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Details

Product Status	Not For New Designs
Core Processor	e200z6
Core Size	32-Bit Single-Core
Speed	80MHz
Connectivity	CANbus, EBI/EMI, Ethernet, SCI, SPI
Peripherals	DMA, POR, PWM, WDT
Number of I/O	192
Program Memory Size	1MB (1M x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	64K x 8
Voltage - Supply (Vcc/Vdd)	1.35V ~ 1.65V
Data Converters	A/D 34x12b
Oscillator Type	External
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	208-BGA
Supplier Device Package	208-BGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/spc5534mvm80r

3 Electrical Characteristics

This section contains detailed information on power considerations, DC/AC electrical characteristics, and AC timing specifications for the MCU.

3.1 Maximum Ratings

Table 2. Absolute Maximum Ratings ¹

Spec	Characteristic	Symbol	Min.	Max.	Unit
1	1.5 V core supply voltage ²	V _{DD}	−0.3	1.7	V
2	Flash program/erase voltage	V _{PP}	−0.3	6.5	V
4	Flash read voltage	V _{FLASH}	−0.3	4.6	V
5	SRAM standby voltage	V _{STBY}	−0.3	1.7	V
6	Clock synthesizer voltage	V _{DDSYN}	−0.3	4.6	V
7	3.3 V I/O buffer voltage	V _{DD33}	−0.3	4.6	V
8	Voltage regulator control input voltage	V _{RC33}	−0.3	4.6	V
9	Analog supply voltage (reference to V _{SSA})	V _{DDA}	−0.3	5.5	V
10	I/O supply voltage (fast I/O pads) ³	V _{DDE}	−0.3	4.6	V
11	I/O supply voltage (slow and medium I/O pads) ³	V _{DDEH}	−0.3	6.5	V
12	DC input voltage ⁴ V _{DDEH} powered I/O pads V _{DDE} powered I/O pads	V _{IN}	−1.0 ⁵ −1.0 ⁵	6.5 ⁶ 4.6 ⁷	V
13	Analog reference high voltage (reference to V _{RL})	V _{RH}	−0.3	5.5	V
14	V _{SS} to V _{SSA} differential voltage	V _{SS} − V _{SSA}	−0.1	0.1	V
15	V _{DD} to V _{DDA} differential voltage	V _{DD} − V _{DDA}	−V _{DDA}	V _{DD}	V
16	V _{REF} differential voltage	V _{RH} − V _{RL}	−0.3	5.5	V
17	V _{RH} to V _{DDA} differential voltage	V _{RH} − V _{DDA}	−5.5	5.5	V
18	V _{RL} to V _{SSA} differential voltage	V _{RL} − V _{SSA}	−0.3	0.3	V
19	V _{DDEH} to V _{DDA} differential voltage	V _{DDEH} − V _{DDA}	−V _{DDA}	V _{DDEH}	V
20	V _{DDF} to V _{DD} differential voltage	V _{DDF} − V _{DD}	−0.3	0.3	V
21	V _{RC33} to V _{DDSYN} differential voltage spec has been moved to Table 9 DC Electrical Specifications, Spec 43a.				
22	V _{SSSYN} to V _{SS} differential voltage	V _{SSSYN} − V _{SS}	−0.1	0.1	V
23	V _{RCVSS} to V _{SS} differential voltage	V _{RCVSS} − V _{SS}	−0.1	0.1	V
24	Maximum DC digital input current ⁸ (per pin, applies to all digital pins) ⁴	I _{MAXD}	−2	2	mA
25	Maximum DC analog input current ⁹ (per pin, applies to all analog pins)	I _{MAXA}	−3	3	mA
26	Maximum operating temperature range ¹⁰ Die junction temperature	T _J	T _L	150.0	°C
27	Storage temperature range	T _{STG}	−55.0	150.0	°C

⁶ The thermal characterization parameter indicates the temperature difference between the package top and the junction temperature per JEDEC JESD51-2.

3.2.1 General Notes for Specifications at Maximum Junction Temperature

An estimation of the device junction temperature, T_J , can be obtained from the equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

T_A = ambient temperature for the package ($^{\circ}\text{C}$)

$R_{\theta JA}$ = junction to ambient thermal resistance ($^{\circ}\text{C}/\text{W}$)

P_D = power dissipation in the package (W)

The thermal resistance values used are based on the JEDEC JESD51 series of standards to provide consistent values for estimations and comparisons. The difference between the values determined for the single-layer (1s) board compared to a four-layer board that has two signal layers, a power and a ground plane (2s2p), demonstrate that the effective thermal resistance is not a constant. The thermal resistance depends on the:

- Construction of the application board (number of planes)
- Effective size of the board which cools the component
- Quality of the thermal and electrical connections to the planes
- Power dissipated by adjacent components

Connect all the ground and power balls to the respective planes with one via per ball. Using fewer vias to connect the package to the planes reduces the thermal performance. Thinner planes also reduce the thermal performance. When the clearance between the vias leave the planes virtually disconnected, the thermal performance is also greatly reduced.

As a general rule, the value obtained on a single-layer board is within the normal range for the tightly packed printed circuit board. The value obtained on a board with the internal planes is usually within the normal range if the application board has:

- One oz. (35 micron nominal thickness) internal planes
- Components are well separated
- Overall power dissipation on the board is less than $0.02 \text{ W}/\text{cm}^2$

The thermal performance of any component depends on the power dissipation of the surrounding components. In addition, the ambient temperature varies widely within the application. For many natural convection and especially closed box applications, the board temperature at the perimeter (edge) of the package is approximately the same as the local air temperature near the device. Specifying the local ambient conditions explicitly as the board temperature provides a more precise description of the local ambient conditions that determine the temperature of the device.

At a known board temperature, the junction temperature is estimated using the following equation:

$$T_J = T_B + (R_{\theta JB} \times P_D)$$

where:

T_J = junction temperature ($^{\circ}\text{C}$)

T_B = board temperature at the package perimeter ($^{\circ}\text{C/W}$)

$R_{\theta JB}$ = junction-to-board thermal resistance ($^{\circ}\text{C/W}$) per JESD51-8

P_D = power dissipation in the package (W)

When the heat loss from the package case to the air does not factor into the calculation, an acceptable value for the junction temperature is predictable. Ensure the application board is similar to the thermal test condition, with the component soldered to a board with internal planes.

The thermal resistance is expressed as the sum of a junction-to-case thermal resistance plus a case-to-ambient thermal resistance:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$$

where:

$R_{\theta JA}$ = junction-to-ambient thermal resistance ($^{\circ}\text{C/W}$)

$R_{\theta JC}$ = junction-to-case thermal resistance ($^{\circ}\text{C/W}$)

$R_{\theta CA}$ = case-to-ambient thermal resistance ($^{\circ}\text{C/W}$)

$R_{\theta JC}$ is device related and is not affected by other factors. The thermal environment can be controlled to change the case-to-ambient thermal resistance, $R_{\theta CA}$. For example, change the air flow around the device, add a heat sink, change the mounting arrangement on the printed circuit board, or change the thermal dissipation on the printed circuit board surrounding the device. This description is most useful for packages with heat sinks where 90% of the heat flow is through the case to heat sink to ambient. For most packages, a better model is required.

A more accurate two-resistor thermal model can be constructed from the junction-to-board thermal resistance and the junction-to-case thermal resistance. The junction-to-case thermal resistance describes when using a heat sink or where a substantial amount of heat is dissipated from the top of the package. The junction-to-board thermal resistance describes the thermal performance when most of the heat is conducted to the printed circuit board. This model can be used to generate simple estimations and for computational fluid dynamics (CFD) thermal models.

To determine the junction temperature of the device in the application on a prototype board, use the thermal characterization parameter (Ψ_{JT}) to determine the junction temperature by measuring the temperature at the top center of the package case using the following equation:

$$T_J = T_T + (\Psi_{JT} \times P_D)$$

where:

T_T = thermocouple temperature on top of the package ($^{\circ}\text{C}$)

Ψ_{JT} = thermal characterization parameter ($^{\circ}\text{C/W}$)

P_D = power dissipation in the package (W)

Table 6. V_{RC} and POR Electrical Specifications (continued)

Spec	Characteristic		Symbol	Min.	Max.	Units
9	Absolute value of slew rate on power supply pins		—	—	50	V/ms
10	Required gain at Tj: $I_{DD} \div I_{VRCCTL}$ (@ $f_{sys} = f_{MAX}$) 6, 7, 8, 9	– 40° C 25° C 150° C	BETA ¹⁰	35 40 50	— — 500	— — —

¹ The internal POR signals are V_{POR15} , V_{POR33} , and V_{POR5} . On power up, assert \overline{RESET} before the internal POR negates. \overline{RESET} must remain asserted until the power supplies are within the operating conditions as specified in Table 9 DC Electrical Specifications. On power down, assert \overline{RESET} before any power supplies fall outside the operating conditions and until the internal POR asserts.

² V_{IL_S} (Table 9, Spec15) is guaranteed to scale with V_{DDEH6} down to V_{POR5} .

³ Supply full operating current for the 1.5 V supply when the 3.3 V supply reaches this range.

⁴ It is possible to reach the current limit during ramp up—do not treat this event as short circuit current.

