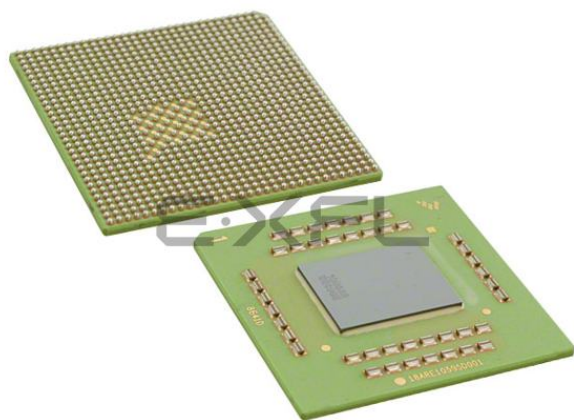




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Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details

Product Status	Obsolete
Core Processor	PowerPC e600
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	1.0GHz
Co-Processors/DSP	-
RAM Controllers	DDR, DDR2
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (4)
SATA	-
USB	-
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	-
Package / Case	994-BCBGA, FCCBGA
Supplier Device Package	994-FCCBGA (33x33)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mc8641vu1000gc

1.1 Key Features

The following lists an overview of the MPC8641 key feature set:

- Major features of the e600 core are as follows:
 - High-performance, 32-bit superscalar microprocessor that implements the PowerPC ISA
 - Eleven independent execution units and three register files
 - Branch processing unit (BPU)
 - Four integer units (IUs) that share 32 GPRs for integer operands
 - 64-bit floating-point unit (FPU)
 - Four vector units and a 32-entry vector register file (VRs)
 - Three-stage load/store unit (LSU)
 - Three issue queues, FIQ, VIQ, and GIQ, can accept as many as one, two, and three instructions, respectively, in a cycle.
 - Rename buffers
 - Dispatch unit
 - Completion unit
 - Two separate 32-Kbyte instruction and data level 1 (L1) caches
 - Integrated 1-Mbyte, eight-way set-associative unified instruction and data level 2 (L2) cache with ECC
 - 36-bit real addressing
 - Separate memory management units (MMUs) for instructions and data
 - Multiprocessing support features
 - Power and thermal management
 - Performance monitor
 - In-system testability and debugging features
 - Reliability and serviceability
- MPX coherency module (MCM)
 - Ten local address windows plus two default windows
 - Optional low memory offset mode for core 1 to allow for address disambiguation
- Address translation and mapping units (ATMUs)
 - Eight local access windows define mapping within local 36-bit address space
 - Inbound and outbound ATMUs map to larger external address spaces
 - Three inbound windows plus a configuration window on PCI Express
 - Four inbound windows plus a default window on serial RapidIO
 - Four outbound windows plus default translation for PCI Express
 - Eight outbound windows plus default translation for serial RapidIO with segmentation and sub-segmentation support

2.1.3 Output Driver Characteristics

Table 3 provides information on the characteristics of the output driver strengths. The values are preliminary estimates.

Table 3. Output Drive Capability

Driver Type	Programmable Output Impedance (Ω)	Supply Voltage	Notes
DDR1 signal	18 36 (half strength mode)	$Dn_GV_{DD} = 2.5\text{ V}$	4, 9
DDR2 signal	18 36 (half strength mode)	$Dn_GV_{DD} = 1.8\text{ V}$	1, 5, 9
Local Bus signals	45 25	$OV_{DD} = 3.3\text{ V}$	2, 6
eTSEC/10/100 signals	45	$T/LV_{DD} = 3.3\text{ V}$	6
	30	$T/LV_{DD} = 2.5\text{ V}$	6
DUART, DMA, Multiprocessor Interrupts, System Control & Clocking, Debug, Test, Power management, JTAG and Miscellaneous I/O voltage	45	$OV_{DD} = 3.3\text{ V}$	6
I ² C	150	$OV_{DD} = 3.3\text{ V}$	7
SRIO, PCI Express	100	$SV_{DD} = 1.1/1.05\text{ V}$	3, 8

Notes:

1. See the DDR Control Driver registers in the MPC8641D reference manual for more information.
2. Only the following local bus signals have programmable drive strengths: LALE, LAD[0:31], LDP[0:3], LA[27:31], LCKE, LCS[1:2], LWE[0:3], LGPL1, LGPL2, LGPL3, LGPL4, LGPL5, LCLK[0:2]. The other local bus signals have a fixed drive strength of 45 Ω . See the POR Impedance Control register in the MPC8641D reference manual for more information about local bus signals and their drive strength programmability.
3. See [Section 17, "Signal Listings,"](#) for details on resistor requirements for the calibration of $SDn_IMP_CAL_TX$ and $SDn_IMP_CAL_RX$ transmit and receive signals.
4. Stub Series Terminated Logic (SSTL-25) type pins.
5. Stub Series Terminated Logic (SSTL-18) type pins.
6. Low Voltage Transistor-Transistor Logic (LVTTTL) type pins.
7. Open Drain type pins.
8. Low Voltage Differential Signaling (LVDS) type pins.
9. The drive strength of the DDR interface in half strength mode is at $T_j = 105^\circ\text{C}$ and at Dn_GV_{DD} (min).

2.2 Power Up/Down Sequence

The MPC8641 requires its power rails to be applied in a specific sequence in order to ensure proper device operation.

NOTE

The recommended maximum ramp up time for power supplies is 20 milliseconds.

The chronological order of power up is as follows:

1. All power rails other than DDR I/O (Dn_GV_{DD} , and Dn_MV_{REF}).

10 Local Bus

This section describes the DC and AC electrical specifications for the local bus interface of the MPC8641.

10.1 Local Bus DC Electrical Characteristics

Table 40 provides the DC electrical characteristics for the local bus interface operating at $OV_{DD} = 3.3$ V DC.

Table 40. Local Bus DC Electrical Characteristics (3.3 V DC)

Parameter	Symbol	Min	Max	Unit
High-level input voltage	V_{IH}	2	$OV_{DD} + 0.3$	V
Low-level input voltage	V_{IL}	-0.3	0.8	V
Input current ($V_{IN}^1 = 0$ V or $V_{IN} = OV_{DD}$)	I_{IN}	—	± 5	μA
High-level output voltage ($OV_{DD} = \min$, $I_{OH} = -2$ mA)	V_{OH}	$OV_{DD} - 0.2$	—	V
Low-level output voltage ($OV_{DD} = \min$, $I_{OL} = 2$ mA)	V_{OL}	—	0.2	V

Note:

- Note that the symbol V_{IN} , in this case, represents the OV_{IN} symbol referenced in Table 1 and Table 2.

10.2 Local Bus AC Electrical Specifications

Table 41 describes the timing parameters of the local bus interface at $OV_{DD} = 3.3$ V with PLL enabled. For information about the frequency range of local bus see Section 18.1, “Clock Ranges.”

