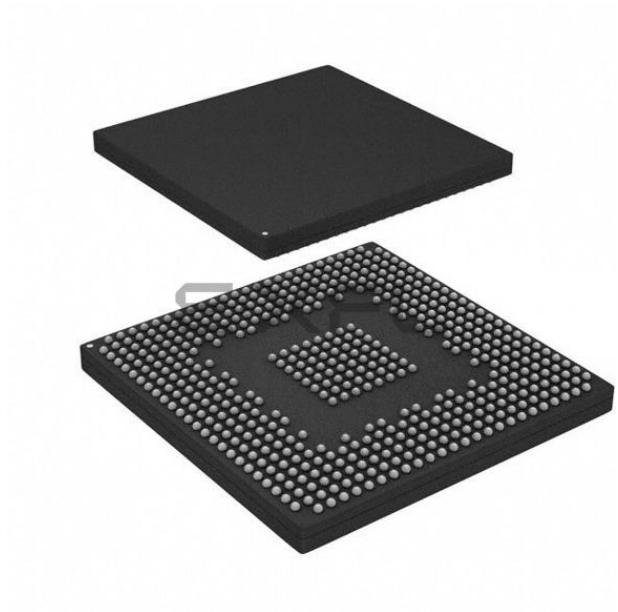

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Details	
Product Status	Active
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Core Size	32-Bit Tri-Core
Speed	264MHz
Connectivity	EBI/EMI, Ethernet, FlexCANbus, LINbus, SCI, SPI
Peripherals	DMA, LVD, POR, Zipwire
Number of I/O	-
Program Memory Size	8MB (8M x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	512K x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	A/D 16b Sigma-Delta, eQADC
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	516-BGA
Supplier Device Package	516-MAPBGA (27x27)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/spc5777crk3mmo3

- Enhanced Modular Input/Output System (eMIOS) supporting 32 unified channels with each channel capable of single action, double action, pulse width modulation (PWM) and modulus counter operation
- Two Enhanced Queued Analog-to-Digital Converter (eQADC) modules with:
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  - Support for a total of 70 analog input pins, expandable to 182 inputs with off-chip multiplexers
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- Four independent 16-bit Sigma-Delta ADCs (SDADCs)
- 10-channel Reaction Module
- Ethernet (FEC)
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- Clock Monitor Units (CMUs)
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- PASS module to support security features
- Nexus development interface (NDI) per IEEE-ISTO 5001-2003 standard, with some support for 2010 standard
- Device and board test support per Joint Test Action Group (JTAG) IEEE 1149.1 and 1149.7
- On-chip voltage regulator controller (VRC) that derives the core logic supply voltage from the high-voltage supply
- On-chip voltage regulator for flash memory
- Self Test capability

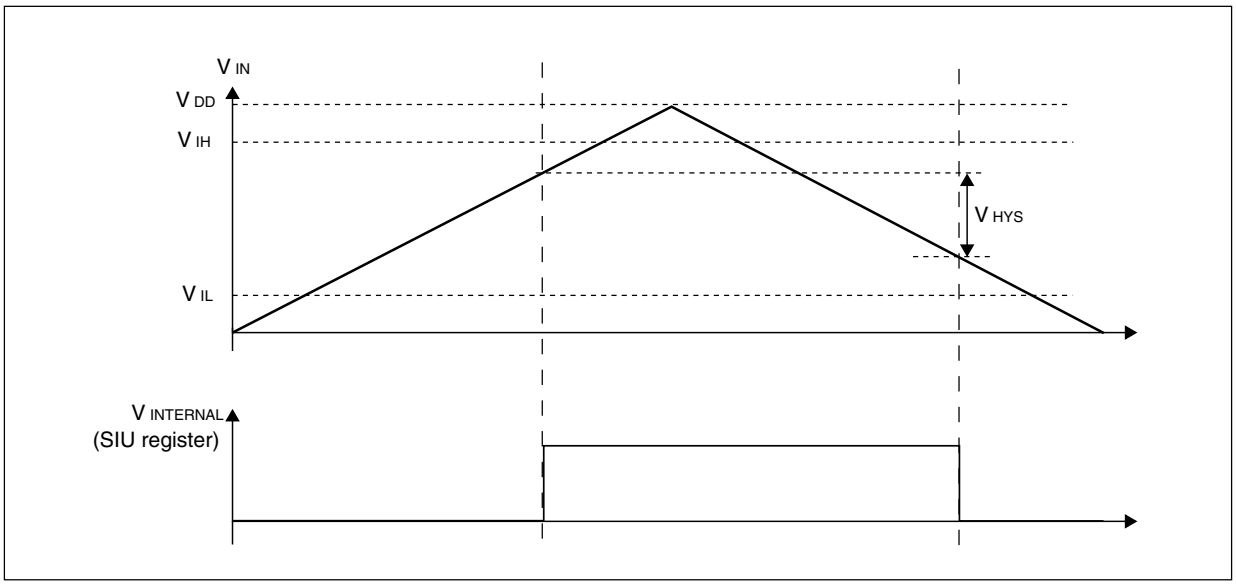


Figure 4. I/O input DC electrical characteristics definition

Table 6. I/O input DC electrical characteristics

Symbol	Parameter	Conditions	Value			Unit
			Min	Typ	Max	
V_{IHCMOS_H}	Input high level CMOS (with hysteresis)	3.0 V < V_{DDEx} < 3.6 V and 4.5 V < V_{DDEx} < 5.5 V	0.65 * V_{DDEx}	—	V_{DDEx} + 0.3	V
V_{IHCMOS}	Input high level CMOS (without hysteresis)	3.0 V < V_{DDEx} < 3.6 V and 4.5 V < V_{DDEx} < 5.5 V	0.55 * V_{DDEx}	—	V_{DDEx} + 0.3	V
V_{ILCMOS_H}	Input low level CMOS (with hysteresis)	3.0 V < V_{DDEx} < 3.6 V and 4.5 V < V_{DDEx} < 5.5 V	–0.3	—	0.35 * V_{DDEx}	V
V_{ILCMOS}	Input low level CMOS (without hysteresis)	3.0 V < V_{DDEx} < 3.6 V and 4.5 V < V_{DDEx} < 5.5 V	–0.3	—	0.4 * V_{DDEx}	V
$V_{HYSCMOS}$	Input hysteresis CMOS	3.0 V < V_{DDEx} < 3.6 V and 4.5 V < V_{DDEx} < 5.5 V	0.1 * V_{DDEx}	—	—	V
Input Characteristics[1]						
I_{LKG}	Digital input leakage	V_{SS} < V_{IN} < V_{DDEx}/V_{DDEHx}	—	—	2.5	µA
I_{LKG_FAST}	Digital input leakage for EBI address/control signal pads	V_{SS} < V_{IN} < V_{DDEx}/V_{DDEHx}	—	—	2.5	µA
I_{LKGA}	Analog pin input leakage (5 V range)	V_{SSA_SD} < V_{IN} < V_{DDA_SD}, V_{SSA_EQ} < V_{IN} < $V_{DDA\_EQA/B}$	—	—	220	nA
C_{IN}	Digital input capacitance	GPIO and EBI input pins	—	—	7	pF

1. For LFAST, microsecond bus, and LVDS input characteristics, see dedicated communication module sections.

Table 7 provides current specifications for weak pullup and pulldown.

5. Below disruptive current conditions, the channel being stressed has conversion values of $3FF for analog inputs greater than V_{RH} and $000 for values less than V_{RL}. Other channels are not affected by non-disruptive conditions.
6. Exceeding limit may cause conversion error on stressed channels and on unstressed channels. Transitions within the limit do not affect device reliability or cause permanent damage.
7. Input must be current limited to the value specified. To determine the value of the required current-limiting resistor, calculate resistance values using $V_{POSCLAMP} = V_{DDA} + 0.5$ V and $V_{NEGCLAMP} = -0.3$ V, then use the larger of the calculated values.
8. Condition applies to two adjacent pins at injection limits.
9. Performance expected with production silicon.
10. All channels have same 10 kΩ < Rs < 100 kΩ Channel under test has Rs = 10 kΩ, $I_{INJ}=I_{INJMAX},I_{INJMIN}$.
11. The TUE specification is always less than the sum of the INL, DNL, offset, and gain errors due to cancelling errors.
12. TUE does not apply to differential conversions.
13. Variable gain is controlled by setting the PRE_GAIN bits in the ADC_ACR1-8 registers to select a gain factor of ×1, ×2, or ×4. Settings are for differential input only. Tested at ×1 gain. Values for other settings are guaranteed as indicated.
14. Guaranteed 10-bit monotonicity.
15. At $V_{RH\_EQ} - V_{RL\_EQ} = 5.12$ V, one LSB = 1.25 mV.