⁵ At peak current for device.

⁶ Requires compliance with Freescale's recommended board requirements and transistor recommendations. Board signal traces/routing from the V_{RCCTL} package signal to the base of the external pass transistor and between the emitter of the pass transistor to the V_{DD} package signals must have a maximum of 100 nH inductance and minimal resistance (less than 1 Ω). V_{RCCTL} must have a nominal 1 μ F phase compensation capacitor to ground. V_{DD} must have a 20 μ F (nominal) bulk capacitor (greater than 4 μ F over all conditions, including lifetime). Place high-frequency bypass capacitors consisting of eight 0.01 μ F, two 0.1 μ F, and one 1 μ F capacitors around the package on the V_{DD} supply signals.

⁷ I_{VRCCTL} is measured at the following conditions: $V_{DD} = 1.35$ V, $V_{RC33} = 3.1$ V, $V_{VRCCTL} = 2.2$ V.

⁸ Refer to Table 1 for the maximum operating frequency.

⁹ Values are based on I_{DD} from high-use applications as explained in the I_{DD} Electrical Specification.

¹⁰ BETA represents the worst-case external transistor. It is measured on a per-part basis and calculated as $(I_{DD} \div I_{VRCCTL})$.

3.7 Power-Up/Down Sequencing

Power sequencing between the 1.5 V power supply and V_{DDSYN} or the \overline{RESET} power supplies is required if using an external 1.5 V power supply with V_{RC33} tied to ground (GND). To avoid power-sequencing, V_{RC33} must be powered up within the specified operating range, even if the on-chip voltage regulator controller is not used. Refer to Section 3.7.2, “Power-Up Sequence (VRC33 Grounded),” and Section 3.7.3, “Power-Down Sequence (VRC33 Grounded).”

Power sequencing requires that V_{DD33} must reach a certain voltage where the values are read as ones before the POR signal negates. Refer to Section 3.7.1, “Input Value of Pins During POR Dependent on V_{DD33} .”

Although power sequencing is not required between V_{RC33} and V_{DDSYN} during power up, V_{RC33} must not lead V_{DDSYN} by more than 600 mV or lag by more than 100 mV for the V_{RC} stage turn-on to operate within specification. Higher spikes in the emitter current of the pass transistor occur if V_{RC33} leads or lags V_{DDSYN} by more than these amounts. The value of that higher spike in current depends on the board power supply circuitry and the amount of board level capacitance.

Furthermore, when all of the PORs negate, the system clock starts to toggle, adding another large increase of the current consumed by V_{RC33} . If V_{RC33} lags V_{DDSYN} by more than 100 mV, the increase in current consumed can drop V_{DD} low enough to assert the 1.5 V POR again. Oscillations are possible when the 1.5 V POR asserts and stops the system clock, causing the voltage on V_{DD} to rise until the 1.5 V POR negates again. All oscillations stop when V_{RC33} is powered sufficiently.

Figure 2 shows an approximate interpolation of the I_{STBY} worst-case specification to estimate values at different voltages and temperatures. The vertical lines shown at 25 °C, 60 °C, and 150 °C in Figure 2 are the actual I_{DD_STBY} specifications (27d) listed in Table 9

Figure 2. fISTBY Worst-case Specifications

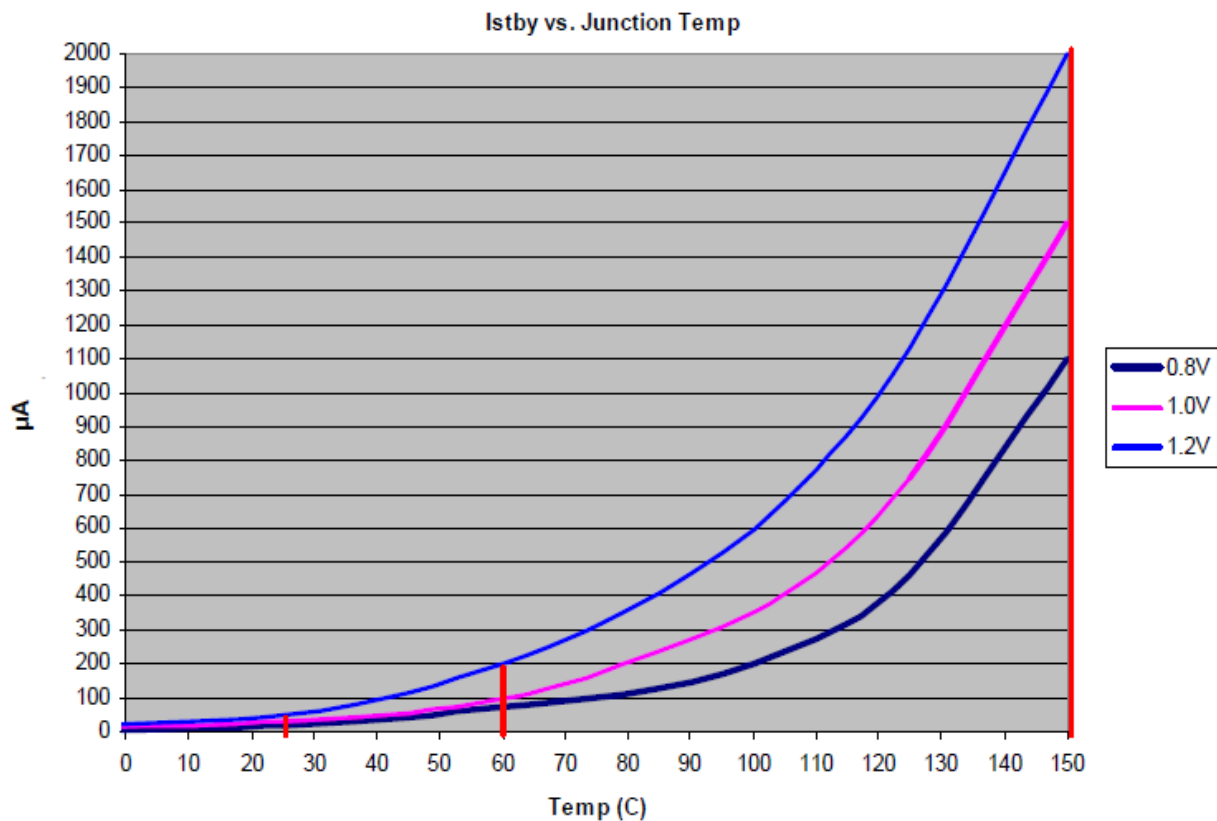


Table 12. FMPLL Electrical Specifications (continued) $(V_{DDSYN} = 3.0\text{--}3.6\text{ V}; V_{SS} = V_{SSSYN} = 0.0\text{ V}; T_A = T_L \text{ to } T_H)$

Spec	Characteristic	Symbol	Minimum	Maximum	Unit
19	CLKOUT period jitter, measured at f_{SYS} max: ^{13, 14} Peak-to-peak jitter (clock edge to clock edge) Long term jitter (averaged over a 2 ms interval)	C_{JITTER}	— —	5.0 0.01	% f_{CLKOUT}
20	Frequency modulation range limit ¹⁵ (do not exceed f_{SYS} maximum)	C_{MOD}	0.8	2.4	% f_{SYS}
21	ICO frequency $f_{ICO} = [f_{ref_crystal} \times (MFD + 4)] \div (PREDIV + 1)$ ¹⁶ $f_{ICO} = [f_{ref_ext} \times (MFD + 4)] \div (PREDIV + 1)$	f_{ICO}	48	80 ¹⁷	MHz
22	Predivider output frequency (to PLL)	f_{PREDIV}	4	20 ¹⁸	MHz

¹ Nominal crystal and external reference values are worst-case not more than 1%. The device operates correctly if the frequency remains within $\pm 5\%$ of the specification limit. This tolerance range allows for a slight frequency drift of the crystals over time. The designer must thoroughly understand the drift margin of the source clock.

² All internal registers retain data at 0 Hz.

³ Up to the maximum frequency rating of the device (refer to Table 1).

⁴ Loss of reference frequency is defined as the reference frequency detected internally, which transitions the PLL into self-clocked mode.

⁵ The PLL operates at self-clocked mode (SCM) frequency when the reference frequency falls below f_{LOR} . SCM frequency is measured on the CLKOUT ball with the divider set to divide-by-two of the system clock.

NOTE: In SCM, the MFD and PREDIV have no effect and the RFD is bypassed.

⁶ Use the EXTAL input high voltage parameter when using the FlexCAN oscillator in crystal mode (no quartz crystals or resonators). ($V_{extal} - V_{xtal}$) must be $\geq 400\text{ mV}$ for the oscillator's comparator to produce the output clock.

⁷ Use the EXTAL input low voltage parameter when using the FlexCAN oscillator in crystal mode (no quartz crystals or resonators). ($V_{xtal} - V_{extal}$) must be $\geq 400\text{ mV}$ for the oscillator's comparator to produce the output clock.

⁸ I_{xtal} is the oscillator bias current out of the XTAL pin with both EXTAL and XTAL pins grounded.

⁹ C_{PCB_EXTAL} and C_{PCB_XTAL} are the measured PCB stray capacitances on EXTAL and XTAL, respectively.