Table 41. Local Bus Timing Parameters ($OV_{DD} = 3.3$ V)m - PLL Enabled

Parameter	Symbol ¹	Min	Max	Unit	Notes
Local bus cycle time	t_{LBK}	7.5	—	ns	2
Local Bus Duty Cycle	t_{LBKH}/t_{LBK}	45	55	%	—
LCLK[n] skew to LCLK[m] or LSYNC_OUT	$t_{LBKSKEW}$	—	150	ps	7, 8
Input setup to local bus clock (except $\overline{LGTA}/LUPWAIT$)	$t_{LBIVKH1}$	1.8	—	ns	3, 4
$\overline{LGTA}/LUPWAIT$ input setup to local bus clock	$t_{LBIVKH2}$	1.7	—	ns	3, 4
Input hold from local bus clock (except $\overline{LGTA}/LUPWAIT$)	$t_{LBIXKH1}$	1.0	—	ns	3, 4
$\overline{LGTA}/LUPWAIT$ input hold from local bus clock	$t_{LBIXKH2}$	1.0	—	ns	3, 4
LALE output transition to LAD/LDP output transition (LATCH hold time)	t_{LBOTOT}	1.5	—	ns	6
Local bus clock to output valid (except LAD/LDP and LALE)	$t_{LBKHOV1}$	—	2.0	ns	—
Local bus clock to data valid for LAD/LDP	$t_{LBKHOV2}$	—	2.2	ns	—
Local bus clock to address valid for LAD	$t_{LBKHOV3}$	—	2.3	ns	—

Table 41. Local Bus Timing Parameters ($OV_{DD} = 3.3\text{ V}$)m - PLL Enabled (continued)

Parameter	Symbol ¹	Min	Max	Unit	Notes
Local bus clock to LALE assertion	$t_{LBKHOV4}$	—	2.3	ns	3
Output hold from local bus clock (except LAD/LDP and LALE)	$t_{LBKHOX1}$	0.7	—	ns	—
Output hold from local bus clock for LAD/LDP	$t_{LBKHOX2}$	0.7	—	ns	3
Local bus clock to output high Impedance (except LAD/LDP and LALE)	$t_{LBKHOZ1}$	—	2.5	ns	5
Local bus clock to output high impedance for LAD/LDP	$t_{LBKHOZ2}$	—	2.5	ns	5

Note:

1. The symbols used for timing specifications herein follow the pattern of $t_{(\text{First two letters of functional block})(\text{signal})(\text{state})}$ (reference)(state) for inputs and $t_{(\text{First two letters of functional block})(\text{reference})(\text{state})(\text{signal})(\text{state})}$ for outputs. For example, $t_{LBIXKH1}$ symbolizes local bus timing (LB) for the input (I) to go invalid (X) with respect to the time the t_{LBK} clock reference (K) goes high (H), in this case for clock one(1). Also, t_{LBKHOX} symbolizes local bus timing (LB) for the t_{LBK} clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.
2. All timings are in reference to LSYNC_IN for PLL enabled and internal local bus clock for PLL bypass mode.
3. All signals are measured from $OV_{DD}/2$ of the rising edge of LSYNC_IN for PLL enabled or internal local bus clock for PLL bypass mode to $0.4 \times OV_{DD}$ of the signal in question for 3.3-V signaling levels.
4. Input timings are measured at the pin.
5. For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
6. t_{LBOTOT} is a measurement of the minimum time between the negation of LALE and any change in LAD. t_{LBOTOT} is programmed with the LBCR[AHD] parameter.
7. Maximum possible clock skew between a clock LCLK[m] and a relative clock LCLK[n]. Skew measured between complementary signals at $BV_{DD}/2$.
8. Guaranteed by design.

Figure 25 provides the AC test load for the local bus.

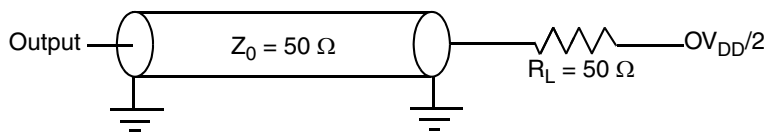


Figure 25. Local Bus AC Test Load

11 JTAG

This section describes the DC and AC electrical specifications for the IEEE 1149.1 (JTAG) interface of the MPC8641/D.

11.1 JTAG DC Electrical Characteristics

Table 43 provides the DC electrical characteristics for the JTAG interface.

Table 43. JTAG DC Electrical Characteristics

Parameter	Symbol	Min	Max	Unit
High-level input voltage	V_{IH}	2	$OV_{DD} + 0.3$	V
Low-level input voltage	V_{IL}	-0.3	0.8	V
Input current ($V_{IN}^1 = 0$ V or $V_{IN} = V_{DD}$)	I_{IN}	—	± 5	μA
High-level output voltage ($OV_{DD} = \text{min}$, $I_{OH} = -100$ μA)	V_{OH}	$OV_{DD} - 0.2$	—	V
Low-level output voltage ($OV_{DD} = \text{min}$, $I_{OL} = 100$ μA)	V_{OL}	—	0.2	V

Note:

1. Note that the symbol V_{IN} , in this case, represents the OV_{IN} symbol referenced in Table 1 and Table 2.

11.2 JTAG AC Electrical Specifications

Table 44 provides the JTAG AC timing specifications as defined in Figure 33 through Figure 35.

Table 44. JTAG AC Timing Specifications (Independent of SYSCLK) ¹

At recommended operating conditions (see Table 3).

Parameter	Symbol ²	Min	Max	Unit	Notes
JTAG external clock frequency of operation	f_{JTG}	0	33.3	MHz	—
JTAG external clock cycle time	t_{JTG}	30	—	ns	—
JTAG external clock pulse width measured at 1.4 V	t_{JTKHKL}	15	—	ns	—
JTAG external clock rise and fall times	t_{JTGR} & t_{JTGF}	0	2	ns	6
$\overline{TRST}$ assert time	t_{TRST}	25	—	ns	3
Input setup times:				ns	
Boundary-scan data TMS, TDI	t_{JTDVKH} t_{JTIVKH}	4 0	— —		4
Input hold times:				ns	
Boundary-scan data TMS, TDI	t_{JTDXKH} t_{JTIXKH}	20 25	— —		4
Valid times:				ns	
Boundary-scan data TDO	t_{JTKLDV} t_{JTKLOV}	4 4	20 25		5

Table 46. I²C AC Electrical Specifications (continued)

All values refer to V_{IH} (min) and V_{IL} (max) levels (see Table 45).

Parameter	Symbol ¹	Min	Max	Unit
Noise margin at the HIGH level for each connected device (including hysteresis)	V _{NH}	0.2 × OV _{DD}	—	V

Note:

- The symbols used for timing specifications herein follow the pattern of t_{(first two letters of functional block)(signal)(state) (reference)(state)} for inputs and t_{(first two letters of functional block)(reference)(state)(signal)(state)} for outputs. For example, t_{I2DV_{KH}} symbolizes I²C timing (I2) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time. Also, t_{I2SX_{KL}} symbolizes I²C timing (I2) for the time that the data with respect to the start condition (S) went invalid (X) relative to the t_{I2C} clock reference (K) going to the low (L) state or hold time. Also, t_{I2PV_{KH}} symbolizes I²C timing (I2) for the time that the data with respect to the stop condition (P) reaching the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
- As a transmitter, the MPC8641 provides a delay time of at least 300 ns for the SDA signal (referred to the V_{IHmin} of the SCL signal) to bridge the undefined region of the falling edge of SCL to avoid unintended generation of Start or Stop condition. When MPC8641 acts as the I²C bus master while transmitting, MPC8641 drives both SCL and SDA. As long as the load on SCL and SDA are balanced, MPC8641 would not cause unintended generation of Start or Stop condition. Therefore, the 300 ns SDA output delay time is not a concern. If, under some rare condition, the 300 ns SDA output delay time is required for MPC8641 as transmitter, the following setting is recommended for the FDR bit field of the I2CFDR register to ensure both the desired I²C SCL clock frequency and SDA output delay time are achieved, assuming that the desired I²C SCL clock frequency is 400 KHz and the Digital Filter Sampling Rate Register (I2CDFSRR) is programmed with its default setting of 0x10 (decimal 16):

I ² C Source Clock Frequency	333 MHz	266 MHz	200 MHz	133 MHz
FDR Bit Setting	0x2A	0x05	0x26	0x00
Actual FDR Divider Selected	896	704	512	384
Actual I ² C SCL Frequency Generated	371 KHz	378 KHz	390 KHz	346 KHz

For the detail of I²C frequency calculation, refer to the application note AN2919 “Determining the I²C Frequency Divider Ratio for SCL”. Note that the I²C Source Clock Frequency is half of the MPX clock frequency for MPC8641.
- The maximum t_{I2DX_{KL}} has only to be met if the device does not stretch the LOW period (t_{I2CL}) of the SCL signal.
- Guaranteed by design.
- C_B = capacitance of one bus line in pF.