3.8.2 Sigma-Delta ADC (SDADC)

The SDADC is a 16-bit Sigma-Delta analog-to-digital converter with a 333 Ksps maximum output conversion rate.

NOTE

The voltage range is 4.5 V to 5.5 V for SDADC specifications, except where noted otherwise.

Table 18. SDADC electrical specifications

Symbol	Parameter	Conditions	Value			Unit
			Min	Typ	Max	
V_{IN}	ADC input signal	—	0	—	V_{DDA_SD}	V
V_{IN_PK2PK}[1]	Input range peak to peak $V_{IN\_PK2PK} = V_{INP}$[2] – V_{INM},[3]	Single ended $V_{INM} = V_{RL\_SD}$	V_{RH_SD}/GAIN			V
		Single ended $V_{INM} = 0.5*V_{RH\_SD}$ GAIN = 1	±0.5*V_{RH_SD}			
		Single ended $V_{INM} = 0.5*V_{RH\_SD}$ GAIN = 2,4,8,16	±V_{RH_SD}/GAIN			
		Differential $0 < V_{IN} < V_{DDEx}$	±V_{RH_SD}/GAIN			
f_{ADCD_M}	SD clock frequency[4]	—	4	14.4	16	MHz
f_{ADCD_S}	Conversion rate	—	—	—	333	Ksps
—	Oversampling ratio	Internal modulator	24	—	256	—
RESOLUTION	SD register resolution[5]	2's complement notation	16			bit

Table continues on the next page...

Table 18. SDADC electrical specifications (continued)

Symbol	Parameter	Conditions	Value			Unit
			Min	Typ	Max	
GAIN	ADC gain	Defined through SDADC_MCR[PGAN]. Only integer powers of 2 are valid gain values.	1	—	16	—
$\lvert\delta_{GAIN}\rvert$	Absolute value of the ADC gain error[6, 7]	Before calibration (applies to gain setting = 1)	—	—	1.5	%
		After calibration $\Delta V_{RH\_SD} < 5\%$, $\Delta V_{DDA\_SD} < 10\%$ $\Delta T_J < 50$ °C	—	—	5	mV
		After calibration $\Delta V_{RH\_SD} < 5\%$, $\Delta V_{DDA\_SD} < 10\%$ $\Delta T_J < 100$ °C	—	—	7.5	
		After calibration $\Delta V_{RH\_SD} < 5\%$, $\Delta V_{DDA\_SD} < 10\%$ $\Delta T_J < 150$ °C	—	—	10	
V_{OFFSET}	Conversion offset[6, 7]	Before calibration (applies to all gain settings: 1, 2, 4, 8, 16)	—	10*(1+1/gain)	20	mV
		After calibration $\Delta V_{DDA\_SD} < 10\%$ $\Delta T_J < 50$ °C	—	—	5	
		After calibration $\Delta V_{DDA\_SD} < 10\%$ $\Delta T_J < 100$ °C	—	—	7.5	
		After calibration $\Delta V_{DDA\_SD} < 10\%$ $\Delta T_J < 150$ °C	—	—	10	

Table continues on the next page...

Table 20. LVDS pad startup and receiver electrical characteristics[1] (continued)

Symbol	Parameter	Conditions	Value			Unit
			Min	Typ	Max	
t_{PD2NM_TX}	Transmitter startup time (power down to Normal mode)[5]	—	—	0.4	2.75	µs
t_{SM2NM_TX}	Transmitter startup time (Sleep mode to Normal mode)[6]	Not applicable to the MSC/DSPI LVDS pad	—	0.2	0.5	µs
t_{PD2NM_RX}	Receiver startup time (power down to Normal mode)[7]	—	—	20	40	ns
t_{PD2SM_RX}	Receiver startup time (power down to Sleep mode)[8]	Not applicable to the MSC/DSPI LVDS pad	—	20	50	ns
I_{LVDS_BIAS}	LVDS bias current consumption	Tx or Rx enabled	—	—	0.95	mA
TRANSMISSION LINE CHARACTERISTICS (PCB Track)						
Z_0	Transmission line characteristic impedance	—	47.5	50	52.5	Ω
Z_{DIFF}	Transmission line differential impedance	—	95	100	105	Ω
RECEIVER						
V_{ICOM}	Common mode voltage	—	0.15[9]	—	1.6[10]	V
$\|\Delta VI\|$	Differential input voltage	—	100	—	—	mV
V_{HYS}	Input hysteresis	—	25	—	—	mV
R_{IN}	Terminating resistance	V_{DDEH} = 3.0 V to 5.5 V	80	125	150	Ω
C_{IN}	Differential input capacitance[11]	—	—	3.5	6.0	pF
I_{LVDS_RX}	Receiver DC current consumption	Enabled	—	—	0.5	mA

1. The LVDS pad startup and receiver electrical characteristics in this table apply to both the LFAST and the MSC/DSPI LVDS pad except where noted in the conditions.
2. All startup times are defined after a 2 peripheral bridge clock delay from writing to the corresponding enable bit in the LVDS control registers (LCR) of the LFAST and High-Speed Debug modules.
3. Startup times are valid for the maximum external loads CL defined in both the LFAST/HSD and MSC/DSPI transmitter electrical characteristic tables.
4. Bias startup time is defined as the time taken by the current reference block to reach the settling bias current after being enabled.
5. Total transmitter startup time from power down to normal mode is t_{STRT_BIAS} + t_{PD2NM_TX} + 2 peripheral bridge clock periods.
6. Total transmitter startup time from sleep mode to normal mode is t_{SM2NM_TX} + 2 peripheral bridge clock periods. Bias block remains enabled in sleep mode.
7. Total receiver startup time from power down to normal mode is t_{STRT_BIAS} + t_{PD2NM_RX} + 2 peripheral bridge clock periods.
8. Total receiver startup time from power down to sleep mode is t_{PD2SM_RX} + 2 peripheral bridge clock periods. Bias block remains enabled in sleep mode.
9. Absolute min = 0.15 V – (285 mV/2) = 0 V
10. Absolute max = 1.6 V + (285 mV/2) = 1.743 V
11. Total internal capacitance including receiver and termination, co-bonded GPIO pads, and package contributions. For bare die devices, subtract the package value given in Figure 11.

Table 21. LFAST transmitter electrical characteristics[1]

Symbol	Parameter	Conditions	Value			Unit
			Min	Typ	Max	
f_{DATA}	Data rate	—	—	—	240	Mbps

Table continues on the next page...

The following table shows the recommended components to be used in LDO regulation mode.

Table 25. Recommended operating characteristics

Part name	Part type	Nominal	Description
Q1	NPN BJT	h_{FE} = 400	NJD2873: ON Semiconductor LDO voltage regulator controller (VRC)
CI	Capacitor	4.7 µF - 20 V	Ceramic capacitor, total ESR < 70 mΩ
CE	Capacitor	0.047–0.049 µF - 7 V	Ceramic—one capacitor for each V_{DD} pin
CV	Capacitor	22 µF - 20 V	Ceramic V_{DDPMC} (optional 0.1 µF)
CD	Capacitor	22 µF - 20 V	Ceramic supply decoupling capacitor, ESR < 50 mΩ (as close as possible to NPN collector)
CB	Capacitor	0.1 µF - 7 V	Ceramic V_{DDPWR}
R	Resistor	Application specific	Optional; reduces thermal loading on the NPN with high V_{DDPMC} levels

The following diagram shows the LDO configuration connection.