¹⁰ This specification applies to the period required for the PLL to relock after changing the MFD frequency control bits in the synthesizer control register (SYNCR). From power up with crystal oscillator reference, the lock time also includes the crystal startup time.

¹¹ PLL is operating in 1:1 PLL mode.

¹² $V_{DDE} = 3.0\text{--}3.6\text{ V}$.

¹³ Jitter is the average deviation from the programmed frequency measured over the specified interval at maximum f_{SYS} . Measurements are made with the device powered by filtered supplies and clocked by a stable external clock signal. Noise injected into the PLL circuitry via V_{DDSYN} and V_{SSSYN} and variation in crystal oscillator frequency increase the jitter percentage for a given interval. CLKOUT divider is set to divide-by-two.

¹⁴ Values are with frequency modulation disabled. If frequency modulation is enabled, jitter is the sum of (jitter + Cmod).

¹⁵ Modulation depth selected must not result in f_{SYS} value greater than the f_{SYS} maximum specified value.

¹⁶ $f_{SYS} = f_{ICO} \div (2^{RFD})$.

¹⁷ The ICO frequency can be higher than the maximum allowable system frequency. For this case, set the FMPLL synthesizer control register reduced frequency divider (FMPLL_SYNCR[RFD]) to divide-by-two (RFD = 0b001). Therefore, for a 40 MHz maximum device (system frequency), program the FMPLL to generate 80 MHz at the ICO output and then divide-by-two the RFD to provide the 40 MHz clock.

¹⁸ Maximum value for dual controller (1:1) mode is $(f_{MAX} \div 2)$ with the predivider set to 1 (FMPLL_SYNCR[PREDIV] = 0b001).

3.10 eQADC Electrical Characteristics

Table 13. eQADC Conversion Specifications ($T_A = T_L$ to T_H)

Spec	Characteristic	Symbol	Minimum	Maximum	Unit
1	ADC clock (ADCLK) frequency ¹	F_{ADCLK}	1	12	MHz
2	Conversion cycles Differential Single ended	CC	13 + 2 (15) 14 + 2 (16)	13 + 128 (141) 14 + 128 (142)	ADCLK cycles
3	Stop mode recovery time ²	T_{SR}	10	—	μ s
4	Resolution ³	—	1.25	—	mV
5	INL: 6 MHz ADC clock	INL6	−4	4	Counts ³
6	INL: 12 MHz ADC clock	INL12	−8	8	Counts
7	DNL: 6 MHz ADC clock	DNL6	−3 ⁴	3 ⁴	Counts
8	DNL: 12 MHz ADC clock	DNL12	−6 ⁴	6 ⁴	Counts
9	Offset error with calibration	OFFWC	−4 ⁵	4 ⁵	Counts
10	Full-scale gain error with calibration	GAINWC	−8 ⁶	8 ⁶	Counts
11	Disruptive input injection current ^{7, 8, 9, 10}	I_{INJ}	−1	1	mA
12	Incremental error due to injection current. All channels are $10\text{ k}\Omega < R_s < 100\text{ k}\Omega$ Channel under test has $R_s = 10\text{ k}\Omega$, $I_{INJ} = I_{INJMAX}, I_{INJMIN}$	E_{INJ}	−4	4	Counts
13	Total unadjusted error (TUE) for single ended conversions with calibration ^{11, 12, 13, 14, 15}	TUE	−4	4	Counts

¹ Conversion characteristics vary with F_{ADCLK} rate. Reduced conversion accuracy occurs at maximum F_{ADCLK} rate. The maximum value is based on 800 KS/s and the minimum value is based on 20 MHz oscillator clock frequency divided by a maximum 16 factor.

² Stop mode recovery time begins when the ADC control register enable bits are set until the ADC is ready to perform conversions.

³ At $V_{RH} - V_{RL} = 5.12\text{ V}$, one least significant bit (LSB) = 1.25, mV = one count.

⁴ Guaranteed 10-bit mono tonicity.

⁵ The absolute value of the offset error without calibration ≤ 100 counts.

⁶ The absolute value of the full scale gain error without calibration ≤ 120 counts.

⁷ Below disruptive current conditions, the channel being stressed has conversion values of: 0x3FF for analog inputs greater than V_{RH} , and 0x000 for values less than V_{RL} . This assumes that $V_{RH} \leq V_{DDA}$ and $V_{RL} \geq V_{SSA}$ due to the presence of the sample amplifier. Other channels are not affected by non-disruptive conditions.

⁸ Exceeding the limit can cause a conversion error on both stressed and unstressed channels. Transitions within the limit do not affect device reliability or cause permanent damage.

⁹ Input must be current limited to the value specified. To determine the value of the required current-limiting resistor, calculate resistance values using $V_{POSCLAMP} = V_{DDA} + 0.5\text{ V}$ and $V_{NEGCLAMP} = -0.3\text{ V}$, then use the larger of the calculated values.

¹⁰ This condition applies to two adjacent pads on the internal pad.

¹¹ The TUE specification is always less than the sum of the INL, DNL, offset, and gain errors due to canceling errors.

¹² TUE does not apply to differential conversions.

¹³ Measured at 6 MHz ADC clock. TUE with a 12 MHz ADC clock is: $-16\text{ counts} < \text{TUE} < 16\text{ counts}$.

¹⁴ TUE includes all internal device errors such as internal reference variation (75% Ref, 25% Ref).

¹⁵ Depending on the input impedance, the analog input leakage current (Table 9. DC Electrical Specifications, spec 35a) can affect the actual TUE measured on analog channels AN[12], AN[13], AN[14], AN[15].

Table 16 shows the FLASH_BIU settings versus frequency of operation. Refer to the device reference manual for definitions of these bit fields.

Table 16. FLASH_BIU Settings vs. Frequency of Operation¹

Target Maximum Frequency (MHz)	APC	RWSC	WWSC	DPFEN ²	IPFEN ²	PFLIM ³	BFEN ²
Up to and including 27 MHz ^{4, 5}	0b000	0b000	0b01	0b0 0b1	0b0 0b1	0b000 to 0b010	0b0 0b1
Up to and including 52 MHz ⁶	0b001	0b001	0b01	0b0 0b1	0b0 0b1	0b000 to 0b010	0b0 0b1
Up to and including 77 MHz ⁷	0b010	0b010	0b01	0b0 0b1	0b0 0b1	0b000 to 0b010	0b0 0b1
Up to and including 82 MHz ⁸	0b011 ⁹	0b011 ⁹	0b01	0b0 0b1	0b0 0b1	0b000 to 0b010	0b0 0b1
Reset values:	0b111	0b111	0b11	0b0	0b0	0b000	0b0

¹ Illegal combinations exist. Use entries from the same row in this table.

² For maximum flash performance, set to 0b1.

³ For maximum flash performance, set to 0b010.

⁴ 27 MHz parts allow for 25 MHz system clock + 2% frequency modulation (FM).

⁵ The APC, RWSC, and WWSC combination requires setting the PRD bit to 1 in the flash MCR register.

⁶ 52 MHz parts allow for 50 MHz system clock + 2% FM.

⁷ 77 MHz parts allow for 75 MHz system clock + 2% FM.

⁸ 82 MHz parts allow for 80 MHz system clock + 2% FM.

⁹ For frequencies up to and including 80 MHz, if VDD is within $\pm 5\%$ of 1.5 V, then APC = RWSC = 0b010 is a valid setting.

3.12 AC Specifications

3.12.1 Pad AC Specifications

Table 17. Pad AC Specifications ($V_{DDEH} = 5.0\text{ V}$, $V_{DDE} = 1.8\text{ V}$)¹

Spec	Pad	SRC / DSC (binary)	Out Delay (ns) ^{2, 3, 4}	Rise / Fall ^{4, 5} (ns)	Load Drive (pF)
1	Slow high voltage (SH)	11	26	15	50
			82	60	200
		01	75	40	50
			137	80	200
		00	377	200	50
			476	260	200

Table 17. Pad AC Specifications ($V_{DDEH} = 5.0\text{ V}$, $V_{DDE} = 1.8\text{ V}$)¹ (continued)

Spec	Pad	SRC / DSC (binary)	Out Delay ^{2, 3, 4} (ns)	Rise / Fall ^{4, 5} (ns)	Load Drive (pF)
2	Medium high voltage (MH)	11	16	8	50
			43	30	200
		01	34	15	50
			61	35	200
		00	192	100	50
			239	125	200
3	Fast	00	3.1	2.7	10
		01		2.5	20
		10		2.4	30
		11		2.3	50
4	Pullup/down (3.6 V max)	—	—	7500	50
5	Pullup/down (5.5 V max)	—	—	9000	50

¹ These are worst-case values that are estimated from simulation (not tested). The values in the table are simulated at: $V_{DD} = 1.35\text{--}1.65\text{ V}$; $V_{DDE} = 1.62\text{--}1.98\text{ V}$; $V_{DDEH} = 4.5\text{--}5.25\text{ V}$; V_{DD33} and $V_{DDSYN} = 3.0\text{--}3.6\text{ V}$; and $T_A = T_L$ to T_H .

² This parameter is supplied for reference and is guaranteed by design (not tested).