Figure 32 provides the AC test load for the I²C.

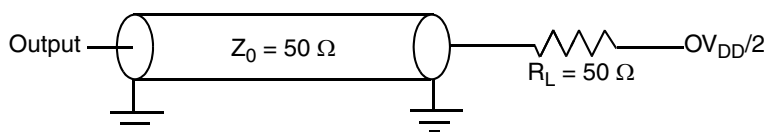

Figure 36. I²C AC Test Load

Figure 43 shows the SerDes reference clock connection reference circuits for HCSL type clock driver. It assumes that the DC levels of the clock driver chip is compatible with MPC8641D SerDes reference clock input's DC requirement.

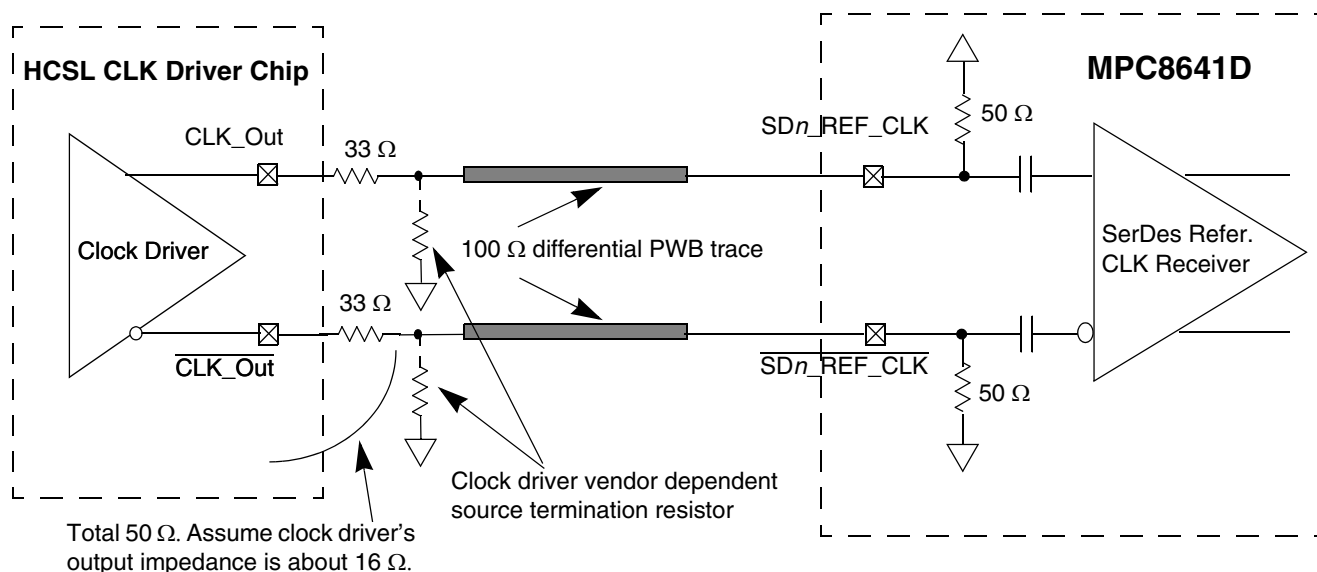


Figure 43. DC-Coupled Differential Connection with HCSL Clock Driver (Reference Only)

Figure 44 shows the SerDes reference clock connection reference circuits for LVDS type clock driver. Since LVDS clock driver's common mode voltage is higher than the MPC8641D SerDes reference clock input's allowed range (100 to 400 mV), AC-coupled connection scheme must be used. It assumes the LVDS output driver features 50- Ω termination resistor. It also assumes that the LVDS transmitter establishes its own common mode level without relying on the receiver or other external component.

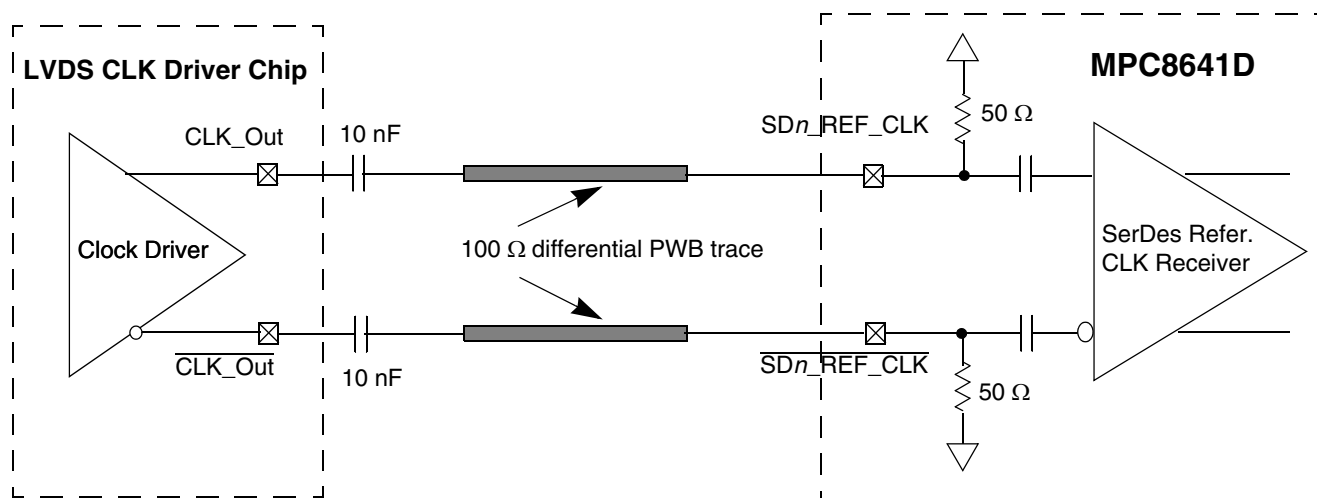


Figure 44. AC-Coupled Differential Connection with LVDS Clock Driver (Reference Only)

Figure 45 shows the SerDes reference clock connection reference circuits for LVPECL type clock driver. Since LVPECL driver's DC levels (both common mode voltages and output swing) are incompatible with

Table 50. Differential Receiver (RX) Input Specifications (continued)

Symbol	Parameter	Min	Nom	Max	Units	Comments
$T_{RX-IDLE-DET-DIFF-ENTERTIME}$	Unexpected Electrical Idle Enter Detect Threshold Integration Time	—	—	10	ms	An unexpected Electrical Idle ($V_{RX-DIFFp-p} < V_{RX-IDLE-DET-DIFFp-p}$) must be recognized no longer than $T_{RX-IDLE-DET-DIFF-ENTERING}$ to signal an unexpected idle condition.
$L_{TX-SKEW}$	Total Skew	—	—	20	ns	Skew across all lanes on a Link. This includes variation in the length of SKP ordered set (for example, COM and one to five Symbols) at the RX as well as any delay differences arising from the interconnect itself.