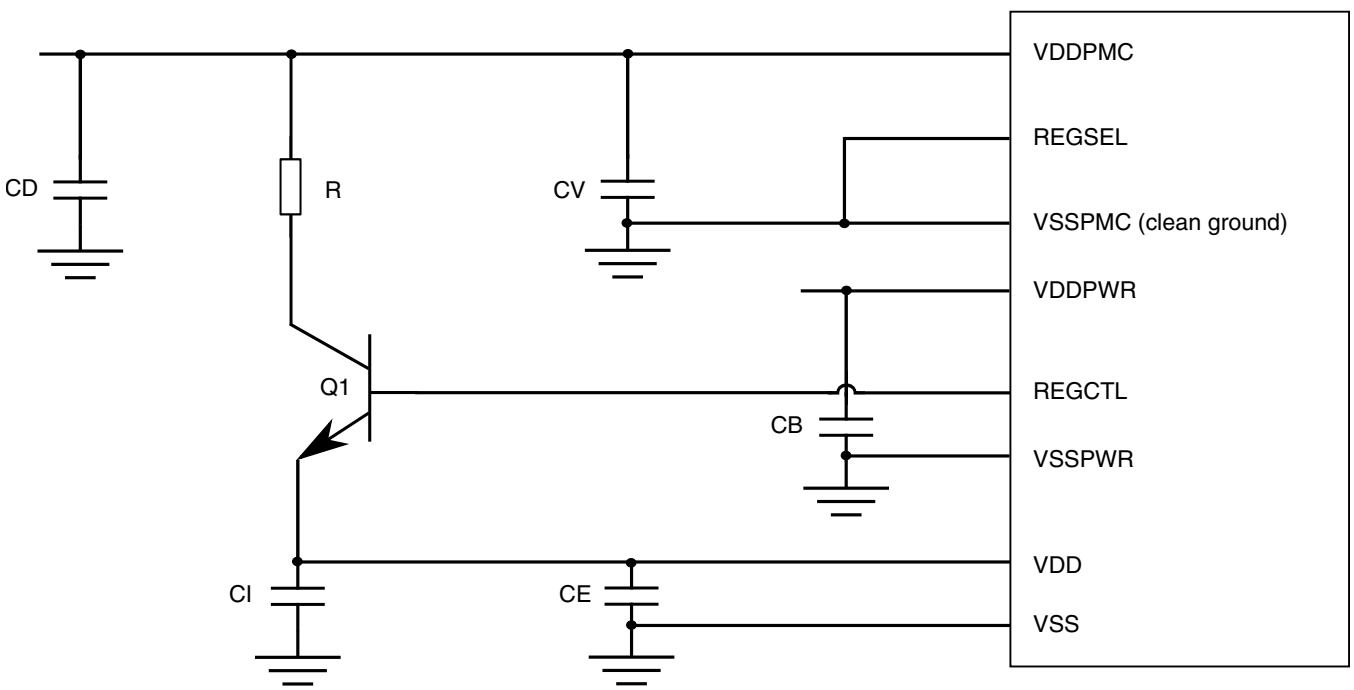


Figure 12. VRC 1.2 V LDO configuration

3.11.1.2 SMPS mode recommended external components and characteristics

The following table shows the recommended components to be used in SMPS regulation mode.

Table 26. Recommended operating characteristics

Part name	Part type	Nominal	Description
Q1	p-MOS	3 A - 20 V	SQ2301ES / FDC642P or equivalent: low threshold p-MOS, Vth < 2.0 V, Rdson @ 4.5 V < 100 mΩ, Cg < 5 nF
D1	Schottky	2 A - 20 V	SS8P3L or equivalent: Vishay™ low Vf Schottky diode
L	Inductor	3−4 µH - 1.5 A	Buck shielded coil low ESR
CI	Capacitor	22 µF - 20 V	Ceramic capacitor, total ESR < 70 mΩ
CE	Capacitor	0.1 µF - 7 V	Ceramic—one capacitor for each V_{DD} pin
CV	Capacitor	22 µF - 20 V	Ceramic V_{DDPMC} (optional 0.1 µF capacitor in parallel)
CD	Capacitor	22 µF - 20 V	Ceramic supply decoupling capacitor, ESR < 50 mΩ (as close as possible to the p-MOS source)
R	Resistor	2.0-4.7 kΩ	Pullup for power p-MOS gate
CB	Capacitor	22 µF - 20 V	Ceramic, connect 100 nF capacitor in parallel (as close as possible to package to reduce current loop from V_{DDPWR} to V_{SSPWR})

The following diagram shows the SMPS configuration connection.

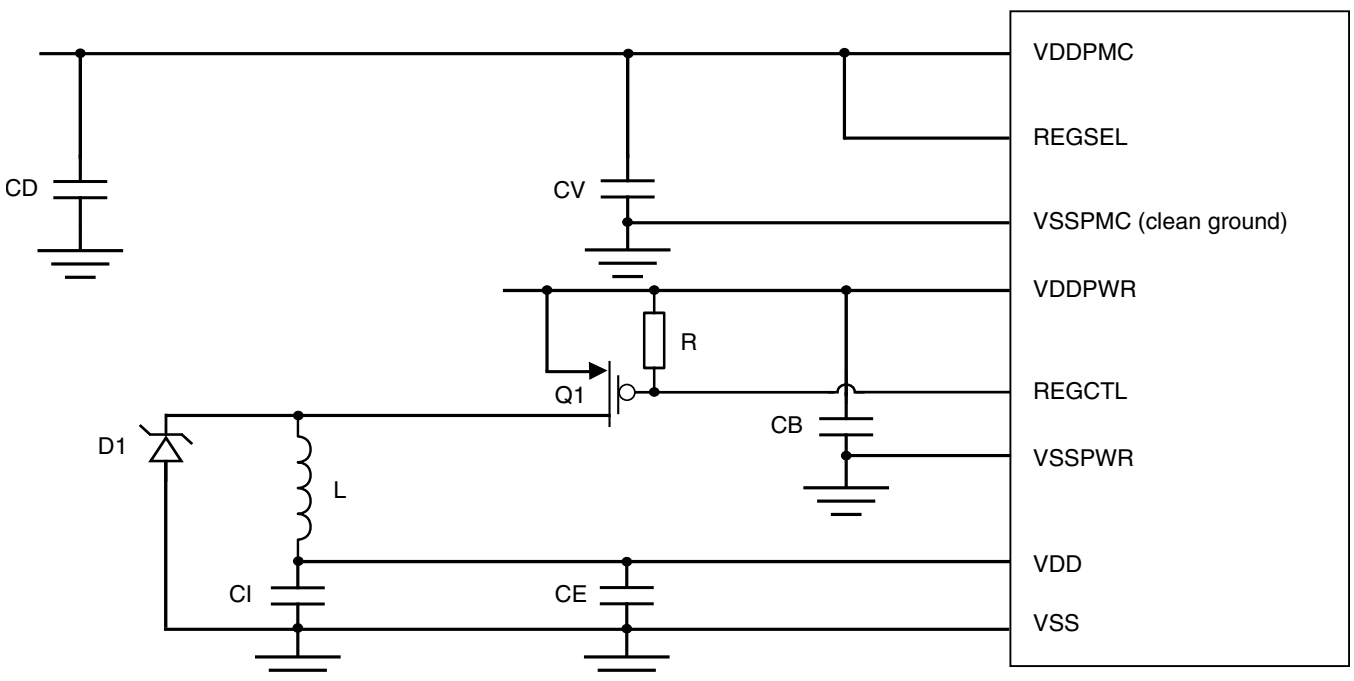


Figure 13. SMPS configuration

NOTE

The REGSEL pin is tied to V_{DDPMC} to select SMPS. If REGSEL is 0, the chip boots with the linear regulator.

See Power sequencing requirements for details about V_{DDPMC} and V_{DDPWR}.

The SMPS regulator characteristics appear in the following table.

Table 27. SMPS electrical characteristics

Symbol	Parameter	Conditions	Value			Unit
			Min	Typ	Max	
$SMPS_{CLOCK}$	SMPS oscillator frequency	Trimmed	825	1000	1220	kHz
$SMPS_{SLOPE}$	SMPS soft-start ramp slope	—	0.01	0.025	0.05	V/µs
$SMPS_{EFF}$	SMPS typical efficiency	—	—	70	—	%

3.11.2 Power management integration

To ensure correct functionality of the device, use the following recommended integration scheme for LDO mode.