³ The output delay is shown in Figure 4. To calculate the output delay with respect to the system clock, add a maximum of one system clock to the output delay.

⁴ The output delay and rise and fall are measured to 20% or 80% of the respective signal.

⁵ This parameter is guaranteed by characterization rather than 100% tested.

Table 18. Derated Pad AC Specifications ($V_{DDEH} = 3.3\text{ V}$, $V_{DDE} = 3.3\text{ V}$)¹

Spec	Pad	SRC/DSC (binary)	Out Delay ^{2, 3, 4} (ns)	Rise / Fall ^{3, 5} (ns)	Load Drive (pF)
1	Slow high voltage (SH)	11	39	23	50
			120	87	200
		01	101	52	50
			188	111	200
		00	507	248	50
			597	312	200
2	Medium high voltage (MH)	11	23	12	50
			64	44	200
		01	50	22	50
			90	50	200
		00	261	123	50
			305	156	200

Table 19. Reset and Configuration Pin Timing ¹ (continued)

Spec	Characteristic	Symbol	Min.	Max.	Unit
3	PLLCFG, BOOTCFG, WKPCFG, RSTCFG setup time to $\overline{\text{RSTOUT}}$ valid	t_{RCSU}	10	—	t_{CYC}
4	PLLCFG, BOOTCFG, WKPCFG, RSTCFG hold time from $\overline{\text{RSTOUT}}$ valid	t_{RCH}	0	—	t_{CYC}

¹ Reset timing specified at: $V_{\text{DDEH}} = 3.0\text{--}5.25\text{ V}$ and $T_{\text{A}} = T_{\text{L}}$ to T_{H} .

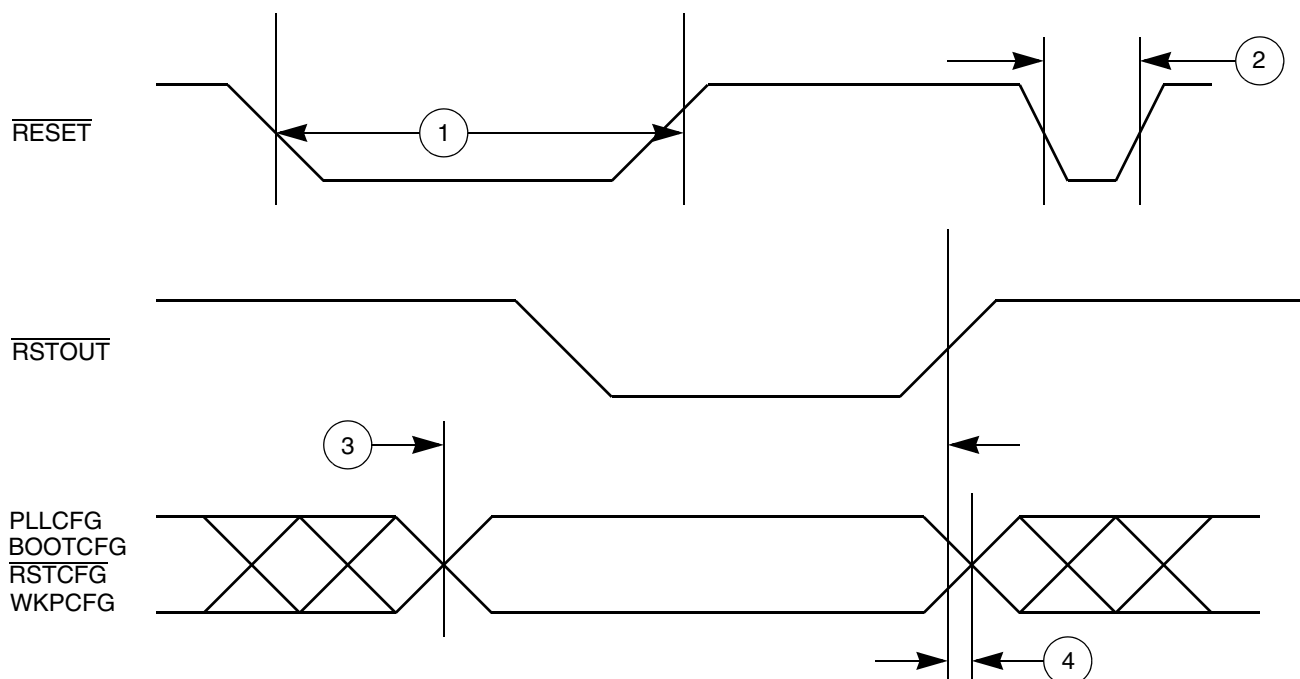


Figure 5. Reset and Configuration Pin Timing

3.13.2 IEEE 1149.1 Interface Timing

Table 20. JTAG Pin AC Electrical Characteristics ¹

Spec	Characteristic	Symbol	Min.	Max.	Unit
1	TCK cycle time	t_{JCYC}	100	—	ns
2	TCK clock pulse width (measured at $V_{\text{DDE}} \div 2$)	t_{JDC}	40	60	ns
3	TCK rise and fall times (40% to 70%)	t_{TCKRISE}	—	3	ns
4	TMS, TDI data setup time	$t_{\text{TMSS}}, t_{\text{TDIS}}$	5	—	ns
5	TMS, TDI data hold time	$t_{\text{TMSH}}, t_{\text{TDIH}}$	25	—	ns
6	TCK low to TDO data valid	t_{TDOV}	—	20	ns
7	TCK low to TDO data invalid	t_{TDOI}	0	—	ns
8	TCK low to TDO high impedance	t_{TDOHZ}	—	20	ns
9	JCOMP assertion time	t_{JCOMPW}	100	—	ns
10	JCOMP setup time to TCK low	t_{JCMPs}	40	—	ns
11	TCK falling-edge to output valid	t_{BSDV}	—	50	ns

Table 20. JTAG Pin AC Electrical Characteristics ¹ (continued)

Spec	Characteristic	Symbol	Min.	Max.	Unit
12	TCK falling-edge to output valid out of high impedance	t_{BSDVZ}	—	50	ns
13	TCK falling-edge to output high impedance (Hi-Z)	t_{BSDHZ}	—	50	ns
14	Boundary scan input valid to TCK rising-edge	t_{BSDST}	50	—	ns
15	TCK rising-edge to boundary scan input invalid	t_{BSDHT}	50	—	ns

¹ These specifications apply to JTAG boundary scan only. JTAG timing specified at: $V_{\text{DDE}} = 3.0\text{--}3.6\text{ V}$ and $T_A = T_L$ to T_H . Refer to [Table 21](#) for Nexus specifications.