Notes:

1. No test load is necessarily associated with this value.
2. Specified at the measurement point and measured over any 250 consecutive UIs. The test load in [Figure 52](#) should be used as the RX device when taking measurements (also refer to the Receiver compliance eye diagram shown in [Figure 51](#)). If the clocks to the RX and TX are not derived from the same reference clock, the TX UI recovered from 3500 consecutive UI must be used as a reference for the eye diagram.
3. A $T_{RX-EYE} = 0.40$ UI provides for a total sum of 0.60 UI deterministic and random jitter budget for the Transmitter and interconnect collected any 250 consecutive UIs. The $T_{RX-EYE-MEDIAN-to-MAX-JITTER}$ specification ensures a jitter distribution in which the median and the maximum deviation from the median is less than half of the total. UI jitter budget collected over any 250 consecutive TX UIs. It should be noted that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value. If the clocks to the RX and TX are not derived from the same reference clock, the TX UI recovered from 3500 consecutive UI must be used as the reference for the eye diagram.
4. The Receiver input impedance shall result in a differential return loss greater than or equal to 15 dB with the D+ line biased to 300 mV and the D- line biased to -300 mV and a common mode return loss greater than or equal to 6 dB (no bias required) over a frequency range of 50 MHz to 1.25 GHz. This input impedance requirement applies to all valid input levels. The reference impedance for return loss measurements for is 50 Ω to ground for both the D+ and D- line (that is, as measured by a Vector Network Analyzer with 50 ohm probes - see [Figure 52](#)). Note: that the series capacitors C_{TX} is optional for the return loss measurement.
5. Impedance during all LTSSM states. When transitioning from a Fundamental Reset to Detect (the initial state of the LTSSM) there is a 5 ms transition time before Receiver termination values must be met on all un-configured Lanes of a Port.
6. The RX DC Common Mode Impedance that exists when no power is present or Fundamental Reset is asserted. This helps ensure that the Receiver Detect circuit will not falsely assume a Receiver is powered on when it is not. This term must be measured at 300 mV above the RX ground.
7. It is recommended that the recovered TX UI is calculated using all edges in the 3500 consecutive UI interval with a fit algorithm using a minimization merit function. Least squares and median deviation fits have worked well with experimental and simulated data.

14.5 Receiver Compliance Eye Diagrams

The RX eye diagram in [Figure 51](#) is specified using the passive compliance/test measurement load (see [Figure 52](#)) in place of any real PCI Express RX component.

Note: In general, the minimum Receiver eye diagram measured with the compliance/test measurement load (see [Figure 52](#)) will be larger than the minimum Receiver eye diagram measured over a range of systems at the input Receiver of any real PCI Express component. The degraded eye diagram at the input Receiver is due to traces internal to the package as well as silicon parasitic characteristics which cause the real PCI Express component to vary in impedance from the compliance/test measurement load. The input Receiver eye diagram is implementation specific and is not specified. RX component designer should

Table 58. Transmitter Differential Output Eye Diagram Parameters

Transmitter Type	V _{DIFFmin} (mV)	V _{DIFFmax} (mV)	A (UI)	B (UI)
1.25 GBaud short range	250	500	0.175	0.39
1.25 GBaud long range	400	800	0.175	0.39
2.5 GBaud short range	250	500	0.175	0.39
2.5 GBaud long range	400	800	0.175	0.39
3.125 GBaud short range	250	500	0.175	0.39
3.125 GBaud long range	400	800	0.175	0.39

15.7 Receiver Specifications

LP-Serial receiver electrical and timing specifications are stated in the text and tables of this section.

Receiver input impedance shall result in a differential return loss better than 10 dB and a common mode return loss better than 6 dB from 100 MHz to (0.8)*(Baud Frequency). This includes contributions from on-chip circuitry, the chip package and any off-chip components related to the receiver. AC coupling components are included in this requirement. The reference impedance for return loss measurements is 100 Ohm resistive for differential return loss and 25 Ohm resistive for common mode.

Table 59. Receiver AC Timing Specifications—1.25 GBaud

Characteristic	Symbol	Range		Unit	Notes
		Min	Max		
Differential Input Voltage	V _{IN}	200	1600	mV p-p	Measured at receiver
Deterministic Jitter Tolerance	J _D	0.37	—	UI p-p	Measured at receiver
Combined Deterministic and Random Jitter Tolerance	J _{DR}	0.55	—	UI p-p	Measured at receiver
Total Jitter Tolerance ¹	J _T	0.65	—	UI p-p	Measured at receiver
Multiple Input Skew	S _{MI}	—	24	ns	Skew at the receiver input between lanes of a multilane link
Bit Error Rate	BER	—	10 ⁻¹²	—	—
Unit Interval	UI	800	800	ps	+/- 100 ppm

Note:

1. Total jitter is composed of three components, deterministic jitter, random jitter and single frequency sinusoidal jitter. The sinusoidal jitter may have any amplitude and frequency in the unshaded region of [Figure 55](#). The sinusoidal jitter component is included to ensure margin for low frequency jitter, wander, noise, crosstalk and other variable system effects.

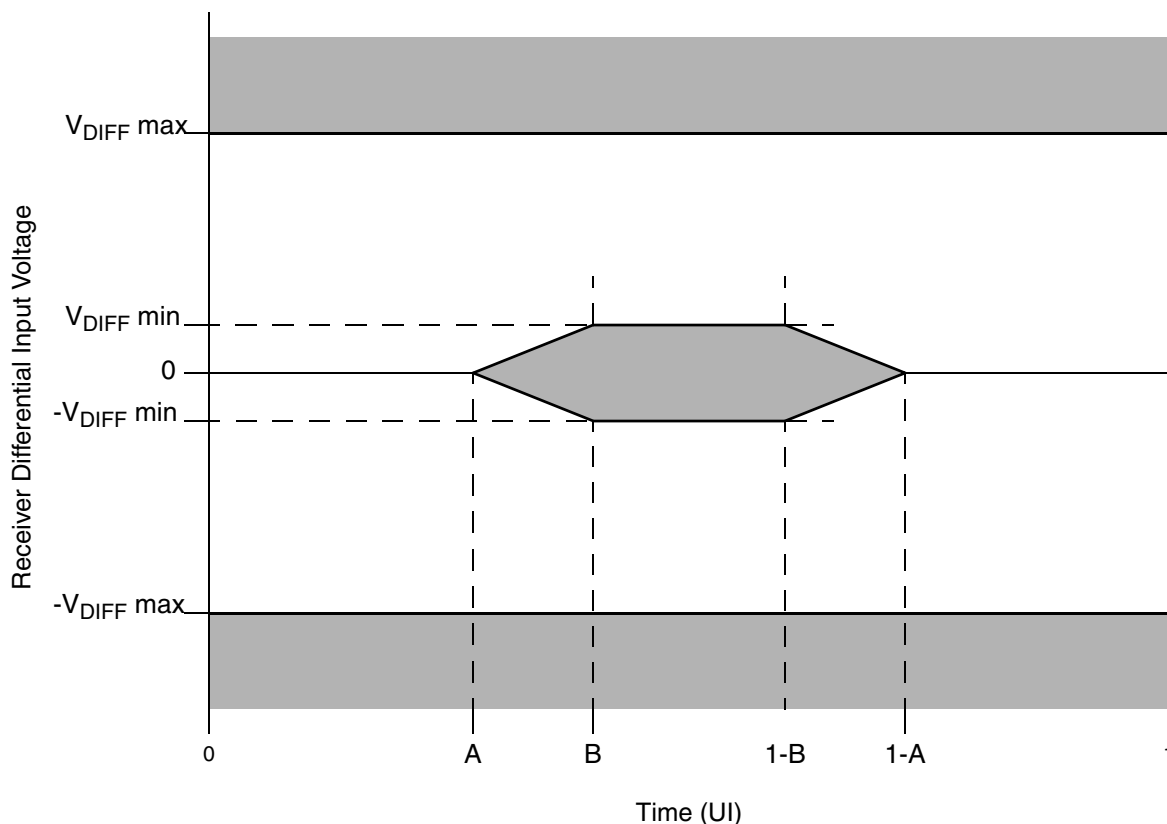


Figure 56. Receiver Input Compliance Mask

Table 62. Receiver Input Compliance Mask Parameters Exclusive of Sinusoidal Jitter