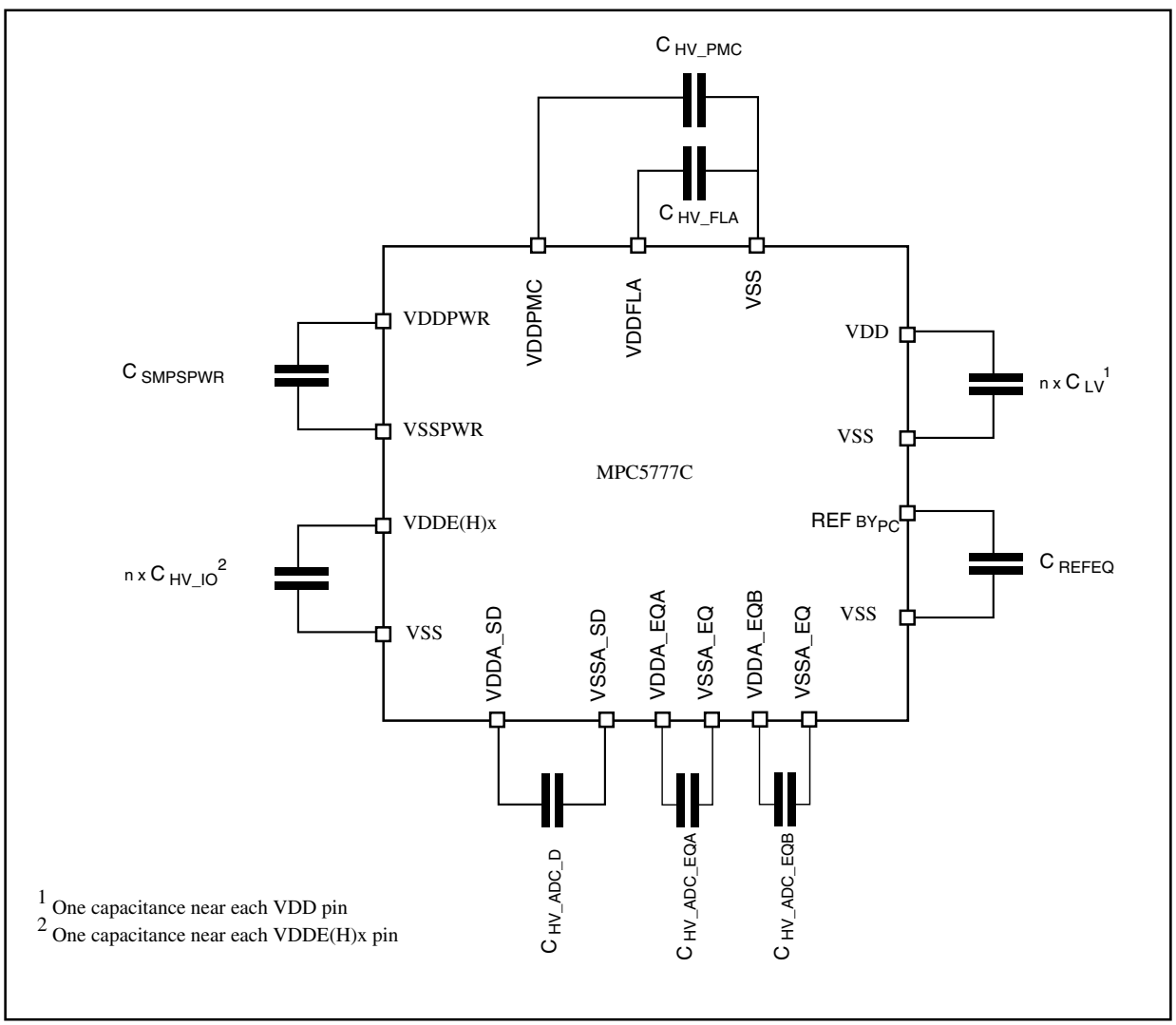


Figure 14. Recommended supply pin circuits

NOTE

In these descriptions, *star route layout* means a track split as close as possible to the power supply source. Each of the split tracks is routed individually to the intended end connection.

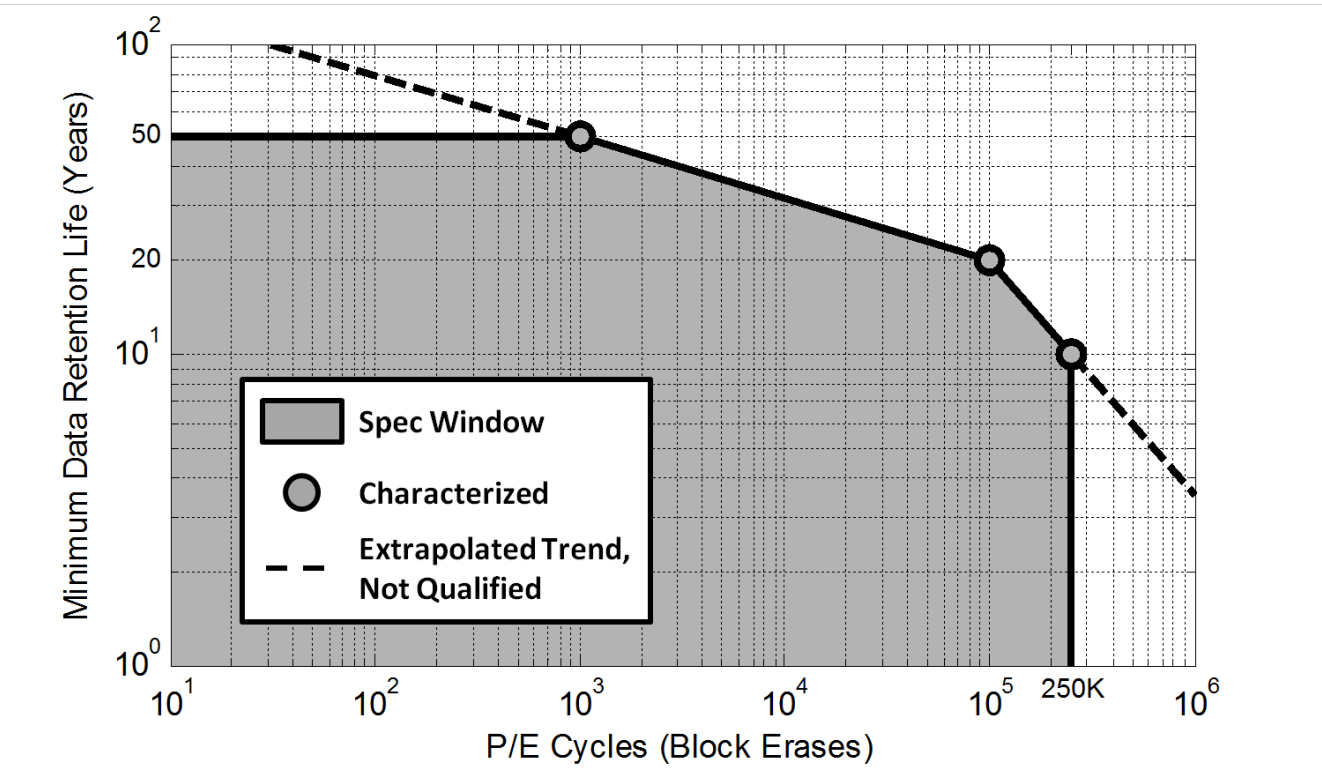
1. For both LDO mode and SMPS mode, V_{DDPMC} and V_{DDPWR} must be connected together (shorted) to ensure aligned voltage ramping up/down. In addition:
   - For SMPS mode, a star route layout of the power track is required to minimize mutual noise. If SMPS mode is not used, the star route layout is not required. V_{DDPWR} is the supply pin for the SMPS circuitry.
   - For 3.3 V operation, V_{DDFLA} must also be star routed and shorted to V_{DDPWR} and V_{DDPMC}. This triple connection is required because 3.3 V does not guarantee correct functionality of the internal V_{DDFLA} regulator. Consequently, V_{DDFLA} is supplied externally.
2. V_{DDA_MISC}: IRC operation is required to provide the clock for chip startup.
   - The V_{DDPMC}, V_{DD}, and V_{DDEH1} (reset pin pad segment) supplies are monitored. They hold IRC until all of them reach operational voltage. In other words, V_{DDA_MISC} must reach its specified minimum operating voltage before or at the same time that all of these monitored voltages reach their respective specified minimum voltages.
   - An alternative is to connect the same supply voltage to both V_{DDEH1} and V_{DDA_MISC}. This alternative approach requires a star route layout to minimize mutual noise.
3. Multiple V_{DDEx} supplies can be powered up in any order.

   During any time when V_{DD} is powered up but V_{DDEx} is not yet powered up: pad outputs are unpowered.