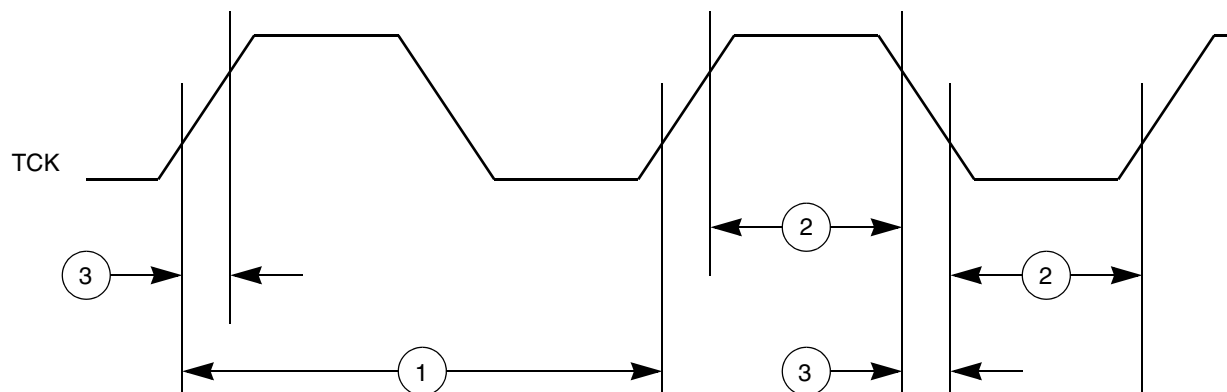


Figure 6. JTAG Test Clock Input Timing

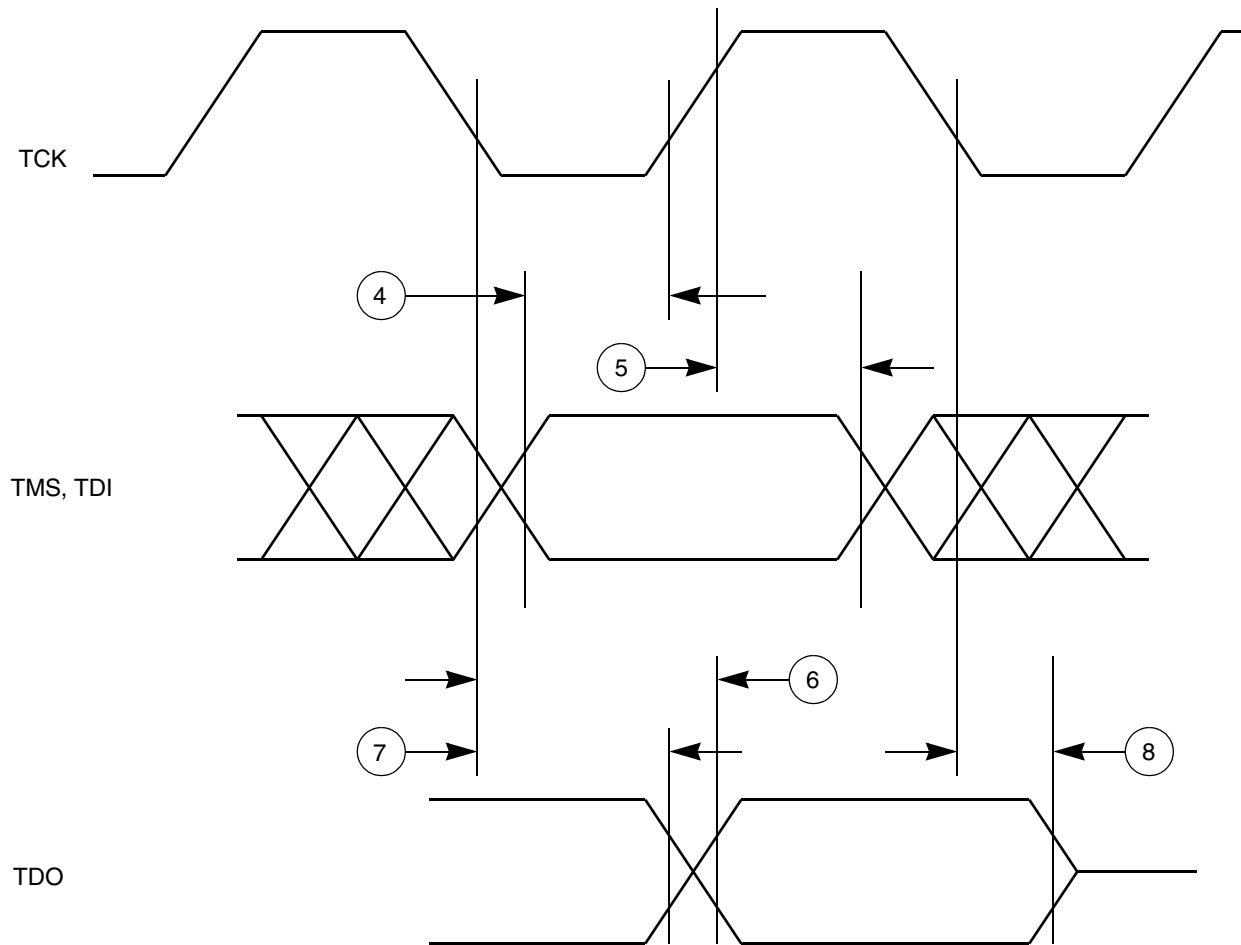


Figure 7. JTAG Test Access Port Timing

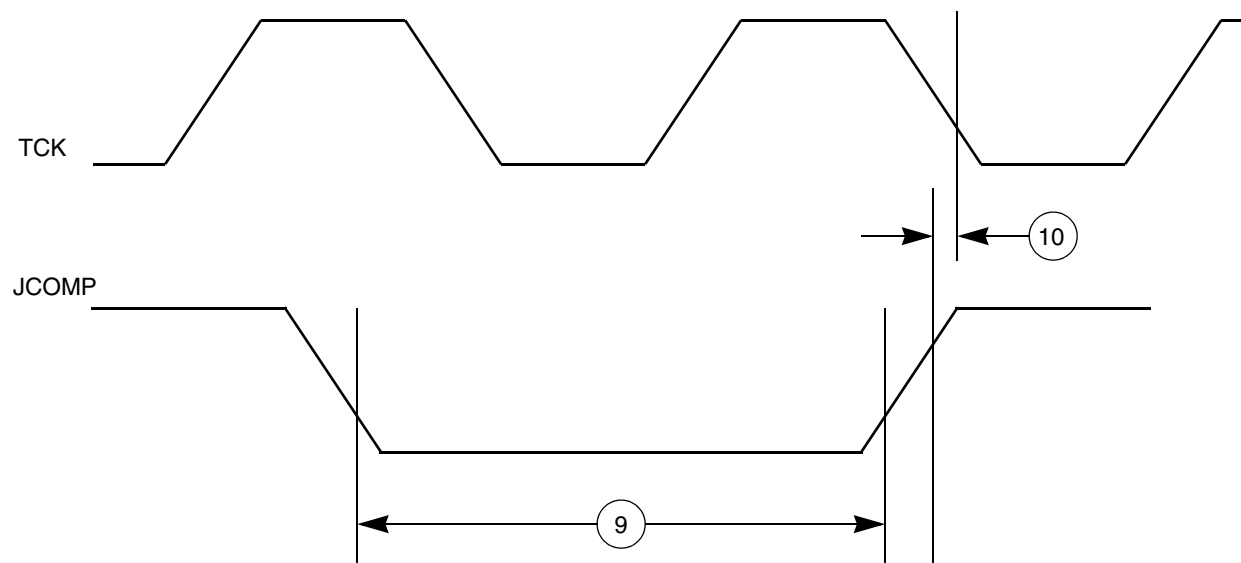


Figure 8. JTAG JCOMP Timing

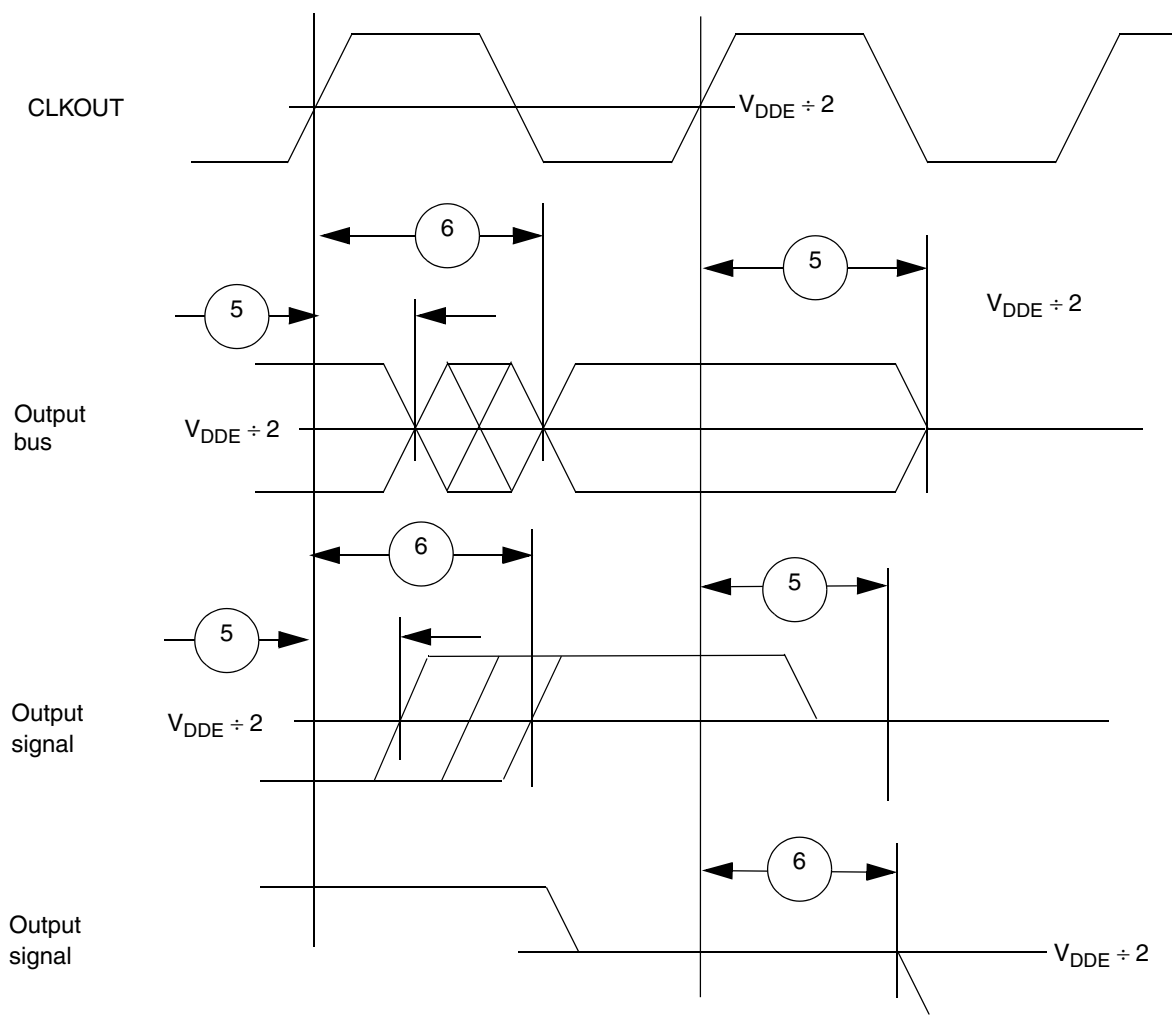


Figure 13. Synchronous Output Timing

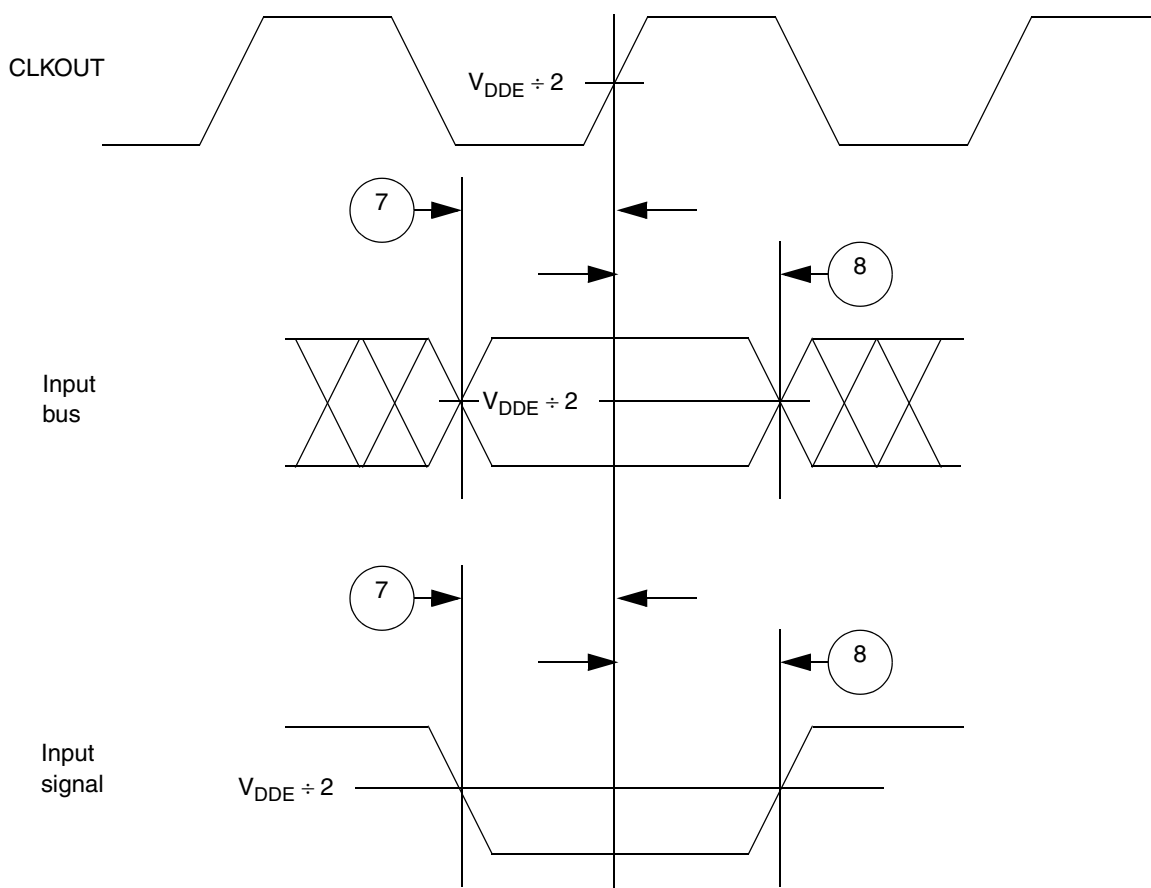


Figure 14. Synchronous Input Timing

3.13.5 External Interrupt Timing (IRQ Signals)

Table 23. External Interrupt Timing ¹

Spec	Characteristic	Symbol	Min.	Max.	Unit
1	IRQ pulse-width low	t_{IPWL}	3	—	t_{CYC}
2	IRQ pulse-width high	T_{IPWH}	3	—	t_{CYC}
3	IRQ edge-to-edge time ²	t_{ICYC}	6	—	t_{CYC}

¹ IRQ timing specified at: $V_{DDEH} = 3.0\text{--}5.25\text{ V}$ and $T_A = T_L$ to T_H .

² Applies when IRQ signals are configured for rising-edge or falling-edge events, but not both.

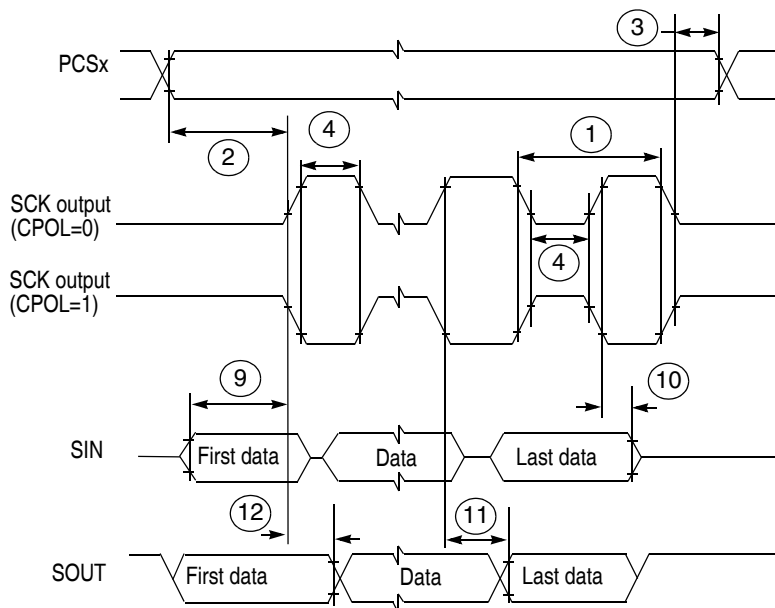


Figure 22. DSPI Modified Transfer Format Timing—Master, CPHA = 0

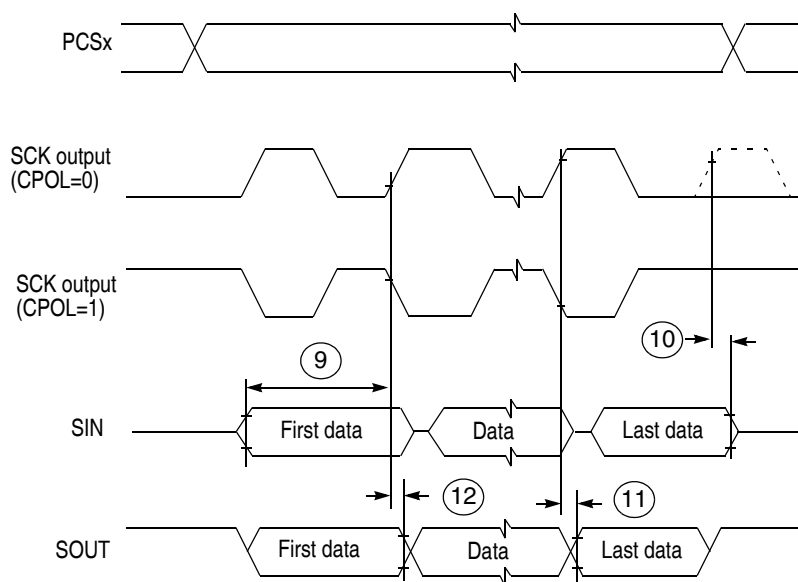


