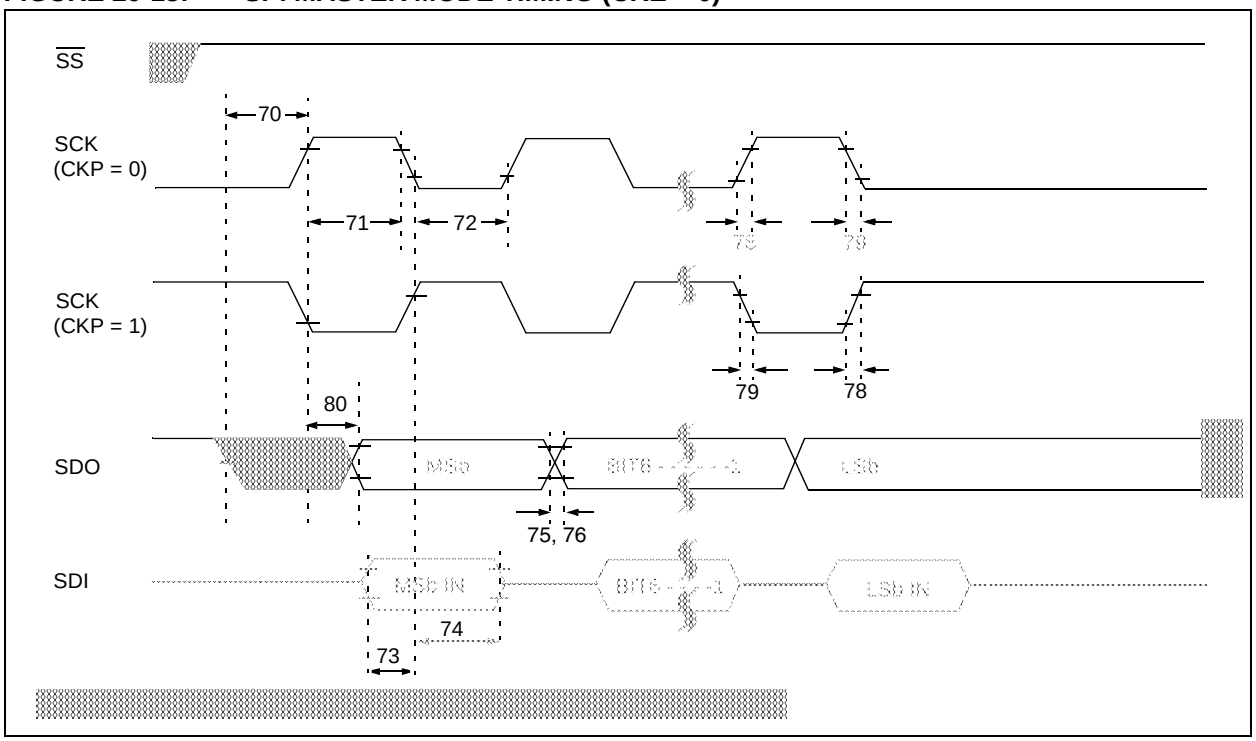


TABLE 20-8: SPI MODE REQUIREMENTS (MASTER MODE, CKE = 0)

Param. No.	Symbol	Characteristic		Min	Typ†	Max	Units	Conditions
70	TssL2scH, TssL2scL	$\overline{SS}\downarrow$ to SCK$\downarrow$ or SCK$\uparrow$ input		Tcy	—	—	ns	
71	TscH	SCK input high time (Slave mode)	Continuous	1.25TCY + 30	—	—	ns	
71A			Single Byte	40	—	—	ns	**(Note 1)**
72	TscL	SCK input low time (Slave mode)	Continuous	1.25TCY + 30	—	—	ns	
72A			Single Byte	40	—	—	ns	**(Note 1)**
73	TdiV2scH, TdiV2scL	Setup time of SDI data input to SCK edge		100	—	—	ns	
73A	TB2B	Last clock edge of Byte1 to the 1st clock edge of Byte2		1.5TCY + 40	—	—	ns	**(Note 1)**
74	TscH2diL, TscL2diL	Hold time of SDI data input to SCK edge		100	—	—	ns	
75	TdoR	SDO data output rise time		—	10	25	ns	
76	TdoF	SDO data output fall time		—	10	25	ns	
78	TscR	SCK output rise time (Master mode)		—	10	25	ns	
79	TscF	SCK output fall time (Master mode)		—	10	25	ns	
80	TscH2doV, TscL2doV	SDO data output valid after SCK edge		—	—	50	ns	

† Data in "Typ" column is at 5V, 25°C unless otherwise stated.

Note 1: Specification 73A is only required if specifications 71A and 72A are used.