   During any time when V_{DDEx} is powered up before all other supplies: all pad output buffers are tristated.
4. Ramp up V_{DDA_EQ} before V_{DD}. Otherwise, a reset might occur.
5. When the device is powering down while using the internal SMPS regulator, V_{DDPMC} and V_{DDPWR} supplies must ramp down through the voltage range from 2.5 V to 1.5 V in less than 1 second. Slower ramp-down times might result in reduced lifetime reliability of the device.

3.12.4 Data retention vs program/erase cycles

Graphically, Data Retention versus Program/Erase Cycles can be represented by the following figure. The spec window represents qualified limits. The extrapolated dotted line demonstrates technology capability, however is beyond the qualification limits.



3.12.5 Flash memory AC timing specifications

Table 33. Flash memory AC timing specifications

Symbol	Characteristic	Min	Typical	Max	Units
t_{psus}	Time from setting the MCR-PSUS bit until MCR-DONE bit is set to a 1.	—	9.4 plus four system clock periods	11.5 plus four system clock periods	µs
t_{esus}	Time from setting the MCR-ESUS bit until MCR-DONE bit is set to a 1.	—	16 plus four system clock periods	20.8 plus four system clock periods	µs
t_{res}	Time from clearing the MCR-ESUS or PSUS bit with EHV = 1 until DONE goes low.	—	—	100	ns

Table continues on the next page...

Table 34. Flash memory read wait-state and address-pipeline control combinations (continued)

Flash memory frequency	RWSC	APC	Flash memory read latency on mini-cache miss (# of f_{PLATF} clock periods)	Flash memory read latency on mini-cache hit (# of f_{PLATF} clock periods)
100 MHz < f_{PLATF} ≤ 133 MHz	3	1	6	1

3.13 AC timing

3.13.1 Generic timing diagrams

The generic timing diagrams in Figure 16 and Figure 17 apply to all I/O pins with pad types SR and FC. See the associated MPC5777C Microsoft Excel® file in the Reference Manual for the pad type for each pin.

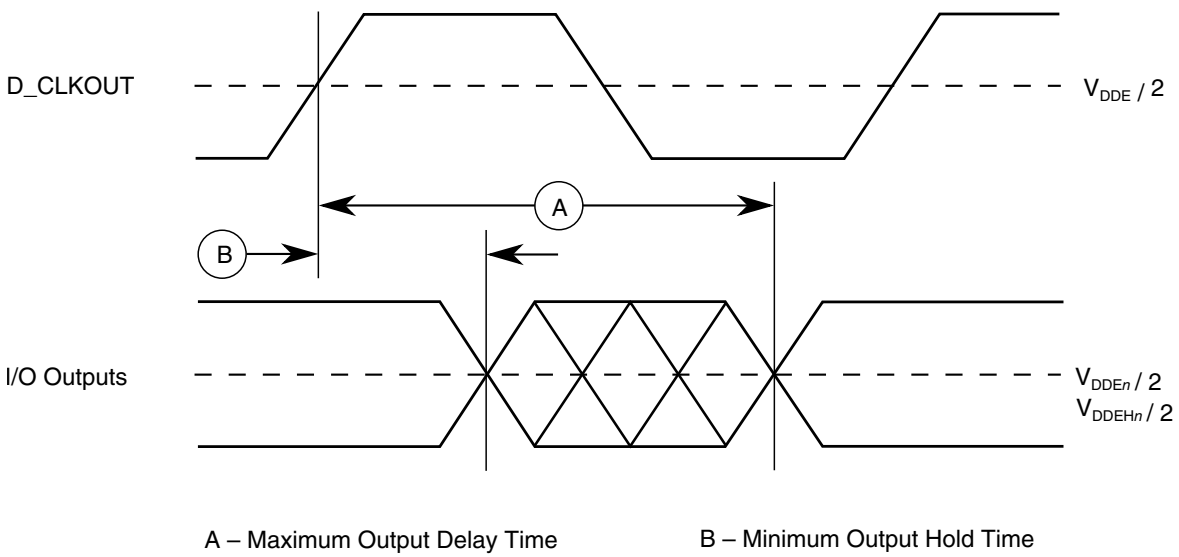


Figure 16. Generic output delay/hold timing

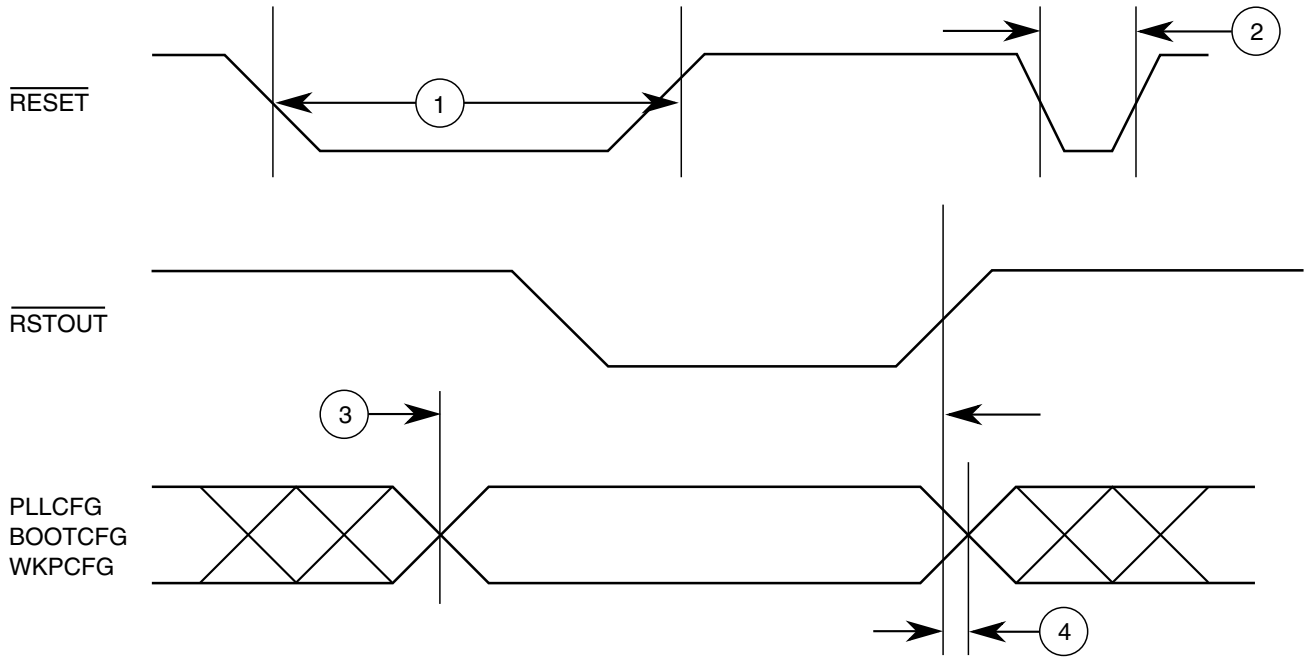


Figure 18. Reset and configuration pin timing

3.13.3 IEEE 1149.1 interface timing

Table 36. JTAG pin AC electrical characteristics[1]

#	Symbol	Characteristic	Value		Unit
			Min	Max	
1	t_{JCYC}	TCK cycle time	100	—	ns
2	t_{JDC}	TCK clock pulse width	40	60	%
3	$t_{TCKRISE}$	TCK rise and fall times (40%–70%)	—	3	ns
4	t_{TMSS}, t_{TDIS}	TMS, TDI data setup time	5	—	ns
5	t_{TMSH}, t_{TDIH}	TMS, TDI data hold time	5	—	ns
6	t_{TDOV}	TCK low to TDO data valid	—	16[2]	ns
7	t_{TDOI}	TCK low to TDO data invalid	0	—	ns
8	t_{TDOHZ}	TCK low to TDO high impedance	—	15	ns
9	t_{JCMPPW}	JCOMP assertion time	100	—	ns
10	t_{JCMPS}	JCOMP setup time to TCK low	40	—	ns
11	t_{BSDV}	TCK falling edge to output valid	—	600[3]	ns
12	t_{BSDVZ}	TCK falling edge to output valid out of high impedance	—	600	ns
13	t_{BSDHZ}	TCK falling edge to output high impedance	—	600	ns
14	t_{BSDST}	Boundary scan input valid to TCK rising edge	15	—	ns
15	t_{BSDHT}	TCK rising edge to boundary scan input invalid	15	—	ns

1. These specifications apply to JTAG boundary scan only. See Table 37 for functional specifications.
2. Timing includes TCK pad delay, clock tree delay, logic delay and TDO output pad delay.
3. Applies to all pins, limited by pad slew rate. Refer to I/O delay and transition specification and add 20 ns for JTAG delay.