Figure 23. DSPI Modified Transfer Format Timing—Master, CPHA = 1

3.13.9 eQADC SSI Timing

Table 27. EQADC SSI Timing Characteristics

Spec	Rating	Symbol	Minimum	Typical	Maximum	Unit
2	FCK period ($t_{FCK} = 1 \div f_{FCK}$) ^{1, 2}	t_{FCK}	2	—	17	t_{SYS_CLK}
3	Clock (FCK) high time	t_{FCKHT}	$t_{SYS_CLK} - 6.5$	—	$9 \times (t_{SYS_CLK} + 6.5)$	ns
4	Clock (FCK) low time	t_{FCKLT}	$t_{SYS_CLK} - 6.5$	—	$8 \times (t_{SYS_CLK} + 6.5)$	ns
5	SDS lead / lag time	t_{SDS_LL}	-7.5	—	+7.5	ns
6	SDO lead / lag time	t_{SDO_LL}	-7.5	—	+7.5	ns
7	EQADC data setup time (inputs)	t_{EQ_SU}	22	—	—	ns
8	EQADC data hold time (inputs)	t_{EQ_HO}	1	—	—	ns

¹ SS timing specified at $V_{DDEH} = 3.0\text{--}5.25\text{ V}$, $T_A = T_L$ to T_H , and $CL = 25\text{ pF}$ with $SRC = 0b11$. Maximum operating frequency varies depending on track delays, master pad delays, and slave pad delays.

² FCK duty cycle is not 50% when it is generated through the division of the system clock by an odd number.