Receiver Type	$V_{DIFFmin}$ (mV)	$V_{DIFFmax}$ (mV)	A (UI)	B (UI)
1.25 GBaud	100	800	0.275	0.400
2.5 GBaud	100	800	0.275	0.400
3.125 GBaud	100	800	0.275	0.400

15.9 Measurement and Test Requirements

Since the LP-Serial electrical specification are guided by the XAUI electrical interface specified in Clause 47 of IEEE 802.3ae-2002, the measurement and test requirements defined here are similarly guided by Clause 47. In addition, the CJPAT test pattern defined in Annex 48A of IEEE 802.3ae-2002 is specified as the test pattern for use in eye pattern and jitter measurements. Annex 48B of IEEE 802.3ae-2002 is recommended as a reference for additional information on jitter test methods.

15.9.1 Eye Template Measurements

For the purpose of eye template measurements, the effects of a single-pole high pass filter with a 3 dB point at (Baud Frequency)/1667 is applied to the jitter. The data pattern for template measurements is the

Table 63. MPC8641 Signal Reference by Functional Block (continued)

Name ¹	Package Pin Number	Pin Type	Power Supply	Notes
D2_MDQ[0:63]	A7, B7, C5, D5, C8, D8, D6, A5, C4, A3, D3, D2, A4, B4, C2, C1, E3, E1, H4, G1, D1, E4, G3, G2, J4, J2, L1, L3, H3, H1, K1, L4, AA4, AA2, AD1, AD2, Y1, AA1, AC1, AC3, AD5, AE1, AG1, AG2, AC4, AD4, AF3, AF4, AH3, AJ1, AM1, AM3, AH1, AH2, AL2, AL3, AK5, AL5, AK7, AM7, AK4, AM4, AM6, AJ7	I/O	D2_GV _{DD}	—
D2_MECC[0:7]	H6, J5, M5, M4, G6, H7, M2, M1	I/O	D2_GV _{DD}	—
D2_MDM[0:8]	C7, B3, F4, J1, AB1, AE2, AK1, AM5, K6	O	D2_GV _{DD}	—
D2_MDQS[0:8]	B6, B1, F1, K2, AB3, AF1, AL1, AL6, L6	I/O	D2_GV _{DD}	—
$\overline{D2_MDQS}[0:8]$	A6, A2, F2, K3, AB2, AE3, AK2, AJ6, K5	I/O	D2_GV _{DD}	—
D2_MBA[0:2]	W5, V5, P3	O	D2_GV _{DD}	—
D2_MA[0:15]	W1, U4, U3, T1, T2, T3, T5, R2, R1, R5, V4, R4, P1, AH5, P4, N1	O	D2_GV _{DD}	—
$\overline{D2_MWE}$	Y4	O	D2_GV _{DD}	—
$\overline{D2_MRAS}$	W3	O	D2_GV _{DD}	—
$\overline{D2_MCAS}$	AB5	O	D2_GV _{DD}	—
$\overline{D2_MCS}[0:3]$	Y3, AF6, AA5, AF7	O	D2_GV _{DD}	—
D2_MCKE[0:3]	N6, N5, N2, N3	O	D2_GV _{DD}	23
D2_MCK[0:5]	U1, F5, AJ3, V2, E7, AG4	O	D2_GV _{DD}	—
$\overline{D2_MCK}[0:5]$	V1, G5, AJ4, W2, E6, AG5	O	D2_GV _{DD}	—
D2_MODT[0:3]	AE6, AG7, AE5, AH6	O	D2_GV _{DD}	—
D2_MDIC[0:1]	F8, F7	IO	D2_GV _{DD}	27
D2_MV _{REF}	A18	DDR Port 2 reference voltage	D2_GV _{DD} / 2	3
High Speed I/O Interface 1 (SERDES 1)⁴				
SD1_TX[0:7]	L26, M24, N26, P24, R26, T24, U26, V24	O	SV _{DD}	—
$\overline{SD1_TX}[0:7]$	L27, M25, N27, P25, R27, T25, U27, V25	O	SV _{DD}	—
SD1_RX[0:7]	J32, K30, L32, M30, T30, U32, V30, W32	I	SV _{DD}	—
$\overline{SD1_RX}[0:7]$	J31, K29, L31, M29, T29, U31, V29, W31	I	SV _{DD}	—
SD1_REF_CLK	N32	I	SV _{DD}	—
$\overline{SD1_REF_CLK}$	N31	I	SV _{DD}	—
SD1_IMP_CAL_TX	Y26	Analog	SV _{DD}	19
SD1_IMP_CAL_RX	J28	Analog	SV _{DD}	30
SD1_PLL_TPD	U28	O	SV _{DD}	13, 17

Table 63. MPC8641 Signal Reference by Functional Block (continued)

Name ¹	Package Pin Number	Pin Type	Power Supply	Notes
TSEC3_TX_EN	AH19	O	TV _{DD}	36
TSEC3_TX_ER	AH17	O	TV _{DD}	—
TSEC3_TX_CLK	AH18	I	TV _{DD}	40
TSEC3_GTX_CLK	AG19	O	TV _{DD}	41
TSEC3_CRS	AE15	I/O	TV _{DD}	37
TSEC3_COL	AF15	I	TV _{DD}	—
TSEC3_RXD[0:7]	AJ17, AE16, AH16, AH14, AJ19, AH15, AG16, AE19	I	TV _{DD}	—
TSEC3_RX_DV	AG15	I	TV _{DD}	—
TSEC3_RX_ER	AF16	I	TV _{DD}	—
TSEC3_RX_CLK	AJ18	I	TV _{DD}	40
eTSEC Port 4 Signals⁵				
TSEC4_TXD[0:3]	AC18, AC16, AD18, AD17	O	TV _{DD}	6
TSEC4_TXD[4]	AD16	O	TV _{DD}	25
TSEC4_TXD[5:7]	AB18, AB17, AB16	O	TV _{DD}	6
TSEC4_TX_EN	AF17	O	TV _{DD}	36
TSEC4_TX_ER	AF19	O	TV _{DD}	—
TSEC4_TX_CLK	AF18	I	TV _{DD}	40
TSEC4_GTX_CLK	AG17	O	TV _{DD}	41
TSEC4_CRS	AB14	I/O	TV _{DD}	37
TSEC4_COL	AC13	I	TV _{DD}	—
TSEC4_RXD[0:7]	AG14, AD13, AF13, AD14, AE14, AB15, AC14, AE17	I	TV _{DD}	—
TSEC4_RX_DV	AC15	I	TV _{DD}	—
TSEC4_RX_ER	AF14	I	TV _{DD}	—
TSEC4_RX_CLK	AG13	I	TV _{DD}	40
Local Bus Signals⁵				
LAD[0:31]	A30, E29, C29, D28, D29, H25, B29, A29, C28, L22, M22, A28, C27, H26, G26, B27, B26, A27, E27, G25, D26, E26, G24, F27, A26, A25, C25, H23, K22, D25, F25, H22	I/O	OV _{DD}	6
LDP[0:3]	A24, E24, C24, B24	I/O	OV _{DD}	6, 22
LA[27:31]	J21, K21, G22, F24, G21	O	OV _{DD}	6, 22
LCS[0:4]	A22, C22, D23, E22, A23	O	OV _{DD}	7
LCS[5]/DMA_DREQ[2]	B23	O	OV _{DD}	7, 9, 10