FIGURE 21-13: TYPICAL AND MAXIMUM ΔIPD vs. VDD (SLEEP MODE, WDT ENABLED, -40°C to +125°C)

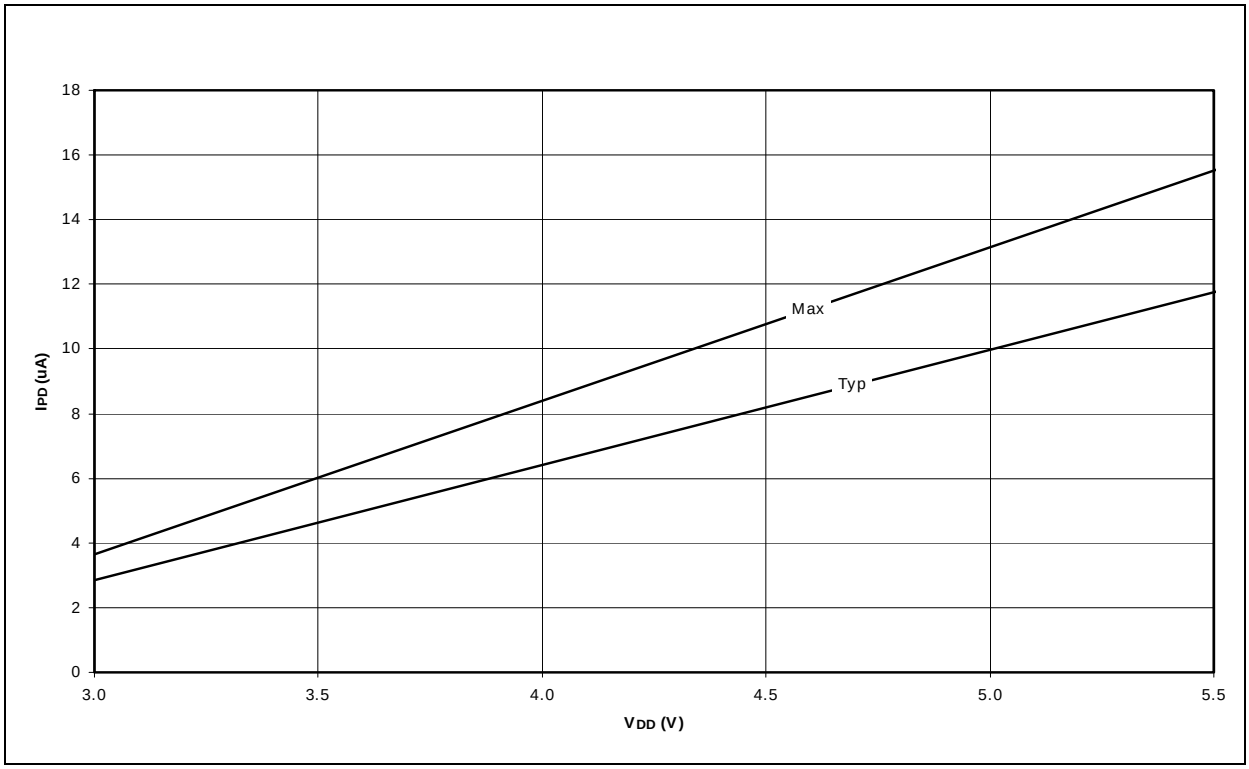


FIGURE 21-14: TYPICAL AND MAXIMUM ΔIRBPU vs. VDD (MEASURED PER INPUT PIN, -40°C TO +125°C)