Table 38. Bus operation timing[1] (continued)

Spec	Characteristic	Symbol	66 MHz (Ext. bus freq.)[2, 3]		Unit	Notes
			Min	Max		
5	D_CLKOUT Posedge to Output Signal Invalid or High Z (Hold Time) D_ADD[9:30] D_BDIP D_CS[0:3] D_DAT[0:15] D_OE $\overline{\text{D\_RD\_WR}}$ D_TA D_TS $\overline{\text{D\_WE}}$[0:3]/$\overline{\text{D\_BE}}$[0:3]	t_{COH}	1.0/1.5	—	ns	Hold time selectable via SIU_ECCR[EBTS] bit: EBTS = 0: 1.0 ns EBTS = 1: 1.5 ns
6	D_CLKOUT Posedge to Output Signal Valid (Output Delay) D_ADD[9:30] D_BDIP D_CS[0:3]	t_{COV}	—	8.5/9.0	ns	Output valid time selectable via SIU_ECCR[EBTS] bit: EBTS = 0: 8.5 ns EBTS = 1: 9.0 ns
	D_DAT[0:15]			11.5		—
	D_OE $\overline{\text{D\_RD\_WR}}$ D_TA D_TS $\overline{\text{D\_WE}}$[0:3]/$\overline{\text{D\_BE}}$[0:3]			8.5/9.0		Output valid time selectable via SIU_ECCR[EBTS] bit: EBTS = 0: 8.5 ns EBTS = 1: 9.0 ns
7	Input Signal Valid to D_CLKOUT Posedge (Setup Time) D_ADD[9:30] D_DAT[0:15] $\overline{\text{D\_RD\_WR}}$ D_TA D_TS	t_{CIS}	7.5	—	ns	—
8	D_CLKOUT Posedge to Input Signal Invalid (Hold Time) D_ADD[9:30] D_DAT[0:15] $\overline{\text{D\_RD\_WR}}$ D_TA D_TS	t_{CIH}	1.0	—	ns	—
9	D_ALE Pulse Width	t_{APW}	6.5	—	ns	The timing is for Asynchronous external memory system.

Table continues on the next page...