Table 63. MPC8641 Signal Reference by Functional Block (continued)

Name ¹	Package Pin Number	Pin Type	Power Supply	Notes
D1_MDVAL/LB_DVAL	J16	O	OV _{DD}	10
D2_MDVAL	D19	O	OV _{DD}	—
Power Management Signals⁵				
ASLEEP	C19	O	OV _{DD}	—
System Clocking Signals⁵				
SYSCLK	G16	I	OV _{DD}	—
RTC	K17	I	OV _{DD}	32
CLK_OUT	B16	O	OV _{DD}	23
Test Signals⁵				
$\overline{\text{LSSD_MODE}}$	C18	I	OV _{DD}	26
TEST_MODE[0:3]	C16, E17, D18, D16	I	OV _{DD}	26
JTAG Signals⁵				
TCK	H18	I	OV _{DD}	—
TDI	J18	I	OV _{DD}	24
TDO	G18	O	OV _{DD}	23
TMS	F18	I	OV _{DD}	24
$\overline{\text{TRST}}$	A17	I	OV _{DD}	24
Miscellaneous⁵				
Spare	J17	—	—	13
GPOUT[0:7]/ TSEC1_TXD[0:7]	AF25, AC23, AG24, AG23, AE24, AE23, AE22, AD22	O	OV _{DD}	6, 10
GPIN[0:7]/ TSEC1_RXD[0:7]	AL25, AL24, AK26, AK25, AM26, AF26, AH24, AG25	I	OV _{DD}	10
GPOUT[8:15]/ TSEC2_TXD[0:7]	AB20, AJ23, AJ22, AD19, AH23, AH21, AG22, AG21	O	OV _{DD}	10
GPIN[8:15]/ TSEC2_RXD[0:7]	AL22, AK22, AM21, AH20, AG20, AF20, AF23, AF22	I	OV _{DD}	10
Additional Analog Signals				
TEMP_ANODE	AA11	Thermal	—	—
TEMP_CATHODE	Y11	Thermal	—	—
Sense, Power and GND Signals				
SENSEV _{DD} _Core0	M14	V _{DD} _Core0 sensing pin	—	31
SENSEV _{DD} _Core1	U20	V _{DD} _Core1 sensing pin	—	12,31, S1

Table 63. MPC8641 Signal Reference by Functional Block (continued)

Name ¹	Package Pin Number	Pin Type	Power Supply	Notes
AGND_SRDS1	P30	SerDes Port 1 Ground pin for AV _{DD} _SRDS1	—	—
AGND_SRDS2	AF30	SerDes Port 2 Ground pin for AV _{DD} _SRDS2	—	—
SGND	H28, H32, J30, K31, L28, L29, M32, N30, R29, T32, U30, V31, W29, Y32 AA30, AB31, AC29, AD32, AE30, AG29, AH32, AJ30, AK31, AL29, AM32	Ground pins for SV _{DD}	—	—
XGND	K27, L25, M26, N24, P27, R25, T26, U24, V27, W25, Y28, AA24, AB27, AC25, AD28, AE26, AF27, AH28, AJ26, AK27, AL26, AM28	Ground pins for XV _{DD} _SRDS _n	—	—
Reset Configuration Signals²⁰				
TSEC1_TXD[0] / cfg_alt_boot_vec	AF25	—	LV _{DD}	—
TSEC1_TXD[1] / cfg_platform_freq	AC23	—	LV _{DD}	21
TSEC1_TXD[2:4] / cfg_device_id[5:7]	AG24, AG23, AE24	—	LV _{DD}	—
TSEC1_TXD[5] / cfg_tsec1_reduce	AE23	—	LV _{DD}	—
TSEC1_TXD[6:7] / cfg_tsec1_prtcl[0:1]	AE22, AD22	—	LV _{DD}	—
TSEC2_TXD[0:3] / cfg_rom_loc[0:3]	AB20, AJ23, AJ22, AD19	—	LV _{DD}	—
TSEC2_TXD[4], TSEC2_TX_ER / cfg_dram_type[0:1]	AH23, AB19	—	LV _{DD}	38
TSEC2_TXD[5] / cfg_tsec2_reduce	AH21	—	LV _{DD}	—
TSEC2_TXD[6:7] / cfg_tsec2_prtcl[0:1]	AG22, AG21	—	LV _{DD}	—
TSEC3_TXD[0:1] / cfg_spare[0:1]	AL21, AJ21	O	TV _{DD}	33
TSEC3_TXD[2] / cfg_core1_enable	AM20	O	TV _{DD}	—
TSEC3_TXD[3] / cfg_core1_lm_offset	AJ20	—	LV _{DD}	—
TSEC3_TXD[5] / cfg_tsec3_reduce	AK21	—	LV _{DD}	—

Table 63. MPC8641 Signal Reference by Functional Block (continued)

Name ¹	Package Pin Number	Pin Type	Power Supply	Notes
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Note:

- Multi-pin signals such as D1_MDQ[0:63] and D2_MDQ[0:63] have their physical package pin numbers listed in order corresponding to the signal names.
- Stub Series Terminated Logic (SSTL-18 and SSTL-25) type pins.
- If a DDR port is not used, it is possible to leave the related power supply (Dn_GVDD, Dn_MVREF) turned off at reset. Note that these power supplies can only be powered up again at reset for functionality to occur on the DDR port.
- Low Voltage Differential Signaling (LVDS) type pins.
- Low Voltage Transistor-Transistor Logic (LVTTTL) type pins.
- This pin is a reset configuration pin and appears again in the Reset Configuration Signals section of this table. See the Reset Configuration Signals section of this table for config name and connection details.
- Recommend a weak pull-up resistor (1–10 k Ω) be placed from this pin to its power supply.
- Recommend a weak pull-down resistor (2–10 k Ω) be placed from this pin to ground.
- This multiplexed pin has input status in one mode and output in another
- This pin is a multiplexed signal for different functional blocks and appears more than once in this table.
- This pin is open drain signal.
- Functional only on the MPC8641D.
- These pins should be left floating.
- These pins should be connected to SV_{DD}.
- These pins should be pulled to ground with a strong resistor (270- Ω to 330- Ω).
- These pins should be connected to OVDD.
- This is a SerDes PLL/DLL digital test signal and is only for factory use.
- This is a SerDes PLL/DLL analog test signal and is only for factory use.
- This pin should be pulled to ground with a 100- Ω resistor.
- The pins in this section are reset configuration pins. Each pin has a weak internal pull-up P-FET which is enabled only when the processor is in the reset state. This pull-up is designed such that it can be overpowered by an external 4.7-k Ω pull-down resistor. However, if the signal is intended to be high after reset, and if there is any device on the net which might pull down the value of the net at reset, then a pullup or active driver is needed.
- Should be pulled down at reset if platform frequency is at 400 MHz.
- These pins require 4.7-k Ω pull-up or pull-down resistors and must be driven as they are used to determine PLL configuration ratios at reset.
- This output is actively driven during reset rather than being tri-stated during reset.
- These JTAG pins have weak internal pull-up P-FETs that are always enabled.
- This pin should NOT be pulled down (or driven low) during reset.
- These are test signals for factory use only and must be pulled up (100- Ω to 1- k Ω) to OVDD for normal machine operation.
- Dn_MDIC[0] should be connected to ground with an 18- Ω resistor $\pm$ 1- Ω and Dn_MDIC[1] should be connected Dn_GVDD with an 18- Ω resistor $\pm$ 1- Ω . These pins are used for automatic calibration of the DDR IOs.
- Pin N18 is recommended as a reference point for determining the voltage of V_{DD}_PLAT and is hence considered as the V_{DD}_PLAT sensing voltage and is called SENSEVDD_PLAT.
- Pin P18 is recommended as the ground reference point for SENSEVDD_PLAT and is called SENSEVSS_PLAT.
- This pin should be pulled to ground with a 200- Ω resistor.
- These pins are connected to the power/ground planes internally and may be used by the core power supply to improve tracking and regulation.
- Must be tied low if unused
- These pins may be used as defined functional reset configuration pins in the future. Please include a resistor pull up/down option to allow flexibility of future designs.
- Used as serial data output for SRIO 1x/4x link.
- Used as serial data input for SRIO 1x/4x link.
- This pin requires an external 4.7-k Ω pull-down resistor to prevent PHY from seeing a valid Transmit Enable before it is actively driven.