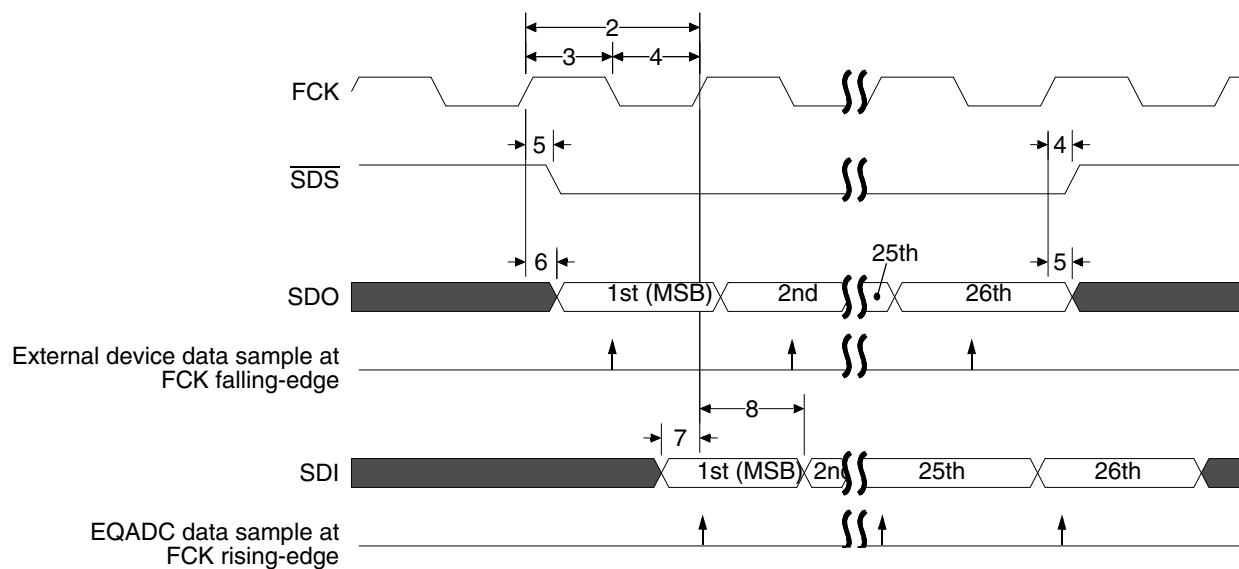


Figure 27. EQADC SSI Timing

4.2 MPC5534 324 PBGA Pinouts

Figure 29 is a pinout for the MPC5534 324 PBGA package.

NOTE

The MPC5500 devices are pin compatible for software portability and use the primary function names to label the pins in the BGA diagram. Although some devices do not support all the primary functions shown in the BGA diagram, the muxed and GPIO signals on those pins remain available. See the signals chapter in the device reference manual for the signal muxing.

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	
A	VSS	VDD	VSTBY	AN37	AN11	VDDA1	VSSA1	AN1	AN5	VRH	VRL	AN27	AN28	AN35	VSSA0	AN12	MDO11	MDO10	MDO8	VDD	VDD33	VSS	A
B	VDD33	VSS	VDD	AN36	AN39	AN19	AN16	AN0	AN4	REF BYPC	AN23	AN26	AN31	AN32	VSSA0	AN13	MDO9	MDO7	MDO4	MDO0	VSS	VDDE7	B
C	ETPUA 30	ETPUA 31	VSS	VDD	AN8	AN17	AN20	AN21	AN3	AN7	AN22	AN25	AN30	AN33	VDDA0	AN14	MDO5	MDO2	MDO1	VSS	VDDE7	VDD	C
D	ETPUA 28	ETPUA 29	ETPUA 26	VSS	VDD	AN38	AN9	AN10	AN18	AN2	AN6	AN24	AN29	AN34	VDDEH 9	AN15	MDO6	MDO3	VSS	VDDE7	TCK	TDI	D
E	ETPUA 24	ETPUA 27	ETPUA 25	ETPUA 21															VDDE7	TMS	TDO	TEST	E
F	ETPUA 23	ETPUA 22	ETPUA 17	ETPUA 18															VDDE7	JCOMP	EVTI	EVTO	F
G	ETPUA 20	ETPUA 19	ETPUA 14	ETPUA 13															RDY	MCKO	MSEO0	MSEO1	G
H	ETPUA 16	ETPUA 15	ETPUA 10	VDDEH 1															VDDEH 10	GPIO 203	GPIO 204	SINB	H
J	ETPUA 12	ETPUA 11	ETPUA 6	ETPUA 9															SOUTB	PCSB3	PCSB0	PCSB1	J
K	ETPUA 8	ETPUA 7	ETPUA 2	ETPUA 5															PCSA3	PCSB4	SCKB	PCSB2	K
L	ETPUA 4	ETPUA 3	ETPUA 0	ETPUA 1															PCSB5	SOUTA	SINA	SCKA	L
M	BDIP	TCRCLK A	CS1	CS0															PCSA1	PCSA0	PCSA2	VPP	M
N	CS3	CS2	WE1	WE0															PCSA4	TXDA	PCSA5	VFLASH	N
P	ADDR 16	ADDR 17	RD_WR	VDD33															CNTXC	RXDA	RSTOUT	RST CFG	P
R	ADDR 18	ADDR 19	VDDE2	TA															WKP CFG	CNRXC	TXDB	RESET	R
T	ADDR 20	ADDR 21	ADDR 12	TS															RXDB	BOOT CFG1	PLL CFG2	VSS SYN	T
U	ADDR 22	ADDR 23	ADDR 13	ADDR 14															VDDEH 6	PLL CFG1	BOOT CFG0	EXTAL	U
V	ADDR 24	ADDR 25	ADDR 15	ADDR 31															VDD	VRC CTL	PLL CFG0	XTAL	V
W	ADDR 26	VDDE2	ADDR 30	VSS	VDD	VDDE2	VDD33	VDDE2	DATA 11	DATA 12	DATA 14	EMIOS 2	EMIOS 8	VDDEH 4	EMIOS 12	EMIOS 21	VDDE5	NC	VSS	VDD	VRC33	VDD SYN	W
Y	ADDR 28	ADDR 27	VSS	VDD	VDDE2	DATA 8	DATA 9	DATA 10	GPIO 207	DATA 13	DATA 15	EMIOS 6	EMIOS 10	EMIOS 15	EMIOS 17	EMIOS 22	CNTXA	VDDE5	NC	VSS	VDD	VDD33	Y
AA	ADDR 29	VSS	VDD	VDDE2	DATA 1	VDDE2	GPIO 206	DATA 5	DATA 7	VDDE2	EMIOS 3	EMIOS 5	EMIOS 9	EMIOS 13	EMIOS 16	EMIOS 19	EMIOS 23	CNRXA	VDDE5	CLKOUT	VSS	VDD	AA
AB	VSS	VDD	VDDE2	DATA 0	DATA 2	DATA 3	DATA 4	DATA 6	OE	EMIOS 0	EMIOS 1	EMIOS 4	EMIOS 7	EMIOS 11	EMIOS 14	EMIOS 18	EMIOS 20	CNTXB	CNRXB	VDDE5	ENG CLK	VSS	AB
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	


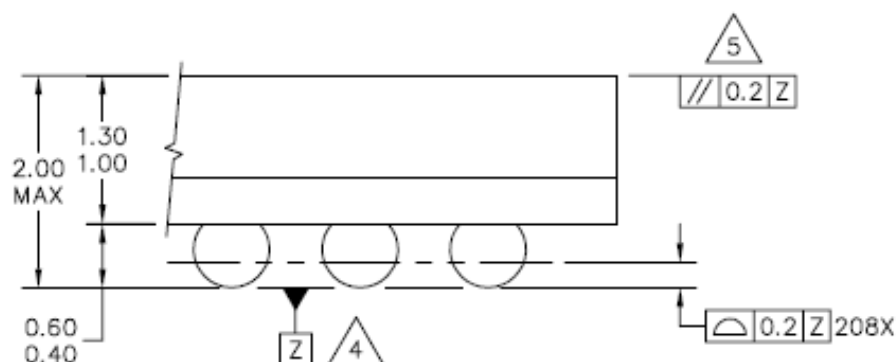
Note:  No connect. Reserved (W18 & Y19 are shorted to each other)

Figure 29. MPC5534 324 Package



DETAIL K
(ROTATED 90° CLOCKWISE)

NOTES:

1. ALL DIMENSIONS ARE IN MILLIMETERS.
2. INTERPRET DIMENSIONS AND TOLERANCES PER ASME Y14.5M-1994.
3. DIMENSION b IS MEASURED AT THE MAXIMUM SOLDER BALL DIAMETER, PARALLEL TO DATUM PLANE Z.
4. DATUM Z (SEATING PLANE) IS DEFINED BY THE SPHERICAL CROWNS OF THE SOLDER BALLS.
5. PARALLELISM MEASUREMENT SHALL EXCLUDE ANY EFFECT OF MARK ON TOP SURFACE OF PACKAGE.

© FREESCALE SEMICONDUCTOR, INC. ALL RIGHTS RESERVED.		MECHANICAL OUTLINE		PRINT VERSION NOT TO SCALE	
TITLE: 208 I/O MAP BGA, 17 X 17 PKG, 1-MM PITCH			DOCUMENT NO: 98ARS23882W		REV: D
			CASE NUMBER: 1159A-01		02 AUG 2005
			STANDARD: JEDEC MO-151 AAF-1		

Figure 30. MPC5534 208 MAP BGA Package (continued)

4.4 MPC5534 324-Pin Package Dimensions

The package drawings of the MPC5534 324-pin TEPBGA package are shown in Figure 31.