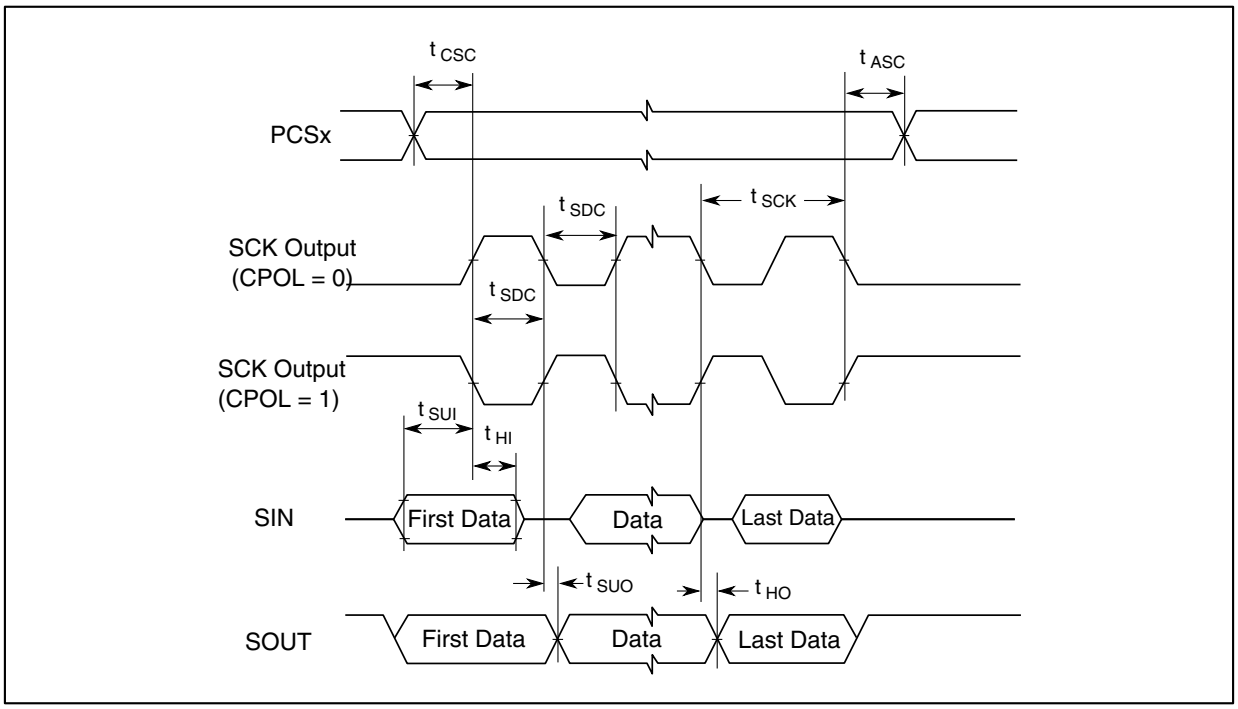


Figure 35. DSPI CMOS master mode – modified timing, CPHA = 0

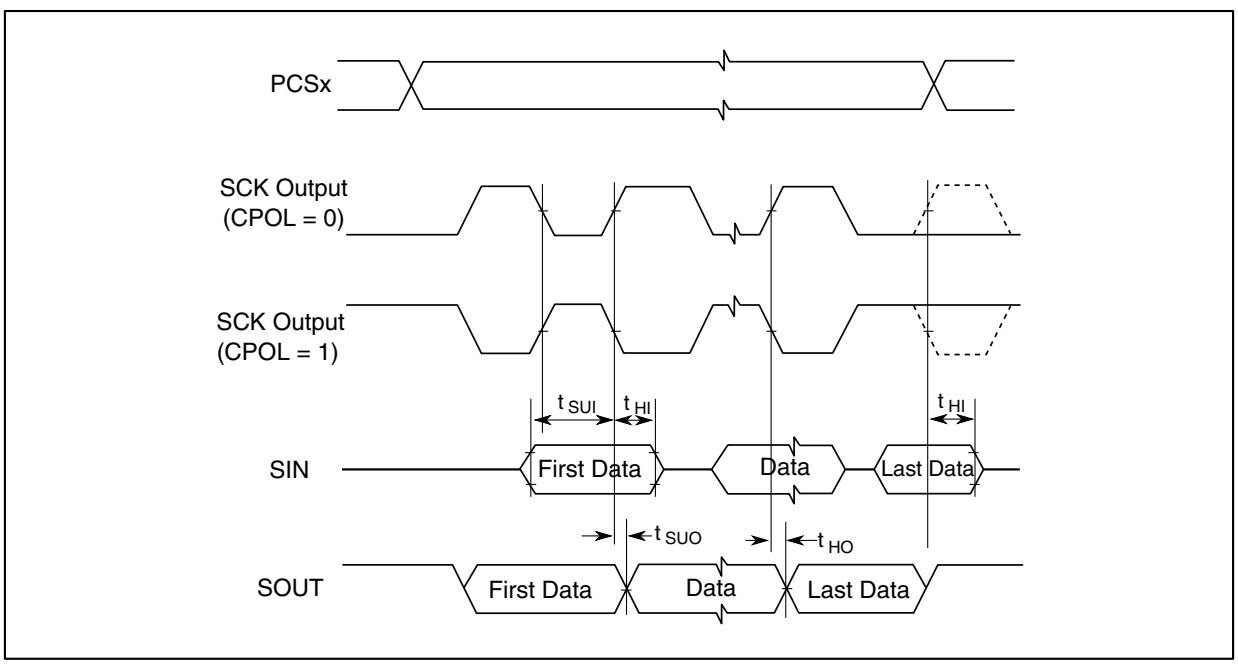


Figure 36. DSPI CMOS master mode – modified timing, CPHA = 1

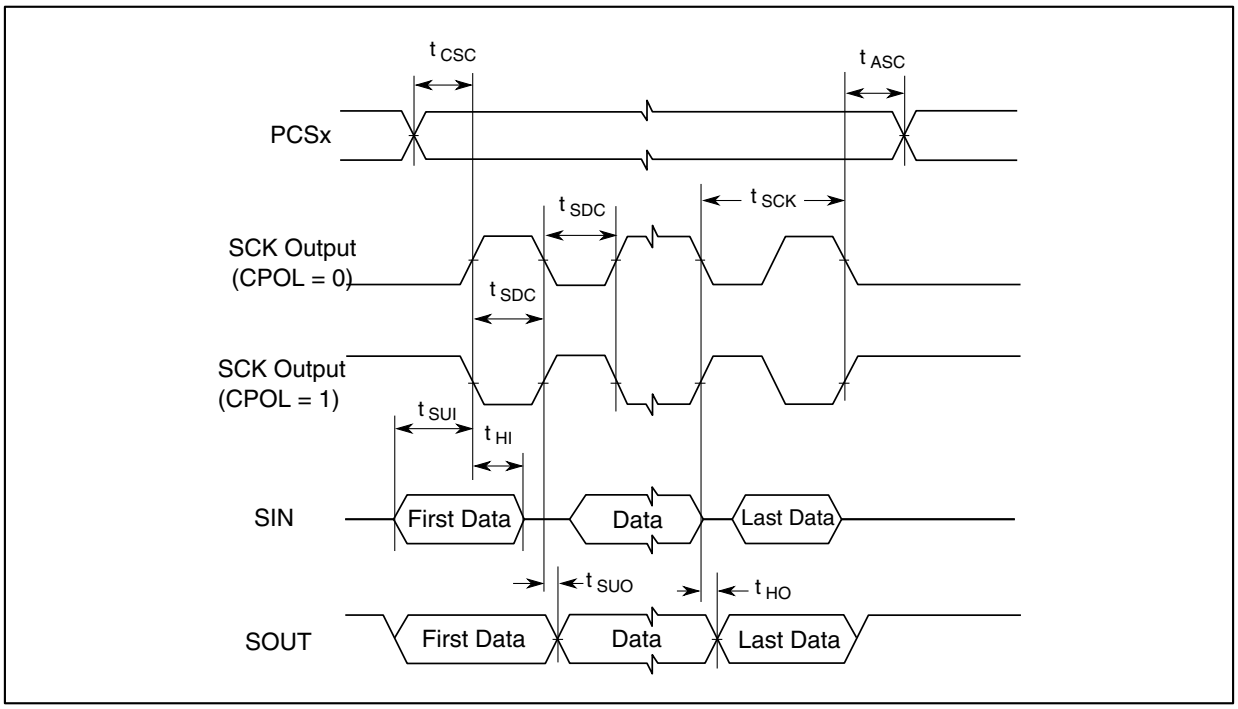


Figure 38. DSPI LVDS master mode – modified timing, CPHA = 0

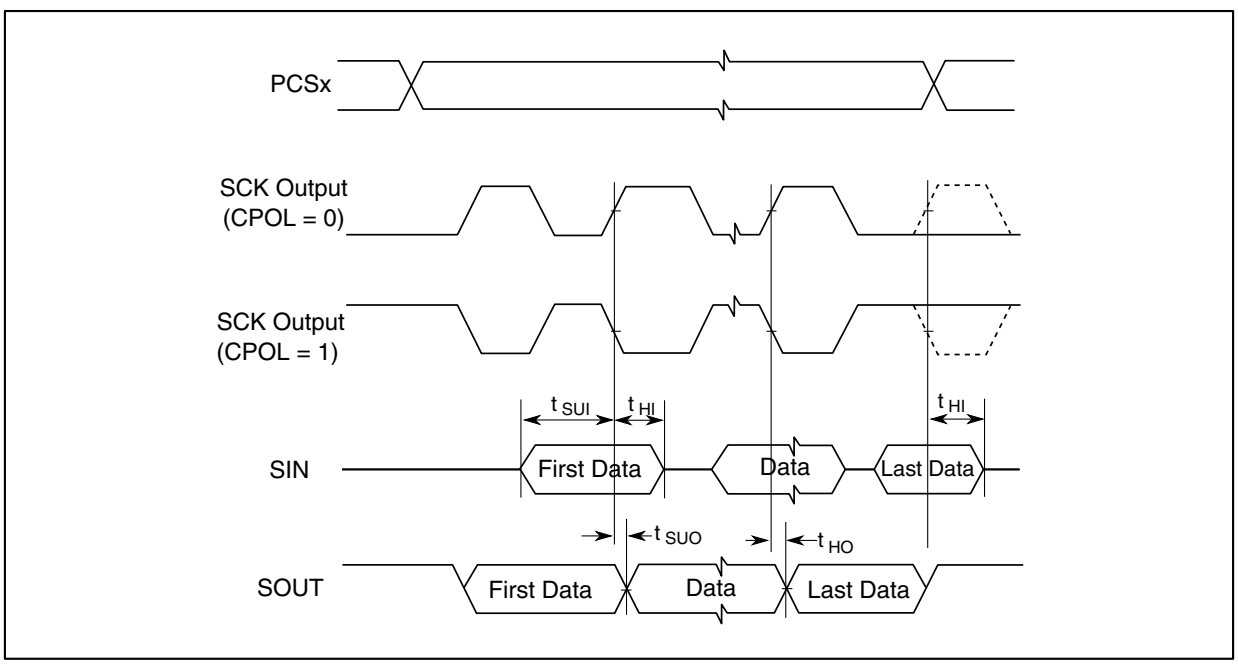


Figure 39. DSPI LVDS master mode – modified timing, CPHA = 1

3.13.9.1.4 DSPI Master Mode – Output Only

Table 46. DSPI LVDS master timing — output only — timed serial bus mode TSB = 1 or ITSB = 1, CPOL = 0 or 1, continuous SCK clock[1, 2]

#	Symbol	Characteristic	Condition[3]		Value[4]		Unit
			Pad drive[5]	Load (C_L)	Min	Max	
1	t_{SCK}	SCK cycle time	LVDS	15 pF to 50 pF differential	25	—	ns
2	t_{CSV}	PCS valid after SCK[6] (SCK with 50 pF differential load cap.)	PCR[SRC]=11b	25 pF	—	8	ns
			PCR[SRC]=10b	50 pF	—	12	ns
3	t_{CSH}	PCS hold after SCK[6] (SCK with 50 pF differential load cap.)	PCR[SRC]=11b	0 pF	–4.0	—	ns
			PCR[SRC]=10b	0 pF	–4.0	—	ns
4	t_{SDC}	SCK duty cycle (SCK with 50 pF differential load cap.)	LVDS	15 pF to 50 pF differential	$1/2t_{SCK} – 2$	$1/2t_{SCK} + 2$	ns
SOUT data valid time (after SCK edge)							
5	t_{SUO}	SOUT data valid time from SCK[7]	LVDS	15 pF to 50 pF differential	—	6	ns
SOUT data hold time (after SCK edge)							
6	t_{HO}	SOUT data hold time after SCK[7]	LVDS	15 pF to 50 pF differential	–7.0	—	ns

1. All DSPI timing specifications apply to pins when using LVDS pads for SCK and SOUT and CMOS pad for PCS with pad driver strength as defined. Timing may degrade for weaker output drivers.
2. TSB = 1 or ITSB = 1 automatically selects MTFE = 1 and CPHA = 1.
3. When a characteristic involves two signals, the pad drive and load conditions apply to each signal's pad, unless specified otherwise.
4. All timing values for output signals in this table are measured to 50% of the output voltage.
5. Pad drive is defined as the PCR[SRC] field setting in the SIU. Timing is guaranteed to same drive capabilities for all signals; mixing of pad drives may reduce operating speeds and may cause incorrect operation.
6. With TSB mode or Continuous SCK clock mode selected, PCS and SCK are driven by the same edge of DSPI_CLKn. This timing value is due to pad delays and signal propagation delays.
7. SOUT Data Valid and Data hold are independent of load capacitance if SCK and SOUT load capacitances are the same value.