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 Internet: www.chipcoolers.com
 Wakefield Engineering
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800-522-6752

603-635-5102

Ultimately, the final selection of an appropriate heat sink depends on many factors, such as thermal performance at a given air velocity, spatial volume, mass, attachment method, assembly, and cost.

19.2.1 Internal Package Conduction Resistance

For the exposed-die packaging technology described in [Table 71](#), the intrinsic conduction thermal resistance paths are as follows:

- The die junction-to-case thermal resistance (the case is actually the top of the exposed silicon die)
- The die junction-to-board thermal resistance

[Figure 60](#) depicts the primary heat transfer path for a package with an attached heat sink mounted to a printed-circuit board.

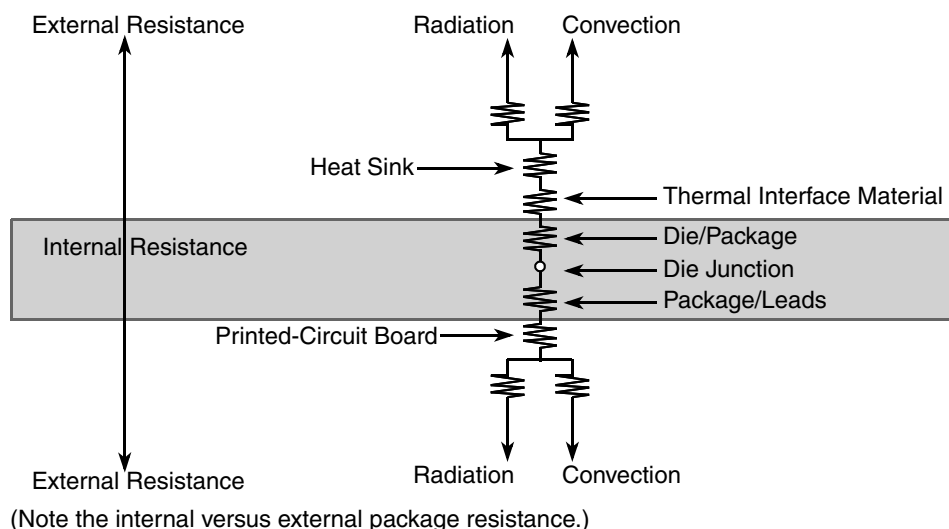


Figure 60. C4 Package with Heat Sink Mounted to a Printed-Circuit Board

Heat generated on the active side of the chip is conducted through the silicon, through the heat sink attach material (or thermal interface material), and finally to the heat sink where it is removed by forced-air convection.

Because the silicon thermal resistance is quite small, the temperature drop in the silicon may be neglected for a first-order analysis. Thus the thermal interface material and the heat sink conduction/convective thermal resistances are the dominant terms.

For other pin pull-up or pull-down recommendations of signals, please see [Section 17, “Signal Listings.”](#)

20.7 Output Buffer DC Impedance

The MPC8641 drivers are characterized over process, voltage, and temperature. For all buses, the driver is a push-pull single-ended driver type (open drain for I²C).

To measure Z_0 for the single-ended drivers, an external resistor is connected from the chip pad to OV_{DD} or GND. Then, the value of each resistor is varied until the pad voltage is $OV_{DD}/2$ (see [Figure 66](#)). The output impedance is the average of two components, the resistances of the pull-up and pull-down devices. When data is held high, SW1 is closed (SW2 is open) and R_P is trimmed until the voltage at the pad equals $OV_{DD}/2$. R_P then becomes the resistance of the pull-up devices. R_P and R_N are designed to be close to each other in value. Then, $Z_0 = (R_P + R_N)/2$.