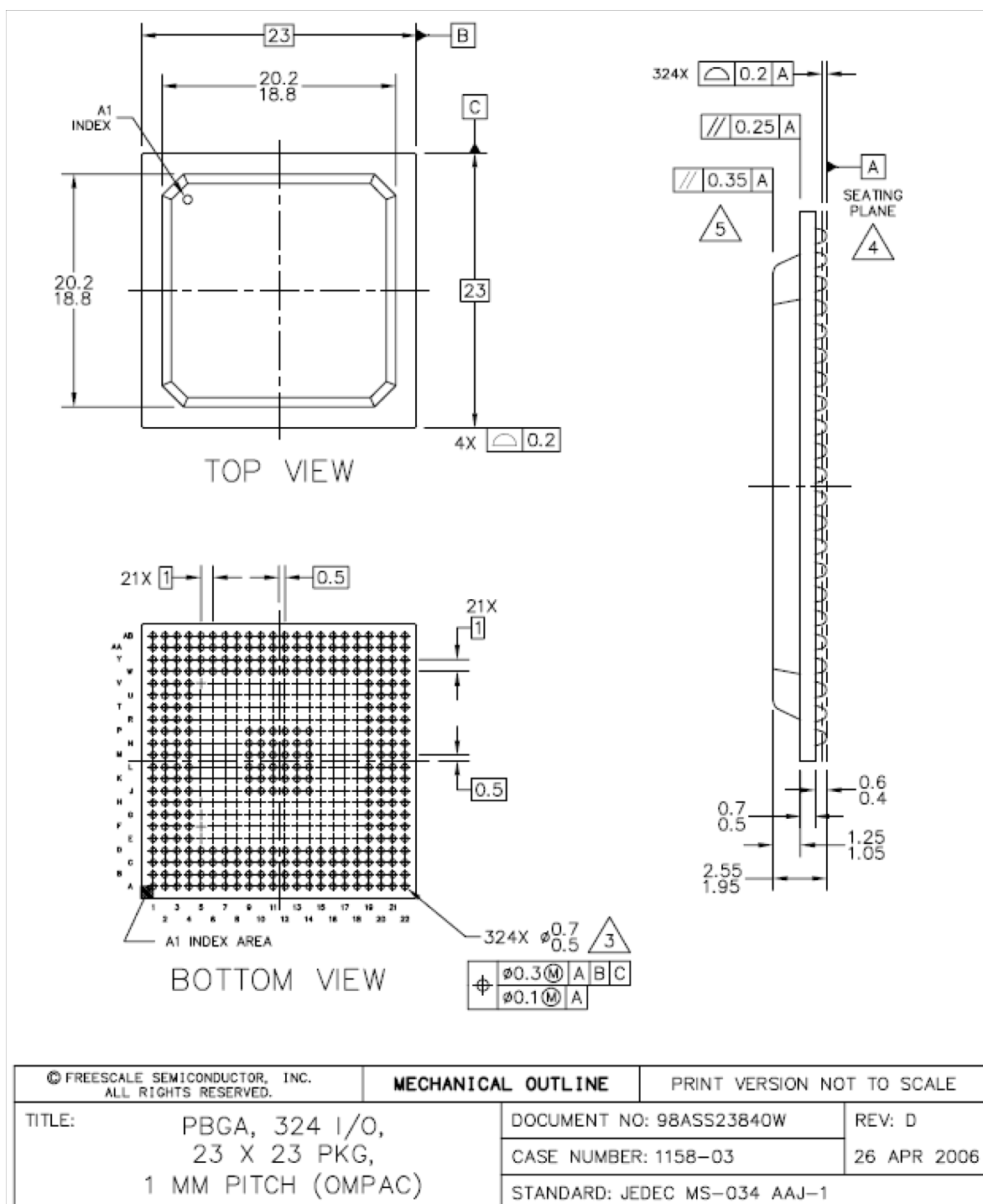


Figure 31. MPC5534 324 TEPBGA Package

Table 30. Table and Figure Changes Between Rev. 3.0 and 4.0 (continued)

Location	Description of Changes								
Table 4, EMI Testing Specifications:	Changed the maximum operating frequency from 82 to f_{MAX} .								
Table 6, VCR/POR Electrical Specifications:	<ul style="list-style-type: none">Added footnote 1 to specs 1, 2, and 3 that reads: On power up, assert \overline{RESET} before V_{POR15}, V_{POR33}, and V_{POR5} negate (internal POR). RESET must remain asserted until the power supplies are within the operating conditions as specified in Table 9 DC Electrical Specifications. On power down, assert \overline{RESET} before any power supplies fall outside the operating conditions and until the internal POR asserts.Subscript all symbol names that appear after the first underscore character.Removed 'Tj ' after '150 C' in Spec 10, Characteristic column:Added footnote 1 to specs 1, 2, and 3 that reads: On power up, assert \overline{RESET} before V_{POR15}, V_{POR33}, and V_{POR5} negate (internal POR). RESET must remain asserted until the power supplies are within the operating conditions as specified in Table 9 DC Electrical Specifications. On power down, assert \overline{RESET} before any power supplies fall outside the operating conditions and until the internal POR asserts.Added to Spec 2:<table><tr><td>3.3 V (V_{DDSYN}) POR negated (ramp down)</td><td>Min 0.0</td><td>Max 0.30</td><td>V</td></tr><tr><td>3.3 V (V_{DDSYN}) POR asserted (ramp up)</td><td>Min 0.0</td><td>Max 0.30</td><td>V</td></tr></table>Spec 3: Added new footnote 3 for both lines: 'It is possible to reach the current limit during ramp up--do not treat this event as a short circuit current.'Spec 5: Changed old Footnote 1 (now footnote 2): 'User must be able to supply full operating current for the 1.5 V supply when the 3.3 V supply reaches this range.' to 'Supply full operating current for the 1.5 V supply when the 3.3 V supply reaches this range.'Specs 7 and 10: added 'at Tj ' at the end of the first line in the second column: Characteristic.Spec 10:<ul style="list-style-type: none">Added cross-reference to footnote 6: 'I_{VRCCTL} is measured at the following conditions: $V_{DD} = 1.35$ V, $V_{RC33} = 3.1$ V, $V_{VRCCTL} = 2.2$ V.' Changed '(@ $V_{DD} = 1.35$ V, $f_{sys} = f_{MAX}$)' to '(@ $f_{sys} = f_{MAX}$).Added old footnote 5 new footnote 6.Added a new footnote 7, 'Refer to Table 1 for the maximum operating frequency.'Rewrote old footnote 8 (new footnote 9) to: Represents the worst-case external transistor BETA. It is measured on a per part basis and calculated as $(I_{DD} \div I_{VRCCTL})$.Deleted footnote 9 Preliminary value, Final specification pending characterization.	3.3 V (V_{DDSYN}) POR negated (ramp down)	Min 0.0	Max 0.30	V	3.3 V (V_{DDSYN}) POR asserted (ramp up)	Min 0.0	Max 0.30	V
3.3 V (V_{DDSYN}) POR negated (ramp down)	Min 0.0	Max 0.30	V						
3.3 V (V_{DDSYN}) POR asserted (ramp up)	Min 0.0	Max 0.30	V						
Table 7, Power Sequence Pin Status for Fast Pads:	Changed title to Pin Status for Fast Pads During the Power Sequence								
Table 8, Power Sequence Pin Status for Medium/Slow Pads:	<ul style="list-style-type: none">Changed title to <i>Pin Status for Medium and Slow Pads During the Power Sequence</i>Updated preceding paragraph.								
Table 9, DC Electrical Specifications:	<ul style="list-style-type: none">Spelled out meaning of the slash '/' as 'and'. Still confusing. Deleted 'I/O' from the specs to improve clarity.Added footnote that reads: V_{DDE2} and V_{DDE3} are limited to 2.25–3.6 V only if EBTS = 0; V_{DDE2} and V_{DDE3} have a range of 1.6–3.6 V if EBTS = 1.Spec 20, column 2, <i>Characteristics</i>: Slow and medium output high voltage ($I_{OH_S} = -2.0$ mA):' Created a left-justified second line and moved '$I_{OH_S} = -2.0$ mA' from the 1st line to the second line and deleted the parentheses. Created a left-justified third line that reads '$I_{OH_S} = -1.0$ mA.'Spec 20, column 4, <i>Min</i>: Added a blank line before and after '$0.80 \times V_{DDEH}$' and put '$0.85 \times V_{DDEH}$' on the last line.Spec 22, column 2, '<i>Slow and medium output low voltage ($I_{OL_S} = 2.0$ mA)</i>:' Created a left-justified second line and moved '$I_{OL_S} = 2.0$ mA.' from the 1st line to the second line and deleted the parentheses. Created a left-justified third line that reads '$I_{OL_S} = 1.0$ mA.'Spec 22, column 5, <i>Max</i>: Added a blank line before and after '$0.20 \times V_{DDEH}$' and put '$0.15 \times V_{DDEH}$' on the last line.Spec 26: Changed 'AN[12]_MA[1]_SDO' to 'AN[13]_MA[1]_SDO'.Spec 28: Changed 82 MHz to f_{MAX} MHz.Footnote 9: Changed from 'Preliminary. Final specification pending characterization.' to ' shows an illustration of the I_{DD_STBY} values interpolated for these temperature values.'								