Table 47. DSPI CMOS master timing – output only – timed serial bus mode TSB = 1 or ITSB = 1, CPOL = 0 or 1, continuous SCK clock [1, 2]

#	Symbol	Characteristic	Condition[3]		Value[4]		Unit
			Pad drive[5]	Load (C_L)	Min	Max	
1	t_{SCK}	SCK cycle time	PCR[SRC]=11b	25 pF	33.0	—	ns
			PCR[SRC]=10b	50 pF	80.0	—	ns
			PCR[SRC]=01b	50 pF	200.0	—	ns
2	t_{CSV}	PCS valid after SCK[6]	PCR[SRC]=11b	25 pF	7	—	ns
			PCR[SRC]=10b	50 pF	8	—	ns
			PCR[SRC]=01b	50 pF	18	—	ns
			PCS: PCR[SRC]=01b SCK: PCR[SRC]=10b	50 pF	45	—	ns

Table continues on the next page...

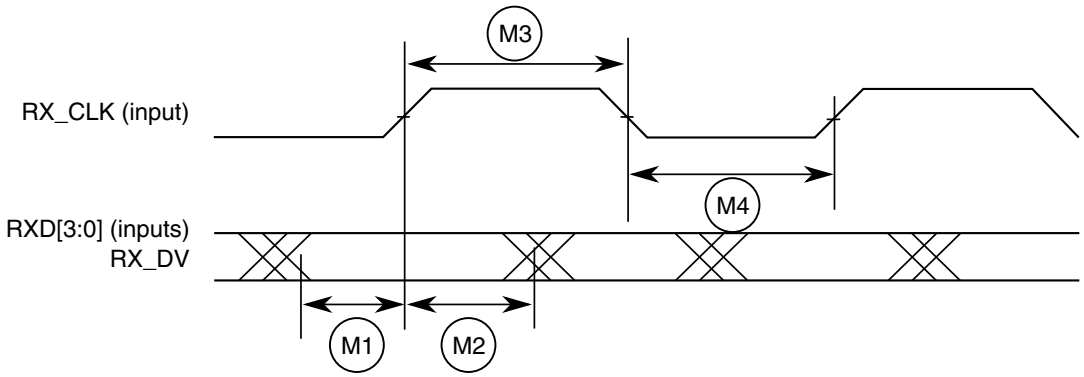


Figure 41. MII receive signal timing diagram

3.13.10.2 MII transmit signal timing (TXD[3:0], TX_EN, and TX_CLK)

The transmitter functions correctly up to a TX_CLK maximum frequency of 25 MHz +1%. There is no minimum frequency requirement. The system clock frequency must be at least equal to or greater than the TX_CLK frequency.

The transmit outputs (TXD[3:0], TX_EN) can be programmed to transition from either the rising or falling edge of TX_CLK, and the timing is the same in either case. This options allows the use of noncompliant MII PHYs.

Refer to the *MPC5777C Microcontroller Reference Manual's* Fast Ethernet Controller (FEC) chapter for details of this option and how to enable it.

Table 49. MII transmit signal timing[1]

Symbol	Characteristic	Value[2]		Unit
		Min	Max	
M5	TX_CLK to TXD[3:0], TX_EN invalid	4.5	—	ns
M6	TX_CLK to TXD[3:0], TX_EN valid	—	25	ns
M7	TX_CLK pulse width high	35%	65%	TX_CLK period
M8	TX_CLK pulse width low	35%	65%	TX_CLK period

1. All timing specifications valid to the pad input levels defined in I/O pad specifications.
2. Output parameters are valid for C_L = 25 pF, where C_L is the external load to the device. The internal package capacitance is accounted for, and does not need to be subtracted from the 25 pF value.

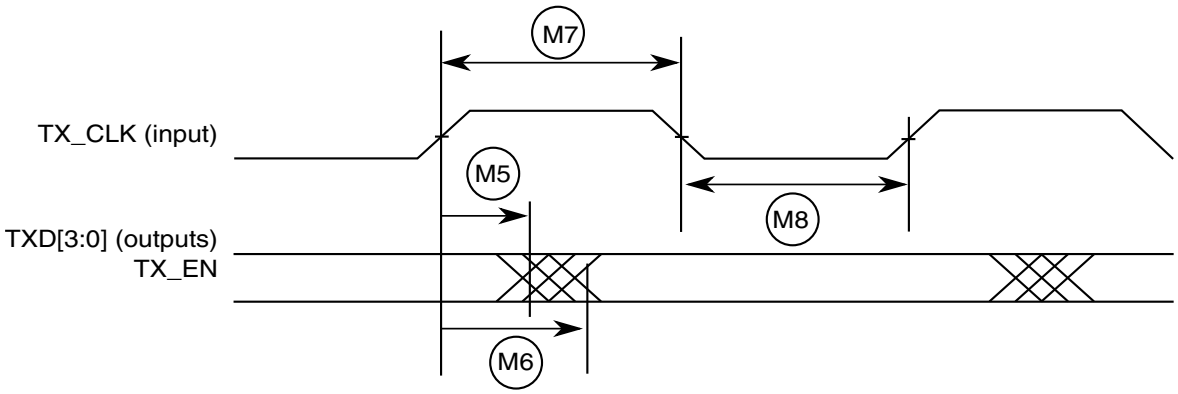


Figure 42. MII transmit signal timing diagram

3.13.10.3 MII async inputs signal timing (CRS)

Table 50. MII async inputs signal timing

Symbol	Characteristic	Value		Unit
		Min	Max	
M9	CRS minimum pulse width	1.5	—	TX_CLK period

Figure 43. MII async inputs timing diagram

3.13.10.4 MII and RMII serial management channel timing (MDIO and MDC)

The FEC functions correctly with a maximum MDC frequency of 2.5 MHz.

Table 51. MII serial management channel timing[1]

Symbol	Characteristic	Value[2]		Unit
		Min	Max	
M10	MDC falling edge to MDIO output invalid (minimum propagation delay)	0	—	ns
M11	MDC falling edge to MDIO output valid (max prop delay)	—	25	ns
M12	MDIO (input) to MDC rising edge setup	10	—	ns
M13	MDIO (input) to MDC rising edge hold	0	—	ns
M14	MDC pulse width high	40%	60%	MDC period
M15	MDC pulse width low	40%	60%	MDC period

1. All timing specifications valid to the pad input levels defined in I/O pad specifications.
2. Output parameters are valid for C_L = 25 pF, where C_L is the external load to the device. The internal package capacitance is accounted for, and does not need to be subtracted from the 25 pF value

6. Thermal characterization parameter indicating the temperature difference between package top and the junction temperature per JEDEC JESD51-2.

Table 55. Thermal characteristics, 516-ball MAPBGA package

Characteristic	Symbol	Value	Unit
Junction to Ambient [1, 2] Natural Convection (Single layer board)	$R_{\Theta JA}$	28.5	°C/W
Junction to Ambient [1, 3] Natural Convection (Four layer board 2s2p)	$R_{\Theta JA}$	20.0	°C/W
Junction to Ambient (@200 ft./min., Single layer board)	$R_{\Theta JMA}$	21.3	°C/W
Junction to Ambient (@200 ft./min., Four layer board 2s2p)	$R_{\Theta JMA}$	15.5	°C/W
Junction to Board [4]	$R_{\Theta JB}$	8.8	°C/W
Junction to Case [5]	$R_{\Theta JC}$	4.8	°C/W
Junction to Package Top [6] Natural Convection	Ψ_{JT}	0.2	°C/W

1. Junction temperature is a function of on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, air flow, power dissipation of other components on the board, and board thermal resistance.
2. Per JEDEC JESD51-2 with the single layer board horizontal. Board meets JESD51-9 specification.
3. Per JEDEC JESD51-6 with the board horizontal.
4. Thermal resistance between the die and the printed circuit board per JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.
5. Indicates the average thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1) with the cold plate temperature used for the case temperature.
6. Thermal characterization parameter indicating the temperature difference between package top and the junction temperature per JEDEC JESD51-2.

4.1.1 General notes for thermal characteristics

An estimation of the chip junction temperature, T_J, can be obtained from the equation:

$$T_J = T_A + (R_{\Theta JA} * P_D)$$

where:

T_A = ambient temperature for the package (°C)

$R_{\Theta JA}$ = junction-to-ambient thermal resistance (°C/W)

P_D = power dissipation in the package (W)

The thermal resistance values used are based on the JEDEC JESD51 series of standards to provide consistent values for estimations and comparisons. The difference between the values determined for the single-layer (1s) board compared to a four-layer board that has two signal layers, a power and a ground plane (2s2p), demonstrate that the effective thermal resistance is not a constant. The thermal resistance depends on the:

- Construction of the application board (number of planes)
- Effective size of the board which cools the component

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