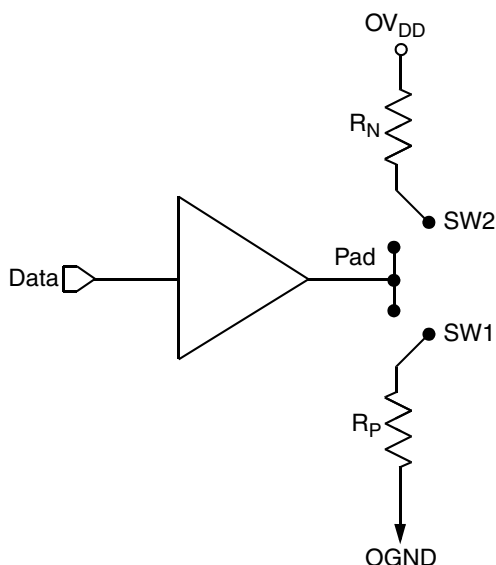


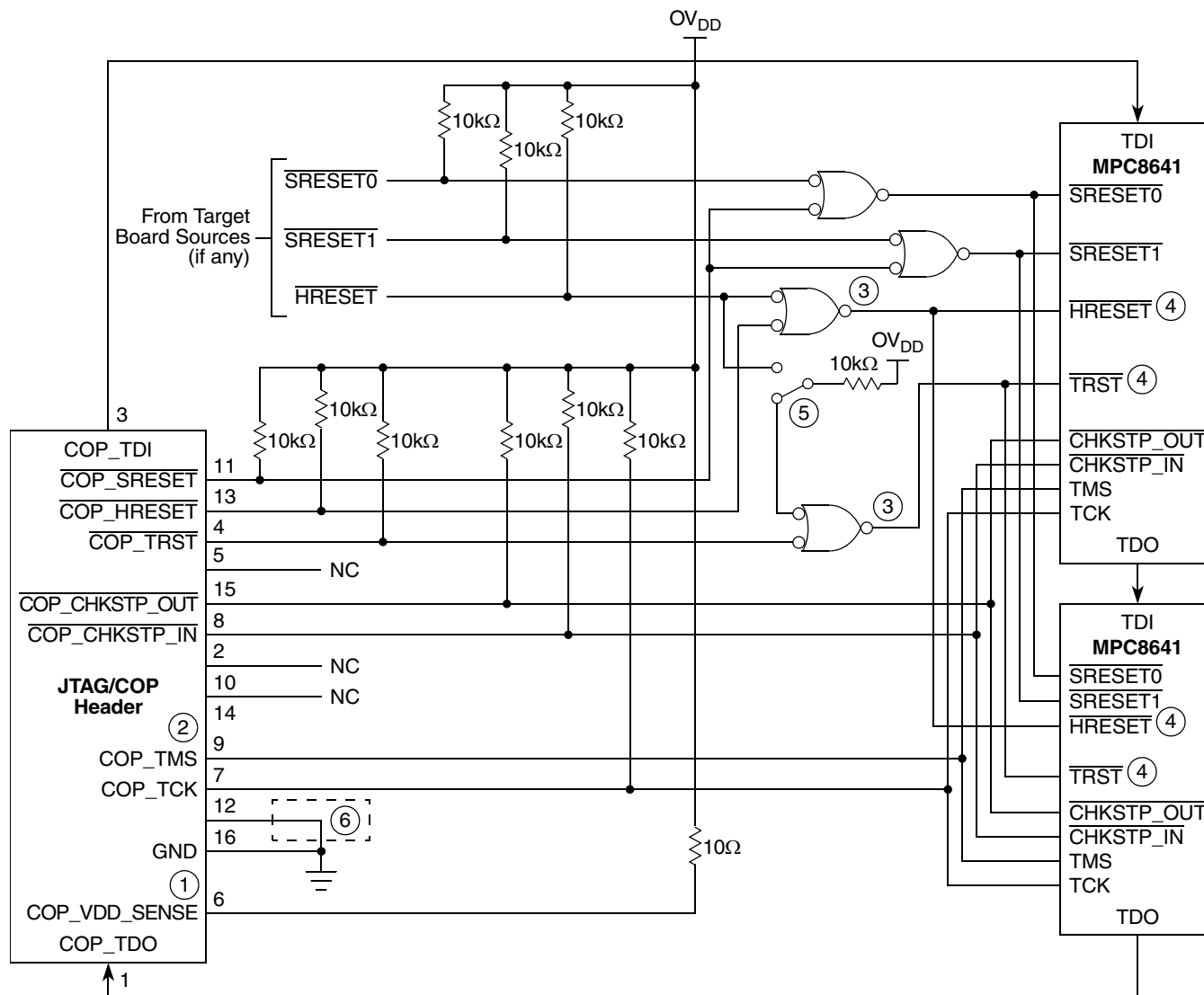
Figure 66. Driver Impedance Measurement

[Table 73](#) summarizes the signal impedance targets. The driver impedances are targeted at minimum V_{DD} , nominal OV_{DD} , 105°C.

Table 73. Impedance Characteristics

Impedance	DUART, Control, Configuration, Power Management	PCI Express	DDR DRAM	Symbol	Unit
R_N	43 Target	25 Target	20 Target	Z_0	W
R_P	43 Target	25 Target	20 Target	Z_0	W

Note: Nominal supply voltages. See [Table 1](#), $T_j = 105^\circ\text{C}$.


Notes:

1. Populate this with a 10Ω resistor for short circuit/current-limiting protection.
2. KEY location; pin 14 is not physically present on the COP header.
3. Use a AND gate with sufficient drive strength to drive two inputs.
4. The COP port and target board should be able to independently assert $\overline{\text{HRESET}}$ and $\overline{\text{TRST}}$ to the processor in order to fully control the processor as shown above.
5. This switch is included as a precaution for BSDL testing. The switch should be open during BSDL testing to avoid accidentally asserting the $\overline{\text{TRST}}$ line. If BSDL testing is not being performed, this switch should be closed or removed.
6. Although pin 12 is defined as a No-Connect, some debug tools may use pin 12 as an additional GND pin for improved signal integrity.

Figure 69. JTAG/COP Interface Connection for Multiple MPC8641 Devices in Daisy Chain Configuration

21 Ordering Information

Ordering information for the parts fully covered by this specification document is provided in [Section 21.1, “Part Numbers Fully Addressed by This Document.”](#)

21.1 Part Numbers Fully Addressed by This Document

[Table 74](#) provides the Freescale part numbering nomenclature for the MPC8641. Note that the individual part numbers correspond to a maximum processor core frequency. For available frequencies, contact your local Freescale sales office. In addition to the processor frequency, the part numbering scheme also includes an application modifier which may specify special application conditions. Each part number also contains a revision code which refers to the die mask revision number.

Table 74. Part Numbering Nomenclature

MC	nnnn	x	xx	nnnn	x	x
Product Code	Part Identifier	Core Count	Package ¹	Core Processor Frequency ² (MHz)	DDR speed (MHz)	Product Revision Level
MC	8641	Blank = Single Core D = Dual Core	HX = High-lead HCTE FC-CBGA VU = RoHS lead-free HCTE FC-CBGA ⁵ VJ = lead-free HCTE FC-CBGA ⁶	1000, 1250, 1333, 1500	N = 500 MHz ⁴ K = 600 MHz J = 533 MHz H = 500 MHz G = 400 MHz	Revision B = 2.0 System Version Register Value for Rev B: 0x8090_0020 - MPC8641 0x8090_0120 - MPC8641D Revision C = 2.1 System Version Register Value for Rev C: 0x8090_0021 - MPC8641 0x8090_0121 - MPC8641D Revision E = 3.0 System Version Register Value for Rev E: 0x8090_0030 - MPC8641 0x8090_0130 - MPC8641D

Notes:

1. See [Section 16, “Package,”](#) for more information on available package types.
2. Processor core frequencies supported by parts addressed by this specification only. Not all parts described in this specification support all core frequencies. Additionally, parts addressed by part number specifications may support other maximum core frequencies.
3. The P prefix in a Freescale part number designates a “Pilot Production Prototype” as defined by Freescale SOP 3-13. These parts have only preliminary reliability and characterization data. Before pilot production prototypes may be shipped, written authorization from the customer must be on file in the applicable sales office acknowledging the qualification status and the fact that product changes may still occur while shipping pilot production prototypes.
4. Part Number MC8641xxx1000NX is our low V_{DD_Coren} device. $V_{DD_Coren} = 0.95\text{ V}$ and $V_{DD_PLAT} = 1.05\text{ V}$.
5. VU part number is RoHS compliant with the permitted exception of the C4 die bumps.
6. VJ part number is entirely lead-free including the C4 die bumps.

Table 75 shows the parts that are available for ordering and their operating conditions.

Table 75. Part Offerings and Operating Conditions

Part Offerings ¹	Operating Conditions
MC8641Dxx1500KX	Dual core Max CPU speed = 1500 MHz, Max DDR = 600 MHz Core Voltage = 1.1 volts
MC8641Dxx1333JX	Dual core Max CPU speed = 1333 MHz, Max DDR = 533 MHz Core Voltage = 1.05 volts
MC8641Dxx1250HX	Dual core Max CPU speed = 1250 MHz, Max DDR = 500 MHz Core Voltage = 1.05 volts
MC8641Dxx1000GX	Dual core Max CPU speed = 1000 MHz, Max DDR = 400 MHz Core Voltage = 1.05 volts
MC8641Dxx1000NX	Dual core MAX CPU speed = 1000 MHz, MAX DDR = 500 MHz Core Voltage = 0.95 volts
MC8641xx1500KX	Single core Max CPU speed = 1500 MHz, Max DDR = 600 MHz Core Voltage = 1.1 volts
MC8641xx1333JX	Single core Max CPU speed = 1333 MHz, Max DDR = 533 MHz Core Voltage = 1.05 volts
MC8641xx1250HX	Single core Max CPU speed = 1250 MHz, Max DDR = 500 MHz Core Voltage = 1.05 volts
MC8641xx1000HX	Single core Max CPU speed = 1000 MHz, Max DDR = 400 MHz Core Voltage = 1.05 volts
MC8641xx1000NX	Single core Max CPU speed = 1000 MHz, Max DDR = 500 MHz Core Voltage = 0.95 volts

¹ Note that the “xx” in the part marking represents the package option. The upper case “X” represents the revision letter. For more information see [Table 74](#).