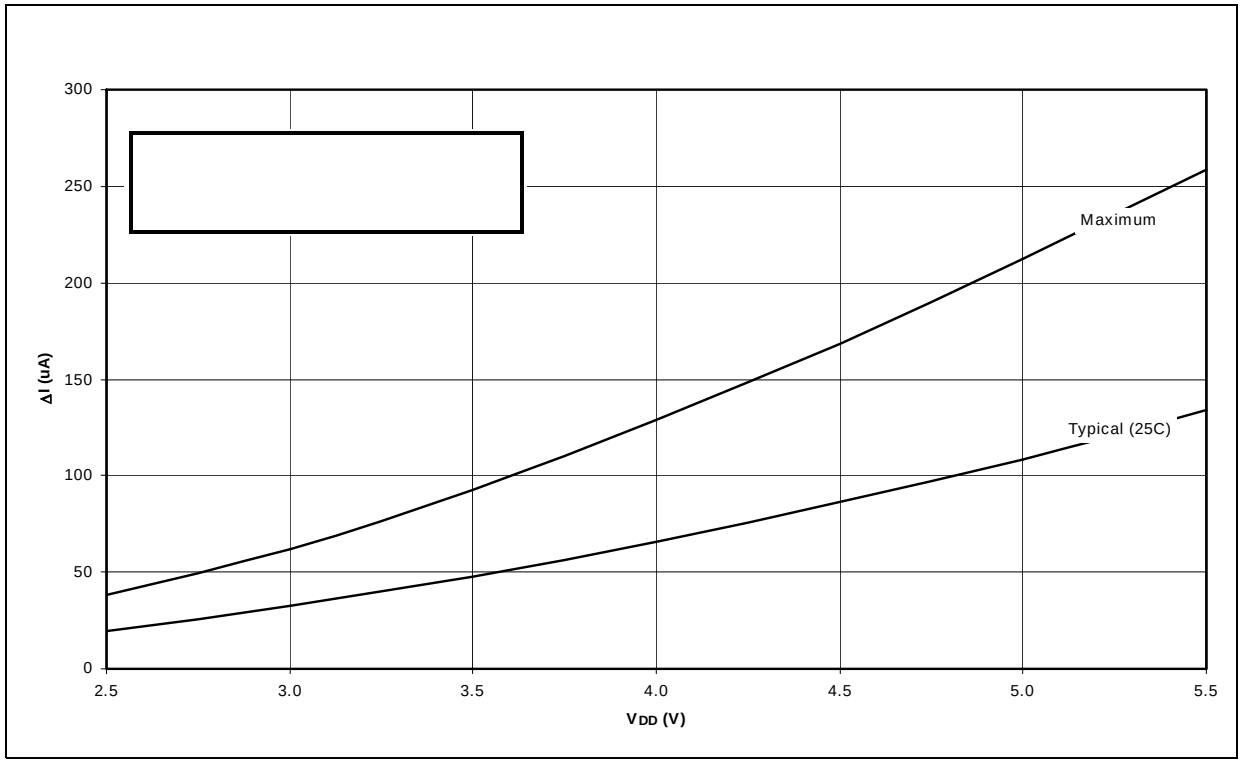
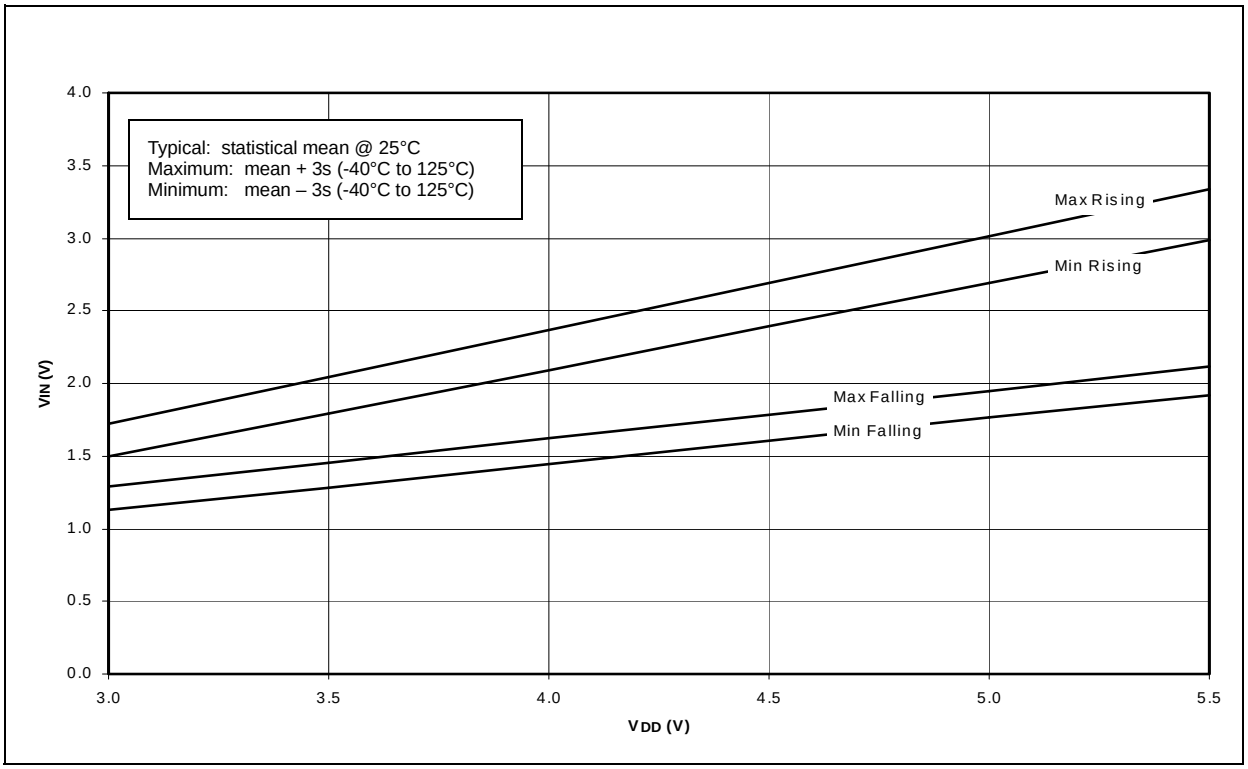


FIGURE 21-23: MAXIMUM AND MINIMUM VIN vs. VDD (I^2C Input, -40°C to +125°C)



APPENDIX C: WHAT'S NEW

This is a new Data Sheet for the Following Devices:

- PIC17C752
- PIC17C756A
- PIC17C762
- PIC17C766

This Data Sheet is based on the PIC17C75X Data Sheet (DS30246A).

APPENDIX D: WHAT'S CHANGED

Clarified the TAD vs. device maximum operating frequency tables in Section 16.2.

Added device characteristic graphs and charts in Section 21.

Removed the "Preliminary" status from the entire document.

Revision C (January 2013)

Added a note to each package outline drawing.